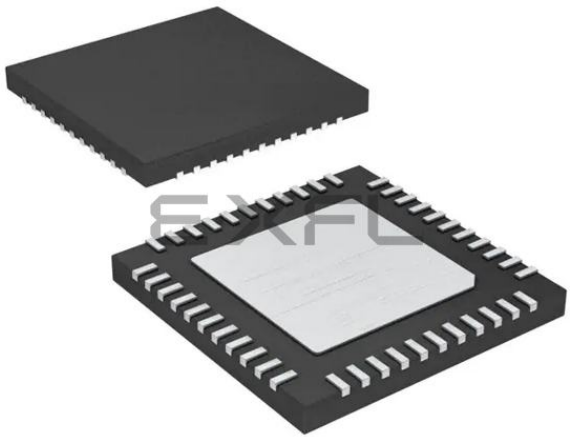

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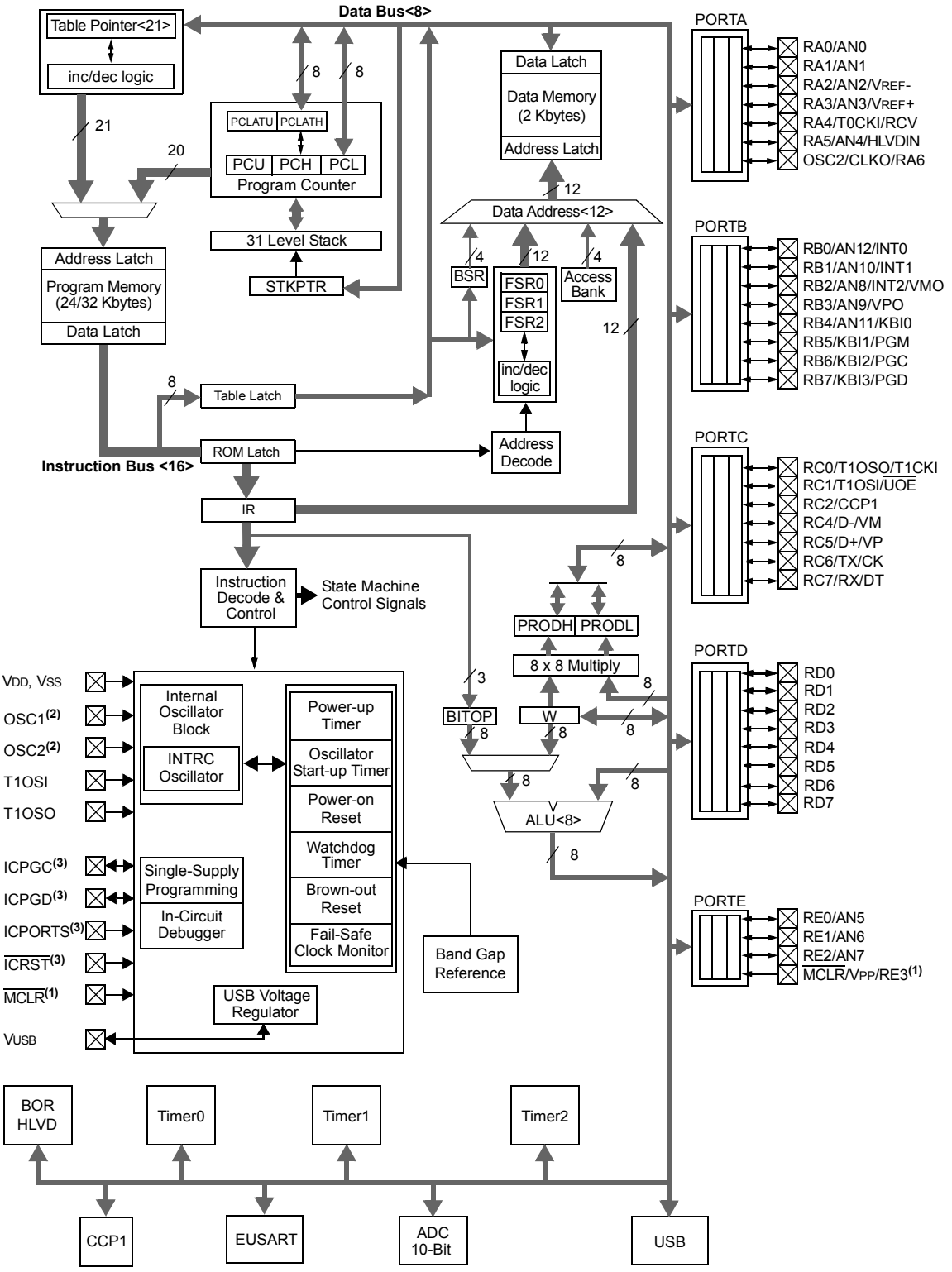
Applications of "Embedded - Microcontrollers"

Details	
Product Status	Obsolete
Core Processor	PIC
Core Size	8-Bit
Speed	48MHz
Connectivity	UART/USART, USB
Peripherals	Brown-out Detect/Reset, HLVD, POR, PWM, WDT
Number of I/O	34
Program Memory Size	16KB (8K x 16)
Program Memory Type	FLASH
EEPROM Size	-
RAM Size	768 x 8
Voltage - Supply (Vcc/Vdd)	4.2V ~ 5.5V
Data Converters	A/D 13x10b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	44-VQFN Exposed Pad
Supplier Device Package	44-QFN (8x8)
Purchase URL	https://www.e-xfl.com/product-detail/microchip-technology/pic18f4450t-i-ml

TABLE 1-1: DEVICE FEATURES

Features	PIC18F2450	PIC18F4450
Operating Frequency	DC – 48 MHz	DC – 48 MHz
Program Memory (Bytes)	16384	16384
Program Memory (Instructions)	8192	8192
Data Memory (Bytes)	768	768
Interrupt Sources	13	13
I/O Ports	Ports A, B, C, (E)	Ports A, B, C, D, E
Timers	3	3
Capture/Compare/PWM Modules	1	1
Enhanced USART	1	1
Universal Serial Bus (USB) Module	1	1
10-Bit Analog-to-Digital Module	10 Input Channels	13 Input Channels
Resets (and Delays)	POR, BOR, RESET Instruction, Stack Full, Stack Underflow (PWRT, OST), MCLR (optional), WDT	POR, BOR, RESET Instruction, Stack Full, Stack Underflow (PWRT, OST), MCLR (optional), WDT
Programmable Low-Voltage Detect	Yes	Yes
Programmable Brown-out Reset	Yes	Yes
Instruction Set	75 Instructions; 83 with Extended Instruction Set enabled	75 Instructions; 83 with Extended Instruction Set enabled
Packages	28-Pin SPDIP 28-Pin SOIC 28-Pin QFN	40-Pin PDIP 44-Pin QFN 44-Pin TQFP

FIGURE 1-2: PIC18F4450 (40/44-PIN) BLOCK DIAGRAM



Note 1: RE3 is multiplexed with $\overline{\text{MCLR}}$ and is only available when the $\overline{\text{MCLR}}$ Resets are disabled.

2: OSC1/CLKI and OSC2/CLKO are only available in select oscillator modes and when these pins are not being used as digital I/O. Refer to **Section 2.0 "Oscillator Configurations"** for additional information.

3: These pins are only available on 44-pin TQFP under certain conditions. Refer to **Section 18.9 "Special ICPORT Features (Designated Packages Only)"** for additional information.

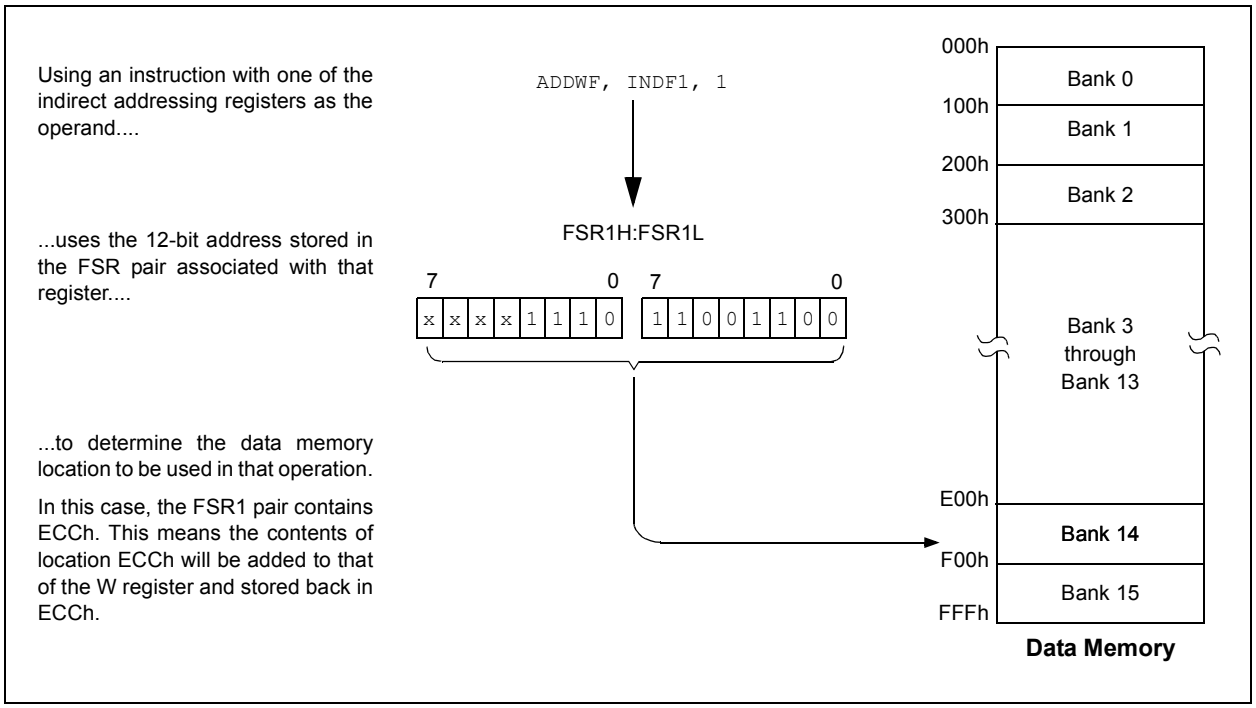
5.4.3.1 FSR Registers and the INDF Operand

At the core of Indirect Addressing are three sets of registers: FSR0, FSR1 and FSR2. Each represents a pair of 8-bit registers: FSRnH and FSRnL. The four upper bits of the FSRnH register are not used, so each FSR pair holds a 12-bit value. This represents a value that can address the entire range of the data memory in a linear fashion. The FSR register pairs, then, serve as pointers to data memory locations.

Indirect Addressing is accomplished with a set of Indirect File Operands, INDF0 through INDF2. These can be thought of as “virtual” registers; they are mapped in the SFR space but are not physically implemented. Reading or writing to a particular INDF register actually accesses its corresponding FSR register pair. A read from INDF1, for example, reads the data at the address indicated by FSR1H:FSR1L. Instructions that use the INDF registers as operands actually use the contents of their corresponding FSR as a pointer to the instruction’s target. The INDF operand is just a convenient way of using the pointer.

Because Indirect Addressing uses a full 12-bit address, data RAM banking is not necessary. Thus, the current contents of the BSR and the Access RAM bit have no effect on determining the target address.

FIGURE 5-7: INDIRECT ADDRESSING



NOTES:

TABLE 11-2: REGISTERS ASSOCIATED WITH TIMER1 AS A TIMER/COUNTER

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on Page:
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF	49
PIR1	—	ADIF	RCIF	TXIF	—	CCP1IF	TMR2IF	TMR1IF	51
PIE1	—	ADIE	RCIE	TXIE	—	CCP1IE	TMR2IE	TMR1IE	51
IPR1	—	ADIP	RCIP	TXIP	—	CCP1IP	TMR2IP	TMR1IP	51
TMR1L	Timer1 Register Low Byte								50
TMR1H	TImer1 Register High Byte								50
T1CON	RD16	T1RUN	T1CKPS1	T1CKPS0	T1OSCEN	$\overline{\text{T1SYNC}}$	TMR1CS	TMR1ON	50

Legend: — = unimplemented, read as '0'. Shaded cells are not used by the Timer1 module.

TABLE 14-1: DIFFERENTIAL OUTPUTS TO TRANSCEIVER

VPO	VMO	Bus State
0	0	Single-Ended Zero
0	1	Differential '0'
1	0	Differential '1'
1	1	Illegal Condition

TABLE 14-2: SINGLE-ENDED INPUTS FROM TRANSCEIVER

VP	VM	Bus State
0	0	Single-Ended Zero
0	1	Low Speed
1	0	High Speed
1	1	Error

The $\overline{UOE}$ signal toggles the state of the external transceiver. This line is pulled low by the device to enable the transmission of data from the SIE to an external device.

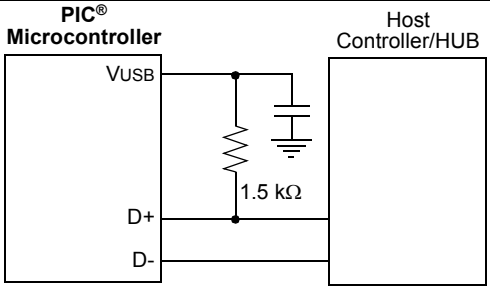
14.2.2.3 Internal Pull-up Resistors

The PIC18F2450/4450 devices have built-in pull-up resistors designed to meet the requirements for low-speed and full-speed USB. The UPUEN bit (UCFG<4>) enables the internal pull-ups. Figure 14-1 shows the pull-ups and their control.

14.2.2.4 Pull-up Resistors

The PIC18F2450/4450 devices require an external pull-up resistor to meet the requirements for low-speed and full-speed USB. Either an external 3.3V supply or the VUSB pin may be used to pull up D+ or D-. The pull-up resistor must be 1.5 kΩ (±5%) as required by the USB specifications. Figure 14-3 shows an example with the VUSB pin.

FIGURE 14-3: EXTERNAL CIRCUITRY



Note: The above setting shows a typical connection for a full-speed configuration using an on-chip regulator and an external pull-up resistor.

14.2.2.5 Ping-Pong Buffer Configuration

The usage of ping-pong buffers is configured using the PPB1:PPB0 bits. Refer to **Section 14.4.4 "Ping-Pong Buffering"** for a complete explanation of the ping-pong buffers.

14.2.2.6 USB Output Enable Monitor

The USB $\overline{OE}$ monitor provides indication as to whether the SIE is listening to the bus or actively driving the bus. This is enabled by default when using an external transceiver or when UCFG<6> = 1.

The USB $\overline{OE}$ monitoring is useful for initial system debugging, as well as scope triggering during eye pattern generation tests.

14.2.2.7 Eye Pattern Test Enable

An automatic eye pattern test can be generated by the module when the UCFG<7> bit is set. The eye pattern output will be observable based on module settings, meaning that the user is first responsible for configuring the SIE clock settings, pull-up resistor and Transceiver mode. In addition, the module has to be enabled.

Once UTEYE is set, the module emulates a switch from a receive to transmit state and will start transmitting a J-K-J-K bit sequence (K-J-K-J for full speed). The sequence will be repeated indefinitely while the Eye Pattern Test mode is enabled.

Note that this bit should never be set while the module is connected to an actual USB system. This test mode is intended for board verification to aid with USB certification tests. It is intended to show a system developer the noise integrity of the USB signals which can be affected by board traces, impedance mismatches and proximity to other system components. It does not properly test the transition from a receive to a transmit state. Although the eye pattern is not meant to replace the more complex USB certification test, it should aid during first order system debugging.

14.2.2.8 Internal Regulator

The PIC18F2450/4450 devices have a built-in 3.3V regulator to provide power to the internal transceiver and provide a source for the external pull-ups. An external 220 nF (±20%) capacitor is required for stability.

Note: The drive from VUSB is sufficient to only drive an external pull-up in addition to the internal transceiver.

The regulator is disabled by default and can be enabled through the VREGEN Configuration bit. When enabled, the voltage is visible on pin VUSB. When the regulator is disabled, a 3.3V source must be provided through the VUSB pin for the internal transceiver. If the internal transceiver is disabled, VUSB is not used.

Note 1: Do not enable the internal regulator if an external regulator is connected to VUSB.

2: VDD must be greater than or equal to VUSB at all times, even with the regulator disabled.

14.4.4 PING-PONG BUFFERING

An endpoint is defined to have a ping-pong buffer when it has two sets of BD entries: one set for an EVEN transfer and one set for an ODD transfer. This allows the CPU to process one BD while the SIE is processing the other BD. Double-buffering BDs in this way allows for maximum throughput to/from the USB.

The USB module supports three modes of operation:

- No ping-pong support
- Ping-pong buffer support for OUT Endpoint 0 only
- Ping-pong buffer support for all endpoints

The ping-pong buffer settings are configured using the PPB1:PPB0 bits in the UCFG register.

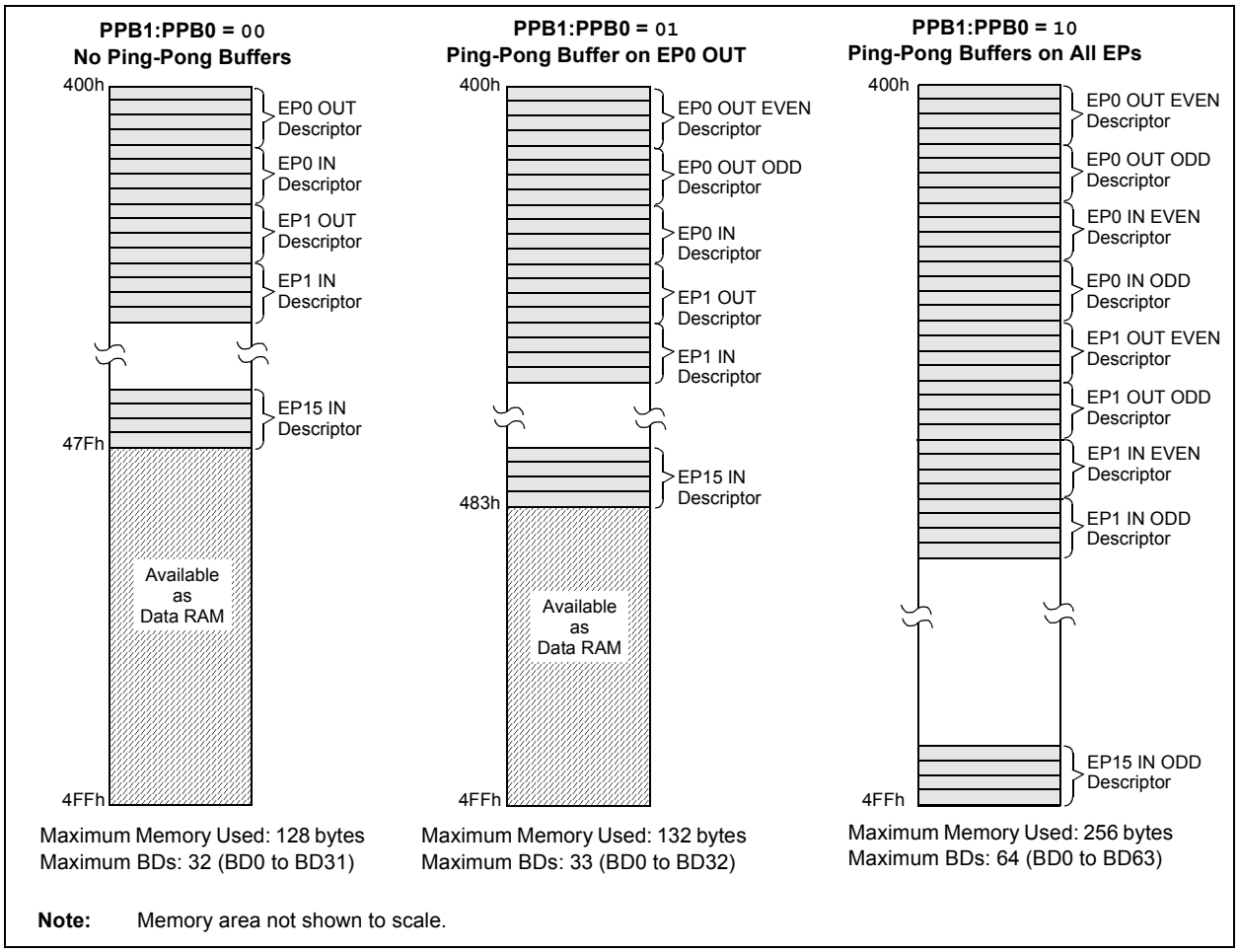
The USB module keeps track of the Ping-Pong Pointer individually for each endpoint. All pointers are initially reset to the EVEN BD when the module is enabled. After the completion of a transaction (UOWN cleared by the SIE), the pointer is toggled to the ODD BD. After the completion of the next transaction, the pointer is toggled back to the EVEN BD and so on.

The EVEN/ODD status of the last transaction is stored in the PPBI bit of the USTAT register. The user can reset all Ping-Pong Pointers to EVEN using the PPBRST bit.

Figure 14-7 shows the three different modes of operation and how USB RAM is filled with the BDs.

BDs have a fixed relationship to a particular endpoint, depending on the buffering configuration. The mapping of BDs to endpoints is detailed in Table 14-4. This relationship also means that gaps may occur in the BDT if endpoints are not enabled contiguously. This theoretically means that the BDs for disabled endpoints could be used as buffer space. In practice, users should avoid using such spaces in the BDT unless a method of validating BD addresses is implemented.

FIGURE 14-7: BUFFER DESCRIPTOR TABLE MAPPING FOR BUFFERING MODES



Note: Memory area not shown to scale.

14.5.1.1 Bus Activity Detect Interrupt Bit (ACTVIF)

The ACTVIF bit cannot be cleared immediately after the USB module wakes up from Suspend or while the USB module is suspended. A few clock cycles are required to synchronize the internal hardware state machine before the ACTVIF bit can be cleared by firmware. Clearing the ACTVIF bit before the internal hardware is synchronized may not have an effect on the value of ACTVIF. Additionally, if the USB module uses the clock from the 96 MHz PLL source, then after clearing the SUSPND bit, the USB module may not be immediately operational while waiting for the 96 MHz PLL to lock. The application code should clear the ACTVIF bit as shown in Example 14-1.

Only one ACTVIF interrupt is generated when resuming from the USB bus Idle condition. If user firmware clears the ACTVIF bit, the bit will not immediately become set again, even when there is continuous bus traffic. Bus traffic must cease long enough to generate another IDLEIF condition before another ACTVIF interrupt can be generated.

EXAMPLE 14-1: CLEARING ACTVIF BIT (UIR<2>)

Assembly:

```
        BCF     UCON, SUSPND
LOOP:
        BTFSS   UIR, ACTVIF
        BRA     DONE
        BCF     UIR, ACTVIF
        BRA     LOOP
DONE
```

C:

```
UCONbits.SUSPND = 0;
while (UIRbits.ACTVIF){UIRbits.ACTVIF = 0};
```

14.5.2 USB INTERRUPT ENABLE REGISTER (UIE)

The USB Interrupt Enable register (Register 14-8) contains the enable bits for the USB status interrupt sources. Setting any of these bits will enable the respective interrupt source in the UIR register.

The values in this register only affect the propagation of an interrupt condition to the microcontroller's interrupt logic. The flag bits are still set by their interrupt conditions, allowing them to be polled and serviced without actually generating an interrupt.

REGISTER 14-8: UIE: USB INTERRUPT ENABLE REGISTER

U-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
—	SOFIE	STALLIE	IDLEIE	TRNIE	ACTVIE	UERRIE	URSTIE
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'	
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7 **Unimplemented:** Read as '0'

bit 6 **SOFIE:** Start-of-Frame Token Interrupt Enable bit
1 = Start-of-Frame token interrupt enabled
0 = Start-of-Frame token interrupt disabled

bit 5 **STALLIE:** STALL Handshake Interrupt Enable bit
1 = STALL interrupt enabled
0 = STALL interrupt disabled

bit 4 **IDLEIE:** Idle Detect Interrupt Enable bit
1 = Idle detect interrupt enabled
0 = Idle detect interrupt disabled

bit 3 **TRNIE:** Transaction Complete Interrupt Enable bit
1 = Transaction interrupt enabled
0 = Transaction interrupt disabled

bit 2 **ACTVIE:** Bus Activity Detect Interrupt Enable bit
1 = Bus activity detect interrupt enabled
0 = Bus activity detect interrupt disabled

bit 1 **UERRIE:** USB Error Interrupt Enable bit
1 = USB error interrupt enabled
0 = USB error interrupt disabled

bit 0 **URSTIE:** USB Reset Interrupt Enable bit
1 = USB Reset interrupt enabled
0 = USB Reset interrupt disabled

14.5.3 USB ERROR INTERRUPT STATUS REGISTER (UEIR)

The USB Error Interrupt Status register (Register 14-9) contains the flag bits for each of the error sources within the USB peripheral. Each of these sources is controlled by a corresponding interrupt enable bit in the UEIE register. All of the USB error flags are ORed together to generate the USB Error Interrupt Flag (UERRIF) at the top level of the interrupt logic.

Each error bit is set as soon as the error condition is detected. Thus, the interrupt will typically not correspond with the end of a token being processed.

Once an interrupt bit has been set by the SIE, it must be cleared by software by writing a '0'.

REGISTER 14-9: UEIR: USB ERROR INTERRUPT STATUS REGISTER

R/C-0	U-0	U-0	R/C-0	R/C-0	R/C-0	R/C-0	R/C-0
BTSEF	—	—	BTOEF	DFN8EF	CRC16EF	CRC5EF	PIDEF
bit 7							bit 0

Legend:			
R = Readable bit	C = Clearable bit	U = Unimplemented bit, read as '0'	
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7 **BTSEF:** Bit Stuff Error Flag bit
1 = A bit stuff error has been detected
0 = No bit stuff error

bit 6-5 **Unimplemented:** Read as '0'

bit 4 **BTOEF:** Bus Turnaround Time-out Error Flag bit
1 = Bus turnaround time-out has occurred (more than 16 bit times of Idle from previous EOP elapsed)
0 = No bus turnaround time-out

bit 3 **DFN8EF:** Data Field Size Error Flag bit
1 = The data field was not an integral number of bytes
0 = The data field was an integral number of bytes

bit 2 **CRC16EF:** CRC16 Failure Flag bit
1 = The CRC16 failed
0 = The CRC16 passed

bit 1 **CRC5EF:** CRC5 Host Error Flag bit
1 = The token packet was rejected due to a CRC5 error
0 = The token packet was accepted

bit 0 **PIDEF:** PID Check Failure Flag bit
1 = PID check failed
0 = PID check passed

REGISTER 15-1: TXSTA: TRANSMIT STATUS AND CONTROL REGISTER

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R-1	R/W-0
CSRC	TX9	TXEN(1)	SYNC	SENDB	BRGH	TRMT	TX9D
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'	
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7 **CSRC:** Clock Source Select bit

Asynchronous mode:
Don't care.

Synchronous mode:
1 = Master mode (clock generated internally from BRG)
0 = Slave mode (clock from external source)

bit 6 **TX9:** 9-Bit Transmit Enable bit

1 = Selects 9-bit transmission
0 = Selects 8-bit transmission

bit 5 **TXEN:** Transmit Enable bit(1)

1 = Transmit enabled
0 = Transmit disabled

bit 4 **SYNC:** EUSART Mode Select bit

1 = Synchronous mode
0 = Asynchronous mode

bit 3 **SENDB:** Send Break Character bit

Asynchronous mode:
1 = Send Sync Break on next transmission (cleared by hardware upon completion)
0 = Sync Break transmission completed

Synchronous mode:
Don't care.

bit 2 **BRGH:** High Baud Rate Select bit

Asynchronous mode:
1 = High speed
0 = Low speed

Synchronous mode:
Unused in this mode.

bit 1 **TRMT:** Transmit Shift Register Status bit

1 = TSR empty
0 = TSR full

bit 0 **TX9D:** 9th bit of Transmit Data

Can be address/data bit or a parity bit.

Note 1: SREN/CREN overrides TXEN in Sync mode with the exception that SREN has no effect in Synchronous Slave mode.

15.1 Baud Rate Generator (BRG)

The BRG is a dedicated, 8-bit or 16-bit generator that supports both the Asynchronous and Synchronous modes of the EUSART. By default, the BRG operates in 8-bit mode. Setting the BRG16 bit (BAUDCON<3>) selects 16-bit mode.

The SPBRGH:SPBRG register pair controls the period of a free-running timer. In Asynchronous mode, bits BRGH (TXSTA<2>) and BRG16 (BAUDCON<3>) also control the baud rate. In Synchronous mode, BRGH is ignored. Table 15-1 shows the formula for computation of the baud rate for different EUSART modes which only apply in Master mode (internally generated clock).

Given the desired baud rate and FOSC, the nearest integer value for the SPBRGH:SPBRG registers can be calculated using the formulas in Table 15-1. From this, the error in baud rate can be determined. An example calculation is shown in Example 15-1. Typical baud rates and error values for the various Asynchronous modes are shown in Table 15-2. It may be advantageous to use the high baud rate (BRGH = 1) or the 16-bit BRG to reduce the baud rate error, or achieve a slow baud rate for a fast oscillator frequency.

Writing a new value to the SPBRGH:SPBRG registers causes the BRG timer to be reset (or cleared). This ensures the BRG does not wait for a timer overflow before outputting the new baud rate.

15.1.1 OPERATION IN POWER-MANAGED MODES

The device clock is used to generate the desired baud rate. When one of the power-managed modes is entered, the new clock source may be operating at a different frequency. This may require an adjustment to the value in the SPBRG register pair.

15.1.2 SAMPLING

The data on the RX pin is sampled three times by a majority detect circuit to determine if a high or a low level is present at the RX pin.

TABLE 15-1: BAUD RATE FORMULAS

Configuration Bits			BRG/EUSART Mode	Baud Rate Formula
SYNC	BRG16	BRGH		
0	0	0	8-Bit/Asynchronous	FOSC/[64 (n + 1)]
0	0	1	8-Bit/Asynchronous	FOSC/[16 (n + 1)]
0	1	0	16-Bit/Asynchronous	
0	1	1	16-Bit/Asynchronous	FOSC/[4 (n + 1)]
1	0	x	8-Bit/Synchronous	
1	1	x	16-Bit/Synchronous	

Legend: x = Don't care, n = Value of SPBRGH:SPBRG register pair

15.2 EUSART Asynchronous Mode

The Asynchronous mode of operation is selected by clearing the SYNC bit (TXSTA<4>). In this mode, the EUSART uses the standard Non-Return-to-Zero (NRZ) format (one Start bit, eight or nine data bits and one Stop bit). The most common data format is eight bits. An on-chip dedicated 8-bit/16-bit Baud Rate Generator can be used to derive standard baud rate frequencies from the oscillator.

The EUSART transmits and receives the LSb first. The EUSART's transmitter and receiver are functionally independent but use the same data format and baud rate. The Baud Rate Generator produces a clock, either x16 or x64 of the bit shift rate depending on the BRGH and BRG16 bits (TXSTA<2> and BAUDCON<3>). Parity is not supported by the hardware but can be implemented in software and stored as the ninth data bit.

When operating in Asynchronous mode, the EUSART module consists of the following important elements:

- Baud Rate Generator
- Sampling Circuit
- Asynchronous Transmitter
- Asynchronous Receiver
- Auto-Wake-up on Sync Break Character
- 12-Bit Break Character Transmit
- Auto-Baud Rate Detection

15.2.1 EUSART ASYNCHRONOUS TRANSMITTER

The EUSART transmitter block diagram is shown in Figure 15-3. The heart of the transmitter is the Transmit (Serial) Shift Register (TSR). The Shift register obtains its data from the Read/Write Transmit Buffer register, TXREG. The TXREG register is loaded with data in software. The TSR register is not loaded until the Stop bit has been transmitted from the previous load. As soon as the Stop bit is transmitted, the TSR is loaded with new data from the TXREG register (if available).

Once the TXREG register transfers the data to the TSR register (occurs in one TCY), the TXREG register is empty and the TXIF flag bit (PIR1<4>) is set. This interrupt can be enabled or disabled by setting or clearing the interrupt enable bit, TXIE (PIE1<4>). TXIF will be set regardless of the state of TXIE; it cannot be cleared in software. TXIF is also not cleared immediately upon loading TXREG but becomes valid in the second instruction cycle following the load instruction. Polling TXIF immediately following a load of TXREG will return invalid results.

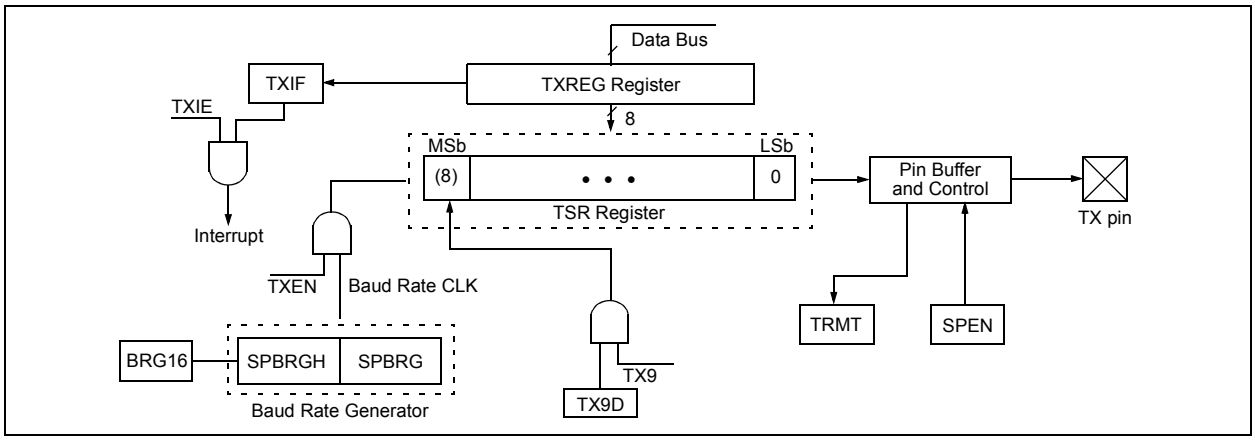
While TXIF indicates the status of the TXREG register, another bit, TRMT (TXSTA<1>), shows the status of the TSR register. TRMT is a read-only bit which is set when the TSR register is empty. No interrupt logic is tied to this bit so the user has to poll this bit in order to determine if the TSR register is empty.

Note 1:	The TSR register is not mapped in data memory so it is not available to the user.
2:	Flag bit, TXIF, is set when enable bit, TXEN, is set.

To set up an Asynchronous Transmission:

1. Initialize the SPBRGH:SPBRG registers for the appropriate baud rate. Set or clear the BRGH and BRG16 bits, as required, to achieve the desired baud rate.
2. Enable the asynchronous serial port by clearing bit, SYNC, and setting bit, SPEN.
3. If interrupts are desired, set enable bit, TXIE.
4. If 9-bit transmission is desired, set transmit bit, TX9. Can be used as address/data bit.
5. Enable the transmission by setting bit, TXEN, which will also set bit, TXIF.
6. If 9-bit transmission is selected, the ninth bit should be loaded in bit, TX9D.
7. Load data to the TXREG register (starts transmission).
8. If using interrupts, ensure that the GIE and PEIE bits in the INTCON register (INTCON<7:6>) are set.

FIGURE 15-3: EUSART TRANSMIT BLOCK DIAGRAM



REGISTER 16-2: ADCON1: A/D CONTROL REGISTER 1

U-0	U-0	R/W-0	R/W-0	R/W-0(1)	R/W(1)	R/W(1)	R/W(1)
—	—	VCFG1	VCFG0	PCFG3	PCFG2	PCFG1	PCFG0
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'	
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-6 **Unimplemented:** Read as '0'

bit 5 **VCFG1:** Voltage Reference Configuration bit (V_{REF-} source)
1 = V_{REF-} (AN2)
0 = V_{SS}

bit 4 **VCFG0:** Voltage Reference Configuration bit (V_{REF+} source)
1 = V_{REF+} (AN3)
0 = V_{DD}

bit 3-0 **PCFG3:PCFG0:** A/D Port Configuration Control bits:

PCFG3: PCFG0	AN12	AN11	AN10	AN9	AN8	AN7(2)	AN6(2)	AN5(2)	AN4	AN3	AN2	AN1	AN0
0000(1)	A	A	A	A	A	A	A	A	A	A	A	A	A
0001	A	A	A	A	A	A	A	A	A	A	A	A	A
0010	A	A	A	A	A	A	A	A	A	A	A	A	A
0011	D	A	A	A	A	A	A	A	A	A	A	A	A
0100	D	D	A	A	A	A	A	A	A	A	A	A	A
0101	D	D	D	A	A	A	A	A	A	A	A	A	A
0110	D	D	D	D	A	A	A	A	A	A	A	A	A
0111(1)	D	D	D	D	D	A	A	A	A	A	A	A	A
1000	D	D	D	D	D	D	A	A	A	A	A	A	A
1001	D	D	D	D	D	D	D	A	A	A	A	A	A
1010	D	D	D	D	D	D	D	D	A	A	A	A	A
1011	D	D	D	D	D	D	D	D	D	A	A	A	A
1100	D	D	D	D	D	D	D	D	D	D	A	A	A
1101	D	D	D	D	D	D	D	D	D	D	D	A	A
1110	D	D	D	D	D	D	D	D	D	D	D	D	A
1111	D	D	D	D	D	D	D	D	D	D	D	D	D

A = Analog input D = Digital I/O

Note 1: The POR value of the PCFG bits depends on the value of the PBADEN Configuration bit. When PBADEN = 1, PCFG<3:0> = 0000; when PBADEN = 0, PCFG<3:0> = 0111.

2: AN5 through AN7 are available only on 40/44-pin devices.

Even when the dedicated port is enabled, the ICSP and ICD functions remain available through the legacy port. When VIHH is seen on the MCLR/VPP/RE3 pin, the state of the ICRST/ICVPP pin is ignored.

Note 1: The ICPRT Configuration bit can only be programmed through the default ICSP port.

2: The ICPRT Configuration bit must be maintained clear for all 28-pin and 40-pin devices; otherwise, unexpected operation may occur.

18.9.2 28-PIN EMULATION

PIC18F4450 devices in 44-pin TQFP packages also have the ability to change their configuration under external control for debugging purposes. This allows the device to behave as if it were a PIC18F2450/4450 28-pin device.

This 28-pin Configuration mode is controlled through a single pin, NC/ICPORTS. Connecting this pin to VSS forces the device to function as a 28-pin device. Features normally associated with the 40/44-pin devices are disabled, along with their corresponding control registers and bits. On the other hand, connecting the pin to VDD forces the device to function in its default configuration.

The configuration option is only available when background debugging and the dedicated ICD/ICSP port are both enabled (DEBUG Configuration bit is clear and ICPRT Configuration bit is set). When disabled, NC/ICPORTS is a No Connect pin.

18.10 Single-Supply ICSP Programming

The LVP Configuration bit enables Single-Supply ICSP Programming (formerly known as *Low-Voltage ICSP Programming* or *LVP)*. When Single-Supply Programming is enabled, the microcontroller can be programmed without requiring high voltage being applied to the MCLR/VPP/RE3 pin, but the RB5/KBI1/PGM pin is then dedicated to controlling Program mode entry and is not available as a general purpose I/O pin.

While programming using Single-Supply Programming, VDD is applied to the MCLR/VPP/RE3 pin as in normal execution mode. To enter Programming mode, VDD is applied to the PGM pin.

Note 1: High-Voltage Programming is always available, regardless of the state of the LVP bit, by applying VIHH to the MCLR pin.

2: While in Low-Voltage ICSP Programming mode, the RB5 pin can no longer be used as a general purpose I/O pin and should be held low during normal operation.

3: When using Low-Voltage ICSP Programming (LVP) and the pull-ups on PORTB are enabled, bit 5 in the TRISB register must be cleared to disable the pull-up on RB5 and ensure the proper operation of the device.

4: If the device Master Clear is disabled, verify that either of the following is done to ensure proper entry into ICSP mode:

a) Disable Low-Voltage Programming (CONFIG4L<2> = 0); or

b) Make certain that RB5/KBI1/PGM is held low during entry into ICSP.

If Single-Supply ICSP Programming mode will not be used, the LVP bit can be cleared. RB5/KBI1/PGM then becomes available as the digital I/O pin, RB5. The LVP bit may be set or cleared only when using standard high-voltage programming (VIHH applied to the MCLR/VPP/RE3 pin). Once LVP has been disabled, only the standard high-voltage programming is available and must be used to program the device.

Memory that is not code-protected can be erased using either a Block Erase, or erased row by row, then written at any specified VDD. If code-protected memory is to be erased, a Block Erase is required. If a Block Erase is to be performed when using Low-Voltage Programming, the device must be supplied with VDD of 4.5V to 5.5V.

BNC **Branch if Not Carry**

Syntax: BNC n

Operands: -128 ≤ n ≤ 127

Operation: if Carry bit is '0',
(PC) + 2 + 2n → PC

Status Affected: None

Encoding:

1110	0011	nnnn	nnnn

Description: If the Carry bit is '0', then the program will branch.
The 2's complement number '2n' is added to the PC. Since the PC will have incremented to fetch the next instruction, the new address will be PC + 2 + 2n. This instruction is then a two-cycle instruction.

Words: 1

Cycles: 1(2)

Q Cycle Activity:
If Jump:

Q1	Q2	Q3	Q4
Decode	Read literal 'n'	Process Data	Write to PC
No operation	No operation	No operation	No operation

If No Jump:

Q1	Q2	Q3	Q4
Decode	Read literal 'n'	Process Data	No operation

Example: `HERE BNC Jump`

Before Instruction
PC = address `(HERE)`
After Instruction
If Carry = 0;
PC = address `(Jump)`
If Carry = 1;
PC = address `(HERE + 2)`

BNN **Branch if Not Negative**

Syntax: BNN n

Operands: -128 ≤ n ≤ 127

Operation: if Negative bit is '0',
(PC) + 2 + 2n → PC

Status Affected: None

Encoding:

1110	0111	nnnn	nnnn

Description: If the Negative bit is '0', then the program will branch.
The 2's complement number '2n' is added to the PC. Since the PC will have incremented to fetch the next instruction, the new address will be PC + 2 + 2n. This instruction is then a two-cycle instruction.

Words: 1

Cycles: 1(2)

Q Cycle Activity:
If Jump:

Q1	Q2	Q3	Q4
Decode	Read literal 'n'	Process Data	Write to PC
No operation	No operation	No operation	No operation

If No Jump:

Q1	Q2	Q3	Q4
Decode	Read literal 'n'	Process Data	No operation

Example: `HERE BNN Jump`

Before Instruction
PC = address `(HERE)`
After Instruction
If Negative = 0;
PC = address `(Jump)`
If Negative = 1;
PC = address `(HERE + 2)`

BRA **Unconditional Branch**

Syntax: BRA n

Operands: $-1024 \le n \le 1023$

Operation: $(PC) + 2 + 2n \rightarrow PC$

Status Affected: None

Encoding:

1101	0nnn	nnnn	nnnn

Description: Add the 2's complement number '2n' to the PC. Since the PC will have incremented to fetch the next instruction, the new address will be PC + 2 + 2n. This instruction is a two-cycle instruction.

Words: 1

Cycles: 2

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	Read literal 'n'	Process Data	Write to PC
No operation	No operation	No operation	No operation

Example: `HERE BRA Jump`

Before Instruction
PC = address `(HERE)`
After Instruction
PC = address `(Jump)`

BSF **Bit Set f**

Syntax: BSF f, b {,a}

Operands: $0 \le f \le 255$
$0 \le b \le 7$
$a \in [0,1]$

Operation: $1 \rightarrow f<b>$

Status Affected: None

Encoding:

1000	bbba	ffff	ffff

Description: Bit 'b' in register 'f' is set.
If 'a' is '0', the Access Bank is selected.
If 'a' is '1', the BSR is used to select the GPR bank (default).
If 'a' is '0' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever $f \le 95$ (5Fh). See **Section 19.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode"** for details.

Words: 1

Cycles: 1

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	Read register 'f'	Process Data	Write register 'f'

Example: `BSF FLAG_REG, 7, 1`

Before Instruction
FLAG_REG = 0Ah
After Instruction
FLAG_REG = 8Ah

TABLE 21-1: MEMORY PROGRAMMING REQUIREMENTS

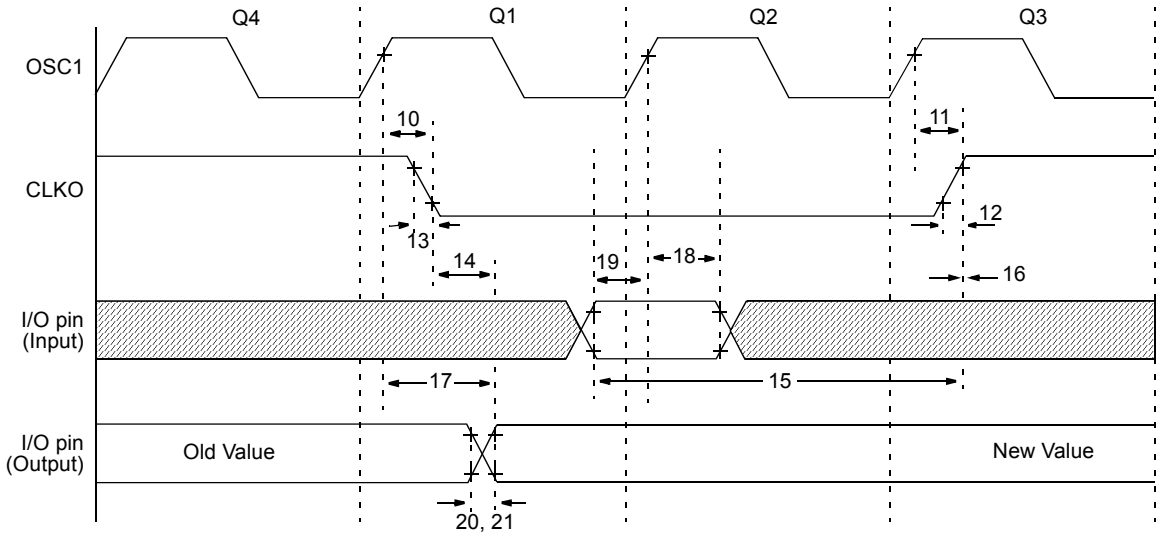
DC Characteristics			Standard Operating Conditions (unless otherwise stated) Operating temperature -40°C ≤ TA ≤ +85°C for industrial				
Param No.	**Sym**	**Characteristic**	**Min**	**Typ†**	**Max**	**Units**	**Conditions**
		Internal Program Memory Programming Specifications(1)					
D110	VIHH	Voltage on $\overline{\text{MCLR}}$/VPP/RE3 pin	9.00	—	13.25	V	**(Note 2)**
D113	IDDP	Supply Current during Programming	—	—	10	mA	
		Program Flash Memory					
D130	EP	Cell Endurance	10K	100K	—	E/W	-40°C to +85°C
D131	VPR	VDD for Read	VMIN	—	5.5	V	VMIN = Minimum operating voltage
D132	VIE	VDD for Block Erase	4.5	—	5.5	V	Using ICSP™ port
D132A	VIW	VDD for Externally Timed Erase or Write	3.0	—	5.5	V	Using ICSP port
D132B	VPEW	VDD for Self-Timed Write	VMIN	—	5.5	V	VMIN = Minimum operating voltage
D133	TIE	ICSP™ Block Erase Cycle Time	—	4	—	ms	VDD > 4.5V
D133A	TIW	ICSP Erase or Write Cycle Time (externally timed)	1	—	—	ms	VDD > 4.5V
D133A	TIW	Self-Timed Write Cycle Time	—	2	—	ms	
D134	TRETD	Characteristic Retention	40	100	—	Year	Provided no other specifications are violated

† Data in “Typ” column is at 5.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: These specifications are for programming the on-chip program memory through the use of table write instructions.

2: Required only if Single-Supply Programming is disabled.

FIGURE 21-6: CLKO AND I/O TIMING



Note: Refer to Figure 21-4 for load conditions.

TABLE 21-9: CLKO AND I/O TIMING REQUIREMENTS

Param No.	Symbol	Characteristic		Min	Typ	Max	Units	Conditions
10	TosH2ckL	OSC1 ↑ to CLKO ↓		—	75	200	ns	**(Note 1)**
11	TosH2ckH	OSC1 ↑ to CLKO ↑		—	75	200	ns	**(Note 1)**
12	TckR	CLKO Rise Time		—	35	100	ns	**(Note 1)**
13	TckF	CLKO Fall Time		—	35	100	ns	**(Note 1)**
14	TckL2ioV	CLKO ↓ to Port Out Valid		—	—	0.5 TCY + 20	ns	**(Note 1)**
15	TioV2ckH	Port In Valid before CLKO ↑		0.25 TCY + 25	—	—	ns	**(Note 1)**
16	TckH2ioI	Port In Hold after CLKO ↑		0	—	—	ns	**(Note 1)**
17	TosH2ioV	OSC1 ↑ (Q1 cycle) to Port Out Valid		—	50	150	ns	
18	TosH2ioI	OSC1 ↑ (Q2 cycle) to Port Input Invalid (I/O in hold time)	PIC18**F**XXXX	100	—	—	ns	
18A			PIC18**LF**XXXX	200	—	—	ns	VDD = 2.0V
19	TioV2osH	Port Input Valid to OSC1 ↑ (I/O in setup time)		0	—	—	ns	
20	TioR	Port Output Rise Time	PIC18**F**XXXX	—	10	25	ns	
20A			PIC18**LF**XXXX	—	—	60	ns	VDD = 2.0V
21	TioF	Port Output Fall Time	PIC18**F**XXXX	—	10	25	ns	
21A			PIC18**LF**XXXX	—	—	60	ns	VDD = 2.0V
22†	TINP	INTx Pin High or Low Time		TCY	—	—	ns	
23†	TRBP	RB7:RB4 Change Interrupt High or Low Time		TCY	—	—	ns	

† These parameters are asynchronous events not related to any internal clock edges.

Note 1: Measurements are taken in RC mode, where CLKO output is 4 x TOSC.

TABLE 21-18: A/D CONVERSION REQUIREMENTS

Param No.	Symbol	Characteristic		Min	Max	Units	Conditions
130	TAD	A/D Clock Period	PIC18**F**XXXX	0.7	25(1)	µs	TOSC based, VREF ≥ 3.0V
			PIC18**LF**XXXX	1.4	25(1)	µs	VDD = 2.0V, TOSC based, VREF full range
			PIC18**F**XXXX	2.0	6.0	µs	A/D RC mode
			PIC18**LF**XXXX	3.0	9.0	µs	VDD = 2.0V, A/D RC mode
131	TCNV	Conversion Time (not including acquisition time)(2)		11	12	TAD	
132	TACQ	Acquisition Time(3)		15 10	— —	µs µs	-40°C to +85°C 0°C ≤ to ≤ +85°C
135	TSWC	Switching Time from Convert → Sample		—	**(Note 4)**		
137	TDIS	Discharge Time		0.2	—	µs	

Note 1: The time of the A/D clock period is dependent on the device frequency and the TAD clock divider.

2: ADRES registers may be read on the following TCY cycle.

3: The time for the holding capacitor to acquire the “New” input voltage when the voltage changes full scale after the conversion (VDD to VSS or VSS to VDD). The source impedance (RS) on the input channels is 50Ω.

4: On the following cycle of the device clock.