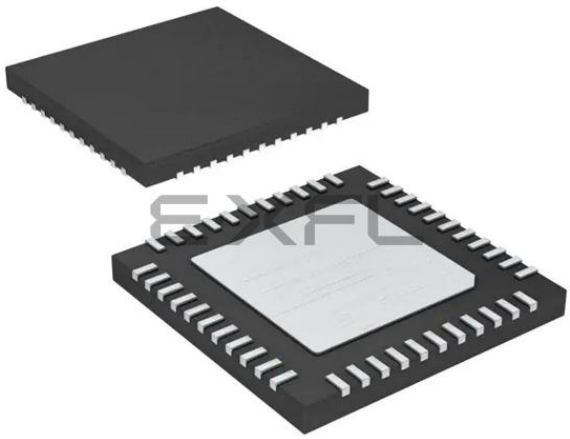

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Applications of "Embedded - Microcontrollers"

Details	
Product Status	Active
Core Processor	PIC
Core Size	8-Bit
Speed	64MHz
Connectivity	I²C, LINbus, SPI, UART/USART
Peripherals	Brown-out Detect/Reset, LVD, POR, PWM, WDT
Number of I/O	36
Program Memory Size	128KB (64K x 16)
Program Memory Type	FLASH
EEPROM Size	1K x 8
RAM Size	3.6K x 8
Voltage - Supply (Vcc/Vdd)	1.8V ~ 3.6V
Data Converters	A/D 35x10b; D/A 1x5b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 125°C (TA)
Mounting Type	Surface Mount
Package / Case	44-VQFN Exposed Pad
Supplier Device Package	44-QFN (8x8)
Purchase URL	https://www.e-xfl.com/product-detail/microchip-technology/pic18lf47k40-e-ml

REGISTER 3-3: Configuration Word 2L (30 0002h): Supervisor

R/W-1	R/W-1	R/W-1	U-1	U-1	U-1	R/W-1	R/W-1
BOREN<1:0>		$\overline{\text{LPBOREN}}$	—	—	—	$\overline{\text{PWRTE}}$	MCLRE
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '1'	
-n = Value for blank device	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-6 **BOREN<1:0>:** Brown-out Reset Enable bits
When enabled, Brown-out Reset Voltage (V_{BOR}) is set by BORV bit
11 = Brown-out Reset enabled, SBOREN bit is ignored
10 = Brown-out Reset enabled while running, disabled in Sleep; SBOREN is ignored
01 = Brown-out Reset enabled according to SBOREN
00 = Brown-out Reset disabled

bit 5 **$\overline{\text{LPBOREN}}$**: Low-Power BOR Enable bit
1 = Low-Power Brown-out Reset is disabled
0 = Low-Power Brown-out Reset is enabled

bit 4-2 **Unimplemented:** Read as '1'

bit 1 **$\overline{\text{PWRTE}}$:** Power-up Timer Enable bit
1 = PWRT disabled
0 = PWRT enabled

bit 0 **MCLRE:** Master Clear ($\overline{\text{MCLR}}$) Enable bit
If LVP = 1
RE3 pin function is $\overline{\text{MCLR}}$
If LVP = 0
1 = $\overline{\text{MCLR}}$ pin is $\overline{\text{MCLR}}$
0 = $\overline{\text{MCLR}}$ pin function is port defined function

10.2.3 LOOK-UP TABLES IN PROGRAM MEMORY

There may be programming situations that require the creation of data structures, or look-up tables, in program memory. For PIC18 devices, look-up tables can be implemented in two ways:

- Computed GOTO
- Table Reads

10.2.3.1 Computed GOTO

A computed GOTO is accomplished by adding an offset to the program counter. An example is shown in Example 10-2.

A look-up table can be formed with an ADDWF PCL instruction and a group of RETLW nn instructions. The W register is loaded with an offset into the table before executing a call to that table. The first instruction of the called routine is the ADDWF PCL instruction. The next instruction executed will be one of the RETLW nn instructions that returns the value 'nn' to the calling function.

The offset value (in WREG) specifies the number of bytes that the program counter should advance and should be multiples of two (LSb = 0).

In this method, only one data byte may be stored in each instruction location and room on the return address stack is required.

EXAMPLE 10-2: COMPUTED GOTO USING AN OFFSET VALUE

```
        MOVF    OFFSET, W
        CALL    TABLE
ORG     nn00h
TABLE   ADDWF   PCL
        RETLW   nnh
        RETLW   nnh
        RETLW   nnh
        .
        .
        .
```

10.2.3.2 Table Reads and Table Writes

A better method of storing data in program memory allows two bytes of data to be stored in each instruction location.

Look-up table data may be stored two bytes per program word by using table reads and writes. The Table Pointer (TBLPTR) register specifies the byte address and the Table Latch (TABLAT) register contains the data that is read from or written to program memory. Data is transferred to or from program memory one byte at a time.

Table read and table write operations are discussed further in **Section 11.1.1 "Table Reads and Table Writes"**.

11.3.6 OPERATION DURING CODE-PROTECT

Data EEPROM Memory has its own code-protect bits in Configuration Words. External read and write operations are disabled if code protection is enabled.

If the Data EEPROM is write-protected or if NVMADR points an invalid address location, the WR bit is cleared without any effect. WRERR is signaled in this scenario.

11.3.7 PROTECTION AGAINST SPURIOUS WRITE

There are conditions when the user may not want to write to the Data EEPROM Memory. To protect against spurious EEPROM writes, various mechanisms have been implemented. On power-up, the WREN bit is cleared. In addition, writes to the EEPROM are blocked during the Power-up Timer period (TPWRT).

The unlock sequence and the WREN bit together help prevent an accidental write during brown-out, power glitch or software malfunction.

11.3.8 ERASING THE DATA EEPROM MEMORY

Data EEPROM Memory can be erased by writing 0xFF to all locations in the Data EEPROM Memory that needs to be erased.

EXAMPLE 11-7: DATA EEPROM REFRESH ROUTINE

```
        CLRF    NVMADRL             ; Clear address low byte register
        CLRF    NVMADRH             ; Clear address high byte register (if applicable)
        BCF     NVMCON1, NVMREG0    ; Set access for EEPROM
        BCF     NVMCON1, NVMREG1    ; Set access for EEPROM
        SETF    NVMDAT              ; Load 0xFF to data register
        BCF     INTCON, GIE         ; Disable interrupts
        BSF     NVMCON1, WREN       ; Enable writes
Loop                                ; Loop to refresh array
        MOVLW   0x55                ; Initiate unlock sequence
        MOVWF   NVMCON2             ;
        MOVLW   0xAA                ;
        MOVWF   NVMCON2             ;
        BSF     NVMCON1, WR         ; Set WR bit to begin write
        BTFSC   NVMCON1, WR         ; Wait for write to complete
        BRA     $-2
        INCFSZ  NVMADRL, F          ; Increment address low byte
        BRA     Loop                ; Not zero, do it again

 //The following 4 lines of code are not needed if the part doesn't have NVMADRH register
        INCF    NVMADRH, F          ; Decrement address high byte
        MOVLW   0x03                ; Move 0x03 to working register
        CPFSGT  NVMADRH             ; Compare address high byte with working register
        BRA     Loop                ; Skip if greater than working register
                                    ; Else go back to erase loop

        BCF     NVMCON1, WREN       ; Disable writes
        BSF     INTCON, GIE         ; Enable interrupts
```

REGISTER 11-2: NVMCON2: NONVOLATILE MEMORY CONTROL 2 REGISTER

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
NVMCON2<7:0>							
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
x = Bit is unknown	'0' = Bit is cleared	'1' = Bit is set
-n = Value at POR		

bit 7-0 **NVMCON2<7:0>:**
Refer to **Section 11.1.4 "NVM Unlock Sequence"**.

Note 1: This register always reads zeros, regardless of data written.

Register 11-3: NVMADRL: Data EEPROM Memory Address Low

R/W-x/0	R/W-x/0	R/W-x/0	R/W-x/0	R/W-x/0	R/W-x/0	R/W-x/0	R/W-x/0
NVMADR<7:0>							
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
x = Bit is unknown	'0' = Bit is cleared	'1' = Bit is set
-n = Value at POR		

bit 7-0 **NVMADR<7:0>:** EEPROM Read Address bits

REGISTER 11-4: NVMADRH: DATA EEPROM MEMORY ADDRESS HIGH

U-0	U-0	U-0	U-0	U-0	U-0	R/W-x/u	R/W-x/u
—	—	—	—	—	—	NVMADR<9:8>	
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
x = Bit is unknown	'0' = Bit is cleared	'1' = Bit is set
-n = Value at POR		

bit 7-2 **Unimplemented:** Read as '0'
bit 1-0 **NVMADR<9:8>:** EEPROM Read Address bits

REGISTER 13-16: SCANHADRH: SCAN HIGH ADDRESS HIGH BYTE REGISTER

R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0
HADR<15:8>(1, 2)							
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7-0 **HADR<15:8>**: Scan End Address bits(1, 2)
Most Significant bits of the address at the end of the designated scan

Note 1: Registers SCANHADRU/H/L form a 22-bit value, but are not guarded for atomic or asynchronous access; registers should only be read or written while SCANGO = 0 (SCANCON0 register).
2: While SCANGO = 1 (SCANCON0 register), writing to this register is ignored.

REGISTER 13-17: SCANHADRL: SCAN HIGH ADDRESS LOW BYTE REGISTER

R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0
HADR<7:0>(1, 2)							
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7-0 **HADR<7:0>**: Scan End Address bits(1, 2)
Least Significant bits of the address at the end of the designated scan

Note 1: Registers SCANHADRU/H/L form a 22-bit value, but are not guarded for atomic or asynchronous access; registers should only be read or written while SCANGO = 0 (SCANCON0 register).
2: While SCANGO = 1 (SCANCON0 register), writing to this register is ignored.

13.6 CRC Check Value

The CRC check value will be located in the CRCACC registers after the CRC calculation has finished. The check value will depend on two mode settings of the CRCCON register: ACCM and SHIFTM.

When the ACCM bit is set, the CRC module augments the data with a number of zeros equal to the length of the polynomial to align the final check value. When the ACCM bit is not set, the CRC will stop at the end of the data. A number of zeros equal to the length of the polynomial can then be entered into CRCDAT to find the same check value as augmented mode. Alternatively, the expected check value can be entered at this point to make the final result equal 0.

When the CRC check value is computed with the SHIFTM bit set, selecting LSb first, and the ACCM bit is set then the final value in the CRCACC registers will be reversed such that the LSb will be in the MSb position and vice versa. This is the expected check value in bit reversed form. If you are creating a check value to be appended to a data stream then a bit reversal must be performed on the final value to achieve the correct checksum. You can use the CRC to do this reversal by the following method:

- Save CRCACC value in user RAM space
- Clear the CRCACC registers
- Clear the CRCXOR registers
- Write the saved CRCACC value to the CRCDAT input

The properly oriented check value will be in the CRCACC registers as the result.

13.7 CRC Interrupt

The CRC will generate an interrupt when the BUSY bit transitions from 1 to 0. The CRCIF Interrupt Flag bit of the PIR7 register is set every time the BUSY bit transitions, regardless of whether or not the CRC interrupt is enabled. The CRCIF bit can only be cleared in software. The CRC interrupt enable is the CRCIE bit of the PIE7 register.

13.8 Configuring the CRC

The following steps illustrate how to properly configure the CRC.

1. Determine if the automatic program memory scan will be used with the scanner or manual calculation through the SFR interface and perform the actions specified in **Section 13.5 "CRC Data Sources"**, depending on which decision was made.
2. If desired, seed a starting CRC value into the CRCACCH/L registers.
3. Program the CRCXORH/L registers with the desired generator polynomial.
4. Program the DLEN<3:0> bits of the CRCCON1 register with the length of the data word - 1 (refer to Example 13-1). This determines how many times the shifter will shift into the accumulator for each data word.
5. Program the PLEN<3:0> bits of the CRCCON1 register with the length of the polynomial -2 (refer to Example 13-1).
6. Determine whether shifting in trailing zeros is desired and set the ACCM bit of the CRCCON0 register appropriately.
7. Likewise, determine whether the MSb or LSb should be shifted first and write the SHIFTM bit of the CRCCON0 register appropriately.
8. Write the CRCGO bit of the CRCCON0 register to begin the shifting process.

9a. If manual SFR entry is used, monitor the FULL bit of the CRCCON0 register. When FULL = 0, another word of data can be written to the CRCDATH/L registers, keeping in mind that CRCDATH should be written first if the data has more than eight bits, as the shifter will begin upon the CRCDATL register being written.

9b. If the scanner is used, the scanner will automatically stuff words into the CRCDATH/L registers as needed, as long as the SCANGO bit is set.

10a. If using the Flash memory scanner, monitor the SCANIF (or the SCANGO bit) for the scanner to finish pushing information into the CRCDATA registers. After the scanner is completed, monitor the BUSY bit to determine that the CRC has been completed and the check value can be read from the CRCACC registers. If both the interrupt flags are set (or both BUSY and SCANGO bits are cleared), the completed CRC calculation can be read from the CRCACCH/L registers.

10b. If manual entry is used, monitor the BUSY bit to determine when the CRCACC registers will hold the check value.

16.0 INTERRUPT-ON-CHANGE

PORTA, PORTB, PORTC and pin RE3 of PORTE can be configured to operate as Interrupt-on-Change (IOC) pins on PIC18(L)F2x/4xK40 family devices. An interrupt can be generated by detecting a signal that has either a rising edge or a falling edge. Any individual port pin, or combination of port pins, can be configured to generate an interrupt. The interrupt-on-change module has the following features:

- Interrupt-on-Change enable (Master Switch)
- Individual pin configuration
- Rising and falling edge detection
- Individual pin interrupt flags

Figure 16-1 is a block diagram of the IOC module.

16.1 Enabling the Module

To allow individual port pins to generate an interrupt, the IOCIE bit of the PIE0 register must be set. If the IOCIE bit is disabled, the edge detection on the pin will still occur, but an interrupt will not be generated.

16.2 Individual Pin Configuration

For each port pin, a rising edge detector and a falling edge detector are present. To enable a pin to detect a rising edge, the associated bit of the IOCxP register is set. To enable a pin to detect a falling edge, the associated bit of the IOCxN register is set.

A pin can be configured to detect rising and falling edges simultaneously by setting both associated bits of the IOCxP and IOCxN registers, respectively.

16.3 Interrupt Flags

The IOCAFx, IOCBFx, IOCCFx and IOCEF3 bits located in the IOCAF, IOCBF, IOCCF and IOCEF registers respectively, are status flags that correspond to the interrupt-on-change pins of the associated port. If an expected edge is detected on an appropriately enabled pin, then the status flag for that pin will be set, and an interrupt will be generated if the IOCIE bit is set. The IOCIF bit of the PIR0 register reflects the status of all IOCAFx, IOCBFx, IOCCFx and IOCEF3 bits.

16.4 Clearing Interrupt Flags

The individual status flags, (IOCAFx, IOCBFx, IOCCFx and IOCEF3 bits), can be cleared by resetting them to zero. If another edge is detected during this clearing operation, the associated status flag will be set at the end of the sequence, regardless of the value actually being written.

In order to ensure that no detected edge is lost while clearing flags, only AND operations masking out known changed bits should be performed. The following sequence is an example of what should be performed.

EXAMPLE 16-1: CLEARING INTERRUPT FLAGS (PORTA EXAMPLE)

```
MOVLW   0xff
XORWF   IOCAF, W
ANDWF   IOCAF, F
```

16.5 Operation in Sleep

The interrupt-on-change interrupt sequence will wake the device from Sleep mode, if the IOCIE bit is set.

If an edge is detected while in Sleep mode, the IOCxF register will be updated prior to the first instruction executed out of Sleep.

REGISTER 17-3: PPSLOCK: PPS LOCK REGISTER

U-0	U-0	U-0	U-0	U-0	U-0	U-0	R/W-0/0
—	—	—	—	—	—	—	PPSLOCKED
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7-1 **Unimplemented:** Read as '0'

bit 0 **PPSLOCKED:** PPS Locked bit

1 = PPS is locked. PPS selections can not be changed.

0 = PPS is not locked. PPS selections can be changed.

TABLE 17-2: SUMMARY OF REGISTERS ASSOCIATED WITH THE PPS MODULE

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Register on page
PPSLOCK	—	—	—	—	—	—	—	PPSLOCKED	219
INT0PPS	—	—	—	INT0PPS<4:0>					216
INT1PPS	—	—	—	INT1PPS<4:0>					216
INT2PPS	—	—	—	INT2PPS<4:0>					216
T0CKIPPS	—	—	—	T0CKIPPS<4:0>					216
T1CKIPPS	—	—	—	T1CKIPPS<4:0>					216
T1GPPS	—	—	—	T1GPPS<4:0>					216
T3CKIPPS	—	—	—	T3CKIPPS<4:0>					216
T3GPPS	—	—	—	T3GPPS<4:0>					216
T5CKIPPS	—	—	—	T5CKIPPS<4:0>					216
T5GPPS	—	—	—	T5GPPS<4:0>					216
T2INPPS	—	—	—	T2INPPS<4:0>					216
T4INPPS	—	—	—	T4INPPS<4:0>					216
T6INPPS	—	—	—	T6INPPS<4:0>					216
CCP1PPS	—	—	—	CCP1PPS<4:0>					216
CCP2PPS	—	—	—	CCP2PPS<4:0>					216
CWG1PPS	—	—	—	CWG1PPS<4:0>					216
MDCARLPPS	—	—	—	MDCARLPPS<4:0>					216
MDCARHPPS	—	—	—	MDCARHPPS<4:0>					216
MDSRCPPS	—	—	—	MDSRCPPS<4:0>					216
ADACTPPS	—	—	—	ADACTPPS<4:0>					216
SSP1CLKPPS	—	—	—	SSP1CLKPPS<4:0>					216
SSP1DATPPS	—	—	—	SSP1DATPPS<4:0>					216
SSP1SSPPS	—	—	—	SSP1SSPPS<4:0>					216
RX1PPS	—	—	—	RX1PPS<4:0>					218
TX1PPS	—	—	—	TX1PPS<4:0>					216
SSP2CLKPPS	—	—	—	SSP2CLKPPS<4:0>					216
SSP2DATPPS	—	—	—	SSP2DATPPS<4:0>					216

19.7 Timer1/3/5 16-Bit Read/Write Mode

Timer1/3/5 can be configured to read and write all 16 bits of data, to and from, the 8-bit TMRxL and TMRxH registers, simultaneously. The 16-bit read and write operations are enabled by setting the RD16 bit of the TxCON register.

To accomplish this function, the TMRxH register value is mapped to a buffer register called the TMRxH buffer register. While in 16-Bit mode, the TMRxH register is not directly readable or writable and all read and write operations take place through the use of this TMRxH buffer register.

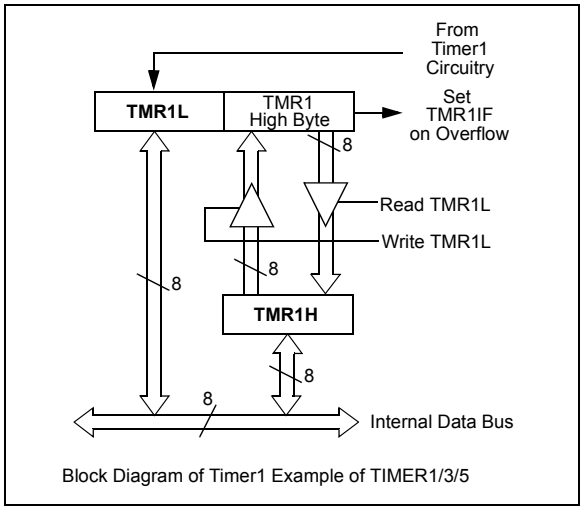
When a read from the TMRxL register is requested, the value of the TMRxH register is simultaneously loaded into the TMRxH buffer register. When a read from the TMRxH register is requested, the value is provided from the TMRxH buffer register instead. This provides the user with the ability to accurately read all 16 bits of the Timer1/3/5 value from a single instance in time. Reference the block diagram in Figure 19-2 for more details.

In contrast, when not in 16-Bit mode, the user must read each register separately and determine if the values have become invalid due to a rollover that may have occurred between the read operations.

When a write request of the TMRxL register is requested, the TMRxH buffer register is simultaneously updated with the contents of the TMRxH register. The value of TMRxH must be preloaded into the TMRxH buffer register prior to the write request for the TMRxL register. This provides the user with the ability to write all 16 bits to the TMRxL:TMRxH register pair at the same time.

Any requests to write to the TMRxH directly does not clear the Timer1/3/5 prescaler value. The prescaler value is only cleared through write requests to the TMRxL register.

FIGURE 19-2: TIMER1/3/5 16-BIT READ/WRITE MODE BLOCK DIAGRAM



Block Diagram of Timer1 Example of TIMER1/3/5

19.8 Timer1/3/5 Gate

Timer1/3/5 can be configured to count freely or the count can be enabled and disabled using Timer1/3/5 gate circuitry. This is also referred to as Timer1/3/5 gate enable.

Timer1/3/5 gate can also be driven by multiple selectable sources.

19.8.1 TIMER1/3/5 GATE ENABLE

The Timer1/3/5 Gate Enable mode is enabled by setting the TMRxGE bit of the TxGCON register. The polarity of the Timer1/3/5 Gate Enable mode is configured using the TxGPOL bit of the TxGCON register.

When Timer1/3/5 Gate Enable mode is enabled, Timer1/3/5 will increment on the rising edge of the Timer1/3/5 clock source. When Timer1/3/5 Gate signal is inactive, the timer will not increment and hold the current count. Enable mode is disabled, no incrementing will occur and Timer1/3/5 will hold the current count. See Figure 19-4 for timing details.

TABLE 19-3: TIMER1/3/5 GATE ENABLE SELECTIONS

TMRxCLK	TxGPOL	TxG	Timer1/3/5 Operation
↑	1	1	Counts
↑	1	0	Holds Count
↑	0	1	Holds Count
↑	0	0	Counts

22.1.5 PWM RESOLUTION

The resolution determines the number of available duty cycles for a given period. For example, a 10-bit resolution will result in 1024 discrete duty cycles, whereas an 8-bit resolution will result in 256 discrete duty cycles.

The maximum PWM resolution is ten bits when PR2 is 255. The resolution is a function of the PR2 register value as shown by Equation 22-4.

EQUATION 22-4: PWM RESOLUTION

$$Resolution = \frac{log[4(PR2 + 1)]}{log(2)} \ bits$$

Note:	If the pulse-width value is greater than the period the assigned PWM pin(s) will remain unchanged.

TABLE 22-1: EXAMPLE PWM FREQUENCIES AND RESOLUTIONS (FOSC = 20 MHz)

PWM Frequency	0.31 kHz	4.88 kHz	19.53 kHz	78.12 kHz	156.3 kHz	208.3 kHz
Timer Prescale	64	4	1	1	1	1
PR2 Value	0xFF	0xFF	0xFF	0x3F	0x1F	0x17
Maximum Resolution (bits)	10	10	10	8	7	6.6

TABLE 22-2: EXAMPLE PWM FREQUENCIES AND RESOLUTIONS (FOSC = 8 MHz)

PWM Frequency	0.31 kHz	4.90 kHz	19.61 kHz	76.92 kHz	153.85 kHz	200.0 kHz
Timer Prescale	64	4	1	1	1	1
PR2 Value	0x65	0x65	0x65	0x19	0x0C	0x09
Maximum Resolution (bits)	8	8	8	6	5	5

22.1.6 OPERATION IN SLEEP MODE

In Sleep mode, the TMR2 register will not increment and the state of the module will not change. If the PWMx pin is driving a value, it will continue to drive that value. When the device wakes up, TMR2 will continue from its previous state.

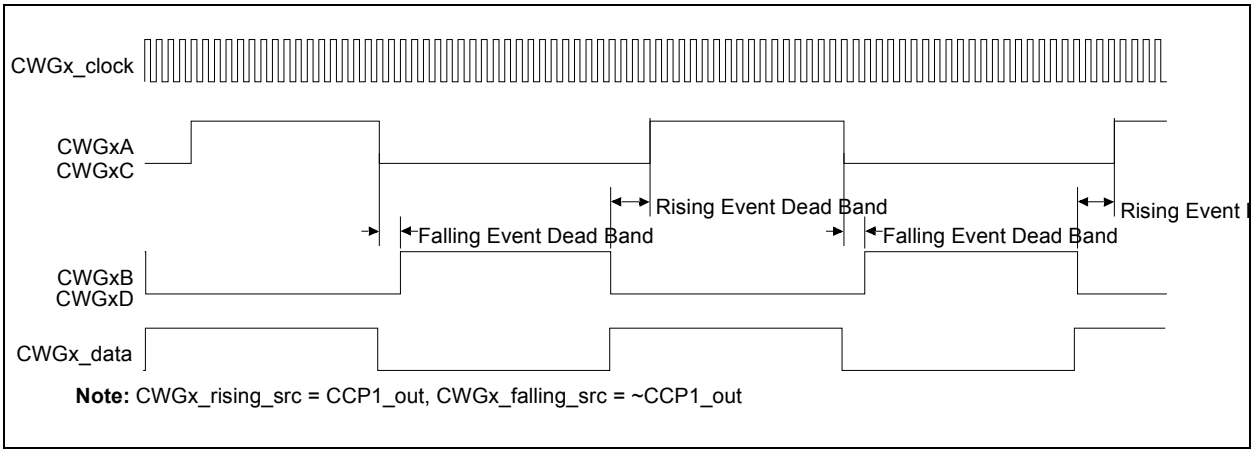
22.1.7 CHANGES IN SYSTEM CLOCK FREQUENCY

The PWM frequency is derived from the system clock frequency (FOSC). Any changes in the system clock frequency will result in changes to the PWM frequency. Refer to **Section 4.0 "Oscillator Module (with Fail-Safe Clock Monitor)"** for additional details.

22.1.8 EFFECTS OF RESET

Any Reset will force all ports to Input mode and the PWM registers to their Reset states.

FIGURE 24-2: CWG1 HALF-BRIDGE MODE OPERATION



Note: CWGx_rising_src = CCP1_out, CWGx_falling_src = ~CCP1_out

24.2.2 PUSH-PULL MODE

In Push-Pull mode, two output signals are generated, alternating copies of the input as illustrated in Figure 24-4. This alternation creates the push-pull effect required for driving some transformer-based power supply designs. Steering modes are not used in Push-Pull mode. A basic block diagram for the Push-Pull mode is shown in Figure 24-3.

The push-pull sequencer is reset whenever EN = 0 or if an auto-shutdown event occurs. The sequencer is clocked by the first input pulse, and the first output appears on CWG1A.

The unused outputs CWG1C and CWG1D drive copies of CWG1A and CWG1B, respectively, but with polarity controlled by the POLC and POLD bits of the CWG1CON1 register, respectively.

FIGURE 26-7: SPI MODE WAVEFORM (SLAVE MODE WITH CKE = 0)

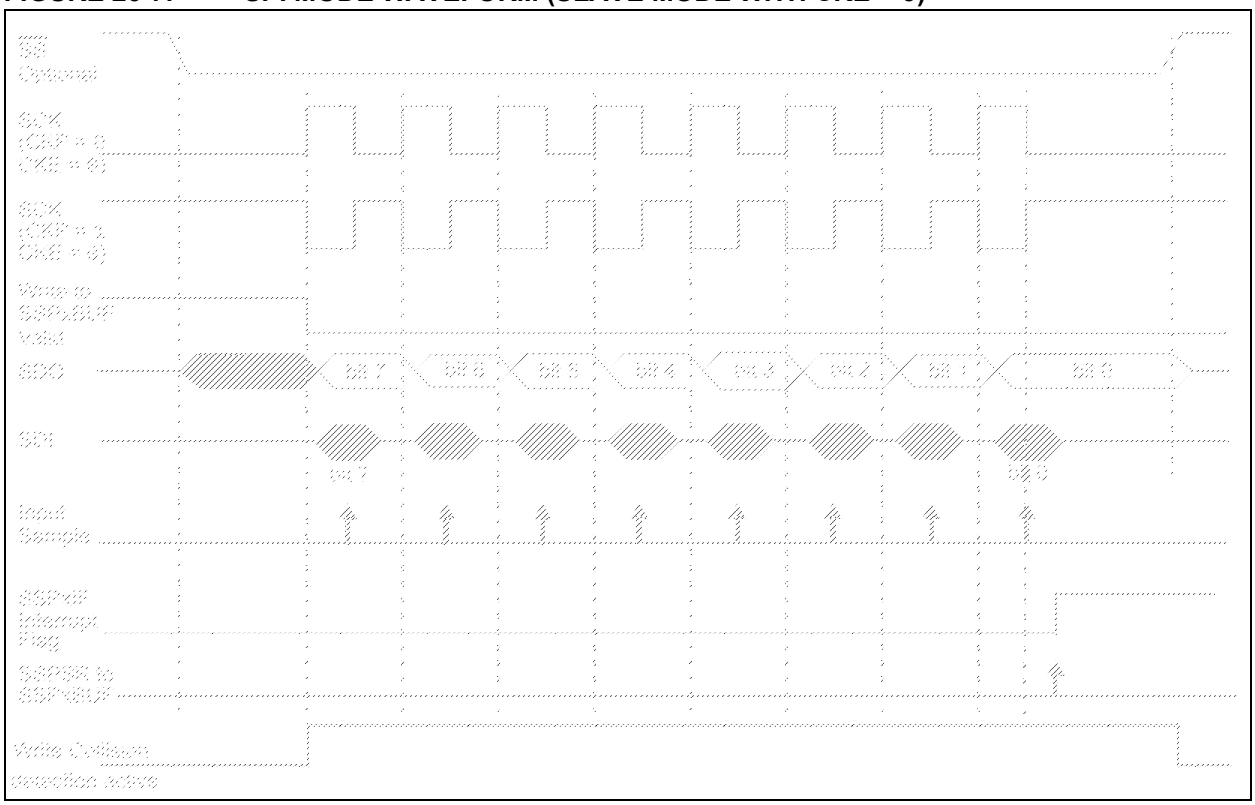


FIGURE 26-8: SPI MODE WAVEFORM (SLAVE MODE WITH CKE = 1)

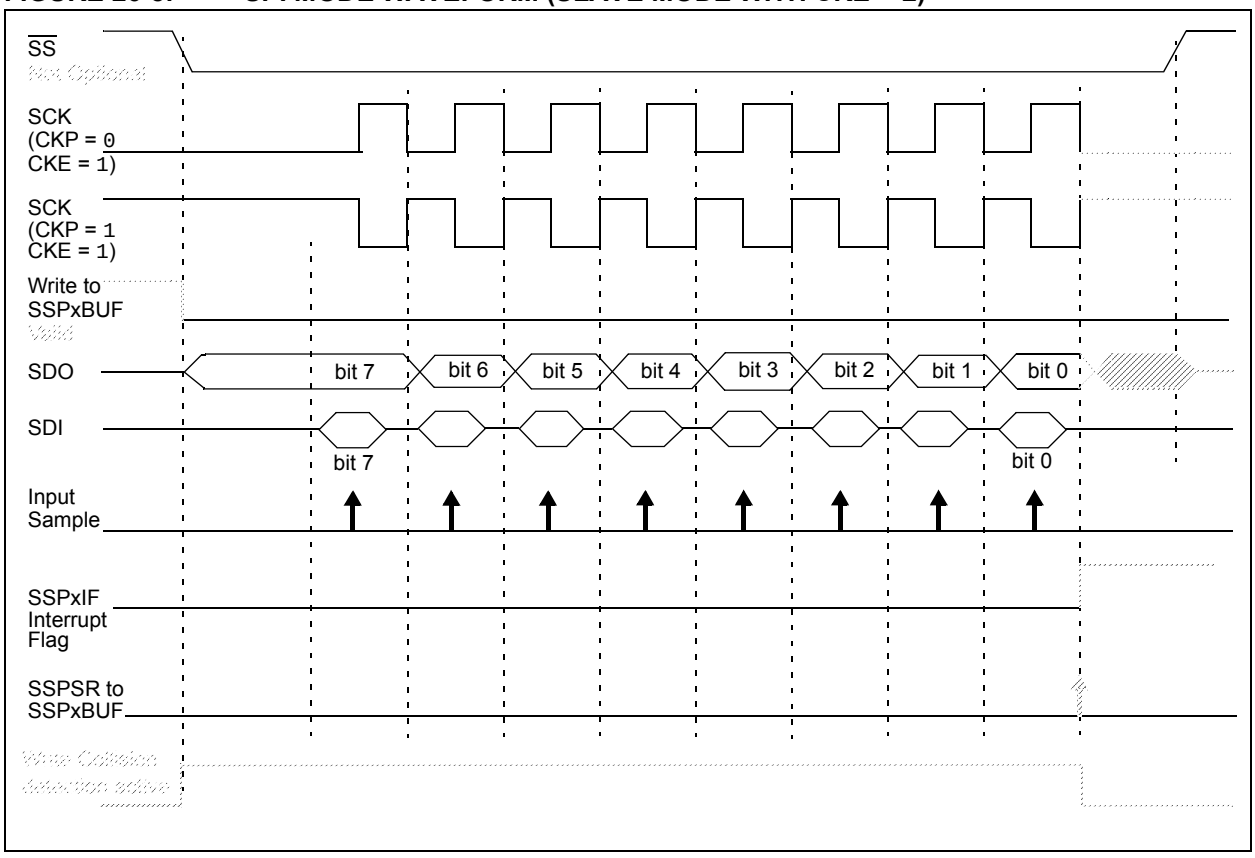
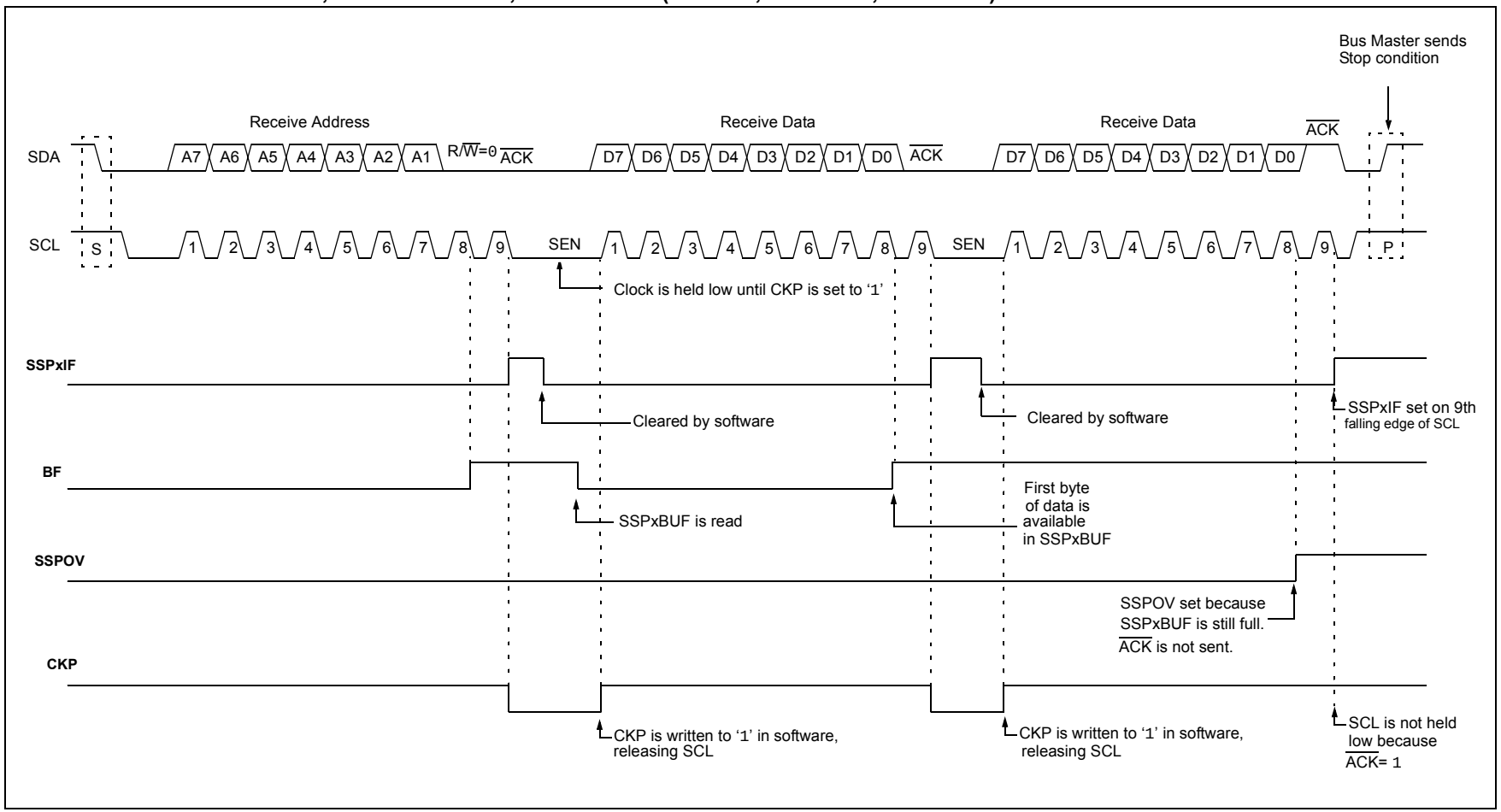


FIGURE 26-15: I²C SLAVE, 7-BIT ADDRESS, RECEPTION (SEN = 1, AHEN = 0, DHEN = 0)



27.4.4 BREAK CHARACTER SEQUENCE

The EUSART module has the capability of sending the special Break character sequences that are required by the LIN bus standard. A Break character consists of a Start bit, followed by 12 '0' bits and a Stop bit.

To send a Break character, set the SENDB and TXEN bits of the TXxSTA register. The Break character transmission is then initiated by a write to the TXxREG. The value of data written to TXxREG will be ignored and all '0's will be transmitted.

The SENDB bit is automatically reset by hardware after the corresponding Stop bit is sent. This allows the user to preload the transmit FIFO with the next transmit byte following the Break character (typically, the Sync character in the LIN specification).

The TRMT bit of the TXxSTA register indicates when the transmit operation is active or idle, just as it does during normal transmission. See Figure 27-9 for the timing of the Break character sequence.

27.4.4.1 Break and Sync Transmit Sequence

The following sequence will start a message frame header made up of a Break, followed by an auto-baud Sync byte. This sequence is typical of a LIN bus master.

1. Configure the EUSART for the desired mode.
2. Set the TXEN and SENDB bits to enable the Break sequence.
3. Load the TXxREG with a dummy character to initiate transmission (the value is ignored).
4. Write '55h' to TXxREG to load the Sync character into the transmit FIFO buffer.
5. After the Break has been sent, the SENDB bit is reset by hardware and the Sync character is then transmitted.

When the TXxREG becomes empty, as indicated by the TXxIF, the next data byte can be written to TXxREG.

27.4.5 RECEIVING A BREAK CHARACTER

The Enhanced EUSART module can receive a Break character in two ways.

The first method to detect a Break character uses the FERR bit of the RCxSTA register and the received data as indicated by RCxREG. The Baud Rate Generator is assumed to have been initialized to the expected baud rate.

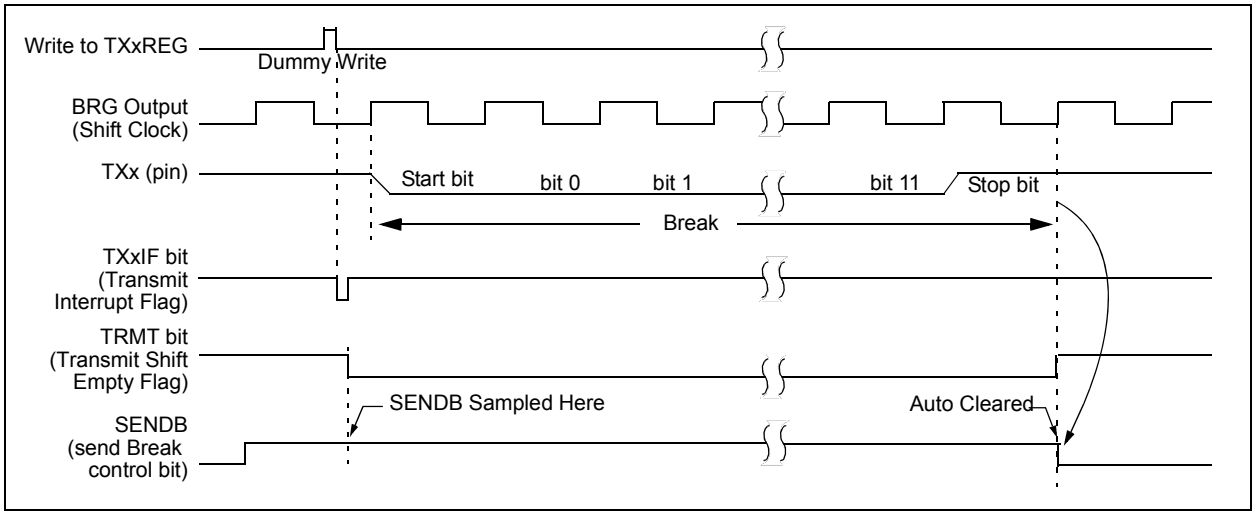
A Break character has been received when;

- RCxIF bit is set
- FERR bit is set
- RCxREG = 00h

The second method uses the Auto-Wake-up feature described in **Section 27.4.3 "Auto-Wake-up on Break"**. By enabling this feature, the EUSART will sample the next two transitions on RX/DT, cause an RCxIF interrupt, and receive the next data byte followed by another interrupt.

Note that following a Break character, the user will typically want to enable the Auto-Baud Detect feature. For both methods, the user can set the ABDEN bit of the BAUDxCON register before placing the EUSART in Sleep mode.

FIGURE 27-9: SEND BREAK CHARACTER SEQUENCE



27.5.2.3 EUSART Synchronous Slave Reception

The operation of the Synchronous Master and Slave modes is identical (**Section 27.5.1.5 "Synchronous Master Reception"**), with the following exceptions:

- Sleep
- CREN bit is always set, therefore the receiver is never idle
- SREN bit, which is a "don't care" in Slave mode

A character may be received while in Sleep mode by setting the CREN bit prior to entering Sleep. Once the word is received, the RSR register will transfer the data to the RCxREG register. If the RCxIE enable bit is set, the interrupt generated will wake the device from Sleep and execute the next instruction. If the GIE bit is also set, the program will branch to the interrupt vector.

27.5.2.4 Synchronous Slave Reception Setup:

1. Set the SYNC and SPEN bits and clear the CSRC bit.
2. Clear the ANSEL bit for both the CKx and DTx pins (if applicable).
3. If interrupts are desired, set the RCxIE bit of the PIE3 register and the GIE and PEIE bits of the INTCON register.
4. If 9-bit reception is desired, set the RX9 bit.
5. Set the CREN bit to enable reception.
6. The RCxIF bit will be set when reception is complete. An interrupt will be generated if the RCxIE bit was set.
7. If 9-bit mode is enabled, retrieve the Most Significant bit from the RX9D bit of the RCxSTA register.
8. Retrieve the eight Least Significant bits from the receive FIFO by reading the RCxREG register.
9. If an overrun error occurs, clear the error by either clearing the CREN bit of the RCxSTA register or by clearing the SPEN bit which resets the EUSART.

TABLE 27-10: SUMMARY OF REGISTERS ASSOCIATED WITH SYNCHRONOUS SLAVE RECEPTION

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Register on Page
BAUDxCON	ABDOVF	RCIDL	—	SCKP	BRG16	—	WUE	ABDEN	395
INTCON	GIE/GIEH	PEIE/GIEL	IPEN	—	—	INT2EDG	INT1EDG	INT0EDG	170
PIE3	RC2IE	TX2IE	RC1IE	TX1IE	BCL2IE	SSP2IE	BCL1IE	SSP1IE	182
PIR3	RC2IF	TX2IF	RC1IF	TX1IF	BCL2IF	SSP2IF	BCL1IF	SSP1IF	174
IPR3	RC2IP	TX2IP	RC1IP	TX1IP	BCL2IP	SSP2IP	BCL1IP	SSP1IP	190
RCxREG	EUSART Receive Data Register								399*
RCxSTA	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D	394
RxyPPS	—	—	—	RxyPPS<4:0>					218
RXxPPS	—	—	—	RXPPS<4:0>					216
TXxSTA	CSRC	TX9	TXEN	SYNC	SENDB	BRGH	TRMT	TX9D	393

Legend: — = unimplemented location, read as '0'. Shaded cells are not used for synchronous slave reception.
* Page provides register information.

31.5.8 CONTINUOUS SAMPLING MODE

Setting the ADCONT bit in the ADCON0 register automatically retriggers a new conversion cycle after updating the ADACC register. That means the ADGO bit is set to generate automatic retriggering, until the device Reset occurs or the A/D Stop-on-interrupt bit (ADSOI in the ADCON3 register) is set (correct logic).

31.5.9 DOUBLE SAMPLE CONVERSION

Double sampling is enabled by setting the ADDSEN bit of the ADCON1 register. When this bit is set, two conversions are required before the module will calculate threshold error (each conversion must still be triggered separately). The first conversion will set the ADMATH bit of the ADSTAT register and update ADACC, but will not calculate ADERR or trigger ADTIF. When the second conversion completes, the first value is transferred to ADPREV (depending on the setting of ADPSIS) and the value of the second conversion is placed into ADRES. Only upon the completion of the second conversion is ADERR calculated and ADTIF triggered (depending on the value of ADCALC).

31.6 Register Definitions: ADC Control

REGISTER 31-1: ADCON0: ADC CONTROL REGISTER 0

R/W-0/0	R/W-0/0	U-0	R/W-0/0	U-0	R/W-0/0	U-0	R/W/HC-0
ADON	ADCONT	—	ADCS	—	ADFM	—	ADGO
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as ‘0’
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
‘1’ = Bit is set	‘0’ = Bit is cleared	HC = Bit is cleared by hardware

bit 7 **ADON:** ADC Enable bit
1 = ADC is enabled
0 = ADC is disabled

bit 6 **ADCONT:** ADC Continuous Operation Enable bit
1 = ADGO is retriggered upon completion of each conversion trigger until ADTIF is set (if ADSOI is set) or until ADGO is cleared (regardless of the value of ADSOI)
0 = ADC is cleared upon completion of each conversion trigger

bit 5 **Unimplemented:** Read as ‘0’

bit 4 **ADCS:** ADC Clock Selection bit
1 = Clock supplied from FRC dedicated oscillator
0 = Clock supplied by FOSC, divided according to ADCLK register

bit 3 **Unimplemented:** Read as ‘0’

bit 2 **ADFM:** ADC results Format/alignment Selection
1 = ADRES and ADPREV data are right-justified
0 = ADRES and ADPREV data are left-justified, zero-filled

bit 1 **Unimplemented:** Read as ‘0’

bit 0 **ADGO:** ADC Conversion Status bit
1 = ADC conversion cycle in progress. Setting this bit starts an ADC conversion cycle. The bit is cleared by hardware as determined by the ADCONT bit
0 = ADC conversion completed/not in progress

REGISTER 31-11: ADCAP: ADC ADDITIONAL SAMPLE CAPACITOR SELECTION REGISTER

U-0	U-0	U-0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0
—	—	—	ADCAP<4:0>				
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7-5 **Unimplemented**: Read as '0'

bit 4-0 **ADCAP<4:0>**: ADC Additional Sample Capacitor Selection bits
11111 = 31 pF
11110 = 30 pF
11101 = 29 pF
•
•
•
00011 = 3 pF
00010 = 2 pF
00001 = 1 pF
00000 = No additional capacitance

REGISTER 31-12: ADRPT: ADC REPEAT SETTING REGISTER

R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0
ADRPT<7:0>							
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7-0 **Unimplemented**: Read as '0'

bit 7-0 **ADRPT<7:0>**: ADC Repeat Threshold bits
Counts the number of times that the ADC has been triggered and is used along with ADCNT to determine when the error threshold is checked when the computation is Low-pass Filter, Burst Average, or Average modes. See Table 31-3 for more details.

CPFSGT **Compare f with W, skip if f > W**

Syntax: CPFSGT f {,a}

Operands: $0 \le f \le 255$
$a \in [0,1]$

Operation: (f) – (W),
skip if (f) > (W)
(unsigned comparison)

Status Affected: None

Encoding:

0110	010a	ffff	ffff

Description: Compares the contents of data memory location 'f' to the contents of the W by performing an unsigned subtraction.
If the contents of 'f' are greater than the contents of WREG, then the fetched instruction is discarded and a NOP is executed instead, making this a 2-cycle instruction.
If 'a' is '0', the Access Bank is selected.
If 'a' is '1', the BSR is used to select the GPR bank.
If 'a' is '0' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever $f \le 95$ (5Fh). See **Section 35.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode"** for details.

Words: 1

Cycles: 1(2)
Note: 3 cycles if skip and followed by a 2-word instruction.

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	Read register 'f'	Process Data	No operation

If skip:

Q1	Q2	Q3	Q4
No operation	No operation	No operation	No operation

If skip and followed by 2-word instruction:

Q1	Q2	Q3	Q4
No operation	No operation	No operation	No operation
No operation	No operation	No operation	No operation

Example:
```
HERE      CPFSGT REG, 0
NGREATER  :
GREATER   :
```
Before Instruction
PC = Address (HERE)
W = ?
After Instruction
If REG > W;
PC = Address (GREATER)
If REG ≤ W;
PC = Address (NGREATER)

CPFSLT **Compare f with W, skip if f < W**

Syntax: CPFSLT f {,a}

Operands: $0 \le f \le 255$
$a \in [0,1]$

Operation: (f) – (W),
skip if (f) < (W)
(unsigned comparison)

Status Affected: None

Encoding:

0110	000a	ffff	ffff

Description: Compares the contents of data memory location 'f' to the contents of W by performing an unsigned subtraction.
If the contents of 'f' are less than the contents of W, then the fetched instruction is discarded and a NOP is executed instead, making this a 2-cycle instruction.
If 'a' is '0', the Access Bank is selected.
If 'a' is '1', the BSR is used to select the GPR bank.

Words: 1

Cycles: 1(2)
Note: 3 cycles if skip and followed by a 2-word instruction.

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	Read register 'f'	Process Data	No operation

If skip:

Q1	Q2	Q3	Q4
No operation	No operation	No operation	No operation

If skip and followed by 2-word instruction:

Q1	Q2	Q3	Q4
No operation	No operation	No operation	No operation
No operation	No operation	No operation	No operation

Example:
```
HERE    CPFSLT REG, 1
NLESS   :
LESS    :
```
Before Instruction
PC = Address (HERE)
W = ?
After Instruction
If REG < W;
PC = Address (LESS)
If REG ≥ W;
PC = Address (NLESS)

35.2 Extended Instruction Set

In addition to the standard 75 instructions of the PIC18 instruction set, PIC18(L)F2x/4xK40 devices also provide an optional extension to the core CPU functionality. The added features include eight additional instructions that augment indirect and indexed addressing operations and the implementation of Indexed Literal Offset Addressing mode for many of the standard PIC18 instructions.

The additional features of the extended instruction set are disabled by default. To enable them, users must set the XINST Configuration bit.

The instructions in the extended set can all be classified as literal operations, which either manipulate the File Select Registers, or use them for indexed addressing. Two of the instructions, ADDFSR and SUBFSR, each have an additional special instantiation for using FSR2. These versions (ADDULNK and SUBULNK) allow for automatic return after execution.

The extended instructions are specifically implemented to optimize re-entrant program code (that is, code that is recursive or that uses a software stack) written in high-level languages, particularly C. Among other things, they allow users working in high-level languages to perform certain operations on data structures more efficiently. These include:

- dynamic allocation and deallocation of software stack space when entering and leaving subroutines
- function pointer invocation
- software Stack Pointer manipulation
- manipulation of variables located in a software stack

A summary of the instructions in the extended instruction set is provided in Table 35-3. Detailed descriptions are provided in **Section 35.2.2 "Extended Instruction Set"**. The opcode field descriptions in Table 35-1 apply to both the standard and extended PIC18 instruction sets.

Note:	The instruction set extension and the Indexed Literal Offset Addressing mode were designed for optimizing applications written in C; the user may likely never use these instructions directly in assembler. The syntax for these commands is provided as a reference for users who may be reviewing code that has been generated by a compiler.

35.2.1 EXTENDED INSTRUCTION SYNTAX

Most of the extended instructions use indexed arguments, using one of the File Select Registers and some offset to specify a source or destination register. When an argument for an instruction serves as part of indexed addressing, it is enclosed in square brackets ("[ ]"). This is done to indicate that the argument is used as an index or offset. MPASM™ Assembler will flag an error if it determines that an index or offset value is not bracketed.

When the extended instruction set is enabled, brackets are also used to indicate index arguments in byte-oriented and bit-oriented instructions. This is in addition to other changes in their syntax. For more details, see **Section 35.2.3.1 "Extended Instruction Syntax with Standard PIC18 Commands"**.

Note:	In the past, square brackets have been used to denote optional arguments in the PIC18 and earlier instruction sets. In this text and going forward, optional arguments are denoted by braces ("{ }").

TABLE 35-3: EXTENSIONS TO THE PIC18 INSTRUCTION SET

Mnemonic, Operands		Description	Cycles	16-Bit Instruction Word				Status Affected
				MSb			LSb	
ADDFSR	f, k	Add literal to FSR	1	1110	1000	ffkk	kkkk	None
ADDULNK	k	Add literal to FSR2 and return	2	1110	1000	11kk	kkkk	None
CALLW		Call subroutine using WREG	2	0000	0000	0001	0100	None
MOVSF	z_s, f_d	Move z_s (source) to 1st word	2	1110	1011	0zzz	zzzz	None
		f_d (destination) 2nd word		1111	ffff	ffff	ffff	
MOVSS	z_s, z_d	Move z_s (source) to 1st word	2	1110	1011	1zzz	zzzz	None
		z_d (destination) 2nd word		1111	xxxx	xzzz	zzzz	
PUSHL	k	Store literal at FSR2, decrement FSR2	1	1110	1010	kkkk	kkkk	None
SUBFSR	f, k	Subtract literal from FSR	1	1110	1001	ffkk	kkkk	None
SUBULNK	k	Subtract literal from FSR2 and return	2	1110	1001	11kk	kkkk	None

TABLE 37-7: EXTERNAL CLOCK/OSCILLATOR TIMING REQUIREMENTS (CONTINUED)

Standard Operating Conditions (unless otherwise stated)							
Param No.	**Sym.**	**Characteristic**	**Min.**	**Typ†**	**Max.**	**Units**	**Conditions**
OS21	F_{CY}	Instruction Frequency	—	FOSC/4	—	MHz	
OS22	T_{CY}	Instruction Period	62.5	$1/F_{CY}$	—	ns	

* These parameters are characterized but not tested.

† Data in "Typ" column is at 3.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: Instruction cycle period (TCY) equals four times the input oscillator time base period. All specified values are based on characterization data for that particular oscillator type under standard operating conditions with the device executing code. Exceeding these specified limits may result in an unstable oscillator operation and/or higher than expected current consumption. All devices are tested to operate at "min" values with an external clock applied to OSC1 pin. When an external clock input is used, the "max" cycle time limit is "DC" (no clock) for all devices.

2: The system clock frequency (FOSC) is selected by the "main clock switch controls" as described in **Section 6.0 "Power-Saving Operation Modes"**.

3: The system clock frequency (FOSC) must meet the voltage requirements defined in the **Section 37.2 "Standard Operating Conditions"**.

4: LP, XT and HS oscillator modes require an appropriate crystal or resonator to be connected to the device. For clocking the device with the external square wave, one of the EC mode selections must be used.