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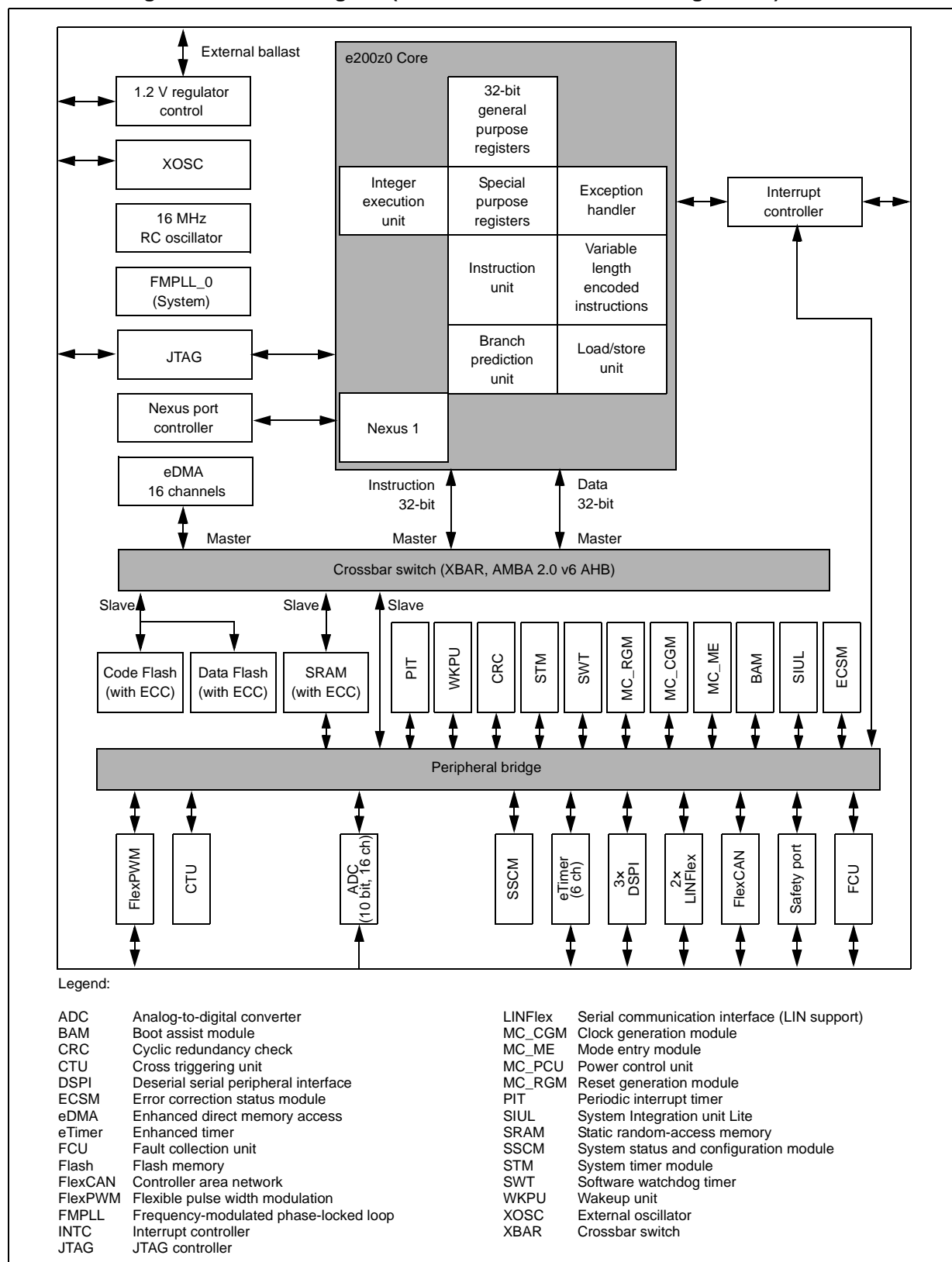
Details

Product Status	Active
Core Processor	e200z0h
Core Size	32-Bit Single-Core
Speed	64MHz
Connectivity	CANbus, LINbus, SPI, UART/USART
Peripherals	DMA, POR, PWM, WDT
Number of I/O	37
Program Memory Size	256KB (256K x 8)
Program Memory Type	FLASH
EEPROM Size	-
RAM Size	20K x 8
Voltage - Supply (Vcc/Vdd)	3V ~ 5.5V
Data Converters	A/D 16x10b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 125°C (TA)
Mounting Type	Surface Mount
Package / Case	64-LQFP
Supplier Device Package	64-LQFP (10x10)
Purchase URL	https://www.e-xfl.com/product-detail/stmicroelectronics/spc560p40l1cefar

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Figure 1. Block diagram (SPC560P40 full-featured configuration)



Digital part:

- 16 input channels
- 4 analog watchdogs comparing ADC results against predefined levels (low, high, range) before results are stored in the appropriate ADC result location
- 2 modes of operation: Motor Control mode or Regular mode
- Regular mode features
 - Register based interface with the CPU: control register, status register and 1 result register per channel
 - ADC state machine managing 3 request flows: regular command, hardware injected command and software injected command
 - Selectable priority between software and hardware injected commands
 - DMA compatible interface
- CTU-controlled mode features
 - Triggered mode only
 - 4 independent result queues (1×16 entries, 2×8 entries, 1×4 entries)
 - Result alignment circuitry (left justified and right justified)
 - 32-bit read mode allows to have channel ID on one of the 16-bit part
 - DMA compatible interfaces

1.5.27 Cross triggering unit (CTU)

The cross triggering unit allows automatic generation of ADC conversion requests on user selected conditions without CPU load during the PWM period and with minimized CPU load for dynamic configuration.

It implements the following features:

- Double buffered trigger generation unit with up to 8 independent triggers generated from external triggers
- Trigger generation unit configurable in sequential mode or in triggered mode
- Each trigger can be appropriately delayed to compensate the delay of external low pass filter
- Double buffered global trigger unit allowing eTimer synchronization and/or ADC command generation
- Double buffered ADC command list pointers to minimize ADC-trigger unit update
- Double buffered ADC conversion command list with up to 24 ADC commands
- Each trigger capable of generating consecutive commands
- ADC conversion command allows to control ADC channel, single or synchronous sampling, independent result queue selection

1.5.28 Nexus Development Interface (NDI)

The NDI (Nexus Development Interface) block is compliant with Nexus Class 1 of the IEEE-ISTO 5001-2003 standard. This development support is supplied for MCUs without requiring external address and data pins for internal visibility. The NDI block is an integration of several individual Nexus blocks that are selected to provide the development support interface for this device. The NDI block interfaces to the host processor and internal busses to provide development support as per the IEEE-ISTO 5001-2003 Nexus Class 1 standard.

Table 5. Supply pins (continued)

Supply		Pin	
Symbol	Description	64-pin	100-pin
$V_{DD_LV_COR2}$	1.2 V supply pins for core logic and code Flash. Decoupling capacitor must be connected between these pins and the nearest $V_{SS_LV_COR}$ pin.	58	92
$V_{SS_LV_COR2}$	1.2 V supply pins for core logic and code Flash. Decoupling capacitor must be connected between these pins and the nearest $V_{DD_LV_COR}$ pin.	59	93

1. Analog supply/ground and high/low reference lines are internally physically separate, but are shorted via a double-bonding connection on $V_{DD_HV_ADCx}/V_{SS_HV_ADCx}$ pins.

2.2.2 System pins

[Table 6](#) and [Table 7](#) contain information on pin functions for the SPC560P34/SPC560P40 devices. The pins listed in [Table 6](#) are single-function pins. The pins shown in [Table 7](#) are multi-function pins, programmable via their respective pad configuration register (PCR) values.

Table 6. System pins

Symbol	Description	Direction	Pad speed ⁽¹⁾		Pin	
			SRC = 0	SRC = 1	64-pin	100-pin
Dedicated pins						
NMI	Non-maskable Interrupt	Input only	Slow	—	1	1
XTAL	Analog output of the oscillator amplifier circuit—needs to be grounded if oscillator is used in bypass mode	—	—	—	11	18
EXTAL	Analog input of the oscillator amplifier circuit, when the oscillator is not in bypass mode Analog input for the clock generator when the oscillator is in bypass mode	—	—	—	12	19
TDI	JTAG test data input	Input only	Slow	—	35	58
TMS	JTAG state machine control	Input only	Slow	—	36	59
TCK	JTAG clock	Input only	Slow	—	37	60
TDO	JTAG test data output	Output only	Slow	Fast	38	61
Reset pin						
$\overline{\text{RESET}}$	Bidirectional reset with Schmitt trigger characteristics and noise filter	Bidirectional	Medium	—	13	20
Test pin						
VPP_TEST	Pin for testing purpose only. To be tied to ground in normal operating mode.	—	—	—	47	74

1. SRC values refer to the value assigned to the Slew Rate Control bits of the pad configuration register.

Table 7. Pin muxing (continued)

Port pin	PCR register	Alternate function ^{(1),(2)}	Functions	Peripheral ⁽³⁾	I/O direction ⁽⁴⁾	Pad speed ⁽⁵⁾		Pin	
						SRC = 0	SRC = 1	64-pin	100-pin
A[11]	PCR[11]	ALT0	GPIO[11]	SIUL	I/O	Slow	Medium	53	82
		ALT1	SCK	DSPI_2	I/O				
		ALT2	A[0]	FlexPWM_0	O				
		ALT3	A[2]	FlexPWM_0	O				
		—	EIRQ[10]	SIUL	I				
A[12]	PCR[12]	ALT0	GPIO[12]	SIUL	I/O	Slow	Medium	54	83
		ALT1	SOUT	DSPI_2	O				
		ALT2	A[2]	FlexPWM_0	O				
		ALT3	B[2]	FlexPWM_0	O				
		—	EIRQ[11]	SIUL	I				
A[13]	PCR[13]	ALT0	GPIO[13]	SIUL	I/O	Slow	Medium	61	95
		ALT1	—	—	—				
		ALT2	B[2]	FlexPWM_0	O				
		ALT3	—	—	—				
		—	SIN	DSPI_2	I				
		—	FAULT[0]	FlexPWM_0	I				
		—	EIRQ[12]	SIUL	I				
A[14]	PCR[14]	ALT0	GPIO[14]	SIUL	I/O	Slow	Medium	63	99
		ALT1	TXD	Safety Port_0	O				
		ALT2	—	—	—				
		ALT3	—	—	—				
		—	EIRQ[13]	SIUL	I				
A[15]	PCR[15]	ALT0	GPIO[15]	SIUL	I/O	Slow	Medium	64	100
		ALT1	—	—	—				
		ALT2	—	—	—				
		ALT3	—	—	—				
		—	RXD	Safety Port_0	I				
		—	EIRQ[14]	SIUL	I				
Port B (16-bit)									
B[0]	PCR[16]	ALT0	GPIO[16]	SIUL	I/O	Slow	Medium	49	76
		ALT1	TXD	FlexCAN_0	O				
		ALT2	—	—	—				
		ALT3	DEBUG[0]	SSCM	—				
		—	EIRQ[15]	SIUL	I				
B[1]	PCR[17]	ALT0	GPIO[17]	SIUL	I/O	Slow	Medium	50	77
		ALT1	—	—	—				
		ALT2	—	—	—				
		ALT3	DEBUG[1]	SSCM	—				
		—	RXD	FlexCAN_0	I				
		—	EIRQ[16]	SIUL	I				

Table 7. Pin muxing (continued)

Port pin	PCR register	Alternate function ^{(1),(2)}	Functions	Peripheral ⁽³⁾	I/O direction ⁽⁴⁾	Pad speed ⁽⁵⁾		Pin	
						SRC = 0	SRC = 1	64-pin	100-pin
B[2]	PCR[18]	ALT0 ALT1 ALT2 ALT3 —	GPIO[18] TXD — DEBUG[2] EIRQ[17]	SIUL LIN_0 — SSCM SIUL	I/O O — — I	Slow	Medium	51	79
B[3]	PCR[19]	ALT0 ALT1 ALT2 ALT3 —	GPIO[19] — — DEBUG[3] RXD	SIUL — — SSCM LIN_0	I/O — — — I	Slow	Medium	—	80
B[6]	PCR[22]	ALT0 ALT1 ALT2 ALT3 —	GPIO[22] CLKOUT CS2 — EIRQ[18]	SIUL Control DSPI_2 — SIUL	I/O O O — I	Slow	Medium	62	96
B[7]	PCR[23]	ALT0 ALT1 ALT2 ALT3 — —	GPIO[23] — — — AN[0] RXD	SIUL — — — ADC_0 LIN_0	Input only	—	—	20	29
B[8]	PCR[24]	ALT0 ALT1 ALT2 ALT3 — —	GPIO[24] — — — AN[1] ETC[5]	SIUL — — — ADC_0 eTimer_0	Input only	—	—	22	31
B[9]	PCR[25]	ALT0 ALT1 ALT2 ALT3 —	GPIO[25] — — — AN[11]	SIUL — — — ADC_0	Input only	—	—	24	35
B[10]	PCR[26]	ALT0 ALT1 ALT2 ALT3 —	GPIO[26] — — — AN[12]	SIUL — — — ADC_0	Input only	—	—	25	36

Table 7. Pin muxing (continued)

Port pin	PCR register	Alternate function ^{(1),(2)}	Functions	Peripheral ⁽³⁾	I/O direction ⁽⁴⁾	Pad speed ⁽⁵⁾		Pin	
						SRC = 0	SRC = 1	64-pin	100-pin
B[11]	PCR[27]	ALT0 ALT1 ALT2 ALT3 —	GPIO[27] — — — AN[13]	SIUL — — — ADC_0	Input only	—	—	26	37
B[12]	PCR[28]	ALT0 ALT1 ALT2 ALT3 —	GPIO[28] — — — AN[14]	SIUL — — — ADC_0	Input only	—	—	27	38
B[13]	PCR[29]	ALT0 ALT1 ALT2 ALT3 — — —	GPIO[29] — — — AN[6] emu. AN[0] RXD	SIUL — — — ADC_0 emu. ADC_1 ⁽⁶⁾ LIN_1	Input only	—	—	30	42
B[14]	PCR[30]	ALT0 ALT1 ALT2 ALT3 — — — —	GPIO[30] — — — AN[7] emu. AN[1] ETC[4] EIRQ[19]	SIUL — — — ADC_0 emu. ADC_1 ⁽⁶⁾ eTimer_0 SIUL	Input only	—	—	—	44
B[15]	PCR[31]	ALT0 ALT1 ALT2 ALT3 — — —	GPIO[31] — — — AN[8] emu. AN[2] EIRQ[20]	SIUL — — — ADC_0 emu. ADC_1 ⁽⁶⁾ SIUL	Input only	—	—	—	43
Port C (16-bit)									
C[0]	PCR[32]	ALT0 ALT1 ALT2 ALT3 — —	GPIO[32] — — — AN[9] emu. AN[3]	SIUL — — — ADC_0 emu. ADC_1 ⁽⁶⁾	Input only	—	—	—	45

3.3 Absolute maximum ratings

Table 9. Absolute maximum ratings⁽¹⁾

Symbol		Parameter	Conditions	Value		Unit
				Min	Max ⁽²⁾	
V _{SS}	S R	Device ground	—	0	0	V
V _{DD_HV_IOx} ⁽³⁾	S R	3.3 V/5.0 V input/output supply voltage (supply). Code flash memory supply with V _{DD_HV_IO3} and data flash memory with V _{DD_HV_IO2}	—	−0.3	6.0	V
V _{SS_HV_IOx}	S R	3.3 V/5.0 V input/output supply voltage (ground). Code flash memory ground with V _{SS_HV_IO3} and data flash memory with V _{SS_HV_IO2}	—	−0.1	0.1	V
V _{DD_HV_OSC}	S R	3.3 V/5.0 V crystal oscillator amplifier supply voltage (supply)	—	−0.3	6.0	V
			Relative to V _{DD_HV_IOx}	−0.3	V _{DD_HV_IOx} + 0.3	
V _{SS_HV_OSC}	S R	3.3 V/5.0 V crystal oscillator amplifier supply voltage (ground)	—	−0.1	0.1	V
V _{DD_HV_ADC0}	S R	3.3 V/5.0 V ADC_0 supply and high-reference voltage	V _{DD_HV_REG} < 2.7 V	−0.3	V _{DD_HV_REG} + 0.3	V
			V _{DD_HV_REG} > 2.7 V	−0.3	6.0	
V _{SS_HV_ADC0}	S R	3.3 V/5.0 V ADC_0 ground and low-reference voltage	—	−0.1	0.1	V
V _{DD_HV_REG}	S R	3.3 V/5.0 V voltage-regulator supply voltage	—	−0.3	6.0	V
			Relative to V _{DD_HV_IOx}	−0.3	V _{DD_HV_IOx} + 0.3	
TV _{DD}	S R	Slope characteristics on all V _{DD} during power up ⁽⁴⁾ with respect to ground (V _{SS})	—	3.0 ⁽⁵⁾	500 × 10 ³ (0.5 [V/μs])	V/s
V _{DD_LV_CORx}	C C	1.2 V supply pins for core logic (supply)	—	−0.1	1.5	V
V _{SS_LV_CORx}	S R	1.2 V supply pins for core logic (ground)	—	−0.1	0.1	V
V _{IN}	S R	Voltage on any pin with respect to ground (V _{SS_HV_IOx})	—	−0.3	6.0	V
			Relative to V _{DD_HV_IOx}	−0.3	V _{DD_HV_IOx} + 0.3 ⁽⁶⁾	
I _{INJPAD}	S R	Input current on any pin during overload condition	—	−10	10	mA

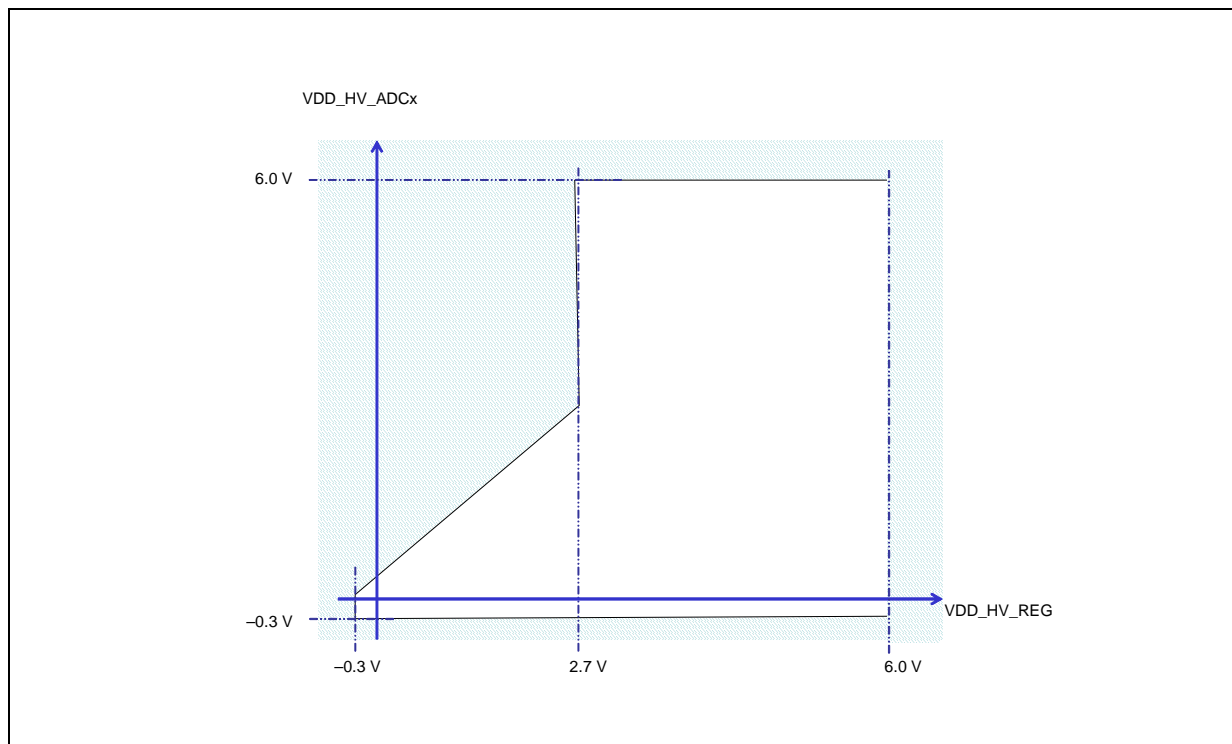


Figure 7. Independent ADC supply ($-0.3\text{ V} \leq V_{DD_HV_REG} \leq 6.0\text{ V}$)

3.4 Recommended operating conditions

Table 10. Recommended operating conditions (5.0 V)

Symbol		Parameter	Conditions	Value		Unit
				Min	Max ⁽¹⁾	
V_{SS}	SR	Device ground	—	0	0	V
$V_{DD_HV_IOx}^{(2)}$	SR	5.0 V input/output supply voltage	—	4.5	5.5	V
$V_{SS_HV_IOx}$	SR	Input/output ground voltage	—	0	0	V
$V_{DD_HV_OSC}$	SR	5.0 V crystal oscillator amplifier supply voltage	—	4.5	5.5	V
			Relative to $V_{DD_HV_IOx}$	$V_{DD_HV_IOx} - 0.1$	$V_{DD_HV_IOx} + 0.1$	
$V_{SS_HV_OSC}$	SR	5.0 V crystal oscillator amplifier reference voltage	—	0	0	V
$V_{DD_HV_REG}$	SR	5.0 V voltage regulator supply voltage	—	4.5	5.5	V
			Relative to $V_{DD_HV_IOx}$	$V_{DD_HV_IOx} - 0.1$	$V_{DD_HV_IOx} + 0.1$	

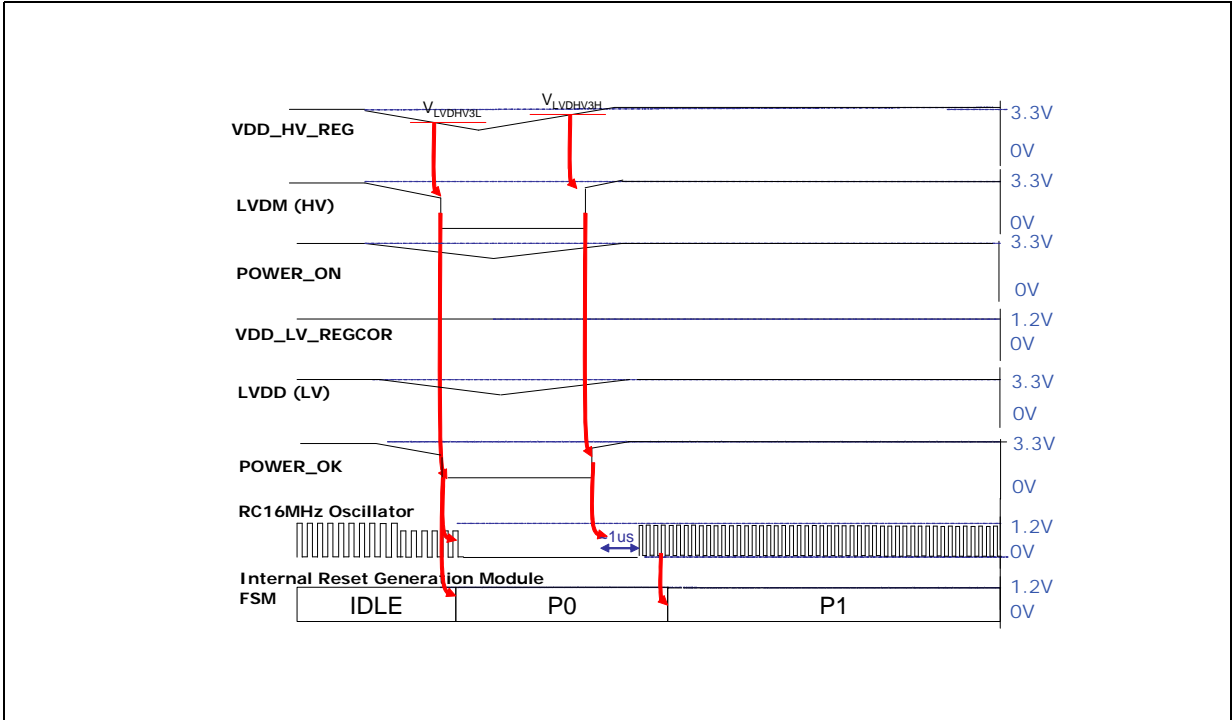
Table 11. Recommended operating conditions (3.3 V) (continued)

Symbol		Parameter	Conditions	Value		Unit
				Min	Max ⁽¹⁾	
V _{DD_HV_REG}	SR	3.3 V voltage regulator supply voltage	—	3.0	3.6	V
			Relative to V _{DD_HV_IOx}	V _{DD_HV_IOx} − 0.1	V _{DD_HV_IOx} + 0.1	
V _{DD_HV_ADC0}	SR	3.3 V ADC_0 supply and high reference voltage	—	3.0	5.5	V
			Relative to V _{DD_HV_REG}	V _{DD_HV_REG} − 0.1	5.5	
V _{SS_HV_ADC0}	SR	ADC_0 ground and low reference voltage	—	0	0	V
V _{DD_LV_REGCOR} ^{(3),(4)}	CC	Internal supply voltage	—	—	—	V
V _{SS_LV_REGCOR} ⁽³⁾	SR	Internal reference voltage	—	0	0	V
V _{DD_LV_CORx} ^{(3),(4)}	CC	Internal supply voltage	—	—	—	V
V _{SS_LV_CORx} ⁽³⁾	SR	Internal reference voltage	—	0	0	V
T _A	SR	Ambient temperature under bias	f _{CPU} = 60 MHz	−40	125	°C
			f _{CPU} = 64 MHz	−40	105	°C

- Full functionality cannot be guaranteed when voltage drops below 3.0 V. In particular, ADC electrical characteristics and I/Os DC electrical specification may not be guaranteed.
- The difference between each couple of voltage supplies must be less than 100 mV, $|V_{DD_HV_IOy} - V_{DD_HV_IOx}| < 100 \text{ mV}$.
- To be connected to emitter of external NPN. Low voltage supplies are not under user control—they are produced by an on-chip voltage regulator—but for the device to function properly the low voltage grounds ($V_{SS_LV_xxx}$) must be shorted to high voltage grounds ($V_{SS_HV_xxx}$) and the low voltage supply pins ($V_{DD_LV_xxx}$) must be connected to the external ballast emitter.
- The low voltage supplies ($V_{DD_LV_xxx}$) are not all independent.
 - $V_{DD_LV_COR1}$ and $V_{DD_LV_COR2}$ are shorted internally via double bonding connections with lines that provide the low voltage supply to the data flash memory module. Similarly, $V_{SS_LV_COR1}$ and $V_{SS_LV_COR2}$ are internally shorted.
 - $V_{DD_LV_REGCOR}$ and $V_{DD_LV_REGCORx}$ are physically shorted internally, as are $V_{SS_LV_REGCOR}$ and $V_{SS_LV_CORx}$.

Figure 8 shows the constraints of the different power supplies.

Figure 13. Brown-out typical sequence



3.10 DC electrical characteristics

3.10.1 NVUSRO register

Portions of the device configuration, such as high voltage supply and watchdog enable/disable after reset are controlled via bit values in the non-volatile user options (NVUSRO) register.

For a detailed description of the NVUSRO register, please refer to the device reference manual.

NVUSRO[PAD3V5V] field description

The DC electrical characteristics are dependent on the PAD3V5V bit value. [Table 18](#) shows how NVUSRO[PAD3V5V] controls the device configuration.

Table 18. PAD3V5V field description

Value ⁽¹⁾	Description
0	High voltage supply is 5.0 V
1	High voltage supply is 3.3 V

1. Default manufacturing value before flash initialization is '1' (3.3 V).

Table 20. Supply current (5.0 V, NVUSRO[PAD3V5V] = 0)

Symbol	C	Parameter	Conditions	Value ⁽¹⁾		Unit
				Typ	Max	
I _{DD_LV_CORx}	T	RUN—Maximum mode ⁽²⁾	V _{DD_LV_CORx} externally forced at 1.3 V	40 MHz	44	55
	P			64 MHz	52	65
	T	RUN—Typical mode ⁽³⁾		40 MHz	38	46
	P			64 MHz	45	54
	P	HALT mode ⁽⁴⁾		—	1.5	10
		STOP mode ⁽⁵⁾		—	1	10
I _{DD_FLASH}	T	Flash during read	V _{DD_HV_FL} at 5.0 V	—	8	10
		Flash during erase operation on 1 flash module	V _{DD_HV_FL} at 5.0 V	—	15	19
I _{DD_ADC}	T	ADC	V _{DD_HV_ADC0} at 5.0 V f _{ADC} = 16 MHz	ADC_0	3	4
I _{DD_OSC}	T	Oscillator	V _{DD_HV_OSC} at 5.0 V	8 MHz	2.6	3.2
I _{DD_HV_REG}	D	Internal regulator module current consumption	V _{DD_HV_REG} at 5.5 V		—	10

1. All values to be confirmed after characterization/data collection.
2. Maximum mode: FlexPWM, ADC, CTU, DSPI, LINFlex, FlexCAN, 15 output pins, PLL_0 enabled, 125 °C ambient. I/O supply current excluded.
3. Typical mode configurations: DSPI, LINFlex, FlexCAN, 15 output pins, PLL_0, 105 °C ambient. I/O supply current excluded.
4. Halt mode configurations: Code fetched from SRAM, code flash memory and data flash memory in low power mode, OSC/PLL_0 are OFF, core clock frozen, all peripherals disabled.
5. STOP "P" mode Device Under Test (DUT) configuration: Code fetched from SRAM, code flash memory and data flash memory off, OSC/PLL_0 are OFF, core clock frozen, all peripherals disabled.

In particular two different transient periods can be distinguished:

- A first and quick charge transfer from the internal capacitance C_{P1} and C_{P2} to the sampling capacitance C_S occurs (C_S is supposed initially completely discharged): considering a worst case (since the time constant in reality would be faster) in which C_{P2} is reported in parallel to C_{P1} (call $C_P = C_{P1} + C_{P2}$), the two capacitances C_P and C_S are in series, and the time constant is

Equation 5

$$\tau_1 = (R_{SW} + R_{AD}) \cdot \frac{C_P \cdot C_S}{C_P + C_S}$$

[Equation 5](#) can again be simplified considering only C_S as an additional worst condition. In reality, the transient is faster, but the A/D converter circuitry has been designed to be robust also in the very worst case: the sampling time T_S is always much longer than the internal time constant:

Equation 6

$$\tau_1 < (R_{SW} + R_{AD}) \cdot C_S \ll T_S$$

The charge of C_{P1} and C_{P2} is redistributed also on C_S , determining a new value of the voltage V_{A1} on the capacitance according to [Equation 7](#):

Equation 7

$$V_{A1} \cdot (C_S + C_{P1} + C_{P2}) = V_A \cdot (C_{P1} + C_{P2})$$

- A second charge transfer involves also C_F (that is typically bigger than the on-chip capacitance) through the resistance R_L : again considering the worst case in which C_{P2} and C_S were in parallel to C_{P1} (since the time constant in reality would be faster), the time constant is:

Equation 8

$$\tau_2 < R_L \cdot (C_S + C_{P1} + C_{P2})$$

In this case, the time constant depends on the external circuit: in particular imposing that the transient is completed well before the end of sampling time T_S , a constraints on R_L sizing is obtained:

Equation 9

$$8.5 \cdot \tau_2 = 8.5 \cdot R_L \cdot (C_S + C_{P1} + C_{P2}) < T_S$$

Of course, R_L shall be sized also according to the current limitation constraints, in combination with R_S (source impedance) and R_F (filter resistance). Being C_F definitively bigger than C_{P1} , C_{P2} and C_S , then the final voltage V_{A2} (at the end of the charge transfer transient) will be much higher than V_{A1} . [Equation 10](#) must be respected (charge balance assuming now C_S already charged at V_{A1}):

Equation 10

$$V_{A2} \cdot (C_S + C_{P1} + C_{P2} + C_F) = V_A \cdot C_F + V_{A1} \cdot (C_{P1} + C_{P2} + C_S)$$

3.14.2 ADC conversion characteristics

Table 30. ADC conversion characteristics

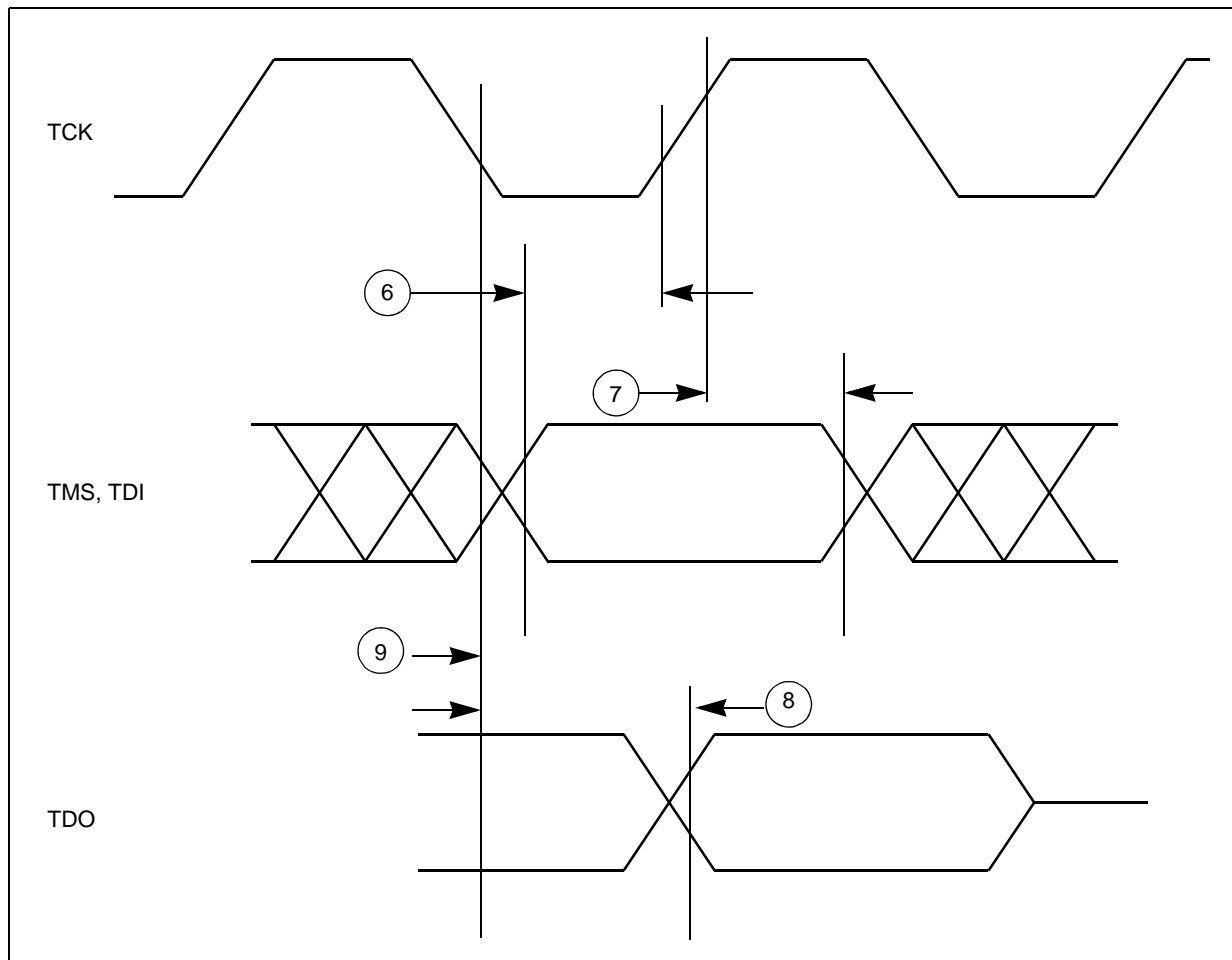
Symbol	C	Parameter	Conditions ⁽¹⁾	Value			Unit
				Min	Typ	Max	
f_{CK}	S R	—	ADC clock frequency (depends on ADC configuration) (The duty cycle depends on ADC clock ⁽²⁾ frequency)	—	—	—	—
f_s	S R	—	Sampling frequency	—	—	1.53	MHz
t_s	—	D	Sampling time ⁽⁴⁾	$f_{ADC} = 20 \text{ MHz}$, INPSAMP = 3	125	—	ns
				$f_{ADC} = 9 \text{ MHz}$, INPSAMP = 255	—	—	28.2 μ s
t_c	—	P	Conversion time ⁽⁵⁾	$f_{ADC} = 20 \text{ MHz}$ ⁽⁶⁾ , INPCMP = 1	0.65 0	—	μ s
$t_{ADC_P_U}$	S R	—	ADC power-up delay (time needed for ADC to settle exiting from software power down; PWDN bit = 0)	—	—	1.5	μ s
C_S ⁽⁷⁾	—	D	ADC input sampling capacitance	—	—	2.5	pF
C_{P1} ⁽⁷⁾	—	D	ADC input pin capacitance 1	—	—	3	pF
C_{P2} ⁽⁷⁾	—	D	ADC input pin capacitance 2	—	—	1	pF
R_{SW} ⁽⁷⁾	—	D	Internal resistance of analog source	$V_{DD_HV_ADC0} = 5 \text{ V} \pm 10\%$	—	—	0.6 k Ω
				$V_{DD_HV_ADC0} = 3.3 \text{ V} \pm 10\%$	—	—	3 k Ω
R_{AD} ⁽⁷⁾	—	D	Internal resistance of analog source	—	—	2	k Ω
I_{INJ}	—	T	Input current injection	Current injection on one ADC input, different from the converted one. Remains within TUE specification	—5	—	5 mA
INL	C C	P	Integral non-linearity	No overload	—1.5	—	1.5 LSB
DNL	C C	P	Differential non-linearity	No overload	—1.0	—	1.0 LSB
E_O	C C	T	Offset error	—	—	± 1	LSB
E_G	C C	T	Gain error	—	—	± 1	LSB
TUE	C C	P	Total unadjusted error without current injection	—	—2.5	—	2.5 LSB
TUE	C C	T	Total unadjusted error with current injection	—	—3	—	3 LSB

1. $V_{DD} = 3.3 \text{ V}$ to 3.6 V / 4.5 V to 5.5 V , $T_A = -40^\circ\text{C}$ to $T_{A \text{ MAX}}$, unless otherwise specified and analog input voltage from $V_{SS_HV_ADC0}$ to $V_{DD_HV_ADC0}$.

2. AD_clk clock is always half of the ADC module input clock defined via the auxiliary clock divider for the ADC.

3. When configured to allow 60 MHz ADC, the minimum ADC clock speed is 9 MHz, below which the precision is lost.

Figure 27. Nexus TDI, TMS, TDO timing



3.17.4 External interrupt timing (IRQ pin)

Table 40. External interrupt timing⁽¹⁾

No.	Symbol	C	D	Parameter	Conditions	Value		Unit
						Min	Max	
1	t_{IPWL}	CC	D	IRQ pulse width low	—	4	—	t_{CYC}
2	t_{IPWH}	CC	D	IRQ pulse width high	—	4	—	t_{CYC}
3	t_{ICYC}	CC	D	IRQ edge to edge time ⁽²⁾	—	$4 + N$ (3)	—	t_{CYC}

1. IRQ timing specified at $f_{SYS} = 64$ MHz and $V_{DD_HV_IOx} = 3.0$ V to 5.5 V, $T_A = T_L$ to T_H , and $C_L = 200$ pF with $SRC = 0b00$

2. Applies when IRQ pins are configured for rising edge or falling edge events, but not both.

3. N = ISR time to clear the flag

Table 41. DSPI timing⁽¹⁾ (continued)

No.	Symbol	C	Parameter	Conditions	Value		Unit
					Min	Max	
11	t_{SUO}	CC	D	Data valid (after SCK edge)			ns
				Master (MTFE = 0)	—	12	
				Slave	—	36	
				Master (MTFE = 1, CPHA = 0)	—	12	
12	t_{HO}	CC	D	Data hold time for outputs			ns
				Master (MTFE = 0)	-2	—	
				Slave	6	—	
				Master (MTFE = 1, CPHA = 0)	6	—	
				Master (MTFE = 1, CPHA = 1)	-2	—	

1. All timing are provided with 50 pF capacitance on output, 1 ns transition time on input signal

Figure 29. DSPI classic SPI timing – Master, CPHA = 0

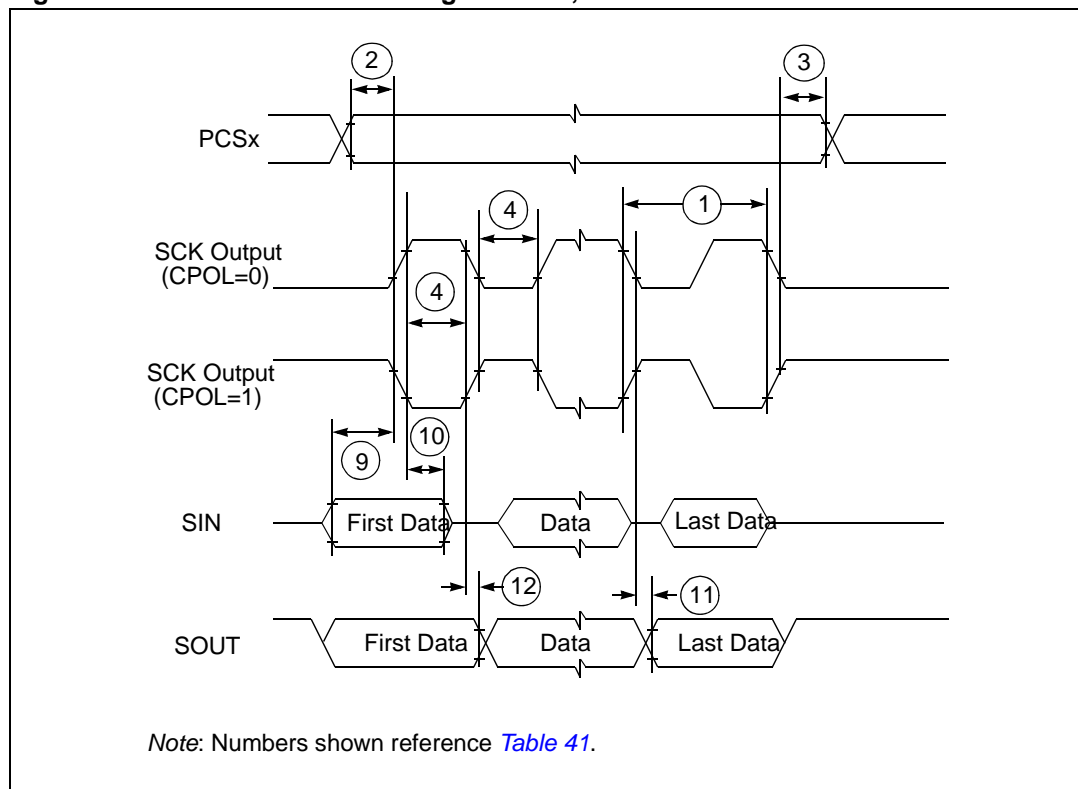


Table 42. LQFP100 package mechanical data

Symbol	Dimensions					
	mm			inches ⁽¹⁾		
	Min	Typ	Max	Min	Typ	Max
A	—	—	1.600	—	—	0.0630
A1	0.050	—	0.150	0.0020	—	0.0059
A2	1.350	1.400	1.450	0.0531	0.0551	0.0571
b	0.170	0.220	0.270	0.0067	0.0087	0.0106
c	0.090	—	0.200	0.0035	—	0.0079
D	15.800	16.000	16.200	0.6220	0.6299	0.6378
D1	13.800	14.000	14.200	0.5433	0.5512	0.5591
D3	—	12.000	—	—	0.4724	—
E	15.800	16.000	16.200	0.6220	0.6299	0.6378
E1	13.800	14.000	14.200	0.5433	0.5512	0.5591
E3	—	12.000	—	—	0.4724	—
e	—	0.500	—	—	0.0197	—
L	0.450	0.600	0.750	0.0177	0.0236	0.0295
L1	—	1.000	—	—	0.0394	—
k	0.0°	3.5°	7.0°	0.0°	3.5°	7.0°
ccc ⁽²⁾	0.08			0.0031		

1. Values in inches are converted from millimeters (mm) and rounded to four decimal digits.

2. Tolerance

4.2.2 LQFP64 mechanical outline drawing

Figure 39. LQFP64 package mechanical drawing

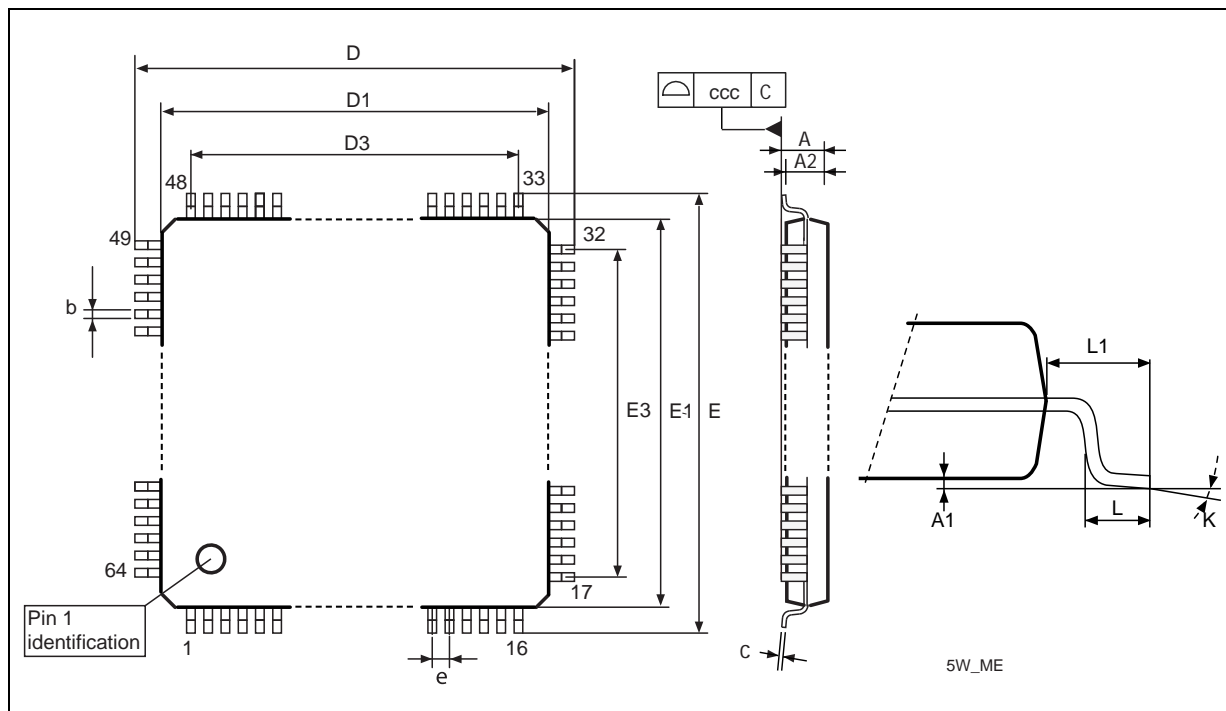


Table 43. LQFP64 package mechanical data

Symbol	Dimensions					
	mm			inches ⁽¹⁾		
	Min	Typ	Max	Min	Typ	Max
A	—	—	1.6	—	—	0.063
A1	0.05	—	0.15	0.002	—	0.0059
A2	1.35	1.4	1.45	0.0531	0.0551	0.0571
b	0.17	0.22	0.27	0.0067	0.0087	0.0106
c	0.09	—	0.2	0.0035	—	0.0079
D	11.8	12	12.2	0.4646	0.4724	0.4803
D1	9.8	10	10.2	0.3858	0.3937	0.4016
D3	—	7.5	—	—	0.2953	—
E	11.8	12	12.2	0.4646	0.4724	0.4803
E1	9.8	10	10.2	0.3858	0.3937	0.4016
E3	—	7.5	—	—	0.2953	—
e	—	0.5	—	—	0.0197	—
L	0.45	0.6	0.75	0.0177	0.0236	0.0295
L1	—	1	—	—	0.0394	—

Table 43. LQFP64 package mechanical data (continued)

Symbol	Dimensions					
	mm			inches ⁽¹⁾		
	Min	Typ	Max	Min	Typ	Max
k	0.0°	3.5°	7.0°	0.0°	3.5°	7.0°
ccc ⁽²⁾	0.08			0.0031		

1. Values in inches are converted from millimeters (mm) and rounded to four decimal digits.

2. Tolerance

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