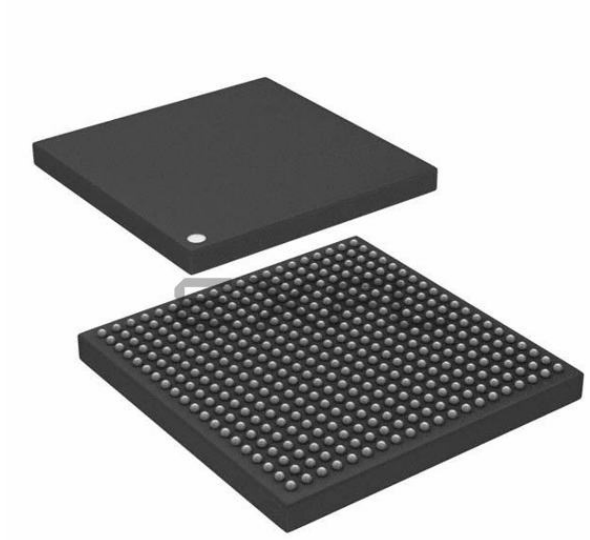

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Understanding Embedded - DSP (Digital Signal Processors)

Embedded - DSP (Digital Signal Processors) are specialized microprocessors designed to perform complex mathematical computations on digital signals in real-time. Unlike general-purpose processors, DSPs are optimized for high-speed numeric processing tasks, making them ideal for applications that require efficient and precise manipulation of digital data. These processors are fundamental in converting and processing signals in various forms, including audio, video, and communication signals, ensuring that data is accurately interpreted and utilized in embedded systems.

Applications of Embedded - DSP (Digital Signal Processors)

Details	
Product Status	Obsolete
Type	Fixed Point
Interface	Host Interface, I²C, UART
Clock Rate	300MHz
Non-Volatile Memory	ROM (8kB)
On-Chip RAM	464kB
Voltage - I/O	3.30V
Voltage - Core	1.20V
Operating Temperature	-40°C ~ 105°C (TJ)
Mounting Type	Surface Mount
Package / Case	400-LFBGA
Supplier Device Package	400-LFBGA (17x17)
Purchase URL	https://www.e-xfl.com/product-detail/nxp-semiconductors/kmsc7119vm1200

Table of Contents

List of Figures

1 Pin Assignments

This section includes diagrams of the MSC7119 package ball grid array layouts and pinout allocation tables.

1.1 MAP-BGA Ball Layout Diagrams

Top and bottom views of the MAP-BGA package are shown in **Figure 2** and **Figure 3** with their ball location index numbers.

Top View

	1	2	3	4	5	6	7	8	9	10	11	12	13	14	15	16	17	18	19	20
A	GND	GND	DQM1	DQS2	CK	$\overline{CK}$	HD15	HD12	HD10	HD7	HD6	HD4	HD1	HD0	GND	BM3	NC	NC	NC	NC
B	V_{DDM}	NC	$\overline{CS0}$	DQM2	DQS3	DQS0	CKE	$\overline{WE}$	HD14	HD11	HD8	HD5	HD2	NC	BM2	NC	NC	NC	NC	NC
C	D24	D30	D25	$\overline{CS1}$	DQM3	DQM0	DQS1	$\overline{RAS}$	$\overline{CAS}$	HD13	HD9	HD3	NC	NC	NC	NC	NC	NC	NC	NC
D	V_{DDM}	D28	D27	GND	V_{DDM}	V_{DDM}	V_{DDM}	V_{DDM}	V_{DDM}	V_{DDM}	V_{DDIO}	V_{DDIO}	V_{DDIO}	V_{DDIO}	V_{DDIO}	V_{DDIO}	V_{DDC}	NC	NC	NC
E	GND	D26	D31	V_{DDM}	V_{DDM}	V_{DDC}	V_{DDC}	V_{DDC}	V_{DDC}	V_{DDM}	V_{DDIO}	V_{DDIO}	V_{DDIO}	V_{DDIO}	V_{DDIO}	V_{DDC}	V_{DDC}	NC	NC	NC
F	V_{DDM}	D15	D29	V_{DDC}	V_{DDC}	V_{DDC}	GND	GND	GND	V_{DDM}	V_{DDM}	GND	GND	GND	V_{DDIO}	V_{DDC}	V_{DDC}	NC	NC	NC
G	GND	D13	GND	V_{DDM}	V_{DDM}	GND	GND	GND	GND	GND	GND	GND	GND	GND	V_{DDIO}	V_{DDIO}	V_{DDC}	NC	NC	NC
H	D14	D12	D11	V_{DDM}	V_{DDM}	GND	GND	GND	GND	GND	GND	GND	GND	GND	V_{DDIO}	V_{DDIO}	V_{DDC}	NC	HA2	HA1
J	D10	V_{DDM}	D9	V_{DDM}	V_{DDM}	V_{DDM}	GND	GND	GND	GND	GND	GND	GND	GND	GND	V_{DDIO}	V_{DDC}	HA3	$\overline{HACK}$	$\overline{HREQ}$
K	D0	GND	D8	V_{DDC}	V_{DDM}	GND	GND	GND	GND	GND	GND	GND	GND	GND	GND	V_{DDIO}	V_{DDIO}	V_{DDC}	HA0	HDDS	$\overline{HDS}$
L	D1	GND	D3	V_{DDC}	V_{DDM}	GND	GND	GND	GND	GND	GND	GND	GND	V_{DDIO}	V_{DDIO}	V_{DDIO}	V_{DDC}	$\overline{HCS2}$	$\overline{HCS1}$	HRW
M	D2	V_{DDM}	D5	V_{DDM}	V_{DDM}	GND	GND	GND	GND	GND	GND	GND	GND	GND	GND	V_{DDC}	V_{DDC}	SDA	UTXD	URXD
N	D4	D6	V_{REF}	V_{DDM}	V_{DDM}	V_{DDM}	GND	GND	GND	GND	GND	GND	GND	GND	V_{DDIO}	V_{DDC}	V_{DDC}	CLKIN	SCL	V_{SSPLL}
P	D7	D17	D16	V_{DDM}	V_{DDM}	V_{DDM}	GND	GND	GND	GND	GND	GND	GND	GND	V_{DDIO}	V_{DDIO}	V_{DDC}	$\overline{PORESET}$	TPSEL	V_{DDPLL}
R	GND	D19	D18	V_{DDM}	V_{DDM}	V_{DDM}	GND	V_{DDM}	GND	V_{DDM}	GND	GND	V_{DDIO}	GND	V_{DDIO}	V_{DDIO}	V_{DDC}	TDO	EE0	TEST0
T	V_{DDM}	D20	D22	V_{DDM}	V_{DDM}	V_{DDC}	V_{DDM}	V_{DDM}	V_{DDC}	V_{DDM}	V_{DDM}	V_{DDIO}	V_{DDIO}	V_{DDIO}	V_{DDIO}	V_{DDC}	V_{DDC}	MDIO	TMS	$\overline{HRESET}$
U	GND	D21	D23	V_{DDM}	V_{DDC}	V_{DDC}	V_{DDC}	V_{DDC}	V_{DDC}	V_{DDC}	V_{DDC}	V_{DDC}	V_{DDC}	V_{DDC}	V_{DDC}	V_{DDC}	V_{DDC}	COL	TCK	$\overline{TRST}$
V	V_{DDM}	NC	A13	A11	A10	A5	A2	BA0	NC	EVNT0	EVNT4	T0TCK	T1RFS	T1TD	TX_ER	RXD2	RXD0	TX_EN	CRS	TDI
W	GND	V_{DDM}	A12	A8	A7	A6	A3	NC	EVNT1	EVNT2	T0RFS	T0TFS	T1RD	T1TFS	TXD2	RXD3	TXD1	TXCLK	RX_ER	MDC
Y	V_{DDM}	GND	A9	A1	A0	A4	BA1	$\overline{NMI}$	EVNT3	T0RCK	T0RD	TOTD	T1RCK	T1TCK	TXD3	RXCLK	TXD0	RXD1	GND	RX_DV

Figure 2. MSC7119 Molded Array Process-Ball Grid Array (MAP-BGA), Top View

Bottom View

	20	19	18	17	16	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1
A	NC	NC	NC	NC	BM3	GND	HD0	HD1	HD4	HD6	HD7	HD10	HD12	HD15	$\overline{CK}$	CK	DQS2	DQM1	GND	GND
B	NC	NC	NC	NC	NC	BM2	NC	HD2	HD5	HD8	HD11	HD14	$\overline{WE}$	CKE	DQS0	DQS3	DQM2	$\overline{CS0}$	NC	V_{DDM}
C	NC	NC	NC	NC	NC	NC	NC	NC	HD3	HD9	HD13	$\overline{CAS}$	$\overline{RAS}$	DQS1	DQM0	DQM3	$\overline{CS1}$	D25	D30	D24
D	NC	NC	NC	V_{DD}	V_{DDIO}	V_{DDIO}	V_{DDIO}	V_{DDIO}	V_{DDIO}	V_{DDIO}	V_{DDM}	V_{DDM}	V_{DDM}	V_{DDM}	V_{DDM}	V_{DDM}	GND	D27	D28	V_{DDM}
E	NC	NC	NC	V_{DD}	V_{DD}	V_{DDIO}	V_{DDIO}	V_{DDIO}	V_{DDIO}	V_{DDIO}	V_{DDM}	V_{DD}	V_{DD}	V_{DD}	V_{DD}	V_{DDM}	V_{DDM}	D31	D26	GND
F	NC	NC	NC	V_{DD}	V_{DD}	V_{DDIO}	GND	GND	GND	V_{DDM}	V_{DDM}	GND	GND	GND	V_{DD}	V_{DD}	V_{DD}	D29	D15	V_{DDM}
G	NC	NC	NC	V_{DD}	V_{DDIO}	V_{DDIO}	GND	GND	GND	GND	GND	GND	GND	GND	GND	V_{DDM}	V_{DDM}	GND	D13	GND
H	HA1	HA2	NC	V_{DD}	V_{DDIO}	V_{DDIO}	GND	GND	GND	GND	GND	GND	GND	GND	GND	V_{DDM}	V_{DDM}	D11	D12	D14
J	$\overline{HREQ}$	$\overline{HACK}$	HA3	V_{DD}	V_{DDIO}	GND	GND	GND	GND	GND	GND	GND	GND	GND	V_{DDM}	V_{DDM}	V_{DDM}	D9	V_{DDM}	D10
K	$\overline{HDS}$	HDDS	HA0	V_{DD}	V_{DDIO}	V_{DDIO}	GND	GND	GND	GND	GND	GND	GND	GND	GND	V_{DDM}	V_{DD}	D8	GND	D0
L	HRW	$\overline{HCS1}$	$\overline{HCS2}$	V_{DD}	V_{DDIO}	V_{DDIO}	V_{DDIO}	GND	GND	GND	GND	GND	GND	GND	GND	V_{DDM}	V_{DD}	D3	GND	D1
M	URXD	UTXD	SDA	V_{DD}	V_{DD}	GND	GND	GND	GND	GND	GND	GND	GND	GND	GND	V_{DDM}	V_{DDM}	D5	V_{DDM}	D2
N	V_{SSPLL}	SCL	CLKIN	V_{DD}	V_{DD}	V_{DDIO}	GND	GND	GND	GND	GND	GND	GND	GND	V_{DDM}	V_{DDM}	V_{DDM}	V_{REF}	D6	D4
P	V_{DDPLL}	TPSEL	$\overline{PORESET}$	V_{DD}	V_{DDIO}	V_{DDIO}	GND	GND	GND	GND	GND	GND	GND	GND	V_{DDM}	V_{DDM}	V_{DDM}	D16	D17	D7
R	TEST0	EE0	TDO	V_{DD}	V_{DDIO}	V_{DDIO}	GND	V_{DDIO}	GND	GND	V_{DDM}	GND	V_{DDM}	GND	V_{DDM}	V_{DDM}	V_{DDM}	D18	D19	GND
T	$\overline{HRESET}$	TMS	MDIO	V_{DD}	V_{DD}	V_{DDIO}	V_{DDIO}	V_{DDIO}	V_{DDIO}	V_{DDM}	V_{DDM}	V_{DD}	V_{DDM}	V_{DDM}	V_{DD}	V_{DDM}	V_{DDM}	D22	D20	V_{DDM}
U	$\overline{TRST}$	TCK	COL	V_{DD}	V_{DD}	V_{DD}	V_{DD}	V_{DD}	V_{DD}	V_{DD}	V_{DD}	V_{DD}	V_{DD}	V_{DD}	V_{DD}	V_{DD}	V_{DDM}	D23	D21	GND
V	TDI	CRS	TX_EN	RXD0	RXD2	TX_ER	T1TD	T1RFS	T0TCK	EVNT4	EVNT0	NC	BA0	A2	A5	A10	A11	A13	NC	V_{DDM}
W	MDC	RX_ER	TXCLK	TXD1	RXD3	TXD2	T1TFS	T1RD	T0TFS	T0RFS	EVNT2	EVNT1	NC	A3	A6	A7	A8	A12	V_{DDM}	GND
Y	RX_DV	GND	RXD1	TXD0	RXCLK	TXD3	T1TCK	T1RCK	TOTD	T0RD	T0RCK	EVNT3	$\overline{NMI}$	BA1	A4	A0	A1	A9	GND	V_{DDM}

Figure 3. MSC7119 Molded Array Process-Ball Grid Array (MAP-BGA), Bottom View

Table 1. MSC7119 Signals by Ball Designator (continued)

Number	Signal Names					
	End of Reset	Software Controlled			Hardware Controlled	
		GPI Enabled (Default)	Interrupt Enabled	GPO Enabled	Primary	Alternate
B14	NC					
B15	BM2	GPID7		GPOD7	reserved	
B16	NC					
B17	NC					
B18	NC					
B19	NC					
B20	NC					
C1	D24					
C2	D30					
C3	D25					
C4	$\overline{CS1}$					
C5	DQM3					
C6	DQM0					
C7	DQS1					
C8	$\overline{RAS}$					
C9	$\overline{CAS}$					
C10	GPIC5			GPOC5	HD13	
C11	GPIC1			GPOC1	HD9	
C12	reserved				HD3	
C13	NC					
C14	NC					
C15	NC					
C16	NC					
C17	NC					
C18	NC					
C19	NC					
C20	NC					
D1	V_{DDM}					
D2	D28					
D3	D27					
D4	GND					
D5	V_{DDM}					
D6	V_{DDM}					
D7	V_{DDM}					
D8	V_{DDM}					
D9	V_{DDM}					

Table 1. MSC7119 Signals by Ball Designator (continued)

Number	Signal Names					
	End of Reset	Software Controlled			Hardware Controlled	
		GPI Enabled (Default)	Interrupt Enabled	GPO Enabled	Primary	Alternate
F6	V_{DDC}					
F7	GND					
F8	GND					
F9	GND					
F10	V_{DDM}					
F11	V_{DDM}					
F12	GND					
F13	GND					
F14	GND					
F15	V_{DDIO}					
F16	V_{DDC}					
F17	V_{DDC}					
F18	NC					
F19	NC					
F20	NC					
G1	GND					
G2	D13					
G3	GND					
G4	V_{DDM}					
G5	V_{DDM}					
G6	GND					
G7	GND					
G8	GND					
G9	GND					
G10	GND					
G11	GND					
G12	GND					
G13	GND					
G14	GND					
G15	V_{DDIO}					
G16	V_{DDIO}					
G17	V_{DDC}					
G18	NC					
G19	NC					
G20	NC					
H1	D14					

Table 1. MSC7119 Signals by Ball Designator (continued)

Number	Signal Names					
	End of Reset	Software Controlled			Hardware Controlled	
		GPI Enabled (Default)	Interrupt Enabled	GPO Enabled	Primary	Alternate
R6	V_{DDM}					
R7	GND					
R8	V_{DDM}					
R9	GND					
R10	V_{DDM}					
R11	GND					
R12	GND					
R13	V_{DDIO}					
R14	GND					
R15	V_{DDIO}					
R16	V_{DDIO}					
R17	V_{DDC}					
R18	TDO					
R19	reserved				EE0/DBREQ	
R20	TEST0					
T1	V_{DDM}					
T2	D20					
T3	D22					
T4	V_{DDM}					
T5	V_{DDM}					
T6	V_{DDC}					
T7	V_{DDM}					
T8	V_{DDM}					
T9	V_{DDC}					
T10	V_{DDM}					
T11	V_{DDM}					
T12	V_{DDIO}					
T13	V_{DDIO}					
T14	V_{DDIO}					
T15	V_{DDIO}					
T16	V_{DDC}					
T17	V_{DDC}					
T18	reserved				MDIO	
T19	TMS					
T20	$\overline{\text{HRESET}}$					
U1	GND					

2.5 AC Timings

This section presents timing diagrams and specifications for individual signals and parallel I/O outputs and inputs. All AC timings are based on a 30 pF load, except where noted otherwise, and a 50 Ω transmission line. For any additional pF, use the following equations to compute the delay:

— Standard interface: 2.45 + (0.054 × C_{load}) ns
— DDR interface: 1.6 + (0.002 × C_{load}) ns

2.5.1 Clock and Timing Signals

The following tables describe clock signal characteristics. **Table 6** shows the maximum frequency values for internal (core, reference, and peripherals) and external (CLKO) clocks. You must ensure that maximum frequency values are not exceeded (see **Section 2.5.2** for the allowable ranges when using the PLL).

Table 6. Maximum Frequencies

Characteristic	Maximum in MHz
Core clock frequency (CLOCK)	300
External output clock frequency (CLKO)	75
Memory clock frequency (CK, $\overline{CK}$)	150
TDM clock frequency (TxRCK, TxTCK)	50

Table 7. Clock Frequencies in MHz

Characteristic	Symbol	Min	Max
CLKIN frequency	F_{CLKIN}	10	100
CLOCK frequency	F_{CORE}	—	300
CK, $\overline{CK}$ frequency	F_{CK}	—	150
TDMxRCK, TDMxTCK frequency	F_{TDMCK}	—	50
CLKO frequency	F_{CKO}	—	75
AHB/IPBus/APB clock frequency	F_{BCK}	—	150
Note: The rise and fall time of external clocks should be 5 ns maximum			

Table 8. System Clock Parameters

Characteristic	Min	Max	Unit
CLKIN frequency	10	100	MHz
CLKIN slope	—	5	ns
CLKIN frequency jitter (peak-to-peak)	—	1000	ps
CLKO frequency jitter (peak-to-peak)	—	150	ps

2.5.4 DDR DRAM Controller Timing

This section provides the AC electrical characteristics for the DDR DRAM interface.

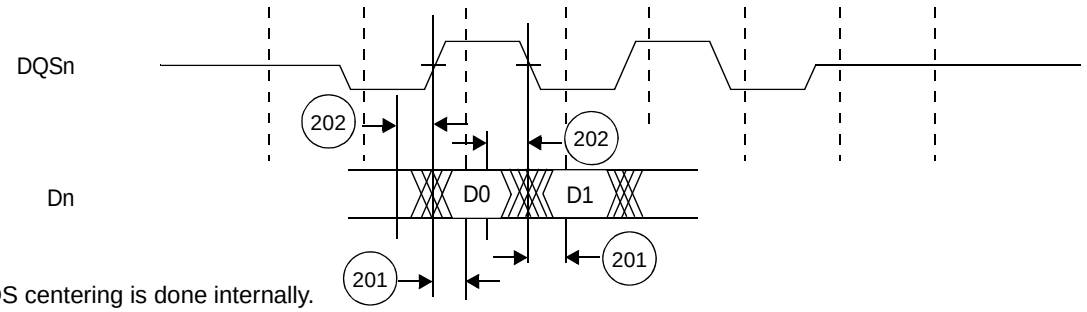
2.5.4.1 DDR DRAM Input AC Timing Specifications

Table 17 provides the input AC timing specifications for the DDR DRAM interface.

Table 17. DDR DRAM Input AC Timing

No.	Parameter	Symbol	Min	Max	Unit
—	AC input low voltage	V_{IL}	—	V_{REF} – 0.31	V
—	AC input high voltage	V_{IH}	V_{REF} + 0.31	V_{DDM} + 0.3	V
201	Maximum Dn input setup skew relative to DQSn input	—	—	900	ps
202	Maximum Dn input hold skew relative to DQSn input	—	—	900	ps

Notes:
1. Maximum possible skew between a data strobe (DQSn) and any corresponding bit of data (D[8n + {0...7}] if 0 ≤ n ≤ 7).
2. See **Table 18** for t_{CK} value.
3. Dn should be driven at the same time as DQSn. This is necessary because the DQSn centering on the DQn data tenure is done internally.



Note: DQS centering is done internally.

Figure 5. DDR DRAM Input Timing Diagram

2.5.4.2 DDR DRAM Output AC Timing Specifications

Table 18 and **Table 19** list the output AC timing specifications and measurement conditions for the DDR DRAM interface.

Table 18. DDR DRAM Output AC Timing

No.	Parameter	Symbol	Min	Max	Unit
200	CK cycle time, (CK/$\overline{CK}$ crossing)[1] • 100 MHz (DDR200) • 150 MHz (DDR300)	t_{CK}	 10 6.67	 — —	 ns ns
204	An/$\overline{RAS}$/$\overline{CAS}$/$\overline{WE}$/CKE output setup with respect to CK	t_{DDKHAS}	0.5 × t_{CK} – 1000	—	ps
205	An/$\overline{RAS}$/$\overline{CAS}$/$\overline{WE}$/CKE output hold with respect to CK	t_{DDKHAX}	0.5 × t_{CK} – 1000	—	ps
206	$\overline{CSn}$ output setup with respect to CK	t_{DDKHCS}	0.5 × t_{CK} – 1000	—	ps
207	$\overline{CSn}$ output hold with respect to CK	t_{DDKHCX}	0.5 × t_{CK} – 1000	—	ps
208	CK to DQSn[2]	t_{DDKHMH}	–600	600	ps

2.5.6.2 Transmit Signal Timing

Table 22. Transmit Signal Timing

No.	Characteristics	Min	Max	Unit
800	Transmit clock period: • MII: TXCLK • RMII: REFCLK	 40 20	 — —	 ns ns
801	Transmit clock pulse width high—as a percent of clock period • MII: RXCLK • RMII: REFCLK	35 14 7	65 — —	% ns ns
802	Transmit clock pulse width low—as a percent of clock period: • MII: RXCLK • RMII: REFCLK	35 14 7	65 — —	% ns ns
805	Transmit clock to TXDn, TX_EN, TX_ER invalid	4	—	ns
806	Transmit clock to TXDn, TX_EN, TX_ER valid	—	14	ns

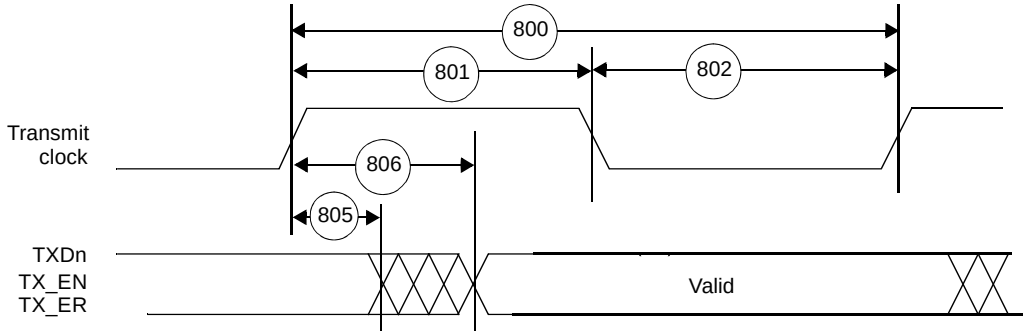


Figure 11. Ethernet Receive Signal Timing

2.5.6.3 Asynchronous Input Signal Timing

Table 23. Asynchronous Input Signal Timing

No.	Characteristics	Min	Max	Unit
807	• MII: CRS and COL minimum pulse width (1.5 × TXCLK period) • RMII: CRS_DV minimum pulse width (1.5 x REFCLK period)	60 30	— —	ns ns

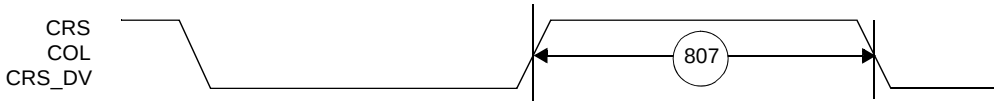


Figure 12. Asynchronous Input Signal Timing

2.5.6.4 Management Interface Timing

Table 24. Ethernet Controller Management Interface Timing

No.	Characteristics	Min	Max	Unit
808	MDC period	400	—	ns
809	MDC pulse width high	160	—	ns
810	MDC pulse width low	160	—	ns
811	MDS falling edge to MDIO output invalid (minimum propagation delay)	0	—	ns
812	MDS falling edge to MDIO output valid (maximum propagation delay)	—	15	ns
813	MDIO input to MDC rising edge setup time	10	—	ns
814	MDC rising edge to MDIO input hold time	10	—	ns

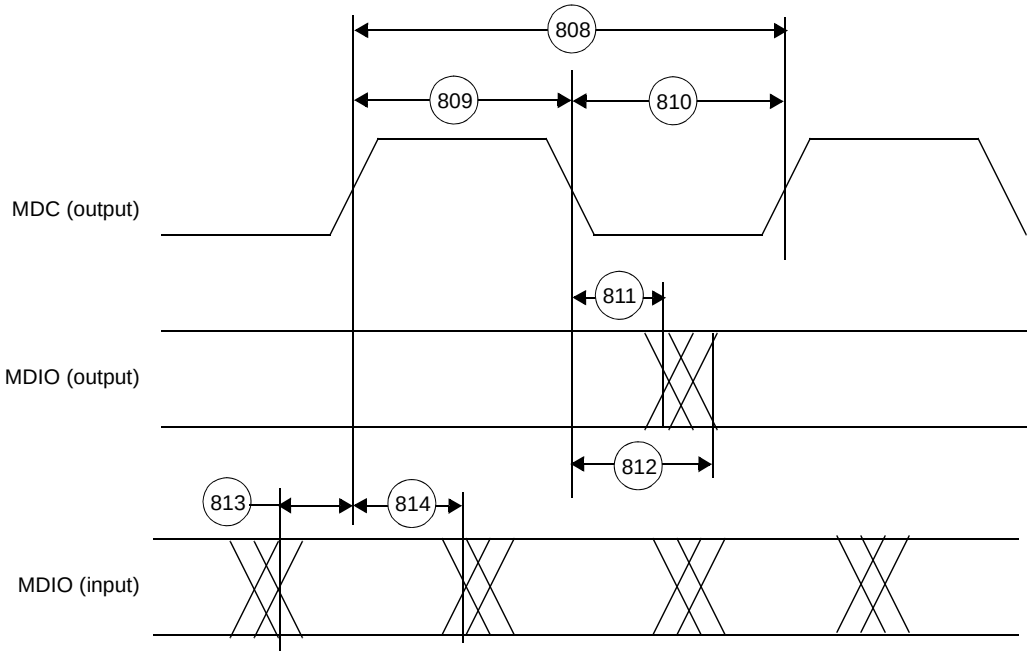


Figure 13. Serial Management Channel Timing

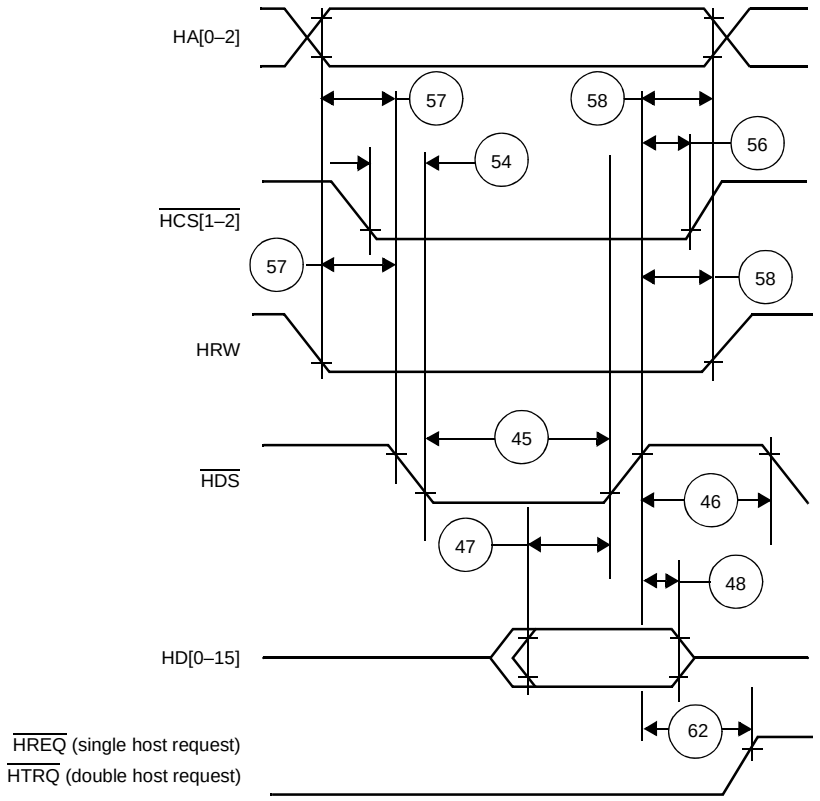


Figure 16. Write Timing Diagram, Single Data Strobe

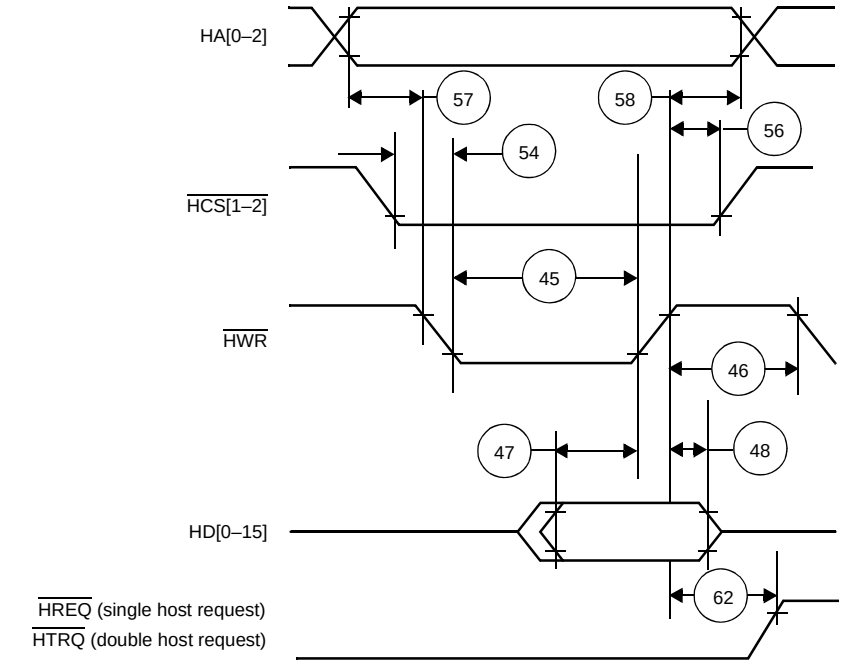


Figure 17. Write Timing Diagram, Double Data Strobe

2.5.9 UART Timing

Table 27. UART Timing

No.	Characteristics	Expression	Min	Max	Unit
—	Internal bus clock (APBCLK)	$F_{CORE}/2$	—	150	MHz
—	Internal bus clock period (1/APBCLK)	T_{APBCLK}	6.67	—	ns
400	URXD and UTXD inputs high/low duration	$16 \times T_{APBCLK}$	106.67	—	ns
401	URXD and UTXD inputs rise/fall time		—	5	ns
402	UTXD output rise/fall time		—	5	ns

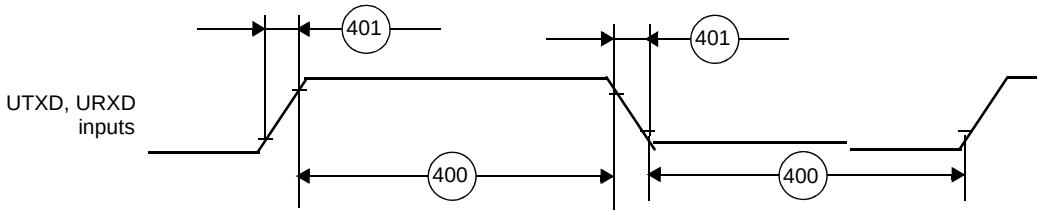


Figure 21. UART Input Timing

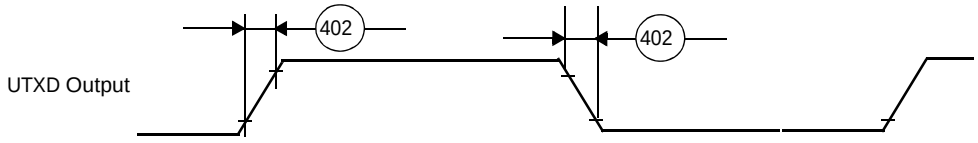


Figure 22. UART Output Timing

2.5.10 EE Timing

Table 28. EE0 Timing

Number	Characteristics	Type	Min
65	EE0 input to the core	Asynchronous	4 core clock periods
66	EE0 output from the core	Synchronous to core clock	1 core clock period

Notes:
1. The core clock is the SC1400 core clock. The ratio between the core clock and CLKOUT is configured during power-on-reset.
2. Configure the direction of the EE pin in the EE_CTRL register (see the *SC140/SC1400 Core Reference Manual* for details.
3. Refer to **Table 1-11** on page 1-16 for details on EE pin functionality.

Figure 24 shows the signal behavior of the EE pin.

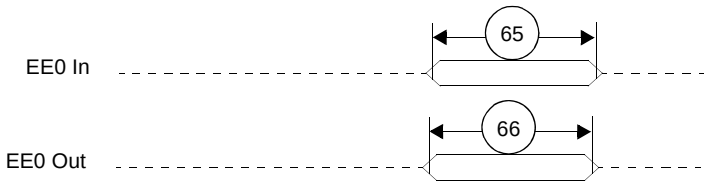


Figure 23. EE Pin Timing

3 Hardware Design Considerations

This section described various areas to consider when incorporating the MSC7119 device into a system design.

3.1 Thermal Design Considerations

An estimation of the chip-junction temperature, T_J, in °C can be obtained from the following:

$$T_J = T_A + (R_{\theta JA} \times P_D) \quad \textit{Eqn. 1}$$

where

T_A = ambient temperature near the package (°C)

$R_{\theta JA}$ = junction-to-ambient thermal resistance (°C/W)

$P_D = P_{INT} + P_{I/O}$ = power dissipation in the package (W)

$P_{INT} = I_{DD} \times V_{DD}$ = internal power dissipation (W)

$P_{I/O}$ = power dissipated from device on output pins (W)

The power dissipation values for the MSC7119 are listed in **Table 4**. The ambient temperature for the device is the air temperature in the immediate vicinity that would cool the device. The junction-to-ambient thermal resistances are JEDEC standard values that provide a quick and easy estimation of thermal performance. There are two values in common usage: the value determined on a single layer board and the value obtained on a board with two planes. The value that more closely approximates a specific application depends on the power dissipated by other components on the printed circuit board (PCB). The value obtained using a single layer board is appropriate for tightly packed PCB configurations. The value obtained using a board with internal planes is more appropriate for boards with low power dissipation (less than 0.02 W/cm^2 with natural convection) and well separated components. Based on an estimation of junction temperature using this technique, determine whether a more detailed thermal analysis is required. Standard thermal management techniques can be used to maintain the device thermal junction temperature below its maximum. If T_J appears to be too high, either lower the ambient temperature or the power dissipation of the chip.

You can verify the junction temperature by measuring the case temperature using a small diameter thermocouple (40 gauge is recommended) or an infrared temperature sensor on a spot on the device case. Use the following equation to determine T_J:

$$T_J = T_T + (\Psi_{JT} \times P_D) \quad \textit{Eqn. 2}$$

where

T_T = thermocouple (or infrared) temperature on top of the package (°C)

Ψ_{JT} = thermal characterization parameter (°C/W)

P_D = power dissipation in the package (W)

3.2.2.1 Case 1

The power-up sequence is as follows:

1. Turn on the V_{DDIO} (3.3 V) supply first.
2. Turn on the V_{DDC} (1.2 V) supply second.
3. Turn on the V_{DDM} (2.5 V) supply third.
4. Turn on the V_{REF} (1.25 V) supply fourth (last).

The power-down sequence is as follows:

1. Turn off the V_{REF} (1.25 V) supply first.
2. Turn off the V_{DDM} (2.5 V) supply second.
3. Turn off the V_{DDC} (1.2 V) supply third.
4. Turn of the V_{DDIO} (3.3 V) supply fourth (last).

Use the following guidelines:

- Make sure that the time interval between the ramp-down of V_{DDIO} and V_{DDC} is less than 10 ms.
- Make sure that the time interval between the ramp-up or ramp-down for V_{DDC} and V_{DDM} is less than 10 ms for power-up and power-down.
- Refer to **Figure 30** for relative timing for power sequencing case 1.

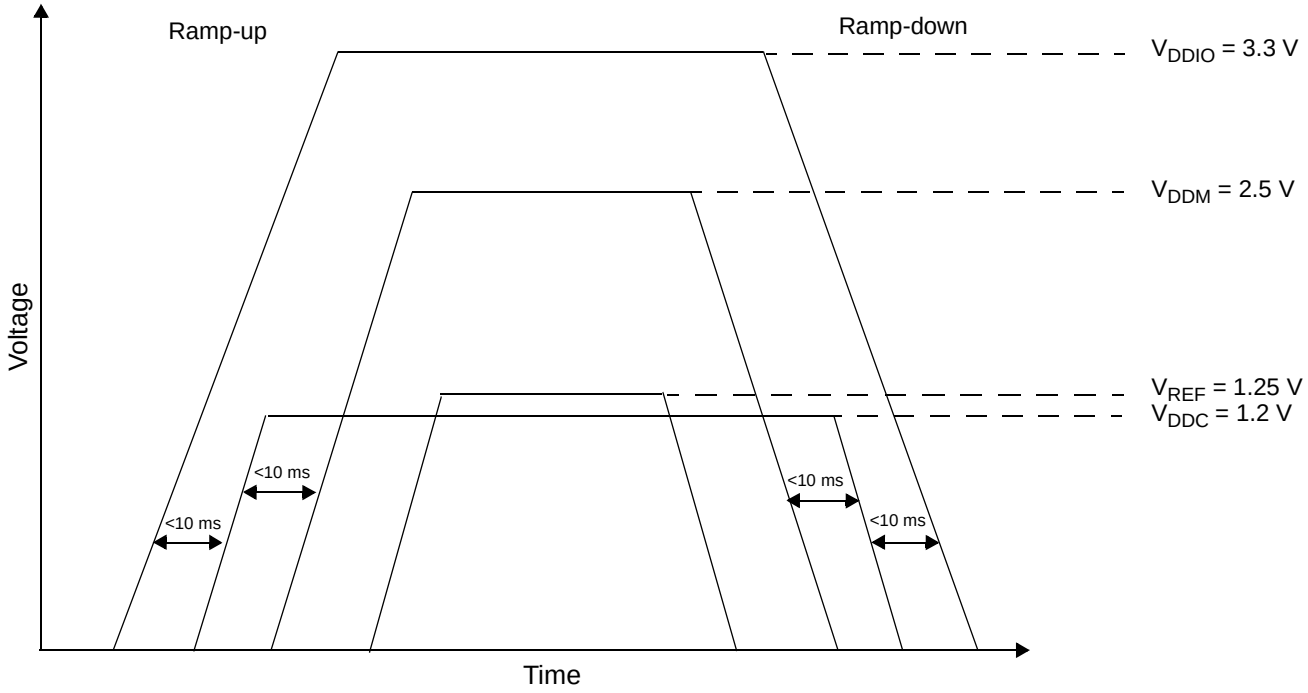


Figure 30. Voltage Sequencing Case 1

3.2.7 Power Supply Design

One of the most common ways to derive power is to use either a simple fixed or adjustable linear regulator. For the system I/O voltage supply, a simple fixed 3.3 V supply can be used. However, a separate adjustable linear regulator supply for the core voltage V_{DDC} should be implemented. For the memory power supply, regulators are available that take care of all DDR power requirements.

Table 33. Recommended Power Supply Ratings

Supply	Symbol	Nominal Voltage	Current Rating
Core	V_{DDC}	1.2 V	1.5 A per device
Memory	V_{DDM}	2.5 V	0.5 A per device
Reference	V_{REF}	1.25 V	10 µA per device
I/O	V_{DDIO}	3.3 V	1.0 A per device

3.3 Estimated Power Usage Calculations

The following equations permit estimated power usage to be calculated for individual design conditions. Overall power is derived by totaling the power used by each of the major subsystems:

$$P_{TOTAL} = P_{CORE} + P_{PERIPHERALS} + P_{DDRIO} + P_{IO} + P_{LEAKAGE} \quad \textit{Eqn. 3}$$

This equation combines dynamic and static power. Dynamic power is determined using the generic equation:

$$C \times V^2 \times F \times 10^{-3}\ mW \quad \textit{Eqn. 4}$$

where,

C = load capacitance in pF

V = peak-to-peak voltage swing in V

F = frequency in MHz

3.3.1 Core Power

Estimation of core power is straightforward. It uses the generic dynamic power equation and assumes that the core load capacitance is 750 pF, core voltage swing is 1.2 V, and the core frequency is 300 MHz. This yields:

$$P_{CORE} = 750\ pF \times (1.2\ V)^2 \times 300\ MHz \times 10^{-3} = 324.0\ mW \quad \textit{Eqn. 5}$$

This equation allows for adjustments to voltage and frequency if necessary.

3.3.6 Example Total Power Consumption

Using the examples in this section and assuming four peripherals and 10 I/O lines active, a total power consumption value is estimated as the following:

$$P_{TOTAL} = 324.0 + (4 \times 4.32) + 326.3 + (10 \times 5.44) + 64 = 784.98\ mW \qquad \textit{Eqn. 13}$$

3.4 Reset and Boot

This section describes the recommendations for configuring the MSC7119 at reset and boot.

3.4.1 Reset Circuit

$\overline{\text{HRESET}}$ is a bidirectional signal and, if driven as an input, should be driven with an open collector or open-drain device. For an open-drain output such as $\overline{\text{HRESET}}$, take care when driving many buffers that implement input bus-hold circuitry. The bus-hold currents can cause enough voltage drop across the pull-up resistor to change the logic level to low. Either a smaller value of pull-up or less current loading from the bus-hold drivers overcomes this issue. To avoid exceeding the MSC7119 output current, the pull-up value should not be too small (a 1 KΩ pull-up resistor is used in the MSC711xADS reference design).

3.4.2 Reset Configuration Pins

Table 34 shows the MSC7119 reset configuration signals. These signals are sampled at the deassertion (rising edge) of $\overline{\text{PORESET}}$. For details, refer to the Reset chapter of the *MSC711x Reference Manual*.

Table 34. Reset Configuration Signals

Signal	Description	Settings
BM[3–0]	Determines boot mode.	See **Table 35** for details.
SWTE	Determines watchdog functionality.	0 Watchdog timer disabled. 1 Watchdog timer enabled.
HDSP	Configures HDI16 strobe polarity.	0 Host Data strobes active low. 1 Host Data strobes active high.
H8BIT	Configures HDI16 operation mode.	0 HDI16 port configured for 16-bit operation. 1 HDI16 port configured for 8-bit operation.

When booting from a power-on reset, the HDI16 is additionally configurable as follows:

- 8- or 16-bit mode as specified by the H8BIT pin.
- Data strobe as specified by the HDSP and HDDS pins.

These pins are sampled only on the deassertion of power-on reset. During a boot from a hard reset, the configuration of these pins is unaffected.

Note: When the HDI16 is used for booting or other purposes, bit 0 is the least significant bit and not the most significant bit as for other DSP products.

3.4.3.2 I^2C Boot

When the MSC7119 device is configured to boot from the I^2C port, the boot program configures the GPIO pins for I^2C operation. Then the MSC7118 device initiates accesses to the I^2C module, downloading data to the MSC7118 device. The I^2C interface is configured as follows:

- PLL is disabled and bypassed so that the I^2C module is clocked with the IPBus clock.
- I^2C interface operates in master mode and polling is used.
- EPROM operates in slave mode.
- Clock divider is set to 128.
- Address of slave during boot is 0xA0.

The IPBus clock is internally divided to generate the bit clock, as follows:

- CLKIN must be a maximum of 100 MHz
- PLL is bypassed.
- IPBus clock = CLKIN/2 is a maximum of 50 MHz.
- I^2C bit clock must be less than or equal to:
  - IPBus clock/I^2C clock divider
  - 50 MHz (max)/128
  - 390.6 KHz

This satisfies the maximum clock rate requirement of 400 kbps for the I^2C interface. For details on the boot procedure, see the "Boot Program" chapter of the *MSC711x Reference Manual*.

3.4.3.3 SPI Boot

When the MSC7119 device is configured to boot from the SPI port, the boot program configures the GPIO pins for SPI operation. Then the MSC7118 device initiates accesses to the SPI module, downloading data to the MSC7118 device. When the SPI routines run in the boot ROM, the MSC7118 is always configured as the SPI master. Booting through the SPI is supported for serial EEPROM devices and serial Flash devices. When a READ_ID instruction is issued to the serial memory device and the device returns a value of 0x00 or 0xFF, the routines for accessing a serial EEPROM are used, at a maximum frequency of 4 Mbps. Otherwise, the routines for accessing a serial Flash are used, and they can run at faster speeds. Booting is performed through one of two sets of pins:

- Main set: BM[2–3], HA3, and HCS2, which allow use of the PLL.
- Alternate set: UTXD, URXD, SDA, and SCL, which cannot be used with the PLL.

In either configuration, an error during SPI boot is flagged on the EVNT3 pin. For details on the boot procedure, see the "Boot Program" chapter of the *MSC711x Reference Manual*.

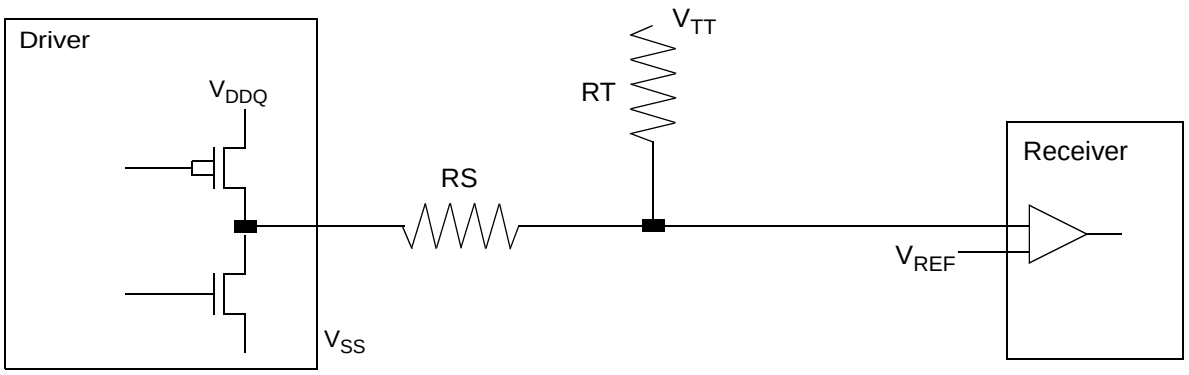


Figure 37. SSTL Power Value

3.5.1 V_{REF} and V_{TT} Design Constraints

V_{TT} and V_{REF} are isolated power supplies at the same voltage, with V_{TT} as a high current power source. This section outlines the voltage supply design needs and goals:

- Minimize the noise on both rails.
- V_{TT} must track variation in the V_{REF} DC offsets. Although they are isolated supplies, one possible solution is to use a single IC to generate both signals.
- Both references should have minimal drift over temperature and source supply.
- It is important to minimize the noise from coupling onto V_{REF} as follows:
  - Isolate V_{REF} and shield it with a ground trace.
  - Use 15–20 mm track.
  - Use 20–30 mm clearance between other traces for isolating.
  - Use the outer layer route when possible.
  - Use distributed decoupling to localize transient currents and return path and decouple with an inductance less than 3 nH.
- Max source/sink transient currents of up to 1.8 A for a 32-bit data bus.
- Use a wide island trace on the outer layer:
  - Place the island at the end of the bus.
  - Decouple both ends of the bus.
  - Use distributed decoupling across the island.
  - Place SSTL termination resistors inside the V_{TT} island and ensure a good, solid connection.
- Place the V_{TT} regulator as closely as possible to the termination island.
  - Reduce inductance and return path.
  - Tie current sense pin at the midpoint of the island.

3.5.2 Decoupling

The DDR decoupling considerations are as follows:

- DDR memory requires significantly more burst current than previous SDRAMs.
- In the worst case, up to 64 drivers may be switching states.
- Pay special attention and decouple discrete ICs per manufacturer guidelines.
- Leverage V_{TT} island topology to minimize the number of capacitors required to supply the burst current needs of the termination rail.
- See the Micron DesignLine publication entitled *Decoupling Capacitor Calculation for a DDR Memory Channel* (`http://download.micron.com/pdf/pubs/designline/3Q00dl1-4.pdf`).

3.5.3 General Routing

The general routing considerations for the DDR are as follows:

- All DDR signals must be routed next to a solid reference:
  - For data, next to solid ground planes.
  - For address/command, power planes if necessary.
- All DDR signals must be impedance controlled. This is system dependent, but typical values are 50–60 ohm.
- Minimize other cross-talk opportunities. As possible, maintain at least a four times the trace width spacing between all DDR signals to non-DDR signals.
- Keep the number of vias to a minimum to eliminate additional stubs and capacitance.
- Signal group routing priorities are as follows:
  - DDR clocks.
  - Route MVTT/MVREF.
  - Data group.
  - Command/address.
- Minimize data bit jitter by trace matching.

3.5.4 Routing Clock Distribution

The DDR clock distribution considerations are as follows:

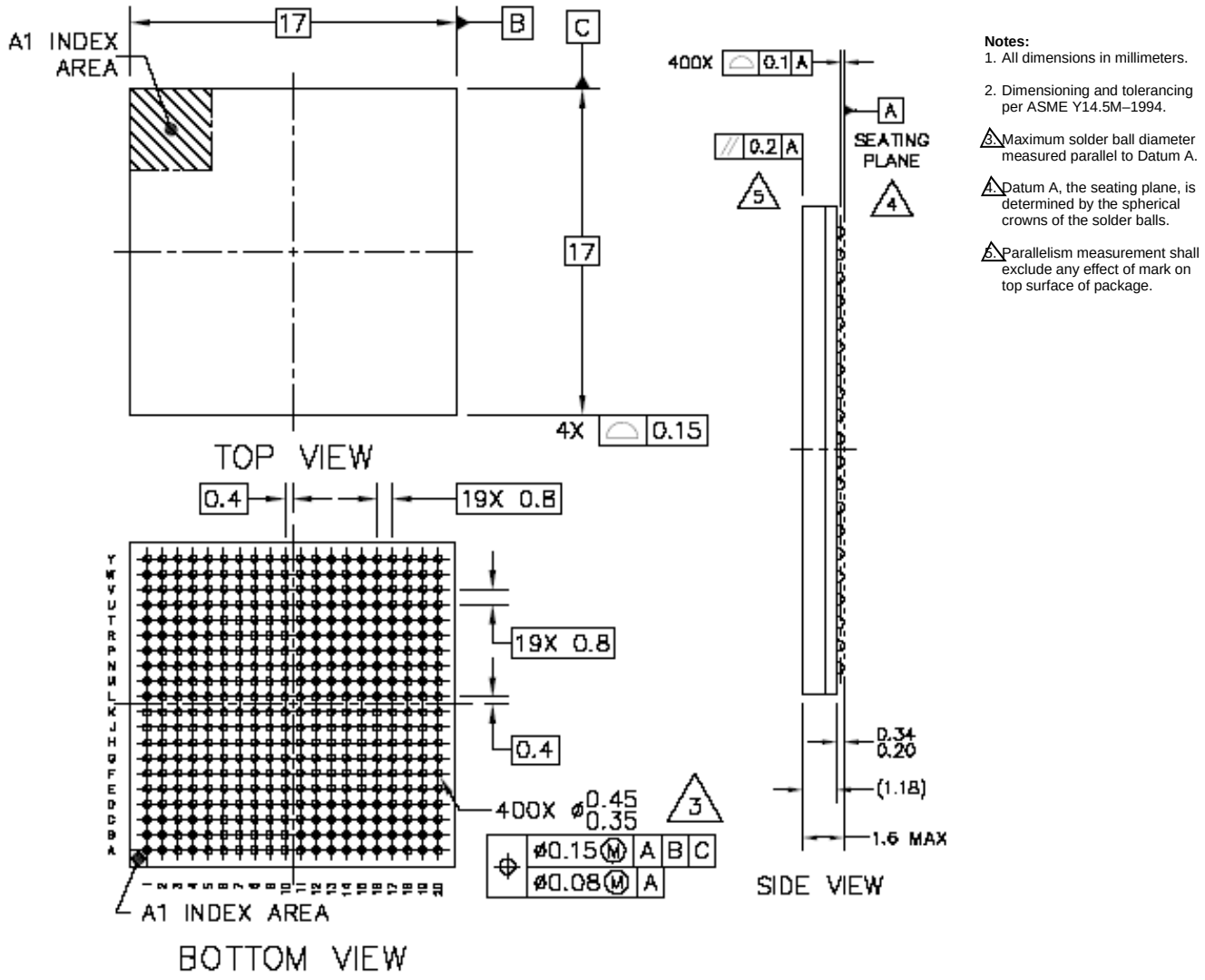
- DDR controller supports six clock pairs:
  - 2 DIMM modules.
  - Up to 36 discrete chips.
- For route traces as for any other differential signals:
  - Maintain proper difference pair spacing.
  - Match pair traces within 25 mm.
- Match all clock traces to within 100 mm.
- Keep all clocks equally loaded in the system.
- Route clocks on inner critical layers.

3.5.5 Data Routing

The DDR data routing considerations are as follows:

- Route each data group (8-bits data + DQS + DM) on the same layer. Avoid switching layers within a byte group.
- Take care to match trace lengths, which is extremely important.
- To make trace matching easier, let adjacent groups be routed on alternate critical layers.
- Pin swap bits within a byte group to facilitate routing (discrete case).
- Tight trace matching is recommended within the DDR data group. Keep each 8-bit datum and its DM signal within ± 25 mm of its respective strobe.
- Minimize lengths across the entire DDR channel:
  - Between all groups maintain a delta of no more than 500 mm.
  - Allows greater flexibility in the design for readjustments as needed.
- DDR data group separation:
  - If stack-up allows, keep DDR data groups away from the address and control nets.
  - Route address and control on separate critical layers.
  - If resistor networks (RNs) are used, attempt to keep data and command lines in separate packages.

5 Package Information



CASE 1568-01

Figure 38. MSC7119 Mechanical Information, 400-pin MAP-BGA Package

6 Product Documentation

- *MSC711x Reference Manual* (MSC711xRM). Includes functional descriptions of the extended cores and all the internal subsystems including configuration and programming information.
- *Application Notes.* Cover various programming topics related to the StarCore DSP core and the MSC7119 device.
- *SC140/SC1400 DSP Core Reference Manual*. Covers the SC140 and SC1400 core architecture, control registers, clock registers, program control, and instruction set.