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Understanding Embedded - DSP (Digital Signal Processors)

Embedded - DSP (Digital Signal Processors) are specialized microprocessors designed to perform complex mathematical computations on digital signals in real-time. Unlike general-purpose processors, DSPs are optimized for high-speed numeric processing tasks, making them ideal for applications that require efficient and precise manipulation of digital data. These processors are fundamental in converting and processing signals in various forms, including audio, video, and communication signals, ensuring that data is accurately interpreted and utilized in embedded systems.

Applications of Embedded - DSP (Digital Signal Processors)

Details	
Product Status	Obsolete
Type	SC140 Core
Interface	Ethernet, I²C, TDM, UART
Clock Rate	400MHz
Non-Volatile Memory	External
On-Chip RAM	1.436MB
Voltage - I/O	3.30V
Voltage - Core	1.10V
Operating Temperature	-40°C ~ 105°C (TJ)
Mounting Type	Surface Mount
Package / Case	431-BFBGA, FCBGA
Supplier Device Package	431-FCPBGA (20x20)
Purchase URL	https://www.e-xfl.com/product-detail/nxp-semiconductors/kmc8113tmp4800v

Table of Contents

List of Figures

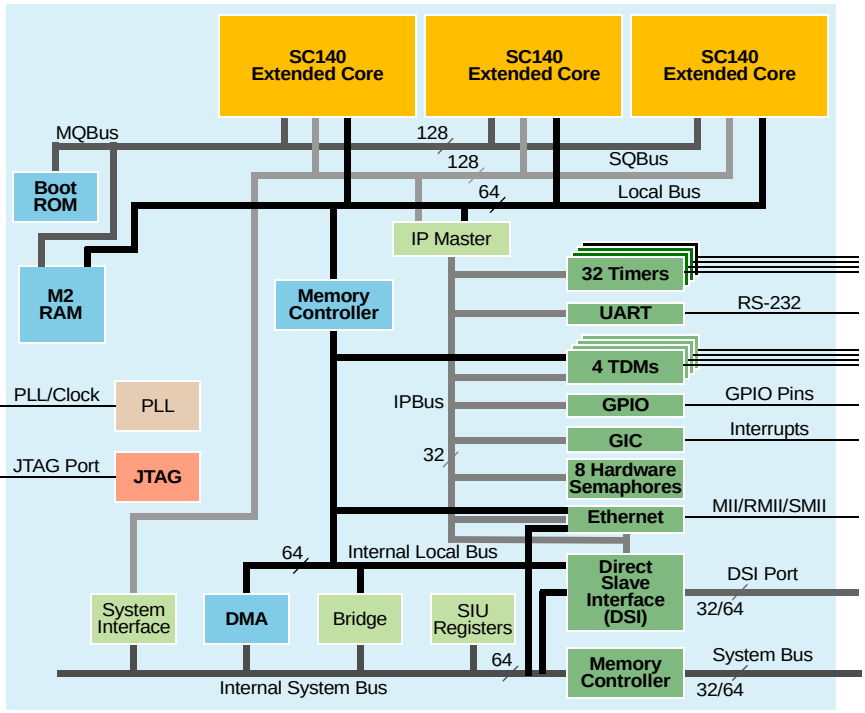
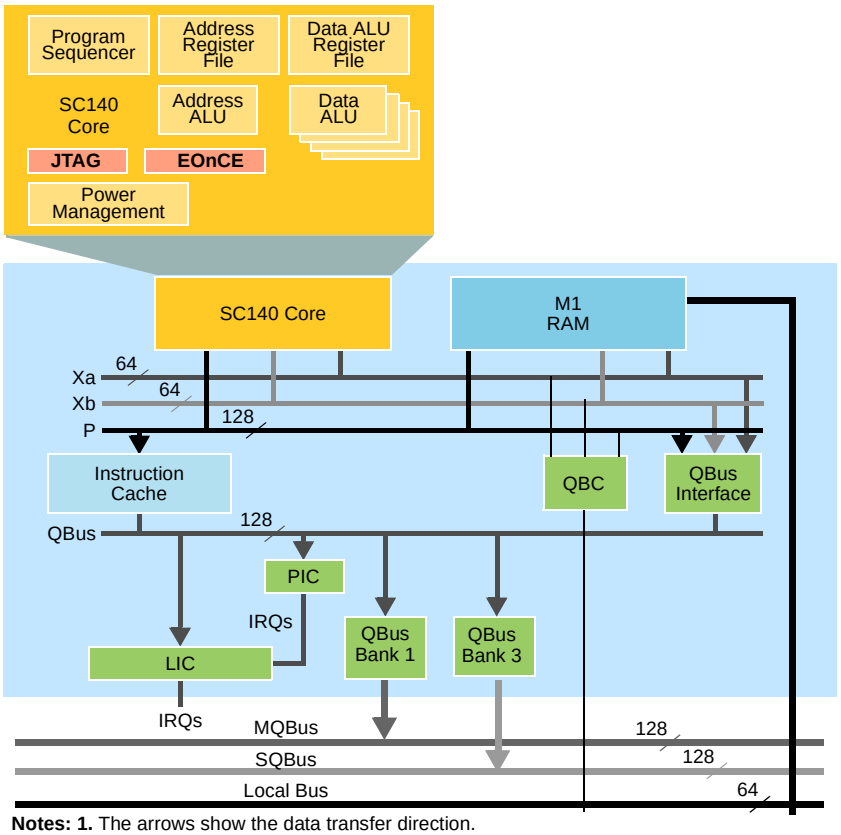


Figure 1. MSC8113 Block Diagram



Notes: **1.** The arrows show the data transfer direction.
2. The QBus interface includes a bus switch, write buffer, fetch unit, and a control unit that defines four QBus banks. In addition, the QBC handles internal memory contentions.

Figure 2. StarCore® SC140 DSP Extended Core Block Diagram

1.2 Signal List By Ball Location

Table 1 presents signal list sorted by ball number. -

Table 1. MSC8113 Signal Listing by Ball Designator

Des.	Signal Name	Des.	Signal Name
B3	V_{DD}	C18	GPIO1/TIMER0/CHIP_ID1/$\overline{IRQ5}$/ETHTXD1
B4	GND	C19	GPIO7/TDM3RCLK/$\overline{IRQ5}$/ETHTXD3
B5	GND	C20	GPIO3/TDM3TSYN/$\overline{IRQ1}$/ETHTXD2
B6	$\overline{NMI\_OUT}$	C21	GPIO5/TDM3TDAT/$\overline{IRQ3}$/ETHRXD3
B7	GND	C22	GPIO6/TDM3RSYN/$\overline{IRQ4}$/ETHRXD2
B8	V_{DD}	D2	TDI
B9	GND	D3	EE0
B10	V_{DD}	D4	EE1
B11	GND	D5	GND
B12	V_{DD}	D6	V_{DDH}
B13	GND	D7	HCID2
B14	V_{DD}	D8	HCID3/HA8
B15	GND	D9	GND
B16	V_{DD}	D10	V_{DD}
B17	GND	D11	GND
B18	V_{DD}	D12	V_{DD}
B19	GPIO0/CHIP_ID0/$\overline{IRQ4}$/ETHTXD0	D13	GND
B20	V_{DD}	D14	V_{DD}
B21	V_{DD}	D15	V_{DD}
B22	GND	D16	GPIO31/TIMER3/SCL
C2	GND	D17	GPIO29/CHIP_ID3/ETHTX_EN
C3	V_{DD}	D18	V_{DDH}
C4	TDO	D19	GPIO4/TDM3TCLK/$\overline{IRQ2}$/ETHTX_ER
C5	$\overline{SRESET}$	D20	V_{DDH}
C6	GPIO28/UTXD/DREQ2	D21	GND
C7	HCID1	D22	GPIO8/TDM3RDAT/$\overline{IRQ6}$/ETHCOL
C8	GND	E2	TCK
C9	V_{DD}	E3	$\overline{TRST}$
C10	GND	E4	TMS
C11	V_{DD}	E5	$\overline{HRESET}$
C12	GND	E6	GPIO27/URXD/DREQ1
C13	V_{DD}	E7	HCID0
C14	GND	E8	GND
C15	GND	E9	V_{DD}
C16	GPIO30/TIMER2/TMCLK/SDA	E10	GND
C17	GPIO2/TIMER1/CHIP_ID2/$\overline{IRQ6}$	E11	V_{DD}

Table 1. MSC8113 Signal Listing by Ball Designator (continued)

Des.	Signal Name	Des.	Signal Name
M15	V_{DDH}	P12	V_{CCSYN}
M16	$\overline{\text{HBRST}}$	P13	GND
M17	V_{DDH}	P14	GND
M18	V_{DDH}	P15	$\overline{\text{TA}}$
M19	GND	P16	$\overline{\text{BR}}$
M20	V_{DDH}	P17	$\overline{\text{TEA}}$
M21	A24	P18	$\overline{\text{PSDVAL}}$
M22	A21	P19	DP0/DREQ1/$\overline{\text{EXT\_BR2}}$
N2	HD26	P20	V_{DDH}
N3	HD30	P21	GND
N4	HD29	P22	A19
N5	HD24	R2	HD18
N6	$\overline{\text{PWE2}}$/$\overline{\text{PSDDQM2}}$/$\overline{\text{PBS2}}$	R3	V_{DDH}
N7	V_{DDH}	R4	GND
N8	$\overline{\text{HWBS0}}$/$\overline{\text{HDBS0}}$/$\overline{\text{HWBE0}}$/$\overline{\text{HDBE0}}$	R5	HD22
N9	$\overline{\text{HBCS}}$	R6	$\overline{\text{HWBS6}}$/$\overline{\text{HDBS6}}$/$\overline{\text{HWBE6}}$/$\overline{\text{HDBE6}}$/$\overline{\text{PWE6}}$/$\overline{\text{PSDDQM6}}$/$\overline{\text{PBS6}}$
N10	GND	R7	$\overline{\text{HWBS4}}$/$\overline{\text{HDBS4}}$/$\overline{\text{HWBE4}}$/$\overline{\text{HDBE4}}$/$\overline{\text{PWE4}}$/$\overline{\text{PSDDQM4}}$/$\overline{\text{PBS4}}$
N14	GND	R8	TSZ1
N15	$\overline{\text{HRDS}}$/HRW/$\overline{\text{HRDE}}$	R9	TSZ3
N16	$\overline{\text{BG}}$	R10	$\overline{\text{IRQ1}}$/$\overline{\text{GBL}}$
N17	$\overline{\text{HCS}}$	R11	V_{DD}
N18	$\overline{\text{CS0}}$	R12	V_{DD}
N19	$\overline{\text{PSDWE}}$/PGPL1	R13	V_{DD}
N20	GPIO26/TDM0RDAT	R14	TT0/HA7
N21	A23	R15	$\overline{\text{IRQ7}}$/DP7/DREQ4
N22	A20	R16	$\overline{\text{IRQ6}}$/DP6/DREQ3
P2	HD20	R17	$\overline{\text{IRQ3}}$/DP3/DREQ2/$\overline{\text{EXT\_BR3}}$
P3	HD27	R18	$\overline{\text{TS}}$
P4	HD25	R19	$\overline{\text{IRQ2}}$/DP2/$\overline{\text{DACK2}}$/$\overline{\text{EXT\_DBG2}}$
P5	HD23	R20	A17
P6	$\overline{\text{HWBS3}}$/$\overline{\text{HDBS3}}$/$\overline{\text{HWBE3}}$/$\overline{\text{HDBE3}}$	R21	A18
P7	$\overline{\text{HWBS2}}$/$\overline{\text{HDBS2}}$/$\overline{\text{HWBE2}}$/$\overline{\text{HDBE2}}$	R22	A16
P8	$\overline{\text{HWBS1}}$/$\overline{\text{HDBS1}}$/$\overline{\text{HWBE1}}$/$\overline{\text{HDBE1}}$	T2	HD17
P9	HCLKIN	T3	HD21
P10	GND	T4	HD1/DSISYNC
P11	GND_{SYN}	T5	HD0/SWTE

Table 1. MSC8113 Signal Listing by Ball Designator (continued)

Des.	Signal Name	Des.	Signal Name
T6	HWBS7/HDBS7/HWBE7/HDBE7/PWE7/PSDDQM7/PBS7	U21	A12
T7	HWBS5/HDBS5/HWBE5/HDBE5/PWE5/PSDDQM5/PBS5	U22	A13
T8	TSZ0	V2	HD3/MODCK1
T9	TSZ2	V3	V_{DDH}
T10	TBST	V4	GND
T11	V_{DD}	V5	D0
T12	D16	V6	D1
T13	TT1	V7	D4
T14	D21	V8	D5
T15	D23	V9	D7
T16	IRQ5/DP5/DACK4/EXT_BG3	V10	D10
T17	IRQ4/DP4/DACK3/EXT_DBG3	V11	D12
T18	IRQ1/DP1/DACK1/EXT_BG2	V12	D13
T19	D30	V13	D18
T20	GND	V14	D20
T21	A15	V15	GND
T22	A14	V16	D24
U2	HD16	V17	D27
U3	HD19	V18	D29
U4	HD2/DSI64	V19	A8
U5	D2	V20	A9
U6	D3	V21	A10
U7	D6	V22	A11
U8	D8	W2	HD6
U9	D9	W3	HD5/CNFGS
U10	D11	W4	HD4/MODCK2
U11	D14	W5	GND
U12	D15	W6	GND
U13	D17	W7	V_{DDH}
U14	D19	W8	V_{DDH}
U15	D22	W9	GND
U16	D25	W10	HDST1/HA10
U17	D26	W11	HDST0/HA9
U18	D28	W12	V_{DDH}
U19	D31	W13	GND
U20	V_{DDH}	W14	HD40/D40/ETHRXD0

2.2 Recommended Operating Conditions

Table 3 lists recommended operating conditions. Proper device operation outside of these conditions is not guaranteed.

Table 3. Recommended Operating Conditions

Rating	Symbol	Value	Unit
Core and PLL supply voltage:	V_{DD} V_{CCSYN}	1.07 to 1.13	V
I/O supply voltage	V_{DDH}	3.135 to 3.465	V
Input voltage	V_{IN}	–0.2 to V_{DDH}+0.2	V
Operating temperature range:	T_J	–40 to 105	°C

2.3 Thermal Characteristics

Table 4 describes thermal characteristics of the MSC8113 for the FC-PBGA packages.

Table 4. Thermal Characteristics for the MSC8113

Characteristic	Symbol	FC-PBGA 20 × 20 mm[5]		Unit
		Natural Convection	200 ft/min (1 m/s) airflow	
Junction-to-ambient[1, 2]	$R_{\theta JA}$	26	21	°C/W
Junction-to-ambient, four-layer board[1, 3]	$R_{\theta JA}$	19	15	°C/W
Junction-to-board (bottom)[4]	$R_{\theta JB}$	9		°C/W
Junction-to-case[5]	$R_{\theta JC}$	0.9		°C/W
Junction-to-package-top[6]	Ψ_{JT}	1		°C/W

Notes:
1. Junction temperature is a function of die size, on-chip power dissipation, package thermal resistance, mounting site (board) temperature, ambient temperature, air flow, power dissipation of other components on the board, and board thermal resistance.
2. Per SEMI G38-87 and JEDEC JESD51-2 with the single layer board horizontal.
3. Per JEDEC JESD51-6 with the board horizontal.
4. Thermal resistance between the die and the printed circuit board per JEDEC JESD 51-8. Board temperature is measured on the top surface of the board near the package.
5. Thermal resistance between the die and the case top surface as measured by the cold plate method (MIL SPEC-883 Method 1012.1).
6. Thermal characterization parameter indicating the temperature difference between package top and the junction temperature per JEDEC JESD51-2.

Section 3.5, *Thermal Considerations* provides a detailed explanation of these characteristics.

2.4 DC Electrical Characteristics

This section describes the DC electrical characteristics for the MSC8113. The measurements in **Table 5** assume the following system conditions:

- T_A = 25 °C
- V_{DD} = 1.1 V nominal = 1.07–1.13 V_{DC}
- V_{DDH} = 3.3 V ± 5% V_{DC}
- GND = 0 V_{DC}

Note: The leakage current is measured for nominal V_{DDH} and V_{DD}.

Table 7. Maximum Frequencies

Characteristic	Maximum in MHz
Internal bus frequency (BLCK)	100/133
DSI clock frequency (HCLKIN) • Core frequency = 300 MHz • Core frequency = 400 MHz	 HCLKIN ≤ (min{70 MHz, CLKOUT}) HCLKIN ≤ (min{100 MHz, CLKOUT})
External clock frequency (CLKIN or CLKOUT)	100/133

Table 8. Clock Frequencies

Characteristics	Symbol	300 MHz Device		400 MHz Device	
		Min	Max	Min	Max
CLKIN frequency	F_{CLKIN}	20	100	20	133.3
BCLK frequency	F_{BCLK}	40	100	40	133.3
Reference clock (REFCLK) frequency	F_{REFCLK}	40	100	40	133.3
Output clock (CLKOUT) frequency	F_{CLKOUT}	40	100	40	133.3
SC140 core clock frequency	F_{CORE}	200	300	200	400
Note: The rise and fall time of external clocks should be 3 ns maximum					

Table 9. System Clock Parameters

Characteristic	Min	Max	Unit
Phase jitter between BCLK and CLKIN	—	0.3	ns
CLKIN frequency	20	see **Table 8**	MHz
CLKIN slope	—	3	ns
PLL input clock (after predivider)	20	100	MHz
PLL output frequency (VCO output) • 300 MHz core • 400 MHz core	800	 1200 1600	MHz MHz MHz
CLKOUT frequency jitter[1]	—	200	ps
CLKOUT phase jitter[1] with CLKIN phase jitter of ±100 ps.	—	500	ps
Notes: **1.** Peak-to-peak. **2.** Not tested. Guaranteed by design.			

2.5.4 Reset Timing

The MSC8113 has several inputs to the reset logic:

- Power-on reset ($\overline{\text{PORESET}}$)
- External hard reset ($\overline{\text{HRESET}}$)
- External soft reset ($\overline{\text{SRESET}}$)
- Software watchdog reset
- Bus monitor reset
- Host reset command through JTAG

All MSC8113 reset sources are fed into the reset controller, which takes different actions depending on the source of the reset. The reset status register indicates the most recent sources to cause a reset. **Table 10** describes the reset sources.

Table 10. Reset Sources

Name	Direction	Description
Power-on reset (PORESET)	Input	Initiates the power-on reset flow that resets the MSC8113 and configures various attributes of the MSC8113. On PORESET, the entire MSC8113 device is reset. SPLL states is reset, HRESET and SRESET are driven, the SC140 extended cores are reset, and system configuration is sampled. The clock mode (MODCK bits), reset configuration mode, boot mode, Chip ID, and use of either a DSI 64 bits port or a System Bus 64 bits port are configured only when PORESET is asserted.
External hard reset (HRESET)	Input/ Output	Initiates the hard reset flow that configures various attributes of the MSC8113. While HRESET is asserted, SRESET is also asserted. HRESET is an open-drain pin. Upon hard reset, HRESET and SRESET are driven, the SC140 extended cores are reset, and system configuration is sampled. The most configurable features are reconfigured. These features are defined in the 32-bit hard reset configuration word described in *Hard Reset Configuration Word* section of the *Reset* chapter in the *MSC8113 Reference Manual*.
External soft reset (SRESET)	Input/ Output	Initiates the soft reset flow. The MSC8113 detects an external assertion of SRESET only if it occurs while the MSC8113 is not asserting reset. SRESET is an open-drain pin. Upon soft reset, SRESET is driven, the SC140 extended cores are reset, and system configuration is maintained.
Software watchdog reset	Internal	When the MSC8113 watchdog count reaches zero, a software watchdog reset is signalled. The enabled software watchdog event then generates an internal hard reset sequence.
Bus monitor reset	Internal	When the MSC8113 bus monitor count reaches zero, a bus monitor hard reset is asserted. The enabled bus monitor event then generates an internal hard reset sequence.
Host reset command through the TAP	Internal	When a host reset command is written through the Test Access Port (TAP), the TAP logic asserts the soft reset signal and an internal soft reset sequence is generated.

Table 11 summarizes the reset actions that occur as a result of the different reset sources.

Table 11. Reset Actions for Each Reset Source

Reset Action/Reset Source	Power-On Reset (PORESET)	Hard Reset (HRESET)	Soft Reset (SRESET)	
	External only	External or Internal (Software Watchdog or Bus Monitor)	External	JTAG Command: EXTEST, CLAMP, or HIGHZ
Configuration pins sampled (Refer to **Section 2.5.4.1** for details)**.**	Yes	No	No	No
SPLL state reset	Yes	No	No	No
System reset configuration write through the DSI	Yes	No	No	No
System reset configuration write though the system bus	Yes	Yes	No	No
HRESET driven	Yes	Yes	No	No
SIU registers reset	Yes	Yes	No	No
IPBus modules reset (TDM, UART, Timers, DSI, IPBus master, GIC, HS, and GPIO)	Yes	Yes	Yes	Yes
SRESET driven	Yes	Yes	Yes	Depends on command
SC140 extended cores reset	Yes	Yes	Yes	Yes
MQBS reset	Yes	Yes	Yes	Yes

2.5.4.1 Power-On Reset (PORESET) Pin

Asserting PORESET initiates the power-on reset flow. PORESET must be asserted externally for at least 16 CLKIN cycles after V_{DD} and V_{DDH} are both at their nominal levels.

2.5.4.2 Reset Configuration

The MSC8113 has two mechanisms for writing the reset configuration:

- Through the direct slave interface (DSI)
- Through the system bus. When the reset configuration is written through the system bus, the MSC8113 acts as a configuration master or a configuration slave. If configuration slave is selected, but no special configuration word is written, a default configuration word is applied.

Fourteen signal levels (see **Chapter 1** for signal description details) are sampled on $\overline{\text{PORESET}}$ deassertion to define the Reset Configuration Mode and boot and operating conditions:

- $\overline{\text{RSTCONF}}$
- CNFGS
- DSISYNC
- DSI64
- CHIP_ID[0–3]
- BM[0–2]
- SWTE
- MODCK[1–2]

2.5.4.3 Reset Timing Tables

Table 12 and **Figure 9** describe the reset timing for a reset configuration write through the direct slave interface (DSI) or through the system bus.

Table 12. Timing for a Reset Configuration Write through the DSI or System Bus

No.	Characteristics	Expression	Min	Max	Unit
1	Required external $\overline{\text{PORESET}}$ duration minimum • CLKIN = 20 MHz • CLKIN = 100 MHz (300 MHz core) • CLKIN = 133 MHz (400 MHz core)	16/CLKIN	 800 160 120	 — — —	 ns ns ns
2	Delay from deassertion of external $\overline{\text{PORESET}}$ to deassertion of internal $\overline{\text{PORESET}}$ • CLKIN = 20 MHz to 133 MHz	1024/CLKIN	 6.17	 51.2	 µs
3	Delay from de-assertion of internal $\overline{\text{PORESET}}$ to SPLL lock • CLKIN = 20 MHz (RDF = 1) • CLKIN = 100 MHz (RDF = 1) (300 MHz core) • CLKIN = 133 MHz (RDF = 2) (400 MHz core)	6400/(CLKIN/RDF) (PLL reference clock-division factor)	 320 64 96	 320 64 96	 µs µs µs
5	Delay from SPLL to $\overline{\text{HRESET}}$ deassertion • REFCLK = 40 MHz to 133 MHz	512/REFCLK	3.08	12.8	µs
6	Delay from SPLL lock to $\overline{\text{SRESET}}$ deassertion • REFCLK = 40 MHz to 133 MHz	515/REFCLK	3.10	12.88	µs
7	Setup time from assertion of $\overline{\text{RSTCONF}}$, CNFGS, DSISYNC, DSI64, CHIP_ID[0–3], BM[0–2], SWTE, and MODCK[1–2] before deassertion of $\overline{\text{PORESET}}$		3	—	ns
8	Hold time from deassertion of $\overline{\text{PORESET}}$ to deassertion of $\overline{\text{RSTCONF}}$, CNFGS, DSISYNC, DSI64, CHIP_ID[0–3], BM[0–2], SWTE, and MODCK[1–2]		5	—	ns
Note:	Timings are not tested, but are guaranteed by design.				

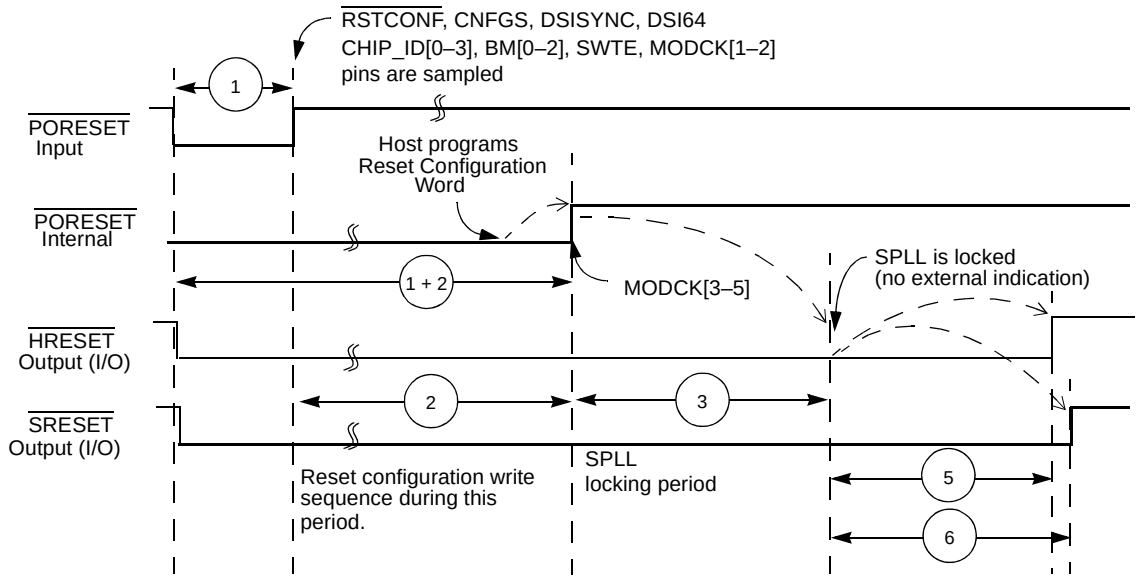


Figure 9. Timing Diagram for a Reset Configuration Write

2.5.5 System Bus Access Timing

2.5.5.1 Core Data Transfers

Generally, all MSC8113 bus and system output signals are driven from the rising edge of the reference clock (REFCLK). The REFCLK is the CLKIN signal. Memory controller signals, however, trigger on four points within a REFCLK cycle. Each cycle is divided by four internal ticks: T1, T2, T3, and T4. T1 always occurs at the rising edge of REFCLK (and T3 at the falling edge), but the spacing of T2 and T4 depends on the PLL clock ratio selected, as **Table 13** shows.

Table 13. Tick Spacing for Memory Controller Signals

BCLK/SC140 clock	Tick Spacing (T1 Occurs at the Rising Edge of REFCLK)		
	T2	T3	T4
1:4, 1:6, 1:8, 1:10	1/4 REFCLK	1/2 REFCLK	3/4 REFCLK
1:3	1/6 REFCLK	1/2 REFCLK	4/6 REFCLK
1:5	2/10 REFCLK	1/2 REFCLK	7/10 REFCLK

Figure 10 is a graphical representation of **Table 13**.

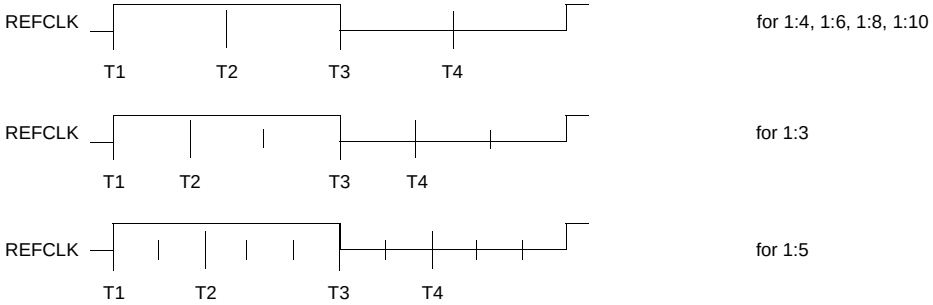


Figure 10. Internal Tick Spacing for Memory Controller Signals

2.5.5.2 CLKIN to CLKOUT Skew

Table 17 describes the CLKOUT-to-CLKIN skew timing.

Table 16. CLKOUT Skew

No.	Characteristic	Min[1]	Max[1]	Units
20	Rise-to-rise skew	0.0	0.95	ns
21	Fall-to-fall skew	–1.5	1.0	ns
23	CLKOUT phase (1.1 V, 133 MHz) • Phase high • Phase low	 2.2 2.2	 — —	 ns ns
24	CLKOUT phase (1.1 V, 100 MHz) • Phase high • Phase low	 3.3 3.3	 — —	 ns ns

Notes:
1. A positive number indicates that CLKOUT precedes CLKIN, A negative number indicates that CLKOUT follows CLKIN.
2. Skews are measured in clock mode 29, with a CLKIN:CLKOUT ratio of 1:1. The same skew is valid for all clock modes.
3. CLKOUT skews are measured using a load of 10 pF.
4. CLKOUT skews and phase are not measured for 500/166 Mhz parts because these parts only use CLKIN mode.

For designs that use the CLKOUT synchronization mode, use the skew values listed in **Table 16** to adjust the rise-to-fall timing values specified for CLKIN synchronization. **Figure 12** shows the relationship between the CLKOUT and CLKIN timings.

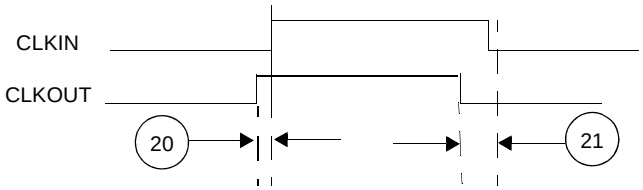


Figure 12. CLKOUT and CLKIN Signals.

2.5.5.3 DMA Data Transfers

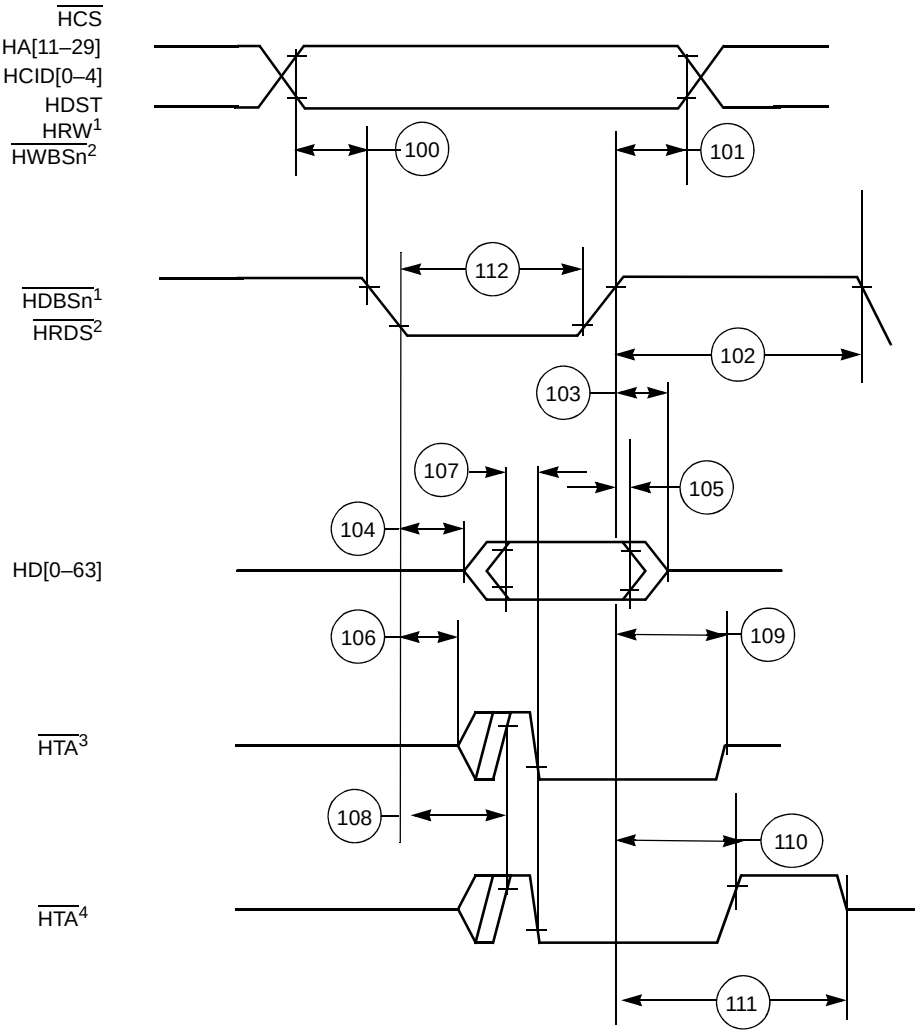
Table 17 describes the DMA signal timing.

Table 17. DMA Signals

No.	Characteristic	Ref = CLKIN		Units
		Min	Max	
37	DREQ set-up time before the 50% level of the falling edge of REFCLK	5.0	—	ns
38	DREQ hold time after the 50% level of the falling edge of REFCLK	0.5	—	ns
39	$\overline{\text{DONE}}$ set-up time before the 50% level of the rising edge of REFCLK	5.0	—	ns
40	$\overline{\text{DONE}}$ hold time after the 50% level of the rising edge of REFCLK	0.5	—	ns
41	$\overline{\text{DACK}}$/$\overline{\text{DRACK}}$/$\overline{\text{DONE}}$ delay after the 50% level of the REFCLK rising edge	0.5	7.5	ns

The DREQ signal is synchronized with REFCLK. To achieve fast response, a synchronized peripheral should assert DREQ according to the timings in **Table 17**. **Figure 13** shows synchronous peripheral interaction.

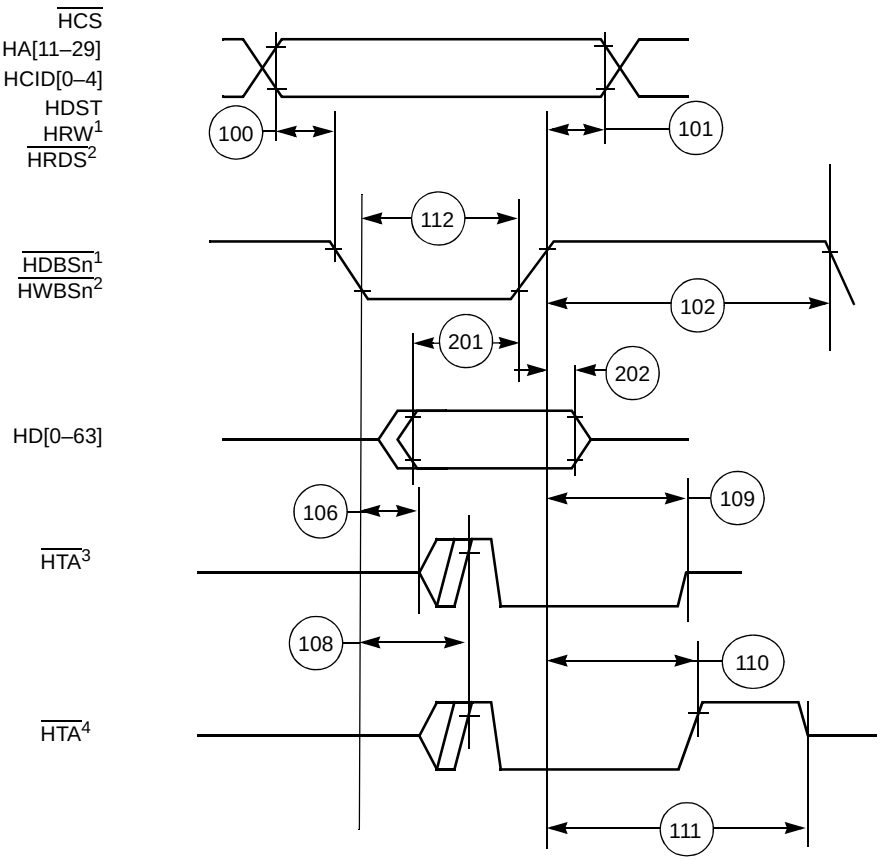
Figure 14 shows DSI asynchronous read signals timing.



Notes:
1. Used for single-strobe mode access.
2. Used for dual-strobe mode access.
3. HTA released at logic 0 (DCR[HTAAD] = 0) at end of access; used with pull-down implementation.
4. HTA released at logic 1 (DCR[HTAAD] = 1) at end of access; used with pull-up implementation.

Figure 14. Asynchronous Single- and Dual-Strobe Modes Read Timing Diagram

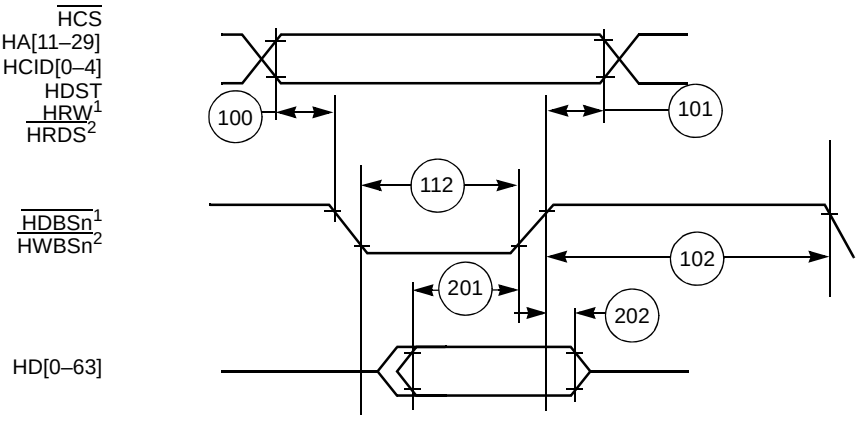
Figure 15 shows DSI asynchronous write signals timing.



Notes:
1. Used for single-strobe mode access.
2. Used for dual-strobe mode access.
3. HTA released at logic 0 (DCR[HTAAD] = 0) at end of access; used with pull-down implementation.
4. HTA released at logic 1 (DCR[HTAAD] = 1) at end of access; used with pull-up implementation.

Figure 15. Asynchronous Single- and Dual-Strobe Modes Write Timing Diagram

Figure 16 shows DSI asynchronous broadcast write signals timing.



Notes:
1. Used for single-strobe mode access.
2. Used for dual-strobe mode access.

Figure 16. Asynchronous Broadcast Write Timing Diagram

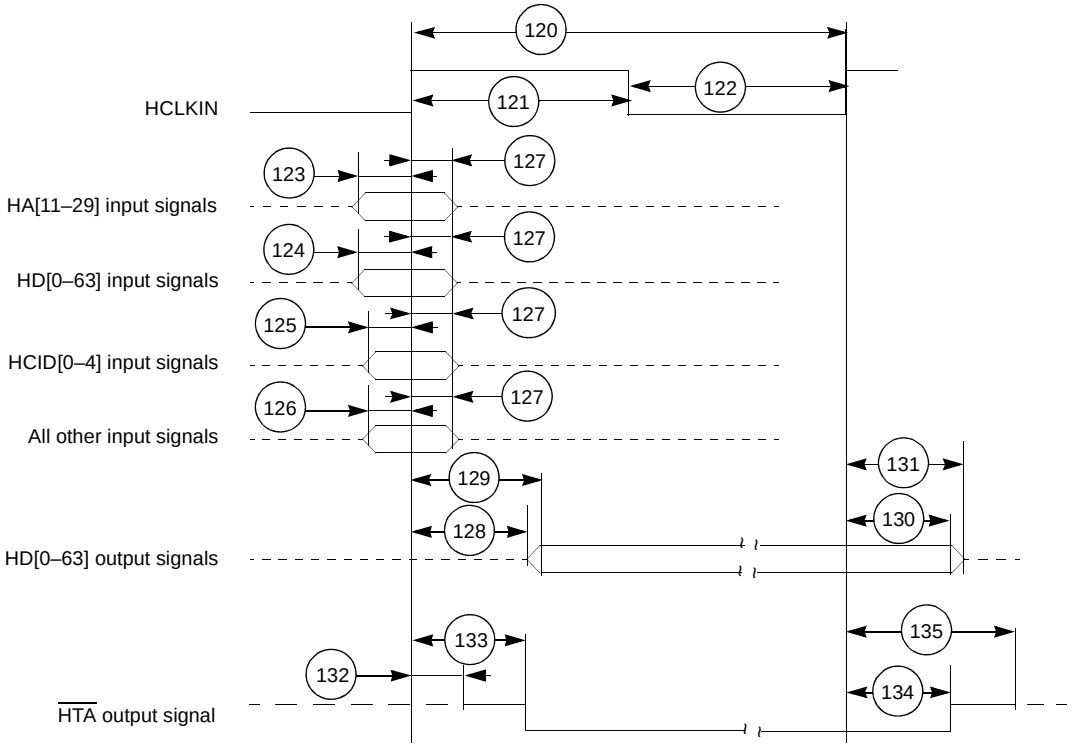


Figure 17. DSI Synchronous Mode Signals Timing Diagram

2.5.7 TDM Timing

Table 21. TDM Timing

No.	Characteristic	Expression	1.1 V Core		Units
			Min	Max	
300	TDMxRCLK/TDMxTCLK	TC[1]	16	—	ns
301	TDMxRCLK/TDMxTCLK high pulse width	(0.5 ± 0.1) × TC	7	—	ns
302	TDMxRCLK/TDMxTCLK low pulse width	(0.5 ± 0.1) × TC	7	—	ns
303	TDM receive all input set-up time		1.3	—	ns
304	TDM receive all input hold time		1.0	—	ns
305	TDMxTCLK high to TDMxTDAT/TDMxRCLK output active[2,3]		2.8	—	ns
306	TDMxTCLK high to TDMxTDAT/TDMxRCLK output		—	10.0	ns
307	All output hold time[4]		2.5	—	ns
308	TDMxTCLK high to TDmXTDAT/TDMxRCLK output high impedance[2,3]		—	10.7	ns
309	TDMxTCLK high to TDMXTSYN output valid[2]		—	9.7	ns
310	TDMxTSYN output hold time[4]		2.5	—	ns

Notes:
1. Values are based on a a maximum frequency of 62.5 MHz. The TDM interface supports any frequency below 62.5 MHz. Devices operating at 300 MHz are limited to a maximum TDMxRCLK/TDMxTCLK frequency of 50 MHz.
2. Values are based on 20 pF capacitive load.
3. When configured as an output, TDMxRCLK acts as a second data link. See the *MSC8113 Reference Manual* for details.
4. Values are based on 10 pF capacitive load.

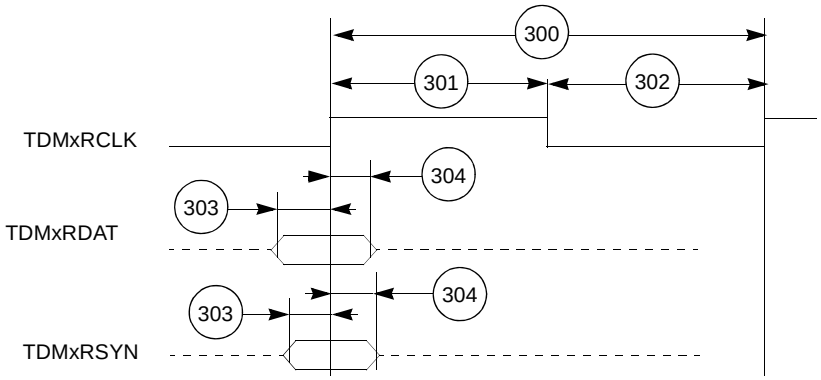


Figure 18. TDM Inputs Signals

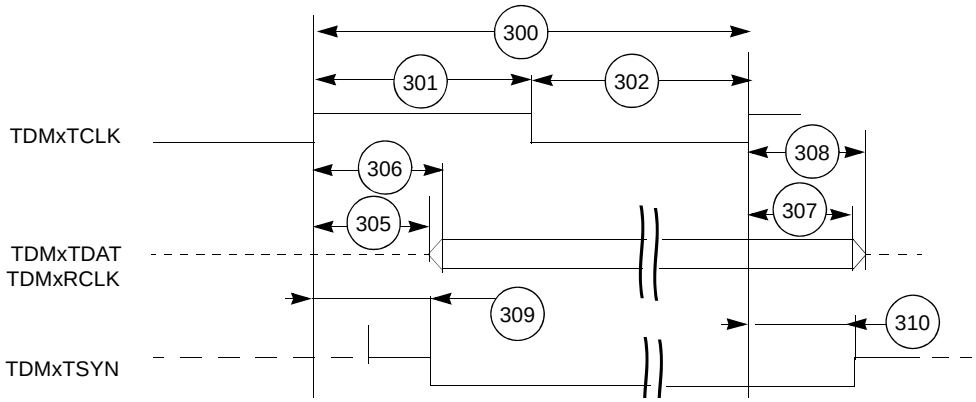


Figure 19. TDM Output Signals

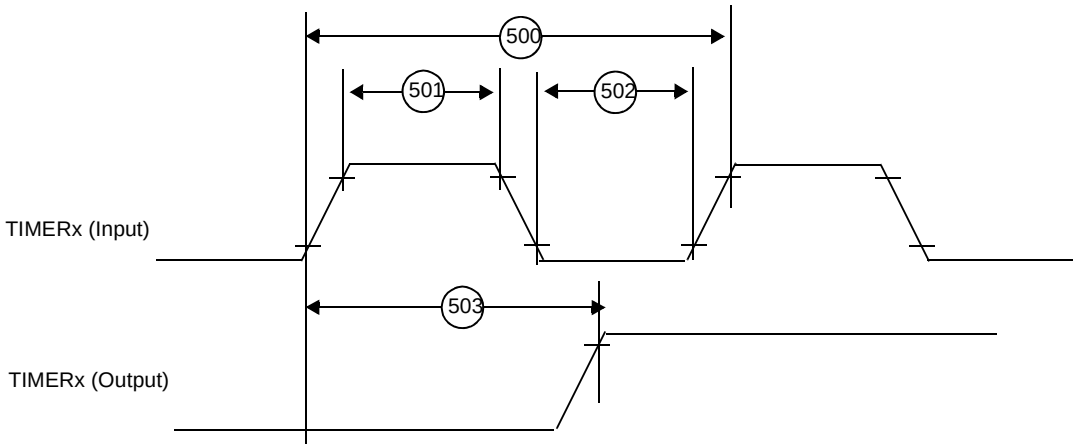


Figure 22. Timer Timing

2.5.10 Ethernet Timing

2.5.10.1 Management Interface Timing

Table 24. Ethernet Controller Management Interface Timing

No.	Characteristics	Min	Max	Unit
801	ETHMDIO to ETHMDC rising edge set-up time	10	—	ns
802	ETHMDC rising edge to ETHMDIO hold time	10	—	ns

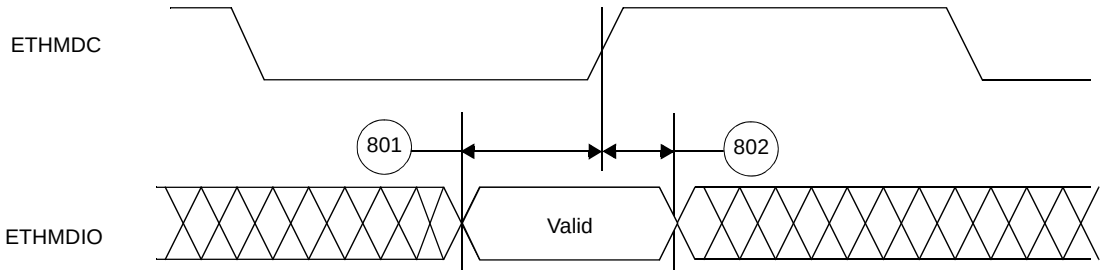


Figure 23. MDIO Timing Relationship to MDC

2.5.10.4 SMII Mode

Table 27. SMII Mode Signal Timing

No.	Characteristics	Min	Max	Unit
808	ETHSYNC_IN, ETHRXD to ETHCLOCK rising edge set-up time	1.0	—	ns
809	ETHCLOCK rising edge to ETHSYNC_IN, ETHRXD hold time	1.0	—	ns
810	ETHCLOCK rising edge to ETHSYNC, ETHTXD output delay	1.5[1]	6.0[2]	ns

Notes: **1.** Measured using a 5 pF load.
2. Measured using a 15 pF load.

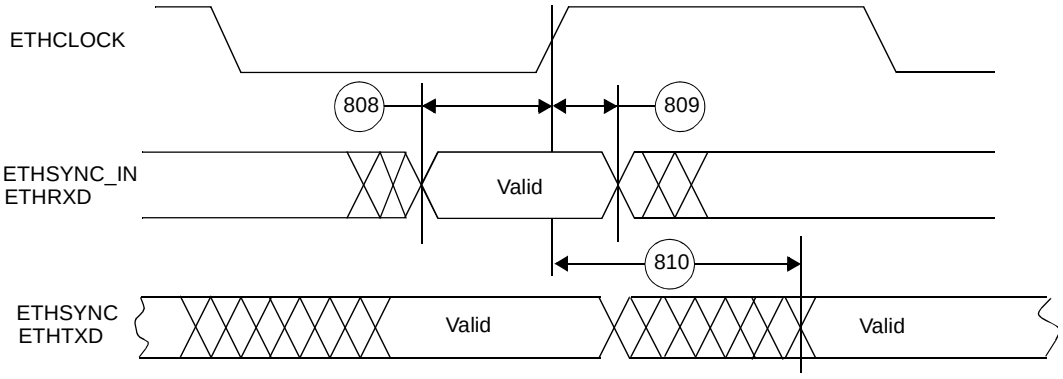


Figure 26. SMII Mode Signal Timing

2.5.11 GPIO Timing

Table 28. GPIO Timing

No.	Characteristics	Ref = CLKIN		Unit
		Min	Max	
601	REFCLK edge to GPIO out valid (GPIO out delay time)	—	6.1	ns
602	REFCLK edge to GPIO out not valid (GPIO out hold time)	1.1	—	ns
603	REFCLK edge to high impedance on GPIO out	—	5.4	ns
604	GPIO in valid to REFCLK edge (GPIO in set-up time)	3.5	—	ns
605	REFCLK edge to GPIO in not valid (GPIO in hold time)	0.5	—	ns

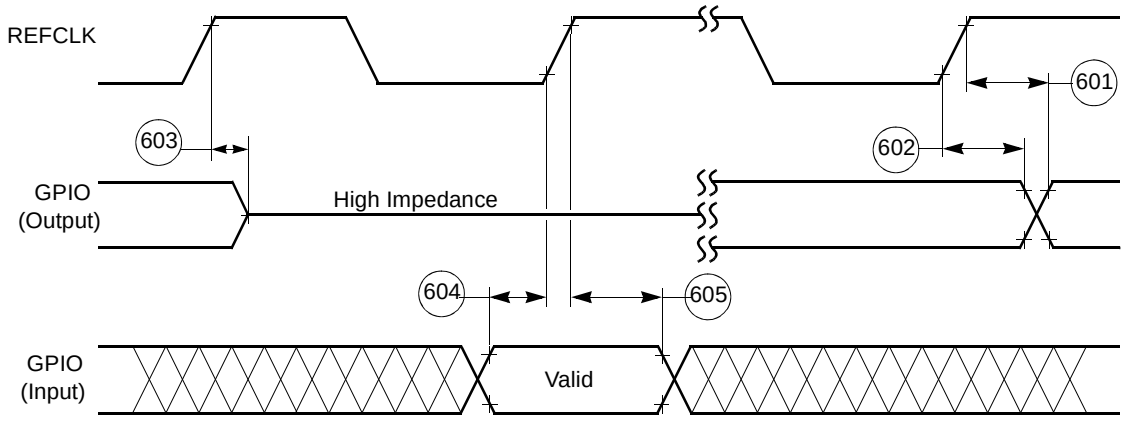


Figure 27. GPIO Timing

2.5.12 EE Signals

Table 29. EE Pin Timing

Number	Characteristics	Type	Min
65	EE0 (input)	Asynchronous	4 core clock periods
66	EE1 (output)	Synchronous to Core clock	1 core clock period

Notes: 1. The core clock is the SC140 core clock. The ratio between the core clock and CLKOUT is configured during power-on-reset.
2. Refer to **Table 1-4** on page 1-6 for details on EE pin functionality.

Figure 28 shows the signal behavior of the EE pins.

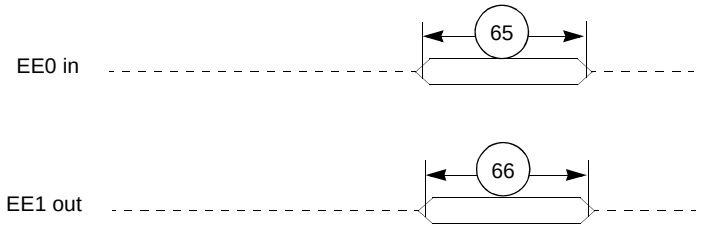


Figure 28. EE Pin Timing

2.5.13 JTAG Signals

Table 30. JTAG Timing

No.	Characteristics	All frequencies		Unit
		Min	Max	
700	TCK frequency of operation (1/(T_C × 4); maximum 25 MHz)	0.0	25	MHz
701	TCK cycle time	40.0	—	ns
702	TCK clock pulse width measured at V_M = 1.6 V • High • Low	 20.0 16.0	 — —	 ns ns
703	TCK rise and fall times	0.0	3.0	ns
704	Boundary scan input data set-up time	5.0	—	ns
705	Boundary scan input data hold time	20.0	—	ns
706	TCK low to output data valid	0.0	30.0	ns
707	TCK low to output high impedance	0.0	30.0	ns
708	TMS, TDI data set-up time	5.0	—	ns
709	TMS, TDI data hold time	20.0	—	ns
710	TCK low to TDO data valid	0.0	20.0	ns
711	TCK low to TDO high impedance	0.0	20.0	ns
712	$\overline{\text{TRST}}$ assert time	100.0	—	ns
713	$\overline{\text{TRST}}$ set-up time to TCK low	30.0	—	ns
Note:	All timings apply to OnCE module data transfers as well as any other transfers via the JTAG port.			

Each V_{CC} and V_{DD} pin on the MSC8113 device should have a low-impedance path to the board power supply. Similarly, each GND pin should have a low-impedance path to the ground plane. The power supply pins drive distinct groups of logic on the chip. The V_{CC} power supply should have at least four 0.1 µF by-pass capacitors to ground located as closely as possible to the four sides of the package. The capacitor leads and associated printed circuit traces connecting to chip V_{CC}, V_{DD}, and GND should be kept to less than half an inch per capacitor lead. A four-layer board is recommended, employing two inner layers as V_{CC} and GND planes.

All output pins on the MSC8113 have fast rise and fall times. PCB trace interconnection length should be minimized to minimize undershoot and reflections caused by these fast output switching times. This recommendation particularly applies to the address and data buses. Maximum PCB trace lengths of six inches are recommended. For the DSI control signals in synchronous mode, ensure that the layout supports the DSI AC timing requirements and minimizes any signal crosstalk. Capacitance calculations should consider all device loads as well as parasitic capacitances due to the PCB traces. Attention to proper PCB layout and bypassing becomes especially critical in systems with higher capacitive loads because these loads create higher transient currents in the V_{CC}, V_{DD}, and GND circuits. Pull up all unused inputs or signals that will be inputs during reset.

Special care should be taken to minimize the noise levels on the PLL supply pins. There is one pair of PLL supply pins: V_{CCSYN}-GND_{SYN}. To ensure internal clock stability, filter the power to the V_{CCSYN} input with a circuit similar to the one in **Figure 34**. For optimal noise filtering, place the circuit as close as possible to V_{CCSYN}. The 0.01-µF capacitor should be closest to V_{CCSYN}, followed by the 10-µF capacitor, the 10-nH inductor, and finally the 10-Ω resistor to V_{DD}. These traces should be kept short and direct. Provide an extremely low impedance path to the ground plane for GND_{SYN}. Bypass GND_{SYN} to V_{CCSYN} by a 0.01-µF capacitor located as close as possible to the chip package. For best results, place this capacitor on the backside of the PCB aligned with the depopulated void on the MSC8113 located in the square defined by positions, L11, L12, L13, M11, M12, M13, N11, N12, and N13.

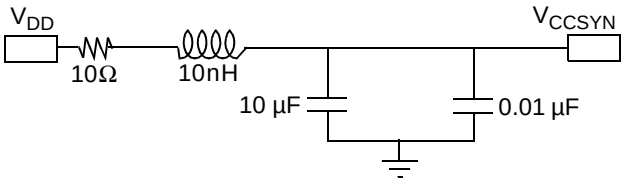


Figure 34. V_{CCSYN} Bypass

3.3 Connectivity Guidelines

Unused output pins can be disconnected, and unused input pins should be connected to the non-active value, via resistors to V_{DDH} or GND, except for the following:

- If the DSI is unused (DDR[DSIDIS] is set), $\overline{\text{HCS}}$ and $\overline{\text{HBCS}}$ must pulled up and all the rest of the DSI signals can be disconnected.
- When the DSI uses synchronous mode, $\overline{\text{HTA}}$ must be pulled up. In asynchronous mode, $\overline{\text{HTA}}$ should be pulled either up or down, depending on design requirements.
- $\overline{\text{HDST}}$ can be disconnected if the DSI is in big-endian mode, or if the DSI is in little-endian mode and the DCR[DSRFA] bit is set.
- When the DSI is in 64-bit data bus mode and DCR[BEM] is cleared, pull up $\overline{\text{HWBS[1–3]}}$/$\overline{\text{HDBS[1–3]}}$/$\overline{\text{HWBE[1–3]}}$/$\overline{\text{HDBE[1–3]}}$ and $\overline{\text{HWBS[4–7]}}$/$\overline{\text{HDBS[4–7]}}$/$\overline{\text{HWBE[4–7]}}$/$\overline{\text{HDBE[4–7]}}$/$\overline{\text{PWE[4–7]}}$/$\overline{\text{PSDDQM[4–7]}}$/$\overline{\text{PBS[4–7]}}$.
- When the DSI is in 32-bit data bus mode and DCR[BEM] is cleared, $\overline{\text{HWBS[1–3]}}$/$\overline{\text{HDBS[1–3]}}$/$\overline{\text{HWBE[1–3]}}$/$\overline{\text{HDBE[1–3]}}$ must be pulled up.
- When the DSI is in asynchronous mode, $\overline{\text{HBRST}}$ and HCLKIN should either be disconnected or pulled up.
- The following signals must be pulled up: $\overline{\text{HRESET}}$, $\overline{\text{SRESET}}$, $\overline{\text{ARTRY}}$, $\overline{\text{TA}}$, $\overline{\text{TEA}}$, $\overline{\text{PSDVAL}}$, and $\overline{\text{AACK}}$.
- In single-master mode (BCR[EBM] = 0) with internal arbitration (PPC_ACR[EARB] = 0):
  - $\overline{\text{BG}}$, $\overline{\text{DBG}}$, and $\overline{\text{TS}}$ can be left unconnected.
  - $\overline{\text{EXT\_BG[2–3]}}$, $\overline{\text{EXT\_DBG[2–3]}}$, and $\overline{\text{GBL}}$ can be left unconnected if they are multiplexed to the system bus functionality. For any other functionality, connect the signal lines based on the multiplexed functionality.
  - $\overline{\text{BR}}$ must be pulled up.
  - $\overline{\text{EXT\_BR[2–3]}}$ must be pulled up if multiplexed to the system bus functionality.

- If there is an external bus master (BCR[EBM] = 1):
  - $\overline{\text{BR}}$, $\overline{\text{BG}}$, $\overline{\text{DBG}}$, and $\overline{\text{TS}}$ must be pulled up.
  - $\overline{\text{EXT\_BR[2–3]}}$, $\overline{\text{EXT\_BG[2–3]}}$, and $\overline{\text{EXT\_DBG[2–3]}}$ must be pulled up if multiplexed to the system bus functionality.
- In single-master mode, $\overline{\text{ABB}}$ and $\overline{\text{DBB}}$ can be selected as $\overline{\text{IRQ}}$ inputs and be connected to the non-active value. In other modes, they must be pulled up.

Note: The MSC8113 does not support DLL-enabled mode. For the following two clock schemes, ensure that the DLL is disabled (that is, the DLLDIS bit in the Hard Reset Configuration Word is set).

- If no system synchronization is required (for example, the design does not use SDRAM), you can use any of the available clock modes.
- In the CLKIN synchronization mode, use the following connections:
  - Connect the oscillator output through a buffer to CLKIN.
  - Connect the CLKIN buffer output to the slave device (for example, SDRAM) making sure that the delay path between the clock buffer to the MSC8113 and the SDRAM is equal (that is, has a skew less than 100 ps).
  - Valid clock modes in this scheme are: 0, 7, 15, 19, 21, 23, 28, 29, 30, and 31.

Note: See the Clock chapter in the *MSC8113 Reference Manual* for details.

- If the 60x-compatible system bus is not used and SIUMCR[PBSE] is set, $\overline{\text{PPBS}}$ can be disconnected. Otherwise, it should be pulled up.
- The following signals: SWTE, DSISYNC, DSI64, MODCK[1–2], CNFGS, CHIPID[0–3], RSTCONF and BM[0–2] are used to configure the MSC8113 and are sampled on the deassertion of the $\overline{\text{PORESET}}$ signal. Therefore, they should be tied to GND or V_{DDH} or through a pull-down or a pull-up resistor until the deassertion of the $\overline{\text{PORESET}}$ signal.
- When they are used, $\overline{\text{INT\_OUT}}$ (if SIUMCR[INTODC] is cleared), $\overline{\text{NMI\_OUT}}$, and $\overline{\text{IRQxx}}$ (if not full drive) signals must be pulled up.
- When the Ethernet controller is enabled and the SMII mode is selected, GPIO10 and GPIO14 must not be connected externally to any signal line.

Note: For details on configuration, see the *MSC8113 User's Guide* and *MSC8113 Reference Manual*. For additional information, refer to the *MSC8113 Design Checklist* (ANxxxx).

3.4 External SDRAM Selection

The external bus speed implemented in a system determines the speed of the SDRAM used on that bus. However, because of differences in timing characteristics among various SDRAM manufacturers, you may have use a faster speed rated SDRAM to assure efficient data transfer across the bus. For example, for 133 MHz operation, you may have to use 133 or 166 MHz SDRAM. Always perform a detailed timing analysis using the MSC8113 bus timing values and the manufacturer specifications for the SDRAM to ensure correct operation within your system design. The output delay listed in SDRAM specifications is usually given for a load of 30 pF. Scale the number to your specific board load using the typical scaling number provided by the SDRAM manufacturer.