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Details

Product Status	Active
Core Processor	PIC
Core Size	8-Bit
Speed	20MHz
Connectivity	-
Peripherals	Brown-out Detect/Reset, LVD, POR, WDT
Number of I/O	11
Program Memory Size	3.5KB (2K x 14)
Program Memory Type	FLASH
EEPROM Size	-
RAM Size	128 x 8
Voltage - Supply (Vcc/Vdd)	2V ~ 5.5V
Data Converters	-
Oscillator Type	Internal
Operating Temperature	-40°C ~ 125°C (TA)
Mounting Type	Surface Mount
Package / Case	14-TSSOP (0.173", 4.40mm Width)
Supplier Device Package	14-TSSOP
Purchase URL	https://www.e-xfl.com/product-detail/microchip-technology/pic16f636-e-st

PIC12F635/PIC16F636/639

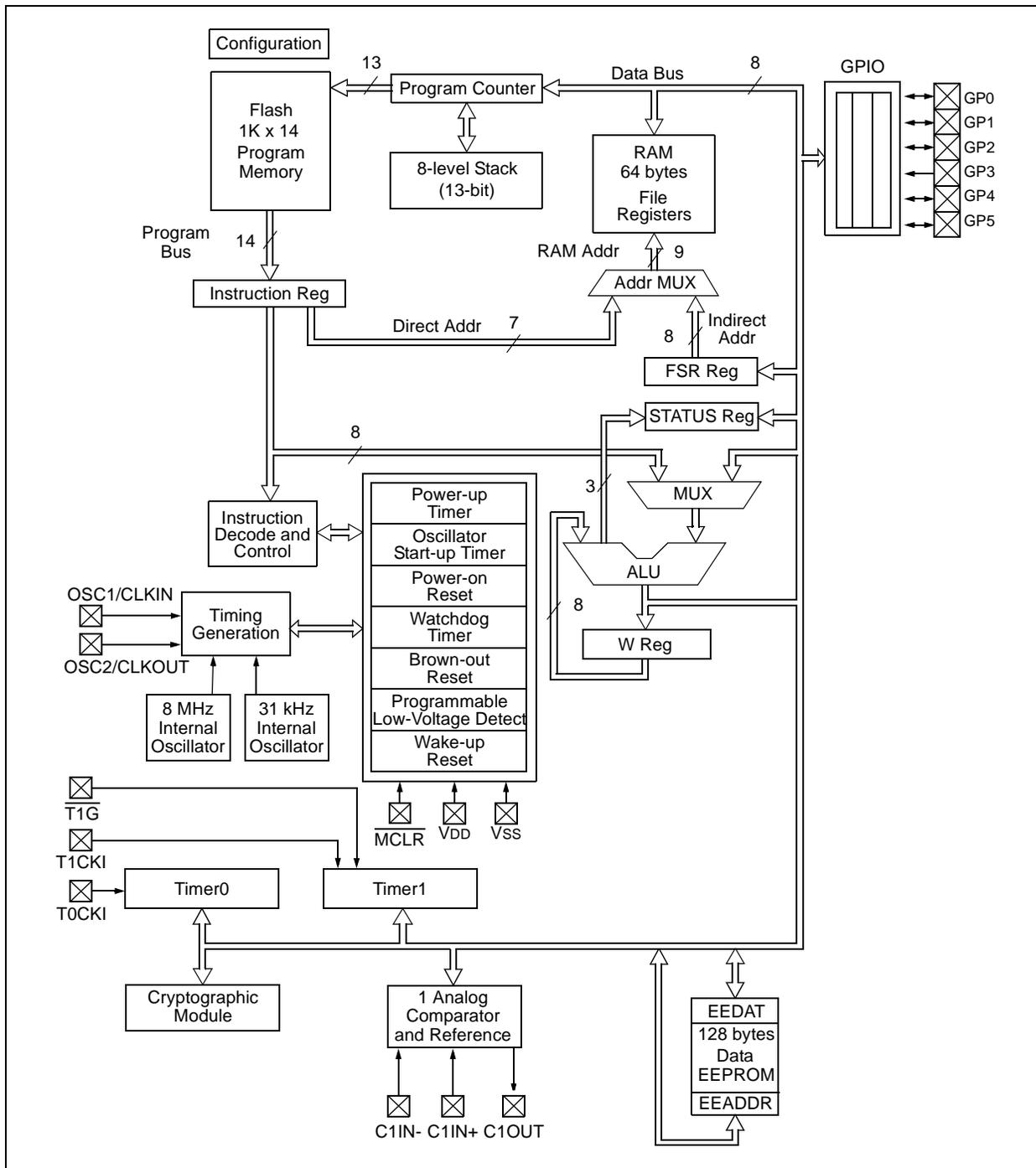
1.0 DEVICE OVERVIEW

This document contains device specific information for the PIC12F635/PIC16F636/639 devices.

Block Diagrams and pinout descriptions of the devices are as follows:

- PIC12F635 (Figure 1-1, Table 1-1)
- PIC16F636 (Figure 1-2, Table 1-2)
- PIC16F639 (Figure 1-3, Table 1-3)

FIGURE 1-1: PIC12F635 BLOCK DIAGRAM



PIC12F635/PIC16F636/639

TABLE 2-1: PIC12F635 SPECIAL FUNCTION REGISTERS SUMMARY BANK 0

Addr	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR/BOR/WUR	Page
Bank 0											
00h	INDF	Addressing this location uses contents of FSR to address data memory (not a physical register)								xxxx xxxx	32,137
01h	TMR0	Timer0 Module Register								xxxx xxxx	61,137
02h	PCL	Program Counter's (PC) Least Significant Byte								0000 0000	32,137
03h	STATUS	IRP	RP1	RP0	\overline{TO}	\overline{PD}	Z	DC	C	0001 1xxx	26,137
04h	FSR	Indirect Data Memory Address Pointer								xxxx xxxx	32,137
05h	GPIO	—	—	GP5	GP4	GP3	GP2	GP1	GP0	--xx xx00	47,137
06h	—	Unimplemented								—	—
07h	—	Unimplemented								—	—
08h	—	Unimplemented								—	—
09h	—	Unimplemented								—	—
0Ah	PCLATH	—	—	—	Write Buffer for upper 5 bits of Program Counter				---	0000	32,137
0Bh	INTCON	GIE	PEIE	TOIE	INTE	RAIE	TOIF	INTF	RAIF ⁽²⁾	0000 000x	28,137
0Ch	PIR1	EEIF	LVDIF	CRIF	—	C1IF	OSFIF	—	TMR1IF	000- 00-0	30,137
0Dh	—	Unimplemented								—	—
0Eh	TMR1L	Holding Register for the Least Significant Byte of the 16-bit TMR1								xxxx xxxx	64,137
0Fh	TMR1H	Holding Register for the Most Significant Byte of the 16-bit TMR1								xxxx xxxx	64,137
10h	T1CON	T1GINV	TMR1GE	T1CKPS1	T1CKPS0	T1OSCEN	$\overline{T1SYNC}$	TMR1CS	TMR1ON	0000 0000	68,137
11h	—	Unimplemented								—	—
12h	—	Unimplemented								—	—
13h	—	Unimplemented								—	—
14h	—	Unimplemented								—	—
15h	—	Unimplemented								—	—
16h	—	Unimplemented								—	—
17h	—	Unimplemented								—	—
18h	WDTCN	—	—	—	WDTPS3	WDTPS2	WDTPS1	WDTPS0	SWDTEN	---0 1000	144,137
19h	CMCON0	—	COUT	—	CINV	CIS	CM2	CM1	CM0	-0-0 0000	79,137
1Ah	CMCON1	—	—	—	—	—	—	T1GSS	CMSYNC	---- --10	82,137
1Bh	—	Unimplemented								—	—
1Ch	—	Unimplemented								—	—
1Dh	—	Unimplemented								—	—
1Eh	—	Unimplemented								—	—
1Fh	—	Unimplemented								—	—

Legend: — = Unimplemented locations read as '0', u = unchanged, x = unknown, q = value depends on condition, shaded = unimplemented

- Note** 1: Other (non Power-up) Resets include \overline{MCLR} Reset and Watchdog Timer Reset during normal operation.
 2: \overline{MCLR} and WDT Reset do not affect the previous value data latch. The RAIF bit will be cleared upon Reset but will set again if the mismatch exists.

PIC12F635/PIC16F636/639

3.3 Clock Source Modes

Clock Source modes can be classified as external or internal.

- External Clock modes rely on external circuitry for the clock source. Examples are: Oscillator modules (EC mode), quartz crystal resonators or ceramic resonators (LP, XT and HS modes) and Resistor-Capacitor (RC) mode circuits.
- Internal clock sources are contained internally within the Oscillator module. The Oscillator module has two internal oscillators: the 8 MHz High-Frequency Internal Oscillator (HFINTOSC) and the 31 kHz Low-Frequency Internal Oscillator (LFINTOSC).

The system clock can be selected between external or internal clock sources via the System Clock Select (SCS) bit of the OSCCON register. See **Section 3.6 “Clock Switching”** for additional information.

3.4 External Clock Modes

3.4.1 OSCILLATOR START-UP TIMER (OST)

If the Oscillator module is configured for LP, XT or HS modes, the Oscillator Start-up Timer (OST) counts 1024 oscillations from OSC1. This occurs following a Power-on Reset (POR) and when the Power-up Timer (PWRT) has expired (if configured), or a wake-up from Sleep. During this time, the program counter does not increment and program execution is suspended. The OST ensures that the oscillator circuit, using a quartz crystal resonator or ceramic resonator, has started and is providing a stable system clock to the Oscillator module. When switching between clock sources, a delay is required to allow the new clock to stabilize. These oscillator delays are shown in Table 3-1.

In order to minimize latency between external oscillator start-up and code execution, the Two-Speed Clock Start-up mode can be selected (see **Section 3.7 “Two-Speed Clock Start-up Mode”**).

TABLE 3-1: OSCILLATOR DELAY EXAMPLES

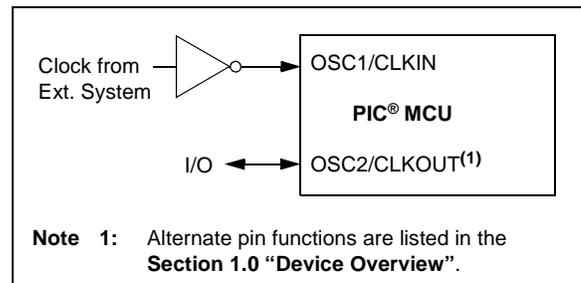
Switch From	Switch To	Frequency	Oscillator Delay
Sleep/POR	LFINTOSC HFINTOSC	31 kHz 125 kHz to 8 MHz	Oscillator Warm-Up Delay (TWARM)
Sleep/POR	EC, RC	DC – 20 MHz	2 instruction cycles
LFINTOSC (31 kHz)	EC, RC	DC – 20 MHz	1 cycle of each
Sleep/POR	LP, XT, HS	32 kHz to 20 MHz	1024 Clock Cycles (OST)
LFINTOSC (31 kHz)	HFINTOSC	125 kHz to 8 MHz	1 μ s (approx.)

3.4.2 EC MODE

The External Clock (EC) mode allows an externally generated logic level as the system clock source. When operating in this mode, an external clock source is connected to the OSC1 input and the OSC2 is available for general purpose I/O. Figure 3-2 shows the pin connections for EC mode.

The Oscillator Start-up Timer (OST) is disabled when EC mode is selected. Therefore, there is no delay in operation after a Power-on Reset (POR) or wake-up from Sleep. Because the PIC[®] MCU design is fully static, stopping the external clock input will have the effect of halting the device while leaving all data intact. Upon restarting the external clock, the device will resume operation as if no time had elapsed.

FIGURE 3-2: EXTERNAL CLOCK (EC) MODE OPERATION



7.11 Comparator Voltage Reference

The Comparator Voltage Reference module provides an internally generated voltage reference for the comparators. The following features are available:

- Independent from Comparator operation
- Two 16-level voltage ranges
- Output clamped to V_{SS}
- Ratiometric with V_{DD}
- Fixed Voltage Reference

The VRCON register (Register 7-5) controls the Voltage Reference module shown in Figure 7-10.

7.11.1 INDEPENDENT OPERATION

The comparator voltage reference is independent of the comparator configuration. Setting the VREN bit of the VRCON register will enable the voltage reference.

7.11.2 OUTPUT VOLTAGE SELECTION

The CVREF voltage reference has 2 ranges with 16 voltage levels in each range. Range selection is controlled by the VRR bit of the VRCON register. The 16 levels are set with the VR<3:0> bits of the VRCON register.

The CVREF output voltage is determined by the following equations:

EQUATION 7-1: CVREF OUTPUT VOLTAGE (INTERNAL CVREF)

$$\begin{aligned}
 &VRR = 1 \text{ (low range):} \\
 &CVREF = (VR<3:0>/24) \times VDD \\
 &VRR = 0 \text{ (high range):} \\
 &CVREF = (VDD/4) + (VR<3:0> \times VDD/32)
 \end{aligned}$$

EQUATION 7-2: CVREF OUTPUT VOLTAGE (EXTERNAL CVREF)

$$\begin{aligned}
 &VRR = 1 \text{ (low range):} \\
 &CVREF = (VR<3:0>/24) \times VLADDER \\
 &VRR = 0 \text{ (high range):} \\
 &CVREF = (VLADDER/4) + (VR<3:0> \times VLADDER/32) \\
 &VLADDER = VDD \text{ or } ([VREF+] - [VREF-]) \text{ or } VREF+
 \end{aligned}$$

The full range of V_{SS} to V_{DD} cannot be realized due to the construction of the module. See Figure 7-10.

7.11.3 OUTPUT CLAMPED TO V_{SS}

The CVREF output voltage can be set to V_{SS} with no power consumption by configuring VRCON as follows:

- VREN = 0
- VRR = 1
- VR<3:0> = 0000

This allows the comparator to detect a zero-crossing while not consuming additional CVREF module current.

7.11.4 OUTPUT RATIOMETRIC TO V_{DD}

The comparator voltage reference is V_{DD} derived and therefore, the CVREF output changes with fluctuations in V_{DD}. The tested absolute accuracy of the Comparator Voltage Reference can be found in **Section 15.0 “Electrical Specifications”**.

11.0 ANALOG FRONT-END (AFE) FUNCTIONAL DESCRIPTION (PIC16F639 ONLY)

The PIC16F639 device consists of the PIC16F636 device and low frequency (LF) Analog Front-End (AFE), with the AFE section containing three analog-input channels for signal detection and LF talk-back. This section describes the Analog Front-End (AFE) in detail.

The PIC16F639 device can detect a 125 kHz input signal as low as 1 mVpp and transmit data by using internal LF talk-back modulation or via an external transmitter. The PIC16F639 can also be used for various bidirectional communication applications. Figure 11-3 and Figure 11-4 show application examples of the device.

Each analog input channel has internal tuning capacitance, sensitivity control circuits, an input signal strength limiter and an LF talk-back modulation transistor. An Automatic Gain Control (AGC) loop is used for all three input channel gains. The output of each channel is OR'd and fed into a demodulator. The digital output is passed to the LFDATA pin. Figure 11-1 shows the block diagram of the AFE and Figure 11-2 shows the LC input path.

There are a total of eight Configuration registers. Six of them are used for AFE operation options, one for column parity bits and one for status indication of AFE operation. Each register has 9 bits including one row parity bit. These registers are readable and writable by SPI (Serial Protocol Interface) commands except for the STATUS register, which is read-only.

11.1 RF Limiter

The RF Limiter limits LC pin input voltage by de-Q'ing the attached LC resonant circuit. The absolute voltage limit is defined by the silicon process's maximum allowed input voltage (see **Section 15.0 "Electrical Specifications"**). The limiter begins de-Q'ing the external LC antenna when the input voltage exceeds VDE_Q, progressively de-Q'ing harder to reduce the antenna input voltage.

The signal levels from all 3 channels are combined such that the limiter attenuates all 3 channels uniformly, in respect to the channel with the strongest signal.

11.2 Modulation Circuit

The modulation circuit consists of a modulation transistor (FET), internal tuning capacitors and external LC antenna components. The modulation transistor and the internal tuning capacitors are connected between the LC input pin and LCCOM pin. Each LC input has its own modulation transistor.

When the modulation transistor turns on, its low Turn-on Resistance (RM) clamps the induced LC antenna voltage. The coil voltage is minimized when the modulation transistor turns-on and maximized when the modulation transistor turns-off. The modulation transistor's low Turn-on Resistance (RM) results in a high modulation depth.

The LF talk-back is achieved by turning on and off the modulation transistor.

The modulation data comes from the microcontroller section via the digital SPI interface as "Clamp On", "Clamp Off" commands. Only those inputs that are enabled will execute the clamp command. A basic block diagram of the modulation circuit is shown in Figure 11-1 and Figure 11-2.

The modulation FET is also shorted momentarily after Soft Reset and Inactivity timer time-out.

11.3 Tuning Capacitor

Each channel has internal tuning capacitors for external antenna tuning. The capacitor values are programmed by the Configuration registers up to 63 pF, 1 pF per step.

Note: The user can control the tuning capacitor by programming the AFE Configuration registers.

11.4 Variable Attenuator

The variable attenuator is used to attenuate, via AGC control, the input signal voltage to avoid saturating the amplifiers and demodulators.

Note: The variable attenuator function is accomplished by the device itself. The user cannot control its function.

11.5 Sensitivity Control

The sensitivity of each channel can be reduced by the channel's Configuration register sensitivity setting. This is used to desensitize the channel from optimum.

Note: The user can desensitize the channel sensitivity by programming the AFE Configuration registers.

TABLE 11-1: TYPICAL OUTPUT ENABLE FILTER TIMING

OEH <1:0>	OEL <1:0>	TOEH (ms)	TOEL (ms)	TOET (ms)
01	00	1	1	3
01	01	1	1	3
01	10	1	2	4
01	11	1	4	6
10	00	2	1	4
10	01	2	1	4
10	10	2	2	5
10	11	2	4	8
11	00	4	1	6
11	01	4	1	6
11	10	4	2	8
11	11	4	4	10
00	xx	Filter Disabled		

Note 1: Typical at room temperature and VDD = 3.0V, 32 kHz oscillator.

TOEH is measured from the rising edge of the demodulator output to the first falling edge. The pulse width must fall within $TOEH \leq t \leq TOET$.

TOEL is measured from the falling edge of the demodulator output to the rising edge of the next pulse. The pulse width must fall within $TOEL \leq t \leq TOET$.

TOET is measured from rising edge to the next rising edge (i.e., the sum of TOEH and TOEL). The pulse width must be $t \leq TOET$. If the Configuration Register 0 (Register 11-1), OEL<8:7> is set to '00', then TOEH must not exceed TOET and TOEL must not exceed TINACT.

The filter will reset, requiring a complete new successive high and low period to enable LFDATA, under the following conditions.

- The received high is not greater than the configured minimum TOEH value.
- During TOEH, a loss of signal > 56 μ s. A loss of signal < 56 μ s may or may not cause a filter Reset.
- The received low is not greater than the configured minimum TOEL value.
- The received sequence exceeds the maximum TOET value:
 - $TOEH + TOEL > TOET$
 - or $TOEH > TOET$
 - or $TOEL > TOET$
- A Soft Reset SPI command is received.

If the filter resets due to a long high ($TOEH > TOET$), the high-pulse timer will not begin timing again until after a gap of TE and another low-to-high transition occurs on the demodulator output.

Disabling the output enable filter disables the TOEH and TOEL requirement and the AFE passes all received LF data. See Figure 11-10, Figure 11-11 and Figure 11-12 for examples.

When viewed from an application perspective, from the pin input, the actual output enable filter timing must factor in the analog delays in the input path (such as demodulator charge and discharge times).

- $TOEH - TDR + TDF$
- $TOEL + TDR - TDF$

The output enable filter starts immediately after TGAP, the gap after AGC stabilization period.

11.16 Input Sensitivity Control

The AFE is designed to have typical input sensitivity of 3 mVPP. This means any input signal with amplitude greater than 3 mVPP can be detected. The AFE's internal AGC loop regulates the detecting signal amplitude when the input level is greater than approximately 20 mVPP. This signal amplitude is called "AGC-active level". The AGC loop regulates the input voltage so that the input signal amplitude range will be kept within the linear range of the detection circuits without saturation. The AGC Active Status bit AGCACT<5>, in the AFE Status Register 7 (Register 11-8) is set if the AGC loop regulates the input voltage.

Table 11-2 shows the input sensitivity comparison when the AGCSIG option is used. When AGCSIG option bit is set, the demodulated output is available only when the AGC loop is active (see Table 11-1). The AFE has also input sensitivity reduction options per each channel. The Configuration Register 3 (Register 11-4), Configuration Register 4 (Register 11-5) and Configuration Register 5 (Register 11-6) have the option to reduce the channel gains from 0 dB to approximately -30 dB.

PIC12F635/PIC16F636/639

12.7 Time-out Sequence

On power-up, the time-out sequence is as follows: first, PWRT time-out is invoked after POR has expired, then OST is activated after the PWRT time-out has expired. The total time-out will vary based on oscillator Configuration and PWRTE bit status. For example, in EC mode with PWRTE bit erased (PWRT disabled), there will be no time-out at all. Figure 12-4, Figure 12-5 and Figure 12-6 depict time-out sequences. The device can execute code from the INTOSC, while OST is active, by enabling Two-Speed Start-up or Fail-Safe Clock Monitor (See Section 3.7.2 “Two-Speed Start-up Sequence” and Section 3.8 “Fail-Safe Clock Monitor”).

Since the time-outs occur from the POR pulse, if MCLR is kept low long enough, the time-outs will expire. Then bringing MCLR high will begin execution immediately (see Figure 12-5). This is useful for testing purposes or to synchronize more than one PIC12F635/PIC16F636/639 device operating in parallel.

Table 12-5 shows the Reset conditions for some special registers, while Table 12-4 shows the Reset conditions for all the registers.

12.8 Power Control (PCON) Register

The Power Control register, PCON (address 8Eh), has two Status bits to indicate what type of Reset that last occurred.

Bit 0 is $\overline{\text{BOR}}$ (Brown-out). $\overline{\text{BOR}}$ is unknown on Power-on Reset. It must then be set by the user and checked on subsequent Resets to see if $\overline{\text{BOR}} = 0$, indicating that a Brown-out has occurred. The $\overline{\text{BOR}}$ Status bit is a “don’t care” and is not necessarily predictable if the brown-out circuit is disabled ($\text{BOREN}\langle 1:0 \rangle = 00$ in the Configuration Word register).

Bit 1 is $\overline{\text{POR}}$ (Power-on Reset). It is a ‘0’ on Power-on Reset and unaffected otherwise. The user must write a ‘1’ to this bit following a Power-on Reset. On a subsequent Reset, if $\overline{\text{POR}}$ is ‘0’, it will indicate that a Power-on Reset has occurred (i.e., VDD may have gone too low).

For more information, see Section 4.2.3 “Ultra Low-Power Wake-up” and Section 12.6 “Brown-out Reset (BOR)”.

TABLE 12-1: TIME-OUT IN VARIOUS SITUATIONS

Oscillator Configuration	Power-up		Brown-out Reset		Wake-up from Sleep
	$\overline{\text{PWRTE}} = 0$	$\overline{\text{PWRTE}} = 1$	$\overline{\text{PWRTE}} = 0$	$\overline{\text{PWRTE}} = 1$	
XT, HS, LP	TPWRT + 1024 • TOSC	1024 • TOSC	TPWRT + 1024 • TOSC	1024 • TOSC	1024 • TOSC
RC, EC, INTOSC	TPWRT	—	TPWRT	—	—

TABLE 12-2: SUMMARY OF REGISTERS ASSOCIATED WITH BROWN-OUT RESET

Name	Bit 9	Bit 8	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	Value on all other Resets ⁽¹⁾
CONFIG ⁽²⁾	BOREN1	BOREN0	$\overline{\text{CPD}}$	$\overline{\text{CP}}$	MCLRE	$\overline{\text{PWRTE}}$	WDTE	FOSC2	FOSC1	FOSC0	—	—
PCON			—	—	ULPWUE	SBOREN	$\overline{\text{WUR}}$	—	$\overline{\text{POR}}$	$\overline{\text{BOR}}$	--01 --qg	--0u --uu
STATUS			IRP	RP1	RP0	$\overline{\text{TO}}$	$\overline{\text{PD}}$	Z	DC	C	0001 1xxx	000q quuu

Legend: u = unchanged, x = unknown, — = unimplemented bit, reads as ‘0’, q = value depends on condition. Shaded cells are not used by BOR.
Note 1: Other (non Power-up) Resets include MCLR Reset and Watchdog Timer Reset during normal operation.
2: See Configuration Word register (Register 12-1) for operation of all register bits.

TABLE 12-3: PCON BITS AND THEIR SIGNIFICANCE

$\overline{\text{POR}}$	$\overline{\text{BOR}}$	$\overline{\text{WUR}}$	$\overline{\text{TO}}$	$\overline{\text{PD}}$	Condition
0	x	x	1	1	Power-on Reset
u	0	u	1	1	Brown-out Reset
u	u	u	0	u	WDT Reset
u	u	u	0	0	WDT Wake-up
u	u	u	u	u	$\overline{\text{MCLR}}$ Reset during normal operation
u	u	u	1	0	$\overline{\text{MCLR}}$ Reset during Sleep
u	u	0	1	0	Wake-up Reset during Sleep
u	0	u	1	1	Brown-out Reset during Sleep

Legend: u = unchanged, x = unknown

PIC12F635/PIC16F636/639

12.9.2 TIMER INTERRUPT

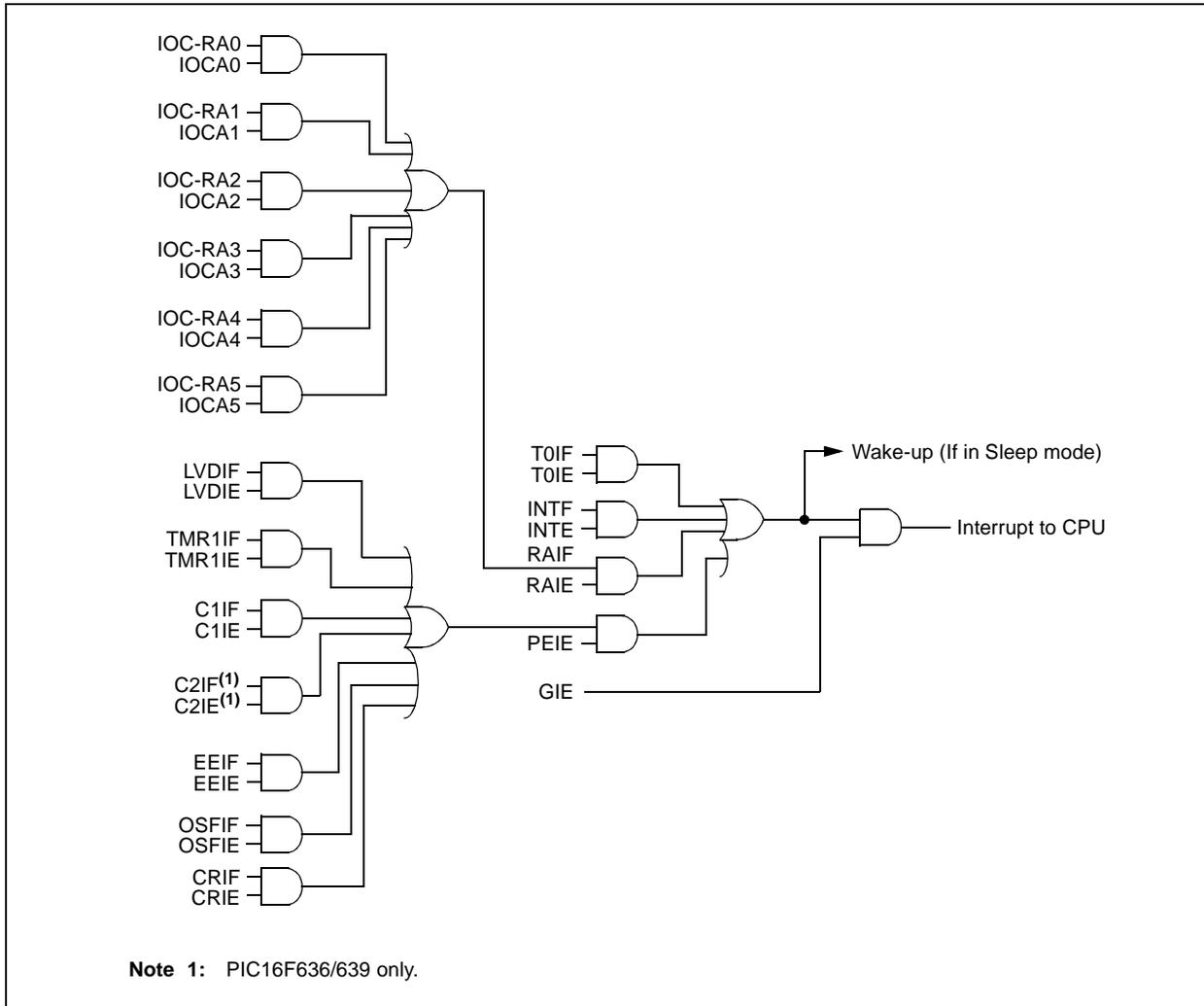
An overflow (FFh → 00h) in the TMR0 register will set the TOIF bit of the INTCON register. The interrupt can be enabled/disabled by setting/clearing TOIE bit of the INTCON register. See **Section 5.0 “Timer0 Module”** for operation of the Timer0 module.

12.9.3 PORTA INTERRUPT

An input change on PORTA change sets the RAIF bit of the INTCON register. The interrupt can be enabled/disabled by setting/clearing the RAIE bit of the INTCON register. Plus, individual pins can be configured through the IOCA register.

Note: If a change on the I/O pin should occur when the read operation is being executed (start of the Q2 cycle), then the RAIF interrupt flag may not get set.

FIGURE 12-7: INTERRUPT LOGIC



PIC12F635/PIC16F636/639

12.15 In-Circuit Serial Programming

The PIC12F635/PIC16F636/639 microcontrollers can be serially programmed while in the end application circuit. This is simply done with two lines for clock and data and three other lines for:

- Power
- Ground
- Programming Voltage

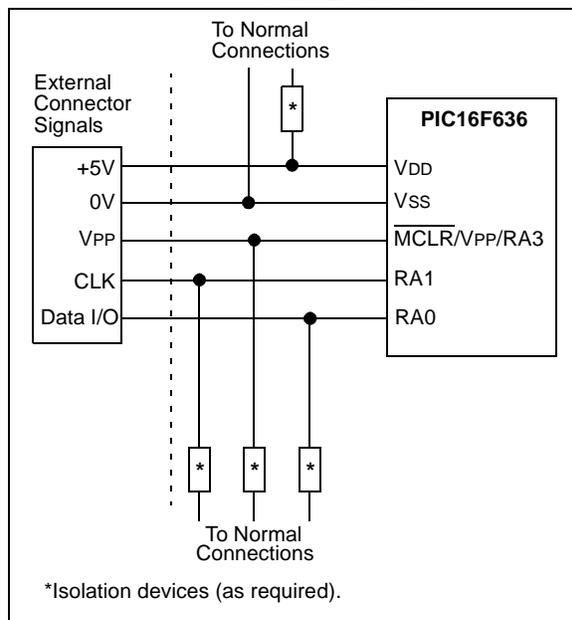
This allows customers to manufacture boards with unprogrammed devices and then program the microcontroller just before shipping the product. This also allows the most recent firmware or a custom firmware to be programmed.

The device is placed into a Program/Verify mode by holding the RA0 and RA1 pins low, while raising the MCLR (VPP) pin from V_{IL} to V_{IH} . See the "PIC12F6XX/16F6XX Memory Programming Specification" (DS41204) for more information. RA0 becomes the programming data and RA1 becomes the programming clock. Both RA0 and RA1 are Schmitt Trigger inputs in this mode.

After Reset, to place the device into Program/Verify mode, the Program Counter (PC) is at location 00h. A 6-bit command is then supplied to the device. Depending on the command, 14 bits of program data are then supplied to or from the device, depending on whether the command was a load or a read. For complete details of serial programming, please refer to the "PIC12F6XX/16F6XX Memory Programming Specification" (DS41204).

A typical In-Circuit Serial Programming connection is shown in Figure 12-11.

FIGURE 12-11: TYPICAL IN-CIRCUIT SERIAL PROGRAMMING CONNECTION



12.16 In-Circuit Debugger

Since in-circuit debugging requires the loss of clock, data and MCLR pins, MPLAB® ICD 2 development with a 14-pin device is not practical. A special 20-pin PIC16F636 ICD device is used with MPLAB ICD 2 to provide separate clock, data and MCLR pins and frees all normally available pins to the user.

Use of the ICD device requires the purchase of a special header. On the top of the header is an MPLAB ICD 2 connector. On the bottom of the header is a 14-pin socket that plugs into the user's target via the 14-pin stand-off connector.

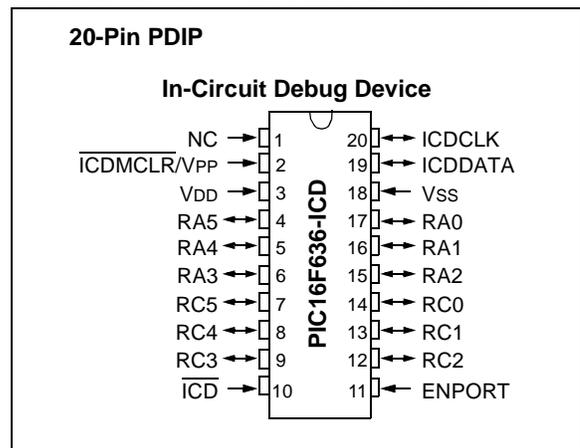
When the \overline{ICD} pin on the PIC16F636 ICD device is held low, the In-Circuit Debugger functionality is enabled. This function allows simple debugging functions when used with MPLAB ICD 2. When the microcontroller has this feature enabled, some of the resources are not available for general use. Table 12-9 shows which features are consumed by the background debugger:

TABLE 12-9: DEBUGGER RESOURCES

Resource	Description
I/O pins	ICDCLK, ICDDATA
Stack	1 level
Program Memory	Address 0h must be NOP 700h-7FFh

For more information, see the "MPLAB® ICD 2 In-Circuit Debugger User's Guide" (DS51331), available on Microchip's web site (www.microchip.com).

FIGURE 12-12: 20-PIN ICD PINOUT



PIC12F635/PIC16F636/639

NOTES:

PIC12F635/PIC16F636/639

BTFSF **Bit Test f, Skip if Set**

Syntax: [*label*] BTFSF f,b
Operands: $0 \leq f \leq 127$
 $0 \leq b < 7$
Operation: skip if (f) = 1
Status Affected: None
Description: If bit 'b' in register 'f' is '0', the next instruction is executed.
 If bit 'b' is '1', then the next instruction is discarded and a NOP is executed instead, making this a two-cycle instruction.

CLRWDT **Clear Watchdog Timer**

Syntax: [*label*] CLRWDT
Operands: None
Operation: 00h → WDT
 0 → WDT prescaler,
 1 → \overline{TO}
 1 → \overline{PD}
Status Affected: \overline{TO} , \overline{PD}
Description: CLRWDT instruction resets the Watchdog Timer. It also resets the prescaler of the WDT.
 Status bits \overline{TO} and \overline{PD} are set.

CALL **Call Subroutine**

Syntax: [*label*] CALL k
Operands: $0 \leq k \leq 2047$
Operation: (PC)+1 → TOS,
 k → PC<10:0>,
 (PCLATH<4:3>) → PC<12:11>
Status Affected: None
Description: Call Subroutine. First, return address (PC + 1) is pushed onto the stack. The eleven-bit immediate address is loaded into PC bits <10:0>. The upper bits of the PC are loaded from PCLATH. CALL is a two-cycle instruction.

COMF **Complement f**

Syntax: [*label*] COMF f,d
Operands: $0 \leq f \leq 127$
 $d \in [0,1]$
Operation: (\bar{f}) → (destination)
Status Affected: Z
Description: The contents of register 'f' are complemented. If 'd' is '0', the result is stored in W. If 'd' is '1', the result is stored back in register 'f'.

CLRF **Clear f**

Syntax: [*label*] CLRF f
Operands: $0 \leq f \leq 127$
Operation: 00h → (f)
 1 → Z
Status Affected: Z
Description: The contents of register 'f' are cleared and the Z bit is set.

DECF **Decrement f**

Syntax: [*label*] DECF f,d
Operands: $0 \leq f \leq 127$
 $d \in [0,1]$
Operation: (f) - 1 → (destination)
Status Affected: Z
Description: Decrement register 'f'. If 'd' is '0', the result is stored in the W register. If 'd' is '1', the result is stored back in register 'f'.

CLRW **Clear W**

Syntax: [*label*] CLRW
Operands: None
Operation: 00h → (W)
 1 → Z
Status Affected: Z
Description: W register is cleared. Zero bit (Z) is set.

PIC12F635/PIC16F636/639

15.2 DC Characteristics: PIC12F635/PIC16F636-I (Industrial)

DC CHARACTERISTICS			Standard Operating Conditions (unless otherwise stated) Operating temperature $-40^{\circ}\text{C} \leq T_A \leq +85^{\circ}\text{C}$ for industrial					
Param No.	Sym	Device Characteristics	Min	Typ†	Max	Units	Conditions	
							V _{DD}	Note
D010	I _{DD}	Supply Current ^(1,2)	—	11	16	μA	2.0	Fosc = 32.768 kHz LP Oscillator mode
			—	18	28	μA	3.0	
			—	35	54	μA	5.0	
D011			—	140	240	μA	2.0	Fosc = 1 MHz XT Oscillator mode
			—	220	380	μA	3.0	
			—	380	550	μA	5.0	
D012			—	260	360	μA	2.0	Fosc = 4 MHz XT Oscillator mode
			—	420	650	μA	3.0	
			—	0.8	1.1	mA	5.0	
D013			—	130	220	μA	2.0	Fosc = 1 MHz EC Oscillator mode
			—	215	360	μA	3.0	
			—	360	520	μA	5.0	
D014			—	220	340	μA	2.0	Fosc = 4 MHz EC Oscillator mode
			—	375	550	μA	3.0	
			—	0.65	1.0	mA	5.0	
D015			—	8	20	μA	2.0	Fosc = 31 kHz LFINTOSC mode
			—	16	40	μA	3.0	
			—	31	65	μA	5.0	
D016			—	340	450	μA	2.0	Fosc = 4 MHz HFINTOSC mode
			—	500	700	μA	3.0	
			—	0.8	1.2	mA	5.0	
D017			—	410	650	μA	2.0	Fosc = 8 MHz HFINTOSC mode
			—	700	950	μA	3.0	
			—	1.30	1.65	mA	5.0	
D018			—	230	400	μA	2.0	Fosc = 4 MHz EXTRC mode
			—	400	680	μA	3.0	
			—	0.63	1.1	mA	5.0	
D019			—	2.6	3.25	mA	4.5	Fosc = 20 MHz HS Oscillator mode
			—	2.6	3.25	mA	5.0	

† Data in "Typ" column is at 5.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: The test conditions for all I_{DD} measurements in Active Operation mode are: OSC1 = external square wave, from rail-to-rail; all I/O pins tri-stated, pulled to V_{DD}; MCLR = V_{DD}; WDT disabled. MCU only, Analog Front-End not included.

2: The supply current is mainly a function of the operating voltage and frequency. Other factors, such as I/O pin loading and switching rate, oscillator type, internal code execution pattern and temperature, also have an impact on the current consumption. MCU only, Analog Front-End not included.

3: The peripheral current is the sum of the base I_{DD} or I_{PD} and the additional current consumed when this peripheral is enabled. The peripheral Δ current can be determined by subtracting the base I_{DD} or I_{PD} current from this limit. Max values should be used when calculating total current consumption.

4: The power-down current in Sleep mode does not depend on the oscillator type. Power-down current is measured with the part in Sleep mode, with all I/O pins in high-impedance state and tied to V_{DD}.

PIC12F635/PIC16F636/639

15.5 DC Characteristics: PIC16F639-I (Industrial)

DC CHARACTERISTICS			Standard Operating Conditions (unless otherwise stated) Operating temperature $-40^{\circ}\text{C} \leq T_A \leq +85^{\circ}\text{C}$ for industrial				
Param No.	Sym	Characteristic	Min	Typ†	Max	Units	Conditions
D001	VDD	Supply Voltage	2.0	—	3.6	V	FOSC \leq 10 MHz
D001A	VDDT	Supply Voltage (AFE)	2.0	—	3.6	V	Analog Front-End VDD voltage. Treated as VDD in this document.
D002	VDR	RAM Data Retention Voltage⁽¹⁾	1.5*	—	—	V	Device in Sleep mode
D003	VPOR	VDD Start Voltage to ensure internal Power-on Reset signal	—	VSS	—	V	See Section 12.3 “Power-on Reset” for details.
D003A	VPORT	VDD Start Voltage (AFE) to ensure internal Power-on Reset signal	—	—	1.8	V	Analog Front-End POR voltage.
D004	SVDD	VDD Rise Rate to ensure internal Power-on Reset signal	0.05*	—	—	V/ms	See Section 12.3 “Power-on Reset” for details.
D005	VBOD	Brown-out Reset	2.0	2.1	2.2	V	
D006	RM	Turn-on Resistance or Modulation Transistor	—	50	100	Ohm	VDD = 3.0V
D007	RPU	Digital Input Pull-Up Resistor CS, SCLK	50	200	350	kOhm	VDD = 3.6V
D008	IAIL	Analog Input Leakage Current LCX, LCY, LCZ LCCOM	— —	— —	± 1 ± 1	μA μA	VDD = 3.6V, VSS \leq VIN \leq VDD, tested at Sleep mode

* These parameters are characterized but not tested.

† Data in “Typ” column is at 3.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: This is the limit to which VDD can be lowered in Sleep mode without losing RAM data.

PIC12F635/PIC16F636/639

FIGURE 15-7: RESET, WATCHDOG TIMER, OSCILLATOR START-UP TIMER AND POWER-UP TIMER TIMING

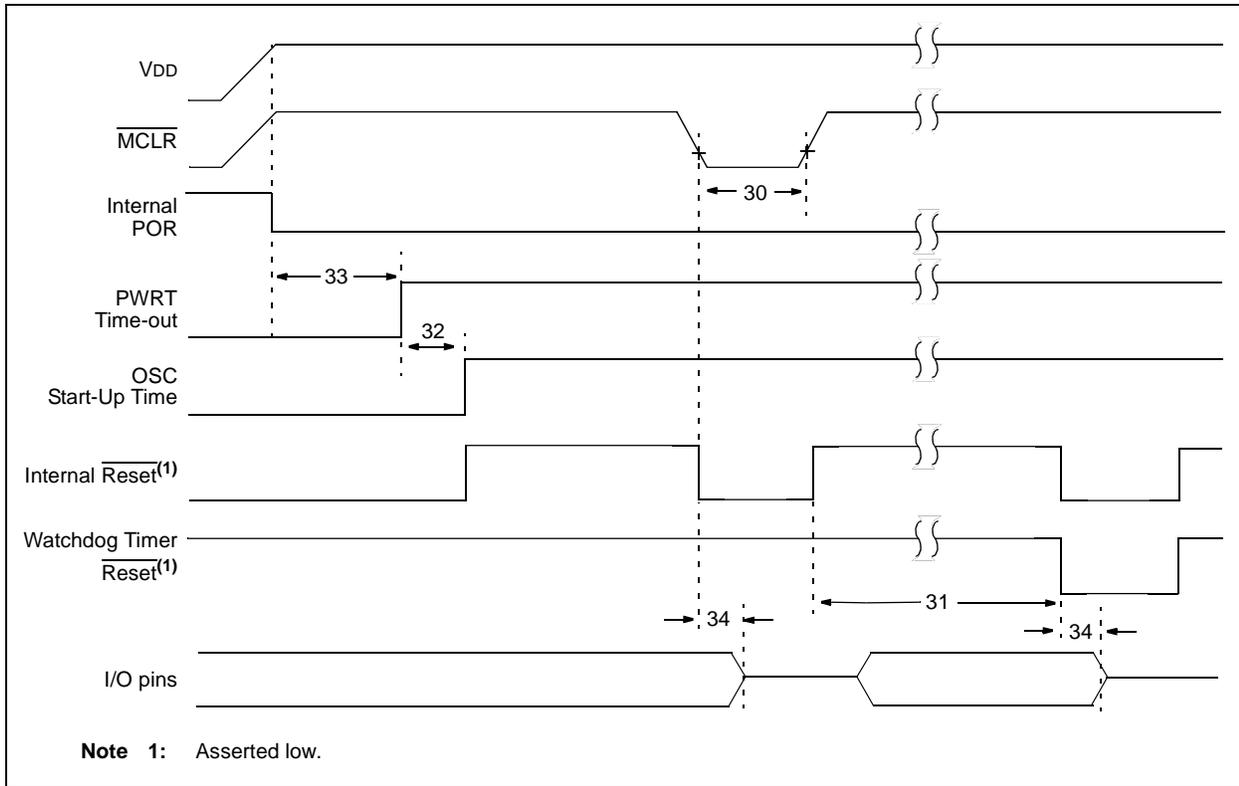
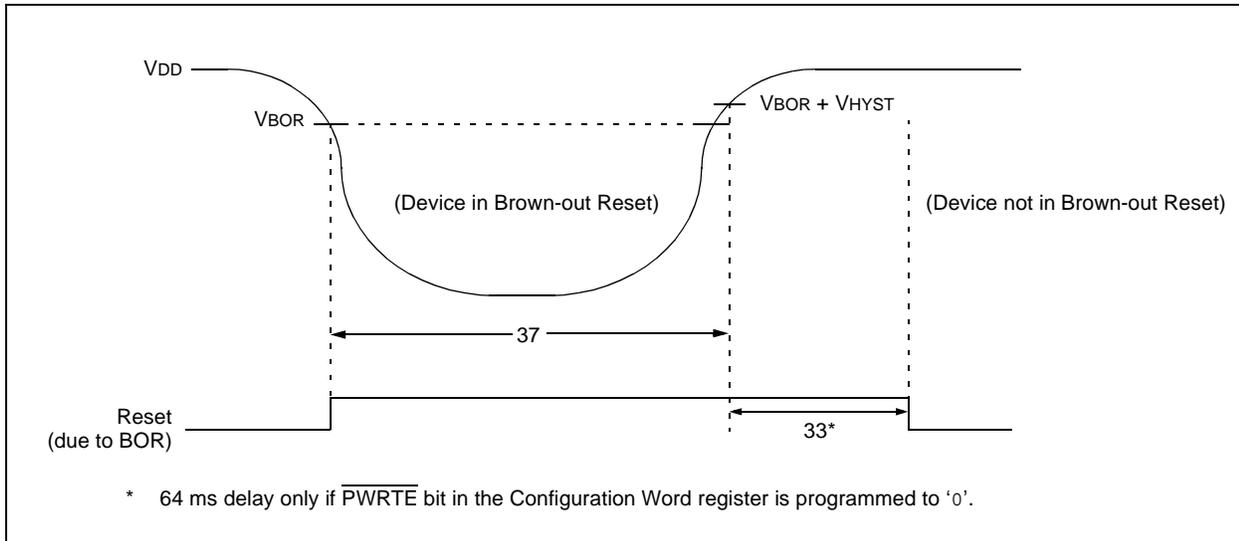


FIGURE 15-8: BROWN-OUT RESET TIMING AND CHARACTERISTICS



PIC12F635/PIC16F636/639

15.12 SPI Timing: Analog Front-End (AFE) for PIC16F639

AC CHARACTERISTICS			Standard Operating Conditions (unless otherwise stated)				
			Supply Voltage $2.0V \leq V_{DD} \leq 3.6V$				
			Operating temperature $-40^{\circ}C \leq T_{AMB} \leq +85^{\circ}C$ for industrial				
			LC Signal Input Sinusoidal 300 mVPP				
			Carrier Frequency 125 kHz				
			LCCOM connected to VSS				
Param	Sym	Characteristic	Min	Typ†	Max	Units	Conditions
AF33	F _{SCLK}	SCLK Frequency	—	—	3	MHz	
AF34	T _{cSSC}	\overline{CS} fall to first SCLK edge setup time	100	—	—	ns	
AF35	T _{su}	SDI setup time	30	—	—	ns	
AF36	T _{HD}	SDI hold time	50	—	—	ns	
AF37	T _{HI}	SCLK high time	150	—	—	ns	
AF38	T _{LO}	SCLK low time	150	—	—	ns	
AF39	T _{DO}	SDO setup time	—	—	150	ns	
AF40	T _{sCCS}	SCLK last edge to \overline{CS} rise setup time	100	—	—	ns	
AF41	T _{CSH}	\overline{CS} high time	500	—	—	ns	
AF42	T _{CS1}	\overline{CS} rise to SCLK edge setup time	50	—	—	ns	
AF43	T _{CS0}	SCLK edge to \overline{CS} fall setup time	50	—	—	ns	SCLK edge when \overline{CS} is high
AF44	T _{SPIR}	Rise time of SPI data (SPI Read command)	—	10	—	ns	V _{DD} = 3.0V. Time is measured from 10% to 90% of amplitude
AF45	T _{SPIF}	Fall time of SPI data (SPI Read command)	—	10	—	ns	V _{DD} = 3.0V. Time is measured from 90% to 10% of amplitude

* Parameter is characterized but not tested.

† Data in "Typ" column is at 3.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

PIC12F635/PIC16F636/639

FIGURE 16-16: BOR IPD vs. VDD OVER TEMPERATURE

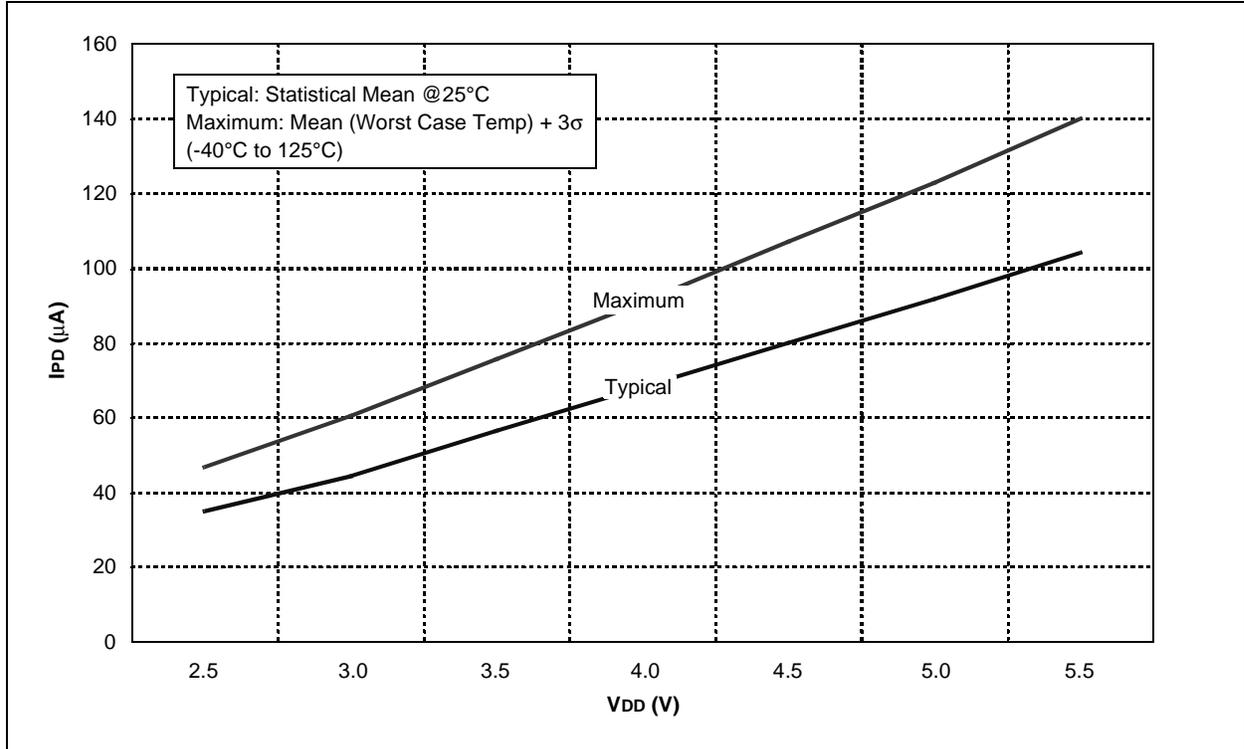
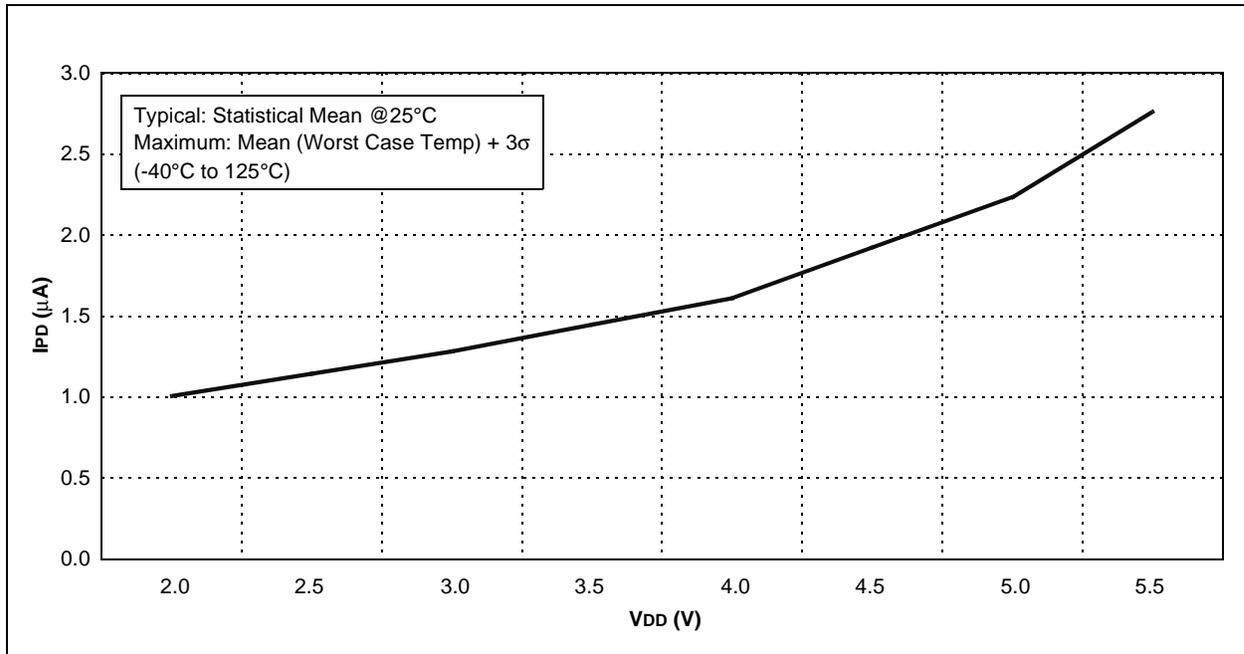


FIGURE 16-17: TYPICAL WDT IPD vs. VDD OVER TEMPERATURE



PIC12F635/PIC16F636/639

FIGURE 16-18: MAXIMUM WDT IPD vs. VDD OVER TEMPERATURE

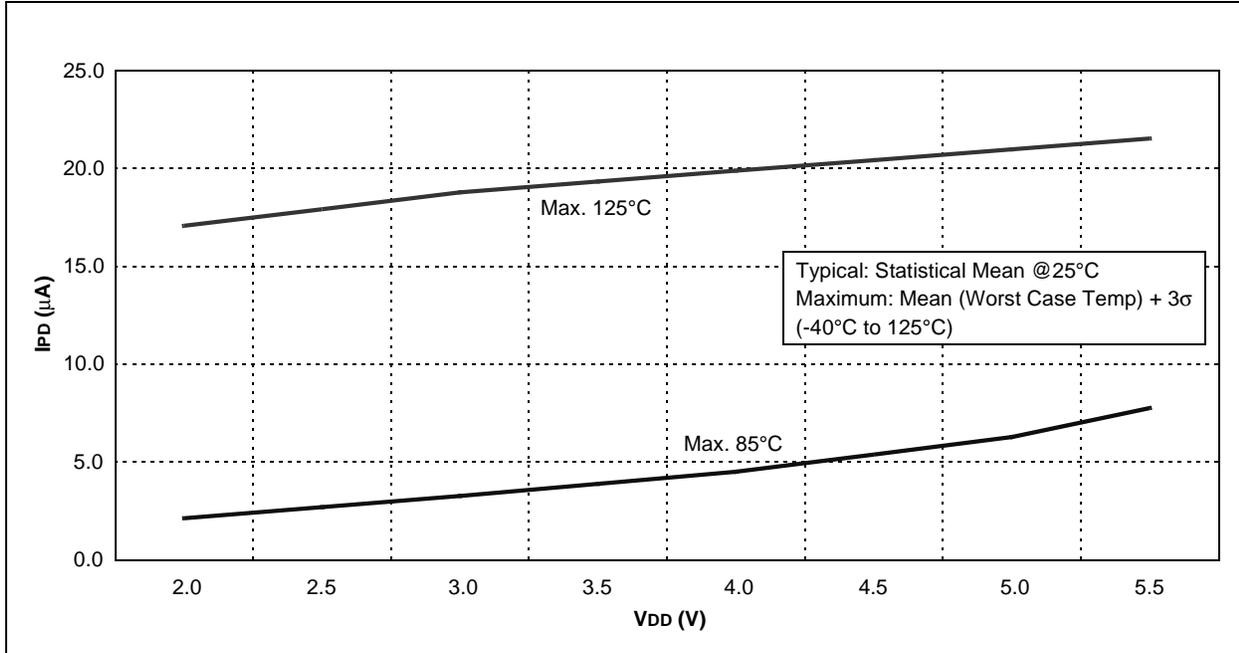
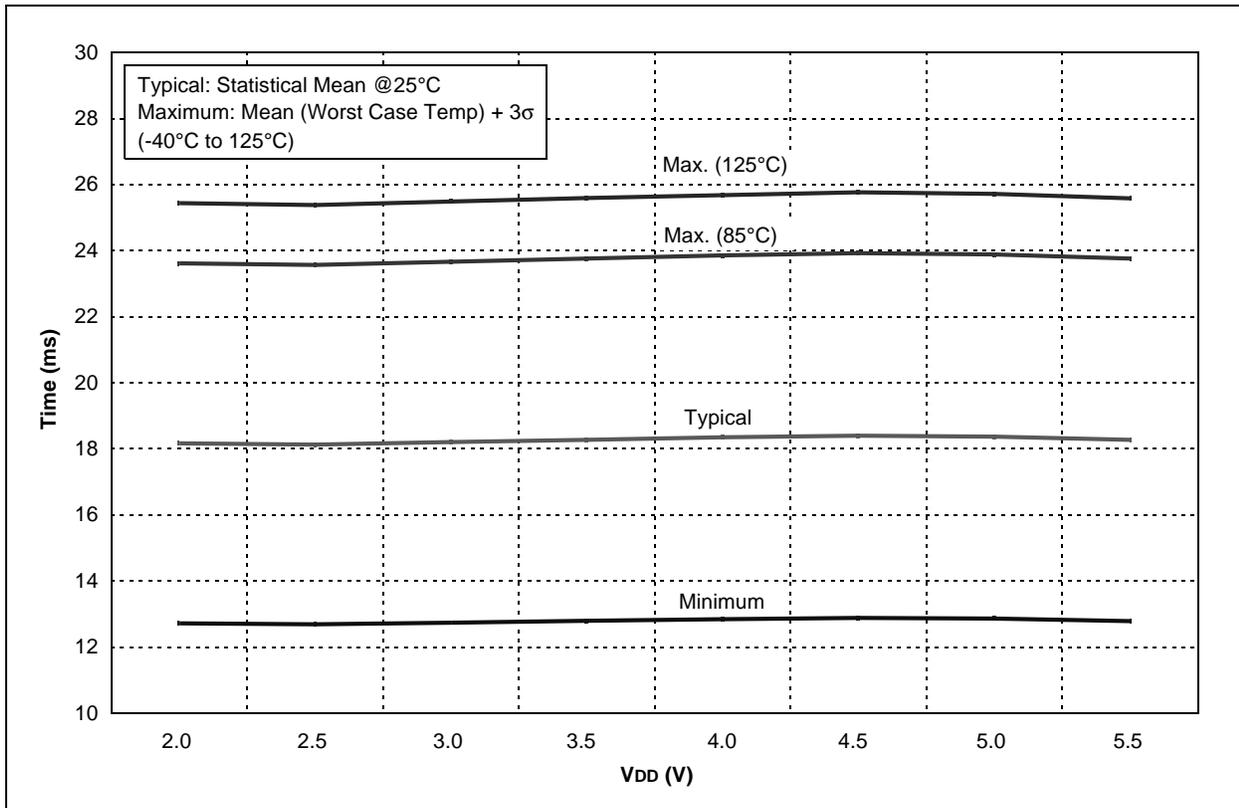


FIGURE 16-19: WDT PERIOD vs. VDD OVER TEMPERATURE



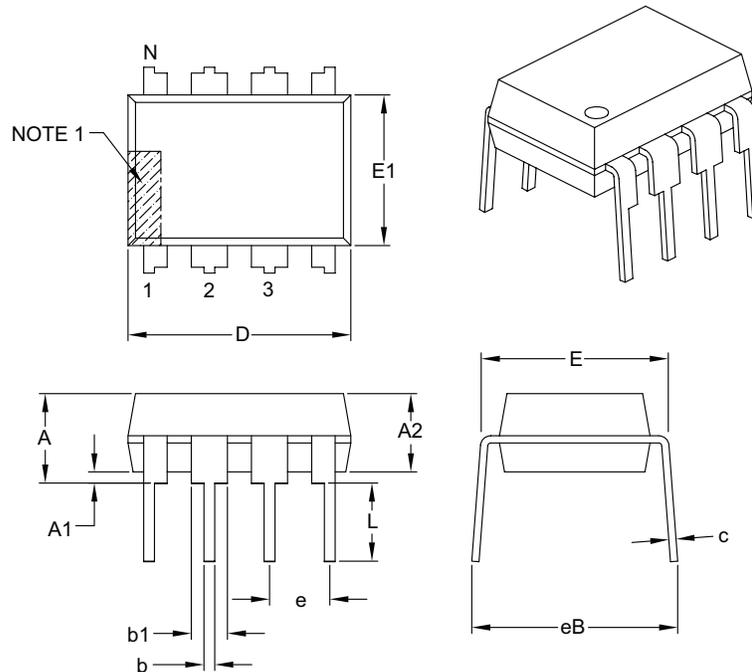
PIC12F635/PIC16F636/639

17.2 Package Details

The following sections give the technical details of the packages.

8-Lead Plastic Dual In-Line (P or PA) – 300 mil Body [PDIP]

Note: For the most current package drawings, please see the Microchip Packaging Specification located at <http://www.microchip.com/packaging>



Dimension Limits	Units	INCHES		
		MIN	NOM	MAX
Number of Pins	N	8		
Pitch	e	.100 BSC		
Top to Seating Plane	A	–	–	.210
Molded Package Thickness	A2	.115	.130	.195
Base to Seating Plane	A1	.015	–	–
Shoulder to Shoulder Width	E	.290	.310	.325
Molded Package Width	E1	.240	.250	.280
Overall Length	D	.348	.365	.400
Tip to Seating Plane	L	.115	.130	.150
Lead Thickness	c	.008	.010	.015
Upper Lead Width	b1	.040	.060	.070
Lower Lead Width	b	.014	.018	.022
Overall Row Spacing §	eB	–	–	.430

Notes:

- Pin 1 visual index feature may vary, but must be located with the hatched area.
- § Significant Characteristic.
- Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed .010" per side.
- Dimensioning and tolerancing per ASME Y14.5M.

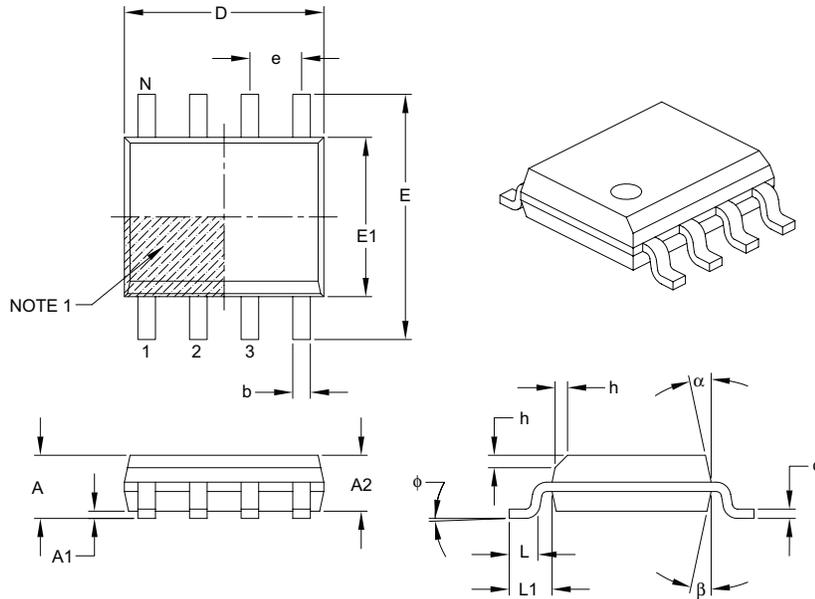
BSC: Basic Dimension. Theoretically exact value shown without tolerances.

Microchip Technology Drawing C04-018B

PIC12F635/PIC16F636/639

8-Lead Plastic Small Outline (SN or OA) – Narrow, 3.90 mm Body [SOIC]

Note: For the most current package drawings, please see the Microchip Packaging Specification located at <http://www.microchip.com/packaging>



Dimension Limits	Units	MILLIMETERS		
		MIN	NOM	MAX
Number of Pins	N	8		
Pitch	e	1.27 BSC		
Overall Height	A	–	–	1.75
Molded Package Thickness	A2	1.25	–	–
Standoff §	A1	0.10	–	0.25
Overall Width	E	6.00 BSC		
Molded Package Width	E1	3.90 BSC		
Overall Length	D	4.90 BSC		
Chamfer (optional)	h	0.25	–	0.50
Foot Length	L	0.40	–	1.27
Footprint	L1	1.04 REF		
Foot Angle	ϕ	0°	–	8°
Lead Thickness	c	0.17	–	0.25
Lead Width	b	0.31	–	0.51
Mold Draft Angle Top	α	5°	–	15°
Mold Draft Angle Bottom	β	5°	–	15°

Notes:

- Pin 1 visual index feature may vary, but must be located within the hatched area.
- § Significant Characteristic.
- Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed 0.15 mm per side.
- Dimensioning and tolerancing per ASME Y14.5M.

BSC: Basic Dimension. Theoretically exact value shown without tolerances.

REF: Reference Dimension, usually without tolerance, for information purposes only.

Microchip Technology Drawing C04-057B