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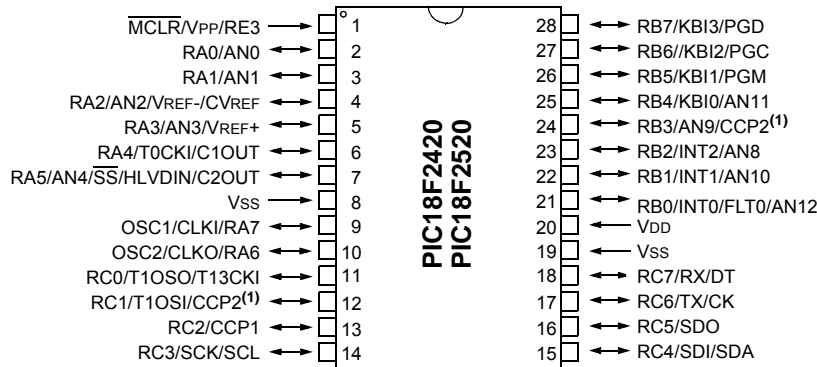
Details

Product Status	Active
Core Processor	PIC
Core Size	8-Bit
Speed	40MHz
Connectivity	I ² C, SPI, UART/USART
Peripherals	Brown-out Detect/Reset, HLVD, POR, PWM, WDT
Number of I/O	36
Program Memory Size	32KB (16K x 16)
Program Memory Type	FLASH
EEPROM Size	256 x 8
RAM Size	1.5K x 8
Voltage - Supply (Vcc/Vdd)	2V ~ 5.5V
Data Converters	A/D 13x10b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	44-TQFP
Supplier Device Package	44-TQFP (10x10)
Purchase URL	https://www.e-xfl.com/product-detail/microchip-technology/pic18lf4520t-i-pt

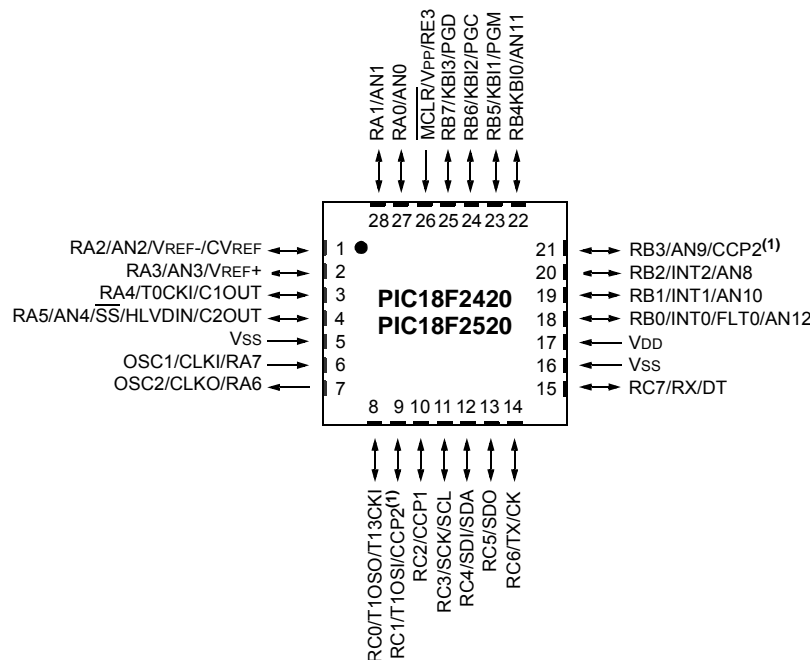
PIC18F2420/2520/4420/4520

Pin Diagrams

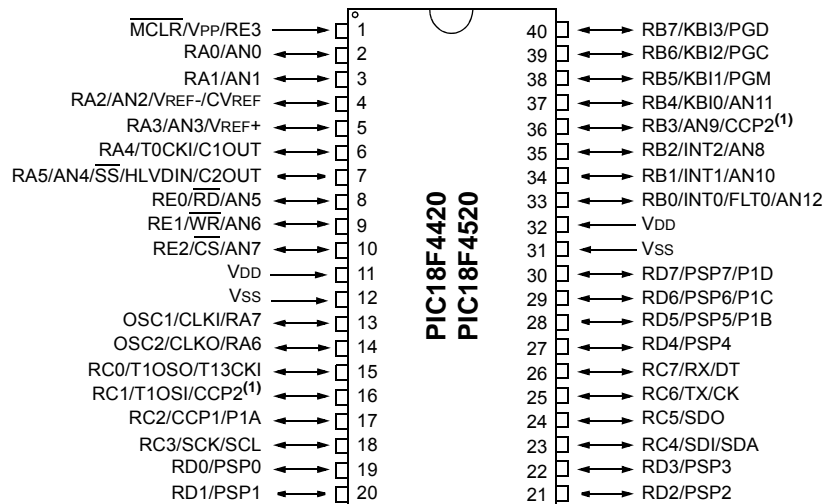
28-Pin SPDIP, SOIC



28-Pin QFN



40-Pin PDIP



Note 1: RB3 is the alternate pin for CCP2 multiplexing.

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TABLE 1-3: PIC18F4420/4520 PINOUT I/O DESCRIPTIONS (CONTINUED)

Pin Name	Pin Number			Pin Type	Buffer Type	Description
	PDIP	QFN	TQFP			
RC0/T1OSO/T13CKI RC0 T1OSO T13CKI	15	34	32	I/O O I	ST — ST	PORTC is a bidirectional I/O port. Digital I/O. Timer1 oscillator output. Timer1/Timer3 external clock input.
RC1/T1OSI/CCP2 RC1 T1OSI CCP2 ⁽²⁾	16	35	35	I/O I I/O	ST CMOS ST	Digital I/O. Timer1 oscillator input. Capture 2 input/Compare 2 output/PWM2 output.
RC2/CCP1/P1A RC2 CCP1 P1A	17	36	36	I/O I/O O	ST ST —	Digital I/O. Capture 1 input/Compare 1 output/PWM1 output. Enhanced CCP1 output.
RC3/SCK/SCL RC3 SCK SCL	18	37	37	I/O I/O I/O	ST ST ST	Digital I/O. Synchronous serial clock input/output for SPI mode. Synchronous serial clock input/output for I ² C™ mode.
RC4/SDI/SDA RC4 SDI SDA	23	42	42	I/O I I/O	ST ST ST	Digital I/O. SPI data in. I ² C data I/O.
RC5/SDO RC5 SDO	24	43	43	I/O O	ST —	Digital I/O. SPI data out.
RC6/TX/CK RC6 TX CK	25	44	44	I/O O I/O	ST — ST	Digital I/O. EUSART asynchronous transmit. EUSART synchronous clock (see related RX/DT).
RC7/RX/DT RC7 RX DT	26	1	1	I/O I I/O	ST ST ST	Digital I/O. EUSART asynchronous receive. EUSART synchronous data (see related TX/CK).

Legend: TTL = TTL compatible input

ST = Schmitt Trigger input with CMOS levels

O = Output

CMOS = CMOS compatible input or output

I = Input

P = Power

Note 1: Default assignment for CCP2 when Configuration bit, CCP2MX, is set.

2: Alternate assignment for CCP2 when Configuration bit, CCP2MX, is cleared.

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3.4.1 PRI_IDLE MODE

This mode is unique among the three low-power Idle modes in that it does not disable the primary device clock. For timing-sensitive applications, this allows for the fastest resumption of device operation with its more accurate primary clock source, since the clock source does not have to “warm-up” or transition from another oscillator.

PRI_IDLE mode is entered from PRI_RUN mode by setting the IDLEN bit and executing a *SLEEP* instruction. If the device is in another Run mode, set IDLEN first, then clear the SCS bits and execute *SLEEP*. Although the CPU is disabled, the peripherals continue to be clocked from the primary clock source specified by the FOSC<3:0> Configuration bits. The OSTS bit remains set (see Figure 3-7).

When a wake event occurs, the CPU is clocked from the primary clock source. A delay of interval T_{CSD} is required between the wake event and when code execution starts. This is required to allow the CPU to become ready to execute instructions. After the wake-up, the OSTS bit remains set. The IDLEN and SCS bits are not affected by the wake-up (see Figure 3-8).

setting the IDLEN bit and executing a *SLEEP* instruction. If the device is in another Run mode, set the IDLEN bit first, then set the SCS<1:0> bits to ‘01’ and execute *SLEEP*. When the clock source is switched to the Timer1 oscillator, the primary oscillator is shut down, the OSTS bit is cleared and the T1RUN bit is set.

When a wake event occurs, the peripherals continue to be clocked from the Timer1 oscillator. After an interval of T_{CSD}, following the wake event, the CPU begins executing code being clocked by the Timer1 oscillator. The IDLEN and SCS bits are not affected by the wake-up; the Timer1 oscillator continues to run (see Figure 3-8).

Note: The Timer1 oscillator should already be running prior to entering SEC_IDLE mode. If the T1OSCEN bit is not set when the *SLEEP* instruction is executed, the *SLEEP* instruction will be ignored and entry to SEC_IDLE mode will not occur. If the Timer1 oscillator is enabled but not yet running, peripheral clocks will be delayed until the oscillator has started. In such situations, initial oscillator operation is far from stable and unpredictable operation may result.

3.4.2 SEC_IDLE MODE

In SEC_IDLE mode, the CPU is disabled but the peripherals continue to be clocked from the Timer1 oscillator. This mode is entered from SEC_RUN by

FIGURE 3-7: TRANSITION TIMING FOR ENTRY TO IDLE MODE

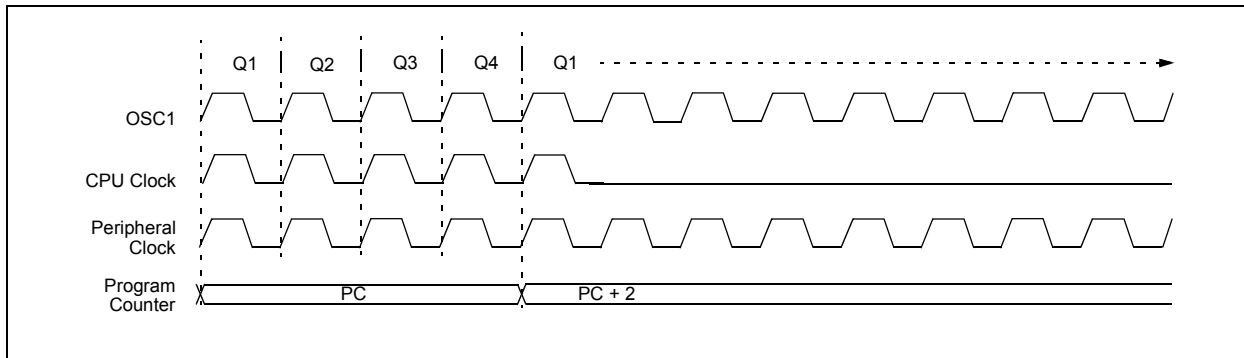
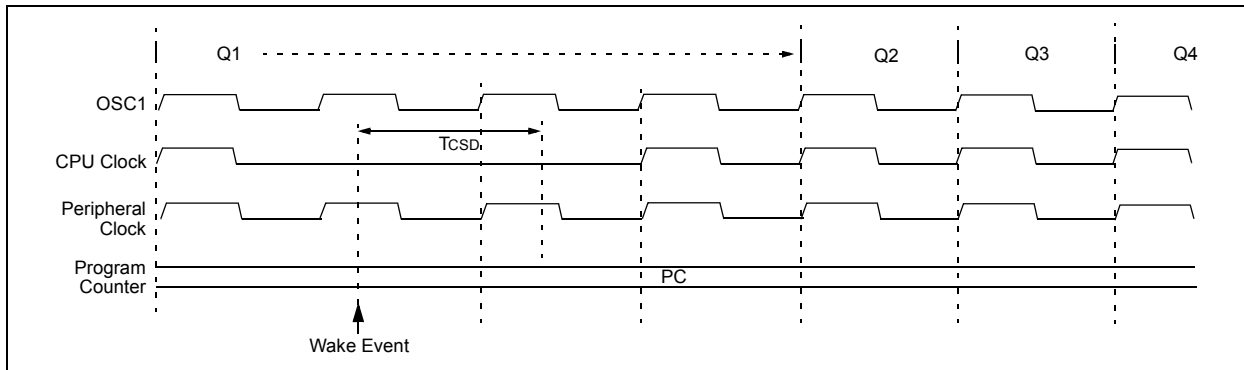


FIGURE 3-8: TRANSITION TIMING FOR WAKE FROM IDLE TO RUN MODE



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4.6 Reset State of Registers

Most registers are unaffected by a Reset. Their status is unknown on POR and unchanged by all other Resets. The other registers are forced to a “Reset state” depending on the type of Reset that occurred.

Most registers are not affected by a WDT wake-up, since this is viewed as the resumption of normal operation. Status bits from the RCON register, \overline{RI} , \overline{TO} , \overline{PD} , \overline{POR} and \overline{BOR} , are set or cleared differently in different Reset situations, as indicated in Table 4-3. These bits are used in software to determine the nature of the Reset.

Table 4-4 describes the Reset states for all of the Special Function Registers. These are categorized by Power-on and Brown-out Resets, Master Clear and WDT Resets and WDT wake-ups.

TABLE 4-3: STATUS BITS, THEIR SIGNIFICANCE AND THE INITIALIZATION CONDITION FOR RCON REGISTER

Condition	Program Counter	RCON Register					STKPTR Register	
		\overline{RI}	\overline{TO}	\overline{PD}	\overline{POR}	\overline{BOR}	STKFUL	STKUNF
Power-on Reset	0000h	1	1	1	0	0	0	0
RESET Instruction	0000h	0	u	u	u	u	u	u
Brown-out Reset	0000h	1	1	1	u	0	u	u
MCLR Reset during Power-Managed Run Modes	0000h	u	1	u	u	u	u	u
MCLR Reset during Power-Managed Idle Modes and Sleep Mode	0000h	u	1	0	u	u	u	u
WDT Time-out during Full Power or Power-Managed Run Mode	0000h	u	0	u	u	u	u	u
MCLR Reset during Full-Power Execution	0000h	u	u	u	u	u	u	u
Stack Full Reset (STVREN = 1)	0000h	u	u	u	u	u	1	u
Stack Underflow Reset (STVREN = 1)	0000h	u	u	u	u	u	u	1
Stack Underflow Error (not an actual Reset, STVREN = 0)	0000h	u	u	u	u	u	u	1
WDT Time-out during Power-Managed Idle or Sleep Modes	PC + 2	u	0	0	u	u	u	u
Interrupt Exit from Power-Managed Modes	PC + 2 ⁽¹⁾	u	u	0	u	u	u	u

Legend: u = unchanged

Note 1: When the wake-up is due to an interrupt and the GIEH or GIEL bits are set, the PC is loaded with the interrupt vector (008h or 0018h).

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5.3.4 SPECIAL FUNCTION REGISTERS

The Special Function Registers (SFRs) are registers used by the CPU and peripheral modules for controlling the desired operation of the device. These registers are implemented as static RAM. SFRs start at the top of data memory (FFFh) and extend downward to occupy the top half of Bank 15 (F80h to FFFh). A list of these registers is given in Table 5-1 and Table 5-2.

The SFRs can be classified into two sets: those associated with the “core” device functionality (ALU, Resets and interrupts) and those related to the peripheral functions. The Reset and Interrupt registers are described in their respective chapters, while the ALU’s STATUS register is described later in this section. Registers related to the operation of a peripheral feature are described in the chapter for that peripheral.

The SFRs are typically distributed among the peripherals whose functions they control. Unused SFR locations are unimplemented and read as ‘0’s.

TABLE 5-1: SPECIAL FUNCTION REGISTER MAP FOR PIC18F2420/2520/4420/4520 DEVICES

Address	Name	Address	Name	Address	Name	Address	Name
FFFh	TOSU	FDFh	INDF2 ⁽¹⁾	FBFh	CCPR1H	F9Fh	IPR1
FFEh	TOSH	FDEh	POSTINC2 ⁽¹⁾	FBEh	CCPR1L	F9Eh	PIR1
FFDh	TOSL	FDDh	POSTDEC2 ⁽¹⁾	FBDh	CCP1CON	F9Dh	PIE1
FFCh	STKPTR	FDCh	PREINC2 ⁽¹⁾	FBCh	CCPR2H	F9Ch	— ⁽²⁾
FFBh	PCLATU	FDBh	PLUSW2 ⁽¹⁾	FBHh	CCPR2L	F9Bh	OSCTUNE
FFAh	PCLATH	FDAh	FSR2H	FBAh	CCP2CON	F9Ah	— ⁽²⁾
FF9h	PCL	FD9h	FSR2L	FB9h	— ⁽²⁾	F99h	— ⁽²⁾
FF8h	TBLPTRU	FD8h	STATUS	FB8h	BAUDCON	F98h	— ⁽²⁾
FF7h	TBLPTRH	FD7h	TMR0H	FB7h	PWM1CON ⁽³⁾	F97h	— ⁽²⁾
FF6h	TBLPTRL	FD6h	TMR0L	FB6h	ECCP1AS ⁽³⁾	F96h	TRISE ⁽³⁾
FF5h	TABLAT	FD5h	T0CON	FB5h	CVRCON	F95h	TRISD ⁽³⁾
FF4h	PRODH	FD4h	— ⁽²⁾	FB4h	CMCON	F94h	TRISC
FF3h	PRODL	FD3h	OSCCON	FB3h	TMR3H	F93h	TRISB
FF2h	INTCON	FD2h	HLVDCON	FB2h	TMR3L	F92h	TRISA
FF1h	INTCON2	FD1h	WDTCON	FB1h	T3CON	F91h	— ⁽²⁾
FF0h	INTCON3	FD0h	RCON	FB0h	SPBRGH	F90h	— ⁽²⁾
FEFh	INDF0 ⁽¹⁾	FCFh	TMR1H	FAFh	SPBRG	F8Fh	— ⁽²⁾
FEEh	POSTINC0 ⁽¹⁾	FCEh	TMR1L	FAEh	RCREG	F8Eh	— ⁽²⁾
FEDh	POSTDEC0 ⁽¹⁾	FCDh	T1CON	FADh	TXREG	F8Dh	LATE ⁽³⁾
FECh	PREINC0 ⁽¹⁾	FCCh	TMR2	FACH	TXSTA	F8Ch	LATD ⁽³⁾
FEBh	PLUSW0 ⁽¹⁾	FCBh	PR2	FABh	RCSTA	F8Bh	LATC
FEAh	FSR0H	FCAh	T2CON	FAAh	— ⁽²⁾	F8Ah	LATB
FE9h	FSR0L	FC9h	SSPBUF	FA9h	EEADR	F89h	LATA
FE8h	WREG	FC8h	SSPADDD	FA8h	EEDATA	F88h	— ⁽²⁾
FE7h	INDF1 ⁽¹⁾	FC7h	SSPSTAT	FA7h	EECON2 ⁽¹⁾	F87h	— ⁽²⁾
FE6h	POSTINC1 ⁽¹⁾	FC6h	SSPCON1	FA6h	EECON1	F86h	— ⁽²⁾
FE5h	POSTDEC1 ⁽¹⁾	FC5h	SSPCON2	FA5h	— ⁽²⁾	F85h	— ⁽²⁾
FE4h	PREINC1 ⁽¹⁾	FC4h	ADRESH	FA4h	— ⁽²⁾	F84h	PORTE ⁽³⁾
FE3h	PLUSW1 ⁽¹⁾	FC3h	ADRESL	FA3h	— ⁽²⁾	F83h	PORTD ⁽³⁾
FE2h	FSR1H	FC2h	ADCON0	FA2h	IPR2	F82h	PORTC
FE1h	FSR1L	FC1h	ADCON1	FA1h	PIR2	F81h	PORTB
FE0h	BSR	FC0h	ADCON2	FA0h	PIE2	F80h	PORTA

- Note 1:** This is not a physical register.
2: Unimplemented registers are read as ‘0’.
3: This register is not available on 28-pin devices.

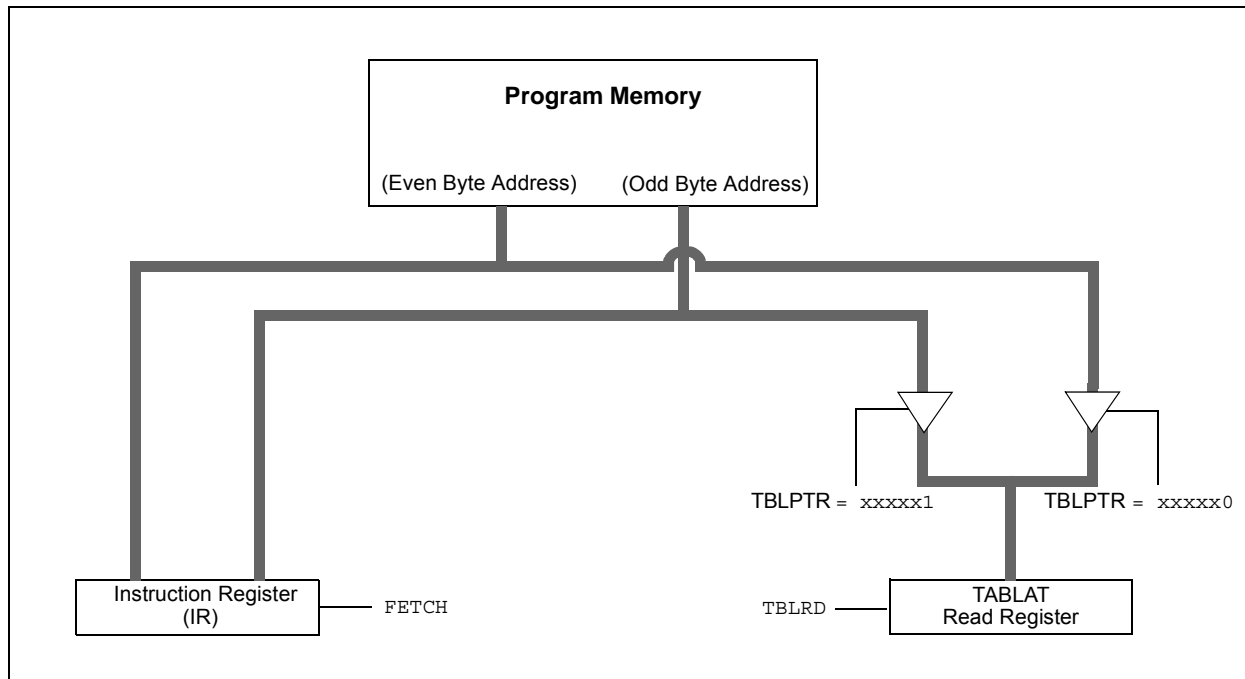
6.3 Reading the Flash Program Memory

The `TBLRD` instruction is used to retrieve data from program memory and places it into data RAM. Table reads from program memory are performed one byte at a time.

TBLPTR points to a byte address in program space. Executing `TBLRD` places the byte pointed to into `TABLAT`. In addition, TBLPTR can be modified automatically for the next table read operation.

The internal program memory is typically organized by words. The Least Significant bit of the address selects between the high and low bytes of the word. Figure 6-4 shows the interface between the internal program memory and the `TABLAT`.

FIGURE 6-4: READS FROM FLASH PROGRAM MEMORY



EXAMPLE 6-1: READING A FLASH PROGRAM MEMORY WORD

```

MOV LW CODE_ADDR_UPPER      ; Load TBLPTR with the base
MOV WF TBLPTRU              ; address of the word
MOV LW CODE_ADDR_HIGH
MOV WF TBLPTRH
MOV LW CODE_ADDR_LOW
MOV WF TBLPTRL

READ_WORD
TBLRD*+                    ; read into TABLAT and increment
MOV F TABLAT, W            ; get data
MOV WF WORD_EVEN
TBLRD*+                    ; read into TABLAT and increment
MOV FW TABLAT, W          ; get data
MOV F WORD_ODD
    
```

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7.6 Operation During Code-Protect

Data EEPROM memory has its own code-protect bits in Configuration Words. External read and write operations are disabled if code protection is enabled.

The microcontroller itself can both read and write to the internal data EEPROM, regardless of the state of the code-protect Configuration bit. Refer to **Section 23.0 “Special Features of the CPU”** for additional information.

7.7 Protection Against Spurious Write

There are conditions when the user may not want to write to the data EEPROM memory. To protect against spurious EEPROM writes, various mechanisms have been implemented. On power-up, the WREN bit is cleared. In addition, writes to the EEPROM are blocked during the Power-up Timer period (TPWRT, parameter 33).

The write initiate sequence and the WREN bit together help prevent an accidental write during brown-out, power glitch or software malfunction.

7.8 Using the Data EEPROM

The data EEPROM is a high-endurance, byte addressable array that has been optimized for the storage of frequently changing information (e.g., program variables or other data that are updated often). Frequently changing values will typically be updated more often than specification D124. If this is not the case, an array refresh must be performed. For this reason, variables that change infrequently (such as constants, IDs, calibration, etc.) should be stored in Flash program memory.

A simple data EEPROM refresh routine is shown in Example 7-3.

Note: If data EEPROM is only used to store constants and/or data that changes rarely, an array refresh is likely not required. See specification D124.

EXAMPLE 7-3: DATA EEPROM REFRESH ROUTINE

```
CLRF    EEADR      ; Start at address 0
BCF     EECON1, CFGS ; Set for memory
BCF     EECON1, EEPGD ; Set for Data EEPROM
BCF     INTCON, GIE  ; Disable interrupts
BSF     EECON1, WREN  ; Enable writes
Loop:   BSF     EECON1, RD      ; Loop to refresh array
        MOVLW  55h           ; Read current address
        MOVWF  EECON2         ; Write 55h
        MOVLW  0AAh          ;
        MOVWF  EECON2         ; Write 0AAh
        BSF     EECON1, WR     ; Set WR bit to begin write
        BTFSC  EECON1, WR     ; Wait for write to complete
        BRA    $-2
        INCF    EEADR, F      ; Increment address
        BRA    LOOP          ; Not zero, do it again

        BCF     EECON1, WREN  ; Disable writes
        BSF     INTCON, GIE    ; Enable interrupts
```


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REGISTER 9-9: IPR2: PERIPHERAL INTERRUPT PRIORITY REGISTER 2

R/W-1	R/W-1	U-0	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1
OSCFIP	CMIP	—	EEIP	BCLIP	HLVDIP	TMR3IP	CCP2IP
bit 7							bit 0

Legend:

R = Readable bit

W = Writable bit

U = Unimplemented bit, read as '0'

-n = Value at POR

'1' = Bit is set

'0' = Bit is cleared

x = Bit is unknown

bit 7 **OSCFIP:** Oscillator Fail Interrupt Priority bit

1 = High priority

0 = Low priority

bit 6 **CMIP:** Comparator Interrupt Priority bit

1 = High priority

0 = Low priority

bit 5 **Unimplemented:** Read as '0'

bit 4 **EEIP:** Data EEPROM/Flash Write Operation Interrupt Priority bit

1 = High priority

0 = Low priority

bit 3 **BCLIP:** Bus Collision Interrupt Priority bit

1 = High priority

0 = Low priority

bit 2 **HLVDIP:** High/Low-Voltage Detect Interrupt Priority bit

1 = High priority

0 = Low priority

bit 1 **TMR3IP:** TMR3 Overflow Interrupt Priority bit

1 = High priority

0 = Low priority

bit 0 **CCP2IP:** CCP2 Interrupt Priority bit

1 = High priority

0 = Low priority

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15.1 CCP Module Configuration

Each Capture/Compare/PWM module is associated with a control register (generically, CCPxCON) and a data register (CCPRx). The data register, in turn, is comprised of two 8-bit registers: CCPRxL (low byte) and CCPRxH (high byte). All registers are both readable and writable.

15.1.1 CCP MODULES AND TIMER RESOURCES

The CCP modules utilize Timers 1, 2 or 3, depending on the mode selected. Timer1 and Timer3 are available to modules in Capture or Compare modes, while Timer2 is available for modules in PWM mode.

TABLE 15-1: CCP MODE – TIMER RESOURCE

CCP/ECCP Mode	Timer Resource
Capture Compare PWM	Timer1 or Timer3 Timer1 or Timer3 Timer2

The assignment of a particular timer to a module is determined by the Timer to CCP enable bits in the T3CON register (Register 14-1). Both modules may be active at any given time and may share the same timer resource if they are configured to operate in the same mode (Capture/Compare or PWM) at the same time. The interactions between the two modules are summarized in Figure 15-1 and Figure 15-2. In Timer1 in Asynchronous Counter mode, the capture operation will not work.

15.1.2 CCP2 PIN ASSIGNMENT

The pin assignment for CCP2 (Capture input, Compare and PWM output) can change, based on device configuration. The CCP2MX Configuration bit determines which pin CCP2 is multiplexed to. By default, it is assigned to RC1 (CCP2MX = 1). If the Configuration bit is cleared, CCP2 is multiplexed with RB3.

Changing the pin assignment of CCP2 does not automatically change any requirements for configuring the port pin. Users must always verify that the appropriate TRIS register is configured correctly for CCP2 operation, regardless of where it is located.

TABLE 15-2: INTERACTIONS BETWEEN CCP1 AND CCP2 FOR TIMER RESOURCES

CCP1 Mode	CCP2 Mode	Interaction
Capture	Capture	Each module can use TMR1 or TMR3 as the time base. The time base can be different for each CCP.
Capture	Compare	CCP2 can be configured for the Special Event Trigger to reset TMR1 or TMR3 (depending upon which time base is used). Automatic A/D conversions on trigger event can also be done. Operation of CCP1 could be affected if it is using the same timer as a time base.
Compare	Capture	CCP1 can be configured for the Special Event Trigger to reset TMR1 or TMR3 (depending upon which time base is used). Operation of CCP2 could be affected if it is using the same timer as a time base.
Compare	Compare	Either module can be configured for the Special Event Trigger to reset the time base. Automatic A/D conversions on CCP2 trigger event can be done. Conflicts may occur if both modules are using the same time base.
Capture	PWM ⁽¹⁾	None
Compare	PWM ⁽¹⁾	None
PWM ⁽¹⁾	Capture	None
PWM ⁽¹⁾	Compare	None
PWM ⁽¹⁾	PWM ⁽¹⁾	Both PWMs will have the same frequency and update rate (TMR2 interrupt).

Note 1: Includes standard and Enhanced PWM operation.

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REGISTER 17-5: SSPCON2: MSSP CONTROL REGISTER 2 (I²C™ MODE)

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
GCEN	ACKSTAT	ACKDT ⁽²⁾	ACKEN ⁽¹⁾	RCEN ⁽¹⁾	PEN ⁽¹⁾	RSEN ⁽¹⁾	SEN ⁽¹⁾
bit 7							bit 0

Legend:

R = Readable bit W = Writable bit U = Unimplemented bit, read as '0'
-n = Value at POR '1' = Bit is set '0' = Bit is cleared x = Bit is unknown

- bit 7 **GCEN:** General Call Enable bit (Slave mode only)
1 = Enables interrupt when a general call address (0000h) is received in the SSPSR
0 = General call address disabled.
- bit 6 **ACKSTAT:** Acknowledge Status bit (Master Transmit mode only)
1 = Acknowledge was not received from slave
0 = Acknowledge was received from slave
- bit 5 **ACKDT:** Acknowledge Data bit (Master Receive mode only)⁽²⁾
1 = Not Acknowledge
0 = Acknowledge
- bit 4 **ACKEN:** Acknowledge Sequence Enable bit (Master Receive mode only)⁽¹⁾
1 = Initiates Acknowledge sequence on SDA and SCL pins and transmit ACKDT data bit. Automatically cleared by hardware.
0 = Acknowledge sequence Idle
- bit 3 **RCEN:** Receive Enable bit (Master mode only)⁽¹⁾
1 = Enables Receive mode for I²C
0 = Receive Idle
- bit 2 **PEN:** Stop Condition Enable bit (Master mode only)⁽¹⁾
1 = Initiates Stop condition on SDA and SCL pins. Automatically cleared by hardware.
0 = Stop condition Idle
- bit 1 **RSEN:** Repeated Start Condition Enable bit (Master mode only)⁽¹⁾
1 = Initiates Repeated Start condition on SDA and SCL pins. Automatically cleared by hardware.
0 = Repeated Start condition Idle
- bit 0 **SEN:** Start Condition Enable/Stretch Enable bit⁽¹⁾
In Master mode:
1 = Initiates Start condition on SDA and SCL pins. Automatically cleared by hardware.
0 = Start condition Idle
In Slave mode:
1 = Clock stretching is enabled for both slave transmit and slave receive (stretch enabled)
0 = Clock stretching is disabled

Note 1: For bits ACKEN, RCEN, PEN, RSEN, SEN: If the I²C module is not in the Idle mode, these bits may not be set (no spooling) and the SSPBUF may not be written (or writes to the SSPBUF are disabled).

2: Value that will be transmitted when the user initiates an Acknowledge sequence at the end of a receive.

18.0 ENHANCED UNIVERSAL SYNCHRONOUS ASYNCHRONOUS RECEIVER TRANSMITTER (EUSART)

The Enhanced Universal Synchronous Asynchronous Receiver Transmitter (EUSART) module is one of the two serial I/O modules. (Generically, the USART is also known as a Serial Communications Interface or SCI.) The EUSART can be configured as a full-duplex asynchronous system that can communicate with peripheral devices, such as CRT terminals and personal computers. It can also be configured as a half-duplex, synchronous system that can communicate with peripheral devices, such as A/D or D/A integrated circuits, serial EEPROMs, etc.

The Enhanced USART module implements additional features, including automatic baud rate detection and calibration, automatic wake-up on Sync Break reception and 12-bit Break character transmit. These make it ideally suited for use in Local Interconnect Network bus (LIN bus) systems.

The EUSART can be configured in the following modes:

- Asynchronous (full duplex) with:
 - Auto-wake-up on character reception
 - Auto-baud calibration
 - 12-bit Break character transmission
- Synchronous – Master (half duplex) with selectable clock polarity
- Synchronous – Slave (half duplex) with selectable clock polarity

The pins of the Enhanced USART are multiplexed with PORTC. In order to configure RC6/TX/CK and RC7/RX/DT as an EUSART:

- bit SPEN (RCSTA<7>) must be set (= 1)
- bit TRISC<7> must be set (= 1)
- bit TRISC<6> must be set (= 1)

Note: The EUSART control will automatically reconfigure the pin from input to output as needed.
--

The operation of the Enhanced USART module is controlled through three registers:

- Transmit Status and Control (TXSTA)
- Receive Status and Control (RCSTA)
- Baud Rate Control (BAUDCON)

These are detailed on the following pages in Register 18-1, Register 18-2 and Register 18-3, respectively.

19.6 A/D Conversions

Figure 19-4 shows the operation of the A/D Converter after the $\text{GO}/\overline{\text{DONE}}$ bit has been set and the $\text{ACQT}<2:0>$ bits are cleared. A conversion is started after the following instruction to allow entry into Sleep mode before the conversion begins.

Figure 19-5 shows the operation of the A/D Converter after the $\text{GO}/\overline{\text{DONE}}$ bit has been set and the $\text{ACQT}<2:0>$ bits are set to '010', and selecting a 4 TAD acquisition time before the conversion starts.

Clearing the $\text{GO}/\overline{\text{DONE}}$ bit during a conversion will abort the current conversion. The A/D Result register pair will NOT be updated with the partially completed A/D conversion sample. This means the $\text{ADRESH}:\text{ADRESL}$ registers will continue to contain the value of the last completed conversion (or the last value written to the $\text{ADRESH}:\text{ADRESL}$ registers).

After the A/D conversion is completed or aborted, a 2 TAD wait is required before the next acquisition can be started. After this wait, acquisition on the selected channel is automatically started.

Note: The $\text{GO}/\overline{\text{DONE}}$ bit should **NOT** be set in the same instruction that turns on the A/D.

19.7 Discharge

The discharge phase is used to initialize the value of the capacitor array. The array is discharged before every sample. This feature helps to optimize the unity-gain amplifier, as the circuit always needs to charge the capacitor array, rather than charge/discharge based on previous measure values.

FIGURE 19-4: A/D CONVERSION TAD CYCLES ($\text{ACQT}<2:0> = 000$, $\text{TACQ} = 0$)

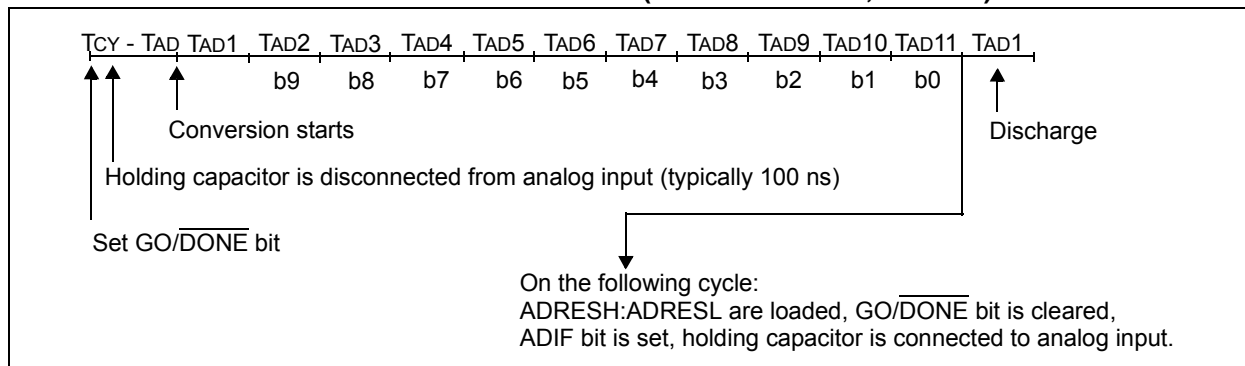
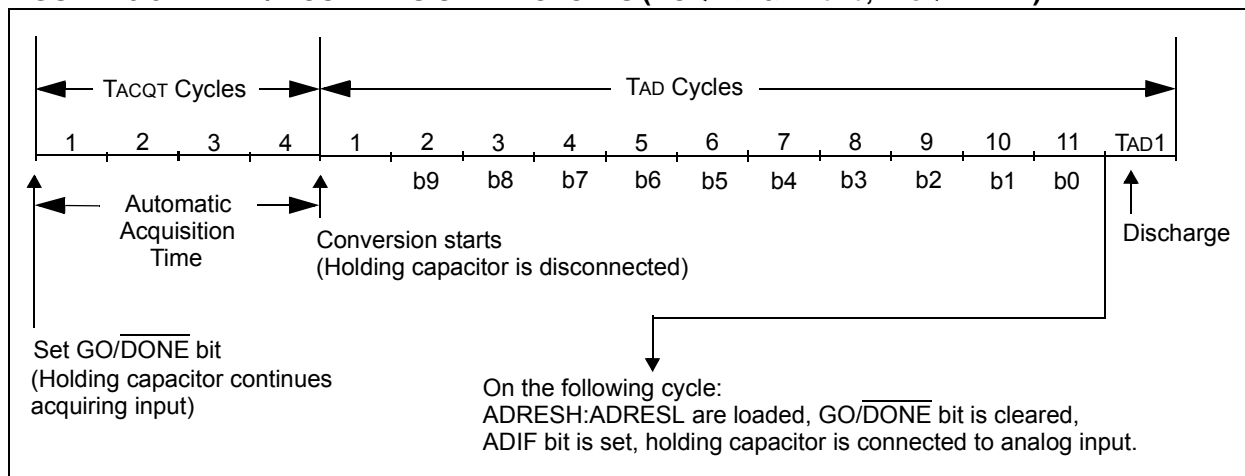


FIGURE 19-5: A/D CONVERSION TAD CYCLES ($\text{ACQT}<2:0> = 010$, $\text{TACQ} = 4 \text{ TAD}$)



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23.2 Watchdog Timer (WDT)

For PIC18F2420/2520/4420/4520 devices, the WDT is driven by the INTRC source. When the WDT is enabled, the clock source is also enabled. The nominal WDT period is 4 ms and has the same stability as the INTRC oscillator.

The 4 ms period of the WDT is multiplied by a 16-bit postscaler. Any output of the WDT postscaler is selected by a multiplexer, controlled by bits in Configuration Register 2H. Available periods range from 4 ms to 131.072 seconds (2.18 minutes). The WDT and postscaler are cleared when any of the following events occur: a `SLEEP` or `CLRWDT` instruction is executed, the IRCF bits (`OSCCON<6:4>`) are changed or a clock failure has occurred.

Note 1: The `CLRWDT` and `SLEEP` instructions clear the WDT and postscaler counts when executed.

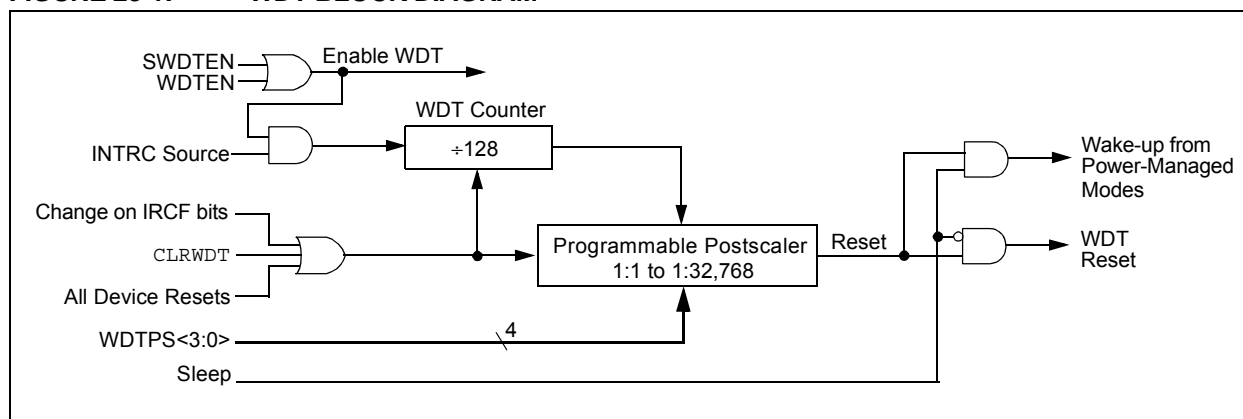
2: Changing the setting of the IRCF bits (`OSCCON<6:4>`) clears the WDT and postscaler counts.

3: When a `CLRWDT` instruction is executed, the postscaler count will be cleared.

23.2.1 CONTROL REGISTER

Register 23-14 shows the WDTCON register. This is a readable and writable register which contains a control bit that allows software to override the WDT enable Configuration bit, but only if the Configuration bit has disabled the WDT.

FIGURE 23-1: WDT BLOCK DIAGRAM



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23.5.1 PROGRAM MEMORY CODE PROTECTION

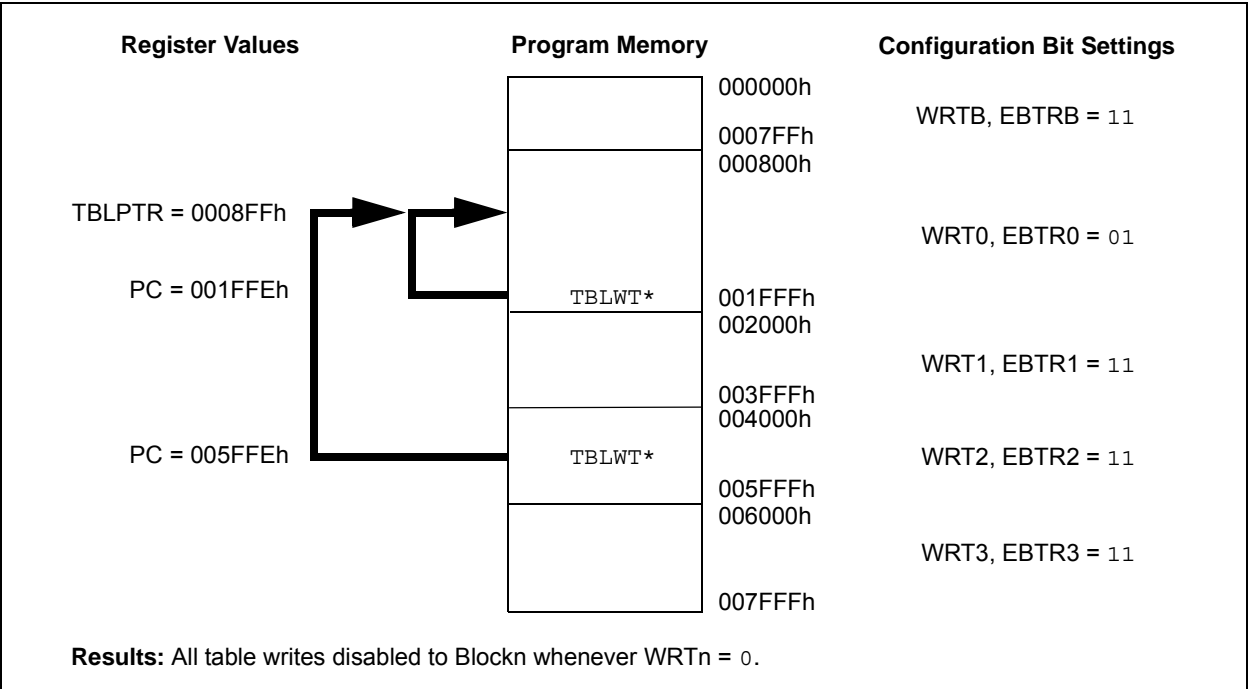
The program memory may be read to, or written from, any location using the table read and table write instructions. The Device ID may be read with table reads. The Configuration registers may be read and written with the table read and table write instructions.

In normal execution mode, the CPn bits have no direct effect. CPn bits inhibit external reads and writes. A block of user memory may be protected from table writes if the WRTn Configuration bit is '0'. The EBTRn bits control table reads. For a block of user memory with the EBTRn bit set to '0', a table read instruction that executes from within that block is allowed to read. A table read

instruction that executes from a location outside of that block is not allowed to read and will result in reading '0's. Figures 23-6 through 23-8 illustrate table write and table read protection.

Note: Code protection bits may only be written to a '0' from a '1' state. It is not possible to write a '1' to a bit in the '0' state. Code protection bits are only set to '1' by a full chip erase or block erase function. The full chip erase and block erase functions can only be initiated via ICSP or an external programmer.

FIGURE 23-6: TABLE WRITE (WRTn) DISALLOWED



PIC18F2420/2520/4420/4520

24.1.1 STANDARD INSTRUCTION SET

ADDLW ADD Literal to W

Syntax:	ADDLW	k								
Operands:	$0 \leq k \leq 255$									
Operation:	$(W) + k \rightarrow W$									
Status Affected:	N, OV, C, DC, Z									
Encoding:	<table border="1"><tr><td>0000</td><td>1111</td><td>kkkk</td><td>kkkk</td></tr></table>		0000	1111	kkkk	kkkk				
0000	1111	kkkk	kkkk							
Description:	The contents of W are added to the 8-bit literal 'k' and the result is placed in W.									
Words:	1									
Cycles:	1									
Q Cycle Activity:	<table><tr><th>Q1</th><th>Q2</th><th>Q3</th><th>Q4</th></tr><tr><td>Decode</td><td>Read literal 'k'</td><td>Process Data</td><td>Write to W</td></tr></table>		Q1	Q2	Q3	Q4	Decode	Read literal 'k'	Process Data	Write to W
Q1	Q2	Q3	Q4							
Decode	Read literal 'k'	Process Data	Write to W							

Example: ADDLW 15h

Before Instruction
W = 10h
After Instruction
W = 25h

ADDWF ADD W to f

Syntax:	ADDWF f {,d {,a}}				
Operands:	$0 \leq f \leq 255$ $d \in [0,1]$ $a \in [0,1]$				
Operation:	$(W) + (f) \rightarrow \text{dest}$				
Status Affected:	N, OV, C, DC, Z				
Encoding:	<table border="1"><tr><td>0010</td><td>01da</td><td>ffff</td><td>ffff</td></tr></table>	0010	01da	ffff	ffff
0010	01da	ffff	ffff		
Description:	<p>Add W to register 'f'. If 'd' is '0', the result is stored in W. If 'd' is '1', the result is stored back in register 'f' (default).</p> <p>If 'a' is '0', the Access Bank is selected. If 'a' is '1', the BSR is used to select the GPR bank (default).</p> <p>If 'a' is '0' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever $f \leq 95$ (5Fh). See Section 24.2.3 “Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode” for details.</p>				
Words:	1				
Cycles:	1				

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	Read register 'f'	Process Data	Write to destination

Example: ADDWF REG, 0, 0

Before Instruction
W = 17h
REG = 0C2h
After Instruction
W = 0D9h
REG = 0C2h

Note: All PIC18 instructions may take an optional label argument preceding the instruction mnemonic for use in symbolic addressing. If a label is used, the instruction format then becomes: {label} instruction argument(s).

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IORLW Inclusive OR Literal with W

Syntax: IORLW k

Operands: $0 \leq k \leq 255$

Operation: $(W) .OR. k \rightarrow W$

Status Affected: N, Z

Encoding:

0000	1001	kkkk	kkkk
------	------	------	------

Description: The contents of W are ORed with the 8-bit literal 'k'. The result is placed in W.

Words: 1

Cycles: 1

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	Read literal 'k'	Process Data	Write to W

Example: IORLW 35h

Before Instruction

W = 9Ah

After Instruction

W = BFh

IORWF Inclusive OR W with f

Syntax: IORWF f {,d {,a}}

Operands: $0 \leq f \leq 255$
 $d \in [0,1]$
 $a \in [0,1]$

Operation: $(W) .OR. (f) \rightarrow \text{dest}$

Status Affected: N, Z

Encoding:

0001	00da	ffff	ffff
------	------	------	------

Description: Inclusive OR W with register 'f'. If 'd' is '0', the result is placed in W. If 'd' is '1', the result is placed back in register 'f' (default).
If 'a' is '0', the Access Bank is selected. If 'a' is '1', the BSR is used to select the GPR bank (default).
If 'a' is '0' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever $f \leq 95$ (5Fh). See **Section 24.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode"** for details.

Words: 1

Cycles: 1

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	Read register 'f'	Process Data	Write to destination

Example: IORWF RESULT, 0, 1

Before Instruction

RESULT = 13h

W = 91h

After Instruction

RESULT = 13h

W = 93h

PIC18F2420/2520/4420/4520

26.2 DC Characteristics: Power-Down and Supply Current PIC18F2420/2520/4420/4520 (Industrial) PIC18LF2420/2520/4420/4520 (Industrial) (Continued)

PIC18LF2420/2520/4420/4520 (Industrial)		Standard Operating Conditions (unless otherwise stated) Operating temperature $-40^{\circ}\text{C} \leq T_A \leq +85^{\circ}\text{C}$ for industrial					
PIC18F2420/2520/4420/4520 (Industrial, Extended)		Standard Operating Conditions (unless otherwise stated) Operating temperature $-40^{\circ}\text{C} \leq T_A \leq +85^{\circ}\text{C}$ for industrial $-40^{\circ}\text{C} \leq T_A \leq +125^{\circ}\text{C}$ for extended					
Param No.	Device	Typ	Max	Units	Conditions		
	Supply Current (I_{DD}) ⁽²⁾						
	PIC18LF2X2X/4X20	165	250	μA	-40°C	$V_{DD} = 2.0\text{V}$	FOSC = 1 MHz (RC_IDLE mode, INTOSC source)
		175	250	μA	$+25^{\circ}\text{C}$		
		190	270	μA	$+85^{\circ}\text{C}$		
	PIC18LF2X2X/4X20	250	360	μA	-40°C	$V_{DD} = 3.0\text{V}$	
		270	360	μA	$+25^{\circ}\text{C}$		
		290	380	μA	$+85^{\circ}\text{C}$		
	All devices	500	700	μA	-40°C	$V_{DD} = 5.0\text{V}$	
		520	700	μA	$+25^{\circ}\text{C}$		
		550	700	μA	$+85^{\circ}\text{C}$		
	Extended devices only	0.6	1	mA	$+125^{\circ}\text{C}$		
	PIC18LF2X2X/4X20	340	500	μA	-40°C	$V_{DD} = 2.0\text{V}$	FOSC = 4 MHz (RC_IDLE mode, INTOSC source)
		350	500	μA	$+25^{\circ}\text{C}$		
		360	500	μA	$+85^{\circ}\text{C}$		
	PIC18LF2X2X/4X20	520	800	μA	-40°C	$V_{DD} = 3.0\text{V}$	
		540	800	μA	$+25^{\circ}\text{C}$		
		580	850	μA	$+85^{\circ}\text{C}$		
	All devices	1.0	1.6	mA	-40°C	$V_{DD} = 5.0\text{V}$	
		1.1	1.4	mA	$+25^{\circ}\text{C}$		
		1.1	1.4	mA	$+85^{\circ}\text{C}$		
Extended devices only	1.1	2.0	mA	$+125^{\circ}\text{C}$			

Legend: Shading of rows is to assist in readability of the table.

Note 1: The power-down current in Sleep mode does not depend on the oscillator type. Power-down current is measured with the part in Sleep mode, with all I/O pins in high-impedance state and tied to V_{DD} or V_{SS} and all features that add delta current disabled (such as WDT, Timer1 Oscillator, BOR, etc.).

2: The supply current is mainly a function of operating voltage, frequency and mode. Other factors, such as I/O pin loading and switching rate, oscillator type and circuit, internal code execution pattern and temperature, also have an impact on the current consumption.

The test conditions for all I_{DD} measurements in active operation mode are:

OSC1 = external square wave, from rail-to-rail; all I/O pins tri-stated, pulled to V_{DD} or V_{SS} ;

MCLR = V_{DD} ; WDT enabled/disabled as specified.

3: When operation below -10°C is expected, use T1OSC High-Power mode, where LPT1OSC (CONFIG3H<2>) = 0. When operation will always be above -10°C , then the low-power Timer1 oscillator may be selected.

4: BOR and HLVD enable internal band gap reference. With both modules enabled, current consumption will be less than the sum of both specifications.

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