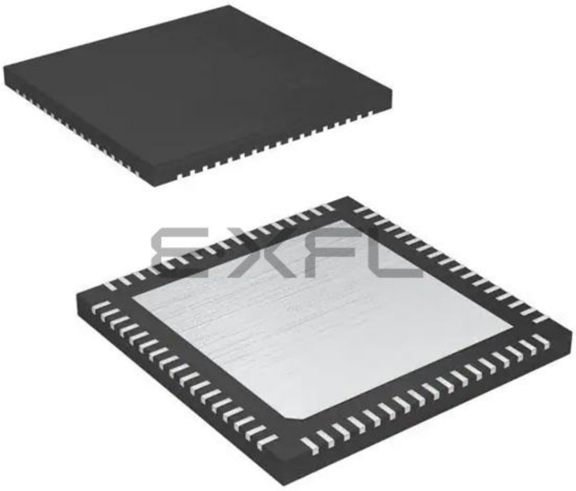
Infineon Technologies - CY8C5867LTI-LP028 Datasheet



Details	
Product Status	Active
Core Processor	ARM® Cortex®-M3
Core Size	32-Bit Single-Core
Speed	67MHz
Connectivity	I²C, LINbus, SPI, UART/USART, USB
Peripherals	CapSense, DMA, LCD, POR, PWM, WDT
Number of I/O	38
Program Memory Size	128KB (128K x 8)
Program Memory Type	FLASH
EEPROM Size	2K x 8
RAM Size	32K x 8
Voltage - Supply (Vcc/Vdd)	1.71V ~ 5.5V
Data Converters	A/D 1x20b, 1x12b; D/A 4x8b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	68-VFQFN Exposed Pad
Supplier Device Package	68-QFN (8x8)
Purchase URL	https://www.e-xfl.com/product-detail/infineon-technologies/cy8c5867lti-lp028

More Information

Cypress provides a wealth of data at www.cypress.com to help you to select the right PSoC device for your design, and to help you to quickly and effectively integrate the device into your design. For a comprehensive list of resources, see the knowledge base article KBA86521, How to Design with PSoC 3, PSoC 4, and PSoC 5LP. Following is an abbreviated list for PSoC 5LP:

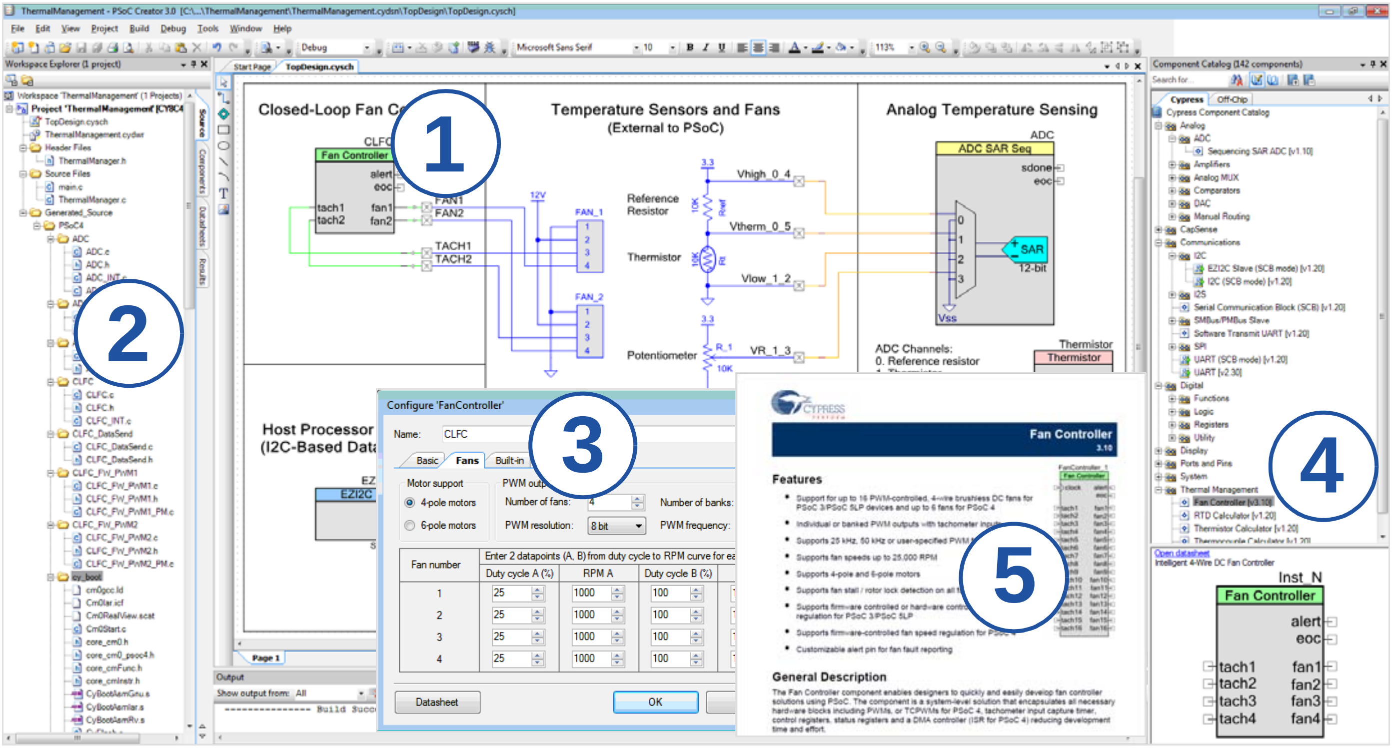
- Overview: PSoC Portfolio, PSoC Roadmap
- Product Selectors: PSoC 1, PSoC 3, PSoC 4, PSoC 5LP
  In addition, PSoC Creator includes a device selection tool.
- Application notes: Cypress offers a large number of PSoC application notes and code examples covering a broad range of topics, from basic to advanced level. Recommended application notes for getting started with PSoC 5LP are:
  - AN77759: Getting Started With PSoC 5LP
  - AN77835: PSoC 3 to PSoC 5LP Migration Guide
  - AN61290: Hardware Design Considerations
  - AN57821: Mixed Signal Circuit Board Layout
  - AN58304: Pin Selection for Analog Designs
  - AN81623: Digital Design Best Practices
  - AN73854: Introduction To Bootloaders
- Development Kits:
  - CY8CKIT-059 is a low-cost platform for prototyping, with a unique snap-away programmer and debugger on the USB connector.
  - CY8CKIT-050 is designed for analog performance, for developing high-precision analog, low-power, and low-voltage applications.
  - CY8CKIT-001 provides a common development platform for any one of the PSoC 1, PSoC 3, PSoC 4, or PSoC 5LP families of devices.
  - The MiniProg3 device provides an interface for flash programming and debug.
- Technical Reference Manuals (TRM)
  - Architecture TRM
  - Registers TRM
- Programming Specification

PSoC Creator

PSoC Creator is a free Windows-based Integrated Design Environment (IDE). It enables concurrent hardware and firmware design of PSoC 3, PSoC 4, and PSoC 5LP based systems. Create designs using classic, familiar schematic capture supported by over 100 pre-verified, production-ready PSoC Components; see the list of component datasheets. With PSoC Creator, you can:

1. Drag and drop component icons to build your hardware system design in the main design workspace
2. Codesign your application firmware with the PSoC hardware, using the PSoC Creator IDE C compiler
3. Configure components using the configuration tools
4. Explore the library of 100+ components
5. Review component datasheets

Figure 1. Multiple-Sensor Example Project in PSoC Creator



ADC Channels:
0. Reference resistor

Configure 'FanController'

Name: CLFC

Basic | **Fans** | Built-in

Motor support
● 4-pole motors
○ 6-pole motors

Number of fans: 4
PWM resolution: 8 bit
Number of banks:
PWM frequency:

Fan number	Duty cycle A (%)	RPM A	Duty cycle B (%)
1	25	1000	100
2	25	1000	100
3	25	1000	100
4	25	1000	100

Enter 2 datapoints (A, B) from duty cycle to RPM curve for ea

Datasheet | OK

Fan Controller
3.10

Features

- Support for up to 16 PWM-controlled, 4-wire brushless DC fans for PSoC 3/PSoC 5LP devices and up to 6 fans for PSoC 4
- Individual or banked PWM outputs with tachometer inputs
- Supports 25 kHz, 50 kHz or user-specified PWM
- Supports fan speeds up to 25,000 RPM
- Supports 4-pole and 6-pole motors
- Supports fan stall / rotor lock detection on all
- Supports firmware controlled or hardware contro regulation for PSoC 3/PSoC 5LP
- Supports firmware-controlled fan speed regulation for PSoC 4
- Customizable alert pin for fan fault reporting

General Description

The Fan Controller component enables designers to quickly and easily develop fan controller solutions using PSoC. The component is a system-level solution that encapsulates all necessary hardware blocks including PWMs, or TCPWMs for PSoC 4, tachometer input capture timer, control registers, status registers and a DMA controller (ISR for PSoC 4) reducing development time and effort.

Figure 2-4. 100-pin TQFP Part Pinout

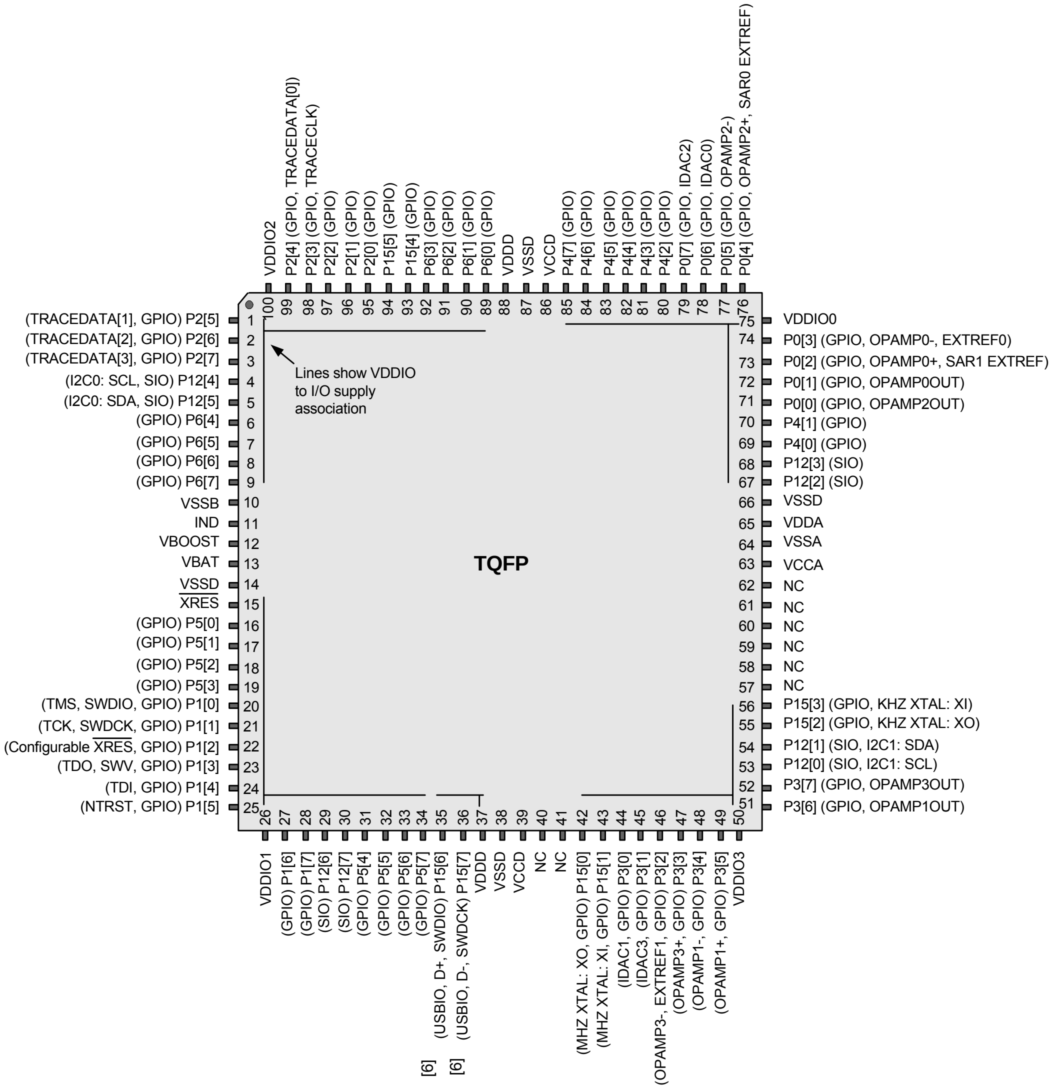


Table 2-1. V_{DDIO} **and Port Pin Associations**

VDDIO	Port Pins
VDDIO0	P0[7:0], P4[7:0], P12[3:2]
VDDIO1	P1[7:0], P5[7:0], P12[7:6]
VDDIO2	P2[7:0], P6[7:0], P12[5:4], P15[5:4]
VDDIO3	P3[7:0], P12[1:0], P15[3:0]
VDDD	P15[7:6] (USB D+, D-)

Note

5. Pins are Do Not Use (DNU) on devices without USB. The pin must be left floating.

3. Pin Descriptions

IDAC0, IDAC1, IDAC2, IDAC3. Low-resistance output pin for high-current DACs (IDAC).

Opamp0out, Opamp1out, Opamp2out, Opamp3out. High current output of uncommitted opamp.[7]

Extref0, Extref1. External reference input to the analog system.

SAR0 EXTREF, SAR1 EXTREF. External references for SAR ADCs

Opamp0-, Opamp1-, Opamp2-, Opamp3-. Inverting input to uncommitted opamp.

Opamp0+, Opamp1+, Opamp2+, Opamp3+. Noninverting input to uncommitted opamp.

GPIO. Provides interfaces to the CPU, digital peripherals, analog peripherals, interrupts, LCD segment drive, and CapSense.[7]

I2C0: SCL, I2C1: SCL. I^{2}C SCL line providing wake from sleep on an address match. Any I/O pin can be used for I^{2}C SCL if wake from sleep is not required.

I2C0: SDA, I2C1: SDA. I^{2}C SDA line providing wake from sleep on an address match. Any I/O pin can be used for I^{2}C SDA if wake from sleep is not required.

Ind. Inductor connection to boost pump.

kHz XTAL: Xo, kHz XTAL: Xi. 32.768-kHz crystal oscillator pin.

MHz XTAL: Xo, MHz XTAL: Xi. 4 to 25-MHz crystal oscillator pin.

nTRST. Optional JTAG Test Reset programming and debug port connection to reset the JTAG connection.

SIO. Provides interfaces to the CPU, digital peripherals and interrupts with a programmable high threshold voltage, analog comparator, high sink current, and high impedance state when the device is unpowered.

SWDCK. SWD Clock programming and debug port connection.

SWDIO. SWD Input and Output programming and debug port connection.

TCK. JTAG Test Clock programming and debug port connection.

TDI. JTAG Test Data In programming and debug port connection.

TDO. JTAG Test Data Out programming and debug port connection.

TMS. JTAG Test Mode Select programming and debug port connection.

TRACECLK. Cortex-M3 TRACEPORT connection, clocks TRACEDATA pins.

TRACEDATA[3:0]. Cortex-M3 TRACEPORT connections, output data.

SWV. SWV output.

USBIO, D+. Provides D+ connection directly to a USB 2.0 bus. May be used as a digital I/O pin; it is powered from VDDD instead of from a VDDIO. Pins are Do Not Use (DNU) on devices without USB.

USBIO, D-. Provides D- connection directly to a USB 2.0 bus. May be used as a digital I/O pin; it is powered from VDDD instead of from a VDDIO. Pins are Do Not Use (DNU) on devices without USB.

VBOOST. Power sense connection to boost pump.

VBAT. Battery supply to boost pump.

VCCA. Output of the analog core regulator or the input to the analog core. Requires a 1uF capacitor to VSSA. The regulator output is not designed to drive external circuits. **Note that if you use the device with an external core regulator (externally regulated mode), the voltage applied to this pin must not exceed the allowable range of 1.71 V to 1.89 V.** When using the internal core regulator, (internally regulated mode, the default), do not tie any power to this pin. For details see Power System on page 26.

VCCD. Output of the digital core regulator or the input to the digital core. The two VCCD pins must be shorted together, with the trace between them as short as possible, and a 1uF capacitor to VSSD. The regulator output is not designed to drive external circuits. **Note that if you use the device with an external core regulator (externally regulated mode), the voltage applied to this pin must not exceed the allowable range of 1.71 V to 1.89 V.** When using the internal core regulator (internally regulated mode, the default), do not tie any power to this pin. For details see Power System on page 26.

VDDA. Supply for all analog peripherals and analog core regulator. **VDDA must be the highest voltage present on the device. All other supply pins must be less than or equal to VDDA.**

VDDD. Supply for all digital peripherals and digital core regulator. VDDD must be less than or equal to VDDA.

VSSA. Ground for all analog peripherals.

VSSB. Ground connection for boost pump.

VSSD. Ground for all digital logic and I/O pins.

VDDIO0, VDDIO1, VDDIO2, VDDIO3. Supply for I/O pins. Each VDDIO must be tied to a valid operating voltage (1.71 V to 5.5 V), and must be less than or equal to VDDA.

XRES. External reset pin. Active low with internal pull-up.

Note

[7] GPIOs with opamp outputs are not recommended for use with CapSense.

The Cortex-M3 does not support ARM instructions for SRAM addresses.

- Bit-band support for the SRAM region. Atomic bit-level write and read operations for SRAM addresses.
- Unaligned data storage and access. Contiguous storage of data of different byte lengths.
- Operation at two privilege levels (privileged and user) and in two modes (thread and handler). Some instructions can only be executed at the privileged level. There are also two stack pointers: Main (MSP) and Process (PSP). These features support a multitasking operating system running one or more user-level processes.
- Extensive interrupt and system exception support.

4.1.2 Cortex-M3 Operating Modes

The Cortex-M3 operates at either the privileged level or the user level, and in either the thread mode or the handler mode. Because the handler mode is only enabled at the privileged level, there are actually only three states, as shown in Table 4-1.

Table 4-1. Operational Level

Condition	Privileged	User
Running an exception	Handler mode	Not used
Running main program	Thread mode	Thread mode

At the user level, access to certain instructions, special registers, configuration registers, and debugging components is blocked. Attempts to access them cause a fault exception. At the privileged level, access to all instructions and registers is allowed.

The processor runs in the handler mode (always at the privileged level) when handling an exception, and in the thread mode when not.

4.1.3 CPU Registers

The Cortex-M3 CPU registers are listed in Table 4-2. Registers R0-R15 are all 32 bits wide.

Table 4-2. Cortex M3 CPU Registers

Register	Description
R0-R12	General purpose registers R0-R12 have no special architecturally defined uses. Most instructions that specify a general purpose register specify R0-R12. ■ Low registers: Registers R0-R7 are accessible by all instructions that specify a general purpose register. ■ High registers: Registers R8-R12 are accessible by all 32-bit instructions that specify a general purpose register; they are not accessible by all 16-bit instructions.
R13	R13 is the stack pointer register. It is a banked register that switches between two 32-bit stack pointers: the main stack pointer (MSP) and the process stack pointer (PSP). The PSP is used only when the CPU operates at the user level in thread mode. The MSP is used in all other privilege levels and modes. Bits[0:1] of the SP are ignored and considered to be 0, so the SP is always aligned to a word (4 byte) boundary.
R14	R14 is the link register (LR). The LR stores the return address when a subroutine is called.
R15	R15 is the program counter (PC). Bit 0 of the PC is ignored and considered to be 0, so instructions are always aligned to a half word (2 byte) boundary.
xPSR	The program status registers are divided into three status registers, which are accessed either together or separately: ■ Application program status register (APSR) holds program execution status bits such as zero, carry, negative, in bits[27:31]. ■ Interrupt program status register (IPSR) holds the current exception number in bits[0:8]. ■ Execution program status register (EPSR) holds control bits for interrupt continuable and IF-THEN instructions in bits[10:15] and [25:26]. Bit 24 is always set to 1 to indicate Thumb mode. Trying to clear it causes a fault exception.
PRIMASK	A 1-bit interrupt mask register. When set, it allows only the nonmaskable interrupt (NMI) and hard fault exception. All other exceptions and interrupts are masked.
FAULTMASK	A 1-bit interrupt mask register. When set, it allows only the NMI. All other exceptions and interrupts are masked.
BASEPRI	A register of up to nine bits that define the masking priority level. When set, it disables all interrupts of the same or higher priority value. If set to 0 then the masking function is disabled.
CONTROL	A 2-bit register for controlling the operating mode. Bit 0: 0 = privileged level in thread mode, 1 = user level in thread mode. Bit 1: 0 = default stack (MSP) is used, 1 = alternate stack is used. If in thread mode or user level then the alternate stack is the PSP. There is no alternate stack for handler mode; the bit must be 0 while in handler mode.

4.2 Cache Controller

The CY8C58LP family has a 1 KB, 4-way set-associative instruction cache between the CPU and the flash memory. This improves instruction execution rate and reduces system power consumption by requiring less frequent flash access.

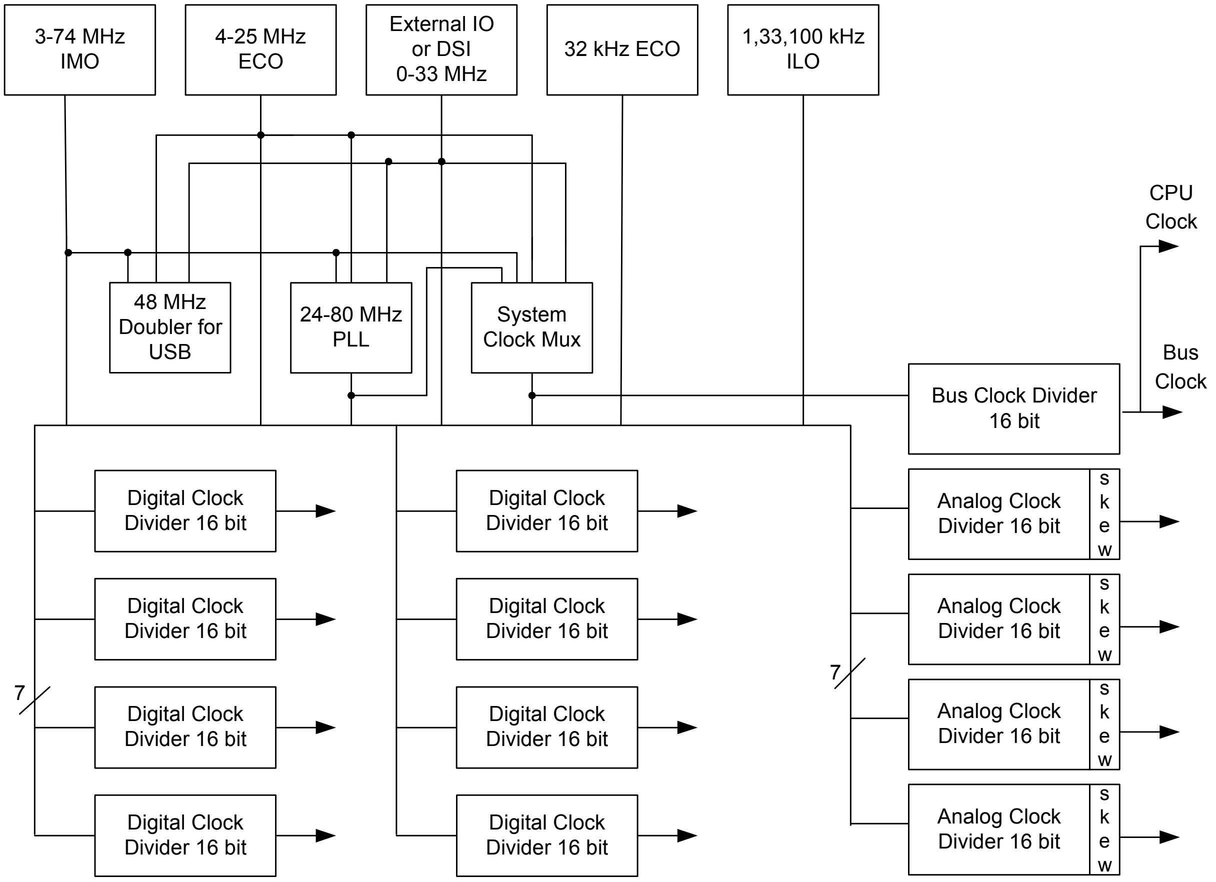
4.3 DMA and PHUB

The PHUB and the DMA controller are responsible for data transfer between the CPU and peripherals, and also data transfers between peripherals. The PHUB and DMA also control device configuration during boot. The PHUB consists of:

- A central hub that includes the DMA controller, arbiter, and router
- Multiple spokes that radiate outward from the hub to most peripherals

There are two PHUB masters: the CPU and the DMA controller. Both masters may initiate transactions on the bus. The DMA channels can handle peripheral communication without CPU intervention. The arbiter in the central hub determines which DMA channel is the highest priority if there are multiple requests.

Figure 6-1. Clocking Subsystem



6.1.1 Internal Oscillators

Figure 6-1 shows that there are two internal oscillators. They can be routed directly or divided. The direct routes may not have a 50% duty cycle. Divided clocks have a 50% duty cycle.

6.1.1.1 Internal Main Oscillator

In most designs the IMO is the only clock source required, due to its ±1% accuracy. The IMO operates with no external components and outputs a stable clock. A factory trim for each frequency range is stored in the device. With the factory trim, tolerance varies from ±1% at 3 MHz, up to ±7% at 74 MHz. The IMO, in conjunction with the PLL, allows generation of CPU and system clocks up to the device's maximum frequency (see USB Clock Domain on page 26). The IMO provides clock outputs at 3, 6, 12, 24, 48, and 74 MHz.

6.1.1.2 Clock Doubler

The clock doubler outputs a clock at twice the frequency of the input clock. The doubler works at input frequency of 24 MHz, providing 48 MHz for the USB. It can be configured to use a clock from the IMO, MHzECO, or the DSI (external pin).

6.1.1.3 Phase-Locked Loop

The PLL allows low frequency, high accuracy clocks to be multiplied to higher frequencies. This is a tradeoff between higher clock frequency and accuracy and, higher power consumption and increased startup time.

The PLL block provides a mechanism for generating clock frequencies based upon a variety of input sources. The PLL outputs clock frequencies in the range of 24 to 80 MHz. Its input and feedback dividers supply 4032 discrete ratios to create almost any desired system clock frequency. The accuracy of the PLL output depends on the accuracy of the PLL input source. The most common PLL use is to multiply the IMO clock at 3 MHz, where it is most accurate, to generate the CPU and system clocks up to the device’s maximum frequency.

The PLL achieves phase lock within 250 µs (verified by bit setting). It can be configured to use a clock from the IMO, MHzECO, or DSI (external pin). The PLL clock source can be used until lock is complete and signaled with a lock bit. The lock signal can be routed through the DSI to generate an interrupt. Disable the PLL before entering low power modes.

6.1.1.4 Internal Low-Speed Oscillator

The ILO provides clock frequencies for low power consumption, including the watchdog timer, and sleep timer. The ILO generates up to three different clocks: 1 kHz, 33 kHz, and 100 kHz.

The 1-kHz clock (CLK1K) is typically used for a background ‘heartbeat’ timer. This clock inherently lends itself to low power supervisory operations such as the watchdog timer and long sleep intervals using the central timewheel (CTW).

The central timewheel is a 1 kHz, free running, 13-bit counter clocked by the ILO. The central timewheel is always enabled except in hibernate mode and when the CPU is stopped during debug on chip mode. It can be used to generate periodic interrupts for timing purposes or to wake the system from a low power mode. Firmware can reset the central timewheel.

The central timewheel can be programmed to wake the system periodically and optionally issue an interrupt. This enables flexible, periodic wakeups from low power modes or coarse timing applications. Systems that require accurate timing should use the RTC capability instead of the central timewheel.

The 100-kHz clock (CLK100K) can be used as a low power system clock to run the CPU. It can also generate time intervals using the fast timewheel.

The fast timewheel is a 5-bit counter, clocked by the 100-kHz clock. It features programmable settings and automatically resets when the terminal count is reached. An optional interrupt can be generated each time the terminal count is reached. This enables flexible, periodic interrupts of the CPU at a higher rate than is allowed using the central timewheel.

The 33-kHz clock (CLK33K) comes from a divide-by-3 operation on CLK100K. This output can be used as a reduced accuracy version of the 32.768-kHz ECO clock with no need for a crystal.

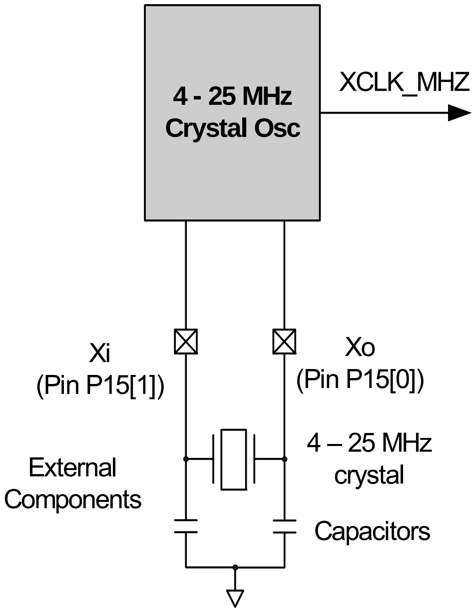
6.1.2 External Oscillators

Figure 6-1 shows that there are two external oscillators. They can be routed directly or divided. The direct routes may not have a 50% duty cycle. Divided clocks have a 50% duty cycle.

6.1.2.1 MHz External Crystal Oscillator

The MHzECO provides high frequency, high precision clocking using an external crystal (see Figure 6-2). It supports a wide variety of crystal types, in the range of 4 to 25 MHz. When used in conjunction with the PLL, it can generate CPU and system clocks up to the device's maximum frequency (see Phase-Locked Loop on page 25). The GPIO pins connecting to the external crystal and capacitors are fixed. MHzECO accuracy depends on the crystal chosen.

Figure 6-2. MHzECO Block Diagram

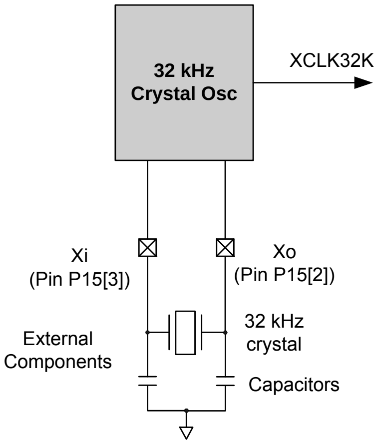


6.1.2.2 32.768 kHz ECO

The 32.768-kHz external crystal oscillator (32kHzECO) provides precision timing with minimal power consumption using an external 32.768-kHz watch crystal (see Figure 6-3). The 32kHzECO also connects directly to the sleep timer and provides the source for the RTC. The RTC uses a 1 second interrupt to implement the RTC functionality in firmware.

The oscillator works in two distinct power modes. This allows users to trade off power consumption with noise immunity from neighboring circuits. The GPIO pins connected to the external crystal and capacitors are fixed.

Figure 6-3. 32kHzECO Block Diagram



It is recommended that the external 32.768-kHz watch crystal have a load capacitance (CL) of 6 pF or 12.5 pF. Check the crystal manufacturer's datasheet. The two external capacitors, CL1 and CL2, are typically of the same value, and their total capacitance, CL1CL2 / (CL1 + CL2), including pin and trace capacitance, should equal the crystal CL value. For more information, refer to application note AN54439: PSoC 3 and PSoC 5 External Oscillators. See also pin capacitance specifications in the "GPIO" section on page 76.

6.1.2.3 Digital System Interconnect

The DSI provides routing for clocks taken from external clock oscillators connected to I/O. The oscillators can also be generated within the device in the digital system and UDBs.

While the primary DSI clock input provides access to all clocking resources, up to eight other DSI clocks (internally or externally generated) may be routed directly to the eight digital clock dividers. This is only possible if there are multiple precision clock sources.

6.1.3 Clock Distribution

All seven clock sources are inputs to the central clock distribution system. The distribution system is designed to create multiple high precision clocks. These clocks are customized for the design's requirements and eliminate the common problems found with limited resolution prescalers attached to peripherals. The clock distribution system generates several types of clock trees.

- The system clock is used to select and supply the fastest clock in the system for general system clock requirements and clock synchronization of the PSoC device.
- Bus clock 16-bit divider uses the system clock to generate the system's bus clock used for data transfers and the CPU. The CPU clock is directly derived from the bus clock.
- Eight fully programmable 16-bit clock dividers generate digital system clocks for general use in the digital system, as configured by the design's requirements. Digital system clocks can generate custom clocks derived from any of the seven clock sources for any purpose. Examples include baud rate generators, accurate PWM periods, and timer clocks, and many others. If more than eight digital clock dividers are required, the UDBs and fixed function timer/counter/PWMs can also generate clocks.
- Four 16-bit clock dividers generate clocks for the analog system components that require clocking, such as ADCs and mixers. The analog clock dividers include skew control to ensure that critical analog events do not occur simultaneously with digital switching events. This is done to reduce analog system noise.

Each clock divider consists of an 8-input multiplexer, a 16-bit clock divider (divide by 2 and higher) that generates ~50% duty cycle clocks, system clock resynchronization logic, and deglitch logic. The outputs from each digital clock tree can be routed into the digital system interconnect and then brought back into the clock system as an input, allowing clock chaining of up to 32 bits.

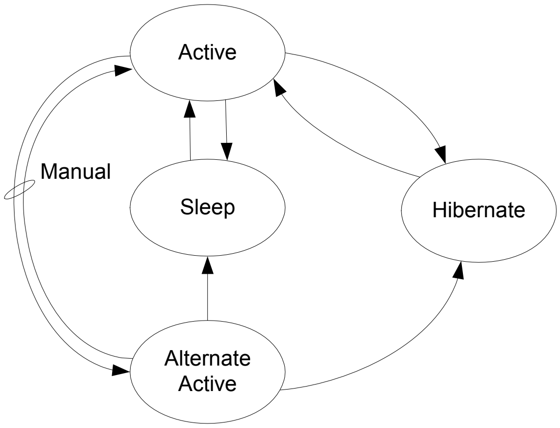
6.1.4 USB Clock Domain

The USB clock domain is unique in that it operates largely asynchronously from the main clock network. The USB logic contains a synchronous bus interface to the chip, while running on an asynchronous clock to process USB data. The USB logic requires a 48-MHz frequency. This frequency can be generated from different sources, including DSI clock at 48 MHz or doubled value of 24 MHz from internal oscillator, DSI signal, or crystal oscillator.

6.2 Power System

The power system consists of separate analog, digital, and I/O supply pins, labeled VDDA, VDDD, and VDDIOX, respectively. It also includes two internal 1.8 V regulators that provide the digital (VCCD) and analog (VCCA) supplies for the internal core logic. The output pins of the regulators (VCCD and VCCA) and the VDDIO pins must have capacitors connected as shown in Figure 6-4. The two V_{CCD} pins must be shorted together, with as short a trace as possible, and connected to a 1 µF ±10% X5R capacitor. The power system also contains a sleep regulator, an I^2C regulator, and a hibernate regulator.

Figure 6-5. Power Mode Transitions



6.2.1.1 Active Mode

Active mode is the primary operating mode of the device. When in active mode, the active configuration template bits control which available resources are enabled or disabled. When a resource is disabled, the digital clocks are gated, analog bias currents are disabled, and leakage currents are reduced as appropriate. User firmware can dynamically control subsystem power by setting and clearing bits in the active configuration template. The CPU can disable itself, in which case the CPU is automatically reenabled at the next wakeup event.

When a wakeup event occurs, the global mode is always returned to active, and the CPU is automatically enabled, regardless of its template settings. Active mode is the default global power mode upon boot.

6.2.1.2 Alternate Active Mode

Alternate Active mode is very similar to Active mode. In alternate active mode, fewer subsystems are enabled, to reduce power consumption. One possible configuration is to turn off the CPU and flash, and run peripherals at full speed.

6.2.1.3 Sleep Mode

Sleep mode reduces power consumption when a resume time of 15 µs is acceptable. The wake time is used to ensure that the regulator outputs are stable enough to directly enter active mode.

6.2.1.4 Hibernate Mode

In hibernate mode nearly all of the internal functions are disabled. Internal voltages are reduced to the minimal level to keep vital systems alive. Configuration state is preserved in hibernate mode and SRAM memory is retained. GPIOs configured as digital outputs maintain their previous values and external GPIO pin interrupt settings are preserved. The device can only return from hibernate mode in response to an external I/O interrupt. The resume time from hibernate mode is less than 100 µs.

To achieve an extremely low current, the hibernate regulator has limited capacity. This limits the frequency of any signal present on the input pins; no GPIO should toggle at a rate greater than 10 kHz while in hibernate mode. If pins must be toggled at a high rate while in a low power mode, use sleep mode instead.

6.2.1.5 Wakeup Events

Wakeup events are configurable and can come from an interrupt or device reset. A wakeup event restores the system to active mode. Firmware enabled interrupt sources include internally generated interrupts, power supervisor, central timewheel, and I/O interrupts. Internal interrupt sources can come from a variety of peripherals, such as analog comparators and UDBs. The central timewheel provides periodic interrupts to allow the system to wake up, poll peripherals, or perform real-time functions. Reset event sources include the external reset pin (XRES), WDT, and Precision Reset (PRES).

6.2.2 Boost Converter

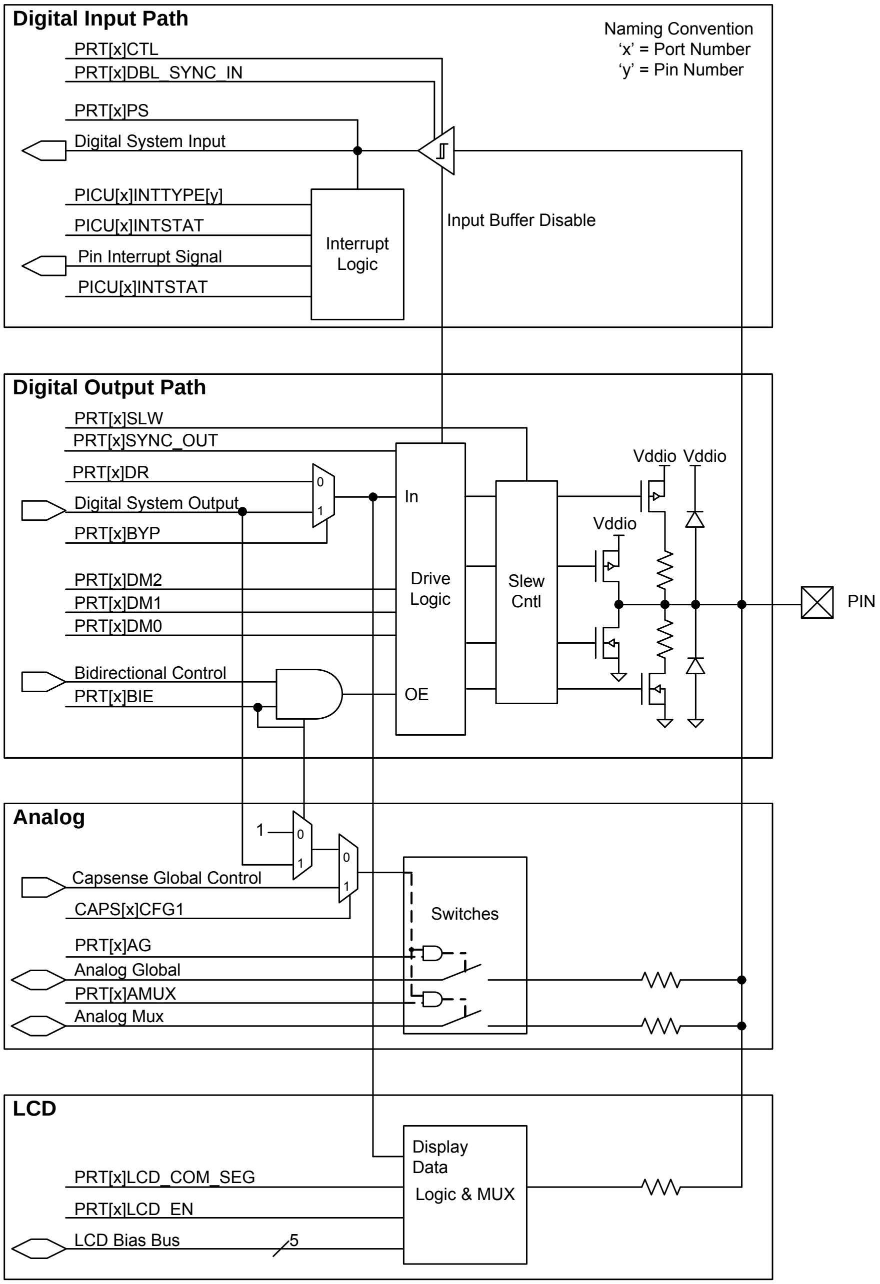
Applications that use a supply voltage of less than 1.71 V, such as solar panels or single cell battery supplies, may use the on-chip boost converter to generate a minimum of 1.8 V supply voltage. The boost converter may also be used in any system that requires a higher operating voltage than the supply provides such as driving 5.0 V LCD glass in a 3.3 V system. With the addition of an inductor, Schottky diode, and capacitors, it produces a selectable output voltage sourcing enough current to operate the PSoC and other on-board components.

The boost converter accepts an input voltage V_{BAT} from 0.5 V to 3.6 V, and can start up with V_{BAT} as low as 0.5 V. The converter provides a user configurable output voltage of 1.8 to 5.0 V (V_{OUT}) in 100 mV increments. V_{BAT} is typically less than V_{OUT}; if V_{BAT} is greater than or equal to V_{OUT}, then V_{OUT} will be slightly less than V_{BAT} due to resistive losses in the boost converter. The block can deliver up to 50 mA (I_{BOOST}) depending on configuration to both the PSoC device and external components. The sum of all current sinks in the design including the PSoC device, PSoC I/O pin loads, and external component loads must be less than the I_{BOOST} specified maximum current.

Four pins are associated with the boost converter: VBAT, VSSB, VBOOST, and IND. The boosted output voltage is sensed at the VBOOST pin and must be connected directly to the chip's supply inputs; VDDA, VDDD, and VDDIO if used to power the PSoC device.

The boost converter requires four components in addition to those required in a non-boost design, as shown in Figure 6-6 on page 30. A 22 µF capacitor (C_{BAT}) is required close to the V_{BAT} pin to provide local bulk storage of the battery voltage and provide regulator stability. A diode between the battery and V_{BAT} pin should not be used for reverse polarity protection because the diodes forward voltage drop reduces the V_{BAT} voltage. Between the V_{BAT} and IND pins, an inductor of 4.7 µH, 10 µH, or 22 µH is required. The inductor value can be optimized to increase the boost converter efficiency based on input voltage, output voltage, temperature, and current. Inductor size is determined by following the design guidance in this chapter and electrical specifications. The inductor must be placed within 1 cm of the V_{BAT} and IND pins and have a minimum saturation current of 750 mA. Between the IND and V_{BOOST} pins a Schottky diode must be placed within 1 cm of the pins. The Schottky diode shall have a forward current rating of at least 1.0 A and a reverse voltage of at least 20 V. A 22 µF bulk capacitor (C_{BOOST}) must be connected close to V_{BOOST} to provide regulator output stability. It is important to sum the total capacitance connected to the V_{BOOST} pin and ensure the maximum C_{BOOST} specification is not exceeded. All capacitors must be rated for a minimum of 10 V to minimize capacitive losses due to voltage de-rating.

Figure 6-8. GPIO Block Diagram



The USBIO pins (P15[7] and P15[6]), when enabled for I/O mode, have limited drive mode control. The drive mode is set using the PRT15.DM0[7, 6] register. A resistive pull option is also available at the USBIO pins, which can be enabled using the PRT15.DM1[7, 6] register. When enabled for USB mode, the drive mode control has no impact on the configuration of the USB pins. Unlike the GPIO and SIO configurations, the port wide configuration registers do not configure the USB drive mode bits. Table 6-7 shows the drive mode configuration for the USBIO pins.

Table 6-7. USBIO Drive Modes (P15[7] and P15[6])

PRT15.DM1[7,6] Pull up enable	PRT15.DM0[7,6] Drive Mode enable	PRT15.DR[7,6] = 1	PRT15.DR[7,6] = 0	Description
0	0	High Z	Strong Low	Open Drain, Strong Low
0	1	Strong High	Strong Low	Strong Outputs
1	0	Res High (5k)	Strong Low	Resistive Pull Up, Strong Low
1	1	Strong High	Strong Low	Strong Outputs

- High impedance analog

  The default reset state with both the output driver and digital input buffer turned off. This prevents any current from flowing in the I/O's digital input buffer due to a floating voltage. This state is recommended for pins that are floating or that support an analog voltage. High impedance analog pins do not provide digital input functionality.

  To achieve the lowest chip current in sleep modes, all I/Os must either be configured to the high impedance analog mode, or have their pins driven to a power supply rail by the PSoC device or by external circuitry.
- High impedance digital

  The input buffer is enabled for digital signal input. This is the standard high impedance (HiZ) state recommended for digital inputs.
- Resistive pull-up or resistive pull-down

  Resistive pull-up or pull-down, respectively, provides a series resistance in one of the data states and strong drive in the other. Pins can be used for digital input and output in these modes. Interfacing to mechanical switches is a common application for these modes. Resistive pull-up and pull-down are not available with SIO in regulated output mode.
- Open drain, drives high and open drain, drives low

  Open drain modes provide high impedance in one of the data states and strong drive in the other. Pins can be used for digital input and output in these modes. A common application for these modes is driving the I^{2}C bus signal lines.
- Strong drive

  Provides a strong CMOS output drive in either high or low state. This is the standard output mode for pins. Strong Drive mode pins must not be used as inputs under normal circumstances. This mode is often used to drive digital output signals or external FETs.
- Resistive pull-up and pull-down

  Similar to the resistive pull-up and resistive pull-down modes except the pin is always in series with a resistor. The high data state is pull-up while the low data state is pull-down. This mode is most often used when other signals that may cause shorts can drive the bus. Resistive pull-up and pull-down are not available with SIO in regulated output mode.

6.4.2 Pin Registers

Registers to configure and interact with pins come in two forms that may be used interchangeably.

All I/O registers are available in the standard port form, where each bit of the register corresponds to one of the port pins. This register form is efficient for quickly reconfiguring multiple port pins at the same time.

I/O registers are also available in pin form, which combines the eight most commonly used port register bits into a single register for each pin. This enables very fast configuration changes to individual pins with a single register write.

6.4.3 Bidirectional Mode

High speed bidirectional capability allows pins to provide both the high impedance digital drive mode for input signals and a second user selected drive mode such as strong drive (set using PRTxDM[2:0] registers) for output signals on the same pin, based on the state of an auxiliary control bus signal. The bidirectional capability is useful for processor busses and communications interfaces such as the SPI Slave MISO pin that requires dynamic hardware control of the output buffer.

The auxiliary control bus routes up to 16 UDB or digital peripheral generated output enable signals to one or more pins.

6.4.4 Slew Rate Limited Mode

GPIO and SIO pins have fast and slow output slew rate options for strong and open drain drive modes, not resistive drive modes. Because it results in reduced EMI, the slow edge rate option is recommended for signals that are not speed critical, generally less than 1 MHz. The fast slew rate is for signals between 1 MHz and 33 MHz. The slew rate is individually configurable for each pin, and is set by the PRTxSLW registers.

6.4.5 Pin Interrupts

All GPIO and SIO pins are able to generate interrupts to the system. All eight pins in each port interface to their own Port Interrupt Control Unit (PICU) and associated interrupt vector. Each pin of the port is independently configurable to detect rising edge, falling edge, both edge interrupts, or to not generate an interrupt.

Depending on the configured mode for each pin, each time an interrupt event occurs on a pin, its corresponding status bit of the interrupt status register is set to “1” and an interrupt request is sent to the interrupt controller. Each PICU has its own interrupt vector in the interrupt controller and the pin status register providing easy determination of the interrupt source down to the pin level.

Port pin interrupts remain active in all sleep modes allowing the PSoC device to wake from an externally generated interrupt. While level sensitive interrupts are not directly supported; Universal Digital Blocks (UDB) provide this functionality to the system when needed.

6.4.6 Input Buffer Mode

GPIO and SIO input buffers can be configured at the port level for the default CMOS input thresholds or the optional LVTTL input thresholds. All input buffers incorporate Schmitt triggers for input hysteresis. Additionally, individual pin input buffers can be disabled in any drive mode.

6.4.7 I/O Power Supplies

Up to four I/O pin power supplies are provided depending on the device and package. Each I/O supply must be less than or equal to the voltage on the chip’s analog (VDDA) pin. This feature allows users to provide different I/O voltage levels for different pins on the device. Refer to the specific device package pinout to determine VDDIO capability for a given port and pin. The SIO port pins support an additional regulated high output capability, as described in Adjustable Output Level.

6.4.8 Analog Connections

These connections apply only to GPIO pins. All GPIO pins may be used as analog inputs or outputs. The analog voltage present on the pin must not exceed the VDDIO supply voltage to which the GPIO belongs. Each GPIO may connect to one of the analog global busses or to one of the analog mux buses to connect any pin to any internal analog resource such as ADC or comparators. In addition, select pins provide direct connections to specific analog features such as the high current DACs or uncommitted opamps.

6.4.9 CapSense

This section applies only to GPIO pins. All GPIO pins may be used to create CapSense buttons and sliders[11]. See the “CapSense” section on page 59 for more information.

6.4.10 LCD Segment Drive

This section applies only to GPIO pins. All GPIO pins may be used to generate Segment and Common drive signals for direct glass drive of LCD glass. See the “LCD Direct Drive” section on page 58 for details.

6.4.11 Adjustable Output Level

This section applies only to SIO pins. SIO port pins support the ability to provide a regulated high output level for interface to external signals that are lower in voltage than the SIO’s respective VDDIO. SIO pins are individually configurable to output either the standard VDDIO level or the regulated output, which is based on an internally generated reference. Typically a voltage DAC (VDAC) is used to generate the reference (see Figure 6-12). The “DAC” section on page 59 has more details on VDAC use and reference routing to the SIO pins. Resistive pull-up and pull-down drive modes are not available with SIO in regulated output mode.

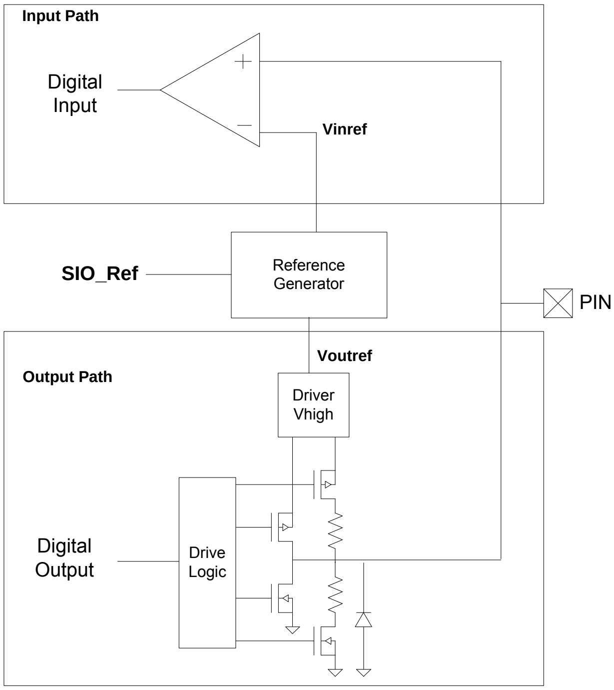
6.4.12 Adjustable Input Level

This section applies only to SIO pins. SIO pins by default support the standard CMOS and LVTTL input levels but also support a differential mode with programmable levels. SIO pins are grouped into pairs. Each pair shares a reference generator block which, is used to set the digital input buffer reference level for interface to external signals that differ in voltage from VDDIO. The reference sets the pins voltage threshold for a high logic level (see Figure 6-12). Available input thresholds are:

- 0.5 × VDDIO
- 0.4 × VDDIO
- 0.5 × VREF
- VREF

Typically a voltage DAC (VDAC) generates the VREF reference. DAC on page 59 has more details on VDAC use and reference routing to the SIO pins.

Figure 6-12. SIO Reference for Input and Output



Note

[11] GPIOs with opamp outputs are not recommended for use with CapSense.

7.2.2.3 Conditionals

Each datapath has two compares, with bit masking options. Compare operands include the two accumulators and the two data registers in a variety of configurations. Other conditions include zero detect, all ones detect, and overflow. These conditions are the primary datapath outputs, a selection of which can be driven out to the UDB routing matrix. Conditional computation can use the built in chaining to neighboring UDBs to operate on wider data widths without the need to use routing resources.

7.2.2.4 Variable MSB

The most significant bit of an arithmetic and shift function can be programmatically specified. This supports variable width CRC and PRS functions, and in conjunction with ALU output masking, can implement arbitrary width timers, counters and shift blocks.

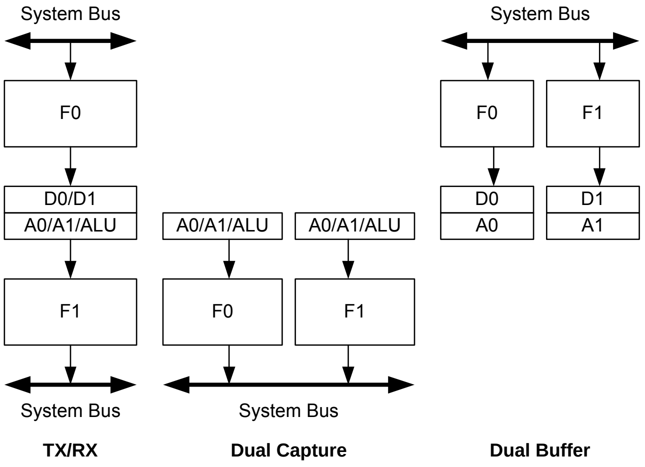
7.2.2.5 Built-in CRC/PRS

The datapath has built in support for single cycle Cyclic Redundancy Check (CRC) computation and Pseudo Random Sequence (PRS) generation of arbitrary width and arbitrary polynomial. CRC/PRS functions longer than 8 bits may be implemented in conjunction with PLD logic, or built in chaining may be use to extend the function into neighboring UDBs.

7.2.2.6 Input/Output FIFOs

Each datapath contains two four-byte deep FIFOs, which can be independently configured as an input buffer (system bus writes to the FIFO, datapath internal reads the FIFO), or an output buffer (datapath internal writes to the FIFO, the system bus reads from the FIFO). The FIFOs generate status that are selectable as datapath outputs and can therefore be driven to the routing, to interact with sequencers, interrupts, or DMA.

Figure 7-5. Example FIFO Configurations



7.2.2.7 Chaining

The datapath can be configured to chain conditions and signals such as carries and shift data with neighboring datapaths to create higher precision arithmetic, shift, CRC/PRS functions.

7.2.2.8 Time Multiplexing

In applications that are over sampled, or do not need high clock rates, the single ALU block in the datapath can be efficiently shared with two sets of registers and condition generators. Carry and shift out data from the ALU are registered and can be selected as inputs in subsequent cycles. This provides support for 16-bit functions in one (8-bit) datapath.

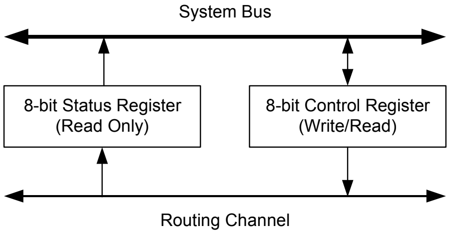
7.2.2.9 Datapath I/O

There are six inputs and six outputs that connect the datapath to the routing matrix. Inputs from the routing provide the configuration for the datapath operation to perform in each cycle, and the serial data inputs. Inputs can be routed from other UDB blocks, other device peripherals, device I/O pins, and so on. The outputs to the routing can be selected from the generated conditions, and the serial data outputs. Outputs can be routed to other UDB blocks, device peripherals, interrupt and DMA controller, I/O pins, and so on.

7.2.3 Status and Control Module

The primary purpose of this circuitry is to coordinate CPU firmware interaction with internal UDB operation.

Figure 7-6. Status and Control Registers



The bits of the control register, which may be written to by the system bus, are used to drive into the routing matrix, and thus provide firmware with the opportunity to control the state of UDB processing. The status register is read-only and it allows internal UDB state to be read out onto the system bus directly from internal routing. This allows firmware to monitor the state of UDB processing. Each bit of these registers has programmable connections to the routing matrix and routing connections are made depending on the requirements of the application.

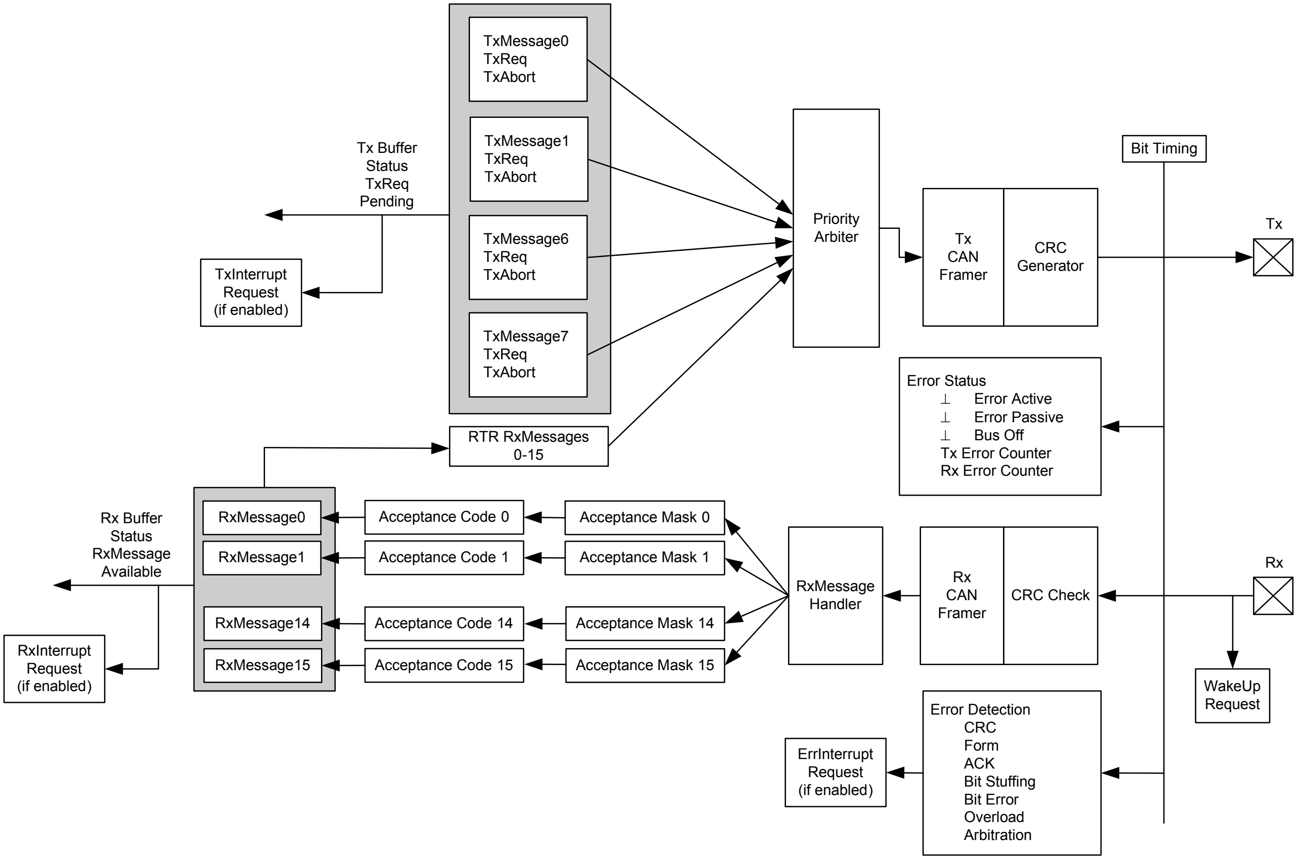
7.2.3.1 Usage Examples

As an example of control input, a bit in the control register can be allocated as a function enable bit. There are multiple ways to enable a function. In one method the control bit output would be routed to the clock control block in one or more UDBs and serve as a clock enable for the selected UDB blocks. A status example is a case where a PLD or datapath block generated a condition, such as a “compare true” condition that is captured and latched by the status register and then read (and cleared) by CPU firmware.

7.2.3.2 Clock Generation

Each subcomponent block of a UDB including the two PLDs, the datapath, and Status and Control, has a clock selection and control block. This promotes a fine granularity with respect to allocating clocking resources to UDB component blocks and allows unused UDB resources to be used by other functions for maximum system efficiency.

Figure 7-15. CAN Controller Block Diagram



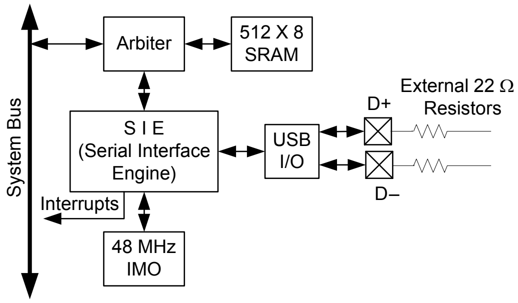
7.6 USB

PSoC includes a dedicated Full-Speed (12 Mbps) USB 2.0 transceiver supporting all four USB transfer types: control, interrupt, bulk, and isochronous. PSoC Creator provides full configuration support. USB interfaces to hosts through two dedicated USBIO pins, which are detailed in the I/O System and Routing on page 33.

USB includes the following features:

- Eight unidirectional data endpoints
- One bidirectional control endpoint 0 (EP0)
- Shared 512-byte buffer for the eight data endpoints
- Dedicated 8-byte buffer for EP0
- Three memory modes
  - Manual Memory Management with No DMA Access
  - Manual Memory Management with Manual DMA Access
  - Automatic Memory Management with Automatic DMA Access
- Internal 3.3 V regulator for transceiver
- Internal 48 MHz oscillator that auto locks to USB bus clock, requiring no external crystal for USB (USB equipped parts only)
- Interrupts on bus and each endpoint event, with device wakeup
- USB Reset, Suspend, and Resume operations
- Bus powered and self powered modes

Figure 7-16. USB



7.7 Timers, Counters, and PWMs

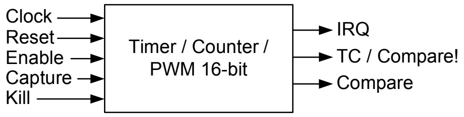
The Timer/Counter/PWM peripheral is a 16-bit dedicated peripheral providing three of the most common embedded peripheral features. As almost all embedded systems use some combination of timers, counters, and PWMs. Four of them have been included on this PSoC device family. Additional and more advanced functionality timers, counters, and PWMs can also be instantiated in Universal Digital Blocks (UDBs) as required. PSoC Creator allows you to choose the timer, counter, and PWM features that you need. The tool set utilizes the most optimal resources available.

The Timer/Counter/PWM peripheral can select from multiple clock sources, with input and output signals connected through the DSI routing. DSI routing allows input and output connections to any device pin and any internal digital signal accessible through the DSI. Each of the four instances has a compare output, terminal count output (optional complementary compare output), and programmable interrupt request line. The Timer/Counter/PWMs are configurable as free running, one shot, or Enable input controlled. The peripheral has timer reset and capture inputs, and a kill input for control of the comparator outputs. The peripheral supports full 16-bit capture.

Timer/Counter/PWM features include:

- 16-bit timer/counter/PWM (down count only)
- Selectable clock source
- PWM comparator (configurable for LT, LTE, EQ, GTE, GT)
- Period reload on start, reset, and terminal count
- Interrupt on terminal count, compare true, or capture
- Dynamic counter reads
- Timer capture mode
- Count while enable signal is asserted mode
- Free run mode
- One-shot mode (stop at end of period)
- Complementary PWM outputs with deadband
- PWM output kill

Figure 7-17. Timer/Counter/PWM



7.8 I^2C

PSoC includes a single fixed-function I^2C peripheral. Additional I^2C interfaces can be instantiated using Universal Digital Blocks (UDBs) in PSoC Creator, as required.

The I^2C peripheral provides a synchronous two-wire interface designed to interface the PSoC device with a two-wire I^2C serial communication bus. It is compatible[13] with I^2C Standard-mode, Fast-mode, and Fast-mode Plus devices as defined in the NXP I^2C-bus specification and user manual (UM10204). The I^2C bus I/O may be implemented with GPIO or SIO in open-drain modes.

To eliminate the need for excessive CPU intervention and overhead, I^2C specific support is provided for status detection and generation of framing bits. I^2C operates as a slave, a master, or multimaster (Slave and Master)[13]. In slave mode, the unit always listens for a start condition to begin sending or receiving data. Master mode supplies the ability to generate the Start and Stop conditions and initiate transactions. Multimaster mode provides clock synchronization and arbitration to allow multiple masters on the same bus. If Master mode is enabled and Slave mode is not enabled, the block does not generate interrupts on externally generated Start conditions. I^2C interfaces through the DSI routing and allows direct connections to any GPIO or SIO pins.

I^2C provides hardware address detect of a 7-bit address without CPU intervention. Additionally the device can wake from low power modes on a 7-bit hardware address match. If wakeup functionality is required, I^2C pin connections are limited to one of two specific pairs of SIO pins. See descriptions of SCL and SDA pins in Pin Descriptions on page 12.

I^2C features include:

- Slave and master, transmitter, and receiver operation
- Byte processing for low CPU overhead
- Interrupt or polling CPU interface
- Support for bus speeds up to 1 Mbps
- 7 or 10-bit addressing (10-bit addressing requires firmware support)
- SMBus operation (through firmware support - SMBus supported in hardware in UDBs)
- 7-bit hardware address compare
- Wake from low power modes on address match
- Glitch filtering (active and alternate-active modes only)

Data transfers follow the format shown in Figure 7-18. After the START condition (S), a slave address is sent. This address is 7 bits long followed by an eighth bit which is a data direction bit (R/W) - a 'zero' indicates a transmission (WRITE), a 'one' indicates a request for data (READ). A data transfer is always terminated by a STOP condition (P) generated by the master.

Notes

[12] The I^2C peripheral is non-compliant with the NXP I^2C specification in the following areas: analog glitch filter, I/O V_{OL}/I_{OL}, I/O hysteresis. The I^2C Block has a digital glitch filter (not available in sleep mode). The Fast-mode minimum fall-time specification can be met by setting the I/Os to slow speed mode. See the I/O Electrical Specifications in Inputs and Outputs on page 76 for details.

[13] Fixed-block I^2C does not support undefined bus conditions, nor does it support Repeated Start in Slave mode. These conditions should be avoided, or the UDB-based I^2C component should be used instead.

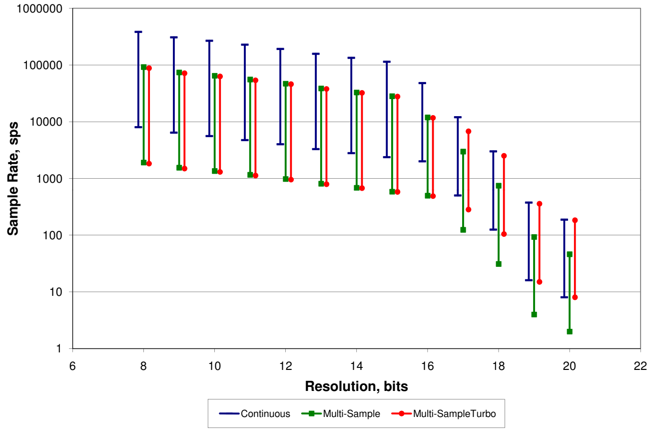
8.2 Delta-sigma ADC

The CY8C58LP device contains one delta-sigma ADC. This ADC offers differential input, high resolution and excellent linearity, making it a good ADC choice for both audio signal processing and measurement applications. The converter's nominal operation is 16 bits at 48 ksps. The ADC can be configured to output 20-bit resolution at data rates of up to 187 sps. At a fixed clock rate, resolution can be traded for faster data rates as shown in Table 8-1 and Figure 8-3.

Table 8-1. Delta-sigma ADC Performance

Bits	Maximum Sample Rate (sps)	SINAD (dB)
20	187	–
16	48 k	84
12	192 k	66
8	384 k	43

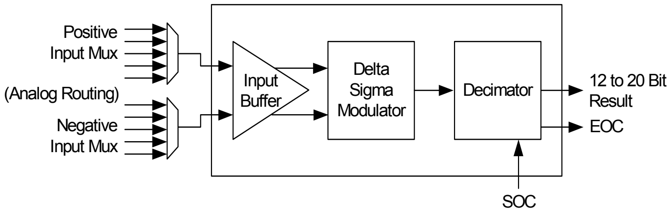
Figure 8-3. Delta-sigma ADC Sample Rates, Range = ±1.024 V



8.2.1 Functional Description

The ADC connects and configures three basic components, input buffer, delta-sigma modulator, and decimator. The basic block diagram is shown in Figure 8-4. The signal from the input muxes is delivered to the delta-sigma modulator either directly or through the input buffer. The delta-sigma modulator performs the actual analog to digital conversion. The modulator over-samples the input and generates a serial data stream output. This high speed data stream is not useful for most applications without some type of post processing, and so is passed to the decimator through the Analog Interface block. The decimator converts the high speed serial data stream into parallel ADC results. The modulator/decimator frequency response is $[(\sin x)/x]^4$.

Figure 8-4. Delta-sigma ADC Block Diagram



Resolution and sample rate are controlled by the Decimator. Data is pipelined in the decimator; the output is a function of the last four samples. When the input multiplexer is switched, the output data is not valid until after the fourth sample after the switch.

8.2.2 Operational Modes

The ADC can be configured by the user to operate in one of four modes: Single Sample, Multi Sample, Continuous, or Multi Sample (Turbo). All four modes are started by either a write to the start bit in a control register or an assertion of the Start of Conversion (SoC) signal. When the conversion is complete, a status bit is set and the output signal End of Conversion (EoC) asserts high and remains high until the value is read by either the DMA controller or the CPU.

8.2.2.1 Single Sample

In Single Sample mode, the ADC performs one sample conversion on a trigger. In this mode, the ADC stays in standby state waiting for the SoC signal to be asserted. When SoC is signaled the ADC performs four successive conversions. The first three conversions prime the decimator. The ADC result is valid and available after the fourth conversion, at which time the EoC signal is generated. To detect the end of conversion, the system may poll a control register for status or configure the external EoC signal to generate an interrupt or invoke a DMA request. When the transfer is done the ADC reenters the standby state where it stays until another SoC event.

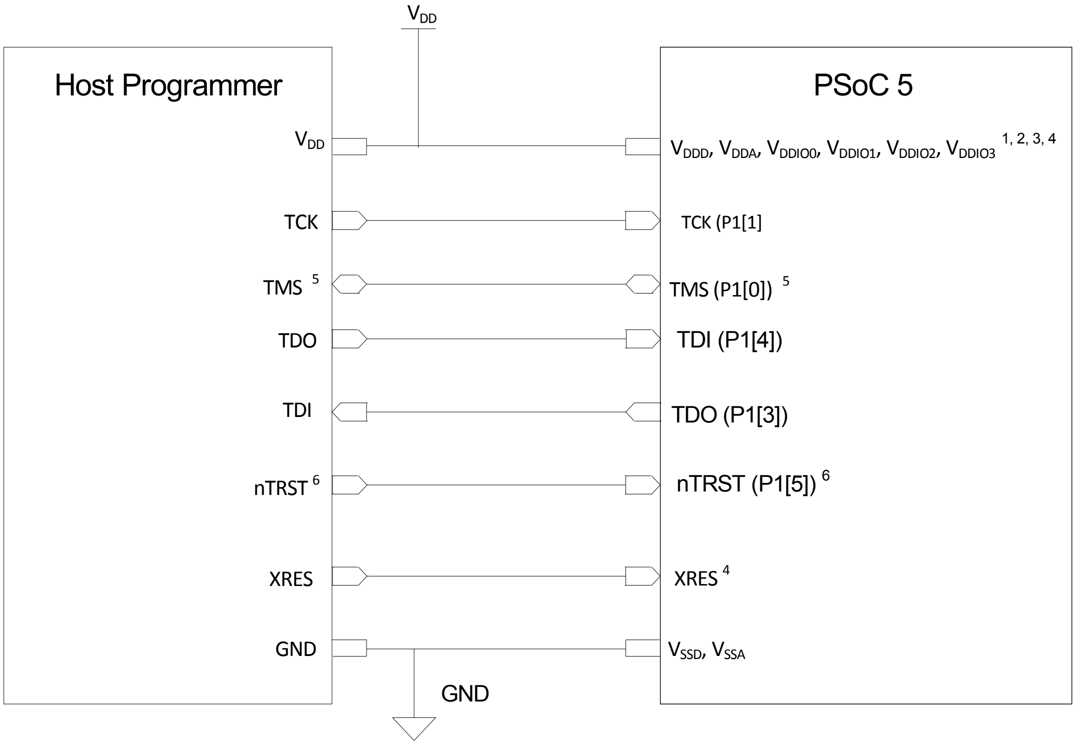
8.2.2.2 Continuous

Continuous sample mode is used to take multiple successive samples of a single input signal. Multiplexing multiple inputs should not be done with this mode. There is a latency of three conversion times before the first conversion result is available. This is the time required to prime the decimator. After the first result, successive conversions are available at the selected sample rate.

8.2.2.3 Multi Sample

Multi sample mode is similar to continuous mode except that the ADC is reset between samples. This mode is useful when the input is switched between multiple signals. The decimator is re-primed between each sample so that previous samples do not affect the current conversion. Upon completion of a sample, the next sample is automatically initiated. The results can be transferred using either firmware polling, interrupt, or DMA.

Figure 9-1. JTAG Interface Connections between PSoC 5LP and Programmer



[1] The voltage levels of Host Programmer and the PSoC 5 voltage domains involved in Programming should be same. The Port 1 JTAG pins and XRES pin are powered by V_{DDIO1}. So, V_{DDIO1} of PSoC 5 should be at same voltage level as host V_{DD}. Rest of PSoC 5 voltage domains (V_{DDD}, V_{DDA}, V_{DDIO0}, V_{DDIO2}, V_{DDIO3}) need not be at the same voltage level as host Programmer.

[2] Vdda must be greater than or equal to all other power supplies (Vddd, Vddio's) in PSoC 5.

[3] For Power cycle mode Programming, XRES pin is not required. But the Host programmer must have the capability to toggle power (Vddd, Vdda, All Vddio's) to PSoC 5. This may typically require external interface circuitry to toggle power which will depend on the programming setup. The power supplies can be brought up in any sequence, however, once stable, VDDA must be greater than or equal to all other supplies.

[4] For JTAG Programming, Device reset can also be done without connecting to the XRES pin or Power cycle mode by using the TMS,TCK,TDI, TDO pins of PSoC 5, and writing to a specific register. But this requires that the DPS setting in NVL is not equal to "Debug Ports Disabled".

[5] By default, PSoC 5 is configured for 4-wire JTAG mode unless user changes the DPS setting. So the TMS pin is unidirectional. But if the DPS setting is changed to non-JTAG mode, the TMS pin in JTAG is bi-directional as the SWD Protocol has to be used for acquiring the PSoC 5 device initially. After switching from SWD to JTAG mode, the TMS pin will be uni-directional. In such a case, unidirectional buffer should not be used on TMS line.

[6] nTRST JTAG pin (P1[5]) cannot be used to reset the JTAG TAP controlller during first time programming of PSoC 5 as the default setting is 4-wire JTAG (nTRST disabled). Use the TMS, TCK pins to do a reset of JTAG TAP controller.

Table 11-11 SIO AC Specifications[39]

Parameter	Description	Conditions	Min	Typ	Max	Units
TriseF	Rise time in fast strong mode (90/10%)	Cload = 25 pF, V_{DDIO} = 3.3 V	–	–	12	ns
TfallF	Fall time in fast strong mode (90/10%)	Cload = 25 pF, V_{DDIO} = 3.3 V	–	–	12	ns
TriseS	Rise time in slow strong mode (90/10%)	Cload = 25 pF, V_{DDIO} = 3.0 V	–	–	75	ns
TfallS	Fall time in slow strong mode (90/10%)	Cload = 25 pF, V_{DDIO} = 3.0 V	–	–	60	ns
Fsioout	SIO output operating frequency					
	2.7 V < V_{DDIO} < 5.5 V, Unregulated output (GPIO) mode, fast strong drive mode	90/10% V_{DDIO} into 25 pF	–	–	33	MHz
	1.71 V < V_{DDIO} < 2.7 V, Unregulated output (GPIO) mode, fast strong drive mode	90/10% V_{DDIO} into 25 pF	–	–	16	MHz
	3.3 V < V_{DDIO} < 5.5 V, Unregulated output (GPIO) mode, slow strong drive mode	90/10% V_{DDIO} into 25 pF	–	–	5	MHz
	1.71 V < V_{DDIO} < 3.3 V, Unregulated output (GPIO) mode, slow strong drive mode	90/10% V_{DDIO} into 25 pF	–	–	4	MHz
	2.7 V < V_{DDIO} < 5.5 V, Regulated output mode, fast strong drive mode	Output continuously switching into 25 pF	–	–	20	MHz
	1.71 V < V_{DDIO} < 2.7 V, Regulated output mode, fast strong drive mode	Output continuously switching into 25 pF	–	–	10	MHz
	1.71 V < V_{DDIO} < 5.5 V, Regulated output mode, slow strong drive mode	Output continuously switching into 25 pF	–	–	2.5	MHz
Fsioin	SIO input operating frequency					
	1.71 V ≤ V_{DDIO} ≤ 5.5 V	90/10% V_{DDIO}	–	–	33	MHz

Note

[39] Based on device characterization (Not production tested).

Figure 11-22. USBIO Output High Voltage and Current, GPIO Mode

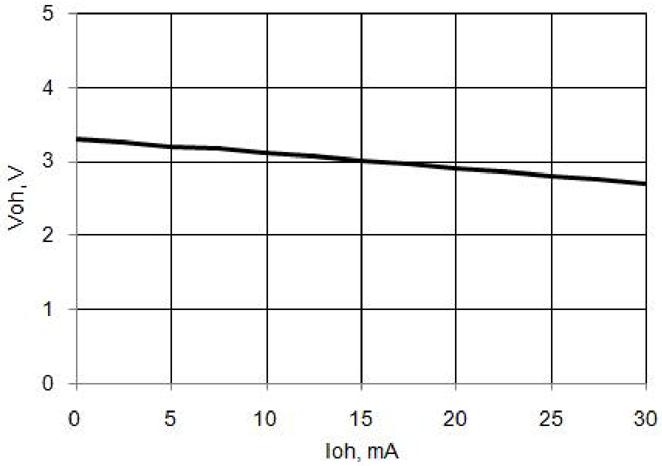


Figure 11-23. USBIO Output Rise and Fall Times, GPIO Mode, V_{DDD} = 3.3 V, 25 pF Load

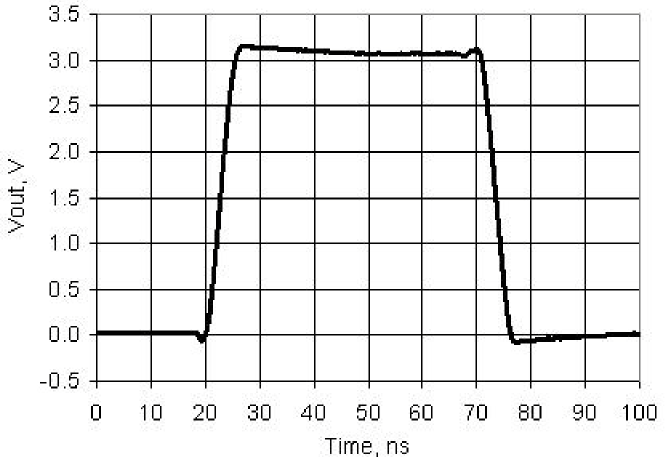


Table 11-14. USBIO AC Specifications[41]

Parameter	Description	Conditions	Min	Typ	Max	Units
Tdrate	Full-speed data rate average bit rate		12 – 0.25%	12	12 + 0.25%	MHz
Tjr1	Receiver data jitter tolerance to next transition		–8	–	8	ns
Tjr2	Receiver data jitter tolerance to pair transition		–5	–	5	ns
Tdj1	Driver differential jitter to next transition		–3.5	–	3.5	ns
Tdj2	Driver differential jitter to pair transition		–4	–	4	ns
Tfdeop	Source jitter for differential transition to SE0 transition		–2	–	5	ns
Tfeopt	Source SE0 interval of EOP		160	–	175	ns
Tfeopr	Receiver SE0 interval of EOP		82	–	–	ns
Tfst	Width of SE0 interval during differential transition		–	–	14	ns
Fgpio_out	GPIO mode output operating frequency	3 V ≤ V_{DDD} ≤ 5.5 V	–	–	20	MHz
		V_{DDD} = 1.71 V	–	–	6	MHz
Tr_gpio	Rise time, GPIO mode, 10%/90% V_{DDD}	V_{DDD} > 3 V, 25 pF load	–	–	12	ns
		V_{DDD} = 1.71 V, 25 pF load	–	–	40	ns
Tf_gpio	Fall time, GPIO mode, 90%/10% V_{DDD}	V_{DDD} > 3 V, 25 pF load	–	–	12	ns
		V_{DDD} = 1.71 V, 25 pF load	–	–	40	ns

Note

[41] Based on device characterization (Not production tested).

11.5.7 Current Digital-to-analog Converter (IDAC)

All specifications are based on use of the low-resistance IDAC output pins (see Pin Descriptions on page 12 for details). See the IDAC component data sheet in PSoC Creator for full electrical specifications and APIs.

Unless otherwise specified, all charts and graphs show typical values.

Table 11-34. IDAC DC Specifications

Parameter	Description	Conditions	Min	Typ	Max	Units
	Resolution		–	–	8	bits
I_{OUT}	Output current at code = 255	Range = 2.04 mA, code = 255, $V_{DDA} \geq 2.7$ V, Rload = 600 Ω	–	2.04	–	mA
		Range = 2.04 mA, High mode, code = 255, $V_{DDA} \leq 2.7$ V, Rload = 300 Ω	–	2.04	–	mA
		Range = 255 µA, code = 255, Rload = 600 Ω	–	255	–	µA
		Range = 31.875 µA, code = 255, Rload = 600 Ω	–	31.875	–	µA
	Monotonicity		–	–	Yes	
Ezs	Zero scale error		–	0	±1	LSB
Eg	Gain error	Range = 2.04 mA	–	–	±2.5	%
		Range = 255 µA	–	–	±2.5	%
		Range = 31.875 µA	–	–	±3.5	%
TC_Eg	Temperature coefficient of gain error	Range = 2.04 mA	–	–	0.045	% / °C
		Range = 255 µA	–	–	0.045	% / °C
		Range = 31.875 µA	–	–	0.05	% / °C
INL	Integral nonlinearity	Sink mode, range = 255 µA, Codes 8–255, Rload = 2.4 kΩ, Cload = 15 pF	–	±0.9	±1	LSB
		Source mode, range = 255 µA, Codes 8–255, Rload = 2.4 kΩ, Cload = 15 pF	–	±1.2	±1.6	LSB
		Source mode, range = 31.875 µA, Codes 8–255, Rload = 20 kΩ, Cload = 15 pF[60]	–	±0.9	±2	LSB
		Sink mode, range = 31.875 µA, Codes 8–255, Rload = 20 kΩ, Cload = 15 pF[60]	–	±0.9	±2	LSB
		Source mode, range = 2.04 mA, Codes 8–255, Rload = 600 Ω, Cload = 15 pF[60]	–	±0.9	±2	LSB
		Sink mode, range = 2.04 mA, Codes 8–255, Rload = 600 Ω, Cload = 15 pF[60]	–	±0.6	±1	LSB

Notes

[60] Based on device characterization (Not production tested).

11.8 PSoC System Resources

Specifications are valid for –40 °C ≤ T_A ≤ 105 °C and T_J ≤ 120 °C, except where noted. Specifications are valid for 1.71 V to 5.5 V, except where noted.

11.8.1 POR with Brown Out

For brown out detect in regulated mode, V_{DDD} and V_{DDA} must be ≥ 2.0 V. Brown out detect is not available in externally regulated mode.

Table 11-71. Precise Low-Voltage Reset (PRES) with Brown Out DC Specifications

Parameter	Description	Conditions	Min	Typ	Max	Units
PRESR	Rising trip voltage	Factory trim	1.64	–	1.68	V
PRESF	Falling trip voltage		1.62	–	1.66	V

Table 11-72. Power-On-Reset (POR) with Brown Out AC Specifications[90]

Parameter	Description	Conditions	Min	Typ	Max	Units
PRES_TR[91]	Response time		–	–	0.5	µs
	V_{DDD}/V_{DDA} droop rate	Sleep mode	–	5	–	V/sec

11.8.2 Voltage Monitors

Table 11-73. Voltage Monitors DC Specifications

Parameter	Description	Conditions	Min	Typ	Max	Units
LVI	Trip voltage					
	LVI_A/D_SEL[3:0] = 0000b		1.68	1.73	1.77	V
	LVI_A/D_SEL[3:0] = 0001b		1.89	1.95	2.01	V
	LVI_A/D_SEL[3:0] = 0010b		2.14	2.20	2.27	V
	LVI_A/D_SEL[3:0] = 0011b		2.38	2.45	2.53	V
	LVI_A/D_SEL[3:0] = 0100b		2.62	2.71	2.79	V
	LVI_A/D_SEL[3:0] = 0101b		2.87	2.95	3.04	V
	LVI_A/D_SEL[3:0] = 0110b		3.11	3.21	3.31	V
	LVI_A/D_SEL[3:0] = 0111b		3.35	3.46	3.56	V
	LVI_A/D_SEL[3:0] = 1000b		3.59	3.70	3.81	V
	LVI_A/D_SEL[3:0] = 1001b		3.84	3.95	4.07	V
	LVI_A/D_SEL[3:0] = 1010b		4.08	4.20	4.33	V
	LVI_A/D_SEL[3:0] = 1011b		4.32	4.45	4.59	V
	LVI_A/D_SEL[3:0] = 1100b		4.56	4.70	4.84	V
	LVI_A/D_SEL[3:0] = 1101b		4.83	4.98	5.13	V
	LVI_A/D_SEL[3:0] = 1110b		5.05	5.21	5.37	V
	LVI_A/D_SEL[3:0] = 1111b		5.30	5.47	5.63	V
HVI	Trip voltage		5.57	5.75	5.92	V

Table 11-74. Voltage Monitors AC Specifications

Parameter	Description	Conditions	Min	Typ	Max	Units
LVI_tr[91]	Response time		–	–	1	µs

Notes

[90] Based on device characterization (Not production tested).

[91] This value is calculated, not measured.