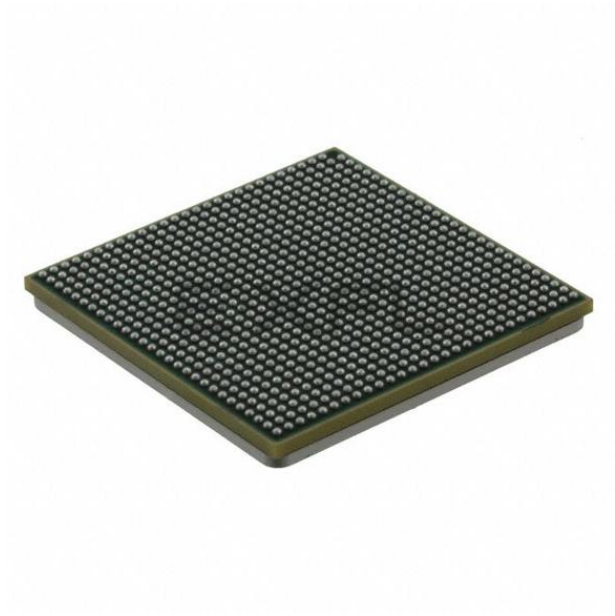

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Understanding Embedded - DSP (Digital Signal Processors)

Embedded - DSP (Digital Signal Processors) are specialized microprocessors designed to perform complex mathematical computations on digital signals in real-time. Unlike general-purpose processors, DSPs are optimized for high-speed numeric processing tasks, making them ideal for applications that require efficient and precise manipulation of digital data. These processors are fundamental in converting and processing signals in various forms, including audio, video, and communication signals, ensuring that data is accurately interpreted and utilized in embedded systems.

Applications of Embedded - DSP (Digital Signal Processors)

Details	
Product Status	Obsolete
Type	SC3850 Six Core
Interface	Ethernet, I²C, PCI, RGMII, Serial RapidIO, SGMII, SPI, UART/USART
Clock Rate	800MHz
Non-Volatile Memory	ROM (96kB)
On-Chip RAM	576kB
Voltage - I/O	2.50V
Voltage - Core	1.00V
Operating Temperature	0°C ~ 105°C (TJ)
Mounting Type	Surface Mount
Package / Case	783-BBGA, FCBGA
Supplier Device Package	783-FCPBGA (29x29)
Purchase URL	https://www.e-xfl.com/product-detail/nxp-semiconductors/msc8256svt800b

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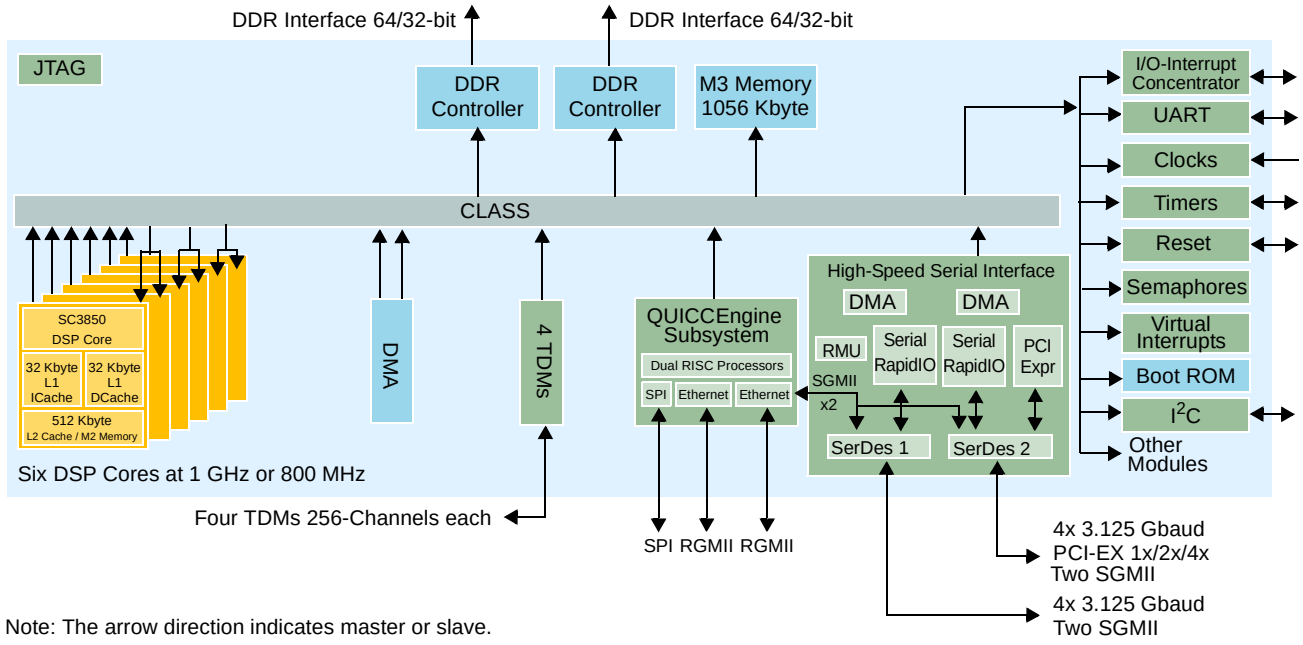


Figure 1. MSC8256 Block Diagram

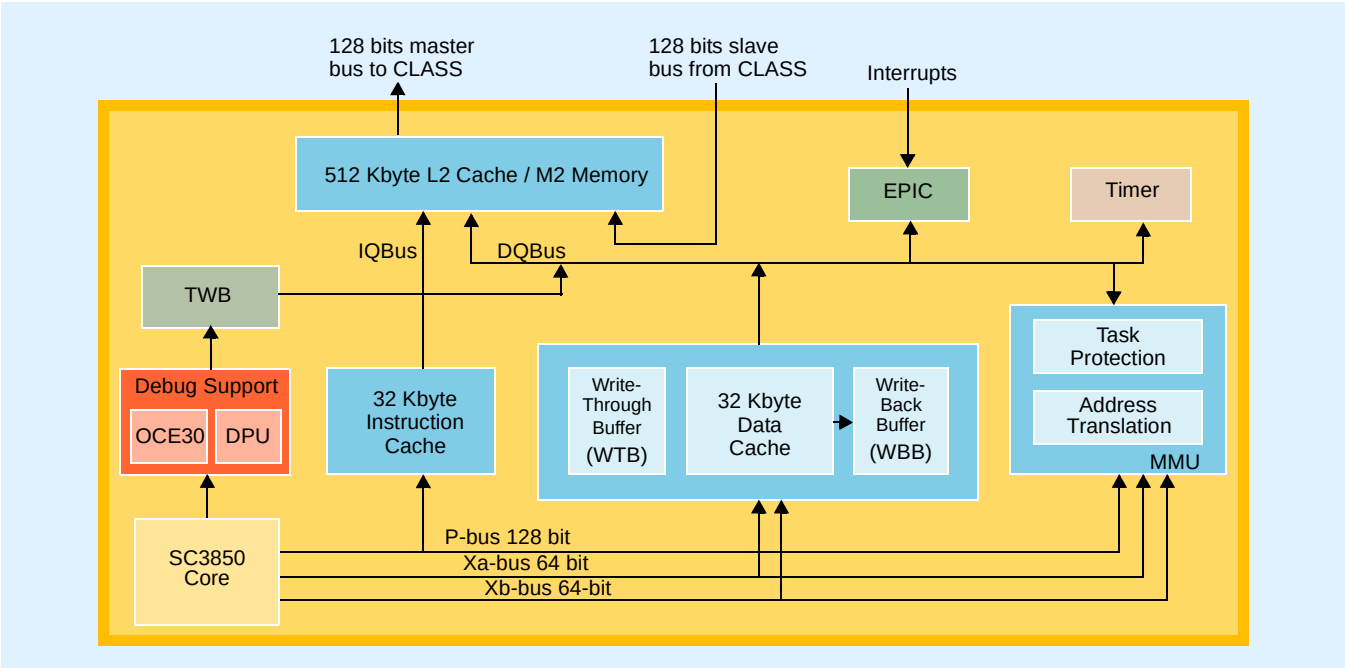


Figure 2. StarCore SC3850 DSP Subsystem Block Diagram

Table 1. Signal List by Ball Number (continued)

Ball Number	Signal Name[1,2]	Pin Type[10]	Power Rail Name
E17	M2DQ56	I/O	GVDD2
E18	M2DQ57	I/O	GVDD2
E19	$\overline{\text{M2DQS7}}$	I/O	GVDD2
E20	Reserved	NC	—
E21	Reserved	NC	—
E22	Reserved	NC	—
E23	SXPVDD1	Power	N/A
E24	SXPVSS1	Ground	N/A
E25	SR1_PLL_AGND[9]	Ground	SXCVSS1
E26	SR1_PLL_AVDD[9]	Power	SXCVDD1
E27	SXCVSS1	Ground	N/A
E28	SXCVDD1	Power	N/A
F1	VSS	Ground	N/A
F2	GVDD2	Power	N/A
F3	M2DQ16	I/O	GVDD2
F4	VSS	Ground	N/A
F5	GVDD2	Power	N/A
F6	M2DQ17	I/O	GVDD2
F7	VSS	Ground	N/A
F8	GVDD2	Power	N/A
F9	M2BA2	O	GVDD2
F10	VSS	Ground	N/A
F11	GVDD2	Power	N/A
F12	M2A4	O	GVDD2
F13	VSS	Ground	N/A
F14	GVDD2	Power	N/A
F15	M2DQ42	I/O	GVDD2
F16	VSS	Ground	N/A
F17	GVDD2	Power	N/A
F18	M2DQ58	I/O	GVDD2
F19	M2DQS7	I/O	GVDD2
F20	GVDD2	Power	N/A
F21	SXPVDD1	Power	N/A
F22	SXPVSS1	Ground	N/A
F23	SR1_TXD2/SG1_TX[4]	O	SXPVDD1
F24	$\overline{\text{SR1\_TXD2}}$/$\overline{\text{SG1\_TX}}$[4]	O	SXPVDD1
F25	SXCVDD1	Power	N/A
F26	SXCVSS1	Ground	N/A
F27	$\overline{\text{SR1\_RXD2}}$/$\overline{\text{SG1\_RX}}$[4]	I	SXCVDD1
F28	SR1_RXD2/SG1_RX[4]	I	SXCVDD1
G1	$\overline{\text{M2DQS2}}$	I/O	GVDD2
G2	M2DQS2	I/O	GVDD2
G3	M2DQ19	I/O	GVDD2
G4	M2DM2	O	GVDD2
G5	M2DQ21	I/O	GVDD2
G6	M2DQ22	I/O	GVDD2

Table 1. Signal List by Ball Number (continued)

Ball Number	Signal Name[1,2]	Pin Type[10]	Power Rail Name
K15	VDD	Power	N/A
K16	VSS	Ground	N/A
K17	VDD	Power	N/A
K18	VSS	Ground	N/A
K19	VDD	Power	N/A
K20	Reserved	NC	—
K21	Reserved	NC	—
K22	Reserved	NC	—
K23	SXPVDD2	Power	N/A
K24	SXPVSS2	Ground	N/A
K25	SXCVDD2	Power	N/A
K26	SXCVSS2	Ground	N/A
K27	SXCVDD2	Power	N/A
K28	SXCVSS2	Ground	N/A
L1	M2DQ9	I/O	GVDD2
L2	M2DQ12	I/O	GVDD2
L3	M2DQ13	I/O	GVDD2
L4	$\overline{\text{M2DQS0}}$	I/O	GVDD2
L5	M2DQS0	I/O	GVDD2
L6	M2DM0	O	GVDD2
L7	M2DQ3	I/O	GVDD2
L8	M2DQ2	I/O	GVDD2
L9	M2DQ4	I/O	GVDD2
L10	VDD	Power	N/A
L11	VSS	Ground	N/A
L12	M3VDD	Power	N/A
L13	VSS	Ground	N/A
L14	VDD	Power	N/A
L15	VSS	Ground	N/A
L16	VDD	Power	N/A
L17	VSS	Ground	N/A
L18	VDD	Power	N/A
L19	VSS	Ground	N/A
L20	Reserved	NC	—
L21	Reserved	NC	—
L22	Reserved	NC	—
L23	SR2_TXD3/PE_TXD3/SG2_TX[4]	O	SXPVDD2
L24	$\overline{\text{SR2\_TXD3}}$/$\overline{\text{PE\_TXD3}}$/$\overline{\text{SG2\_TX}}$[4]	O	SXPVDD2
L25	SXCVSS2	Ground	N/A
L26	SXCVDD2	Power	N/A
L27	$\overline{\text{SR2\_RXD3}}$/$\overline{\text{PE\_RXD3}}$/$\overline{\text{SG2\_RX}}$[4]	I	SXCVDD2
L28	SR2_RXD3/PE_RXD3/SG2_RX[4]	I	SXCVDD2
M1	M2DQ8	I/O	GVDD2
M2	VSS	Ground	N/A
M3	GVDD2	Power	N/A
M4	M2DQ15	I/O	GVDD2

Table 1. Signal List by Ball Number (continued)

Ball Number	Signal Name[1,2]	Pin Type[10]	Power Rail Name
M5	M2DQ1	I/O	GVDD2
M6	VSS	Ground	N/.A
M7	GVDD2	Power	N/A
M8	M2DQ7	I/O	GVDD2
M9	M2DQ6	I/O	GVDD2
M10	VSS	Ground	N/A
M11	VDD	Power	N/A
M12	VSS	Ground	N/A
M13	VDD	Power	N/A
M14	VSS	Ground	N/A
M15	VDD	Power	N/A
M16	VSS	Ground	N/A
M17	VDD	Power	N/A
M18	VSS	Ground	N/A
M19	VDD	Power	N/A
M20	Reserved	NC	—
M21	Reserved	NC	—
M22	Reserved	NC	—
M23	SXPVSS2	Ground	N/A
M24	SXPVDD2	Power	N/A
M25	SR2_IMP_CAL_TX	I	SXCVDD2
M26	SXCVSS2	Ground	N/A
M27	Reserved	NC	—
M28	Reserved	NC	—
N1	VSS	Ground	N/A
N2	$\overline{\text{TRST}}$[7]	I	QVDD
N3	$\overline{\text{PORESET}}$[7]	I	QVDD
N4	VSS	Ground	N/A
N5	TMS[7]	I	QVDD
N6	CLKOUT	O	QVDD
N7	VSS	Ground	N/A
N8	VSS	Ground	N/A
N9	VSS	Ground	N/A
N10	VDD	Power	N/A
N11	VSS	Ground	N/A
N12	M3VDD	Power	N/A
N13	VSS	Ground	N/A
N14	VDD	Power	N/A
N15	VSS	Ground	N/A
N16	VDD	Power	N/A
N17	VSS	Ground	N/A
N18	VDD	Power	N/A
N19	VSS	Ground	N/A
N20	Reserved	NC	—
N21	SXPVDD2	Power	N/A
N22	SXPVSS2	Ground	N/A

Table 1. Signal List by Ball Number (continued)

Ball Number	Signal Name[1,2]	Pin Type[10]	Power Rail Name
AB1	$\overline{\text{M1DQS2}}$	I/O	GVDD1
AB2	M1DQS2	I/O	GVDD1
AB3	M1DQ19	I/O	GVDD1
AB4	M1DM2	O	GVDD1
AB5	M1DQ21	I/O	GVDD1
AB6	M1DQ22	I/O	GVDD1
AB7	M1CKE0	O	GVDD1
AB8	M1A11	O	GVDD1
AB9	M1A7	O	GVDD1
AB10	$\overline{\text{M1CK2}}$	O	GVDD1
AB11	M1APAR_OUT	O	GVDD1
AB12	M1ODT1	O	GVDD1
AB13	$\overline{\text{M1APAR\_IN}}$	I	GVDD1
AB14	M1DQ43	I/O	GVDD1
AB15	M1DM5	O	GVDD1
AB16	M1DQ44	I/O	GVDD1
AB17	M1DQ40	I/O	GVDD1
AB18	M1DQ59	I/O	GVDD1
AB19	M1DM7	O	GVDD1
AB20	M1DQ60	I/O	GVDD1
AB21	VSS	Ground	N/A
AB22	GPIO31/I2C_SDA[5,8]	I/O	NVDD
AB23	GPIO27/TMR4/RCW_SRC0[5,8]	I/O	NVDD
AB24	GPIO25/TMR2/RCW_SRC1[5,8]	I/O	NVDD
AB25	GPIO24/TMR1/RCW_SRC2[5,8]	I/O	NVDD
AB26	GPIO10/$\overline{\text{IRQ10}}$/RC10[5,8]	I/O	NVDD
AB27	GPIO5/$\overline{\text{IRQ5}}$/RC5[5,8]	I/O	NVDD
AB28	GPIO0/$\overline{\text{IRQ0}}$/RC0[5,8]	I/O	NVDD
AC1	VSS	Ground	N/A
AC2	GVDD1	Power	N/A
AC3	M1DQ16	I/O	GVDD1
AC4	VSS	Ground	N/A
AC5	GVDD1	Power	N/A
AC6	M1DQ17	I/O	GVDD1
AC7	VSS	Ground	N/A
AC8	GVDD1	Power	N/A
AC9	M1BA2	O	GVDD1
AC10	VSS	Ground	N/A
AC11	GVDD1	Power	N/A
AC12	M1A4	O	GVDD1
AC13	VSS	Ground	N/A
AC14	GVDD1	Power	N/A
AC15	M1DQ42	I/O	GVDD1
AC16	VSS	Ground	N/A
AC17	GVDD1	Power	N/A
AC18	M1DQ58	I/O	GVDD1

Table 1. Signal List by Ball Number (continued)

Ball Number	Signal Name[1,2]	Pin Type[10]	Power Rail Name
AF27	TDM2TDT/GE1_TX_CLK[3]	I/O	NVDD
AF28	TDM3RSN/GE1_RD1[3]	I/O	NVDD
AG1	M1DQ24	I/O	GVDD1
AG2	GVDD1	Power	N/A
AG3	M1DQ25	I/O	GVDD1
AG4	VSS	Ground	N/A
AG5	GVDD1	Power	N/A
AG6	M1ECC1	I/O	GVDD1
AG7	VSS	Ground	N/A
AG8	GVDD1	Power	N/A
AG9	M1A13	O	GVDD1
AG10	VSS	Ground	N/A
AG11	GVDD1	Power	N/A
AG12	$\overline{\text{M1CS1}}$	O	GVDD1
AG13	VSS	Ground	N/A
AG14	GVDD1	Power	N/A
AG15	M1DQ35	I/O	GVDD1
AG16	VSS	Ground	N/A
AG17	GVDD1	Power	N/A
AG18	M1DQ51	I/O	GVDD1
AG19	VSS	Ground	N/A
AG20	GVDD1	Power	N/A
AG21	NVDD	Power	N/A
AG22	TDM1TSN/GE2_TD1[3]	I/O	NVDD
AG23	TDM1RDT/GE2_TX_CLK[3]	I/O	NVDD
AG24	TDM0TCK/GE2_GTX_CLK[3]	I/O	NVDD
AG25	TDM1TDT/GE2_TD0[3]	I/O	NVDD
AG26	VSS	Ground	N/A
AG27	NVDD	Power	N/A
AG28	TDM3RDT/GE1_RD0[3]	I/O	NVDD
AH1	Reserved.	NC	—
AH2	$\overline{\text{M1DQS3}}$	I/O	GVDD1
AH3	M1DQS3	I/O	GVDD1
AH4	M1ECC0	I/O	GVDD1
AH5	$\overline{\text{M1DQS8}}$	I/O	GVDD1
AH6	M1DQS8	I/O	GVDD1
AH7	M1A5	O	GVDD1
AH8	$\overline{\text{M1CK1}}$	O	GVDD1
AH9	M1CK1	O	GVDD1
AH10	$\overline{\text{M1CS0}}$	O	GVDD1
AH11	M1BA0	O	GVDD1
AH12	$\overline{\text{M1CAS}}$	O	GVDD1
AH13	M1DQ34	I/O	GVDD1
AH14	$\overline{\text{M1DQS4}}$	I/O	GVDD1
AH15	M1DQS4	I/O	GVDD1
AH16	M1DQ50	I/O	GVDD1

2.5.1.3 DDR2/DDR3 SDRAM Capacitance

Table 8 provides the DDR controller interface capacitance for DDR2 and DDR3 memory.

Note: At recommended operating conditions (see Table 3) with V_{DDDDR} = 1.8 V for DDR2 memory or V_{DDDDR} = 1.5 V for DDR3 memory.

Table 8. DDR2/DDR3 SDRAM Capacitance

Parameter	Symbol	Min	Max	Unit
I/O capacitance: DQ, DQS, $\overline{DQS}$	C_{IO}	6	8	pF
Delta I/O capacitance: DQ, DQS, $\overline{DQS}$	C_{DIO}	—	0.5	pF
Note: Guaranteed by FAB process and micro-construction.				

2.5.1.4 DDR Reference Current Draw

Table 9 lists the current draw characteristics for MV_{REF}.

Note: Values when used at recommended operating conditions (see Table 3).

Table 9. Current Draw Characteristics for MV_{REF}

Parameter / Condition	Symbol	Min	Max	Unit
Current draw for MV_{REFn} • DDR2 SDRAM • DDR3 SDRAM	I_{MVREFn}	—	 300 250	 µA µA

2.5.2 High-Speed Serial Interface (HSSI) DC Electrical Characteristics

The MSC8256 features an HSSI that includes two 4-channel SerDes ports used for high-speed serial interface applications (PCI Express, Serial RapidIO interfaces, and SGMII). This section and its subsections describe the common portion of the SerDes DC, including the DC requirements for the SerDes reference clocks and the SerDes data lane transmitter (Tx) and receiver (Rx) reference circuits. The data lane circuit specifications are specific for each supported interface, and they have individual subsections by protocol. The selection of individual data channel functionality is done via the Reset Configuration Word High Register (RCWHR) SerDes Protocol selection fields (S1P and S2P). Specific AC electrical characteristics are defined in Section 2.6.2, "HSSI AC Timing Specifications."

2.5.2.1 Signal Term Definitions

The SerDes interface uses differential signalling to transfer data across the serial link. This section defines terms used in the description and specification of differential signals. Figure 4 shows how the signals are defined. For illustration purposes only, one SerDes lane is used in the description. Figure 4 shows the waveform for either a transmitter output (SR[1–2]_TX and

$\overline{\text{SR[1–2]\_TX}}$) or a receiver input (SR[1–2]_RX and $\overline{\text{SR[1–2]\_RX}}$). Each signal swings between A volts and B volts where A > B.

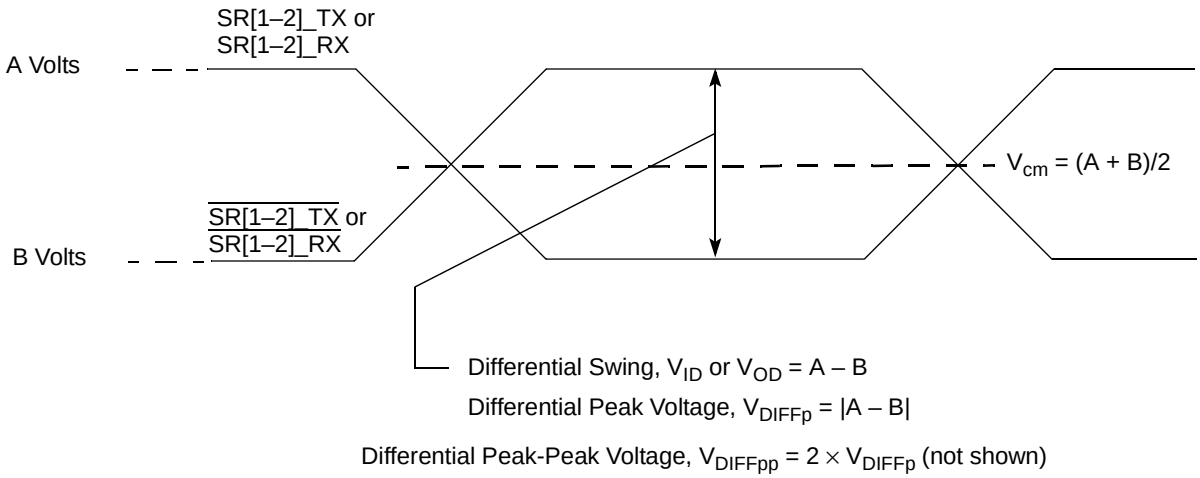


Figure 4. Differential Voltage Definitions for Transmitter or Receiver

Using this waveform, the definitions are listed in Table 10. To simplify the illustration, the definitions assume that the SerDes transmitter and receiver operate in a fully symmetrical differential signalling environment.

Table 10. Differential Signal Definitions

Term	Definition
Single-Ended Swing	The transmitter output signals and the receiver input signals SR[1–2]_TX, $\overline{\text{SR[1–2]\_TX}}$, SR[1–2]_RX and $\overline{\text{SR[1–2]\_RX}}$ each have a peak-to-peak swing of A – B volts. This is also referred to as each signal wire's single-ended swing.
Differential Output Voltage, V_{OD} (or **Differential Output Swing**):	The differential output voltage (or swing) of the transmitter, V_{OD}, is defined as the difference of the two complimentary output voltages: $V_{SR[1–2]\_TX} - V_{\overline{SR[1–2]\_TX}}$. The V_{OD} value can be either positive or negative.
Differential Input Voltage, V_{ID} (or **Differential Input Swing**)	The differential input voltage (or swing) of the receiver, V_{ID}, is defined as the difference of the two complimentary input voltages: $V_{SR[1–2]\_RX} - V_{\overline{SR[1–2]\_RX}}$. The V_{ID} value can be either positive or negative.
Differential Peak Voltage, V_{DIFFp}	The peak value of the differential transmitter output signal or the differential receiver input signal is defined as the differential peak voltage, V_{DIFFp} = |A – B| volts.
Differential Peak-to-Peak, $V_{DIFFp-p}$	Since the differential output signal of the transmitter and the differential input signal of the receiver each range from A – B to –(A – B) volts, the peak-to-peak value of the differential transmitter output signal or the differential receiver input signal is defined as differential peak-to-peak voltage, $V_{DIFFp-p}$ = 2 × V_{DIFFp} = 2 × |(A – B)| volts, which is twice the differential swing in amplitude, or twice of the differential peak. For example, the output differential peak-peak voltage can also be calculated as $V_{TX-DIFFp-p}$ = 2 × |V_{OD}|.
Differential Waveform	The differential waveform is constructed by subtracting the inverting signal ($\overline{\text{SR[1–2]\_TX}}$, for example) from the non-inverting signal (SR[1–2]_TX, for example) within a differential pair. There is only one signal trace curve in a differential waveform. The voltage represented in the differential waveform is not referenced to ground. Refer to Figure 16 as an example for differential waveform.
Common Mode Voltage, V_{cm}	The common mode voltage is equal to half of the sum of the voltages between each conductor of a balanced interchange circuit and ground. In this example, for SerDes output, V_{cm_out} = ($V_{SR[1–2]\_TX}$ + $V_{\overline{SR[1–2]\_TX}}$) ÷ 2 = (A + B) ÷ 2, which is the arithmetic mean of the two complimentary output voltages within a differential pair. In a system, the common mode voltage may often differ from one component's output to the other's input. It may be different between the receiver input and driver output circuits within the same component. It is also referred to as the DC offset on some occasions.

To illustrate these definitions using real values, consider the example of a current mode logic (CML) transmitter that has a common mode voltage of 2.25 V and outputs, TD and $\overline{\text{TD}}$. If these outputs have a swing from 2.0 V to 2.5 V, the peak-to-peak voltage swing of each signal (TD or $\overline{\text{TD}}$) is 500 mV p-p, which is referred to as the single-ended swing for each signal. Because the differential signaling environment is fully symmetrical in this example, the transmitter output differential swing (V_{OD}) has the same amplitude as each signal single-ended swing. The differential output signal ranges between 500 mV and –500 mV. In other words, V_{OD} is 500 mV in one phase and –500 mV in the other phase. The peak differential voltage (V_{DIFFp}) is 500 mV. The peak-to-peak differential voltage ($V_{DIFFp-p}$) is 1000 mV p-p.

2.5.2.2 SerDes Reference Clock Receiver Characteristics

The SerDes reference clock inputs are applied to an internal PLL whose output creates the clock used by the corresponding SerDes lanes. The SerDes reference clock inputs are SR1_REF_CLK/$\overline{\text{SR1\_REF\_CLK}}$ or SR2_REF_CLK/$\overline{\text{SR2\_REF\_CLK}}$. Figure 5 shows a receiver reference diagram of the SerDes reference clocks.

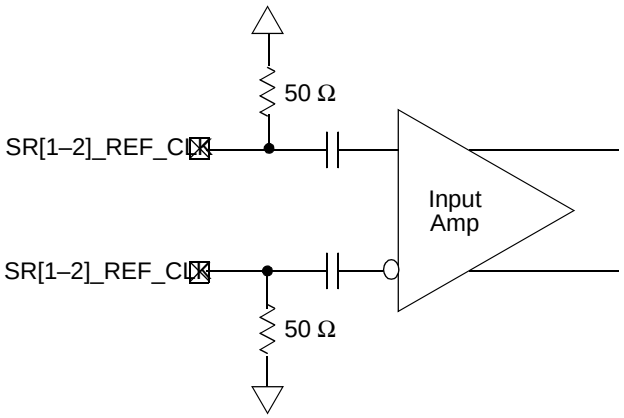


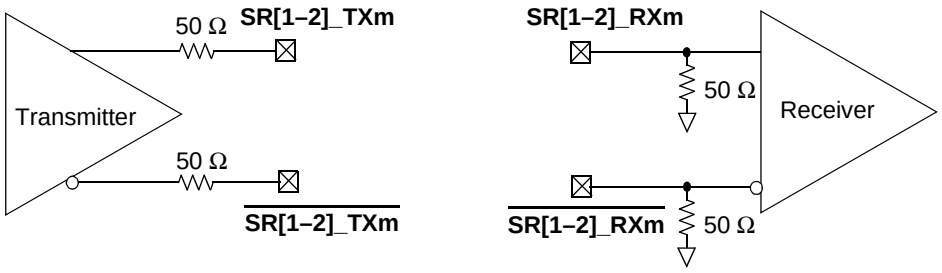
Figure 5. Receiver of SerDes Reference Clocks

The characteristics of the clock signals are as follows:

- The supply voltage requirements for V_{DDSXC} are as specified in **Table 3.**
- The SerDes reference clock receiver reference circuit structure is as follows:
  - The SR[1–2]_REF_CLK and $\overline{\text{SR[1–2]\_REF\_CLK}}$ are internally AC-coupled differential inputs as shown in Figure 5. Each differential clock input (SR[1–2]_REF_CLK or $\overline{\text{SR[1–2]\_REF\_CLK}}$) has on-chip 50-Ω termination to GND_{SXC} followed by on-chip AC-coupling.
  - The external reference clock driver must be able to drive this termination.
  - The SerDes reference clock input can be either differential or single-ended. Refer to the differential mode and single-ended mode descriptions below for detailed requirements.
- The maximum average current requirement also determines the common mode voltage range.
  - When the SerDes reference clock differential inputs are DC coupled externally with the clock driver chip, the maximum average current allowed for each input pin is 8 mA. In this case, the exact common mode input voltage is not critical as long as it is within the range allowed by the maximum average current of 8 mA because the input is AC-coupled on-chip.
  - This current limitation sets the maximum common mode input voltage to be less than 0.4 V (0.4 V / 50 = 8 mA) while the minimum common mode input level is 0.1 V above GND_{SXC}. For example, a clock with a 50/50 duty cycle can be produced by a clock driver with output driven by its current source from 0 mA to 16 mA (0–0.8 V), such that each phase of the differential input has a single-ended swing from 0 V to 800 mV with the common mode voltage at 400 mV.
  - If the device driving the SR[1–2]_REF_CLK and $\overline{\text{SR[1–2]\_REF\_CLK}}$ inputs cannot drive 50 Ω to GND_{SXC} DC or the drive strength of the clock driver chip exceeds the maximum input current limitations, it must be AC-coupled externally.
- The input amplitude requirement is described in detail in the following sections.

2.5.2.3 SerDes Transmitter and Receiver Reference Circuits

Figure 6 shows the reference circuits for SerDes data lane transmitter and receiver.



Note: The [1–2] indicates the specific SerDes Interface (1 or 2) and the m indicates the specific channel within that interface (0,1,2,3). Actual signals are assigned by the HRCW assignments at reset (see **Chapter 5**, *Reset* in the reference manual for details)

Figure 6. SerDes Transmitter and Receiver Reference Circuits

2.5.3 DC-Level Requirements for SerDes Interfaces

The following subsections define the DC-level requirements for the SerDes reference clocks, the PCI Express data lines, the Serial RapidIO data lines, and the SGMII data lines.

2.5.3.1 DC-Level Requirements for SerDes Reference Clocks

The DC-level requirement for the SerDes reference clock inputs is different depending on the signaling mode used to connect the clock driver chip and SerDes reference clock inputs, as described below:

- Differential Mode
  - The input amplitude of the differential clock must be between 400 mV and 1600 mV differential peak-peak (or between 200 mV and 800 mV differential peak). In other words, each signal wire of the differential pair must have a single-ended swing of less than 800 mV and greater than 200 mV. This requirement is the same for both external DC-coupled or AC-coupled connection.
  - For an external DC-coupled connection, the maximum average current requirements sets the requirement for average voltage (common mode voltage) as between 100 mV and 400 mV. Figure 7 shows the SerDes reference clock input requirement for DC-coupled connection scheme.

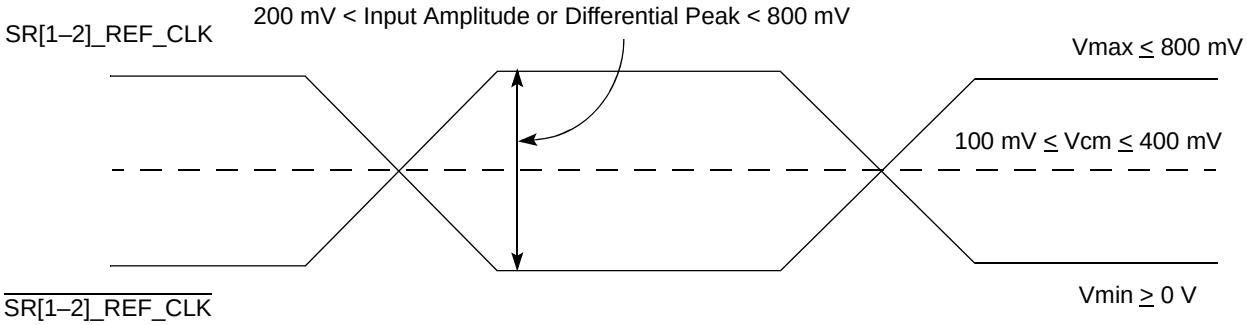


Figure 7. Differential Reference Clock Input DC Requirements (External DC-Coupled)

Note: Specifications are valid at the recommended operating conditions listed in Table 3.

Table 11. PCI Express (2.5 Gbps) Differential Transmitter (Tx) Output DC Specifications

Parameter	Symbol	Min	Typical	Max	Units	Notes
Differential peak-to-peak output voltage	$V_{TX-DIFFp-p}$	800	1000	1200	mV	1
De-emphasized differential output voltage (ratio)	$V_{TX-DE-RATIO}$	3.0	3.5	4.0	dB	2
DC differential Tx impedance	$Z_{TX-DIFF-DC}$	80	100	120	Ω	3
Transmitter DC impedance	Z_{TX-DC}	40	50	60	Ω	4

Notes:
1. $V_{TX-DIFFp-p} = 2 \times |V_{TX-D+} - V_{TX-D-}|$ Measured at the package pins with a test load of 50 Ω to GND on each pin.
2. Ratio of the $V_{TX-DIFFp-p}$ of the second and following bits after a transition divided by the $V_{TX-DIFFp-p}$ of the first bit after a transition. Measured at the package pins with a test load of 50 Ω to GND on each pin.
3. Tx DC differential mode low impedance
4. Required Tx D+ as well as D– DC Impedance during all states

Table 12. PCI Express (2.5 Gbps) Differential Receiver (Rx) Input DC Specifications

Parameter	Symbol	Min	Typical	Max	Units	Notes
Differential input peak-to-peak voltage	$V_{RX-DIFFp-p}$	120	1000	1200	mV	1
DC differential Input Impedance	$Z_{RX-DIFF-DC}$	80	100	120	Ω	2
DC input impedance	Z_{RX-DC}	40	50	60	Ω	3
Powered down DC input impedance	$Z_{RX-HIGH-IMP-DC}$	50	—	—	KΩ	4
Electrical idle detect threshold	$V_{RX-IDLE-DET-DIFFp-p}$	65	—	175	mV	5

Notes:
1. $V_{RX-DIFFp-p} = 2 \times |V_{RX-D+} - V_{RX-D-}|$ Measured at the package pins with a test load of 50 Ω to GND on each pin.
2. Rx DC differential mode impedance. Impedance during all LTSSM states. When transitioning from a fundamental reset to detect (the initial state of the LTSSM), there is a 5 ms transition time before the receiver termination values must be met on all unconfigured lanes of a port.
3. Required Rx D+ as well as D– DC Impedance (50 ±20% tolerance). Measured at the package pins with a test load of 50 Ω to GND on each pin. Impedance during all LTSSM states. When transitioning from a fundamental reset to detect (the initial state of the LTSSM), there is a 5 ms transition time before the receiver termination values must be met on all unconfigured lanes of a port.
4. Required Rx D+ as well as D– DC Impedance when the receiver terminations do not have power. The Rx DC common mode impedance that exists when no power is present or fundamental reset is asserted. This helps ensure that the receiver detect circuit does not falsely assume a receiver is powered on when it is not. This term must be measured at 300 mV above the Rx ground.
5. $V_{RX-IDLE-DET-DIFFp-p} = 2 \times |V_{RX-D+} - V_{RX-D-}|$. Measured at the package pins of the receiver

2.5.3.3 DC-Level Requirements for Serial RapidIO Configurations

This sections provided various DC-level requirements for Serial RapidIO Configurations.

Note: Specifications are valid at the recommended operating conditions listed in Table 3.

Table 13. Serial RapidIO Transmitter DC Specifications

Parameter	Symbol	Min	Typical	Max	Units	Notes
Output voltage	V_O	–0.40	—	2.30	V	1
Long run differential output voltage	V_{DIFFPP}	800	—	1600	mVp-p	—
Short run differential output voltage	V_{DIFFPP}	500	—	1000	mVp-p	—

Note: Voltage relative to COMMON of either signal comprising a differential pair.

Table 14. Serial RapidIO Receiver DC Specifications

Parameter	Symbol	Min	Typical	Max	Units	Notes
Differential input voltage	V_{IN}	200	—	1600	mVp-p	1

Notes: **1.** Measured at receiver.

2.5.3.4 DC-Level Requirements for SGMII Configurations

Note: Specifications are valid at the recommended operating conditions listed in Table 3

Table 15 describes the SGMII SerDes transmitter AC-coupled DC electrical characteristics. Transmitter DC characteristics are measured at the transmitter outputs (SR[1–2]_TX[n] and $\overline{\text{SR[1–2]\_TX[n]}}$) as shown in Figure 10.

Table 15. SGMII DC Transmitter Electrical Characteristics

Parameter	Symbol	Min	Typ	Max	Unit	Notes
Output high voltage	V_{OH}	—	—	$XV_{DD\_SRDS\text{-}Typ}/2 + \lvert V_{OD}\rvert_{\text{-}max}/2$	mV	1
Output low voltage	V_{OL}	$XV_{DD\_SRDS\text{-}Typ}/2 - \lvert V_{OD}\rvert_{\text{-}max}/2$	—	—	mV	1
Output differential voltage ($XV_{DD\text{-}Typ}$ at 1.0 V)	$\lvert V_{OD}\rvert$	323	500	725	mV	2,3,4
		296	459	665		2,3,5
		269	417	604		2,3,6
		243	376	545		2,3,7
		215	333	483		2,3,8
		189	292	424		2,3,9
		162	250	362		2,3,10
Output impedance (single-ended)	R_O	40	50	60	Ω	—

Notes:
1. This does not align to DC-coupled SGMII. $XV_{DD\_SRDS2\text{-}Typ}$ = 1.1 V.
2. The $\lvert V_{OD}\rvert$ value shown in the table assumes full multitude by setting srd_smit_lvl as 000 and the following transmit equalization setting in the XMITEQ**AB** (for lanes A and B) or XMITEQ**EF** (for lanes E and F) bit field of Control Register:
• The MSB (bit 0) of the above bit field is set to zero (selecting the full $V_{DD\text{-}DIFF\text{-}p\text{-}p}$ amplitude which is power up default);
• The LSB (bit [1–3]) of the above bit field is set based on the equalization settings listed in notes 4 through 10.
3. The $\lvert V_{OD}\rvert$ value shown in the Typ column is based on the condition of $XV_{DD\_SRDS2\text{-}Typ}$ = 1.0 V, no common mode offset variation (V_{OS} =500mV), SerDes transmitter is terminated with 100-Ω differential load between
4. Equalization setting: 1.0x: 0000.
5. Equalization setting: 1.09x: 1000.
6. Equalization setting: 1.2x: 0100.
7. Equalization setting: 1.33x: 1100.
8. Equalization setting: 1.5x: 0010.
9. Equalization setting: 1.71x: 1010.
10. Equalization setting: 2.0x: 0110.
11. $\lvert V_{OD}\rvert = \lvert V_{SR[1–2]\_TXn} - V_{\overline{SR[1–2]\_TXn}}\rvert$. $\lvert V_{OD}\rvert$ is also referred to as output differential peak voltage. $V_{TX\text{-}DIFFp\text{-}p} = 2*\lvert V_{OD}\rvert$.

2.6 AC Timing Characteristics

This section describes the AC timing characteristics for the MSC8256.

2.6.1 DDR SDRAM AC Timing Specifications

This section describes the AC electrical characteristics for the DDR SDRAM interface.

2.6.1.1 DDR SDRAM Input AC Timing Specifications

Table 18 provides the input AC timing specifications for the DDR SDRAM when V_{DDDDR} (typ) = 1.8 V.

Table 18. DDR2 SDRAM Input AC Timing Specifications for 1.8 V Interface

Parameter	Symbol	Min	Max	Unit
AC input low voltage	V_{IL}	—	MV_{REF} – 0.20	V
AC input high voltage	V_{IH}	MV_{REF} + 0.20	—	V
Note: At recommended operating conditions with V_{DDDDR} of 1.8 ± 5%.				

Table 19 provides the input AC timing specifications for the DDR SDRAM when V_{DDDDR} (typ) = 1.5 V.

Table 19. DDR3 SDRAM Input AC Timing Specifications for 1.5 V Interface

Parameter	Symbol	Min	Max	Unit
AC input low voltage	V_{IL}	—	MV_{REF} – 0.175	V
AC input high voltage	V_{IH}	MV_{REF} + 0.175	—	V
Note: At recommended operating conditions with V_{DDDDR} of 1.5 ± 5%.				

Table 20 provides the input AC timing specifications for the DDR SDRAM interface.

Table 20. DDR SDRAM Input AC Timing Specifications

Parameter	Symbol	Min	Max	Unit	Notes
Controller Skew for MDQS—MDQ/MECC/MDM	t_{CISKEW}				1, 2
• 800 MHz data rate		–200	200	ps	
• 667 MHz data rate		–240	240	ps	
Tolerated Skew for MDQS—MDQ/MECC/MDM	t_{DISKEW}				2, 3
• 800 MHz data rate		–425	425	ps	
• 667 MHz data rate		–510	510	ps	

Notes:
1. t_{CISKEW} represents the total amount of skew consumed by the controller between MDQS[n] and any corresponding bit that is captured with MDQS[n]. Subtract this value from the total timing budget.
2. At recommended operating conditions with V_{DDDDR} (1.8 V or 1.5 V) ± 5%
3. The amount of skew that can be tolerated from MDQS to a corresponding MDQ signal is called t_{DISKEW}.This can be determined by the following equation: $t_{DISKEW} = \pm(T \div 4 - abs(t_{CISKEW}))$ where T is the clock period and $abs(t_{CISKEW})$ is the absolute value of t_{CISKEW}.

Figure 11 shows the DDR2 and DDR3 SDRAM interface input timing diagram.

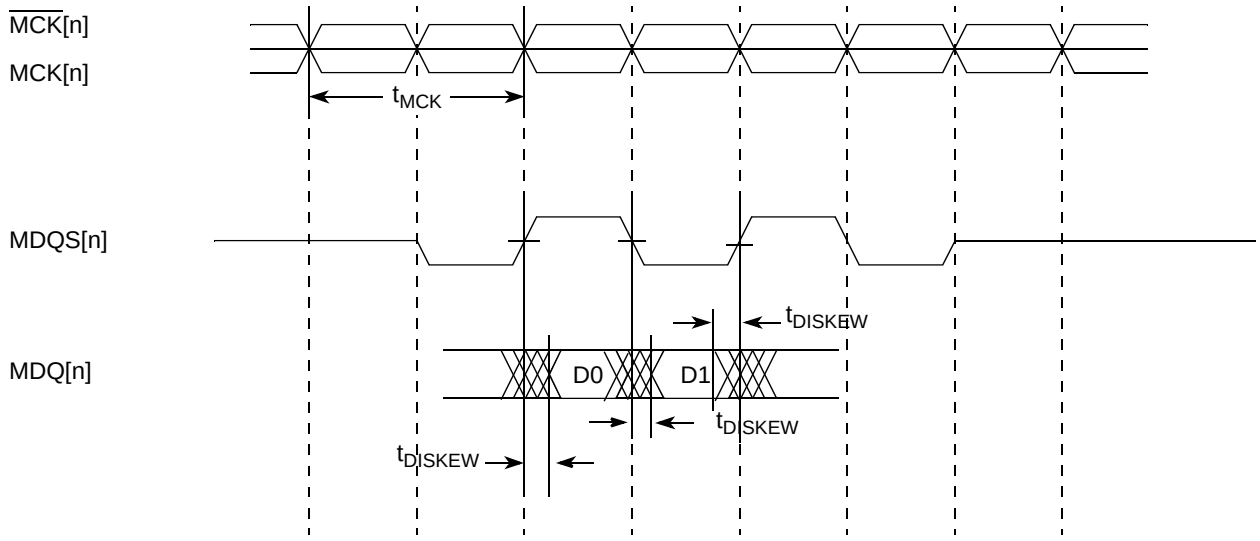


Figure 11. DDR2 and DDR3 SDRAM Interface Input Timing Diagram

2.6.1.2 DDR SDRAM Output AC Timing Specifications

Table 21 provides the output AC timing specifications for the DDR SDRAM interface.

Table 21. DDR SDRAM Output AC Timing Specifications

Parameter	Symbol [1]	Min	Max	Unit	Notes
MCK[n] cycle time	t_{MCK}	2.5	5	ns	2
ADDR/CMD output setup with respect to MCK • 800 MHz data rate • 667 MHz data rate	t_{DDKHAS}	 0.917 1.10	 — —	 ns ns	3
ADDR/CMD output hold with respect to MCK • 800 MHz data rate • 667 MHz data rate	t_{DDKHAX}	 0.767 1.02	 — —	 ns ns	3
$\overline{\text{MCS}}$n output setup with respect to MCK • 800 MHz data rate • 667 MHz data rate	t_{DDKHCS}	 0.917 1.10	 — —	 ns ns	3
$\overline{\text{MCS}}$n output hold with respect to MCK • 800 MHz data rate • 667 MHz data rate	t_{DDKHCX}	 0.767 1.02	 — —	 ns ns	3
MCK to MDQS Skew • 800 MHz data rate • 667 MHz data rate	t_{DDKHMH}	 –0.4 –0.6	 0.375 0.6	ns	4
MDQ/MECC/MDM output setup with respect to MDQS • 800 MHz • 667 MHz	t_{DDKHDS}, t_{DDKLDS}	 300 375	 — —	 ps ps	5
MDQ/MECC/MDM output hold with respect to MDQS • 800 MHz • 667 MHz	t_{DDKHDX}, t_{DDKLDX}	 300 375	 — —	 ps ps	5
MDQS preamble	t_{DDKHMP}	$-0.9 \times t_{MCK}$	—	ns	—
MDQS postamble	t_{DDKHME}	$-0.4 \times t_{MCK}$	$-0.6 \times t_{MCK}$	ns	—

Table 24. SR[1–2]_REF_CLK and $\overline{\text{SR[1–2]\_REF\_CLK}}$ Input Clock Requirements (continued)

Parameter	Symbol	Min	Typical	Max	Units	Notes

Notes:
1. Caution: Only 100 and 125 have been tested. Other values will not work correctly with the rest of the system.
2. Limits from PCI Express CEM Rev 1.0a
3. Measured from –200 mV to +200 mV on the differential waveform (derived from SR[1–2]_REF_CLK minus $\overline{\text{SR[1–2]\_REF\_CLK}}$). The signal must be monotonic through the measurement region for rise and fall time. The 400 mV measurement window is centered on the differential zero crossing. See Figure 16.
4. Measurement taken from differential waveform
5. Measurement taken from single-ended waveform
6. Matching applies to rising edge for SR[1–2]_REF_CLK and falling edge rate for $\overline{\text{SR[1–2]\_REF\_CLK}}$. It is measured using a 200 mV window centered on the median cross point where SR[1–2]_REF_CLK rising meets $\overline{\text{SR[1–2]\_REF\_CLK}}$ falling. The median cross point is used to calculate the voltage thresholds that the oscilloscope uses for the edge rate calculations. The rise edge rate of SR[1–2]_REF_CLK should be compared to the fall edge rate of $\overline{\text{SR[1–2]\_REF\_CLK}}$; the maximum allowed difference should not exceed 20% of the slowest edge rate. See Figure 17.

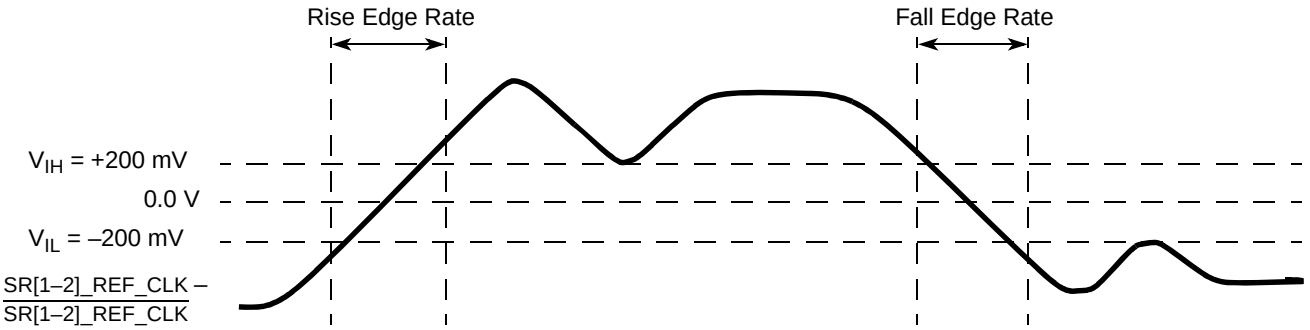


Figure 16. Differential Measurement Points for Rise and Fall Time

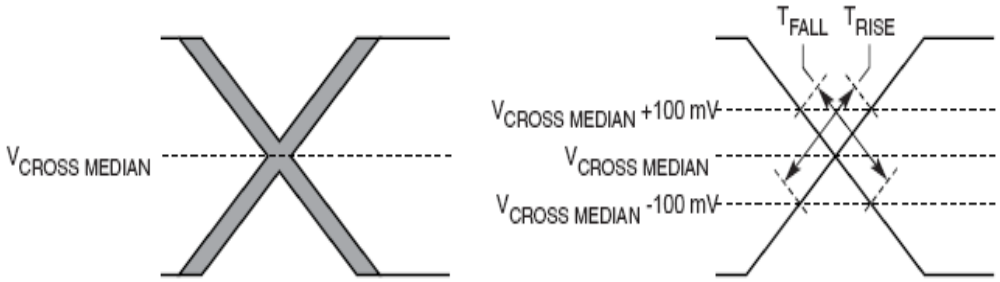


Figure 17. Single-Ended Measurement Points for Rise and Fall Time Matching

3.1.2 Power-On Ramp Time

This section describes the AC electrical specification for the power-on ramp rate requirements for all voltage supplies (including GVDD/SXPVDD/SXCVDD/QVDD/GVDD/NVDD, all VDD supplies, MVREF, and all AVDD supplies). Controlling the power-on ramp time is required to avoid falsely triggering the ESD circuitry. Table 39 defines the power supply ramp time specification.

Table 39. Power Supply Ramp Rate

Parameter	Min	Max	Unit
Required ramp rate.	—	36000	V/s

Notes:
1. Ramp time is specified as a linear ramp from 10% to 90% of nominal voltage of the specific voltage supply. If the ramp is non-linear (for example, exponential), the maximum rate of change from 200 to 500 mV is the most critical because this range might falsely trigger the ESD circuitry.
2. Required over the full recommended operating temperature range (see Table 3).
3. All supplies must be at their stable values within 50 ms.
4. The GVDD pins can be held low on the application board at powerup. If GVDD is not held low, then GVDD will rise to a voltage level that depends on the board-level impedance-to-ground. If the impedance is high (that is, infinite), then theoretically, GVDD can rise up close to the VDD levels.

3.1.3 Power Supply Guidelines

Use the following guidelines for power-up sequencing:

- Couple M3VDD with the VDD power rail using an extremely low impedance path.
- Couple inputs PLL1_AVDD, PLL2_AVDD and PLL3_AVDD with the VDD power rail using an RC filter (see Figure 37).
- There is no dependency in power-on/power-off sequence between the GVDD1, GVDD2, NVDD, and QVDD power rails.
- Couple inputs M1VREF and M2VREF with the GVDD1 and GVDD2 power rails, respectively. They should rise at the same time as or after their respective power rail.
- There is no dependency between RapidIO supplies: SXCVDD1, SXCVDD2, SXPVDD1 and SXPVDD2 and other MSC8256 supplies in the power-on/power-off sequence
- Couple inputs SR1_PLL_AVDD and SR2_PLL_AVDD with SXCVDD1 and SXCVDD2 power rails, respectively, using an RC filter (see Figure 38).

External voltage applied to any input line must not exceed the I/O supply voltage related to this line by more than 0.6 V at any time, including during power-up. Some designs require pull-up voltages applied to selected input lines during power-up for configuration purposes. This is an acceptable exception to the rule during start-up. However, each such input can draw up to 80 mA per input pin per MSC8256 device in the system during power-up. An assertion of the inputs to the high voltage level before power-up should be with slew rate less than 4 V/ns.

The device power rails should rise in the following sequence:

1. VDD (and all coupled supplies)

2. After the above rails rise to 90% of their nominal voltage, the following I/O power rails may rise in any sequence (see Figure 34): QVDD, NVDD, GVDD1, and GVDD2.

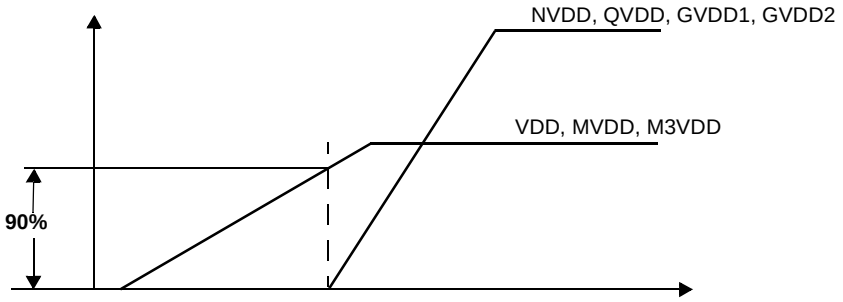


Figure 34. Supply Ramp-Up Sequence

Notes:
1. If the M3 memory is not used, M3VDD can be tied to GND.
2. If the HSSI port1 is not used, SXCVDD1and SXPVDD1 must be connected to the designated power supplies.
3. If the HSSI port2 is not used, SXCVDD2 and SXPVDD2 must be connected to the designated power supplies.
4. If the DDR port 1 interface is not used, it is recommended that GVDD1 be left unconnected.
5. If the DDR port 2 interface is not used, it is recommended that GVDD2 be left unconnected.

3.1.4 Reset Guidelines

When a debugger is not used, implement the connection scheme shown in Figure 35.

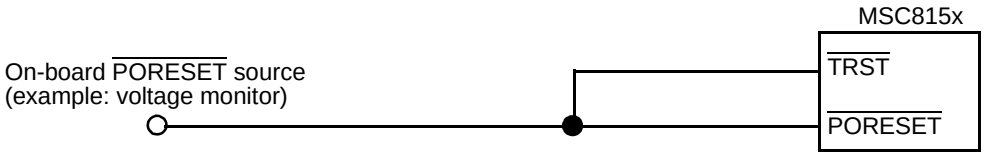


Figure 35. Reset Connection in Functional Application

When a debugger is used, implement the connection scheme shown in Figure 36.

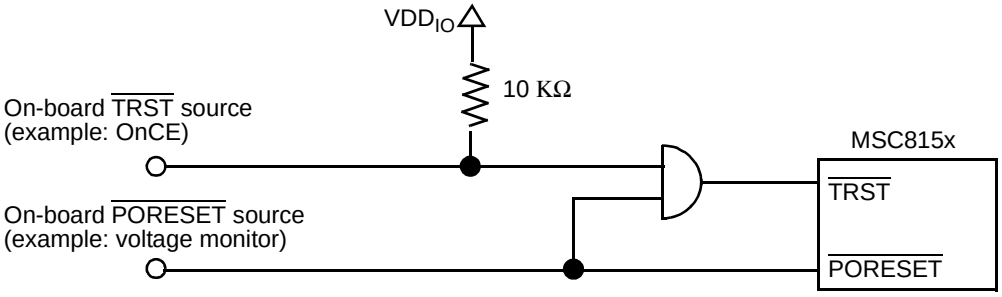


Figure 36. Reset Connection in Debugger Application

3.2 PLL Power Supply Design Considerations

Each global PLL power supply must have an external RC filter for the PLLn_AVDD input (see Figure 37) in which the following components are defined as listed:

- R = 5 Ω ± 5%
- C1 = 10 µF ± 10%, 0603, X5R, with ESL ≤ 0.5 nH, low ESL Surface Mount Capacitor.
- C2 = 1.0 µF ± 10%, 0402, X5R, with ESL ≤ 0.5 nH, low ESL Surface Mount Capacitor.

Note: A higher capacitance value for C2 may be used to improve the filter as long as the other C2 parameters do not change.

All three PLLs can connect to a single supply voltage source (such as a voltage regulator) as long as the external RC filter is applied to each PLL separately. For optimal noise filtering, place the circuit as close as possible to its PLLn_AVDD inputs.

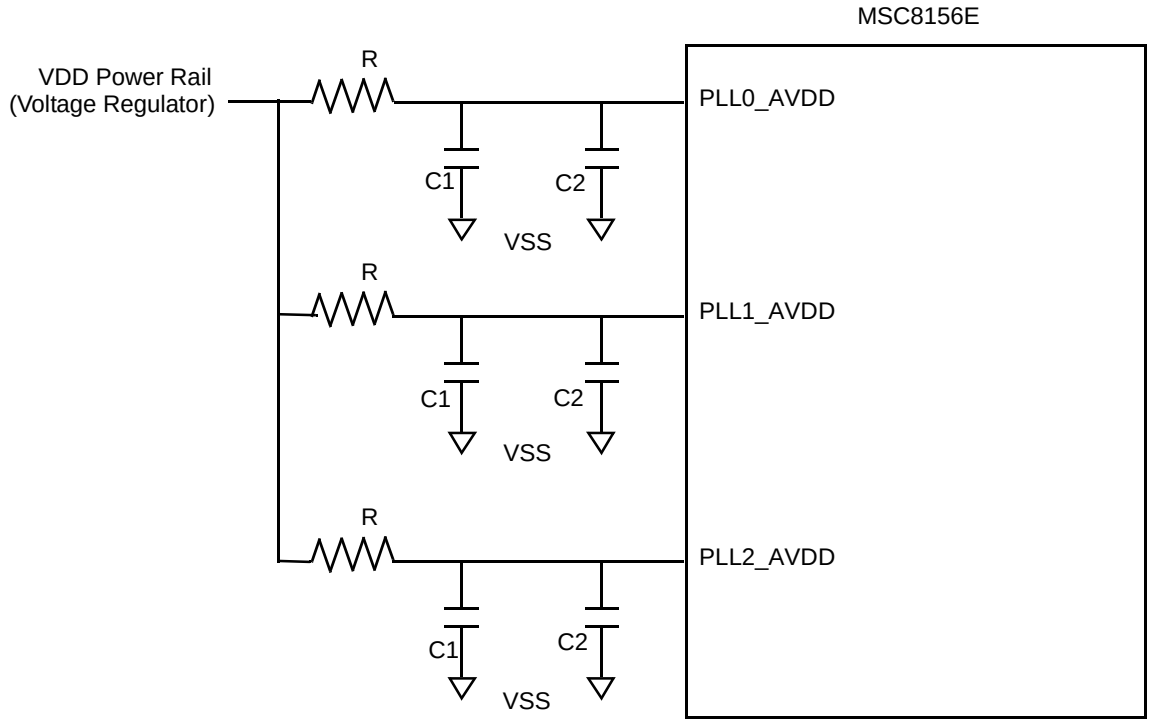


Figure 37. PLL Supplies

Each SerDes PLL power supply must be filtered using a circuit similar to the one shown in Figure 38, to ensure stability of the internal clock. For maximum effectiveness, the filter circuit should be placed as closely as possible to the SRn_PLL_AVDD ball to ensure it filters out as much noise as possible. The ground connection should be near the SRn_PLL_AVDD ball. The 0.003 µF capacitor is closest to the ball, followed by the two 2.2 µF capacitors, and finally the 1 Ω resistor to the board supply plane. The capacitors are connected from SRn_PLL_AVDD to the ground plane. Use ceramic chip capacitors with the highest possible self-resonant frequency. All trances should be kept short, wide, and direct.

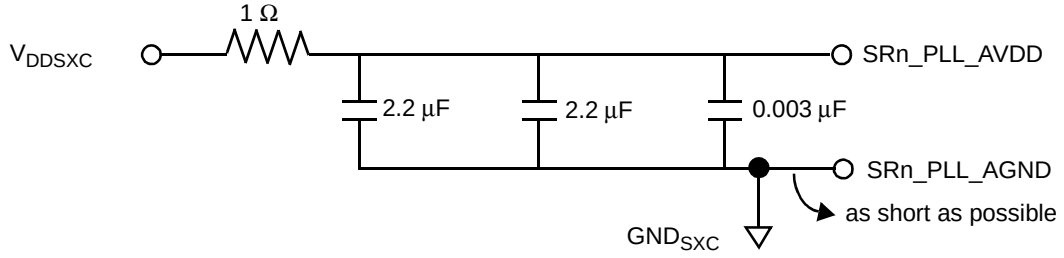


Figure 38. SerDes PLL Supplies

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