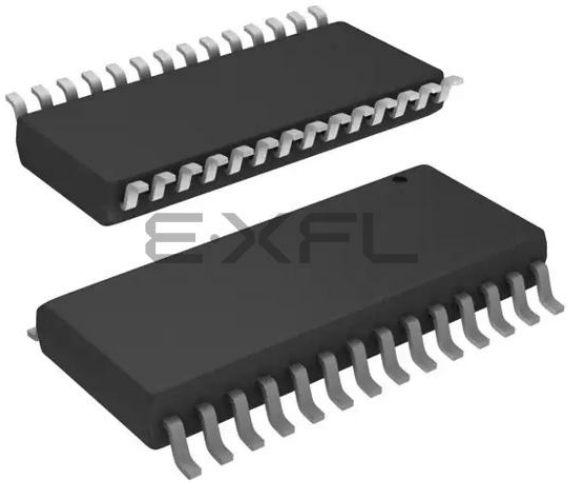

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Applications of "Embedded - Microcontrollers"

Details	
Product Status	Active
Core Processor	PIC
Core Size	8-Bit
Speed	48MHz
Connectivity	UART/USART, USB
Peripherals	Brown-out Detect/Reset, HLVD, POR, PWM, WDT
Number of I/O	23
Program Memory Size	16KB (8K x 16)
Program Memory Type	FLASH
EEPROM Size	-
RAM Size	768 x 8
Voltage - Supply (Vcc/Vdd)	4.2V ~ 5.5V
Data Converters	A/D 10x10b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	28-SOIC (0.295", 7.50mm Width)
Supplier Device Package	28-SOIC
Purchase URL	https://www.e-xfl.com/product-detail/microchip-technology/pic18f2450-i-so

REGISTER 2-1: OSCCON: OSCILLATOR CONTROL REGISTER

R/W-0	U-0	U-0	U-0	R(1)	U-0	R/W-0	R/W-0
IDLEN	—	—	—	OSTS	—	SCS1	SCS0
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'	
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7 **IDLEN:** Idle Enable bit
- 1 = Device enters Idle mode on SLEEP instruction
- 0 = Device enters Sleep mode on SLEEP instruction

bit 6-4 **Unimplemented:** Read as '0'

bit 3 **OSTS:** Oscillator Start-up Time-out Status bit(1)
- 1 = Oscillator Start-up Timer time-out has expired; primary oscillator is running
- 0 = Oscillator Start-up Timer time-out is running; primary oscillator is not ready

bit 2 **Unimplemented:** Read as '0'

bit 1-0 **SCS1:SCS0:** System Clock Select bits
- 1x = Internal oscillator
- 01 = Timer1 oscillator
- 00 = Primary oscillator

Note 1: Depends on the state of the IESO Configuration bit.

3.5.4 EXIT WITHOUT AN OSCILLATOR START-UP DELAY

Certain exits from power-managed modes do not invoke the OST at all. There are two cases:

- PRI_IDLE mode, where the primary clock source is not stopped; and
- The primary clock source is not any of the XT or HS modes

In these instances, the primary clock source either does not require an oscillator start-up delay, since it is already running (PRI_IDLE), or normally does not require an oscillator start-up delay (EC and any internal oscillator modes). However, a fixed delay of interval TCSD following the wake event is still required when leaving Sleep and Idle modes to allow the CPU to prepare for execution. Instruction execution resumes on the first clock cycle following this delay.

TABLE 3-2: EXIT DELAY ON WAKE-UP BY RESET FROM SLEEP MODE OR ANY IDLE MODE (BY CLOCK SOURCES)

Microcontroller Clock Source		Exit Delay	Clock Ready Status Bit (OSCCON)
Before Wake-up	**After Wake-up**		
Primary Device Clock (PRI_IDLE mode)	XT, HS	None	OSTS
	XTPLL, HSPLL		
	EC		
	INTRC(1)		
T1OSC or INTRC(1)	XT, HS	TOST(3)	OSTS
	XTPLL, HSPLL	TOST + t_{rc}(3)	
	EC	TCSD(2)	
	INTRC(1)	TIOBST(4)	
INTRC(1)	XT, HS	TOST(3)	OSTS
	XTPLL, HSPLL	TOST + t_{rc}(3)	
	EC	TCSD(2)	
	INTRC(1)	None	
None (Sleep mode)	XT, HS	TOST(3)	OSTS
	XTPLL, HSPLL	TOST + t_{rc}(3)	
	EC	TCSD(2)	
	INTRC(1)	TIOBST(4)	

Note 1: In this instance, refers specifically to the 31 kHz INTRC clock source.

2: TCSD (parameter 38, Table 21-10) is a required delay when waking from Sleep and all Idle modes and runs concurrently with any other required delays (see **Section 3.4 "Idle Modes"**).

3: TOST is the Oscillator Start-up Timer period (parameter 32, Table 21-10). t_{rc} is the PLL lock time-out (parameter F12, Table 21-7); it is also designated as TPLL.

4: Execution continues during TIOBST (parameter 39, Table 21-10), the INTRC stabilization period.

4.4 Brown-out Reset (BOR)

PIC18F2450/4450 devices implement a BOR circuit that provides the user with a number of configuration and power-saving options. The BOR is controlled by the BORV1:BORV0 and BOREN1:BOREN0 Configuration bits. There are a total of four BOR configurations which are summarized in Table 4-1.

The BOR threshold is set by the BORV1:BORV0 bits. If BOR is enabled (any values of BOREN1:BOREN0 except '00'), any drop of VDD below VBOR (parameter D005, **Section 269 "DC Characteristics: Supply Voltage"**) for greater than TBOR (parameter 35, Table 21-10) will reset the device. A Reset may or may not occur if VDD falls below VBOR for less than TBOR. The chip will remain in Brown-out Reset until VDD rises above VBOR.

If the Power-up Timer is enabled, it will be invoked after VDD rises above VBOR; it then will keep the chip in Reset for an additional time delay, TPWRT (parameter 33, Table 21-10). If VDD drops below VBOR while the Power-up Timer is running, the chip will go back into a Brown-out Reset and the Power-up Timer will be initialized. Once VDD rises above VBOR, the Power-up Timer will execute the additional time delay.

BOR and the Power-on Timer (PWRT) are independently configured. Enabling BOR Reset does not automatically enable the PWRT.

4.4.1 SOFTWARE ENABLED BOR

When BOREN1:BOREN0 = 01, the BOR can be enabled or disabled by the user in software. This is done with the control bit, SBOREN (RCON<6>). Setting SBOREN enables the BOR to function as previously described. Clearing SBOREN disables the BOR entirely. The SBOREN bit operates only in this mode; otherwise, it is read as '0'.

Placing the BOR under software control gives the user the additional flexibility of tailoring the application to its environment without having to reprogram the device to change BOR configuration. It also allows the user to tailor device power consumption in software by eliminating the incremental current that the BOR consumes. While the BOR current is typically very small, it may have some impact in low-power applications.

Note:	Even when BOR is under software control, the BOR Reset voltage level is still set by the BORV1:BORV0 Configuration bits. It cannot be changed in software.

4.4.2 DETECTING BOR

When Brown-out Reset is enabled, the $\overline{\text{BOR}}$ bit always resets to '0' on any Brown-out Reset or Power-on Reset event. This makes it difficult to determine if a Brown-out Reset event has occurred just by reading the state of $\overline{\text{BOR}}$ alone. A more reliable method is to simultaneously check the state of both $\overline{\text{POR}}$ and $\overline{\text{BOR}}$. This assumes that the $\overline{\text{POR}}$ bit is reset to '1' in software immediately after any Power-on Reset event. IF $\overline{\text{BOR}}$ is '0' while $\overline{\text{POR}}$ is '1', it can be reliably assumed that a Brown-out Reset event has occurred.

4.4.3 DISABLING BOR IN SLEEP MODE

When BOREN1:BOREN0 = 10, the BOR remains under hardware control and operates as previously described. Whenever the device enters Sleep mode, however, the BOR is automatically disabled. When the device returns to any other operating mode, BOR is automatically re-enabled.

This mode allows for applications to recover from brown-out situations, while actively executing code, when the device requires BOR protection the most. At the same time, it saves additional power in Sleep mode by eliminating the small incremental BOR current.

TABLE 4-1: BOR CONFIGURATIONS

BOR Configuration		Status of SBOREN (RCON<6>)	BOR Operation
BOREN1	BOREN0		
0	0	Unavailable	BOR disabled; must be enabled by reprogramming the Configuration bits.
0	1	Available	BOR enabled in software; operation controlled by SBOREN.
1	0	Unavailable	BOR enabled in hardware in Run and Idle modes, disabled during Sleep mode.
1	1	Unavailable	BOR enabled in hardware; must be disabled by reprogramming the Configuration bits.

5.3.5 SPECIAL FUNCTION REGISTERS

The Special Function Registers (SFRs) are registers used by the CPU and peripheral modules for controlling the desired operation of the device. These registers are implemented as static RAM in the data memory space. SFRs start at the top of data memory and extend downward to occupy the top segment of Bank 15, from F60h to FFFh. A list of these registers is given in Table 5-1 and Table 5-2.

The SFRs can be classified into two sets: those associated with the "core" device functionality (ALU, Resets and interrupts) and those related to the peripheral functions. The Reset and interrupt registers are described in their respective chapters, while the ALU's STATUS register is described later in this section. Registers related to the operation of a peripheral feature are described in the chapter for that peripheral.

The SFRs are typically distributed among the peripherals whose functions they control. Unused SFR locations are unimplemented and read as '0's.

TABLE 5-1: SPECIAL FUNCTION REGISTER MAP FOR PIC18F2450/4450 DEVICES

Address	Name	Address	Name	Address	Name	Address	Name	Address	Name
FFFh	TOSU	FDFh	INDF2(1)	FBFh	CCPR1H	F9Fh	IPR1	F7Fh	UEP15
FFEh	TOSH	FDEh	POSTINC2(1)	FBEh	CCPR1L	F9Eh	PIR1	F7Eh	UEP14
FFDh	TOSL	FDDh	POSTDEC2(1)	FBDh	CCP1CON	F9Dh	PIE1	F7Dh	UEP13
FFCh	STKPTR	FDCh	PREINC2(1)	FBCh	—(2)	F9Ch	—(2)	F7Ch	UEP12
FFBh	PCLATU	FDBh	PLUSW2(1)	FBBh	—(2)	F9Bh	—(2)	F7Bh	UEP11
FFAh	PCLATH	FDAh	FSR2H	FBAh	—(2)	F9Ah	—(2)	F7Ah	UEP10
FF9h	PCL	FD9h	FSR2L	FB9h	—(2)	F99h	—(2)	F79h	UEP9
FF8h	TBLPTRU	FD8h	STATUS	FB8h	BAUDCON	F98h	—(2)	F78h	UEP8
FF7h	TBLPTRH	FD7h	TMR0H	FB7h	—(2)	F97h	—(2)	F77h	UEP7
FF6h	TBLPTRL	FD6h	TMR0L	FB6h	—(2)	F96h	TRISE(3)	F76h	UEP6
FF5h	TABLAT	FD5h	T0CON	FB5h	—(2)	F95h	TRISD(3)	F75h	UEP5
FF4h	PRODH	FD4h	—(2)	FB4h	—(2)	F94h	TRISC	F74h	UEP4
FF3h	PRODL	FD3h	OSCCON	FB3h	—(2)	F93h	TRISB	F73h	UEP3
FF2h	INTCON	FD2h	HLVDCON	FB2h	—(2)	F92h	TRISA	F72h	UEP2
FF1h	INTCON2	FD1h	WDTCON	FB1h	—(2)	F91h	—(2)	F71h	UEP1
FF0h	INTCON3	FD0h	RCON	FB0h	SPBRGH	F90h	—(2)	F70h	UEP0
FEFh	INDF0(1)	FCFh	TMR1H	FAFh	SPBRG	F8Fh	—(2)	F6Fh	UCFG
FEEh	POSTINC0(1)	FCEh	TMR1L	FAEh	RCREG	F8Eh	—(2)	F6Eh	UADDR
FEDh	POSTDEC0(1)	FCDh	T1CON	FADh	TXREG	F8Dh	LATE(3)	F6Dh	UCON
FECh	PREINC0(1)	FCCh	TMR2	FACh	TXSTA	F8Ch	LATD(3)	F6Ch	USTAT
FEBh	PLUSW0(1)	FCBh	PR2	FABh	RCSTA	F8Bh	LATC	F6Bh	UEIE
FEAh	FSR0H	FCAh	T2CON	FAAh	—(2)	F8Ah	LATB	F6Ah	UEIR
FE9h	FSR0L	FC9h	—(2)	FA9h	—(2)	F89h	LATA	F69h	UIE
FE8h	WREG	FC8h	—(2)	FA8h	—(2)	F88h	—(2)	F68h	UIR
FE7h	INDF1(1)	FC7h	—(2)	FA7h	EECON2(1)	F87h	—(2)	F67h	UFRMH
FE6h	POSTINC1(1)	FC6h	—(2)	FA6h	EECON1	F86h	—(2)	F66h	UFRML
FE5h	POSTDEC1(1)	FC5h	—(2)	FA5h	—(2)	F85h	—(2)	F65h	—(2)
FE4h	PREINC1(1)	FC4h	ADRESH	FA4h	—(2)	F84h	PORTE	F64h	—(2)
FE3h	PLUSW1(1)	FC3h	ADRESL	FA3h	—(2)	F83h	PORTD(3)	F63h	—(2)
FE2h	FSR1H	FC2h	ADCON0	FA2h	IPR2	F82h	PORTC	F62h	—(2)
FE1h	FSR1L	FC1h	ADCON1	FA1h	PIR2	F81h	PORTB	F61h	—(2)
FE0h	BSR	FC0h	ADCON2	FA0h	PIE2	F80h	PORTA	F60h	—(2)

Note 1: Not a physical register.
2: Unimplemented registers are read as '0'.
3: These registers are implemented only on 40/44-pin devices.

6.5 Writing to Flash Program Memory

The minimum programming block is 8 words or 16 bytes. Word or byte programming is not supported.

Table writes are used internally to load the holding registers needed to program the Flash memory. There are 16 holding registers used by the table writes for programming.

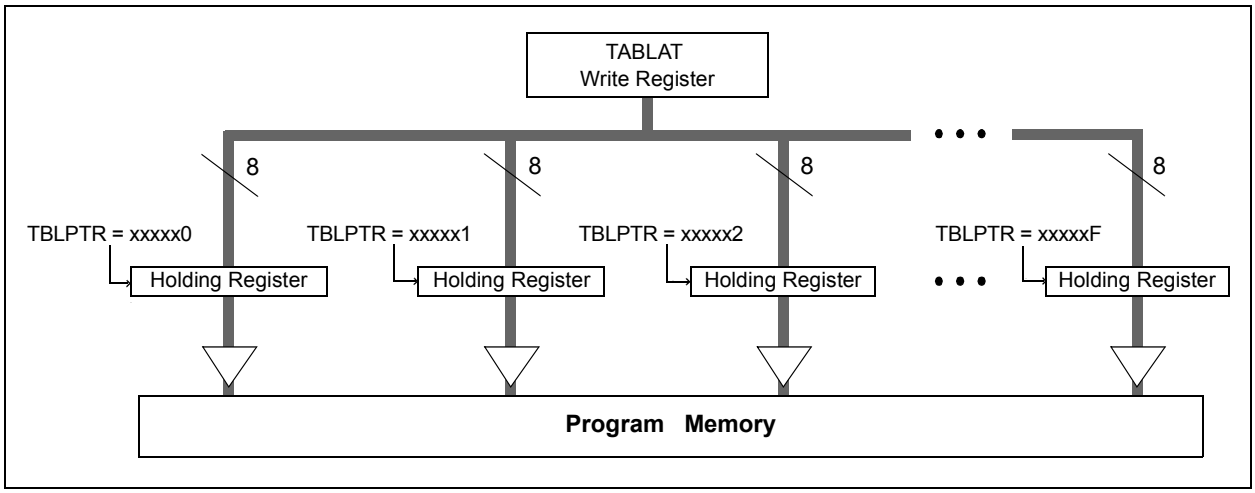
Since the Table Latch (TABLAT) is only a single byte, the TBLWT instruction may need to be executed 16 times for each programming operation. All of the table write operations will essentially be short writes because only the holding registers are written. At the end of updating the 16 holding registers, the EECON1 register must be written to in order to start the programming operation with a long write.

The long write is necessary for programming the internal Flash. Instruction execution is halted while in a long write cycle. The long write will be terminated by the internal programming timer.

The write/erase voltages are generated by an on-chip charge pump, rated to operate over the voltage range of the device.

Note:	The default value of the holding registers on device Resets and after write operations is FFh. A write of FFh to a holding register does not modify that byte. This means that individual bytes of program memory may be modified, provided that the change does not attempt to change any bit from a '0' to a '1'. When modifying individual bytes, it is not necessary to load all 16 holding registers before executing a write operation.

FIGURE 6-5: TABLE WRITES TO FLASH PROGRAM MEMORY



6.5.1 FLASH PROGRAM MEMORY WRITE SEQUENCE

The sequence of events for programming an internal program memory location should be:

1. Read 64 bytes into RAM.
2. Update data values in RAM as necessary.
3. Load Table Pointer register with address being erased.
4. Execute the Row Erase procedure.
5. Load Table Pointer register with address of first byte being written.
6. Write 16 bytes into the holding registers with auto-increment.
7. Set the EECON1 register for the write operation:
   - clear the CFGS bit to access program memory;
   - set WREN to enable byte writes.
8. Disable interrupts.
9. Write 55h to EECON2.
10. Write 0AAh to EECON2.
11. Set the WR bit. This will begin the write cycle.
12. The CPU will stall for duration of the write (about 2 ms using internal timer).
13. Re-enable interrupts.
14. Repeat steps 6 through 14 once more to write 64 bytes.
15. Verify the memory (table read).

This procedure will require about 8 ms to update one row of 64 bytes of memory. An example of the required code is given in Example 6-3.

Note:	Before setting the WR bit, the Table Pointer address needs to be within the intended address range of the 16 bytes in the holding register.

8.4 PIE Registers

The PIE registers contain the individual enable bits for the peripheral interrupts. Due to the number of peripheral interrupt sources, there are two Peripheral Interrupt Enable registers (PIE1 and PIE2). When IPEN = 0, the PEIE bit must be set to enable any of these peripheral interrupts.

REGISTER 8-6: PIE1: PERIPHERAL INTERRUPT ENABLE REGISTER 1

U-0	R/W-0	R/W-0	R/W-0	U-0	R/W-0	R/W-0	R/W-0
—	ADIE	RCIE	TXIE	—	CCP1IE	TMR2IE	TMR1IE
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'	
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7 **Unimplemented:** Read as '0'

bit 6 **ADIE:** A/D Converter Interrupt Enable bit
- 1 = Enables the A/D interrupt
- 0 = Disables the A/D interrupt

bit 5 **RCIE:** EUSART Receive Interrupt Enable bit
- 1 = Enables the EUSART receive interrupt
- 0 = Disables the EUSART receive interrupt

bit 4 **TXIE:** EUSART Transmit Interrupt Enable bit
- 1 = Enables the EUSART transmit interrupt
- 0 = Disables the EUSART transmit interrupt

bit 3 **Unimplemented:** Read as '0'

bit 2 **CCP1IE:** CCP1 Interrupt Enable bit
- 1 = Enables the CCP1 interrupt
- 0 = Disables the CCP1 interrupt

bit 1 **TMR2IE:** TMR2 to PR2 Match Interrupt Enable bit
- 1 = Enables the TMR2 to PR2 match interrupt
- 0 = Disables the TMR2 to PR2 match interrupt

bit 0 **TMR1IE:** TMR1 Overflow Interrupt Enable bit
- 1 = Enables the TMR1 overflow interrupt
- 0 = Disables the TMR1 overflow interrupt

When UOWN is set, the user can no longer depend on the values that were written to the BDs. From this point, the SIE updates the BDs as necessary, overwriting the original BD values. The BDnSTAT register is updated by the SIE with the token PID and the transfer count, BDnCNT, is updated.

The BDnSTAT byte of the BDT should always be the last byte updated when preparing to arm an endpoint. The SIE will clear the UOWN bit when a transaction has completed. The only exception to this is when KEN is enabled and/or BSTALL is enabled.

No hardware mechanism exists to block access when the UOWN bit is set. Thus, unexpected behavior can occur if the microcontroller attempts to modify memory when the SIE owns it. Similarly, reading such memory may produce inaccurate data until the USB peripheral returns ownership to the microcontroller.

14.4.1.2 BDnSTAT Register (CPU Mode)

When UOWN = 0, the microcontroller core owns the BD. At this point, the other seven bits of the register take on control functions.

The Data Toggle Sync Enable bit, DTSEN (BDnSTAT<3>), controls data toggle parity checking. Setting DTSEN enables data toggle synchronization by the SIE. When enabled, it checks the data packet's parity against the value of DTS (BDnSTAT<6>). If a packet arrives with an incorrect synchronization, the data will essentially be ignored. It will not be written to the USB RAM and the USB transfer complete interrupt flag will not be set. The SIE will send an ACK token back to the host to Acknowledge receipt, however. The effects of the DTSEN bit on the SIE are summarized in Table 14-3.

The Buffer Stall bit, BSTALL (BDnSTAT<2>), provides support for control transfers, usually one-time stalls on Endpoint 0. It also provides support for the SET_FEATURE/CLEAR_FEATURE commands specified in Chapter 9 of the USB specification; typically, continuous STALLs to any endpoint other than the default control endpoint.

The BSTALL bit enables buffer stalls. Setting BSTALL causes the SIE to return a STALL token to the host if a received token would use the BD in that location. The EPSTALL bit in the corresponding UEPn control register is set and a STALL interrupt is generated when a STALL is issued to the host. The UOWN bit remains set and the BDs are not changed unless a SETUP token is received. In this case, the STALL condition is cleared and the ownership of the BD is returned to the microcontroller core.

The BD9:BD8 bits (BDnSTAT<1:0>) store the two most significant digits of the SIE byte count; the lower 8 digits are stored in the corresponding BDnCNT register. See **Section 14.4.2 "BD Byte Count"** for more information.

TABLE 14-3: EFFECT OF DTSEN BIT ON ODD/EVEN (DATA0/DATA1) PACKET RECEPTION

OUT Packet from Host	BDnSTAT Settings		Device Response after Receiving Packet			
	DTSEN	DTS	Handshake	UOWN	TRNIF	BDnSTAT and USTAT Status
DATA0	1	0	ACK	0	1	Updated
DATA1	1	0	ACK	1	0	Not Updated
DATA1	1	1	ACK	0	1	Updated
DATA0	1	1	ACK	1	0	Not Updated
Either	0	x	ACK	0	1	Updated
Either, with error	x	x	NAK	1	0	Not Updated

Legend: x = don't care

14.6 USB Power Modes

Many USB applications will likely have several different sets of power requirements and configuration. The most common power modes encountered are Bus Power Only, Self-Power Only and Dual Power with Self-Power Dominance. The most common cases are presented here.

14.6.1 BUS POWER ONLY

In Bus Power Only mode, all power for the application is drawn from the USB (Figure 14-10). This is effectively the simplest power method for the device.

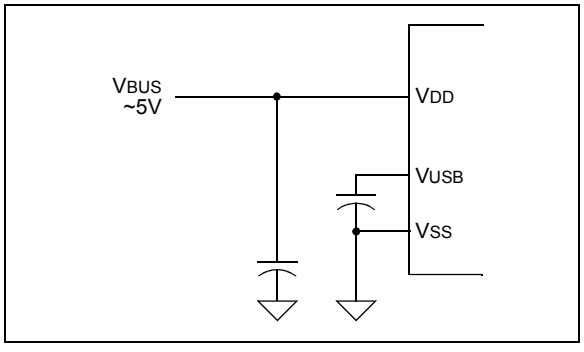
In order to meet the inrush current requirements of the USB 2.0 specifications, the total effective capacitance appearing across VBUS and ground must be no more than 10 μF; otherwise, some kind of inrush limiting is required. For more details, see Section 7.2.4 of the USB 2.0 specification.

According to the USB 2.0 specification, all USB devices must also support a Low-Power Suspend mode. In the USB Suspend mode, devices must consume no more than 500 A (or 2.5 mA for high-powered devices that are capable of remote wake-up) from the 5V VBUS line of the USB cable.

The host signals the USB device to enter the Suspend mode by stopping all USB traffic to that device for more than 3 ms. This condition will set the IDLEIF bit in the UIR register.

During the USB Suspend mode, the D+ or D- pull-up resistor must remain active, which will consume some of the allowed suspend current: 500A/2.5 mA budget.

FIGURE 14-10: BUS POWER ONLY



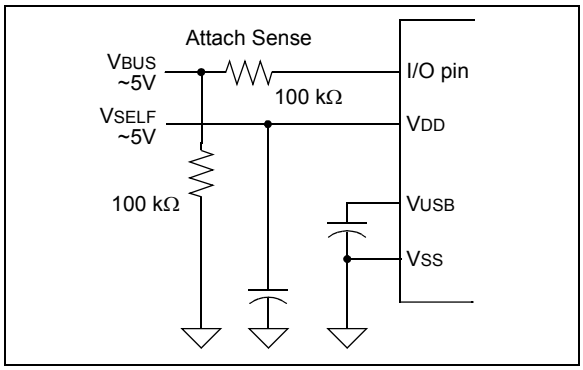
14.6.2 SELF-POWER ONLY

In Self-Power Only mode, the USB application provides its own power, with very little power being pulled from the USB. Figure 14-11 shows an example. Note that an attach indication is added to show when the USB has been connected and the host is actively powering VBUS.

In order to meet compliance specifications, the USB module (and the D+ or D- pull-up resistor) should not be enabled until the host actively drives VBUS high. One of the I/O pins may be used for this purpose.

The application should never source any current onto the 5V VBUS pin of the USB cable.

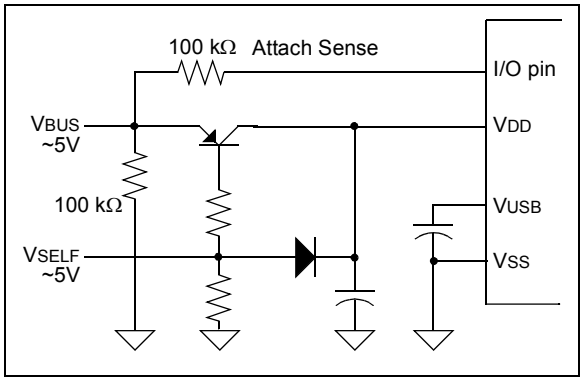
FIGURE 14-11: SELF-POWER ONLY



14.6.3 DUAL POWER WITH SELF-POWER DOMINANCE

Some applications may require a dual power option. This allows the application to use internal power primarily, but switch to power from the USB when no internal power is available. Figure 14-12 shows a simple Dual Power with Self-Power Dominance example, which automatically switches between Self-Power Only and USB Bus Power Only modes.

FIGURE 14-12: DUAL POWER EXAMPLE



Dual power devices must also meet all of the special requirements for inrush current and Suspend mode current, and must not enable the USB module until VBUS is driven high. For descriptions of those requirements, see **Section 14.6.1 "Bus Power Only"** and **Section 14.6.2 "Self-Power Only"**. Additionally, dual power devices must never source current onto the 5V VUSB pin of the USB cable.

Note:	Users should keep in mind the limits for devices drawing power from the USB. According to USB Specification 2.0, this cannot exceed 100 mA per low-power device or 500 mA per high-power device.

15.3 EUSART Synchronous Master Mode

The Synchronous Master mode is entered by setting the CSRC bit (TXSTA<7>). In this mode, the data is transmitted in a half-duplex manner (i.e., transmission and reception do not occur at the same time). When transmitting data, the reception is inhibited and vice versa. Synchronous mode is entered by setting bit, SYNC (TXSTA<4>). In addition, enable bit, SPEN (RCSTA<7>), is set in order to configure the TX and RX pins to CK (clock) and DT (data) lines, respectively.

The Master mode indicates that the processor transmits the master clock on the CK line. Clock polarity is selected with the SCKP bit (BAUDCON<4>). Setting SCKP sets the Idle state on CK as high, while clearing the bit sets the Idle state as low. This option is provided to support Microwire devices with this module.

15.3.1 EUSART SYNCHRONOUS MASTER TRANSMISSION

The EUSART transmitter block diagram is shown in Figure 15-3. The heart of the transmitter is the Transmit (Serial) Shift Register (TSR). The Shift register obtains its data from the Read/Write Transmit Buffer register, TXREG. The TXREG register is loaded with data in software. The TSR register is not loaded until the last bit has been transmitted from the previous load. As soon as the last bit is transmitted, the TSR is loaded with new data from the TXREG (if available).

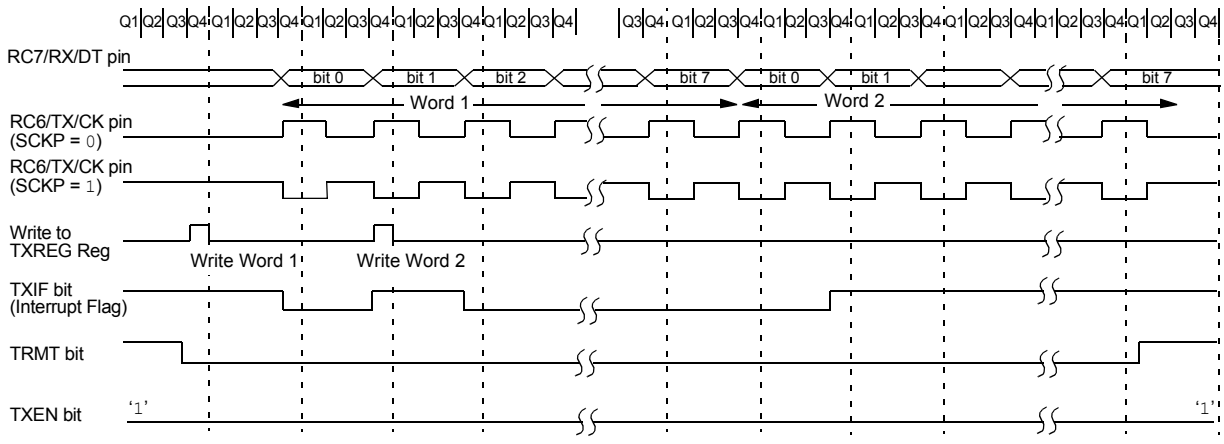
Once the TXREG register transfers the data to the TSR register (occurs in one TCYCLE), the TXREG register is empty and the TXIF flag bit (PIR1<4>) is set. The interrupt can be enabled or disabled by setting or clearing the interrupt enable bit, TXIE (PIE1<4>). TXIF is set regardless of the state of enable bit, TXIE; it cannot be cleared in software. It will reset only when new data is loaded into the TXREG register.

While flag bit, TXIF, indicates the status of the TXREG register, another bit, TRMT (TXSTA<1>), shows the status of the TSR register. TRMT is a read-only bit which is set when the TSR is empty. No interrupt logic is tied to this bit so the user must poll this bit in order to determine if the TSR register is empty. The TSR is not mapped in data memory so it is not available to the user.

To set up a Synchronous Master Transmission:

1. Initialize the SPBRGH:SPBRG registers for the appropriate baud rate. Set or clear the BRG16 bit, as required, to achieve the desired baud rate.
2. Enable the synchronous master serial port by setting bits, SYNC, SPEN and CSRC.
3. If interrupts are desired, set enable bit, TXIE.
4. If 9-bit transmission is desired, set bit, TX9.
5. Enable the transmission by setting bit, TXEN.
6. If 9-bit transmission is selected, the ninth bit should be loaded in bit, TX9D.
7. Start transmission by loading data to the TXREG register.
8. If using interrupts, ensure that the GIE and PEIE bits in the INTCON register (INTCON<7:6>) are set.

FIGURE 15-11: SYNCHRONOUS TRANSMISSION



Note: Sync Master mode (SPBRG = 0), continuous transmission of two 8-bit words.

15.3.2 EUSART SYNCHRONOUS MASTER RECEPTION

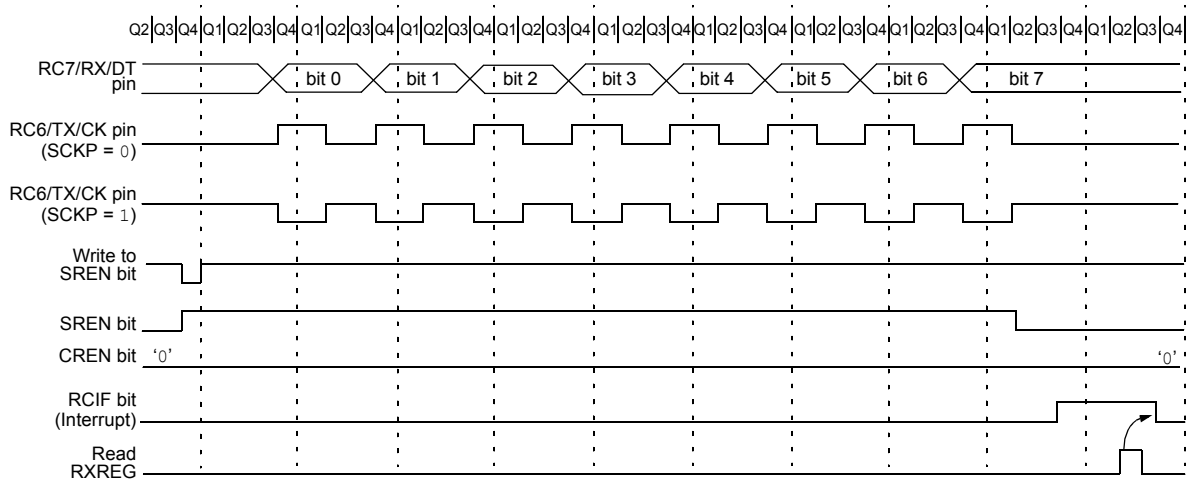
Once Synchronous mode is selected, reception is enabled by setting either the Single Receive Enable bit, SREN (RCSTA<5>), or the Continuous Receive Enable bit, CREN (RCSTA<4>). Data is sampled on the RX pin on the falling edge of the clock.

If enable bit, SREN, is set, only a single word is received. If enable bit, CREN, is set, the reception is continuous until CREN is cleared. If both bits are set, then CREN takes precedence.

To set up a Synchronous Master Reception:

1. Initialize the SPBRGH:SPBRG registers for the appropriate baud rate. Set or clear the BRG16 bit, as required, to achieve the desired baud rate.
2. Enable the synchronous master serial port by setting bits, SYNC, SPEN and CSRC.
3. Ensure bits, CREN and SREN, are clear.
4. If interrupts are desired, set enable bit, RCIE.
5. If 9-bit reception is desired, set bit, RX9.
6. If a single reception is required, set bit, SREN. For continuous reception, set bit, CREN.
7. Interrupt flag bit, RCIF, will be set when reception is complete and an interrupt will be generated if the enable bit, RCIE, was set.
8. Read the RCSTA register to get the ninth bit (if enabled) and determine if any error occurred during reception.
9. Read the 8-bit received data by reading the RCREG register.
10. If any error occurred, clear the error by clearing bit, CREN.
11. If using interrupts, ensure that the GIE and PEIE bits in the INTCON register (INTCON<7:6>) are set.

FIGURE 15-13: SYNCHRONOUS RECEPTION (MASTER MODE, SREN)



Note: Timing diagram demonstrates Sync Master mode with SREN bit = 1 and BRGH bit = 0.

TABLE 15-8: REGISTERS ASSOCIATED WITH SYNCHRONOUS MASTER RECEPTION

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on Page:
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF	49
PIR1	—	ADIF	RCIF	TXIF	—	CCP1IF	TMR2IF	TMR1IF	51
PIE1	—	ADIE	RCIE	TXIE	—	CCP1IE	TMR2IE	TMR1IE	51
IPR1	—	ADIP	RCIP	TXIP	—	CCP1IP	TMR2IP	TMR1IP	51
RCSTA	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D	51
RCREG	EUSART Receive Register								50
TXSTA	CSRC	TX9	TXEN	SYNC	SENDB	BRGH	TRMT	TX9D	51
BAUDCON	ABDOVF	RCIDL	—	SCKP	BRG16	—	WUE	ABDEN	51
SPBRGH	EUSART Baud Rate Generator Register High Byte								51
SPBRG	EUSART Baud Rate Generator Register Low Byte								50

Legend: — = unimplemented, read as '0'. Shaded cells are not used for synchronous master reception.

REGISTER 16-3: ADCON2: A/D CONTROL REGISTER 2

R/W-0	U-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
ADFM	—	ACQT2	ACQT1	ACQT0	ADCS2	ADCS1	ADCS0
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'	
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7 **ADFM:** A/D Result Format Select bit
1 = Right justified
0 = Left justified

bit 6 **Unimplemented:** Read as '0'

bit 5-3 **ACQT2:ACQT0:** A/D Acquisition Time Select bits
111 = 20 TAD
110 = 16 TAD
101 = 12 TAD
100 = 8 TAD
011 = 6 TAD
010 = 4 TAD
001 = 2 TAD
000 = 0 TAD(1)

bit 2-0 **ADCS2:ADCS0:** A/D Conversion Clock Select bits
111 = FRC (clock derived from A/D RC oscillator)(1)
110 = FOSC/64
101 = FOSC/16
100 = FOSC/4
011 = FRC (clock derived from A/D RC oscillator)(1)
010 = FOSC/32
001 = FOSC/8
000 = FOSC/2

Note 1: If the A/D FRC clock source is selected, a delay of one TCY (instruction cycle) is added before the A/D clock starts. This allows the SLEEP instruction to be executed before starting a conversion.

The value in the ADRESH:ADRESL registers is not modified for a Power-on Reset. The ADRESH:ADRESL registers will contain unknown data after a Power-on Reset.

After the A/D module has been configured as desired, the selected channel must be acquired before the conversion is started. The analog input channels must have their corresponding TRIS bits selected as an input. To determine acquisition time, see **Section 16.1 "A/D Acquisition Requirements"**. After this acquisition time has elapsed, the A/D conversion can be started. An acquisition time can be programmed to occur between setting the GO/$\overline{\text{DONE}}$ bit and the actual start of the conversion.

The following steps should be followed to perform an A/D conversion:

1. Configure the A/D module:
   - Configure analog pins, voltage reference and digital I/O (ADCON1)
   - Select A/D input channel (ADCON0)
   - Select A/D acquisition time (ADCON2)
   - Select A/D conversion clock (ADCON2)
   - Turn on A/D module (ADCON0)
2. Configure A/D interrupt (if desired):
   - Clear ADIF bit
   - Set ADIE bit
   - Set GIE bit
3. Wait the required acquisition time (if required).
4. Start conversion:
   - Set GO/$\overline{\text{DONE}}$ bit (ADCON0 register)
5. Wait for A/D conversion to complete, by either:
   - Polling for the GO/$\overline{\text{DONE}}$ bit to be cleared

   OR
   - Waiting for the A/D interrupt
6. Read A/D Result registers (ADRESH:ADRESL); clear bit, ADIF, if required.
7. For next conversion, go to step 1 or step 2, as required. The A/D conversion time per bit is defined as TAD. A minimum wait of 3 TAD is required before the next acquisition starts.

FIGURE 16-2: A/D TRANSFER FUNCTION

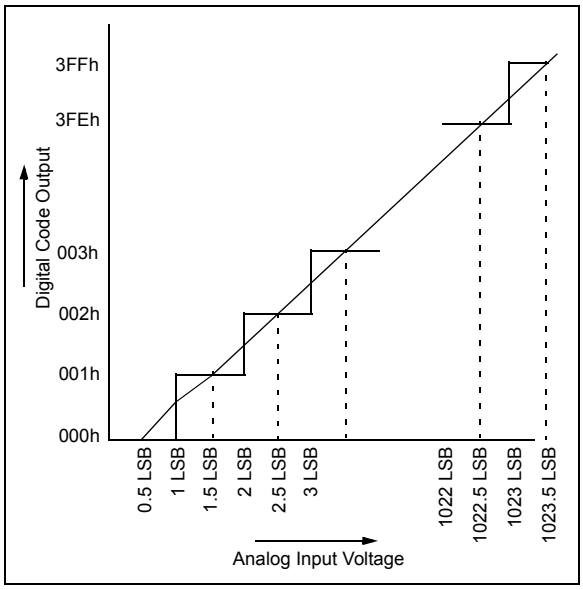
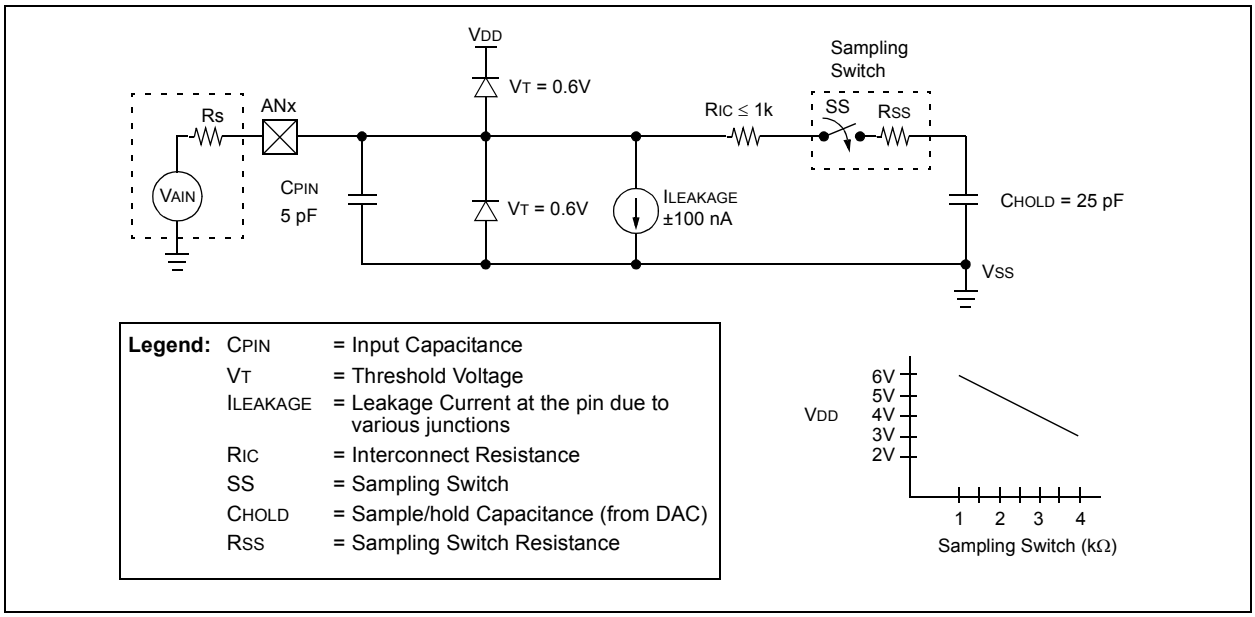


FIGURE 16-3: ANALOG INPUT MODEL



16.8 Use of the CCP1 Trigger

An A/D conversion can be started by the Special Event Trigger of the CCP1 module. This requires that the CCP1M3:CCP1M0 bits (CCP1CON<3:0>) be programmed as '1011' and that the A/D module is enabled (ADON bit is set). When the trigger occurs, the GO/$\overline{\text{DONE}}$ bit will be set, starting the A/D acquisition and conversion, and the Timer1 counter will be reset to zero. Timer1 is reset to automatically repeat the A/D acquisition period with minimal software overhead (moving ADRESH:ADRESL to the desired location). The appropriate analog input channel must be selected and the minimum acquisition period is either timed by the user, or an appropriate TACQ time selected before the Special Event Trigger sets the GO/$\overline{\text{DONE}}$ bit (starts a conversion).

If the A/D module is not enabled (ADON is cleared), the Special Event Trigger will be ignored by the A/D module but will still reset the Timer1 counter.

TABLE 16-2: REGISTERS ASSOCIATED WITH A/D OPERATION

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on Page:
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF	49
PIR1	—	ADIF	RCIF	TXIF	—	CCP1IF	TMR2IF	TMR1IF	51
PIE1	—	ADIE	RCIE	TXIE	—	CCP1IE	TMR2IE	TMR1IE	51
IPR1	—	ADIP	RCIP	TXIP	—	CCP1IP	TMR2IP	TMR1IP	51
PIR2	OSCFIF	—	USBIF	—	—	HLVDIF	—	—	51
PIE2	OSCFIE	—	USBIE	—	—	HLVDIE	—	—	51
IPR2	OSCFIP	—	USBIP	—	—	HLVDIP	—	—	51
ADRESH	A/D Result Register High Byte								50
ADRESL	A/D Result Register Low Byte								50
ADCON0	—	—	CHS3	CHS2	CHS1	CHS0	GO/$\overline{\text{DONE}}$	ADON	50
ADCON1	—	—	VCFG1	VCFG0	PCFG3	PCFG2	PCFG1	PCFG0	50
ADCON2	ADFM	—	ACQT2	ACQT1	ACQT0	ADCS2	ADCS1	ADCS0	50
PORTA	—	RA6(2)	RA5	RA4	RA3	RA2	RA1	RA0	51
TRISA	—	TRISA6(2)	TRISA5	TRISA4	TRISA3	TRISA2	TRISA1	TRISA0	51
PORTB	RB7	RB6	RB5	RB4	RB3	RB2	RB1	RB0	51
TRISB	TRISB7	TRISB6	TRISB5	TRISB4	TRISB3	TRISB2	TRISB1	TRISB0	51
LATB	LATB7	LATB6	LATB5	LATB4	LATB3	LATB2	LATB1	LATB0	51
PORTE	—	—	—	—	RE3(1,3)	RE2(4)	RE1(4)	RE0(4)	51
TRISE(4)	—	—	—	—	—	TRISE2(4)	TRISE1(4)	TRISE0(4)	51
LATE(4)	—	—	—	—	—	LATE2(4)	LATE1(4)	LATE0(4)	51

Legend: — = unimplemented, read as '0'. Shaded cells are not used for A/D conversion.

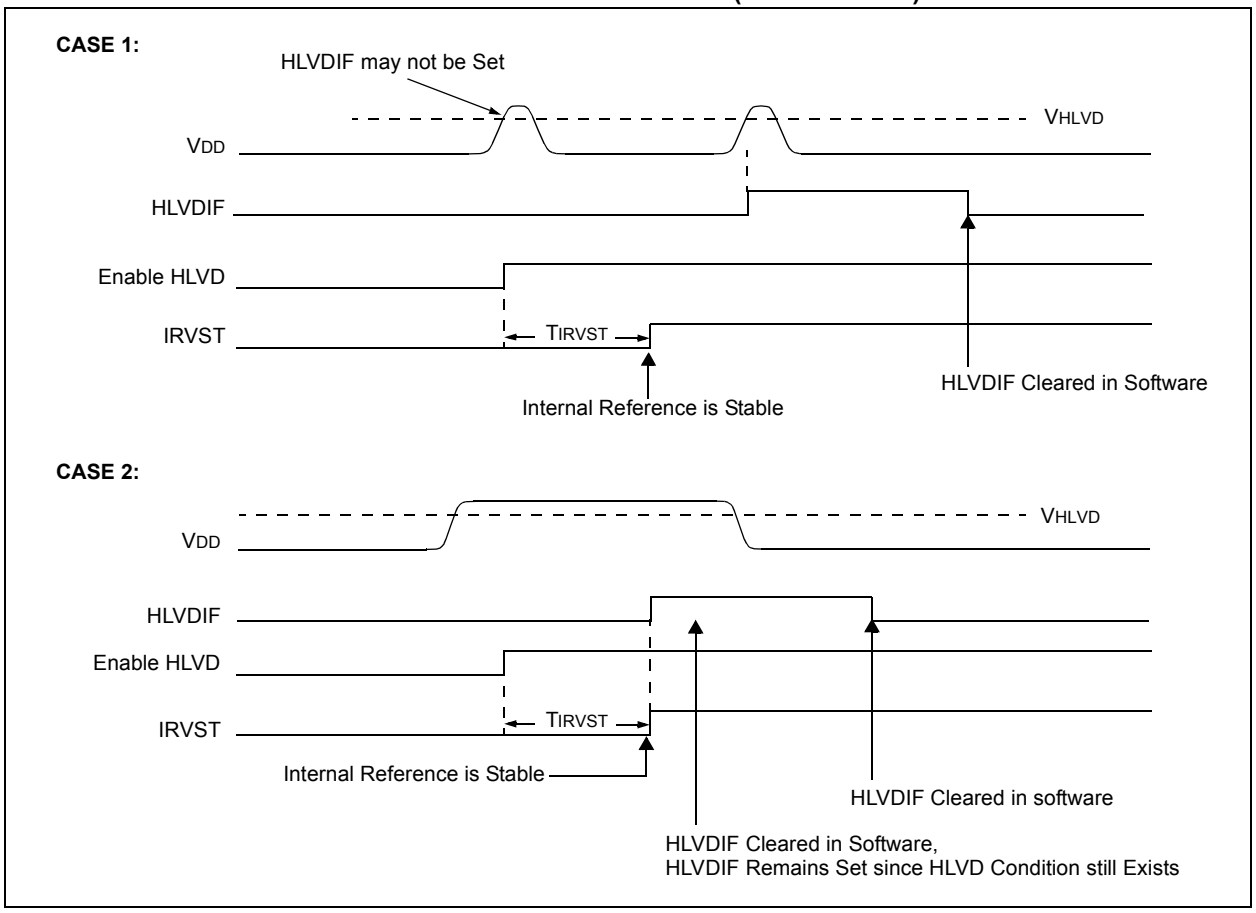
Note 1: Implemented only when Master Clear functionality is disabled (MCLRE Configuration bit = 0).

2: RA6 and its associated latch and data direction bits are enabled as I/O pins based on oscillator configuration; otherwise, they are read as '0'.

3: RE3 port bit is available only as an input pin when the MCLRE Configuration bit is '0'.

4: These registers and/or bits are not implemented on 28-pin devices.

FIGURE 17-3: HIGH-VOLTAGE DETECT OPERATION (VDIRMAG = 1)

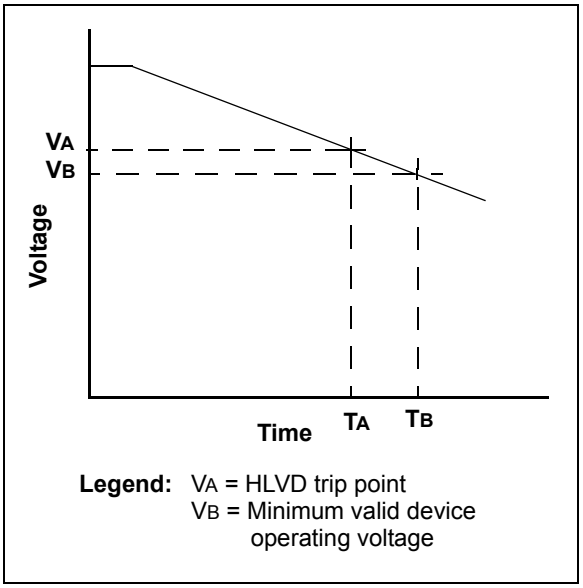


17.5 Applications

In many applications, the ability to detect a drop below or rise above a particular threshold is desirable. For example, the HLVD module could be periodically enabled to detect Universal Serial Bus (USB) attach or detach. This assumes the device is powered by a lower voltage source than the USB when detached. An attach would indicate a high-voltage detect from, for example, 3.3V to 5V (the voltage on USB) and vice versa for a detach. This feature could save a design a few extra components and an attach signal (input pin).

For general battery applications, Figure 17-4 shows a possible voltage curve. Over time, the device voltage decreases. When the device voltage reaches voltage, VA, the HLVD logic generates an interrupt at time, TA. The interrupt could cause the execution of an ISR, which would allow the application to perform "house-keeping tasks" and perform a controlled shutdown before the device voltage exits the valid operating range at TB. The HLVD, thus, would give the application a time window, represented by the difference between TA and TB, to safely exit.

FIGURE 17-4: TYPICAL HIGH/LOW-VOLTAGE DETECT APPLICATION



Legend: VA = HLVD trip point
VB = Minimum valid device operating voltage

REGISTER 18-9: CONFIG6L: CONFIGURATION REGISTER 6 LOW (BYTE ADDRESS 30000Ah)

U-0	U-0	U-0	U-0	U-0	U-0	R/C-1	R/C-1
—	—	—	—	—	—	WRT1	WRT0
bit 7							bit 0

Legend:		
R = Readable bit	C = Clearable bit	U = Unimplemented bit, read as '0'
-n = Value when device is unprogrammed		u = Unchanged from programmed state

bit 7-2 **Unimplemented:** Read as '0'

bit 1 **WRT1:** Write Protection bit
1 = Block 1 (002000-003FFFh) is not write-protected
0 = Block 1 (002000-003FFFh) is write-protected

bit 0 **WRT0:** Write Protection bit
1 = Block 0 (000800-001FFFh) or (001000-001FFFh) is not write-protected
0 = Block 0 (000800-001FFFh) or (001000-001FFFh) is write-protected

REGISTER 18-10: CONFIG6H: CONFIGURATION REGISTER 6 HIGH (BYTE ADDRESS 30000Bh)

U-0	R/C-1	R-1	U-0	U-0	U-0	U-0	U-0
—	WRTB	WRTC(1)	—	—	—	—	—
bit 7							bit 0

Legend:		
R = Readable bit	C = Clearable bit	U = Unimplemented bit, read as '0'
-n = Value when device is unprogrammed		u = Unchanged from programmed state

bit 7 **Unimplemented:** Read as '0'

bit 6 **WRTB:** Boot Block Write Protection bit
1 = Boot block (000000-0007FFh) or (000000-000FFFh) is not write-protected
0 = Boot block (000000-0007FFh) or (000000-000FFFh) is write-protected

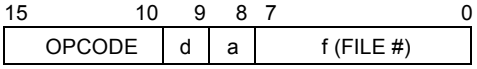
bit 5 **WRTC:** Configuration Register Write Protection bit(1)
1 = Configuration registers (300000-3000FFh) are not write-protected
0 = Configuration registers (300000-3000FFh) are write-protected

bit 4-0 **Unimplemented:** Read as '0'

Note 1: This bit is read-only in normal execution mode; it can be written only in Program mode.

FIGURE 19-1: GENERAL FORMAT FOR INSTRUCTIONS

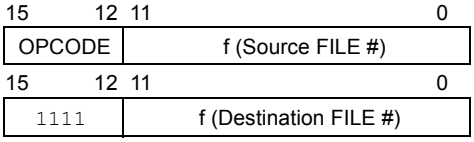
Byte-oriented file register operations — **Example Instruction**



`ADDWF MYREG, W, B`

d = 0 for result destination to be WREG register
d = 1 for result destination to be file register (f)
a = 0 to force Access Bank
a = 1 for BSR to select bank
f = 8-bit file register address

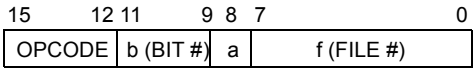
Byte to Byte move operations (2-word)



`MOVFF MYREG1, MYREG2`

f = 12-bit file register address

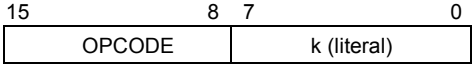
Bit-oriented file register operations



`BSF MYREG, bit, B`

b = 3-bit position of bit in file register (f)
a = 0 to force Access Bank
a = 1 for BSR to select bank
f = 8-bit file register address

Literal operations

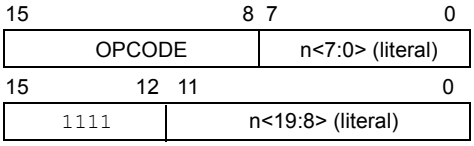


`MOVLW 7Fh`

k = 8-bit immediate value

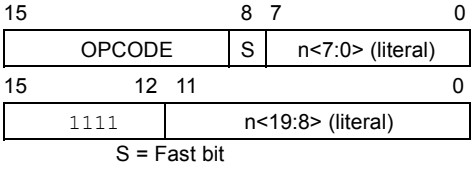
Control operations

CALL, GOTO and Branch operations



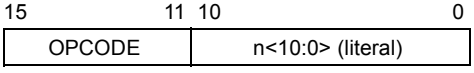
`GOTO Label`

n = 20-bit immediate value

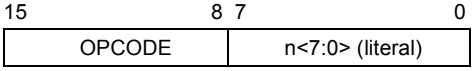


`CALL MYFUNC`

S = Fast bit



`BRA MYFUNC`



`BC MYFUNC`

BTG **Bit Toggle f**

Syntax: BTG f, b {,a}

Operands: $0 \le f \le 255$
$0 \le b < 7$
$a \in [0,1]$

Operation: $(\overline{f<b>}) \rightarrow f<b>$

Status Affected: None

Encoding:

0111	bbba	ffff	ffff

Description: Bit 'b' in data memory location 'f' is inverted.
If 'a' is '0', the Access Bank is selected.
If 'a' is '1', the BSR is used to select the GPR bank (default).
If 'a' is '0' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever $f \le 95$ (5Fh). See **Section 19.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode"** for details.

Words: 1

Cycles: 1

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	Read register 'f'	Process Data	Write register 'f'

Example: `BTG PORTC, 4, 0`

Before Instruction:
PORTC = 0111 0101 [75h]
After Instruction:
PORTC = 0110 0101 [65h]

BOV **Branch if Overflow**

Syntax: BOV n

Operands: $-128 \le n \le 127$

Operation: if Overflow bit is '1',
$(PC) + 2 + 2n \rightarrow PC$

Status Affected: None

Encoding:

1110	0100	nnnn	nnnn

Description: If the Overflow bit is '1', then the program will branch.
The 2's complement number '2n' is added to the PC. Since the PC will have incremented to fetch the next instruction, the new address will be PC + 2 + 2n. This instruction is then a two-cycle instruction.

Words: 1

Cycles: 1(2)

Q Cycle Activity:
If Jump:

Q1	Q2	Q3	Q4
Decode	Read literal 'n'	Process Data	Write to PC
No operation	No operation	No operation	No operation

If No Jump:

Q1	Q2	Q3	Q4
Decode	Read literal 'n'	Process Data	No operation

Example: `HERE BOV Jump`

Before Instruction
PC = address (HERE)
After Instruction
If Overflow = 1;
PC = address (Jump)
If Overflow = 0;
PC = address (HERE + 2)

19.2 Extended Instruction Set

In addition to the standard 75 instructions of the PIC18 instruction set, PIC18F2450/4450 devices also provide an optional extension to the core CPU functionality. The added features include eight additional instructions that augment indirect and indexed addressing operations and the implementation of Indexed Literal Offset Addressing mode for many of the standard PIC18 instructions.

The additional features of the extended instruction set are disabled by default. To enable them, users must set the XINST Configuration bit.

The instructions in the extended set can all be classified as literal operations, which either manipulate the File Select Registers, or use them for Indexed Addressing. Two of the instructions, `ADDFSR` and `SUBFSR`, each have an additional special instantiation for using FSR2. These versions (`ADDULNK` and `SUBULNK`) allow for automatic return after execution.

The extended instructions are specifically implemented to optimize re-entrant program code (that is, code that is recursive or that uses a software stack) written in high-level languages, particularly C. Among other things, they allow users working in high-level languages to perform certain operations on data structures more efficiently. These include:

- Dynamic allocation and deallocation of software stack space when entering and leaving subroutines
- Function Pointer invocation
- Software Stack Pointer manipulation
- Manipulation of variables located in a software stack

A summary of the instructions in the extended instruction set is provided in Table 19-3. Detailed descriptions are provided in **Section 19.2.2 "Extended Instruction Set"**. The opcode field descriptions in Table 19-1 (page 214) apply to both the standard and extended PIC18 instruction sets.

Note:	The instruction set extension and the Indexed Literal Offset Addressing mode were designed for optimizing applications written in C; the user may likely never use these instructions directly in assembler. The syntax for these commands is provided as a reference for users who may be reviewing code that has been generated by a compiler.

19.2.1 EXTENDED INSTRUCTION SYNTAX

Most of the extended instructions use indexed arguments, using one of the File Select Registers and some offset to specify a source or destination register. When an argument for an instruction serves as part of Indexed Addressing, it is enclosed in square brackets ("[ ]"). This is done to indicate that the argument is used as an index or offset. The MPASM™ Assembler will flag an error if it determines that an index or offset value is not bracketed.

When the extended instruction set is enabled, brackets are also used to indicate index arguments in byte-oriented and bit-oriented instructions. This is in addition to other changes in their syntax. For more details, see **Section 19.2.3.1 "Extended Instruction Syntax with Standard PIC18 Commands"**.

Note:	In the past, square brackets have been used to denote optional arguments in the PIC18 and earlier instruction sets. In this text and going forward, optional arguments are denoted by braces ("{ }").

TABLE 19-3: EXTENSIONS TO THE PIC18 INSTRUCTION SET

Mnemonic, Operands		Description	Cycles	16-Bit Instruction Word				Status Affected
				MSb			LSb	
ADDFSR	f, k	Add Literal to FSR	1	1110	1000	ffkk	kkkk	None
ADDULNK	k	Add Literal to FSR2 and Return	2	1110	1000	11kk	kkkk	None
CALLW		Call Subroutine using WREG	2	0000	0000	0001	0100	None
MOVSF	z_s, f_d	Move z_s (source) to 1st word f_d (destination) 2nd word	2	1110 1111	1011 ffff	0zzz ffff	zzzz ffff	None
MOVSS	z_s, z_d	Move z_s (source) to 1st word z_d (destination) 2nd word	2	1110 1111	1011 xxxx	1zzz xzzz	zzzz zzzz	None
PUSHL	k	Store Literal at FSR2, Decrement FSR2	1	1110	1010	kkkk	kkkk	None
SUBFSR	f, k	Subtract Literal from FSR	1	1110	1001	ffkk	kkkk	None
SUBULNK	k	Subtract Literal from FSR2 and Return	2	1110	1001	11kk	kkkk	None

19.2.3 BYTE-ORIENTED AND BIT-ORIENTED INSTRUCTIONS IN INDEXED LITERAL OFFSET MODE

Note:	Enabling the PIC18 instruction set extension may cause legacy applications to behave erratically or fail entirely.

In addition to eight new commands in the extended set, enabling the extended instruction set also enables Indexed Literal Offset Addressing mode (**Section 5.6.1 "Indexed Addressing with Literal Offset"**). This has a significant impact on the way that many commands of the standard PIC18 instruction set are interpreted.

When the extended set is disabled, addresses embedded in opcodes are treated as literal memory locations: either as a location in the Access Bank ('a' = 0) or in a GPR bank designated by the BSR ('a' = 1). When the extended instruction set is enabled and 'a' = 0, however, a file register argument of 5Fh or less is interpreted as an offset from the pointer value in FSR2 and not as a literal address. For practical purposes, this means that all instructions that use the Access RAM bit as an argument – that is, all byte-oriented and bit-oriented instructions, or almost half of the core PIC18 instructions – may behave differently when the extended instruction set is enabled.

When the content of FSR2 is 00h, the boundaries of the Access RAM are essentially remapped to their original values. This may be useful in creating backward compatible code. If this technique is used, it may be necessary to save the value of FSR2 and restore it when moving back and forth between C and assembly routines in order to preserve the Stack Pointer. Users must also keep in mind the syntax requirements of the extended instruction set (see **Section 19.2.3.1 "Extended Instruction Syntax with Standard PIC18 Commands"**).

Although the Indexed Literal Offset Addressing mode can be very useful for dynamic stack and pointer manipulation, it can also be very annoying if a simple arithmetic operation is carried out on the wrong register. Users who are accustomed to the PIC18 programming must keep in mind that, when the extended instruction set is enabled, register addresses of 5Fh or less are used for Indexed Literal Offset Addressing.

Representative examples of typical byte-oriented and bit-oriented instructions in the Indexed Literal Offset Addressing mode are provided on the following page to show how execution is affected. The operand conditions shown in the examples are applicable to all instructions of these types.

19.2.3.1 Extended Instruction Syntax with Standard PIC18 Commands

When the extended instruction set is enabled, the file register argument, 'f', in the standard byte-oriented and bit-oriented commands is replaced with the literal offset value, 'k'. As already noted, this occurs only when 'f' is less than or equal to 5Fh. When an offset value is used, it must be indicated by square brackets ("[ ]"). As with the extended instructions, the use of brackets indicates to the compiler that the value is to be interpreted as an index or an offset. Omitting the brackets, or using a value greater than 5Fh within brackets, will generate an error in the MPASM Assembler.

If the index argument is properly bracketed for Indexed Literal Offset Addressing mode, the Access RAM argument is never specified; it will automatically be assumed to be '0'. This is in contrast to standard operation (extended instruction set disabled) when 'a' is set on the basis of the target address. Declaring the Access RAM bit in this mode will also generate an error in the MPASM Assembler.

The destination argument, 'd', functions as before.

In the latest versions of the MPASM assembler, language support for the extended instruction set must be explicitly invoked. This is done with either the command line option, /y, or the PE directive in the source listing.

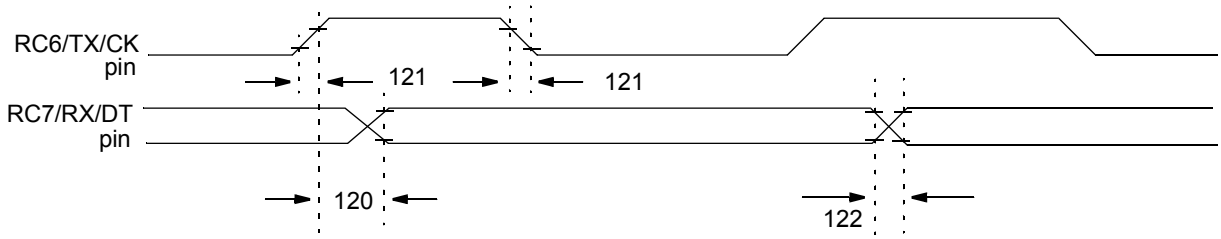
19.2.4 CONSIDERATIONS WHEN ENABLING THE EXTENDED INSTRUCTION SET

It is important to note that the extensions to the instruction set may not be beneficial to all users. In particular, users who are not writing code that uses a software stack may not benefit from using the extensions to the instruction set.

Additionally, the Indexed Literal Offset Addressing mode may create issues with legacy applications written to the PIC18 assembler. This is because instructions in the legacy code may attempt to address registers in the Access Bank below 5Fh. Since these addresses are interpreted as literal offsets to FSR2 when the instruction set extension is enabled, the application may read or write to the wrong data addresses.

When porting an application to the PIC18F2450/4450, it is very important to consider the type of code. A large, re-entrant application that is written in 'C' and would benefit from efficient compilation will do well when using the instruction set extensions. Legacy applications that heavily use the Access Bank will most likely not benefit from using the extended instruction set.

FIGURE 21-11: EUSART SYNCHRONOUS TRANSMISSION (MASTER/SLAVE) TIMING

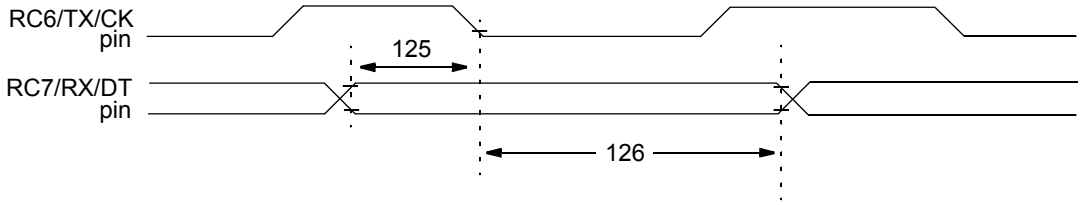


Note: Refer to Figure 21-4 for load conditions.

TABLE 21-13: EUSART SYNCHRONOUS TRANSMISSION REQUIREMENTS

Param No.	Symbol	Characteristic		Min	Max	Units	Conditions
120	TckH2dtV	SYNC XMIT (MASTER & SLAVE) Clock High to Data Out Valid	PIC18**F**XXXX	—	40	ns	
			PIC18**LF**XXXX	—	100	ns	VDD = 2.0V
121	Tckrf	Clock Out Rise Time and Fall Time (Master mode)	PIC18**F**XXXX	—	20	ns	
			PIC18**LF**XXXX	—	50	ns	VDD = 2.0V
122	Tdtrf	Data Out Rise Time and Fall Time	PIC18**F**XXXX	—	20	ns	
			PIC18**LF**XXXX	—	50	ns	VDD = 2.0V

FIGURE 21-12: EUSART SYNCHRONOUS RECEIVE (MASTER/SLAVE) TIMING



Note: Refer to Figure 21-4 for load conditions.

TABLE 21-14: EUSART SYNCHRONOUS RECEIVE REQUIREMENTS

Param. No.	Symbol	Characteristic	Min	Max	Units	Conditions
125	TDTV2CKL	SYNC RCV (MASTER & SLAVE) Data Hold before CK ↓ (DT hold time)	10	—	ns	
126	TCKL2DTL	Data Hold after CK ↓ (DT hold time)	15	—	ns	