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Details

Product Status	Obsolete
Core Processor	ST7
Core Size	8-Bit
Speed	8MHz
Connectivity	I ² C, SCI, SPI
Peripherals	LVD, POR, PWM, WDT
Number of I/O	48
Program Memory Size	48KB (48K x 8)
Program Memory Type	FLASH
EEPROM Size	-
RAM Size	1.5K x 8
Voltage - Supply (Vcc/Vdd)	3.8V ~ 5.5V
Data Converters	A/D 16x10b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	64-TQFP
Supplier Device Package	-
Purchase URL	https://www.e-xfl.com/product-detail/stmicroelectronics/st72f321ar7t6

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PIN DESCRIPTION (Cont'd)

For external pin connection guidelines, refer to See “ELECTRICAL CHARACTERISTICS” on page 138.

Legend / Abbreviations for Table 2 :

Type: I = input, O = output, S = supply

Input level: A = Dedicated analog input

In/Output level: C = CMOS 0.3V_{DD}/0.7V_{DD}
 C_T= CMOS 0.3V_{DD}/0.7V_{DD} with input trigger
 T_T= TTL 0.8V / 2V with Schmitt trigger

Output level: HS = 20mA high sink (on N-buffer only)

Port and control configuration:

- Input: float = floating, wpu = weak pull-up, int = interrupt ¹⁾, ana = analog
- Output: OD = open drain ²⁾, PP = push-pull

Refer to “I/O PORTS” on page 46 for more details on the software configuration of the I/O ports.

The RESET configuration of each pin is shown in bold. This configuration is valid as long as the device is in reset state.

Table 2. Device Pin Description

Pin n°			Pin Name	Type	Level		Port						Main function (after reset)	Alternate function
LQFP64	LQFP44	LQFP32			Input	Output	Input				Output			
							float	wpu	int	ana	OD	PP		
1	-	-	PE4 (HS)	I/O	C _T	HS	X	X			X	X	Port E4	
2	-	-	PE5 (HS)	I/O	C _T	HS	X	X			X	X	Port E5	
3	-	-	PE6 (HS)	I/O	C _T	HS	X	X			X	X	Port E6	
4	-	-	PE7 (HS)	I/O	C _T	HS	X	X			X	X	Port E7	
5	2	28	PB0/PWM3	I/O	C _T		X		ei2		X	X	Port B0	PWM Output 3
6	3	-	PB1/PWM2	I/O	C _T		X		ei2		X	X	Port B1	PWM Output 2
7	4	-	PB2/PWM1	I/O	C _T		X		ei2		X	X	Port B2	PWM Output 1
8	5	29	PB3/PWM0	I/O	C _T		X		ei2		X	X	Port B3	PWM Output 0
9	6	30	PB4 (HS)/ARTCLK	I/O	C _T	HS	X		ei3		X	X	Port B4	PWM-ART External Clock
10	-	-	PB5 / ARTIC1	I/O	C _T		X		ei3		X	X	Port B5	PWM-ART Input Capture 1
11	-	-	PB6 / ARTIC2	I/O	C _T		X		ei3		X	X	Port B6	PWM-ART Input Capture 2
12	-	-	PB7	I/O	C _T		X		ei3		X	X	Port B7	
13	7	31	PD0/AIN0	I/O	C _T		X	X		X	X	X	Port D0	ADC Analog Input 0
14	8	32	PD1/AIN1	I/O	C _T		X	X		X	X	X	Port D1	ADC Analog Input 1
15	9	-	PD2/AIN2	I/O	C _T		X	X		X	X	X	Port D2	ADC Analog Input 2
16	10	-	PD3/AIN3	I/O	C _T		X	X		X	X	X	Port D3	ADC Analog Input 3
17	11	-	PD4/AIN4	I/O	C _T		X	X		X	X	X	Port D4	ADC Analog Input 4
18	12	-	PD5/AIN5	I/O	C _T		X	X		X	X	X	Port D5	ADC Analog Input 5
19	-	-	PD6/AIN6	I/O	C _T		X	X		X	X	X	Port D6	ADC Analog Input 6
20	-	-	PD7/AIN7	I/O	C _T		X	X		X	X	X	Port D7	ADC Analog Input 7
21	13	1	V _{AREF}	I									Analog Reference Voltage for ADC	
22	14	2	V _{SSA}	S									Analog Ground Voltage	

ISTICS for more details.

3. OSC1 and OSC2 pins connect a crystal/ceramic resonator, or an external source to the on-chip oscillator; see Section 1 DESCRIPTION and Section 12.5 CLOCK AND TIMING CHARACTERISTICS for more details.

4. On the chip, each I/O port may have up to 8 pads:

– ads that are not bonded to external pins are forced by hardware in input pull-up configuration after reset. The configuration of these pads must be kept at reset state to avoid added current consumption.

5. Pull-up always activated on PE2 see limitation Section 15.4.6.

6. It is mandatory to connect all available V_{DD} and V_{REF} pins to the supply voltage and all V_{SS} and V_{SSA} pins to ground.

FLASH PROGRAM MEMORY (Cont'd)

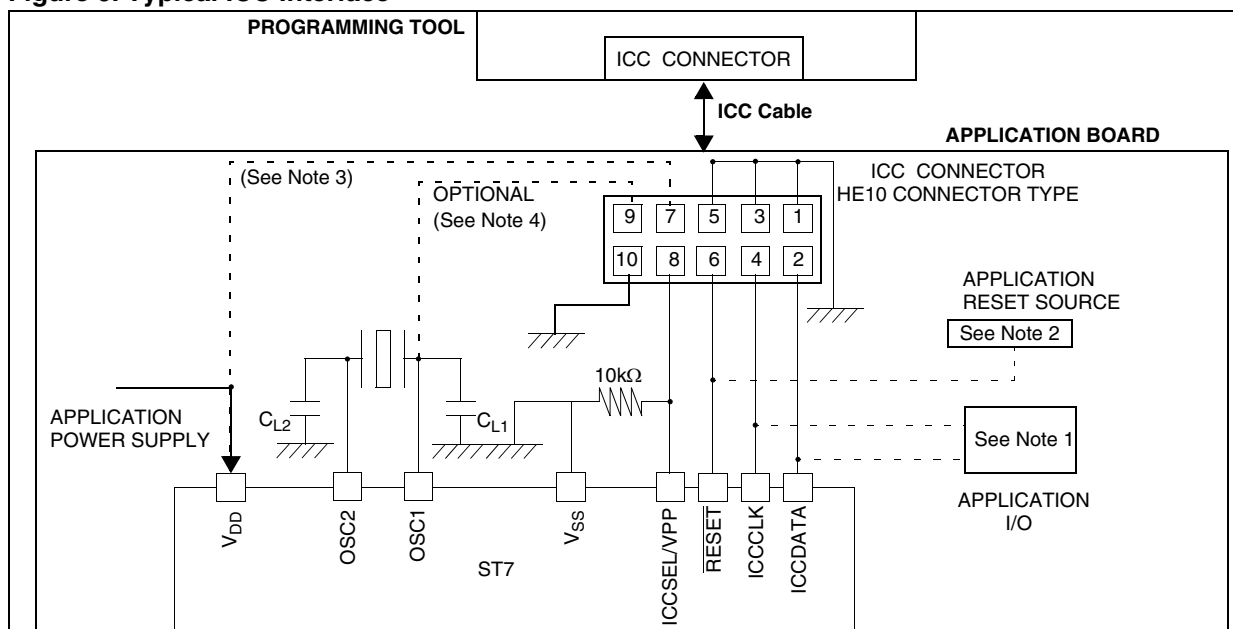
4.4 ICC Interface

ICC needs a minimum of 4 and up to 6 pins to be connected to the programming tool (see Figure 6). These pins are:

- $\overline{\text{RESET}}$: device reset
- V_{SS} : device power supply ground

- ICCCLK: ICC output serial clock pin
- ICCDATA: ICC input/output serial data pin
- ICCSEL/ V_{PP} : programming voltage
- OSC1(or OSCIN): main clock input for external source (optional)
- V_{DD} : application board power supply (optional, see Figure 6, Note 3)

Figure 6. Typical ICC Interface



Notes:

1. If the ICCCLK or ICCDATA pins are only used as outputs in the application, no signal isolation is necessary. As soon as the Programming Tool is plugged to the board, even if an ICC session is not in progress, the ICCCLK and ICCDATA pins are not available for the application. If they are used as inputs by the application, isolation such as a serial resistor has to be implemented in case another device forces the signal. Refer to the Programming Tool documentation for recommended resistor values.

2. During the ICC session, the programming tool must control the $\overline{\text{RESET}}$ pin. This can lead to conflicts between the programming tool and the application reset circuit if it drives more than 5mA at high level (push pull output or pull-up resistor $< 1K$). A schottky diode can be used to isolate the application RESET circuit in this case. When using a classical RC network with $R > 1K$ or a reset man-

agement IC with open drain output and pull-up resistor $> 1K$, no additional components are needed. In all cases the user must ensure that no external reset is generated by the application during the ICC session.

3. The use of Pin 7 of the ICC connector depends on the Programming Tool architecture. This pin must be connected when using most ST Programming Tools (it is used to monitor the application power supply). Please refer to the Programming Tool manual.

4. Pin 9 has to be connected to the OSC1 or OSCIN pin of the ST7 when the clock is not available in the application or if the selected clock option is not programmed in the option byte. ST7 devices with multi-oscillator capability need to have OSC2 grounded in this case.

SYSTEM INTEGRITY MANAGEMENT (Cont'd)

6.4.2 Auxiliary Voltage Detector (AVD)

The Voltage Detector function (AVD) is based on an analog comparison between a $V_{IT-(AVD)}$ and $V_{IT+(AVD)}$ reference value and the V_{DD} main supply or the external EVD pin voltage level (V_{EVD}). The V_{IT-} reference value for falling voltage is lower than the V_{IT+} reference value for rising voltage in order to avoid parasitic detection (hysteresis).

The output of the AVD comparator is directly readable by the application software through a real time status bit (AVDF) in the SICSR register. This bit is read only.

Caution: The AVD function is active only if the LVD is enabled through the option byte.

6.4.2.1 Monitoring the V_{DD} Main Supply

This mode is selected by clearing the AVDS bit in the SICSR register.

The AVD voltage threshold value is relative to the selected LVD threshold configured by option byte (see section 14.1 on page 175).

If the AVD interrupt is enabled, an interrupt is generated when the voltage crosses the $V_{IT+(AVD)}$ or $V_{IT-(AVD)}$ threshold (AVDF bit toggles).

In the case of a drop in voltage, the AVD interrupt acts as an early warning, allowing software to shut down safely before the LVD resets the microcontroller. See Figure 16.

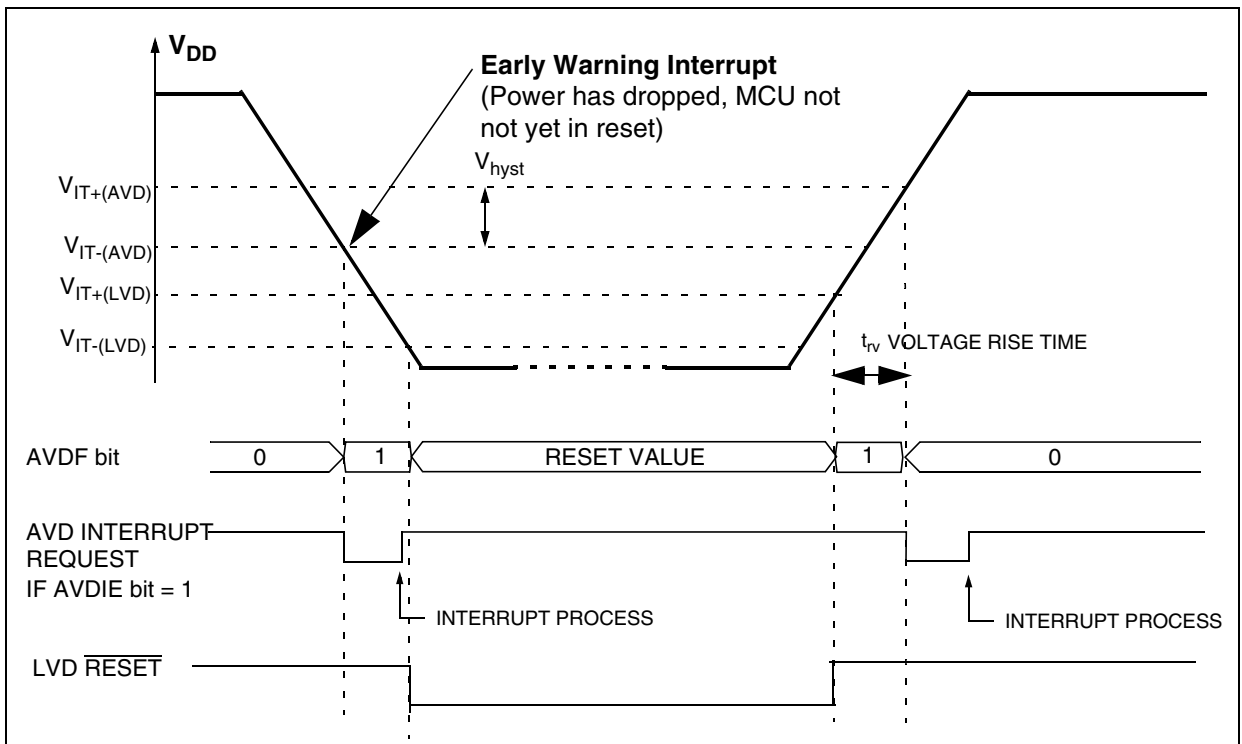
The interrupt on the rising edge is used to inform the application that the V_{DD} warning state is over.

If the voltage rise time t_{rv} is less than 256 or 4096 CPU cycles (depending on the reset delay selected by option byte), no AVD interrupt will be generated when $V_{IT+(AVD)}$ is reached.

If t_{rv} is greater than 256 or 4096 cycles then:

- If the AVD interrupt is enabled before the $V_{IT+(AVD)}$ threshold is reached, then 2 AVD interrupts will be received: the first when the AVDIE bit is set, and the second when the threshold is reached.
- If the AVD interrupt is enabled after the $V_{IT+(AVD)}$ threshold is reached then only one AVD interrupt will occur.

Figure 16. Using the AVD to Monitor V_{DD} (AVDS bit=0)



INTERRUPTS (Cont'd)

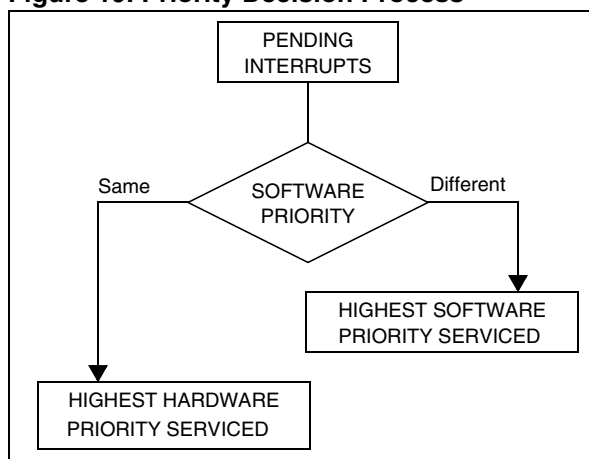
Servicing Pending Interrupts

As several interrupts can be pending at the same time, the interrupt to be taken into account is determined by the following two-step process:

- the highest software priority interrupt is serviced,
- if several interrupts have the same software priority then the interrupt with the highest hardware priority is serviced first.

Figure 19 describes this decision process.

Figure 19. Priority Decision Process



When an interrupt request is not serviced immediately, it is latched and then processed when its software priority combined with the hardware priority becomes the highest one.

Note 1: The hardware priority is exclusive while the software one is not. This allows the previous process to succeed with only one interrupt.

Note 2: TLI, RESET and TRAP can be considered as having the highest software priority in the decision process.

Different Interrupt Vector Sources

Two interrupt source types are managed by the ST7 interrupt controller: the non-maskable type (RESET, TRAP) and the maskable type (external or from internal peripherals).

Non-Maskable Sources

These sources are processed regardless of the state of the I1 and I0 bits of the CC register (see Figure 18). After stacking the PC, X, A and CC registers (except for RESET), the corresponding vector is loaded in the PC register and the I1 and I0 bits of the CC are set to disable interrupts (level 3). These sources allow the processor to exit HALT mode.

■ TRAP (Non Maskable Software Interrupt)

This software interrupt is serviced when the TRAP instruction is executed. It will be serviced according to the flowchart in Figure 18.

Caution: TRAP can be interrupted by a TLI.

■ RESET

The RESET source has the highest priority in the ST7. This means that the first current routine has the highest software priority (level 3) and the highest hardware priority.

See the RESET chapter for more details.

Maskable Sources

Maskable interrupt vector sources can be serviced if the corresponding interrupt is enabled and if its own interrupt software priority (in ISPRx registers) is higher than the one currently being serviced (I1 and I0 in CC register). If any of these two conditions is false, the interrupt is latched and thus remains pending.

■ TLI (Top Level Hardware Interrupt)

This hardware interrupt occurs when a specific edge is detected on the dedicated TLI pin. It will be serviced according to the flowchart in Figure 18 as a trap.

Caution: A TRAP instruction must not be used in a TLI service routine.

■ External Interrupts

External interrupts allow the processor to exit from HALT low power mode. External interrupt sensitivity is software selectable through the External Interrupt Control register (EICR).

External interrupt triggered on edge will be latched and the interrupt request automatically cleared upon entering the interrupt service routine.

If several input pins of a group connected to the same interrupt line are selected simultaneously, these will be logically ORed.

■ Peripheral Interrupts

Usually the peripheral interrupts cause the MCU to exit from HALT mode except those mentioned in the "Interrupt Mapping" table. A peripheral interrupt occurs when a specific flag is set in the peripheral status registers and if the corresponding enable bit is set in the peripheral control register. The general sequence for clearing an interrupt is based on an access to the status register followed by a read or write to an associated register.

Note: The clearing sequence resets the internal latch. A pending interrupt (i.e. waiting for being serviced) will therefore be lost if the clear sequence is executed.

POWER SAVING MODES (Cont'd)

8.4 ACTIVE-HALT AND HALT MODES

ACTIVE-HALT and HALT modes are the two lowest power consumption modes of the MCU. They are both entered by executing the 'HALT' instruction. The decision to enter either in ACTIVE-HALT or HALT mode is given by the MCC/RTC interrupt enable flag (OIE bit in MCCR register).

MCCR OIE bit	Power Saving Mode entered when HALT instruction is executed
0	HALT mode
1	ACTIVE-HALT mode

8.4.1 ACTIVE-HALT MODE

ACTIVE-HALT mode is the lowest power consumption mode of the MCU with a real time clock available. It is entered by executing the 'HALT' instruction when the OIE bit of the Main Clock Controller Status register (MCCR) is set (see section 10.2 on page 57 for more details on the MCCR register).

The MCU can exit ACTIVE-HALT mode on reception of an MCC/RTC interrupt or a RESET. When exiting ACTIVE-HALT mode by means of an interrupt, no 256 or 4096 CPU cycle delay occurs. The CPU resumes operation by servicing the interrupt or by fetching the reset vector which woke it up (see Figure 27).

When entering ACTIVE-HALT mode, the I[1:0] bits in the CC register are forced to '10b' to enable interrupts. Therefore, if an interrupt is pending, the MCU wakes up immediately.

In ACTIVE-HALT mode, only the main oscillator and its associated counter (MCC/RTC) are running to keep a wake-up time base. All other peripherals are not clocked except those which get their clock supply from another clock generator (such as external or auxiliary oscillator).

The safeguard against staying locked in ACTIVE-HALT mode is provided by the oscillator interrupt.

Note: As soon as the interrupt capability of one of the oscillators is selected (MCCR.OIE bit set), entering ACTIVE-HALT mode while the Watchdog is active does not generate a RESET.

This means that the device cannot spend more than a defined delay in this power saving mode.

CAUTION: When exiting ACTIVE-HALT mode following an MCC/RTC interrupt, OIE bit of MCCR register must not be cleared before t_{DELAY} after the interrupt occurs ($t_{\text{DELAY}} = 256 \text{ or } 4096 t_{\text{CPU}}$ de-

lay depending on option byte). Otherwise, the ST7 enters HALT mode for the remaining t_{DELAY} period.

Figure 26. ACTIVE-HALT Timing Overview

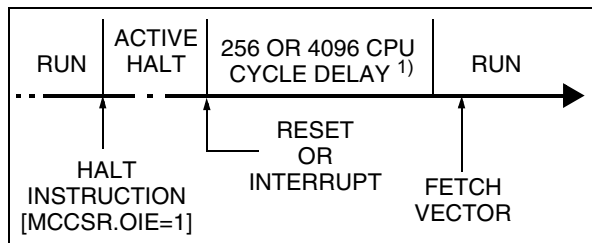
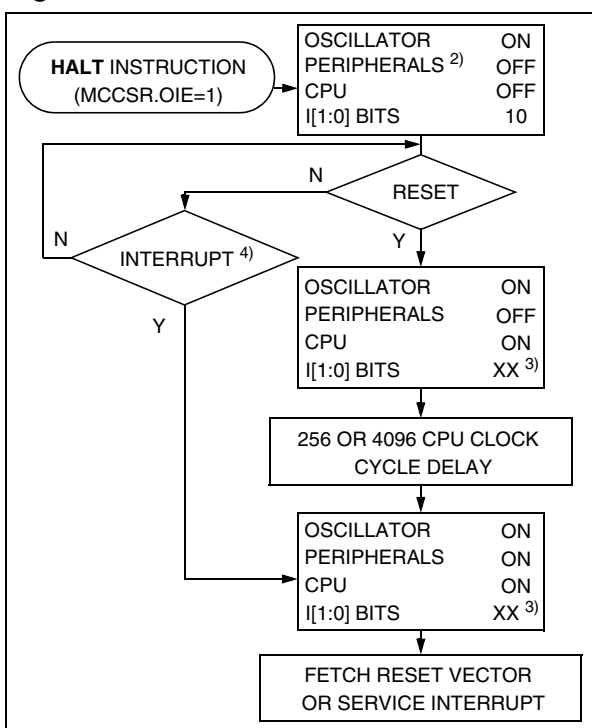


Figure 27. ACTIVE-HALT Mode Flow-chart

**Notes:**

1. This delay occurs only if the MCU exits ACTIVE-HALT mode by means of a RESET.
2. Peripheral clocked with an external clock source can still be active.
3. Before servicing an interrupt, the CC register is pushed on the stack. The I[1:0] bits of the CC register are set to the current software priority level of the interrupt routine and restored when the CC register is popped.
4. Only the MCC/RTC interrupt can exit the MCU from ACTIVE-HALT mode.

MAIN CLOCK CONTROLLER WITH REAL TIME CLOCK (Cont'd)

10.2.5 Low Power Modes

Mode	Description
WAIT	No effect on MCC/RTC peripheral. MCC/RTC interrupt cause the device to exit from WAIT mode.
ACTIVE-HALT	No effect on MCC/RTC counter (OIE bit is set), the registers are frozen. MCC/RTC interrupt cause the device to exit from ACTIVE-HALT mode.
HALT	MCC/RTC counter and registers are frozen. MCC/RTC operation resumes when the MCU is woken up by an interrupt with "exit from HALT" capability.

10.2.6 Interrupts

The MCC/RTC interrupt event generates an interrupt if the OIE bit of the MCCR register is set and the interrupt mask in the CC register is not active (RIM instruction).

Interrupt Event	Event Flag	Enable Control Bit	Exit from Wait	Exit from Halt
Time base overflow event	OIF	OIE	Yes	No ¹⁾

Note:

The MCC/RTC interrupt wakes up the MCU from ACTIVE-HALT mode, not from HALT mode.

10.2.7 Register Description

MCC CONTROL/STATUS REGISTER (MCCR)

Read/Write

Reset Value: 0000 0000 (00h)

7

0

MCO	CP1	CP0	SMS	TB1	TB0	OIE	OIF
-----	-----	-----	-----	-----	-----	-----	-----

Bit 7 = **MCO** Main clock out selection

This bit enables the MCO alternate function on the PF0 I/O port. It is set and cleared by software.

0: MCO alternate function disabled (I/O pin free for general-purpose I/O)

1: MCO alternate function enabled (f_{CPU} on I/O port)

Note: To reduce power consumption, the MCO function is not active in ACTIVE-HALT mode.

Bit 6:5 = **CP[1:0]** CPU clock prescaler

These bits select the CPU clock prescaler which is applied in the different slow modes. Their action is conditioned by the setting of the SMS bit. These two bits are set and cleared by software

f_{CPU} in SLOW mode	CP1	CP0
$f_{OSC2} / 2$	0	0
$f_{OSC2} / 4$	0	1
$f_{OSC2} / 8$	1	0
$f_{OSC2} / 16$	1	1

Bit 4 = **SMS** Slow mode select

This bit is set and cleared by software.

0: Normal mode. $f_{CPU} = f_{OSC2}$

1: Slow mode. f_{CPU} is given by CP1, CP0

See Section 8.2 SLOW MODE and Section 10.2 MAIN CLOCK CONTROLLER WITH REAL TIME CLOCK AND BEEPER (MCC/RTC) for more details.

Bit 3:2 = **TB[1:0]** Time base control

These bits select the programmable divider time base. They are set and cleared by software.

Counter Prescaler	Time Base		TB1	TB0
	$f_{OSC2} = 4\text{MHz}$	$f_{OSC2} = 8\text{MHz}$		
16000	4ms	2ms	0	0
32000	8ms	4ms	0	1
80000	20ms	10ms	1	0
200000	50ms	25ms	1	1

A modification of the time base is taken into account at the end of the current period (previously set) to avoid an unwanted time shift. This allows to use this time base as a real time clock.

Bit 1 = **OIE** Oscillator interrupt enable

This bit set and cleared by software.

0: Oscillator interrupt disabled

1: Oscillator interrupt enabled

This interrupt can be used to exit from ACTIVE-HALT mode.

When this bit is set, calling the ST7 software HALT instruction enters the ACTIVE-HALT power saving mode.

16-BIT TIMER (Cont'd)

Figure 43. Counter Timing Diagram, Internal Clock Divided by 2

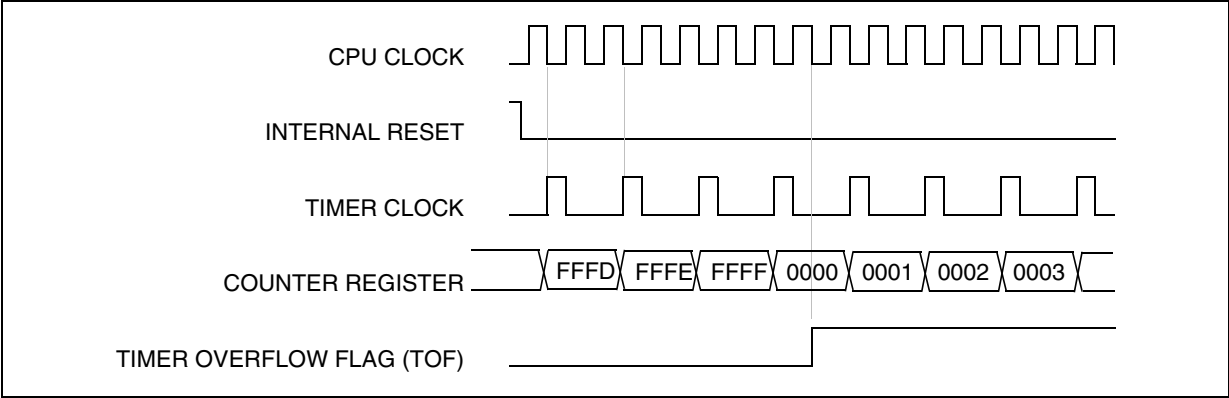


Figure 44. Counter Timing Diagram, Internal Clock Divided by 4

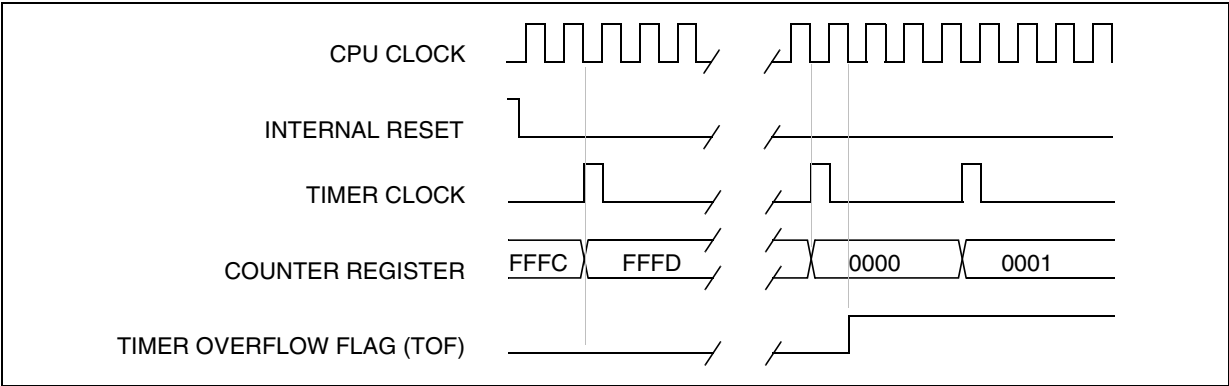
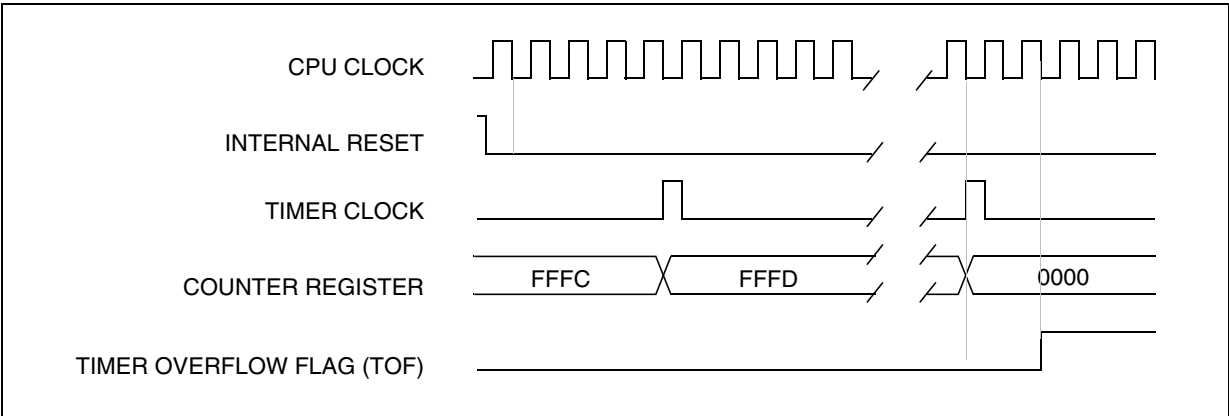


Figure 45. Counter Timing Diagram, Internal Clock Divided By 8



Note: The MCU is in reset state when the internal reset signal is high, when it is low the MCU is running.

SERIAL PERIPHERAL INTERFACE (Cont'd)**10.5.4 Clock Phase and Clock Polarity**

Four possible timing relationships may be chosen by software, using the CPOL and CPHA bits (See Figure 57).

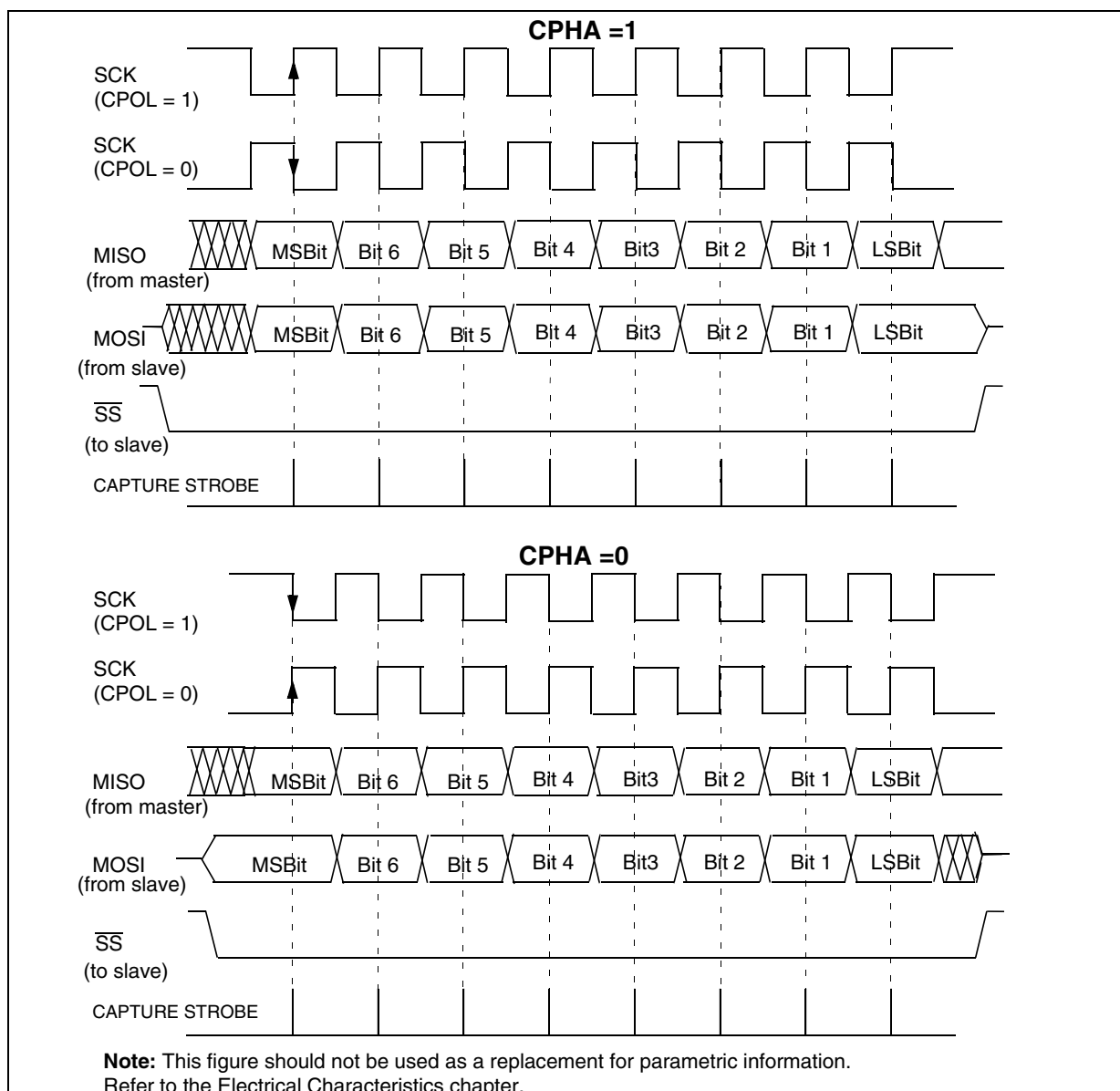
Note: The idle state of SCK must correspond to the polarity selected in the SPICSR register (by pulling up SCK if CPOL=1 or pulling down SCK if CPOL=0).

The combination of the CPOL clock polarity and CPHA (clock phase) bits selects the data capture clock edge

Figure 57, shows an SPI transfer with the four combinations of the CPHA and CPOL bits. The diagram may be interpreted as a master or slave timing diagram where the SCK pin, the MISO pin, the MOSI pin are directly connected between the master and the slave device.

Note: If CPOL is changed at the communication byte boundaries, the SPI must be disabled by re-setting the SPE bit.

Figure 57. Data Clock Timing Diagram



SERIAL COMMUNICATIONS INTERFACE (Cont'd)**10.6.7 Register Description****STATUS REGISTER (SCISR)**

Read Only

Reset Value: 1100 0000 (C0h)

7							0
TDRE	TC	RDRF	IDLE	OR	NF	FE	PE

Bit 7 = TDRE *Transmit data register empty.*

This bit is set by hardware when the content of the TDR register has been transferred into the shift register. An interrupt is generated if the TIE bit = 1 in the SCICR2 register. It is cleared by a software sequence (an access to the SCISR register followed by a write to the SCIDR register).

0: Data is not transferred to the shift register

1: Data is transferred to the shift register

Note: Data is not transferred to the shift register unless the TDRE bit is cleared.

Bit 6 = TC *Transmission complete.*

This bit is set by hardware when transmission of a frame containing Data is complete. An interrupt is generated if TCIE = 1 in the SCICR2 register. It is cleared by a software sequence (an access to the SCISR register followed by a write to the SCIDR register).

0: Transmission is not complete

1: Transmission is complete

Note: TC is not set after the transmission of a Preamble or a Break.

Bit 5 = RDRF *Received data ready flag.*

This bit is set by hardware when the content of the RDR register has been transferred to the SCIDR register. An interrupt is generated if RIE = 1 in the SCICR2 register. It is cleared by a software sequence (an access to the SCISR register followed by a read to the SCIDR register).

0: Data is not received

1: Received data is ready to be read

Bit 4 = IDLE *Idle line detect.*

This bit is set by hardware when an Idle Line is detected. An interrupt is generated if the ILIE = 1 in the SCICR2 register. It is cleared by a software sequence (an access to the SCISR register followed by a read to the SCIDR register).

0: No Idle Line is detected

1: Idle Line is detected

Note: The IDLE bit is not set again until the RDRF bit has been set itself (that is, a new idle line occurs).

Bit 3 = OR *Overrun error.*

This bit is set by hardware when the word currently being received in the shift register is ready to be transferred into the RDR register while RDRF = 1. An interrupt is generated if RIE = 1 in the SCICR2 register. It is cleared by a software sequence (an access to the SCISR register followed by a read to the SCIDR register).

0: No Overrun error

1: Overrun error is detected

Note: When this bit is set RDR register content is not lost but the shift register is overwritten.

Bit 2 = NF *Noise flag.*

This bit is set by hardware when noise is detected on a received frame. It is cleared by a software sequence (an access to the SCISR register followed by a read to the SCIDR register).

0: No noise is detected

1: Noise is detected

Note: This bit does not generate interrupt as it appears at the same time as the RDRF bit which itself generates an interrupt.

Bit 1 = FE *Framing error.*

This bit is set by hardware when a de-synchronization, excessive noise or a break character is detected. It is cleared by a software sequence (an access to the SCISR register followed by a read to the SCIDR register).

0: No Framing error is detected

1: Framing error or break character is detected

Note: This bit does not generate interrupt as it appears at the same time as the RDRF bit which itself generates an interrupt. If the word currently being transferred causes both frame error and overrun error, it will be transferred and only the OR bit will be set.

Bit 0 = PE *Parity error.*

This bit is set by hardware when a parity error occurs in receiver mode. It is cleared by a software sequence (a read to the status register followed by an access to the SCIDR data register). An interrupt is generated if PIE = 1 in the SCICR1 register.

0: No parity error

1: Parity error

SERIAL COMMUNICATIONS INTERFACE (Cont'd)**CONTROL REGISTER 2 (SCICR2)**

Read/Write

Reset Value: 0000 0000 (00h)

7							0
TIE	TCIE	RIE	ILIE	TE	RE	RWU	SBK

Bit 7 = TIE *Transmitter interrupt enable.*

This bit is set and cleared by software.

0: Interrupt is inhibited

1: An SCI interrupt is generated whenever TDRE=1 in the SCISR register

Bit 6 = TCIE *Transmission complete interrupt enable*

This bit is set and cleared by software.

0: Interrupt is inhibited

1: An SCI interrupt is generated whenever TC=1 in the SCISR register

Bit 5 = RIE *Receiver interrupt enable.*

This bit is set and cleared by software.

0: Interrupt is inhibited

1: An SCI interrupt is generated whenever OR=1 or RDRF=1 in the SCISR register

Bit 4 = ILIE *Idle line interrupt enable.*

This bit is set and cleared by software.

0: Interrupt is inhibited

1: An SCI interrupt is generated whenever IDLE=1 in the SCISR register.

Bit 3 = TE *Transmitter enable.*

This bit enables the transmitter. It is set and cleared by software.

0: Transmitter is disabled

1: Transmitter is enabled

Notes:

– During transmission, a “0” pulse on the TE bit (“0” followed by “1”) sends a preamble (idle line) after the current word.

– When TE is set there is a 1 bit-time delay before the transmission starts.

CAUTION: The TDO pin is free for general purpose I/O only when the TE and RE bits are both cleared (or if TE is never set).

Bit 2 = RE *Receiver enable.*

This bit enables the receiver. It is set and cleared by software.

0: Receiver is disabled

1: Receiver is enabled and begins searching for a start bit

Bit 1 = RWU *Receiver wake-up.*

This bit determines if the SCI is in mute mode or not. It is set and cleared by software and can be cleared by hardware when a wake-up sequence is recognized.

0: Receiver in Active mode

1: Receiver in Mute mode

Note: Before selecting Mute mode (setting the RWU bit), the SCI must receive some data first, otherwise it cannot function in Mute mode with wake-up by idle line detection.

Bit 0 = SBK *Send break.*

This bit set is used to send break characters. It is set and cleared by software.

0: No break character is transmitted

1: Break characters are transmitted

Note: If the SBK bit is set to “1” and then to “0”, the transmitter sends a BREAK word at the end of the current word.

10-BIT A/D CONVERTER (ADC) (Cont'd)

10.8.3 Functional Description

The conversion is monotonic, meaning that the result never decreases if the analog input does not and never increases if the analog input does not.

If the input voltage (V_{AIN}) is greater than V_{AREF} (high-level voltage reference) then the conversion result is FFh in the ADCDRH register and 03h in the ADCDRL register (without overflow indication).

If the input voltage (V_{AIN}) is lower than V_{SSA} (low-level voltage reference) then the conversion result in the ADCDRH and ADCDRL registers is 00 00h.

The A/D converter is linear and the digital result of the conversion is stored in the ADCDRH and ADCDRL registers. The accuracy of the conversion is described in the Electrical Characteristics Section.

R_{AIN} is the maximum recommended impedance for an analog input signal. If the impedance is too high, this will result in a loss of accuracy due to leakage and sampling not being completed in the allotted time.

10.8.3.1 A/D Converter Configuration

The analog input ports must be configured as input, no pull-up, no interrupt. Refer to the «I/O ports» chapter. Using these pins as analog inputs does not affect the ability of the port to be read as a logic input.

In the ADCCSR register:

- Select the CS[3:0] bits to assign the analog channel to convert.

10.8.3.2 Starting the Conversion

In the ADCCSR register:

- Set the ADON bit to enable the A/D converter and to start the conversion. From this time on, the ADC performs a continuous conversion of the selected channel.

When a conversion is complete:

- The EOC bit is set by hardware.
- The result is in the ADCDR registers.

A read to the ADCDRH or a write to any bit of the ADCCSR register resets the EOC bit.

To read the 10 bits, perform the following steps:

1. Poll the EOC bit
2. Read the ADCDRL register
3. Read the ADCDRH register. This clears EOC automatically.

Note: The data is not latched, so both the low and the high data register must be read before the next conversion is complete, so it is recommended to disable interrupts while reading the conversion result.

To read only 8 bits, perform the following steps:

1. Poll the EOC bit
2. Read the ADCDRH register. This clears EOC automatically.

10.8.3.3 Changing the conversion channel

The application can change channels during conversion. When software modifies the CH[3:0] bits in the ADCCSR register, the current conversion is stopped, the EOC bit is cleared, and the A/D converter starts converting the newly selected channel.

10.8.4 Low Power Modes

Note: The A/D converter may be disabled by resetting the ADON bit. This feature allows reduced power consumption when no conversion is needed and between single shot conversions.

Mode	Description
WAIT	No effect on A/D Converter
HALT	A/D Converter disabled. After wakeup from Halt mode, the A/D Converter requires a stabilization time t_{STAB} (see Electrical Characteristics) before accurate conversions can be performed.

10.8.5 Interrupts

None.

INSTRUCTION SET OVERVIEW (Cont'd)**11.1.6 Indirect Indexed (Short, Long)**

This is a combination of indirect and short indexed addressing modes. The operand is referenced by its memory address, which is defined by the unsigned addition of an index register value (X or Y) with a pointer value located in memory. The pointer address follows the opcode.

The indirect indexed addressing mode consists of two submodes:

Indirect Indexed (Short)

The pointer address is a byte, the pointer size is a byte, thus allowing 00 - 1FE addressing space, and requires 1 byte after the opcode.

Indirect Indexed (Long)

The pointer address is a byte, the pointer size is a word, thus allowing 64 Kbyte addressing space, and requires 1 byte after the opcode.

Table 27. Instructions Supporting Direct, Indexed, Indirect and Indirect Indexed Addressing Modes

Long and Short Instructions	Function
LD	Load
CP	Compare
AND, OR, XOR	Logical Operations
ADC, ADD, SUB, SBC	Arithmetic Additions/Subtractions operations
BCP	Bit Compare

Short Instructions Only	Function
CLR	Clear
INC, DEC	Increment/Decrement
TNZ	Test Negative or Zero
CPL, NEG	1 or 2 Complement
BSET, BRES	Bit Operations
BTJT, BTJF	Bit Test and Jump Operations
SLL, SRL, SRA, RLC, RRC	Shift and Rotate Operations
SWAP	Swap Nibbles
CALL, JP	Call or Jump subroutine

11.1.7 Relative mode (Direct, Indirect)

This addressing mode is used to modify the PC register value, by adding an 8-bit signed offset to it.

Available Relative Direct/Indirect Instructions	Function
JRxx	Conditional Jump
CALLR	Call Relative

The relative addressing mode consists of two submodes:

Relative (Direct)

The offset is following the opcode.

Relative (Indirect)

The offset is defined in memory, which address follows the opcode.

INSTRUCTION SET OVERVIEW (Cont'd)

Mnemo	Description	Function/Example	Dst	Src	I1	H	I0	N	Z	C
JRULE	Jump if (C + Z = 1)	Unsigned <=								
LD	Load	dst <= src	reg, M	M, reg				N	Z	
MUL	Multiply	X,A = X * A	A, X, Y	X, Y, A		0				0
NEG	Negate (2's compl)	neg \$10	reg, M					N	Z	C
NOP	No Operation									
OR	OR operation	A = A + M	A	M				N	Z	
POP	Pop from the Stack	pop reg	reg	M						
		pop CC	CC	M	I1	H	I0	N	Z	C
PUSH	Push onto the Stack	push Y	M	reg, CC						
RCF	Reset carry flag	C = 0								0
RET	Subroutine Return									
RIM	Enable Interrupts	I1:0 = 10 (level 0)			1		0			
RLC	Rotate left true C	C <= A <= C	reg, M					N	Z	C
RRC	Rotate right true C	C => A => C	reg, M					N	Z	C
RSP	Reset Stack Pointer	S = Max allowed								
SBC	Subtract with Carry	A = A - M - C	A	M				N	Z	C
SCF	Set carry flag	C = 1								1
SIM	Disable Interrupts	I1:0 = 11 (level 3)			1		1			
SLA	Shift left Arithmetic	C <= A <= 0	reg, M					N	Z	C
SLL	Shift left Logic	C <= A <= 0	reg, M					N	Z	C
SRL	Shift right Logic	0 => A => C	reg, M					0	Z	C
SRA	Shift right Arithmetic	A7 => A => C	reg, M					N	Z	C
SUB	Subtraction	A = A - M	A	M				N	Z	C
SWAP	SWAP nibbles	A7-A4 <=> A3-A0	reg, M					N	Z	
TNZ	Test for Neg & Zero	tnz lbl1						N	Z	
TRAP	S/W trap	S/W interrupt			1		1			
WFI	Wait for Interrupt				1		0			
XOR	Exclusive OR	A = A XOR M	A	M				N	Z	

12.5 CLOCK AND TIMING CHARACTERISTICS

Subject to general operating conditions for V_{DD} , f_{CPU} , and T_A .

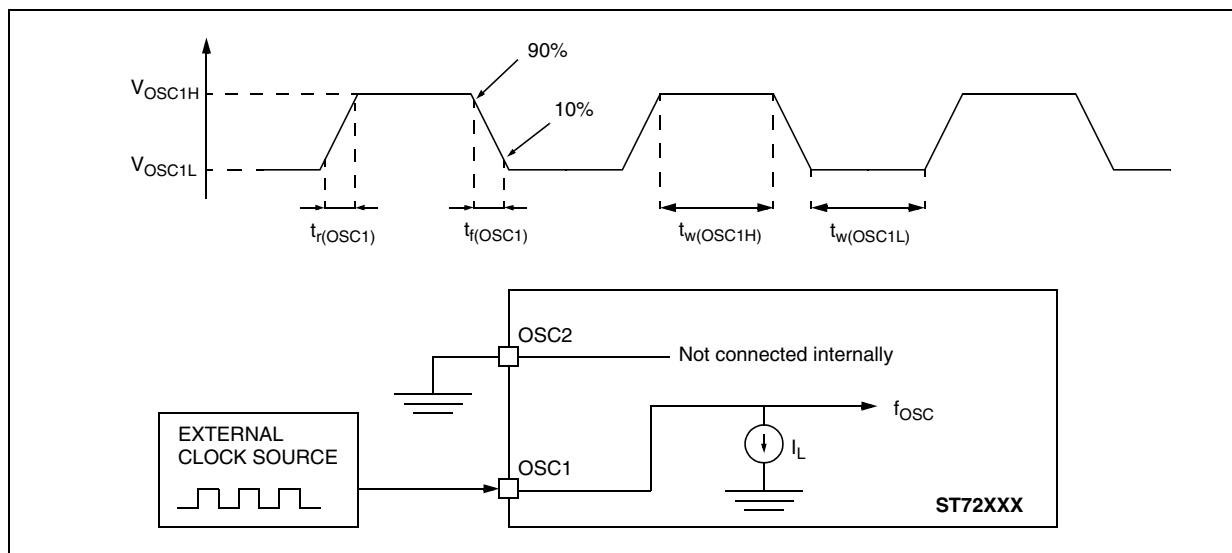
12.5.1 General Timings

Symbol	Parameter	Conditions	Min	Typ ¹⁾	Max	Unit
$t_{c(INST)}$	Instruction cycle time		2	3	12	t_{CPU}
		$f_{CPU}=8MHz$	250	375	1500	ns
$t_{v(IT)}$	Interrupt reaction time ²⁾ $t_{v(IT)} = \Delta t_{c(INST)} + 10$		10		22	t_{CPU}
		$f_{CPU}=8MHz$	1.25		2.75	μs

12.5.2 External Clock Source

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
V_{OSC1H}	OSC1 input pin high level voltage	see Figure 76	$0.7 \times V_{DD}$		V_{DD}	V
V_{OSC1L}	OSC1 input pin low level voltage		V_{SS}		$0.3 \times V_{DD}$	
$t_w(OSC1H)$ $t_w(OSC1L)$	OSC1 high or low time ³⁾		5			ns
$t_r(OSC1)$ $t_f(OSC1)$	OSC1 rise or fall time ³⁾				15	
I_L	OSC1 Input leakage current	$V_{SS} \leq V_{IN} \leq V_{DD}$			± 1	μA

Figure 76. Typical Application with an External Clock Source



Notes:

1. Data based on typical application software.
2. Time measured between interrupt event and interrupt vector fetch. $\Delta t_{c(INST)}$ is the number of t_{CPU} cycles needed to finish the current instruction execution.
3. Data based on design simulation and/or technology characteristics, not tested in production.

CLOCK AND TIMING CHARACTERISTICS (Cont'd)**12.5.3 Crystal and Ceramic Resonator Oscillators**

The ST7 internal clock can be supplied with four different Crystal/Ceramic resonator oscillators. All the information given in this paragraph is based on characterization results with specified typical external components. In the application, the resonator and the load capacitors have to be placed as

close as possible to the oscillator pins in order to minimize output distortion and start-up stabilization time. Refer to the crystal/ceramic resonator manufacturer for more details (frequency, package, accuracy...).

Symbol	Parameter	Conditions	Min	Max	Unit
f_{OSC}	Oscillator Frequency ¹⁾	LP: Low power oscillator MP: Medium power oscillator MS: Medium speed oscillator HS: High speed oscillator	1 >2 >4 >8	2 4 8 16	MHz
R_F	Feedback resistor ²⁾		20	40	k Ω
C_{L1} C_{L2}	Recommended load capacitance versus equivalent serial resistance of the crystal or ceramic resonator (R_S)	$R_S=200\Omega$ LP oscillator $R_S=200\Omega$ MP oscillator $R_S=200\Omega$ MS oscillator $R_S=100\Omega$ HS oscillator	22 22 18 15	56 46 33 33	pF

Symbol	Parameter	Conditions	Typ	Max	Unit
i_2	OSC2 driving current	$V_{DD}=5V$ $V_{IN}=V_{SS}$ LP oscillator MP oscillator MS oscillator HS oscillator	80 160 310 610	150 250 460 910	μA

Notes:

1. The oscillator selection can be optimized in terms of supply current using an high quality resonator with small R_S value. Refer to crystal/ceramic resonator manufacturer for more details.
2. Data based on characterisation results, not tested in production.

12.9 CONTROL PIN CHARACTERISTICS

12.9.1 Asynchronous $\overline{\text{RESET}}$ Pin

Subject to general operating conditions for V_{DD} , f_{CPU} , and T_A unless otherwise specified.

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
V_{IL}	Input low level voltage ¹⁾				$0.16 \times V_{DD}$	V
V_{IH}	Input high level voltage ¹⁾		$0.85 \times V_{DD}$			
V_{hys}	Schmitt trigger voltage hysteresis ²⁾			2.5		V
V_{OL}	Output low level voltage ³⁾	$V_{DD}=5V$ $I_{IO}=+2mA$		0.2	0.5	
I_{IO}	Input current on $\overline{\text{RESET}}$ pin			2		mA
R_{ON}	Weak pull-up equivalent resistor		20	30	120	k Ω
$t_{w(RSTL)out}$	Generated reset pulse duration	Stretch applied on external pulse	0		$42^{6)}$	μs
		Internal reset sources	20	30	$42^{6)}$	μs
$t_{h(RSTL)in}$	External reset pulse hold time ⁴⁾		2.5			μs
$t_{g(RSTL)in}$	Filtered glitch duration ⁵⁾			200		ns

Notes:

1. Data based on characterization results, not tested in production.
2. Hysteresis voltage between Schmitt trigger switching levels.
3. The I_{IO} current sunk must always respect the absolute maximum rating specified in Section 12.2.2 and the sum of I_{IO} (I/O ports and control pins) must not exceed I_{VSS} .
4. To guarantee the reset of the device, a minimum pulse has to be applied to the $\overline{\text{RESET}}$ pin. All short pulses applied on the $\overline{\text{RESET}}$ pin with a duration below $t_{h(RSTL)in}$ can be ignored.
5. The reset network (the resistor and two capacitors) protects the device against parasitic resets, especially in noisy environments.
6. Data guaranteed by design, not tested in production.

12.12 10-BIT ADC CHARACTERISTICS

Subject to general operating conditions for V_{DD} , f_{CPU} , and T_A unless otherwise specified.

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
f _{ADC}	ADC clock frequency		0.4		2	MHz
V _{AREF}	Analog reference voltage	0.7*V _{DD} ≤ V _{AREF} ≤ V _{DD}	3.8		V _{DD}	V
V _{AIN}	Conversion voltage range ¹⁾		V _{SSA}		V _{AREF}	
R _{AIN}	External input impedance				see	kΩ
C _{AIN}	External capacitor on analog input				Figure 96 and Figure 97	pF
f _{AIN}	Variation freq. of analog input signal					Hz
C _{ADC}	Internal sample and hold capacitor			12		pF
t _{ADC}	Conversion time (Sample+Hold) f _{CPU} =8MHz, SPEED=0 f _{ADC} =2MHz		7.5			μs
t _{ADC}	- No of sample capacitor loading cycles		4			1/f _{ADC}
	- No. of Hold conversion cycles		11			

Notes:

1. Any added external serial resistor will downgrade the ADC accuracy (especially for resistance greater than 10k Ω). Data based on characterization results, not tested in production.

2. Injecting negative current on any of the analog input pins significantly reduces the accuracy of any conversion being performed on any analog input. Analog pins can be protected against negative injection by adding a Schottky diode (pin to ground). Injecting negative current on digital input pins degrades ADC accuracy especially if performed on a pin close to the analog input pins. Any positive injection current within the limits specified for $I_{INJ(PIN)}$ and $\Sigma I_{INJ(PIN)}$ in Section 12.8 does not affect the ADC accuracy.