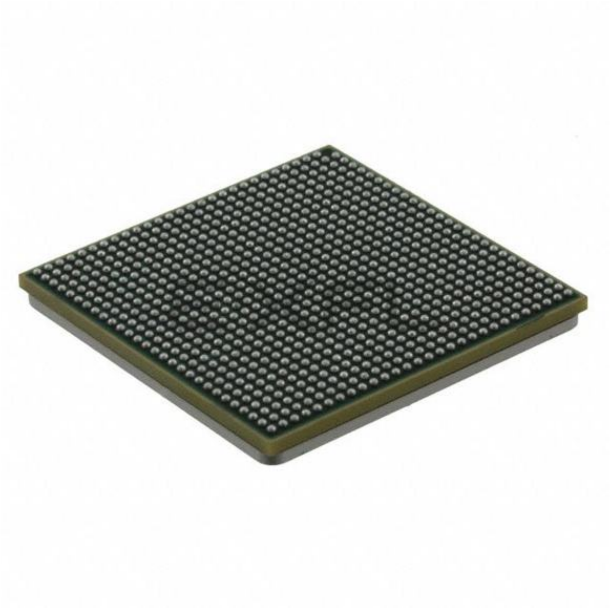

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Understanding Embedded - DSP (Digital Signal Processors)

Embedded - DSP (Digital Signal Processors) are specialized microprocessors designed to perform complex mathematical computations on digital signals in real-time. Unlike general-purpose processors, DSPs are optimized for high-speed numeric processing tasks, making them ideal for applications that require efficient and precise manipulation of digital data. These processors are fundamental in converting and processing signals in various forms, including audio, video, and communication signals, ensuring that data is accurately interpreted and utilized in embedded systems.

Applications of Embedded - DSP (Digital Signal Processors)

Details	
Product Status	Obsolete
Type	SC3850 Single Core
Interface	Ethernet, I²C, PCI, RGMII, Serial RapidIO, SGMII, SPI, UART/USART
Clock Rate	1GHz
Non-Volatile Memory	ROM (96kB)
On-Chip RAM	576kB
Voltage - I/O	2.50V
Voltage - Core	1.00V
Operating Temperature	0°C ~ 105°C (TJ)
Mounting Type	Surface Mount
Package / Case	783-BBGA, FCBGA
Supplier Device Package	783-FCPBGA (29x29)
Purchase URL	https://www.e-xfl.com/product-detail/nxp-semiconductors/msc8251sag1000b

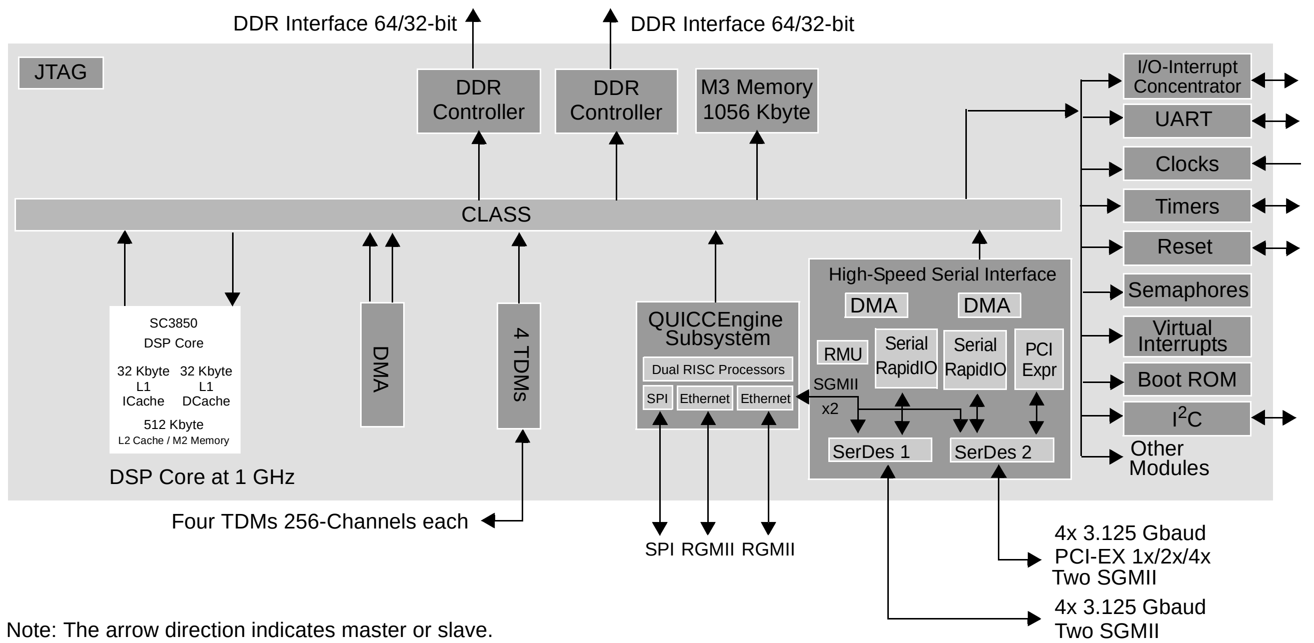


Figure 1. MSC8251 Block Diagram

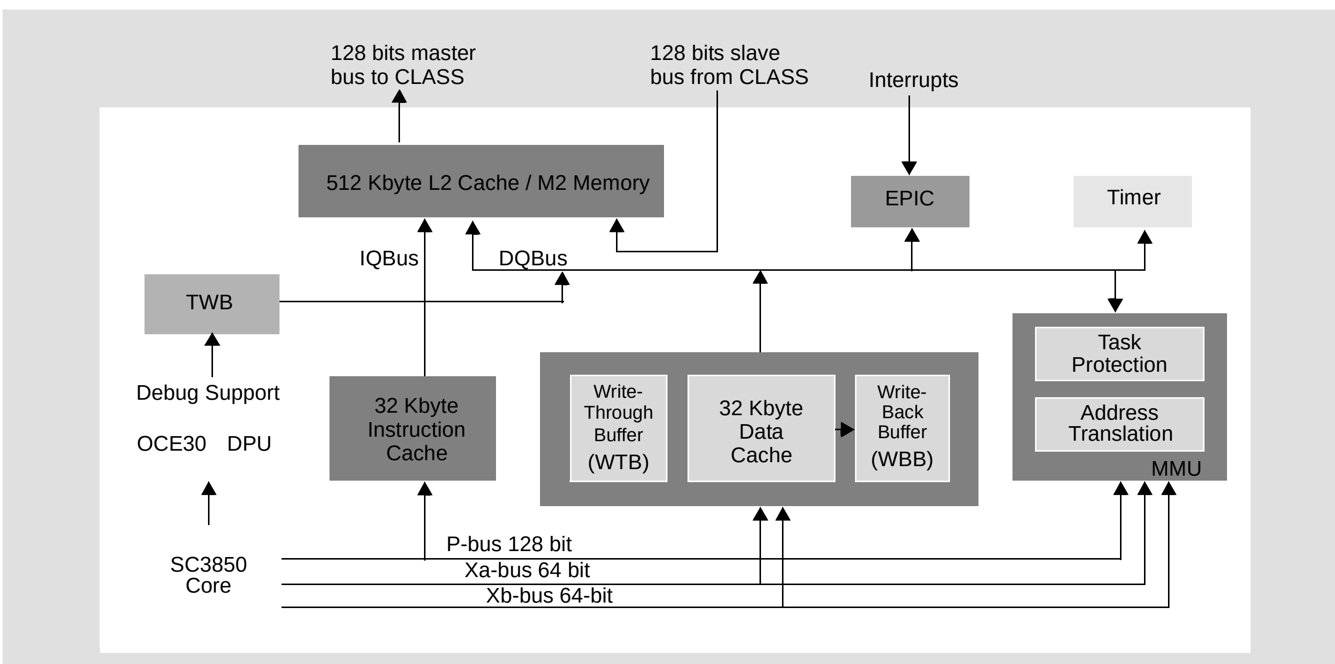


Figure 2. StarCore SC3850 DSP Subsystem Block Diagram

Table 1. Signal List by Ball Number (continued)

Ball Number	Signal Name[1,2]	Pin Type[10]	Power Rail Name
H25	SXCVSS1	Ground	N/A
H26	SXCVDD1	Power	N/A
H27	$\overline{\text{SR1\_RXD3}}$/$\overline{\text{SG2\_RX}}$[4]	I	SXCVDD1
H28	SR1_RXD3/SG2_RX[4]	I	SXCVDD1
J1	M2DQS1	I/O	GVDD2
J2	$\overline{\text{M2DQS1}}$	I/O	GVDD2
J3	M2DQ10	I/O	GVDD2
J4	M2DQ11	I/O	GVDD2
J5	M2DQ14	I/O	GVDD2
J6	M2DQ23	I/O	GVDD2
J7	M2ODT0	O	GVDD2
J8	M2A12	O	GVDD2
J9	M2A14	O	GVDD2
J10	VSS	Ground	N/A
J11	GVDD2	Power	N/A
J12	VSS	Ground	N/A
J13	GVDD2	Power	N/A
J14	VSS	Ground	N/A
J15	GVDD2	Power	N/A
J16	VSS	Ground	N/A
J17	GVDD2	Power	N/A
J18	VSS	Ground	N/A
J19	GVDD2	Power	N/A
J20	Reserved	NC	—
J21	Reserved	NC	—
J22	Reserved	NC	—
J23	SXPVDD1	Power	N/A
J24	SXPVSS1	Ground	N/A
J25	SXCVDD1	Power	N/A
J26	SXCVSS1	Ground	N/A
J27	SXCVDD1	Power	N/A
J28	SXCVSS1	Ground	N/A
K1	VSS	Ground	N/A
K2	GVDD2	Power	N/A
K3	M2DM1	O	GVDD2
K4	VSS	Ground	N/A
K5	GVDD2	Power	N/A
K6	M2DQ0	I/O	GVDD2
K7	VSS	Ground	N/A
K8	GVDD2	Power	N/A
K9	M2DQ5	I/O	GVDD2
K10	VSS	Ground	N/A
K11	VDD	Power	N/A
K12	VSS	Ground	N/A
K13	VDD	Power	N/A
K14	VSS	Ground	N/A

Table 1. Signal List by Ball Number (continued)

Ball Number	Signal Name[1,2]	Pin Type[10]	Power Rail Name
K15	VDD	Power	N/A
K16	VSS	Ground	N/A
K17	VSS	Ground	N/A
K18	VSS	Ground	N/A
K19	VDD	Power	N/A
K20	Reserved	NC	—
K21	Reserved	NC	—
K22	Reserved	NC	—
K23	SXPVDD2	Power	N/A
K24	SXPVSS2	Ground	N/A
K25	SXCVDD2	Power	N/A
K26	SXCVSS2	Ground	N/A
K27	SXCVDD2	Power	N/A
K28	SXCVSS2	Ground	N/A
L1	M2DQ9	I/O	GVDD2
L2	M2DQ12	I/O	GVDD2
L3	M2DQ13	I/O	GVDD2
L4	$\overline{\text{M2DQS0}}$	I/O	GVDD2
L5	M2DQS0	I/O	GVDD2
L6	M2DM0	O	GVDD2
L7	M2DQ3	I/O	GVDD2
L8	M2DQ2	I/O	GVDD2
L9	M2DQ4	I/O	GVDD2
L10	VDD	Power	N/A
L11	VSS	Ground	N/A
L12	M3VDD	Power	N/A
L13	VSS	Ground	N/A
L14	VSS	Ground	N/A
L15	VSS	Ground	N/A
L16	VSS	Ground	N/A
L17	VSS	Ground	N/A
L18	VDD	Power	N/A
L19	VSS	Ground	N/A
L20	Reserved	NC	—
L21	Reserved	NC	—
L22	Reserved	NC	—
L23	SR2_TXD3/PE_TXD3/SG2_TX[4]	O	SXPVDD2
L24	$\overline{\text{SR2\_TXD3}}$/$\overline{\text{PE\_TXD3}}$/$\overline{\text{SG2\_TX}}$[4]	O	SXPVDD2
L25	SXCVSS2	Ground	N/A
L26	SXCVDD2	Power	N/A
L27	$\overline{\text{SR2\_RXD3}}$/$\overline{\text{PE\_RXD3}}$/$\overline{\text{SG2\_RX}}$[4]	I	SXCVDD2
L28	SR2_RXD3/PE_RXD3/SG2_RX[4]	I	SXCVDD2
M1	M2DQ8	I/O	GVDD2
M2	VSS	Ground	N/A
M3	GVDD2	Power	N/A
M4	M2DQ15	I/O	GVDD2

Table 1. Signal List by Ball Number (continued)

Ball Number	Signal Name[1,2]	Pin Type[10]	Power Rail Name
N23	SR2_TXD2/PE_TXD2/SG1_TX[4]	O	SXPVDD2
N24	$\overline{\text{SR2\_TXD2}}$/$\overline{\text{PE\_TXD2}}$/$\overline{\text{SG1\_TX}}$[4]	O	SXPVDD2
N25	SXCVDD2	Power	N/A
N26	SXCVSS2	Ground	N/A
N27	$\overline{\text{SR2\_RXD2}}$/$\overline{\text{PE\_RXD2}}$/$\overline{\text{SG1\_RX}}$[4]	I	SXCVDD2
N28	SR2_RXD2/PE_RXD2/SG1_RX[4]	I	SXCVDD2
P1	CLKIN	I	QVDD
P2	EE0	I	QVDD
P3	QVDD	Power	N/A
P4	VSS	Ground	N/A
P5	STOP_BS	I	QVDD
P6	QVDD	Power	N/A
P7	VSS	Ground	N/A
P8	PLL0_AVDD[9]	Power	VDD
P9	PLL2_AVDD[9]	Power	VDD
P10	VSS	Ground	N/A
P11	VDD	Power	N/A
P12	VSS	Ground	N/A
P13	VDD	Power	N/A
P14	VSS	Ground	N/A
P15	VSS	Ground	N/A
P16	VSS	Ground	N/A
P17	VSS	Ground	N/A
P18	VSS	Ground	N/A
P19	VDD	Power	N/A
P20	Reserved	NC	—
P21	Reserved	NC	—
P22	Reserved	NC	—
P23	SXPVDD2	Power	N/A
P24	SXPVSS2	Ground	N/A
P25	SR2_PLL_AGND[9]	Ground	SXCVSS2
P26	SR2_PLL_AVDD[9]	Power	SXCVDD2
P27	SXCVSS2	Ground	N/A
P28	SXCVDD2	Power	N/A
R1	VSS	Ground	N/A
R2	$\overline{\text{NMI}}$	I	QVDD
R3	$\overline{\text{NMI\_OUT}}$[6]	O	QVDD
R4	$\overline{\text{HRESET}}$[6,7]	I/O	QVDD
R5	$\overline{\text{INT\_OUT}}$[6]	O	QVDD
R6	EE1	O	QVDD
R7	VSS	Ground	N/A
R8	PLL1_AVDD[9]	Power	VDD
R9	VSS	Ground	N/A
R10	VDD	Power	N/A
R11	VSS	Non-user	N/A
R12	VDD	Power	N/A

Table 1. Signal List by Ball Number (continued)

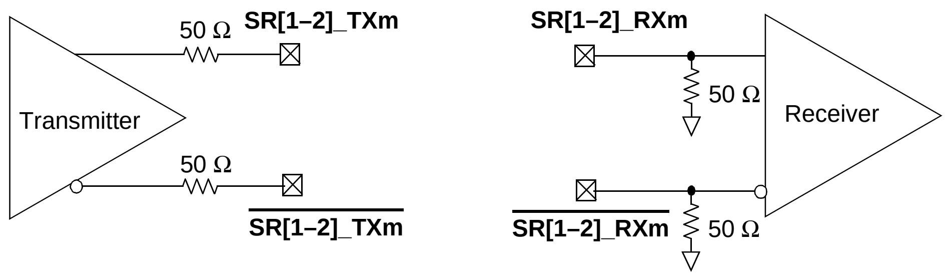
Ball Number	Signal Name[1,2]	Pin Type[10]	Power Rail Name
R13	VSS	Ground	N/A
R14	VDD	Power	N/A
R15	VSS	Ground	N/A
R16	VSS	Ground	N/A
R17	VSS	Ground	N/A
R18	VDD	Power	N/A
R19	VSS	Ground	N/A
R20	VSS	Non-user	N/A
R21	SXPVSS2	Ground	N/A
R22	SXPVDD2	Power	N/A
R23	SR2_TXD1/PE_TXD1[4]	O	SXPVDD2
R24	SR2_TXD1/PE_TXD1[4]	O	SXPVDD2
R25	SXCVSS2	Ground	N/A
R26	SXCVDD2	Power	N/A
R27	SR2_RXD1/PE_RXD1[4]	I	SXCVDD2
R28	SR2_RXD1/PE_RXD1[4]	I	SXCVDD2
T1	VSS	Ground	N/A
T2	TCK	I	QVDD
T3	SRESET[6,7]	I/O	QVDD
T4	TDI	I	QVDD
T5	VSS	Ground	N/A
T6	TDO	O	QVDD
T7	VSS	Ground	N/A
T8	VSS	Ground	N/A
T9	QVDD	Power	N/A
T10	VSS	Ground	N/A
T11	VDD	Power	N/A
T12	VSS	Ground	N/A
T13	M3VDD	Power	N/A
T14	VSS	Ground	N/A
T15	VDD	Power	N/A
T16	VSS	Ground	N/A
T17	VSS	Ground	N/A
T18	VSS	Ground	N/A
T19	VDD	Power	N/A
T20	VSS	Ground	N/A
T21	VSS	Non-user	N/A
T22	SR2_IMP_CAL_RX	I	SXCVDD2
T23	SXPVSS2	Ground	N/A
T24	SXPVDD2	Power	N/A
T25	SR2_REF_CLK	I	SXCVDD2
T26	SR2_REF_CLK	I	SXCVDD2
T27	Reserved	NC	—
T28	Reserved	NC	—
U1	M1DQ8	I/O	GVDD1
U2	VSS	Ground	N/A

Table 1. Signal List by Ball Number (continued)

Ball Number	Signal Name[1,2]	Pin Type[10]	Power Rail Name
AB1	$\overline{\text{M1DQS2}}$	I/O	GVDD1
AB2	M1DQS2	I/O	GVDD1
AB3	M1DQ19	I/O	GVDD1
AB4	M1DM2	O	GVDD1
AB5	M1DQ21	I/O	GVDD1
AB6	M1DQ22	I/O	GVDD1
AB7	M1CKE0	O	GVDD1
AB8	M1A11	O	GVDD1
AB9	M1A7	O	GVDD1
AB10	$\overline{\text{M1CK2}}$	O	GVDD1
AB11	M1APAR_OUT	O	GVDD1
AB12	M1ODT1	O	GVDD1
AB13	$\overline{\text{M1APAR\_IN}}$	I	GVDD1
AB14	M1DQ43	I/O	GVDD1
AB15	M1DM5	O	GVDD1
AB16	M1DQ44	I/O	GVDD1
AB17	M1DQ40	I/O	GVDD1
AB18	M1DQ59	I/O	GVDD1
AB19	M1DM7	O	GVDD1
AB20	M1DQ60	I/O	GVDD1
AB21	VSS	Ground	N/A
AB22	GPIO31/I2C_SDA[5,8]	I/O	NVDD
AB23	GPIO27/TMR4/RCW_SRC0[5,8]	I/O	NVDD
AB24	GPIO25/TMR2/RCW_SRC1[5,8]	I/O	NVDD
AB25	GPIO24/TMR1/RCW_SRC2[5,8]	I/O	NVDD
AB26	GPIO10/$\overline{\text{IRQ10}}$/RC10[5,8]	I/O	NVDD
AB27	GPIO5/$\overline{\text{IRQ5}}$/RC5[5,8]	I/O	NVDD
AB28	GPIO0/$\overline{\text{IRQ0}}$/RC0[5,8]	I/O	NVDD
AC1	VSS	Ground	N/A
AC2	GVDD1	Power	N/A
AC3	M1DQ16	I/O	GVDD1
AC4	VSS	Ground	N/A
AC5	GVDD1	Power	N/A
AC6	M1DQ17	I/O	GVDD1
AC7	VSS	Ground	N/A
AC8	GVDD1	Power	N/A
AC9	M1BA2	O	GVDD1
AC10	VSS	Ground	N/A
AC11	GVDD1	Power	N/A
AC12	M1A4	O	GVDD1
AC13	VSS	Ground	N/A
AC14	GVDD1	Power	N/A
AC15	M1DQ42	I/O	GVDD1
AC16	VSS	Ground	N/A
AC17	GVDD1	Power	N/A
AC18	M1DQ58	I/O	GVDD1

2.5.2.3 SerDes Transmitter and Receiver Reference Circuits

Figure 6 shows the reference circuits for SerDes data lane transmitter and receiver.



Note: The [1–2] indicates the specific SerDes Interface (1 or 2) and the m indicates the specific channel within that interface (0,1,2,3). Actual signals are assigned by the HRCW assignments at reset (see **Chapter 5**, *Reset* in the reference manual for details)

Figure 6. SerDes Transmitter and Receiver Reference Circuits

2.5.3 DC-Level Requirements for SerDes Interfaces

The following subsections define the DC-level requirements for the SerDes reference clocks, the PCI Express data lines, the Serial RapidIO data lines, and the SGMII data lines.

2.5.3.1 DC-Level Requirements for SerDes Reference Clocks

The DC-level requirement for the SerDes reference clock inputs is different depending on the signaling mode used to connect the clock driver chip and SerDes reference clock inputs, as described below:

- Differential Mode
  - — The input amplitude of the differential clock must be between 400 mV and 1600 mV differential peak-peak (or between 200 mV and 800 mV differential peak). In other words, each signal wire of the differential pair must have a single-ended swing of less than 800 mV and greater than 200 mV. This requirement is the same for both external DC-coupled or AC-coupled connection.
  - — For an external DC-coupled connection, the maximum average current requirements sets the requirement for average voltage (common mode voltage) as between 100 mV and 400 mV. Figure 7 shows the SerDes reference clock input requirement for DC-coupled connection scheme.

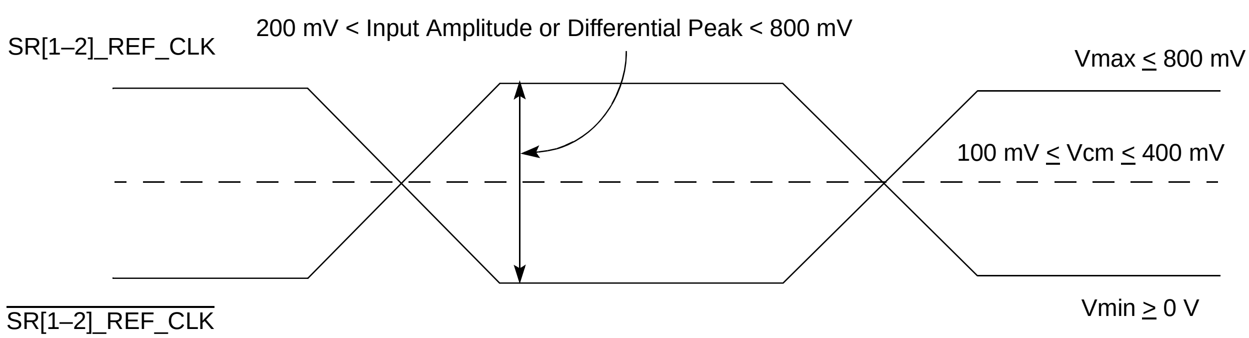


Figure 7. Differential Reference Clock Input DC Requirements (External DC-Coupled)

Figure 11 shows the DDR2 and DDR3 SDRAM interface input timing diagram.

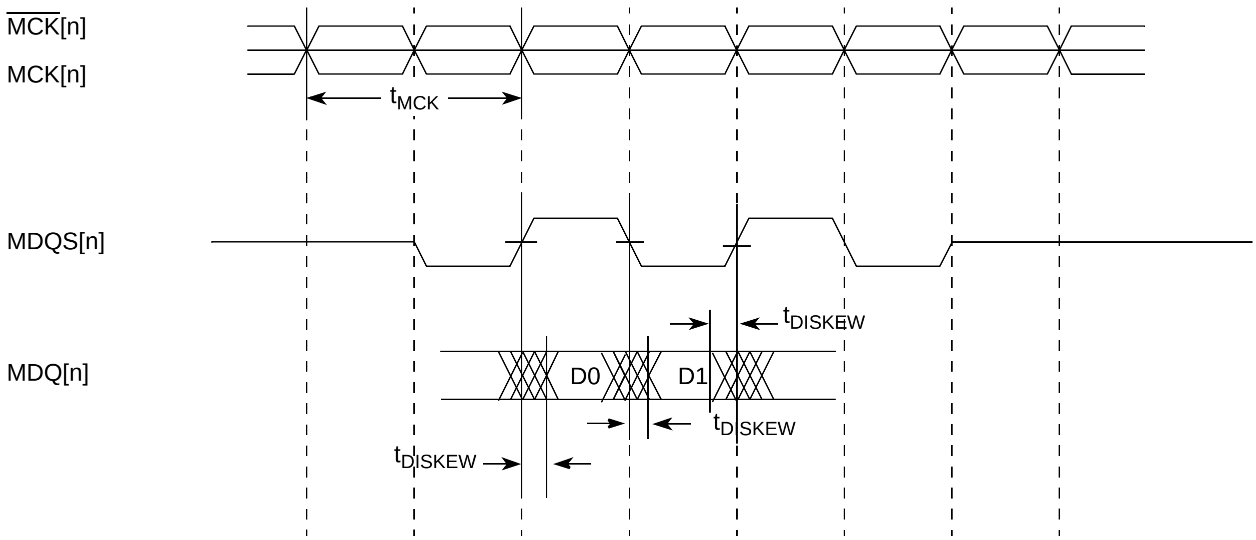


Figure 11. DDR2 and DDR3 SDRAM Interface Input Timing Diagram

2.6.1.2 DDR SDRAM Output AC Timing Specifications

Table 21 provides the output AC timing specifications for the DDR SDRAM interface.

Table 21. DDR SDRAM Output AC Timing Specifications

Parameter	Symbol [1]	Min	Max	Unit	Notes
MCK[n] cycle time	t_{MCK}	2.5	5	ns	2
ADDR/CMD output setup with respect to MCK • 800 MHz data rate • 667 MHz data rate	t_{DDKHAS}	 0.917 1.10	 — —	 ns ns	3
ADDR/CMD output hold with respect to MCK • 800 MHz data rate • 667 MHz data rate	t_{DDKHAX}	 0.767 1.02	 — —	 ns ns	3
$\overline{MCSn}$ output setup with respect to MCK • 800 MHz data rate • 667 MHz data rate	t_{DDKHCS}	 0.917 1.10	 — —	 ns ns	3
$\overline{MCSn}$ output hold with respect to MCK • 800 MHz data rate • 667 MHz data rate	t_{DDKHCX}	 0.767 1.02	 — —	 ns ns	3
MCK to MDQS Skew • 800 MHz data rate • 667 MHz data rate	t_{DDKHMH}	 –0.4 –0.6	 0.375 0.6	ns	4
MDQ/MECC/MDM output setup with respect to MDQS • 800 MHz • 667 MHz	t_{DDKHDS}, t_{DDKLDS}	 300 375	 — —	 ps ps	5
MDQ/MECC/MDM output hold with respect to MDQS • 800 MHz • 667 MHz	t_{DDKHDX}, t_{DDKLDX}	 300 375	 — —	 ps ps	5
MDQS preamble	t_{DDKHMP}	$-0.9 \times t_{MCK}$	—	ns	—
MDQS postamble	t_{DDKHME}	$-0.4 \times t_{MCK}$	$-0.6 \times t_{MCK}$	ns	—

Table 21. DDR SDRAM Output AC Timing Specifications (continued)

Parameter	Symbol [1]	Min	Max	Unit	Notes
Notes: **1.** The symbols used for timing specifications follow the pattern of $t_{\text{(first two letters of functional block)(signal)(state) (reference)(state)}}$ for inputs and $t_{\text{(first two letters of functional block)(reference)(state)(signal)(state)}}$ for outputs. Output hold time can be read as DDR timing (DD) from the rising or falling edge of the reference clock (KH or KL) until the output went invalid (AX or DX). For example, t_{DDKHAS} symbolizes DDR timing (DD) for the time t_{MCK} memory clock reference (K) goes from the high (H) state until outputs (A) are setup (S) or output valid time. Also, t_{DDKLDX} symbolizes DDR timing (DD) for the time t_{MCK} memory clock reference (K) goes low (L) until data outputs (D) are invalid (X) or data output hold time. **2.** All MCK/$\overline{\text{MCK}}$ referenced measurements are made from the crossing of the two signals. **3.** ADDR/CMD includes all DDR SDRAM output signals except MCK/$\overline{\text{MCK}}$, $\overline{\text{MCS}}$, and MDQ/MECC/MDM/MDQS. **4.** Note that t_{DDKHMH} follows the symbol conventions described in note 1. For example, t_{DDKHMH} describes the DDR timing (DD) from the rising edge of the MCK(n) clock (KH) until the MDQS signal is valid (MH). t_{DDKHMH} can be modified through control of the DQSS override bits in the TIMING_CFG_2 register. This will typically be set to the same delay as the clock adjust in the CLK_CNTL register. The timing parameters listed in the table assume that these two parameters have been set to the same adjustment value. See the *MSC8251 Reference Manual* for a description and understanding of the timing modifications enabled by use of these bits. **5.** Determined by maximum possible skew between a data strobe (MDQS) and any corresponding bit of data (MDQ), ECC (MECC), or data mask (MDM). The data strobe should be centered inside of the data eye at the pins of the MSC8251. **6.** At recommended operating conditions with V_{DDDDR} (1.5 V or 1,8 V) ± 5%.					

Note: For the ADDR/CMD setup and hold specifications in Table 21, it is assumed that the clock control register is set to adjust the memory clocks by ½ applied cycle.

Figure 12 shows the DDR SDRAM output timing for the MCK to MDQS skew measurement (t_{DDKHMH}).

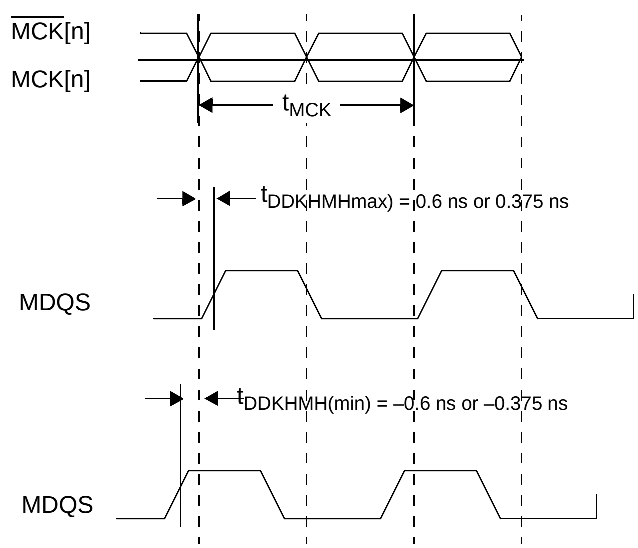


Figure 12. MCK to MDQS Timing

Figure 13 shows the DDR SDRAM output timing diagram.

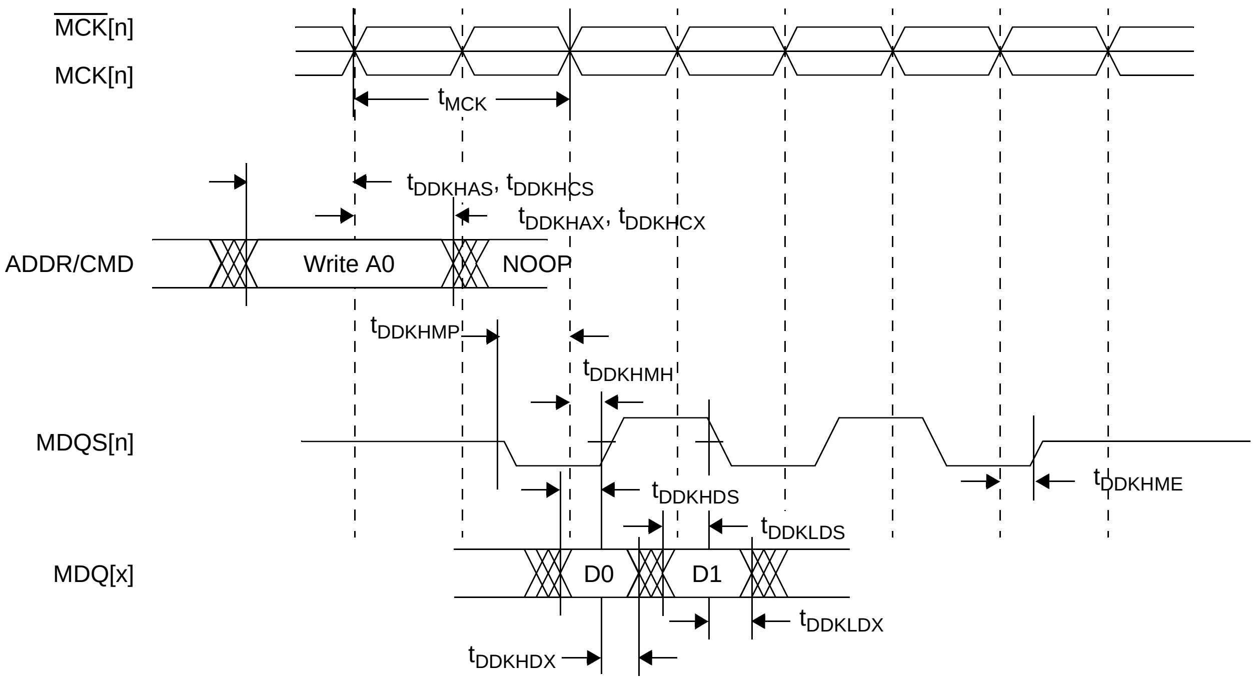


Figure 13. DDR SDRAM Output Timing

Figure 14 provides the AC test load for the DDR2 and DDR3 controller bus.

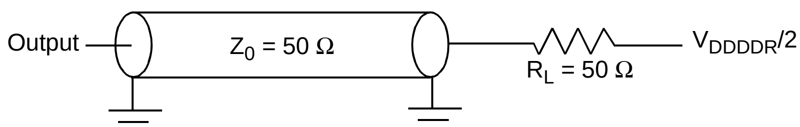


Figure 14. DDR2 and DDR3 Controller Bus AC Test Load

2.6.1.3 DDR2 and DDR3 SDRAM Differential Timing Specifications

This section describes the DC and AC differential timing specifications for the DDR2 and DDR3 SDRAM controller interface. Figure 15 shows the differential timing specification.

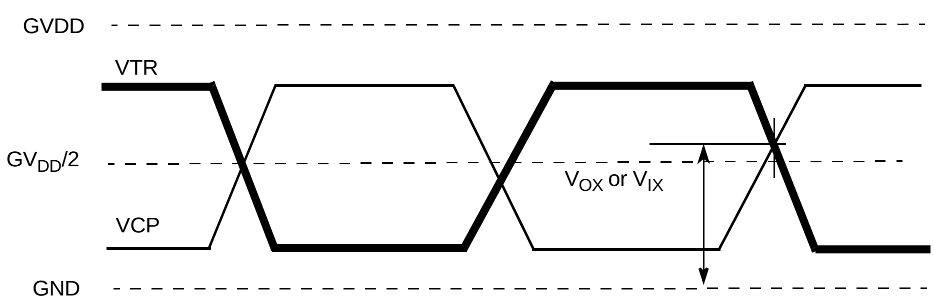


Figure 15. DDR2 and DDR3 SDRAM Differential Timing Specifications

Note: VTR specifies the true input signal (such as MCK or MDQS) and VCP is the complementary input signal (such as $\overline{\text{MCK}}$ or $\overline{\text{MDQS}}$).

Table 22 provides the DDR2 differential specifications for the differential signals MDQS/$\overline{\text{MDQS}}$ and MCK/$\overline{\text{MCK}}$.

Table 22. DDR2 SDRAM Differential Electrical Characteristics

Parameter	Symbol	Min	Max	Unit
Input AC differential cross-point voltage	V_{IXAC}	0.5 × GVDD – 0.175	0.5 × GVDD + 0.175	V
Output AC differential cross-point voltage	V_{OXAC}	0.5 × GVDD – 0.125	0.5 × GVDD + 0.125	V

Table 23 provides the DDR3 differential specifications for the differential signals MDQS/$\overline{\text{MDQS}}$ and MCK/$\overline{\text{MCK}}$.

Table 23. DDR3 SDRAM Differential Electrical Characteristics

Parameter	Symbol	Min	Max	Unit
Input AC differential cross-point voltage	V_{IXAC}	0.5 × GVDD – 0.150	0.5 × GVDD + 0.150	V
Output AC differential cross-point voltage	V_{OXAC}	0.5 × GVDD – 0.115	0.5 × GVDD + 0.115	V

2.6.2 HSSI AC Timing Specifications

The following subsections define the AC timing requirements for the SerDes reference clocks, the PCI Express data lines, the Serial RapidIO data lines, and the SGMII data lines.

2.6.2.1 AC Requirements for SerDes Reference Clock

Table 24 lists AC requirements for the SerDes reference clocks.

Note: Specifications are valid at the recommended operating conditions listed in Table 3.

Table 24. SR[1–2]_REF_CLK and $\overline{\text{SR[1–2]\_REF\_CLK}}$ Input Clock Requirements

Parameter	Symbol	Min	Typical	Max	Units	Notes
SR[1–2]_REF_CLK/$\overline{\text{SR[1–2]\_REF\_CLK}}$ frequency range	t_{CLK_REF}	—	100/125	—	MHz	1
SR[1–2]_REF_CLK/$\overline{\text{SR[1–2]\_REF\_CLK}}$ clock frequency tolerance	t_{CLK_TOL}	–350	—	350	ppm	—
SR[1–2]_REF_CLK/$\overline{\text{SR[1–2]\_REF\_CLK}}$ reference clock duty cycle (measured at 1.6 V)	t_{CLK_DUTY}	40	50	60	%	—
SR[1–2]_REF_CLK/$\overline{\text{SR[1–2]\_REF\_CLK}}$ max deterministic peak-peak jitter at 10^{-6} BER	t_{CLK_DJ}	—	—	42	ps	—
SR[1–2]_REF_CLK/$\overline{\text{SR[1–2]\_REF\_CLK}}$ total reference clock jitter at 10^{-6} BER (peak-to-peak jitter at ref_clk input)	t_{CLK_TJ}	—	—	86	ps	2
SR[1–2]_REF_CLK/$\overline{\text{SR[1–2]\_REF\_CLK}}$ rising/falling edge rate	t_{CLKRR}/t_{CLKFR}	1	—	4	V/ns	3
Differential input high voltage	V_{IH}	200	—	—	mV	4
Differential input low voltage	V_{IL}	—	—	–200	mV	4
Rising edge rate (SR[1–2]_REF_CLK) to falling edge rate (SR[1–2]_REF_CLK) matching	Rise-Fall Matching	—	—	20	%	5, 6

Table 24. SR[1–2]_REF_CLK and SR[1–2]_REF_CLK Input Clock Requirements (continued)

Parameter	Symbol	Min	Typical	Max	Units	Notes
Notes: **1.** Caution: Only 100 and 125 have been tested. Other values will not work correctly with the rest of the system. **2.** Limits from PCI Express CEM Rev 1.0a **3.** Measured from –200 mV to +200 mV on the differential waveform (derived from SR[1–2]_REF_CLK minus SR[1–2]_REF_CLK). The signal must be monotonic through the measurement region for rise and fall time. The 400 mV measurement window is centered on the differential zero crossing. See Figure 16. **4.** Measurement taken from differential waveform **5.** Measurement taken from single-ended waveform **6.** Matching applies to rising edge for SR[1–2]_REF_CLK and falling edge rate for SR[1–2]_REF_CLK. It is measured using a 200 mV window centered on the median cross point where SR[1–2]_REF_CLK rising meets SR[1–2]_REF_CLK falling. The median cross point is used to calculate the voltage thresholds that the oscilloscope uses for the edge rate calculations. The rise edge rate of SR[1–2]_REF_CLK should be compared to the fall edge rate of SR[1–2]_REF_CLK; the maximum allowed difference should not exceed 20% of the slowest edge rate. See Figure 17.						

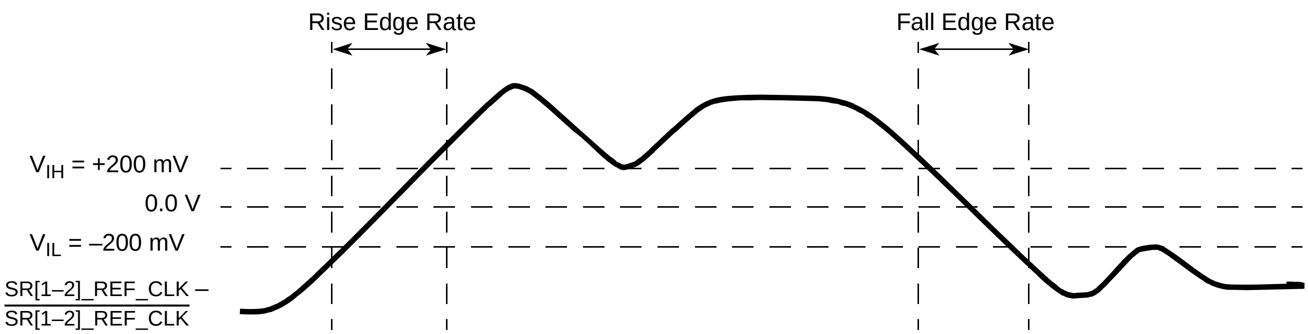


Figure 16. Differential Measurement Points for Rise and Fall Time

Figure 17. Single-Ended Measurement Points for Rise and Fall Time Matching

2.6.2.2 PCI Express AC Physical Layer Specifications

The AC requirements for PCI Express implementations have separate requirements for the Tx and Rx lines. The MSC8251 supports a 2.5 Gbps PCI Express interface defined by the *PCI Express Base Specification, Revision 1.0a*. The transmitter specifications are defined in Table 25 and the receiver specifications are defined in Table 26. The parameters are specified at the component pins. the AC timing specifications do not include REF_CLK jitter.

Note: Specifications are valid at the recommended operating conditions listed in Table 3.

Table 25. PCI Express (2.5 Gbps) Differential Transmitter (Tx) Output AC Specifications

Parameter	Symbol	Min	Typical	Max	Units	Notes
Unit interval	UI	399.88	400.00	400.12	ps	1
Minimum Tx eye width	$T_{TX\text{-}EYE}$	0.70	—	—	UI	2, 3
Maximum time between the jitter median and maximum deviation from the median.	$T_{TX\text{-}EYE\text{-}MEDIAN\text{-}to\text{-}MAX\text{-}JITTER}$	—	—	0.15	UI	3, 4
AC coupling capacitor	C_{TX}	75	—	200	nF	5

Notes:
1. Each UI is 400 ps ± 300 ppm. UI does not account for spread spectrum clock dictated variations. No test load is necessarily associated with this value.
2. The maximum transmitter jitter can be derived as $T_{TX\text{-}MAX\text{-}JITTER} = 1 - T_{TX\text{-}EYE} = 0.3$ UI.
3. Specified at the measurement point into a timing and voltage compliance test load as shown in Figure 8 and measured over any 250 consecutive Tx UIs. A $T_{TX\text{-}EYE}$ = 0.70 UI provides for a total sum of deterministic and random jitter budget of $T_{TX\text{-}JITTER\text{-}MAX}$ = 0.30 UI for the transmitter collected over any 250 consecutive Tx UIs. The $T_{TX\text{-}EYE\text{-}MEDIAN\text{-}to\text{-}MAX\text{-}JITTER}$ median is less than half of the total Tx jitter budget collected over any 250 consecutive Tx UIs. It should be noted that the median is not the same as the mean. The jitter median describes the point in time where the number of jitter points on either side is approximately equal as opposed to the averaged time value. Jitter is defined as the measurement variation of the crossing points ($V_{TX\text{-}DIFFp\text{-}p}$ = 0 V) in relation to a recovered Tx UI. A recovered Tx UI is calculated over 3500 consecutive unit intervals of sample data.
4. Jitter is measured using all edges of the 250 consecutive UI in the center of the 3500 UI used for calculating the Tx UI.
5. All transmitters shall be AC-coupled. The AC coupling is required either within the media or within the transmitting component itself. The SerDes transmitter does not have built-in Tx capacitance. An external AC coupling capacitor is required.

Table 26. PCI Express (2.5 Gbps) Differential Receiver (Rx) Input AC Specifications

Parameter	Symbol	Min	Typical	Max	Units	Notes
Unit Interval	UI	399.88	400.00	400.12	ps	1
Minimum receiver eye width	$T_{RX\text{-}EYE}$	0.4	—	—	UI	2, 3, 4
Maximum time between the jitter median and maximum deviation from the median.	$T_{RX\text{-}EYE\text{-}MEDIAN\text{-}to\text{-}MAX\text{-}JITTER}$	—	—	0.3	UI	3, 4, 5

Notes:
1. Each UI is 400 ps ± 300 ppm. UI does not account for spread spectrum clock dictated variations. No test load is necessarily associated with this value.
2. The maximum interconnect media and transmitter jitter that can be tolerated by the receiver can be derived as $T_{RX\text{-}MAX\text{-}JITTER} = 1 - T_{RX\text{-}EYE} = 0.6$ UI.
3. Specified at the measurement point and measured over any 250 consecutive UIs. The test load in Figure 8 should be used as the Rx device when taking measurements. If the clocks to the Rx and Tx are not derived from the same reference clock, the Tx UI recovered from 3500 consecutive UI must be used as a reference for the eye diagram.
4. A $T_{RX\text{-}EYE}$ = 0.40 UI provides for a total sum of 0.60 UI deterministic and random jitter budget for the transmitter and interconnect collected any 250 consecutive UIs. The $T_{RX\text{-}EYE\text{-}MEDIAN\text{-}to\text{-}MAX\text{-}JITTER}$ specification ensures a jitter distribution in which the median and the maximum deviation from the median is less than half of the total. UI jitter budget collected over any 250 consecutive Tx UIs. It should be noted that the median is not the same as the mean. The jitter median describes the point in time where the number of jitter points on either side is approximately equal as opposed to the averaged time value. If the clocks to the Rx and Tx are not derived from the same reference clock, the Tx UI recovered from 3500 consecutive UI must be used as the reference for the eye diagram.
5. Jitter is defined as the measurement variation of the crossing points ($V_{RX\text{-}DIFFp\text{-}p}$ = 0 V) in relation to a recovered Tx UI. A recovered Tx UI is calculated over 3500 consecutive unit intervals of sample data. Jitter is measured using all edges of the 250 consecutive UI in the center of the 3500 UI used for calculating the Tx UI. It is recommended that the recovered Tx UI is calculated using all edges in the 3500 consecutive UI interval with a fit algorithm using a minimization merit function. Least squares and median deviation fits have worked well with experimental and simulated data.

2.6.5 Ethernet Timing

This section describes the AC electrical characteristics for the Ethernet interface.

There are programmable delay units (PDU) that should be programmed differently for each interface to meet timing. There is a general configuration register 4 (GCR4) used to configure the timing. For additional information, see the *MSC8251 Reference Manual*.

2.6.5.1 Management Interface Timing

Table 33 lists the timer input Ethernet controller management interface timing specifications shown in Figure 24.

Table 33. Ethernet Controller Management Interface Timing

Characteristics	Symbol	Min	Max	Unit
GE_MDC frequency	f_{MDC}	—	2.5	MHz
GE_MDC period	t_{MDC}	400	—	ns
GE_MDC clock pulse width high	t_{MDC_H}	160	—	ns
GE_MDC clock pulse width low	t_{MDC_L}	160	—	ns
GE_MDC to GE_MDIO delay[2]	t_{MDKHDX}	10	70	ns
GE_MDIO to GE_MDC rising edge setup time	t_{MDDVKH}	20	—	ns
GE_MDC rising edge to GE_MDIO hold time	t_{MDDXKH}	0	—	ns

Notes:
1. Program the GE_MDC frequency (f_{MDC}) to a maximum value of 2.5 MHz (400 ns period for t_{MDC}). The value depends on the source clock and configuration of MIIMCFG[MCS] and UPSMR[MDCP]. For example, for a source clock of 400 MHz to achieve f_{MDC} = 2.5 MHz, program MIIMCFG[MCS] = 0x4 and UPSMR[MDCP] = 0. See the *MSC8251 Reference Manual* for configuration details.
2. The value depends on the source clock. For example, for a source clock of 267 MHz, the delay is 70 ns. For a source clock of 333 MHz, the delay is 58 ns.

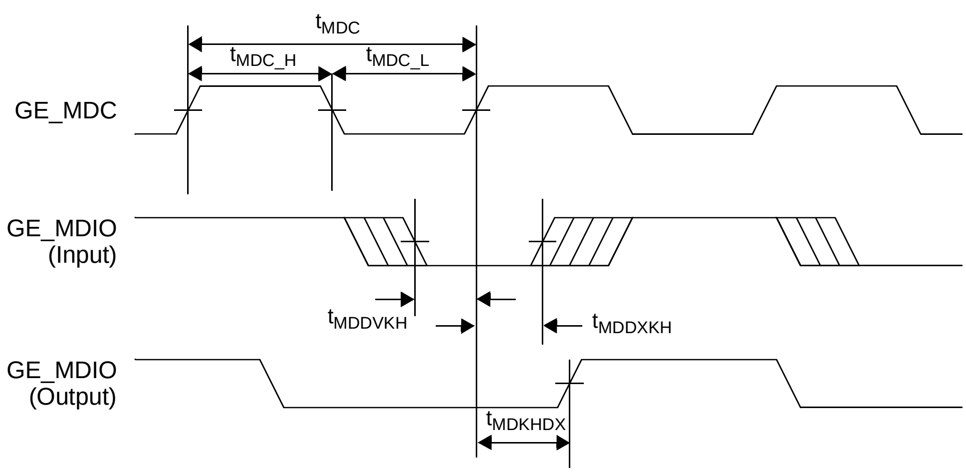


Figure 24. MII Management Interface Timing

2.6.5.2 RGMII AC Timing Specifications

Table 34 presents the RGMII AC timing specifications for applications requiring an on-board delayed clock.

Table 34. RGMII at 1 Gbps[2] with On-Board Delay[3] AC Timing Specifications

Parameter/Condition	Symbol	Min	Typ	Max	Unit
Data to clock output skew (at transmitter)[4]	t_{SKEWT}	–0.5	—	0.5	ns
Data to clock input skew (at receiver) [4]	t_{SKEWR}	1	—	2.6	ns

Notes:
1. At recommended operating conditions with V_{DDIO} of 2.5 V ± 5%.
2. RGMII at 100 Mbps support is guaranteed by design.
3. Program GCR4 as 0x00000000.
4. This implies that PC board design requires clocks to be routed such that an additional trace delay of greater than 1.5 ns and less than 2.0 ns is added to the associated clock signal.

Table 35 presents the RGMII AC timing specification for applications required non-delayed clock on board.

Table 35. RGMII at 1 Gbps[2] with No On-Board Delay[3] AC Timing Specifications

Parameter/Condition	Symbol	Min	Typ	Max	Unit
Data to clock output skew (at transmitter)[4]	t_{SKEWT}	–2.6	—	–1.0	ns
Data to clock input skew (at receiver)[4]	t_{SKEWR}	–0.5	—	0.5	ns

Notes:
1. At recommended operating conditions with V_{DDIO} of 2.5 V ± 5%.
2. RGMII at 100 Mbps support is guaranteed by design.
3. GCR4 should be programmed as 0x000CC330.
4. This implies that PC board design requires clocks to be routed with no additional trace delay

Figure 25 shows the RGMII AC timing and multiplexing diagrams.

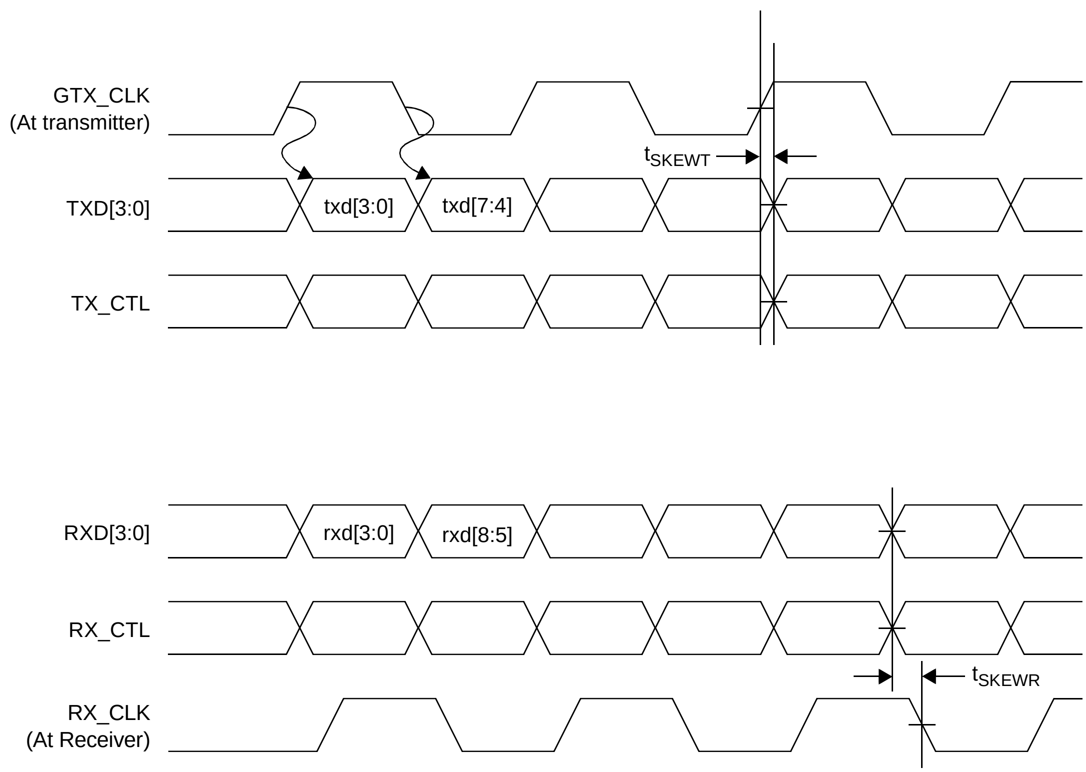


Figure 25. RGMII AC Timing and Multiplexing

2.6.7 Asynchronous Signal Timing

Table 35 lists the asynchronous signal timing specifications.

Table 37. Signal Timing

Characteristics	Symbol	Type	Min
Input	t_{IN}	Asynchronous	One CLKIN cycle
Output	t_{OUT}	Asynchronous	Application dependent
Note: Input value relevant for EE0, $\overline{\text{IRQ[15–0]}}$, and $\overline{\text{NMI}}$ only.			

The following interfaces use the specified asynchronous signals:

- *GPIO*. Signals GPIO[31–0], when used as GPIO signals, that is, when the alternate multiplexed special functions are not selected.

Note: When used as a general purpose input (GPI), the input signal should be driven until it is acknowledged by the MSC8251 device, that is, when the expected input value is read from the GPIO data register.

- *EE port*. Signals EE0, EE1.
- *Boot function*. Signal STOP_BS.
- *I^2C interface*. Signals I2C_SCL and I2C_SDA.
- *Interrupt inputs*. Signals $\overline{\text{IRQ[15–0]}}$ and $\overline{\text{NMI}}$.
- *Interrupt outputs*. Signals $\overline{\text{INT\_OUT}}$ and $\overline{\text{NMI\_OUT}}$ (minimum pulse width is 32 ns).

2.6.8 JTAG Signals

Table 38 lists the JTAG timing specifications shown in Figure 29 through Figure 32.

Table 38. JTAG Timing

Characteristics	Symbol	All frequencies		Unit
		Min	Max	
TCK cycle time	t_{TCKX}	36.0	—	ns
TCK clock high phase measured at $V_M = V_{DDIO}/2$	t_{TCKH}	15.0	—	ns
Boundary scan input data setup time	t_{BSVKH}	0.0	—	ns
Boundary scan input data hold time	t_{BSXKH}	15.0	—	ns
TCK fall to output data valid	t_{TCKHOV}	—	20.0	ns
TCK fall to output high impedance	t_{TCKHOZ}	—	24.0	ns
TMS, TDI data setup time	t_{TDIVKH}	0.0	—	ns
TMS, TDI data hold time	t_{TDIXKH}	5.0	—	ns
TCK fall to TDO data valid	t_{TDOHOV}	—	10.0	ns
TCK fall to TDO high impedance	t_{TDOHOZ}	—	12.0	ns
$\overline{\text{TRST}}$ assert time	t_{TRST}	100.0	—	ns
Note: All timings apply to OnCE module data transfers as well as any other transfers via the JTAG port.				

Figure 29 shows the test clock input timing diagram.

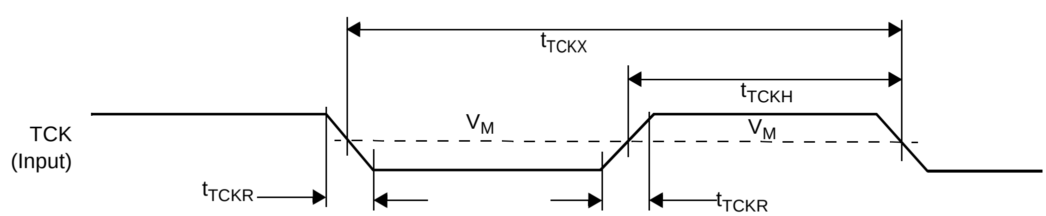


Figure 29. Test Clock Input Timing

3 Hardware Design Considerations

The following sections discuss areas to consider when the MSC8251 device is designed into a system.

3.1 Power Supply Ramp-Up Sequence

The following subsections describe the required device initialization sequence.

3.1.1 Clock, Reset, and Supply Coordination

Starting the device requires coordination between several inputs including: clock, reset, and power supplies. follow this guidelines when starting up an MSC8251 device:

- $\overline{\text{PORESET}}$ and $\overline{\text{TRST}}$ must be asserted externally for the duration of the supply ramp-up, using the V_{DDIO} supply. $\overline{\text{TRST}}$ deassertion does not have to be synchronized with $\overline{\text{PORESET}}$ deassertion. However, $\overline{\text{TRST}}$ must be deasserted before normal operation begins to ensure correct functionality of the device.
- CLKIN should toggle at least 32 cycles before $\overline{\text{PORESET}}$ deassertion to guarantee correct device operation. The 32 cycles should only be counted from the time after V_{DDIO} reaches its nominal value (see timing 1 in Figure 33).
- CLKIN should either be stable low during ramp-up of V_{DDIO} supply (and start its swings after ramp-up) or should swing within V_{DDIO} range during V_{DDIO} ramp-up, so its amplitude grows as V_{DDIO} grows during ramp-up.

Figure 33 shows a sequence in which V_{DDIO} ramps-up after V_{DD} and CLKIN begins to toggle with the raise of V_{DDIO} supply.

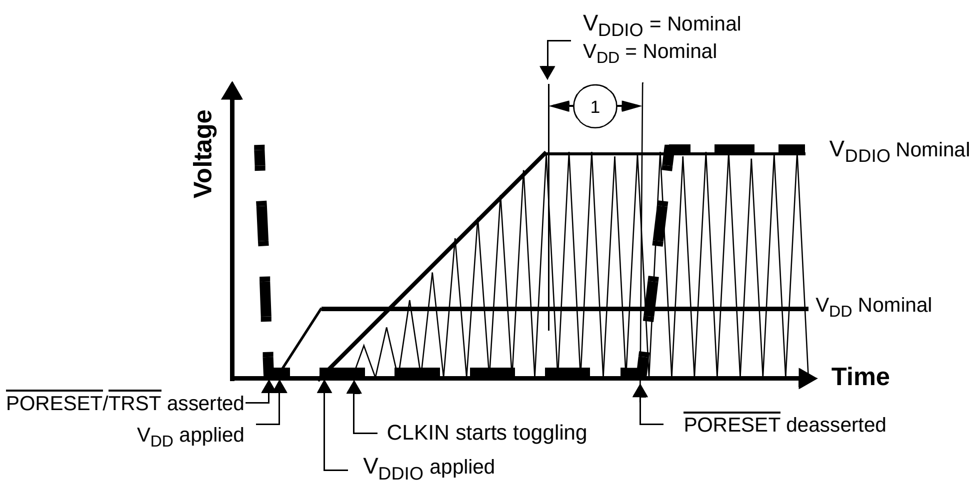


Figure 33. Supply Ramp-Up Sequence with V_{DD} Ramping Before V_{DDIO} and CLKIN Starting With V_{DDIO}

Note: For details on power-on reset flow and duration, see the *Reset* chapter in the *MSC8251 Reference Manual*.

3.1.2 Power-On Ramp Time

This section describes the AC electrical specification for the power-on ramp rate requirements for all voltage supplies (including GVDD/SXPVDD/SXCVDD/QVDD/GVDD/NVDD, all VDD supplies, MVREF, and all AVDD supplies). Controlling the power-on ramp time is required to avoid falsely triggering the ESD circuitry. Table 39 defines the power supply ramp time specification.

Table 39. Power Supply Ramp Rate

Parameter	Min	Max	Unit
Required ramp rate.	—	36000	V/s

Notes:
1. Ramp time is specified as a linear ramp from 10% to 90% of nominal voltage of the specific voltage supply. If the ramp is non-linear (for example, exponential), the maximum rate of change from 200 to 500 mV is the most critical because this range might falsely trigger the ESD circuitry.
2. Required over the full recommended operating temperature range (see Table 3).
3. All supplies must be at their stable values within 50 ms.
4. The GVDD pins can be held low on the application board at powerup. If GVDD is not held low, then GVDD will rise to a voltage level that depends on the board-level impedance-to-ground. If the impedance is high (that is, infinite), then theoretically, GVDD can rise up close to the VDD levels.

3.1.3 Power Supply Guidelines

Use the following guidelines for power-up sequencing:

- Couple M3VDD with the VDD power rail using an extremely low impedance path.
- Couple inputs PLL1_AVDD, PLL2_AVDD and PLL3_AVDD with the VDD power rail using an RC filter (see Figure 37).
- There is no dependency in power-on/power-off sequence between the GVDD1, GVDD2, NVDD, and QVDD power rails.
- Couple inputs M1VREF and M2VREF with the GVDD1 and GVDD2 power rails, respectively. They should rise at the same time as or after their respective power rail.
- There is no dependency between RapidIO supplies: SXCVDD1, SXCVDD2, SXPVDD1 and SXPVDD2 and other MSC8251 supplies in the power-on/power-off sequence.
- Couple inputs SR1_PLL_AVDD and SR2_PLL_AVDD with SXCVDD1 and SXCVDD2 power rails, respectively, using an RC filter (see Figure 38).

External voltage applied to any input line must not exceed the I/O supply voltage related to this line by more than 0.6 V at any time, including during power-up. Some designs require pull-up voltages applied to selected input lines during power-up for configuration purposes. This is an acceptable exception to the rule during start-up. However, each such input can draw up to 80 mA per input pin per MSC8251 device in the system during power-up. An assertion of the inputs to the high voltage level before power-up should be with slew rate less than 4 V/ns.

The device power rails should rise in the following sequence:

1. VDD (and all coupled supplies)

3.3 Clock and Timing Signal Board Layout Considerations

When laying out the system board, use the following guidelines:

- Keep clock and timing signal paths as short as possible and route with 50 Ω impedance.
- Use a serial termination resistor placed close to the clock buffer to minimize signal reflection. Use the following equation to compute the resistor value:

  Rterm = Rim – Rbuf

  where Rim = trace characteristic impedance

  Rbuf = clock buffer internal impedance.

3.4 SGMII AC-Coupled Serial Link Connection Example

Figure 39 shows an example of a 4-wire AC-coupled serial link connection. For additional layout suggestions, see *AN3556 MSC815x High Speed Serial Interface Hardware Design Considerations*, available on the Freescale website or from your local sales office or representative.

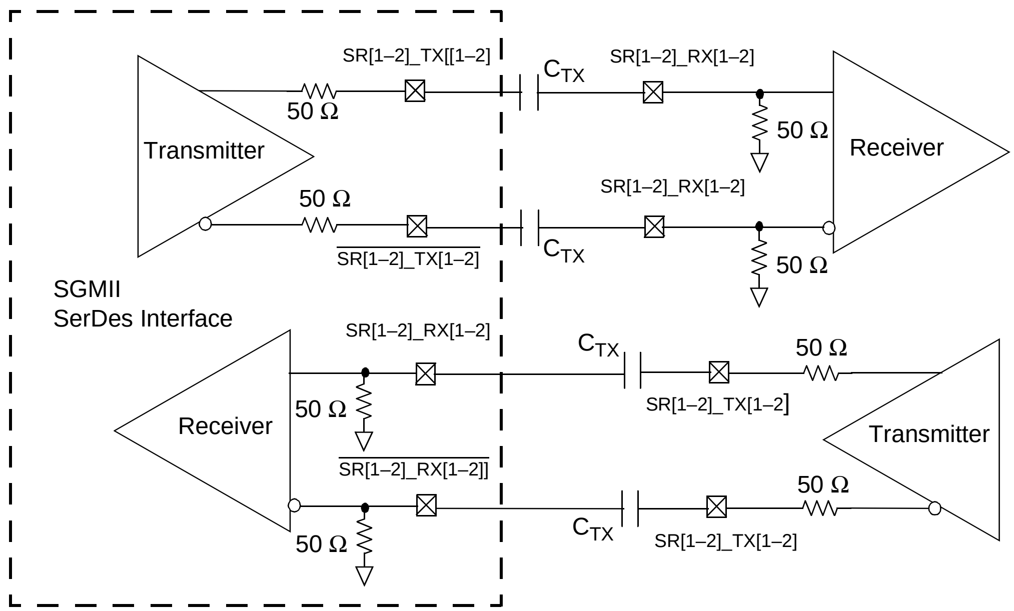


Figure 39. 4-Wire AC-Coupled SGMII Serial Link Connection Example

3.5.1 DDR Memory Related Pins

This section discusses the various scenarios that can be used with either of the MSC8251 DDR ports.

Note: The signal names in Table 40, Table 41 and Table 42 are generic names for a DDR SDRAM interface. For actual pin names refer to Table 1.

3.5.1.1 DDR Interface Is Not Used

Table 40. Connectivity of DDR Related Pins When the DDR Interface Is Not Used

Signal Name	Pin Connection
MDQ[0–63]	NC
MDQS[7–0]	NC
$\overline{\text{MDQS[7–0]}}$	NC
MA[15–0]	NC
MCK[0–2]	NC
$\overline{\text{MCK[0–2]}}$	NC
$\overline{\text{MCS[1–0]}}$	NC
MDM[7–0]	NC
MBA[2–0]	NC
$\overline{\text{MCAS}}$	NC
MCKE[1–0]	NC
MODT[1–0]	NC
MMDIC[1–0]	NC
$\overline{\text{MRAS}}$	NC
$\overline{\text{MWE}}$	NC
MECC[7–0]	NC
MDM8	NC
MDQS8	NC
$\overline{\text{MDQS8}}$	NC
MAPAR_OUT	NC
$\overline{\text{MAPAR\_IN}}$	NC
MVREF[3]	NC
GVDD1/GVDD2[3]	NC

Notes:
1. For the signals listed in this table, the initial M stands for M1 or M2 depending on which DDR controller is not used.
2. If the DDR controller is not used, disable the internal DDR clock by setting the appropriate bit in the System Clock Control Register (SCCR) and put all DDR I/O in sleep mode by setting DRx_GCR[DDRx_DOZE] (for DDR controller x). See the *Clocks* and *General Configuration Registers* chapters in the ***MSC8251 Reference Manual*** for details.
3. For MSC8251 Revision 1 silicon, these pins were connected to GND. For newer revisions of the MSC8251, connecting these pins to GND increases device power consumption.