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What is "Embedded - Microcontrollers"?

"Embedded - Microcontrollers" refer to small, integrated circuits designed to perform specific tasks within larger systems. These microcontrollers are essentially compact computers on a single chip, containing a processor core, memory, and programmable input/output peripherals. They are called "embedded" because they are embedded within electronic devices to control various functions, rather than serving as standalone computers. Microcontrollers are crucial in modern electronics, providing the intelligence and control needed for a wide range of applications.

Applications of "Embedded - Microcontrollers"

Details	
Product Status	Active
Core Processor	8051
Core Size	8-Bit
Speed	50MHz
Connectivity	EBI/EMI, I²C, LINbus, SPI, UART/USART
Peripherals	CapSense, DMA, POR, PWM, WDT
Number of I/O	62
Program Memory Size	32KB (32K x 8)
Program Memory Type	FLASH
EEPROM Size	1K x 8
RAM Size	4K x 8
Voltage - Supply (Vcc/Vdd)	1.71V ~ 5.5V
Data Converters	A/D 16x12b; D/A 2x8b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	100-LQFP
Supplier Device Package	100-TQFP (14x14)
Purchase URL	https://www.e-xfl.com/product-detail/infineon-technologies/cy8c3445axi-104t

Contents

The device provides a PLL to generate clock frequencies up to 50 MHz from the IMO, external crystal, or external reference clock. It also contains a separate, very low-power Internal low-speed oscillator (ILO) for the sleep and watchdog timers. A 32.768-kHz external watch crystal is also supported for use in RTC applications. The clocks, together with programmable clock dividers, provide the flexibility to integrate most timing requirements.

The CY8C34 family supports a wide supply operating range from 1.71 V to 5.5 V. This allows operation from regulated supplies such as 1.8 V ± 5 percent, 2.5 V ±10 percent, 3.3 V ± 10 percent, or 5.0 V ± 10 percent, or directly from a wide range of battery types. In addition, it provides an integrated high efficiency synchronous boost converter that can power the device from supply voltages as low as 0.5 V. This enables the device to be powered directly from a single battery or solar cell. In addition, you can use the boost converter to generate other voltages required by the device, such as a 3.3-V supply for LCD glass drive. The boost's output is available on the V_{BOOST} pin, allowing other devices in the application to be powered from the PSoC.

PSoC supports a wide range of low-power modes. These include a 200-nA hibernate mode with RAM retention and a 1-µA sleep mode with RTC. In the second mode the optional 32.768-kHz watch crystal runs continuously and maintains an accurate RTC.

Power to all major functional blocks, including the programmable digital and analog peripherals, can be controlled independently by firmware. This allows low-power background processing when some peripherals are not in use. This, in turn, provides a total device current of only 1.2 mA when the CPU is running at 6 MHz, or 0.8 mA running at 3 MHz.

The details of the PSoC power modes are covered in the "Power System" section on page 30 of this data sheet.

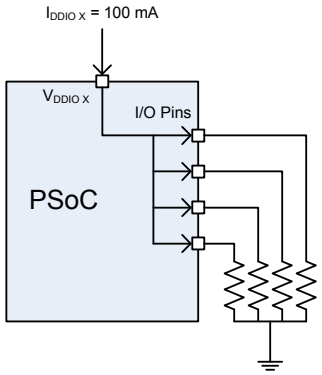
PSoC uses JTAG (4-wire) or SWD (2-wire) interfaces for programming, debug, and test. The 1-wire SWV may also be used for 'printf' style debugging. By combining SWD and SWV, you can implement a full debugging interface with just three pins. Using these standard interfaces enables you to debug or program the PSoC with a variety of hardware solutions from Cypress or third party vendors. PSoC supports on-chip break points and 4 KB instruction and data race memory for debug. Details of the programming, test, and debugging interfaces are discussed in the "Programming, Debug Interfaces, Resources" section on page 65 of this data sheet.

2. Pinouts

Each VDDIO pin powers a specific set of I/O pins. (The USBIOs are powered from VDDD.) Using the VDDIO pins, a single PSoC can support multiple voltage levels, reducing the need for off-chip level shifters. The black lines drawn on the pinout diagrams in Figure 2-3 through Figure 2-6, as well as Table 2-1, show the pins that are powered by each VDDIO.

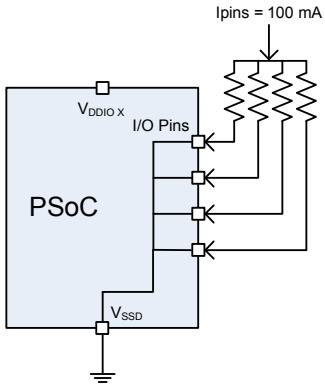
Each VDDIO may source up to 100 mA total to its associated I/O pins, as shown in Figure 2-1.

Figure 2-1. VDDIO Current Limit



Conversely, for the 100-pin and 68-pin devices, the set of I/O pins associated with any VDDIO may sink up to 100 mA total, as shown in Figure 2-2.

Figure 2-2. I/O Pins Current Limit



For the 48-pin devices, the set of I/O pins associated with VDDIO0 plus VDDIO2 may sink up to 100 mA total. The set of I/O pins associated with VDDIO1 plus VDDIO3 may sink up to a total of 100 mA.

4.3.1.2 Logical Instructions

The logical instructions perform Boolean operations such as AND, OR, XOR on bytes, rotate of accumulator contents, and swap of nibbles in an accumulator. The Boolean operations on the bytes are performed on the bit-by-bit basis. Table 4-2 shows the list of logical instructions and their description.

Table 4-2. Logical Instructions

Mnemonic	Description	Bytes	Cycles
ANL A,Rn	AND register to accumulator	1	1
ANL A,Direct	AND direct byte to accumulator	2	2
ANL A,@Ri	AND indirect RAM to accumulator	1	2
ANL A,#data	AND immediate data to accumulator	2	2
ANL Direct, A	AND accumulator to direct byte	2	3
ANL Direct, #data	AND immediate data to direct byte	3	3
ORL A,Rn	OR register to accumulator	1	1
ORL A,Direct	OR direct byte to accumulator	2	2
ORL A,@Ri	OR indirect RAM to accumulator	1	2
ORL A,#data	OR immediate data to accumulator	2	2
ORL Direct, A	OR accumulator to direct byte	2	3
ORL Direct, #data	OR immediate data to direct byte	3	3
XRL A,Rn	XOR register to accumulator	1	1
XRL A,Direct	XOR direct byte to accumulator	2	2
XRL A,@Ri	XOR indirect RAM to accumulator	1	2
XRL A,#data	XOR immediate data to accumulator	2	2
XRL Direct, A	XOR accumulator to direct byte	2	3
XRL Direct, #data	XOR immediate data to direct byte	3	3
CLR A	Clear accumulator	1	1
CPL A	Complement accumulator	1	1
RL A	Rotate accumulator left	1	1
RLC A	Rotate accumulator left through carry	1	1
RR A	Rotate accumulator right	1	1
RRC A	Rotate accumulator right though carry	1	1
SWAP A	Swap nibbles within accumulator	1	1

Table 4-4. Boolean Instructions

Mnemonic	Description	Bytes	Cycles
CLR C	Clear carry	1	1
CLR bit	Clear direct bit	2	3
SETB C	Set carry	1	1
SETB bit	Set direct bit	2	3
CPL C	Complement carry	1	1
CPL bit	Complement direct bit	2	3
ANL C, bit	AND direct bit to carry	2	2
ANL C, /bit	AND complement of direct bit to carry	2	2
ORL C, bit	OR direct bit to carry	2	2
ORL C, /bit	OR complement of direct bit to carry	2	2
MOV C, bit	Move direct bit to carry	2	2
MOV bit, C	Move carry to direct bit	2	3
JC rel	Jump if carry is set	2	3
JNC rel	Jump if no carry is set	2	3
JB bit, rel	Jump if direct bit is set	3	5
JNB bit, rel	Jump if direct bit is not set	3	5
JBC bit, rel	Jump if direct bit is set and clear bit	3	5

4.3.1.5 Program Branching Instructions

The 8051 supports a set of conditional and unconditional jump instructions that help to modify the program execution flow. Table 4-5 shows the list of jump instructions.

Table 4-5. Jump Instructions

Mnemonic	Description	Bytes	Cycles
ACALL addr11	Absolute subroutine call	2	4
LCALL addr16	Long subroutine call	3	4
RET	Return from subroutine	1	4
RETI	Return from interrupt	1	4
AJMP addr11	Absolute jump	2	3
LJMP addr16	Long jump	3	4
SJMP rel	Short jump (relative address)	2	3
JMP @A + DPTR	Jump indirect relative to DPTR	1	5
JZ rel	Jump if accumulator is zero	2	4
JNZ rel	Jump if accumulator is nonzero	2	4
CJNE A,Direct, rel	Compare direct byte to accumulator and jump if not equal	3	5
CJNE A, #data, rel	Compare immediate data to accumulator and jump if not equal	3	4
CJNE Rn, #data, rel	Compare immediate data to register and jump if not equal	3	4
CJNE @Ri, #data, rel	Compare immediate data to indirect RAM and jump if not equal	3	5
DJNZ Rn,rel	Decrement register and jump if not zero	2	4
DJNZ Direct, rel	Decrement direct byte and jump if not zero	3	5
NOP	No operation	1	1

debug on chip mode. It can be used to generate periodic interrupts for timing purposes or to wake the system from a low-power mode. Firmware can reset the central timewheel. Systems that require accurate timing should use the RTC capability instead of the central timewheel.

The 100-kHz clock (CLK100K) can be used as a low power master clock. It can also generate time intervals using the fast timewheel.

The fast timewheel is a 5-bit counter, clocked by the 100-kHz clock. It features programmable settings and automatically resets when the terminal count is reached. An optional interrupt can be generated each time the terminal count is reached. This enables flexible, periodic interrupts of the CPU at a higher rate than is allowed using the central timewheel.

The 33 kHz clock (CLK33K) comes from a divide-by-3 operation on CLK100K. This output can be used as a reduced accuracy version of the 32.768-kHz ECO clock with no need for a crystal.

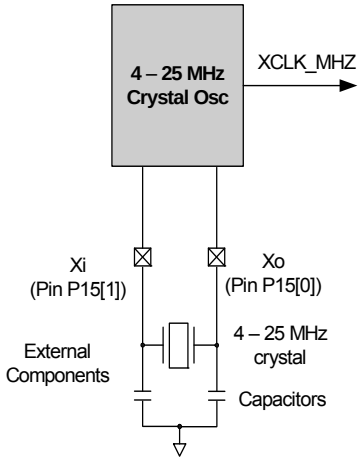
6.1.2 External Oscillators

Figure 6-1 shows that there are two external oscillators. They can be routed directly or divided. The direct routes may not have a 50% duty cycle. Divided clocks have a 50% duty cycle.

6.1.2.1 MHz External Crystal Oscillator

The MHzECO provides high frequency, high precision clocking using an external crystal (see Figure 6-2). It supports a wide variety of crystal types, in the range of 4 to 25 MHz. When used in conjunction with the PLL, it can generate other clocks up to the device's maximum frequency (see "Phase-Locked Loop" section on page 28). The GPIO pins connecting to the external crystal and capacitors are fixed. MHzECO accuracy depends on the crystal chosen.

Figure 6-2. MHzECO Block Diagram

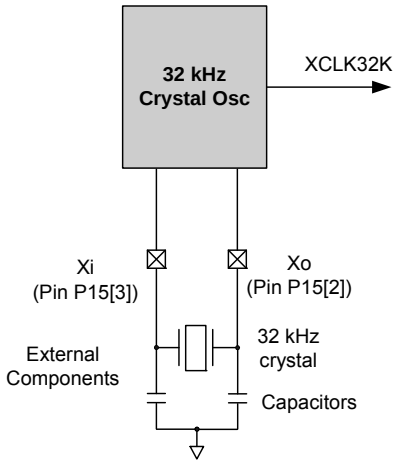


6.1.2.2 32.768-kHz ECO

The 32.768-kHz External Crystal Oscillator (32kHzECO) provides precision timing with minimal power consumption using an external 32.768-kHz watch crystal (see Figure 6-3). The 32kHzECO also connects directly to the sleep timer and provides the source for the RTC. The RTC uses a 1-second interrupt to implement the RTC functionality in firmware.

The oscillator works in two distinct power modes. This allows users to trade off power consumption with noise immunity from neighboring circuits. The GPIO pins connected to the external crystal and capacitors are fixed.

Figure 6-3. 32kHzECO Block Diagram



It is recommended that the external 32.768-kHz watch crystal have a load capacitance (CL) of 6 pF or 12.5 pF. Check the crystal manufacturer's datasheet. The two external capacitors, CL1 and CL2, are typically of the same value, and their total capacitance, CL1CL2 / (CL1 + CL2), including pin and trace capacitance, should equal the crystal CL value. For more information, refer to application note AN54439: PSoC 3 and PSoC 5 External Oscillators. See also pin capacitance specifications in the "GPIO" section on page 79.

6.1.2.3 Digital System Interconnect

The DSI provides routing for clocks taken from external clock oscillators connected to I/O. The oscillators can also be generated within the device in the digital system and Universal Digital Blocks.

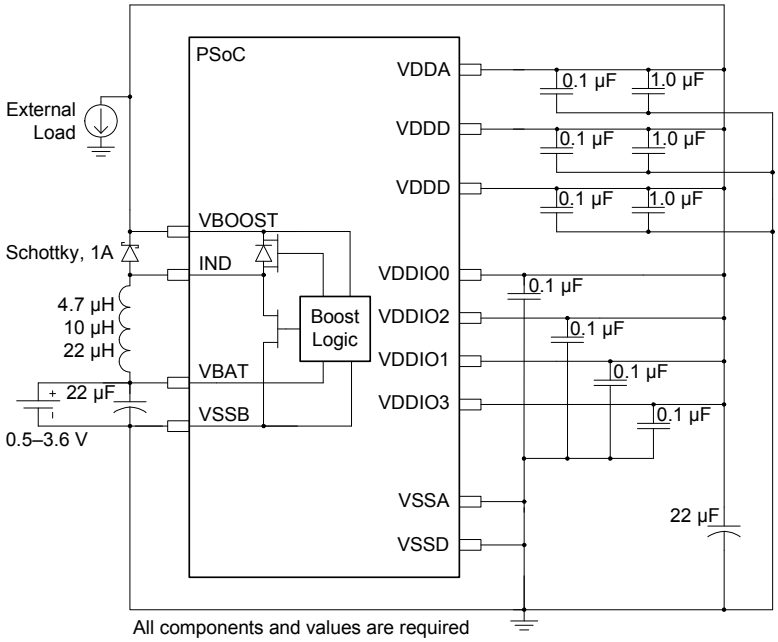
While the primary DSI clock input provides access to all clocking resources, up to eight other DSI clocks (internally or externally generated) may be routed directly to the eight digital clock dividers. This is only possible if there are multiple precision clock sources.

6.1.3 Clock Distribution

All seven clock sources are inputs to the central clock distribution system. The distribution system is designed to create multiple high precision clocks. These clocks are customized for the design's requirements and eliminate the common problems found with limited resolution prescalers attached to peripherals. The clock distribution system generates several types of clock trees.

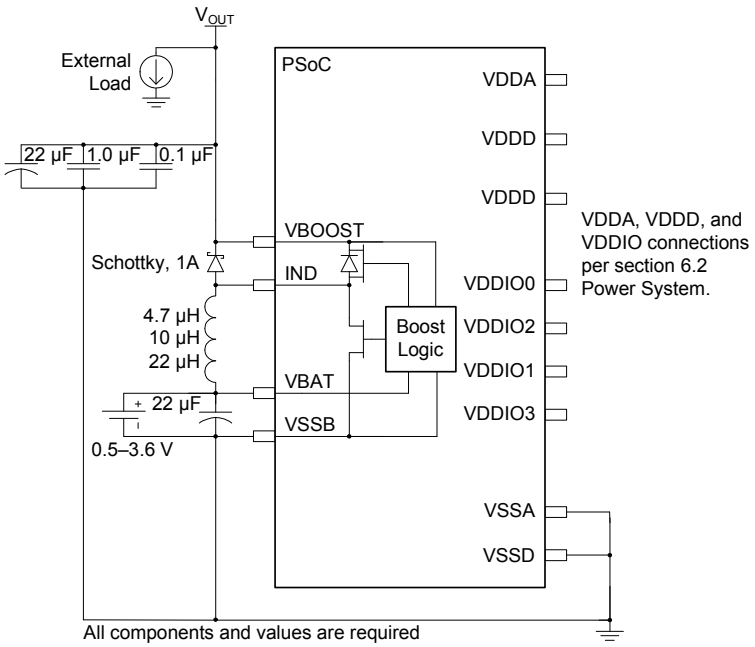
- The master clock is used to select and supply the fastest clock in the system for general clock requirements and clock synchronization of the PSoC device.
- Bus Clock 16-bit divider uses the master clock to generate the bus clock used for data transfers. Bus clock is the source clock for the CPU clock divider.
- Eight fully programmable 16-bit clock dividers generate digital system clocks for general use in the digital system, as configured by the design's requirements. Digital system clocks

Figure 6-6. Application of Boost Converter powering PSoC device



The boost converter may also generate a supply that is not used directly by the PSoC device. An example of this use case is boosting a 1.8 V supply to 4.0 V to drive a white LED. If the boost converter is not supplying the PSoC devices V_{DDA}, V_{DDD}, and V_{DDIO} it must comply with the same design rules as supplying the PSoC device, but with a change to the bulk capacitor requirements. A parallel arrangement 22 µF, 1.0 µF, and 0.1 µF capacitors are all required on the Vout supply and must be placed within 1 cm of the VBOOST pin to ensure regulator stability.

Figure 6-7. Application of Boost Converter not powering PSoC device



The switching frequency is set to 400 kHz using an oscillator integrated into the boost converter. The boost converter can be operated in two different modes: active and standby. Active mode is the normal mode of operation where the boost regulator actively generates a regulated output voltage. In standby mode, most boost functions are disabled, thus reducing power consumption of the boost circuit. Only minimal power is provided, typically < 5 µA to power the PSoC device in Sleep mode. The

boost typically draws 250 µA in active mode and 25 µA in standby mode. The boost operating modes must be used in conjunction with chip power modes to minimize total power consumption. Table 6-4 lists the boost power modes available in different chip power modes.

Table 6-4. Chip and Boost Power Modes Compatibility

Chip Power Modes	Boost Power Modes
Chip-active or alternate active mode	Boost must be operated in its active mode.
Chip-sleep mode	Boost can be operated in either active or standby mode. In boost standby mode, the chip must wake up periodically for boost active-mode refresh.
Chip-hibernate mode	Boost can be operated in its active mode. However, it is recommended not to use the boost in chip hibernate mode due to the higher current consumption in boost active mode.

6.2.2.1 Boost Firmware Requirements

To ensure boost inrush current is within specification at startup, the **Enable Fast IMO During Startup** value must be unchecked in the PSoC Creator IDE. The **Enable Fast IMO During Startup** option is found in PSoC Creator in the design wide resources (cydwr) file **System** tab. Un-checking this option configures the device to run at 12 MHz vs 48 MHz during startup while configuring the device. The slower clock speed results in reduced current draw through the boost circuit.

6.2.2.2 Boost Design Process

Correct operation of the boost converter requires specific component values determined for each designs unique operating conditions. The C_{BAT} capacitor, Inductor, Schottky diode, and C_{BOOST} capacitor components are required with the values specified in the electrical specifications, Table 11-7 on page 77. The only variable component value is the inductor L_{BOOST} which is primarily sized for correct operation of the boost across operating conditions and secondarily for efficiency. Additional operating region constraints exist for V_{OUT}, V_{BAT}, I_{OUT}, and T_A.

The following steps must be followed to determine boost converter operating parameters and L_{BOOST} value.

1. Choose desired V_{BAT}, V_{OUT}, T_A, and I_{OUT} operating condition ranges for the application.
2. Determine if V_{BAT} and V_{OUT} ranges fit the boost operating range based on the **T_A range over V_{BAT} and V_{OUT}** chart, Figure 11-8 on page 77. If the operating ranges are not met, modify the operating conditions or use an external boost regulator.
3. Determine if the desired ambient temperature (T_A) range fits the ambient temperature operating range based on the **T_A range over V_{BAT} and V_{OUT}** chart, Figure 11-8 on page 77. If the temperature range is not met, modify the operating conditions and return to step 2, or use an external boost regulator.
4. Determine if the desired output current (I_{OUT}) range fits the output current operating range based on the **I_{OUT} range over V_{BAT} and V_{OUT}** chart, Figure 11-9 on page 77. If the output current range is not met, modify the operating conditions and return to step 2, or use an external boost regulator.
5. Find the allowed inductor values based on the **L_{BOOST} values over V_{BAT} and V_{OUT}** chart, Figure 11-10 on page 77.
6. Based on the allowed inductor values, inductor dimensions, inductor cost, boost efficiency, and V_{RIPPLE} choose the optimum inductor value for the system. Boost efficiency and V_{RIPPLE} typical values are provided in the **Efficiency vs V_{BAT}** and **V_{RIPPLE} vs V_{BAT}** charts, Figure 11-11 on page 78 through Figure 11-14 on page 78. In general, if high efficiency and low V_{RIPPLE} are most important, then the highest allowed inductor value should be used. If low inductor cost or small inductor size are most important, then one of the smaller allowed inductor values should be used. If the allowed inductor(s) efficiency, V_{RIPPLE}, cost or dimensions are not acceptable for the application than an external boost regulator should be used.

6.3 Reset

CY8C34 has multiple internal and external reset sources available. The reset sources are:

- Power source monitoring – The analog and digital power voltages, VDDA, VDDD, VCCA, and VCCD are monitored in several different modes during power up, active mode, and sleep mode (buzzing). If any of the voltages goes outside predetermined ranges then a reset is generated. The monitors are programmable to generate an interrupt to the processor under certain conditions before reaching the reset thresholds.
- External – The device can be reset from an external source by pulling the reset pin ($\overline{\text{XRES}}$) low. The $\overline{\text{XRES}}$ pin includes an internal pull-up to VDDIO1. VDDD, VDDA, and VDDIO1 must all have voltage applied before the part comes out of reset.
- Watchdog timer – A watchdog timer monitors the execution of instructions by the processor. If the watchdog timer is not reset by firmware within a certain period of time, the watchdog timer generates a reset.
- Software – The device can be reset under program control.

6.4.19 JTAG Boundary Scan

The device supports standard JTAG boundary scan chains on all I/O pins for board level test.

7. Digital Subsystem

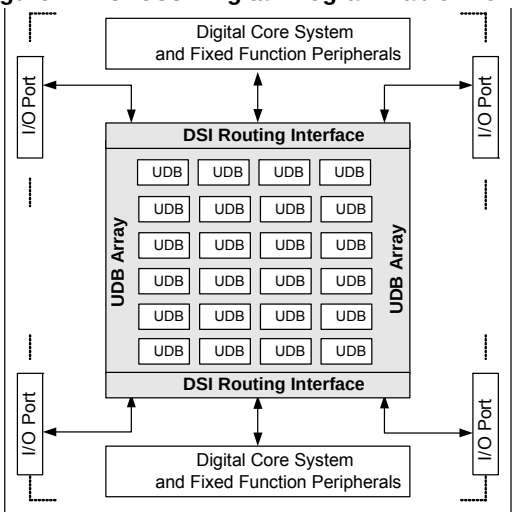
The digital programmable system creates application specific combinations of both standard and advanced digital peripherals and custom logic functions. These peripherals and logic are then interconnected to each other and to any pin on the device, providing a high level of design flexibility and IP security.

The features of the digital programmable system are outlined here to provide an overview of capabilities and architecture. You do not need to interact directly with the programmable digital system at the hardware and register level. PSoC Creator provides a high level schematic capture graphical interface to automatically place and route resources similar to PLDs.

The main components of the digital programmable system are:

- Universal Digital Blocks (UDB) – These form the core functionality of the digital programmable system. UDBs are a collection of uncommitted logic (PLD) and structural logic (Datapath) optimized to create all common embedded peripherals and customized functionality that are application or design specific.
- Universal Digital Block Array – UDB blocks are arrayed within a matrix of programmable interconnect. The UDB array structure is homogeneous and allows for flexible mapping of digital functions onto the array. The array supports extensive and flexible routing interconnects between UDBs and the Digital System Interconnect.
- Digital System Interconnect (DSI) – Digital signals from Universal Digital Blocks (UDBs), fixed function peripherals, I/O pins, interrupts, DMA, and other system core signals are attached to the Digital System Interconnect to implement full featured device connectivity. The DSI allows any digital function to any pin or other feature routability when used with the Universal Digital Block Array.

Figure 7-1. CY8C34 Digital Programmable Architecture



7.1 Example Peripherals

The flexibility of the CY8C34 family's Universal Digital Blocks (UDBs) and Analog Blocks allow the user to create a wide range of components (peripherals). The most common peripherals were built and characterized by Cypress and are shown in the PSoC Creator component catalog, however, users may also create their own custom components using PSoC Creator. Using PSoC Creator, users may also create their own components for reuse within their organization, for example sensor interfaces, proprietary algorithms, and display interfaces.

The number of components available through PSoC Creator is too numerous to list in the data sheet, and the list is always growing. An example of a component available for use in CY8C34 family, but, not explicitly called out in this data sheet is the UART component.

7.1.1 Example Digital Components

The following is a sample of the digital components available in PSoC Creator for the CY8C34 family. The exact amount of hardware resources (UDBs, routing, RAM, flash) used by a component varies with the features selected in PSoC Creator for the component.

- Communications
  - I^2C
  - UART
  - SPI
- Functions
  - EMIF
  - PWMs
  - Timers
  - Counters
- Logic
  - NOT
  - OR
  - XOR
  - AND

7.1.2 Example Analog Components

The following is a sample of the analog components available in PSoC Creator for the CY8C34 family. The exact amount of hardware resources (SC/CT blocks, routing, RAM, flash) used by a component varies with the features selected in PSoC Creator for the component.

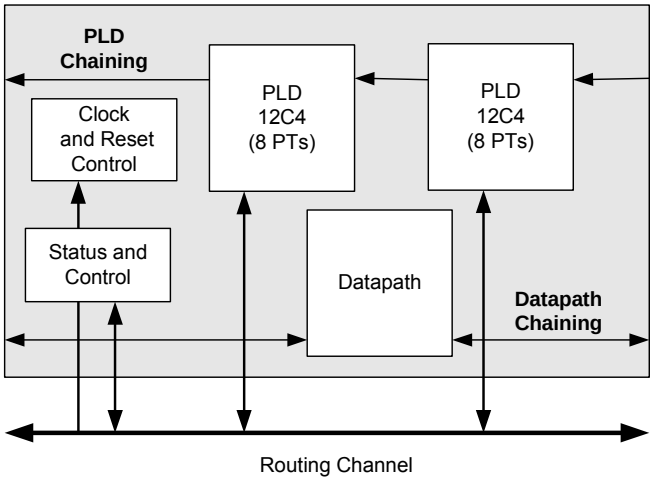
- Amplifiers
  - TIA
  - PGA
  - opamp
- ADC
  - Delta-Sigma
- DACs
  - Current
  - Voltage
  - PWM
- Comparators
- Mixers

7.2 Universal Digital Block

The Universal Digital Block (UDB) represents an evolutionary step to the next generation of PSoC embedded digital peripheral functionality. The architecture in first generation PSoC digital blocks provides coarse programmability in which a few fixed functions with a small number of options are available. The new UDB architecture is the optimal balance between configuration granularity and efficient implementation. A cornerstone of this approach is to provide the ability to customize the devices digital operation to match application requirements.

To achieve this, UDBs consist of a combination of uncommitted logic (PLD), structured logic (Datapath), and a flexible routing scheme to provide interconnect between these elements, I/O connections, and other peripherals. UDB functionality ranges from simple self contained functions that are implemented in one UDB, or even a portion of a UDB (unused resources are available for other functions), to more complex functions that require multiple UDBs. Examples of basic functions are timers, counters, CRC generators, PWMs, dead band generators, and communications functions, such as UARTs, SPI, and I^2C. Also, the PLD blocks and connectivity provide full featured general purpose programmable logic within the limits of the available resources.

Figure 7-2. UDB Block Diagram



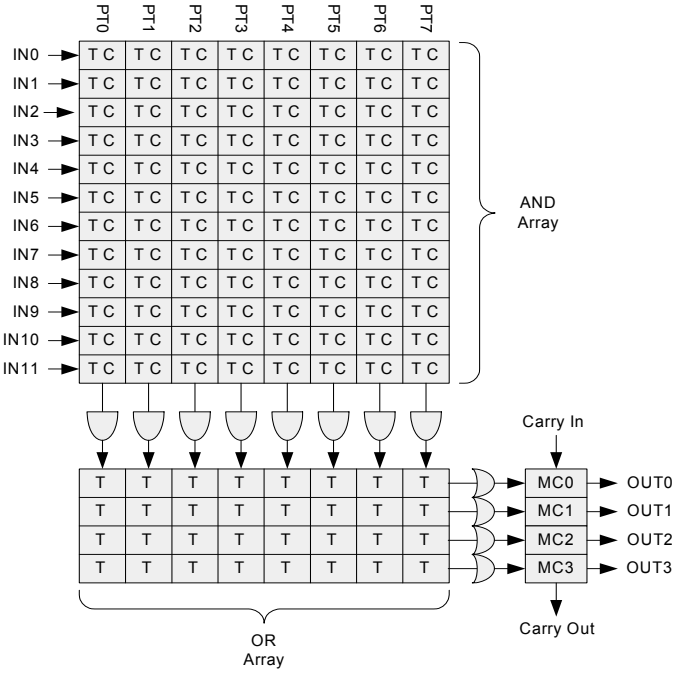
The main component blocks of the UDB are:

- PLD blocks – There are two small PLDs per UDB. These blocks take inputs from the routing array and form registered or combinational sum-of-products logic. PLDs are used to implement state machines, state bits, and combinational logic equations. PLD configuration is automatically generated from graphical primitives.
- Datapath Module – This 8-bit wide datapath contains structured logic to implement a dynamically configurable ALU, a variety of compare configurations and condition generation. This block also contains input/output FIFOs, which are the primary parallel data interface between the CPU/DMA system and the UDB.
- Status and Control Module – The primary role of this block is to provide a way for CPU firmware to interact and synchronize with UDB operation.
- Clock and Reset Module – This block provides the UDB clocks and reset selection and control.

7.2.1 PLD Module

The primary purpose of the PLD blocks is to implement logic expressions, state machines, sequencers, lookup tables, and decoders. In the simplest use model, consider the PLD blocks as a standalone resource onto which general purpose RTL is synthesized and mapped. The more common and efficient use model is to create digital functions from a combination of PLD and datapath blocks, where the PLD implements only the random logic and state portion of the function while the datapath (ALU) implements the more structured elements.

Figure 7-3. PLD 12C4 Structure



One 12C4 PLD block is shown in Figure 7-3. This PLD has 12 inputs, which feed across eight product terms. Each product term (AND function) can be from 1 to 12 inputs wide, and in a given product term, the true (T) or complement (C) of each input can be selected. The product terms are summed (OR function) to create the PLD outputs. A sum can be from 1 to 8 product terms wide. The 'C' in 12C4 indicates that the width of the OR gate (in this case 8) is constant across all outputs (rather than variable as in a 22V10 device). This PLA like structure gives maximum flexibility and insures that all inputs and outputs are permutable for ease of allocation by the software tools. There are two 12C4 PLDs in each UDB.

Figure 7-8. Function Mapping Example in a Bank of UDBs

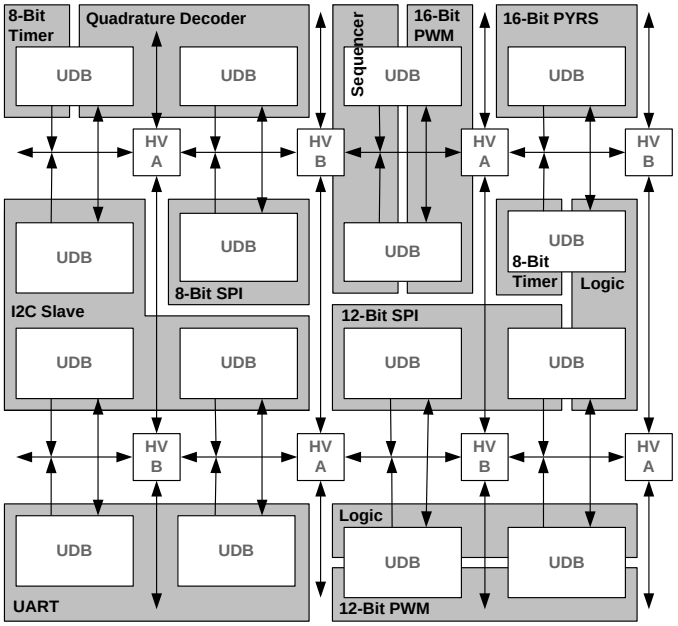
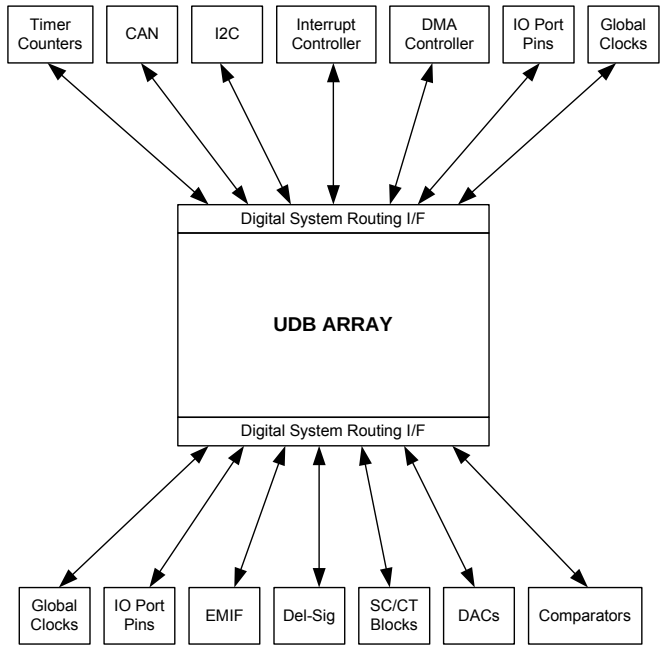


Figure 7-9. Digital System Interconnect



7.4 DSI Routing Interface Description

The DSI routing interface is a continuation of the horizontal and vertical routing channels at the top and bottom of the UDB array core. It provides general purpose programmable routing between device peripherals, including UDBs, I/Os, analog peripherals, interrupts, DMA and fixed function peripherals.

Figure 7-9 illustrates the concept of the digital system interconnect, which connects the UDB array routing matrix with other device peripherals. Any digital core or fixed function peripheral that needs programmable routing is connected to this interface.

Signals in this category include:

- Interrupt requests from all digital peripherals in the system.
- DMA requests from all digital peripherals in the system.
- Digital peripheral data signals that need flexible routing to I/Os.
- Digital peripheral data signals that need connections to UDBs.
- Connections to the interrupt and DMA controllers.
- Connection to I/O pins.
- Connection to analog system digital signals.

Interrupt and DMA routing is very flexible in the CY8C34 programmable architecture. In addition to the numerous fixed function peripherals that can generate interrupt requests, any data signal in the UDB array routing can also be used to generate a request. A single peripheral may generate multiple independent interrupt requests simplifying system and firmware design. Figure 7-10 shows the structure of the IDMUX (Interrupt/DMA Multiplexer).

Figure 7-10. Interrupt and DMA Processing in the IDMUX

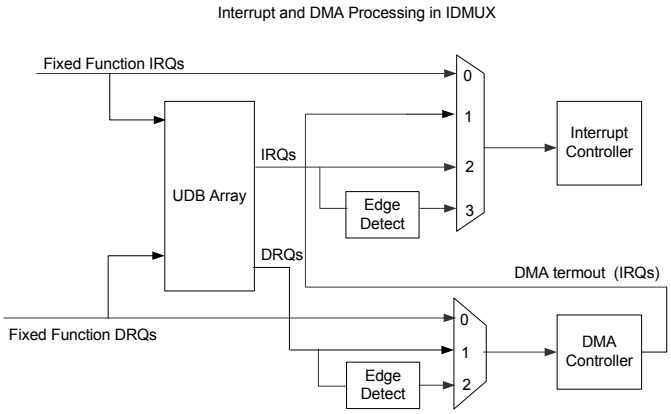
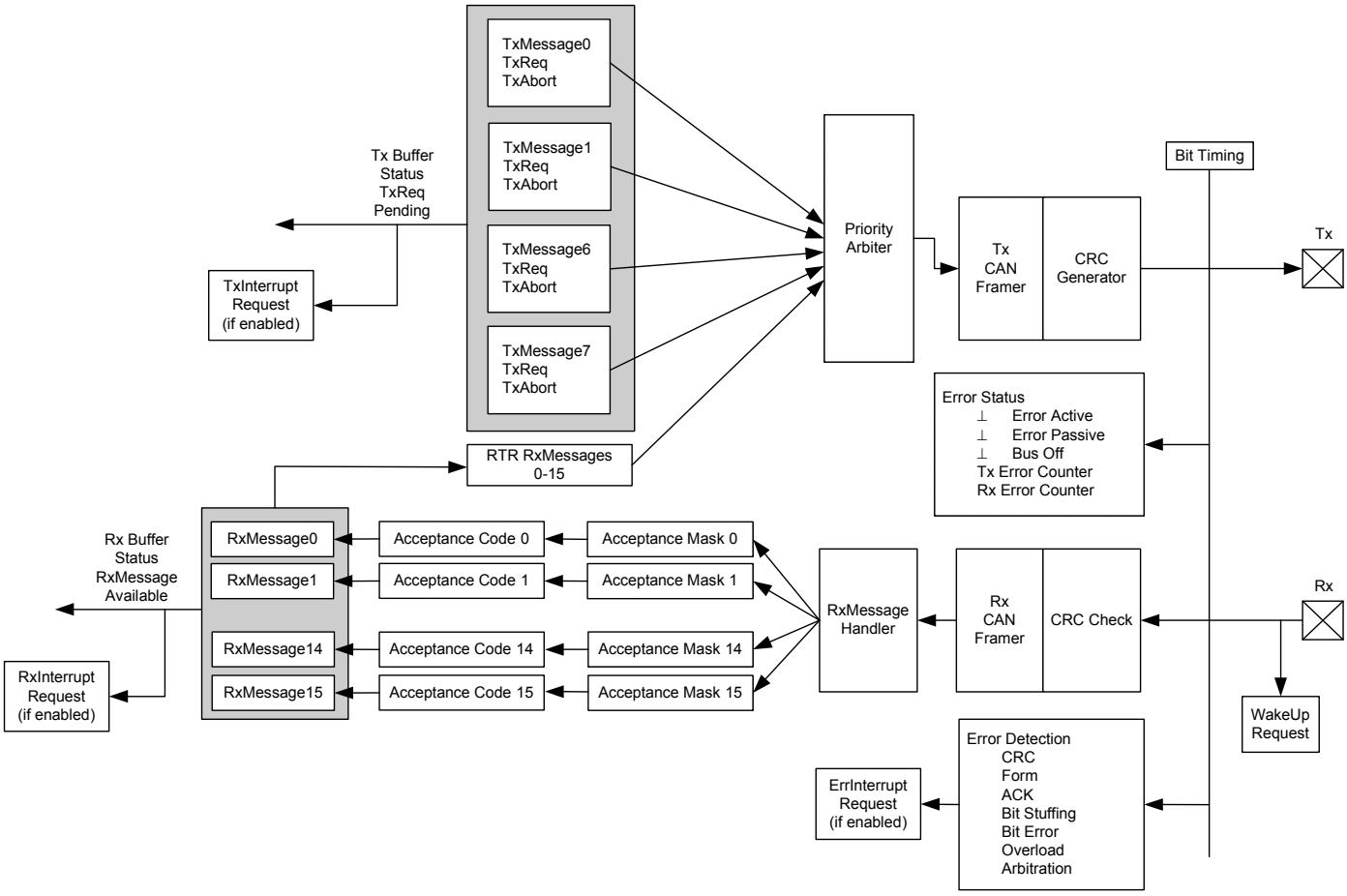


Figure 7-15. CAN Controller Block Diagram



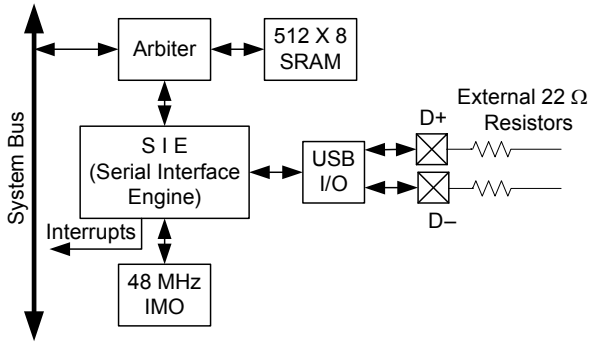
7.6 USB

PSoC includes a dedicated Full-Speed (12 Mbps) USB 2.0 transceiver supporting all four USB transfer types: control, interrupt, bulk, and isochronous. PSoC Creator provides full configuration support. USB interfaces to hosts through two dedicated USBIO pins, which are detailed in the "I/O System and Routing" section on page 36.

USB includes the following features:

- Eight unidirectional data endpoints
- One bidirectional control endpoint 0 (EP0)
- Shared 512-byte buffer for the eight data endpoints
- Dedicated 8-byte buffer for EP0
- Three memory modes
  - Manual Memory Management with No DMA Access
  - Manual Memory Management with Manual DMA Access
  - Automatic Memory Management with Automatic DMA Access
- Internal 3.3-V regulator for transceiver
- Internal 48 MHz main oscillator mode that auto locks to USB bus clock, requiring no external crystal for USB (USB equipped parts only)
- Interrupts on bus and each endpoint event, with device wakeup
- USB Reset, Suspend, and Resume operations
- Bus powered and self powered modes

Figure 7-16. USB



The TIA configuration is used for applications where an external sensor's output is current as a function of some type of stimulus such as temperature, light, magnetic flux etc. In a common application, the voltage DAC output can be connected to the V_{REF} TIA input to allow calibration of the external sensor bias current by adjusting the voltage DAC output voltage.

8.6 LCD Direct Drive

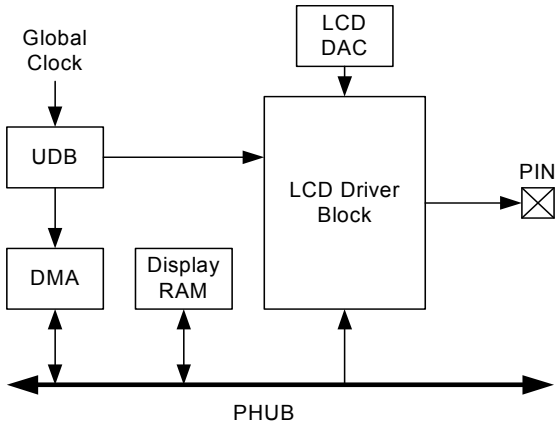
The PSoC Liquid Crystal Display (LCD) driver system is a highly configurable peripheral designed to allow PSoC to directly drive a broad range of LCD glass. All voltages are generated on chip, eliminating the need for external components. With a high multiplex ratio of up to 1/16, the CY8C34 family LCD driver system can drive a maximum of 736 segments. The PSoC LCD driver module was also designed with the conservative power budget of portable devices in mind, enabling different LCD drive modes and power down modes to conserve power.

PSoC Creator provides an LCD segment drive component. The component wizard provides easy and flexible configuration of LCD resources. You can specify pins for segments and commons along with other options. The software configures the device to meet the required specifications. This is possible because of the programmability inherent to PSoC devices.

Key features of the PSoC LCD segment system are:

- LCD panel direct driving
- Type A (standard) and Type B (low-power) waveform support
- Wide operating voltage range support (2 V to 5 V) for LCD panels
- Static, 1/2, 1/3, 1/4, 1/5 bias voltage levels
- Internal bias voltage generation through internal resistor ladder
- Up to 62 total common and segment outputs
- Up to 1/16 multiplex for a maximum of 16 backplane/common outputs
- Up to 62 front plane/segment outputs for direct drive
- Drives up to 736 total segments (16 backplane × 46 front plane)
- Up to 64 levels of software controlled contrast
- Ability to move display data from memory buffer to LCD driver through DMA (without CPU intervention)
- Adjustable LCD refresh rate from 10 Hz to 150 Hz
- Ability to invert LCD display for negative image
- Three LCD driver drive modes, allowing power optimization

Figure 8-10. LCD System



8.6.1 LCD Segment Pin Driver

Each GPIO pin contains an LCD driver circuit. The LCD driver buffers the appropriate output of the LCD DAC to directly drive the glass of the LCD. A register setting determines whether the pin is a common or segment. The pin's LCD driver then selects one of the six bias voltages to drive the I/O pin, as appropriate for the display data.

8.6.2 Display Data Flow

The LCD segment driver system reads display data and generates the proper output voltages to the LCD glass to produce the desired image. Display data resides in a memory buffer in the system SRAM. Each time you need to change the common and segment driver voltages, the next set of pixel data moves from the memory buffer into the Port Data Registers via DMA.

8.6.3 UDB and LCD Segment Control

A UDB is configured to generate the global LCD control signals and clocking. This set of signals is routed to each LCD pin driver through a set of dedicated LCD global routing channels. In addition to generating the global LCD control signals, the UDB also produces a DMA request to initiate the transfer of the next frame of LCD data.

8.6.4 LCD DAC

The LCD DAC generates the contrast control and bias voltage for the LCD system. The LCD DAC produces up to five LCD drive voltages plus ground, based on the selected bias ratio. The bias voltages are driven out to GPIO pins on a dedicated LCD bias bus, as required.

Table 11-2. DC Specifications *(continued)*

Parameter	Description	Conditions		Min	Typ[25]	Max	Units
	Sleep Mode[28]						
	CPU = OFF RTC = ON (= ECO32K ON, in low-power mode) Sleep timer = ON (= ILO ON at 1 kHz)[29] WDT = OFF I2C Wake = OFF Comparator = OFF POR = ON Boost = OFF SIO pins in single ended input, unregulated output mode	$V_{DD} = V_{DDIO}$ = 4.5 V - 5.5 V	T = –40 °C	–	1.1	2.3	µA
			T = 25 °C	–	1.1	2.2	
			T = 85 °C	–	15	30	
		$V_{DD} = V_{DDIO}$ = 2.7 V – 3.6 V	T = –40 °C	–	1	2.2	
			T = 25 °C	–	1	2.1	
			T = 85 °C	–	12	28	
		$V_{DD} = V_{DDIO}$ = 1.71 V – 1.95 V[30]	T = 25 °C	–	2.2	4.2	
	Comparator = ON CPU = OFF RTC = OFF Sleep timer = OFF WDT = OFF I2C Wake = OFF POR = ON Boost = OFF SIO pins in single ended input, unregulated output mode	$V_{DD} = V_{DDIO}$ = 2.7 V – 3.6 V[31]	T = 25 °C	–	2.2	2.7	
	I2C Wake = ON CPU = OFF RTC = OFF Sleep timer = OFF WDT = OFF Comparator = OFF POR = ON Boost = OFF SIO pins in single ended input, unregulated output mode	$V_{DD} = V_{DDIO}$ = 2.7 V – 3.6 V[31]	T = 25 °C	–	2.2	2.8	
	Hibernate Mode[28]						
	Hibernate mode current All regulators and oscillators off SRAM retention GPIO interrupts are active Boost = OFF SIO pins in single ended input, unregulated output mode	$V_{DD} = V_{DDIO}$ = 4.5 V - 5.5 V	T = –40 °C	–	0.2	1.5	µA
			T = 25 °C	–	0.5	1.5	
			T = 85 °C	–	4.1	5.3	
		$V_{DD} = V_{DDIO}$ = 2.7 V – 3.6 V	T = –40 °C	–	0.2	1.5	
			T = 25 °C	–	0.2	1.5	
			T = 85 °C	–	3.2	4.2	
		$V_{DD} = V_{DDIO}$ = 1.71 V – 1.95 V[30]	T = –40 °C	–	0.2	1.5	
			T = 25 °C	–	0.3	1.5	
			T = 85 °C	–	3.3	4.3	
I_{DDAR}	Analog current consumption while device is reset[32]	$V_{DDA} \leq 3.6$ V		–	0.3	0.6	mA
		$V_{DDA} > 3.6$ V		–	1.4	3.3	mA
I_{DDDR}	Digital current consumption while device is reset[32]	$V_{DDD} \leq 3.6$ V		–	1.1	3.1	mA
		$V_{DDD} > 3.6$ V		–	0.7	3.1	mA

Figure 11-1. Active Mode Current vs F_{CPU}, V_{DD} = 3.3 V, Temperature = 25 °C

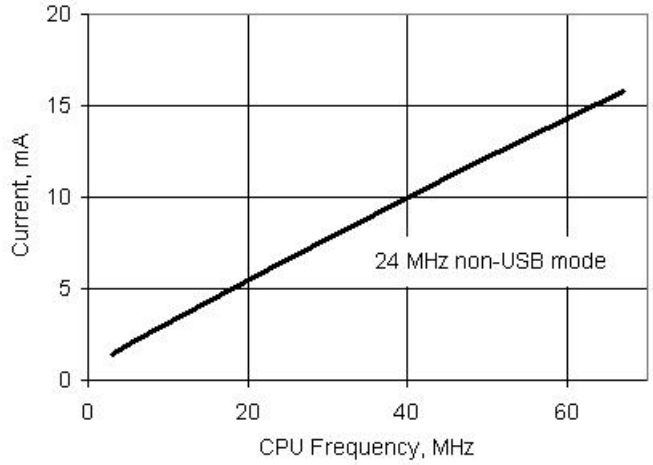


Figure 11-2. Active Mode Current vs Temperature and F_{CPU}, V_{DD} = 3.3 V

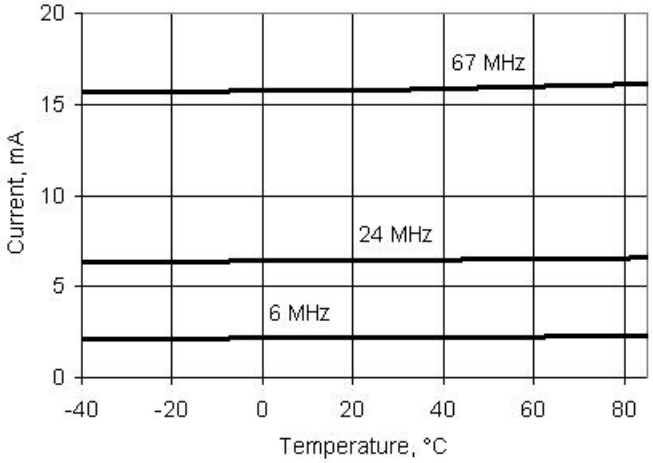
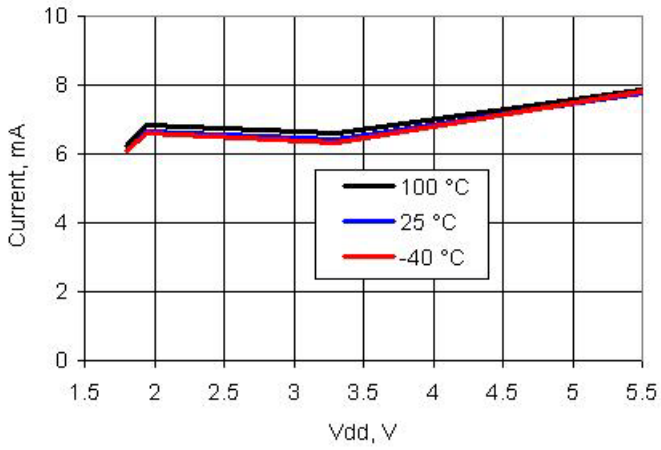


Figure 11-3. Active Mode Current vs V_{DD} and Temperature, F_{CPU} = 24 MHz



Notes
28. If V_{CCD} and V_{CCA} are externally regulated, the voltage difference between V_{CCD} and V_{CCA} must be less than 50 mV.
29. Sleep timer generates periodic interrupts to wake up the CPU. This specification applies only to those times that the CPU is off.
30. Externally regulated mode.
31. Based on device characterization (not production tested).
32. Based on device characterization (not production tested). USBIO pins tied to ground (VSSD).

Table 11-7. Recommended External Components for Boost Circuit

Parameter	Description	Conditions	Min	Typ	Max	Units
L_{BOOST}	Boost inductor	4.7 µH nominal	3.7	4.7	5.7	µH
		10 µH nominal	8.0	10.0	12.0	µH
		22 µH nominal	17.0	22.0	27.0	µH
C_{BOOST}	Total capacitance sum of V_{DDD}, V_{DDA}, V_{DDIO}[37]		17.0	26.0	31.0	µF
C_{BAT}	Battery filter capacitor		17.0	22.0	27.0	µF
I_F	Schottky diode average forward current		1.0	–	–	A
V_R	Schottky reverse voltage		20.0	–	–	V

Figure 11-8. T_A range over V_{BAT} and V_{OUT}

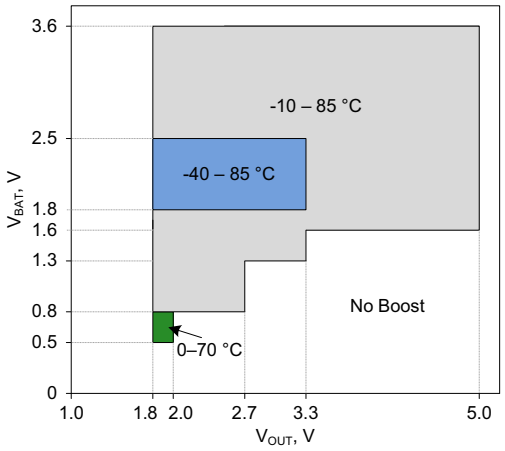


Figure 11-9. I_{OUT} range over V_{BAT} and V_{OUT}

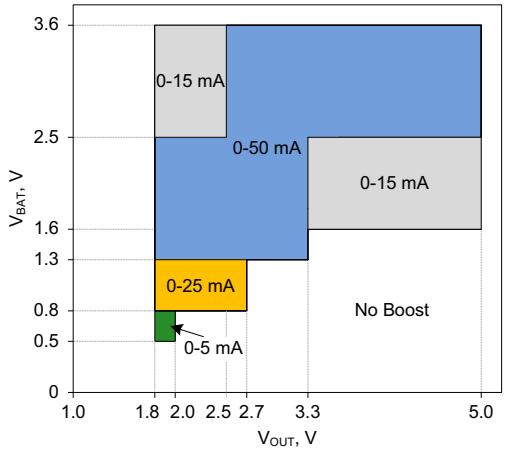
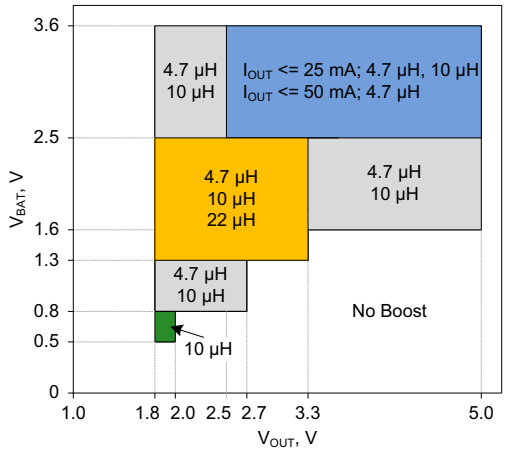


Figure 11-10. L_{BOOST} values over V_{BAT} and V_{OUT}



Note
37. Based on device characterization (Not production tested).

Figure 11-40. IDAC Full Scale Error vs Temperature, Range = 255 µA, Source Mode

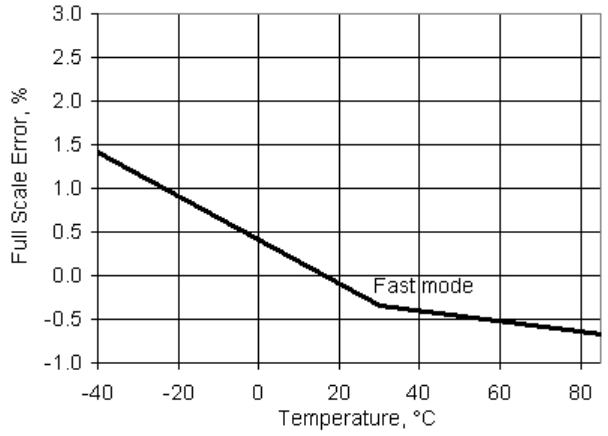


Figure 11-41. IDAC Full Scale Error vs Temperature, Range = 255 µA, Sink Mode

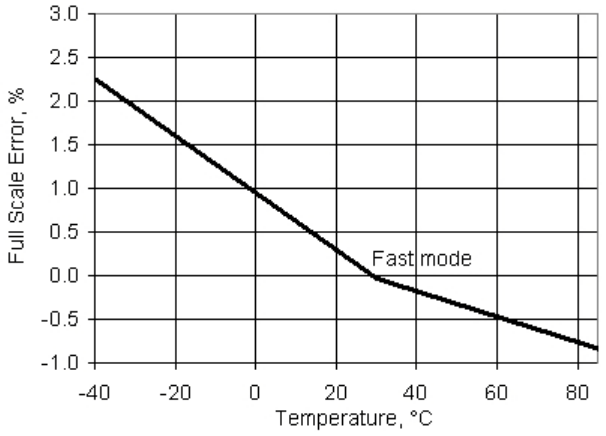


Figure 11-42. IDAC Operating Current vs Temperature, Range = 255 µA, Code = 0, Source Mode

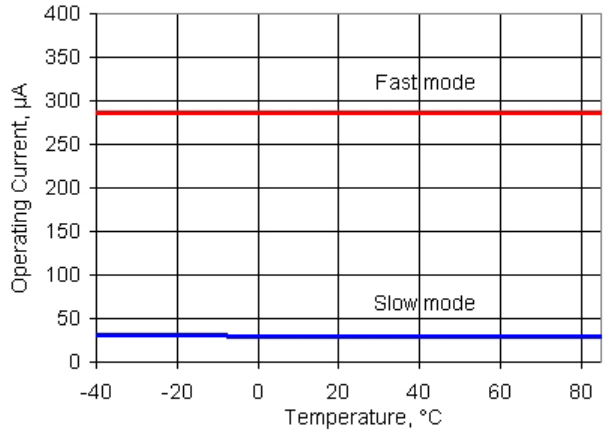


Figure 11-43. IDAC Operating Current vs Temperature, Range = 255 µA, Code = 0, Sink Mode

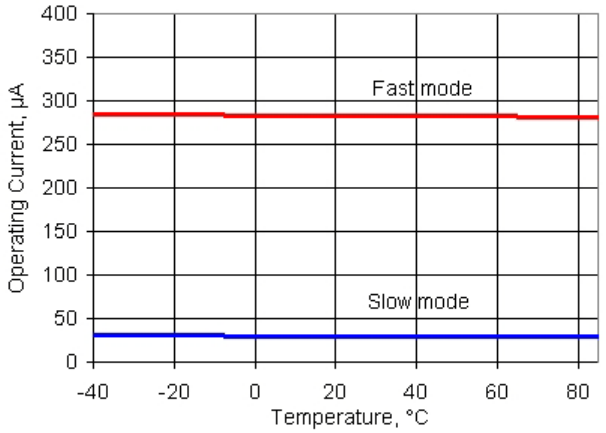


Figure 11-65. Synchronous Write and Read Cycle Timing, No Wait States

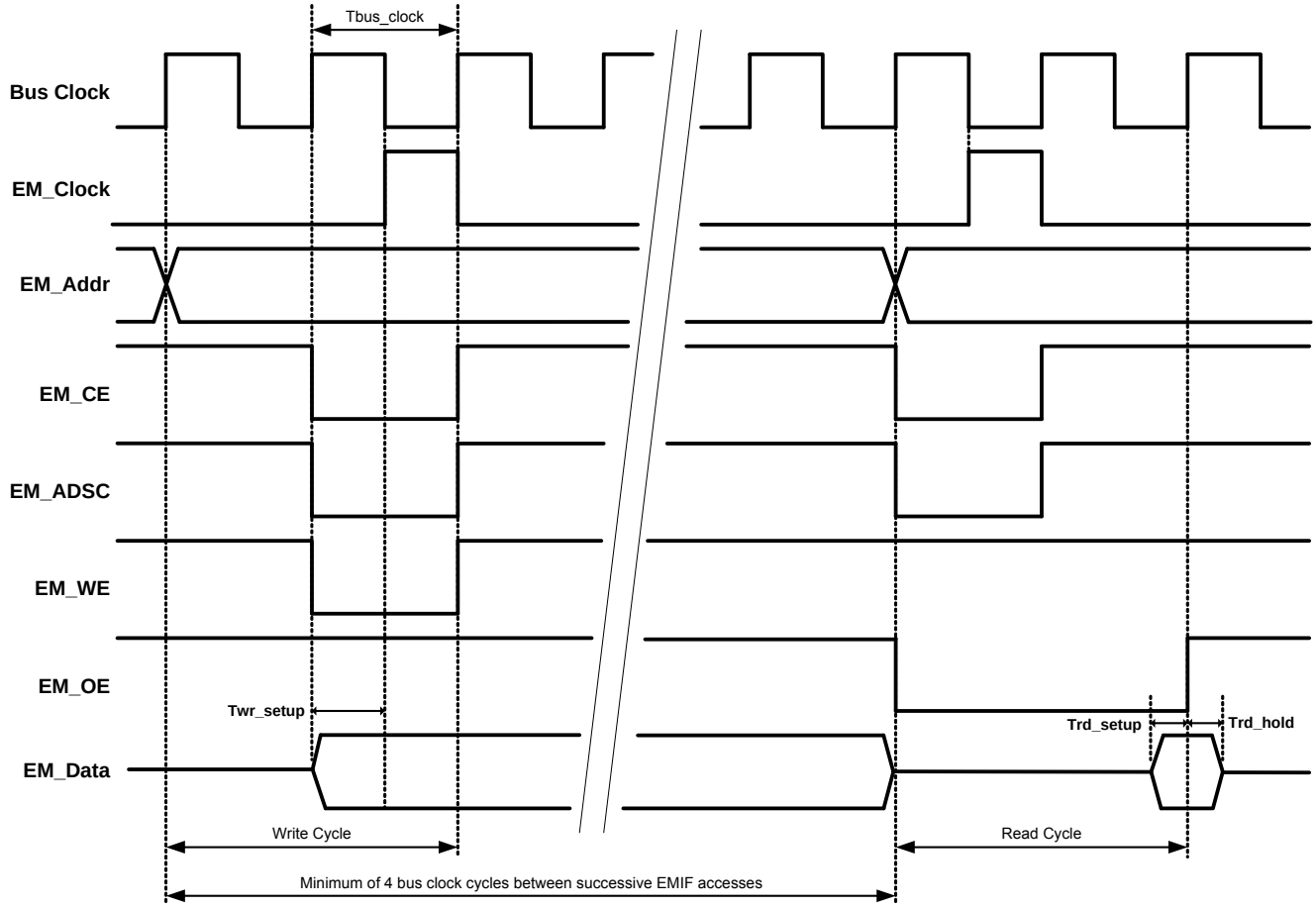


Table 11-62. Synchronous Write and Read Timing Specifications[63]

Parameter	Description	Conditions	Min	Typ	Max	Units
Fbus_clock	Bus clock frequency[64]		–	–	33	MHz
Tbus_clock	Bus clock period[65]		30.3	–	–	ns
Twr_Setup	Time from EM_data valid to rising edge of EM_Clock		Tbus_clock – 10	–	–	ns
Trd_setup	Time that EM_data must be valid before rising edge of EM_OE		5	–	–	ns
Trd_hold	Time that EM_data must be valid after rising edge of EM_OE		5	–	–	ns

Notes
63. Based on device characterization (Not production tested).
64. EMIF signal timings are limited by GPIO frequency limitations. See "GPIO" section on page 79.
65. EMIF output signals are generally synchronized to bus clock, so EMIF signal timings are dependent on bus clock frequency.

12. Ordering Information

In addition to the features listed in Table 12-1, every CY8C34 device includes: a precision on-chip voltage reference, precision oscillators, flash, ECC, DMA, a fixed function I^2C, 4 KB trace RAM, JTAG/SWD programming and debug, external memory interface, and more. In addition to these features, the flexible UDBs and analog subsection support a wide range of peripherals. To assist you in selecting the ideal part, PSoC Creator makes a part recommendation after you choose the components required by your application. All CY8C34 derivatives incorporate device and flash security in user-selectable security levels; see the TRM for details.

Table 12-1. CY8C34 Family with Single Cycle 8051

Part Number	MCU Core				Analog								Digital				I/O[81]				Package	JTAG ID[82]
	CPU Speed (MHz)	Flash (KB)	SRAM (KB)	EEPROM (KB)	LCD Segment Drive	ADC	DAC	Comparator	SC/CT Analog Blocks[79]	Opamps	DFB	CapSense	UDBs[80]	16-bit Timer/PWM	FS USB	CAN 2.0b	Total I/O	GPIO	SIO	USBIO		
16 KB Flash																						
CY8C3444LTI-110	50	16	2	0.5	✔	12-bit Del-Sig	2	4	2	2	–	✔	16	4	–	–	46	38	8	0	68-pin QFN	0×1E06E069
CY8C3444LTI-119	50	16	2	0.5	✔	12-bit Del-Sig	2	4	2	2	–	✔	16	4	–	–	29	25	4	0	48-pin QFN	0×1E077069
CY8C3444PVI-100	50	16	2	0.5	✔	12-bit Del-Sig	2	4	2	2	–	✔	16	4	–	–	29	25	4	0	48-pin SSOP	0×1E064069
32 KB Flash																						
CY8C3445AXI-104	50	32	4	1	✔	12-bit Del-Sig	2	4	2	2	–	✔	20	4	–	–	70	62	8	0	100-pin TQFP	0×1E068069
CY8C3445LTI-079	50	32	4	1	✔	12-bit Del-Sig	2	4	2	2	–	✔	20	4	–	–	46	38	8	0	68-pin QFN	0×1E04F069
CY8C3445LTI-078	50	32	4	1	✔	12-bit Del-Sig	2	4	2	2	–	✔	20	4	–	–	29	25	4	0	48-pin QFN	0×1E04E069
CY8C3445PVI-094	50	32	4	1	✔	12-bit Del-Sig	2	4	2	2	–	✔	20	4	–	–	29	25	4	0	48-pin SSOP	0×1E05E069
CY8C3445AXI-108	50	32	4	1	✔	12-bit Del-Sig	2	4	2	2	–	✔	20	4	✔	–	72	62	8	2	100-pin TQFP	0×1E06C069
CY8C3445LTI-081	50	32	4	1	✔	12-bit Del-Sig	2	4	2	2	–	✔	20	4	✔	–	48	38	8	2	68-pin QFN	0×1E051069
CY8C3445PVI-090	50	32	4	1	✔	12-bit Del-Sig	2	4	2	2	–	✔	20	4	✔	–	31	25	4	2	48-pin SSOP	0×1E05A069
64 KB Flash																						
CY8C3446LTI-073	50	64	8	2	✔	12-bit Del-Sig	2	4	2	2	–	✔	24	4	✔	–	31	25	4	2	48-pin QFN	0×1E049069
CY8C3446LTI-074	50	64	8	2	✔	12-bit Del-Sig	2	4	2	2	–	✔	24	4	–	–	46	38	8	0	68-pin QFN	0×1E04A069
CY8C3446LTI-083	50	64	8	2	✔	12-bit Del-Sig	2	4	2	2	–	✔	24	4	–	–	29	25	4	0	48-pin QFN	0x1E053069
CY8C3446AXI-099	50	64	8	2	✔	12-bit Del-Sig	2	4	2	2	–	✔	24	4	✔	–	72	62	8	2	100-pin TQFP	0×1E063069
CY8C3446AXI-105	50	64	8	2	✔	12-bit Del-Sig	2	4	2	2	–	✔	24	4	–	–	70	62	8	0	100-pin TQFP	0x1E069069
CY8C3446LTI-085	50	64	8	2	✔	12-bit Del-Sig	2	4	2	2	–	✔	24	4	✔	–	48	38	8	2	68-pin QFN	0×1E055069
CY8C3446PVI-076	50	64	8	2	✔	12-bit Del-Sig	2	4	2	2	–	✔	24	4	✔	–	31	25	4	2	48-pin SSOP	0×1E04C069
CY8C3446PVI-102	50	64	8	2	✔	12-bit Del-Sig	2	4	2	2	–	✔	24	4	–	✔	29	25	4	0	48-pin SSOP	0x1E066069

Notes

79. Analog blocks support a wide variety of functionality including TIA, PGA, and mixers. See the Example Peripherals on page 43 for more information on how analog blocks can be used.
80. UDBs support a wide variety of functionality including SPI, LIN, UART, timer, counter, PWM, PRS, and others. Individual functions may use a fraction of a UDB or multiple UDBs. Multiple functions can share a single UDB. See the Example Peripherals on page 43 for more information on how UDBs can be used.
81. The I/O Count includes all types of digital I/O: GPIO, SIO, and the two USB I/O. See the I/O System and Routing on page 36 for details on the functionality of each of these types of I/O.
82. The JTAG ID has three major fields. The most significant nibble (left digit) is the version, followed by a 2 byte part number and a 3 nibble manufacturer ID.

16. Document Conventions

16.1 Units of Measure

Table 16-1. Units of Measure

Symbol	Unit of Measure
°C	degrees Celsius
dB	decibels
fF	femtofarads
Hz	hertz
KB	1024 bytes
kbps	kilobits per second
Khr	kilohours
kHz	kilohertz
kΩ	kilohms
ksps	kilosamples per second
LSB	least significant bit
Mbps	megabits per second
MHz	megahertz
MΩ	megaohms
Msps	megasamples per second
µA	microamperes
µF	microfarads
µH	microhenrys
µs	microseconds
µV	microvolts
µW	microwatts
mA	milliamperes
ms	milliseconds
mV	millivolts
nA	nanoamperes
ns	nanoseconds
nV	nanovolts
Ω	ohms
pF	picofarads
ppm	parts per million
ps	picoseconds
s	seconds
sps	samples per second
sqrtHz	square root of hertz
V	volts

Description Title: PSoC® 3: CY8C34 Family Datasheet Programmable System-on-Chip (PSoC®) *(continued)* Document Number: 001-53304				
Revision	**ECN**	**Submission Date**	**Orig. of Change**	**Description of Change**
*M	3464258	12/14/2011	MKEA	Updated Analog Global specs Updated IDAC range Updated TIA section Modified VDDIO description in Section 3 Added note on Sleep and Hibernate modes in the Power Modes section Updated Boost Converter section Updated conditions for Inductive boost AC specs Added VDAC/IDAC noise graphs and specs Added pin capacitance specs for ECO pins Removed C_L from 32 kHz External Crystal DC Specs table. Added reference to AN54439 in Section 6.1.2.2 Deleted T_SWDO_hold row from the SWD Interface AC Specifications table Removed Pin 46 connections in "Example Schematic for 100-pin TQFP Part with Power Connections" Updated Active Mode IDD description in Table 11-2. Added I_{DDDR} and I_{DDAR} specs in Table 11-2. Replaced "total device program time" with T_{PROG} in Flash AC specs table Added I_{GPIO}, I_{SIO} and I_{USBIO} specs in Absolute Maximum Ratings Added conditions to I_{CC} spec in 32 kHz External Crystal DC Specs table. Updated TCV_{OS} value Removed Boost Efficiency vs V_{OUT} graph Updated boost graphs Updated min value of GPIO input edge rate Removed 3.4 Mbps in UDBs from I2C section Updated USBIO Block diagram; added USBIO drive mode description Updated Analog Interconnect diagram Changed max IMO startup time to 12 µs Added note for I_{IL} spec in USBIO DC specs table Updated GPIO Block diagram Updated voltage reference specs Added text explaining power supply ramp up in Section 11-4.