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Applications of "<u>Embedded -</u> <u>Microcontrollers</u>"

Details

E·XFI

Product Status	Active
Core Processor	PIC
Core Size	8-Bit
Speed	48MHz
Connectivity	I ² C, SPI, UART/USART
Peripherals	Brown-out Detect/Reset, POR, PWM, WDT
Number of I/O	34
Program Memory Size	32KB (16K x 16)
Program Memory Type	FLASH
EEPROM Size	-
RAM Size	3.8K x 8
Voltage - Supply (Vcc/Vdd)	2V ~ 3.6V
Data Converters	A/D 13x10b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	44-TQFP
Supplier Device Package	44-TQFP (10x10)
Purchase URL	https://www.e-xfl.com/product-detail/microchip-technology/pic18lf45j11t-i-pt

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Address: Room A, 16/F, Full Win Commercial Centre, 573 Nathan Road, Mongkok, Hong Kong

	Pin Nu	umber			
Pin Name	28-SPDIP/ SSOP/ SOIC	28-QFN	Pin Type	Buffer Type	Description
MCLR	1	26	I	ST	Master Clear (Reset) input. This pin is an active-low Reset to the device.
OSC1/CLKI/RA7	9	6			
OSC1			I	ST	Oscillator crystal input or external clock source input. ST buffer when configured in RC mode; CMOS otherwise. Main oscillator input connection.
CLKI			Ι	CMOS	External clock source input; always associated with pin function OSC1 (see related OSC1/CLKI pins).
RA7 ⁽¹⁾			I/O	TTL	Digital I/O.
OSC2/CLKO/RA6 OSC2	10	7	0	_	Oscillator crystal or clock output. Oscillator crystal output. Connects to crystal or resonator in Crystal Oscillator mode.
CLKO			0	_	Main oscillator feedback output connection. In RC mode, OSC2 pin outputs CLKO, which has 1/4 the frequency of OSC1 and denotes the instruction cycle rate.
RA6 ⁽¹⁾			I/O	TTL	Digital I/O.
Legend: TTL = TTL compatient	tible input			С	MOS = CMOS compatible input or output
SI = Schmitt Trig	iger input w	Ith CMOS	levels	A	nalog = Analog input
I = Input				0	= Output = Onen Drain (no R diado to Von)
DIG = Digital outpu	ıt			0	

TABLE 1-3:PIC18F2XJ11 PINOUT I/O DESCRIPTIONS

Note 1: RA7 and RA6 will be disabled if OSC1 and OSC2 are used for the clock function.

4.4.3 RC_IDLE MODE

In RC_IDLE mode, the CPU is disabled but the peripherals continue to be clocked from the internal oscillator block. This mode allows for controllable power conservation during Idle periods.

From RC_RUN, this mode is entered by setting the IDLEN bit and executing a SLEEP instruction. If the device is in another Run mode, first set IDLEN, then clear the SCS bits and execute SLEEP. When the clock source is switched to the INTOSC block, the primary oscillator is shut down and the OSTS bit is cleared.

When a wake event occurs, the peripherals continue to be clocked from the internal oscillator block. After the wake event, the CPU begins executing code being clocked by the INTRC. The IDLEN and SCS bits are not affected by the wake-up. The INTRC source will continue to run if either the WDT or the FSCM is enabled.

4.5 Exiting Idle and Sleep Modes

An exit from Sleep mode, or any of the Idle modes, is triggered by an interrupt, a Reset or a WDT time-out. This section discusses the triggers that cause exits from power-managed modes. The clocking subsystem actions are discussed in each of the power-managed modes sections (see Section 4.2 "Run Modes", Section 4.3 "Sleep Mode" and Section 4.4 "Idle Modes").

4.5.1 EXIT BY INTERRUPT

Any of the available interrupt sources can cause the device to exit from an Idle mode, or the Sleep mode, to a Run mode. To enable this functionality, an interrupt source must be enabled by setting its enable bit in one of the INTCON or PIE registers. The exit sequence is initiated when the corresponding interrupt flag bit is set.

On all exits from Idle or Sleep modes by interrupt, code execution branches to the interrupt vector if the GIE/GIEH bit (INTCON<7>) is set. Otherwise, code execution continues or resumes without branching (see Section 9.0 "Interrupts").

4.5.2 EXIT BY WDT TIME-OUT

A WDT time-out will cause different actions depending on which power-managed mode the device is in when the time-out occurs. If the device is not executing code (all Idle modes and Sleep mode), the time-out will result in an exit from the power-managed mode (see Section 4.2 "Run Modes" and Section 4.3 "Sleep Mode"). If the device is executing code (all Run modes), the time-out will result in a WDT Reset (see Section 26.2 "Watchdog Timer (WDT)").

The WDT and postscaler are cleared by one of the following events:

- Executing a SLEEP or CLRWDT instruction
- The loss of a currently selected clock source (if the FSCM is enabled)

4.5.3 EXIT BY RESET

Exiting an Idle or Sleep mode by Reset automatically forces the device to run from the INTRC.

4.5.4 EXIT WITHOUT AN OSCILLATOR START-UP DELAY

Certain exits from power-managed modes do not invoke the OST at all. There are two cases:

- PRI_IDLE mode (where the primary clock source is not stopped) and the primary clock source is the EC mode
- PRI_IDLE mode and the primary clock source is the ECPLL mode

In these instances, the primary clock source either does not require an oscillator start-up delay, since it is already running (PRI_IDLE), or normally does not require an oscillator start-up delay (EC).

4.6 Deep Sleep Mode

Deep Sleep mode brings the device into its lowest power consumption state without requiring the use of external switches to remove power from the device. During deep sleep, the on-chip VDDCORE voltage regulator is powered down, effectively disconnecting power to the core logic of the microcontroller.

Note: Since Deep Sleep mode powers down the microcontroller by turning off the on-chip VDDCORE voltage regulator, Deep Sleep capability is available only on PIC18FXXJ members in the device family. The on-chip voltage regulator is not available in PIC18LFXXJ members of the device family, and therefore, they do not support Deep Sleep.

TABLE 5-2.	INITIALIZAT		INS FOR ALL REGI	STERS (CONTINUEL	<i>'</i>)
Register	Applicabl	e Devices	Power-on Reset, Brown-out Reset, Wake From Deep Sleep	MCLR Resets WDT Reset RESET Instruction Stack Resets CM Resets	Wake-up via WDT or Interrupt
PMDOUT2H ⁽⁵⁾		PIC18F4XJ11	0000 0000	0000 0000	uuuu uuuu
PMDOUT2L ⁽⁵⁾	—	PIC18F4XJ11	0000 0000	0000 0000	uuuu uuuu
PMDIN2H ⁽⁵⁾	—	PIC18F4XJ11	0000 0000	0000 0000	uuuu uuuu
PMDIN2L ⁽⁵⁾	—	PIC18F4XJ11	0000 0000	0000 0000	uuuu uuuu
PMEH ⁽⁵⁾	_	PIC18F4XJ11	0000 0000	0000 0000	uuuu uuuu
PMEL ⁽⁵⁾	—	PIC18F4XJ11	0000 0000	0000 0000	uuuu uuuu
PMSTATH ⁽⁵⁾	—	PIC18F4XJ11	00 0000	00 0000	uu uuuu
PMSTATL ⁽⁵⁾	—	PIC18F4XJ11	10 1111	10 1111	uu uuuu
CVRCON	PIC18F2XJ11	PIC18F4XJ11	0000 0000	0000 0000	uuuu uuuu
TCLKCON	PIC18F2XJ11	PIC18F4XJ11	000	0uu	uuu
DSGPR1 ⁽⁶⁾	PIC18F2XJ11	PIC18F4XJ11	uuuu uuuu	սսսս սսսս	uuuu uuuu
DSGPR0 ⁽⁶⁾	PIC18F2XJ11	PIC18F4XJ11	uuuu uuuu	uuuu uuuu	uuuu uuuu
DSCONH ⁽⁶⁾	PIC18F2XJ11	PIC18F4XJ11	0000	0uuu	uuuu
DSCONL ⁽⁶⁾	PIC18F2XJ11	PIC18F4XJ11	000	u00	uuu
DSWAKEH ⁽⁶⁾	PIC18F2XJ11	PIC18F4XJ11	0	0	u
DSWAKEL ⁽⁶⁾	PIC18F2XJ11	PIC18F4XJ11	0-00 00-1	0-00 00-0	u-uu uu-u
ANCON1	PIC18F2XJ11	PIC18F4XJ11	00-0 0000	00-0 0000	uu-u uuuu
ANCON0	PIC18F2XJ11	PIC18F4XJ11	0000 0000	0000 0000	uuuu uuuu
ODCON1	PIC18F2XJ11	PIC18F4XJ11	00	uu	uu
ODCON2	PIC18F2XJ11	PIC18F4XJ11	00	uu	uu
ODCON3	PIC18F2XJ11	PIC18F4XJ11	00	uu	uu
RTCCFG	PIC18F2XJ11	PIC18F4XJ11	0-00 0000	u-uu uuuu	u-uu uuuu
RTCCAL	PIC18F2XJ11	PIC18F4XJ11	0000 0000	uuuu uuuu	uuuu uuuu
REFOCON	PIC18F2XJ11	PIC18F4XJ11	0-00 0000	0-00 0000	u-uu uuuu
PADCFG1	PIC18F2XJ11	PIC18F4XJ11	000	000	uuu
PPSCON	PIC18F2XJ11	PIC18F4XJ11	0	0	u
RPINR24	PIC18F2XJ11	PIC18F4XJ11	1 1111	1 1111	u uuuu
RPINR23	PIC18F2XJ11	PIC18F4XJ11	1 1111	1 1111	u uuuu
RPINR22	PIC18F2XJ11	PIC18F4XJ11	1 1111	1 1111	u uuuu
RPINR21	PIC18F2XJ11	PIC18F4XJ11	1 1111	1 1111	u uuuu
RPINR17	PIC18F2XJ11	PIC18F4XJ11	1 1111	1 1111	u uuuu
RPINR16	PIC18F2XJ11	PIC18F4XJ11	1 1111	1 1111	u uuuu

TABLE 5-2: INITIALIZATION CONDITIONS FOR ALL REGISTERS (CONTINUED)

Legend: u = unchanged, x = unknown, - = unimplemented bit, read as '0', q = value depends on condition.

Note 1: When the wake-up is due to an interrupt and the GIEL or GIEH bit is set, the TOSU, TOSH and TOSL are updated with the current value of the PC. The STKPTR is modified to point to the next location in the hardware stack.

2: When the wake-up is due to an interrupt and the GIEL or GIEH bit is set, the PC is loaded with the interrupt vector (0008h or 0018h).

3: One or more bits in the INTCONx or PIRx registers will be affected (to cause wake-up).

- 4: See Table 5-1 for Reset value for specific condition.
- 5: Not implemented for PIC18F2XJ11 devices.
- 6: Not implemented on "LF" devices.

6.6.1 INDEXED ADDRESSING WITH LITERAL OFFSET

Enabling the PIC18 extended instruction set changes the behavior of Indirect Addressing using the FSR2 register pair and its associated file operands. Under proper conditions, instructions that use the Access Bank, that is, most bit and byte-oriented instructions, can invoke a form of Indexed Addressing using an offset specified in the instruction. This special addressing mode is known as Indexed Addressing with Literal Offset, or Indexed Literal Offset mode.

When using the extended instruction set, this addressing mode requires the following:

- The use of the Access Bank is forced ('a' = 0); and
- The file address argument is less than or equal to 5Fh.

Under these conditions, the file address of the instruction is not interpreted as the lower byte of an address (used with the BSR in Direct Addressing) or as an 8-bit address in the Access Bank. Instead, the value is interpreted as an offset value to an Address Pointer specified by FSR2. The offset and the contents of FSR2 are added to obtain the target address of the operation.

6.6.2 INSTRUCTIONS AFFECTED BY INDEXED LITERAL OFFSET MODE

Any of the core PIC18 instructions that can use Direct Addressing are potentially affected by the Indexed Literal Offset Addressing mode. This includes all byte and bit-oriented instructions, or almost one-half of the standard PIC18 instruction set. Instructions that only use Inherent or Literal Addressing modes are unaffected.

Additionally, byte and bit-oriented instructions are not affected if they do not use the Access Bank (Access RAM bit is '1') or include a file address of 60h or above. Instructions meeting these criteria will continue to execute as before. A comparison of the different possible addressing modes when the extended instruction set is enabled is provided in Figure 6-9.

Those who desire to use byte or bit-oriented instructions in the Indexed Literal Offset mode should note the changes to assembler syntax for this mode. This is described in more detail in **Section 27.2.1 "Extended Instruction Syntax"**.

FIGURE 6-9: COMPARING ADDRESSING OPTIONS FOR BIT-ORIENTED AND BYTE-ORIENTED **INSTRUCTIONS (EXTENDED INSTRUCTION SET ENABLED)**

EXAMPLE INSTRUCTION: ADDWF, f, d, a (Opcode: 0010 01da ffff fff) 000h When a = 0 and $f \ge 60h$: The instruction executes in 060h Direct Forced mode. 'f' is Bank 0 interpreted as a location in the 100h Access RAM between 060h 00h Bank 1 and FFFh. This is the same as through Bank 14 60ŀ locations F60h to FFFh Valid range (Bank 15) of data memory. for 'f Locations below 060h are not F00h Access RAM available in this addressing Bank 15 mode. F60h SFRs FFFh Data Memory When a = 0 and f < 5Fh: 000h Bank 0 The instruction executes in 060h 100h 001001da ffffffff Bank 1 Ŧ through Bank 14 FSR2L FSR2H F00h Note that in this mode, the Bank 15 correct syntax is: F60h ADDWF [k], d SFRs where 'k' is same as 'f'. FFFh Data Memory BSR When a = 1 (all values of f): 000h 00000000 Bank 0 The instruction executes in 060h Direct mode (also known as Direct Long mode). 'f' is 100h interpreted as a location in one of the 16 banks of the data 001001da fffffff Bank 1 through memory space. The bank is Bank 14 designated by the Bank Select Register (BSR). The address can be in any implemented F00h bank in the data memory Bank 15 space. F60h SFRs FFFh

Data Memory

Indexed Literal Offset mode. 'f' is interpreted as an offset to the address value in FSR2. The two are added together to obtain the address of the target register for the instruction. The address can be anywhere in the data memory space.

R/W-0	R/W-0	R/W-0	U-0	R/W-0	R/W-0	R/W-0	R/W-0
OSCFIE	CM2IE	CM1IE	—	BCL1IE	LVDIE	TMR3IE	CCP2IE
bit 7							bit 0
Legend:							
R = Readable	e bit	W = Writable	bit	U = Unimplem	nented bit, read	d as '0'	
-n = Value at	POR	'1' = Bit is set		'0' = Bit is clea	ared	x = Bit is unkn	iown
bit 7	OSCFIE: Osc 1 = Enabled 0 = Disabled	illator Fail Inter	rupt Enable bi	t			
bit 6	CM2IE: Comp 1 = Enabled 0 = Disabled	barator 2 Interro	upt Enable bit				
bit 5	CM1IE: Comp 1 = Enabled 0 = Disabled	parator 1 Interro	upt Enable bit				
bit 4	Unimplemen	ted: Read as ')'				
bit 3	BCL1IE: Bus 1 = Enabled 0 = Disabled	Collision Interr	upt Enable bit	(MSSP1 modul	e)		
bit 2	LVDIE: High/l 1 = Enabled 0 = Disabled	_ow-Voltage De	etect Interrupt I	Enable bit			
bit 1	TMR3IE: TMF 1 = Enabled 0 = Disabled	R3 Overflow Int	errupt Enable	bit			
bit 0	CCP2IE: ECC 1 = Enabled 0 = Disabled	CP2 Interrupt E	nable bit				

REGISTER 9-8: PIE2: PERIPHERAL INTERRUPT ENABLE REGISTER 2 (ACCESS FA0h)

10.0 I/O PORTS

Depending on the device selected and features enabled, there are up to five ports available. Some pins of the I/O ports are multiplexed with an alternate function from the peripheral features on the device. In general, when a peripheral is enabled, that pin may not be used as a general purpose I/O pin.

Each port has three registers for its operation. These registers are:

- TRIS register (Data Direction register)
- PORT register (reads the levels on the pins of the device)
- LAT register (Data Latch)

The Data Latch (LAT register) is useful for read-modifywrite operations on the value that the I/O pins are driving.

Figure 10-1 displays a simplified model of a generic I/O port, without the interfaces to other peripherals.

FIGURE 10-1: GENERIC I/O PORT OPERATION



10.1 I/O Port Pin Capabilities

When developing an application, the capabilities of the port pins must be considered. Outputs on some pins have higher output drive strength than others. Similarly, some pins can tolerate higher than VDD input levels.

10.1.1 PIN OUTPUT DRIVE

The output pin drive strengths vary for groups of pins intended to meet the needs for a variety of applications. PORTB and PORTC are designed to drive higher loads, such as LEDs. All other ports are designed for small loads, typically indication only. Table 10-1 summarizes the output capabilities. Refer to **Section 29.0 "Electrical Characteristics"** for more details.

TABLE 10-1: OUTPUT DRIVE LEVELS

Port	Drive	Description
PORTA (except RA6)		
PORTD	Minimum	Intended for indication.
PORTE		
PORTB		
PORTC	High	Suitable for direct LED
PORTA<6>		

10.1.2 INPUT PINS AND VOLTAGE CONSIDERATIONS

The voltage tolerance of pins used as device inputs is dependent on the pin's input function. Pins that are used as digital only inputs are able to handle DC voltages up to 5.5V; a level typical for digital logic circuits. In contrast, pins that also have analog input functions of any kind can only tolerate voltages up to VDD. Voltage excursions beyond VDD on these pins should be avoided. Table 10-2 summarizes the input capabilities. Refer to **Section 29.0 "Electrical Characteristics"** for more details.

TABLE 10-2: INPUT VOLTAGE LEVELS

Port or Pin	Tolerated Input	Description
PORTA<7:0>		
PORTB<3:0>	Vpp	Only VDD input levels
PORTC<2:0>	VDD	tolerated.
PORTE<2:0>		
PORTB<7:4>		Tolerates input levels
PORTC<7:3>	5.5V	above VDD, useful for
PORTD<7:0>		most standard logic.

Pin	Function	TRIS Setting	I/O	I/O Type	Description
RD0/PMD0/	RD0	1	Ι	ST	PORTD<0> data input.
SCL2		0	0	DIG	LATD<0> data output.
	PMD0	1	Ι	ST/TTL	Parallel Master Port data in.
		0	0	DIG	Parallel Master Port data out.
	SCL2	1	I	I ² C/ SMB	I ² C [™] clock input (MSSP2 module); input type depends on module setting.
		0	0	DIG	I ² C [™] clock output (MSSP2 module); takes priority over port data.
RD1/PMD1/	RD1	1	I	ST	PORTD<1> data input.
SDA2		0	0	DIG	LATD<1> data output.
	PMD1	1	I	ST/TTL	Parallel Master Port data in.
		0	0	DIG	Parallel Master Port data out.
	SDA2	1	I	I ² C/ SMB	I ² C data input (MSSP2 module); input type depends on module setting.
		0	0	DIG	I ² C data output (MSSP2 module); takes priority over port data.
RD2/PMD2/	RD2	1	Ι	ST	PORTD<2> data input.
RP19		0	0	DIG	LATD<2> data output.
	PMD2	1	Ι	ST/TTL	Parallel Master Port data in.
		0	0	DIG	Parallel Master Port data out.
	RP19	1	I	ST	Remappable peripheral pin 19 input.
		0	0	DIG	Remappable peripheral pin 19 output.
RD3/PMD3/	RD3	1	Ι	DIG	PORTD<3> data input.
RP20		0	0	DIG	LATD<3> data output.
	PMD3	1	I	ST/TTL	Parallel Master Port data in.
		0	0	DIG	Parallel Master Port data out.
	RP20	1	I	ST	Remappable peripheral pin 20 input.
		0	0	DIG	Remappable peripheral pin 20 output.
RD4/PMD4/	RD4	1	Ι	ST	PORTD<4> data input.
RP21		0	0	DIG	LATD<4> data output.
	PMD4	1	Ι	ST/TTL	Parallel Master Port data in.
		0	0	DIG	Parallel Master Port data out.
	RP21	1	I	ST	Remappable peripheral pin 21 input.
		0	0	DIG	Remappable peripheral pin 21 output.
RD5/PMD5/	RD5	1	Ι	ST	PORTD<5> data input.
RP22		0	0	DIG	LATD<5> data output.
	PMD5	1	I	ST/TTL	Parallel Master Port data in.
		0	0	DIG	Parallel Master Port data out.
	RP22	1	I	ST	Remappable peripheral pin 22 input.
		0	0	DIG	Remappable peripheral pin 22 output.

TABLE 10-9: PORTD I/O SUMMARY

 0
 0
 DIG
 Remappable peripheral pin 22 output.

 Legend:
 DIG = Digital level output; TTL = TTL input buffer; ST = Schmitt Trigger input buffer; I²C/SMB = I²C/SMBus input buffer; x = Don't care (TRIS bit does not affect port direction or is overridden for this option).

10.6 PORTE, TRISE and LATE Registers

Note:	PORTE	is	available	only	in	44-pin
	devices.					

Depending on the particular PIC18F46J11 family device selected, PORTE is implemented in two different ways.

For 44-pin devices, PORTE is a 3-bit wide port. Three pins (RE0/AN5/PMRD, RE1/AN6/PMWR and RE2/ AN7/PMCS) are individually configurable as inputs or outputs. These pins have Schmitt Trigger input buffers. When selected as analog inputs, these pins will read as '0's.

The corresponding Data Direction register is TRISE. Setting a TRISE bit (= 1) will make the corresponding PORTE pin an input (i.e., put the corresponding output driver in a high-impedance mode). Clearing a TRISE bit (= 0) will make the corresponding PORTE pin an output (i.e., put the contents of the output latch on the selected pin).

TRISE controls the direction of the RE pins, even when they are being used as analog inputs. The user must make sure to keep the pins configured as inputs when using them as analog inputs.

Note: On a POR, RE<2:0> are configured as analog inputs.

The Data Latch register (LATE) is also memory mapped. Read-modify-write operations on the LATE register read and write the latched output value for PORTE.

EXAMPLE 10-6:	INITIALIZING PORTE

CLRF	LATE	;	Initialize LATE
		;	to clear output
		;	data latches
MOVLW	0xE0	;	Configure REx
MOVWF	ANCON0	;	for digital inputs
MOVLW	0x03	;	Value used to
		;	initialize data
		;	direction
MOVWF	TRISE	;	Set RE<0> as inputs
		;	RE<1> as outputs
		;	RE<2> as inputs

Each of the PORTE pins has a weak internal pull-up. A single control bit can turn on all the pull-ups. This is performed by setting bit, REPU (PORTE<6>). The weak pull-up is automatically turned off when the port pin is configured as an output. The pull-ups are disabled on a POR.

Note that the pull-ups can be used for any set of features, similar to the pull-ups found on PORTB.

13.4 Timer1 16-Bit Read/Write Mode

Timer1 can be configured for 16-bit reads and writes. When the RD16 control bit (T1CON<1>) is set, the address for TMR1H is mapped to a buffer register for the high byte of Timer1. A read from TMR1L loads the contents of the high byte of Timer1 into the Timer1 High Byte Buffer register. This provides the user with the ability to accurately read all 16 bits of Timer1 without having to determine whether a read of the high byte, followed by a read of the low byte, has become invalid due to a rollover between reads.

A write to the high byte of Timer1 must also take place through the TMR1H Buffer register. The Timer1 high byte is updated with the contents of TMR1H when a write occurs to TMR1L. This allows a user to write all 16 bits to both the high and low bytes of Timer1 at once.

The high byte of Timer1 is not directly readable or writable in this mode. All reads and writes must take place through the Timer1 High Byte Buffer register. Writes to TMR1H do not clear the Timer1 prescaler. The prescaler is only cleared on writes to TMR1L.

13.5 Timer1 Oscillator

An on-chip crystal oscillator circuit is incorporated between pins, T1OSI (input) and T1OSO (amplifier output). It is enabled by setting the Timer1 Oscillator Enable bit, T1OSCEN (T1CON<3>). The oscillator is a low-power circuit rated for 32 kHz crystals. It will continue to run during all power-managed modes. The circuit for a typical LP oscillator is depicted in Figure 13-2. Table 13-2 provides the capacitor selection for the Timer1 oscillator.

The user must provide a software time delay to ensure proper start-up of the Timer1 oscillator.

FIGURE 13-2: EXTERNAL COMPONENTS FOR THE TIMER1 LP OSCILLATOR



TABLE 13-2:CAPACITOR SELECTION FOR
THE TIMER
OSCILLATOR^(2,3,4,5)

Oscillator Type	Freq.	C1	C2
LP	32 kHz	12 pF ⁽¹⁾	12 pF ⁽¹⁾

- Note 1: Microchip suggests these values as a starting point in validating the oscillator circuit.
 - 2: Higher capacitance increases the stability of the oscillator but also increases the start-up time.
 - 3: Since each resonator/crystal has its own characteristics, the user should consult the resonator/crystal manufacturer for appropriate values of external components.
 - 4: Capacitor values are for design guidance only. Values listed would be typical of a CL = 10 pF rated crystal, when LPT1OSC = 1.
 - Incorrect capacitance value may result in a frequency not meeting the crystal manufacturer's tolerance specification.

The Timer1 crystal oscillator drive level is determined based on the LPT1OSC (CONFIG2L<4>) Configuration bit. The higher drive level mode, LPT1OSC = 1, is intended to drive a wide variety of 32.768 kHz crystals with a variety of load capacitance (CL) ratings.

The lower drive level mode is highly optimized for extremely low-power consumption. It is not intended to drive all types of 32.768 kHz crystals. In the low drive level mode, the crystal oscillator circuit may not work if excessively large discrete capacitors are placed on the T1OSI and T1OSO pins. This mode is only designed to work with discrete capacitances of approximately 3 pF-10 pF on each pin.

Crystal manufacturers usually specify a CL (load capacitance) rating for their crystals. This value is related to, but not necessarily the same as, the values that should be used for C1 and C2 in Figure 13-2. See the crystal manufacturer's applications' information for more details on how to select the optimum C1 and C2 for a given crystal. The optimum value depends in part on the amount of parasitic capacitance in the circuit, which is often unknown. Therefore, after values have been selected, it is highly recommended that thorough testing and validation of the oscillator be performed.

FIGURE 13-7: TIMER1 GATE SINGLE PULSE AND TOGGLE COMBINED MODE TMR1GE T1GPOL T1GSPM T1GTM Cleared by Hardware on T1GGO/ Set by Software Falling Edge of T1GVAL T1DONE Counting Enabled on Rising Edge of T1G T1G_IN T1CKI T1GVAL Timer1 N + 1 Ν N + 2 N + 3 N + 4 Cleared by Software Set by Hardware on Cleared by Software Falling Edge of T1GVAL RTCCIF

|--|

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on Page:
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF	90
PIR1	PMPIF ⁽¹⁾	ADIF	RC1IF	TX1IF	SSP1IF	CCP1IF	TMR2IF	TMR1IF	92
PIE1	PMPIE ⁽¹⁾	ADIE	RC1IE	TX1IE	SSP1IE	CCP1IE	TMR2IE	TMR1IE	92
IPR1	PMPIP ⁽¹⁾	ADIP	RC1IP	TX1IP	SSP1IP	CCP1IP	TMR2IP	TMR1IP	92
TMR1L	Timer1 Register Low Byte						91		
TMR1H	Timer1 Register High Byte							91	
T1CON	TMR1CS1	TMR1CS0	T1CKPS1	T1CKPS0	T1OSCEN	T1SYNC	RD16	TMR10N	91
T1GCON	TMR1GE	T1GPOL	T1GTM	T1GSPM	T1GGO/ T1DONE	T1GVAL	T1GSS1	T1GSS0	92
TCLKCON		_	_	T1RUN			T3CCP2	T3CCP1	94

Legend: Shaded cells are not used by the Timer1 module.

Note 1: These bits are only available in 44-pin devices.

17.2.2 CLOCK SOURCE

As mentioned earlier, the RTCC module is intended to be clocked by an external Real-Time Clock crystal oscillating at 32.768 kHz, but also can be clocked by the INTRC oscillator. The RTCC clock selection is decided by the RTCOSC bit (CONFIG3L<1>). Calibration of the crystal can be done through this module to yield an error of 3 seconds or less per month. (For further details, see **Section 17.2.9 "Calibration**".)



17.2.2.1 Real-Time Clock Enable

The RTCC module can be clocked by an external, 32.768 kHz crystal (Timer1 oscillator) or the INTRC oscillator, which can be selected in CONFIG3L<1>.

If the Timer1 oscillator will be used as the clock source for the RTCC, make sure to enable it by setting T1CON<3> (T1OSCEN). The selected clock can be brought out to the RTCC pin by the RTSECSEL<1:0> bits in the PADCFG1 register.

17.2.3 DIGIT CARRY RULES

This section explains which timer values are affected when there is a rollover.

- Time of Day: From 23:59:59 to 00:00:00 with a carry to the Day field
- Month: From 12/31 to 01/01 with a carry to the Year field
- Day of Week: From 6 to 0 with no carry (see Table 17-1)
- Year Carry: From 99 to 00; this also surpasses the use of the RTCC

For the day to month rollover schedule, see Table 17-2.

Considering that the following values are in BCD format, the carry to the upper BCD digit will occur at a count of 10 and not at 16 (SECONDS, MINUTES, HOURS, WEEKDAY, DAYS and MONTHS).

TABLE 17-1: DAY OF WEEK SCHEDULE

Day of Week						
Sunday	0					
Monday	1					
Tuesday	2					
Wednesday	3					
Thursday	4					
Friday	5					
Saturday	6					





18.5.7.1 Steering Synchronization

The STRSYNC bit of the PSTRxCON register gives the user two selections of when the steering event will happen. When the STRSYNC bit is '0', the steering event will happen at the end of the instruction that writes to the PSTRxCON register. In this case, the output signal at the Px<D:A> pins may be an incomplete PWM waveform. This operation is useful when the user firmware needs to immediately remove a PWM signal from the pin.

When the STRSYNC bit is '1', the effective steering update will happen at the beginning of the next PWM period. In this case, steering on/off the PWM output will always produce a complete PWM waveform.

Figures 18-19 and 18-20 illustrate the timing diagrams of the PWM steering depending on the STRSYNC setting.

FIGURE 18-19: EXAMPLE OF STEERING EVENT AT END OF INSTRUCTION (STRSYNC = 0)



FIGURE 18-20: EXAMPLE OF STEERING EVENT AT BEGINNING OF INSTRUCTION (STRSYNC = 1)



REGISTER 19-2:	SSPxCON1: MSSPx CONTROL	REGISTER 1 – SPI MODE	(ACCESS FC6H/F72h)
----------------	-------------------------	-----------------------	--------------------

R/W-0	R/C-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	
WCOL	SSPOV ⁽¹⁾	SSPEN ⁽²⁾	CKP	SSPM3 ⁽³⁾	SSPM2 ⁽³⁾	SSPM1 ⁽³⁾	SSPM0 ⁽³⁾	
bit 7							bit 0	
Legend:								
R = Read	able bit	W = Writable	oit	U = Unimplemented bit, read as '0'				
-n = Value	e at POR	'1' = Bit is set		'0' = Bit is cleared		x = Bit is unknown		
bit 7 WCOL: Write Collision Detect bit 1 = The SSPxBUF register is written while it is still transmitting the previous word (must be cleared i software) 0 = No collision						t be cleared in		
bit 6	bit 6 SSPOV: Receive Overflow Indicator bit ⁽¹⁾ <u>SPI Slave mode:</u> 1 = A new byte is received while the SSPxBUF register is still holding the previous data. In case of over flow, the data in SSPxSR is lost. Overflow can only occur in Slave mode. The user must read the SSPxBUF, even if only transmitting data, to avoid setting overflow (must be cleared in software).						n case of over- must read the in software).	
bit 5	SSPEN: Master Synchronous Serial Port Enable bit ⁽²⁾ 1 = Enables serial port and configures SCKx, SDOx, SDIx and \overline{SSx} as serial port pins 0 = Disables serial port and configures these pins as I/O port pins							
bit 4	CKP: Clock Polarity Select bit 1 = Idle state for clock is a high level 0 = Idle state for clock is a low level							
bit 3-0	t 3-0 SSPM<3:0>: Master Synchronous Serial Port Mode Select bits ⁽³⁾ 0101 = SPI Slave mode, clock = SCKx pin, <u>SSx</u> pin control disabled, <u>SSx</u> can be used as I/O pin 0100 = SPI Slave mode, clock = SCKx pin, <u>SSx</u> pin control enabled 0011 = SPI Master mode, clock = TMR2 output/2 0010 = SPI Master mode, clock = Fosc/64 0001 = SPI Master mode, clock = Fosc/16 0000 = SPI Master mode, clock = Fosc/4						as I/O pin	
Note 1:	In Master mode, t writing to the SSF	the overflow bit PxBUF register.	is not set sind	ce each new rec	eption (and tra	insmission) is ir	nitiated by	

- 2: When enabled, this pin must be properly configured as input or output.
- **3:** Bit combinations not specifically listed here, are either reserved or implemented in I^2C^{TM} mode only.

24.1 Operation

When the HLVD module is enabled, a comparator uses an internally generated reference voltage as the set point. The set point is compared with the trip point, where each node in the resistor divider represents a trip point voltage. The "trip point" voltage is the voltage level at which the device detects a high or low-voltage event, depending on the configuration of the module.

When the supply voltage is equal to the trip point, the voltage tapped off of the resistor array is equal to the internal reference voltage generated by the voltage reference module. The comparator then generates an interrupt signal by setting the LVDIF bit.

The trip point voltage is software programmable to any one of 8 values. The trip point is selected by programming the HLVDL<3:0> bits (HLVDCON<3:0>).

Additionally, the HLVD module allows the user to supply the trip voltage to the module from an external source. This mode is enabled when bits, HLVDL<3:0>, are set to '1111'. In this state, the comparator input is multiplexed from the external input pin, HLVDIN. This gives users flexibility because it allows them to configure the HLVD interrupt to occur at any voltage in the valid operating range.

FIGURE 24-1: HLVD MODULE BLOCK DIAGRAM (WITH EXTERNAL INPUT)



25.4 Measuring Capacitance with the CTMU

There are two separate methods of measuring capacitance with the CTMU. The first is the absolute method, in which the actual capacitance value is desired. The second is the relative method, in which the actual capacitance is not needed, rather an indication of a change in capacitance is required.

25.4.1 ABSOLUTE CAPACITANCE MEASUREMENT

For absolute capacitance measurements, both the current and capacitance calibration steps found in **Section 25.3 "Calibrating the CTMU Module"** should be followed. Capacitance measurements are then performed using the following steps:

- 1. Initialize the A/D Converter.
- 2. Initialize the CTMU.
- 3. Set EDG1STAT.
- 4. Wait for a fixed delay, *T*.
- 5. Clear EDG1STAT.
- 6. Perform an A/D conversion.
- 7. Calculate the total capacitance, CTOTAL = (I * T)/V, where *I* is known from the current source measurement step (see **Section 25.3.1 "Current Source Calibration"**), *T* is a fixed delay and *V* is measured by performing an A/D conversion.
- 8. Subtract the stray and A/D capacitance (*C*OFFSET from **Section 25.3.2** "**Capacitance Calibration**") from *CTOTAL* to determine the measured capacitance.

25.4.2 RELATIVE CHARGE MEASUREMENT

An application may not require precise capacitance measurements. For example, when detecting a valid press of a capacitance-based switch, detecting a relative change of capacitance is of interest. In this type of application, when the switch is open (or not touched), the total capacitance is the capacitance of the combination of the board traces, the A/D Converter, etc. A larger voltage will be measured by the A/D Converter. When the switch is closed (or is touched), the total capacitance is larger due to the addition of the capacitances, and a smaller voltage will be measured by the A/D Converter.

Detecting capacitance changes is easily accomplished with the CTMU using these steps:

- 1. Initialize the A/D Converter and the CTMU.
- 2. Set EDG1STAT.
- 3. Wait for a fixed delay.
- 4. Clear EDG1STAT.
- 5. Perform an A/D conversion.

The voltage measured by performing the A/D conversion is an indication of the relative capacitance. Note that in this case, no calibration of the current source or circuit capacitance measurement is needed. See Example 25-4 for a sample software routine for a capacitive touch switch.

26.5.2 EXITING FAIL-SAFE OPERATION

The Fail-Safe Clock Monitor condition is terminated by either a device Reset or by entering a power-managed mode. On Reset, the controller starts the primary clock source specified in Configuration Register 2H (with any required start-up delays that are required for the oscillator mode, such as OST or PLL timer). The INTRC oscillator provides the device clock until the primary clock source becomes ready (similar to a Two-Speed Start-up). The clock source is then switched to the primary clock (indicated by the OSTS bit in the OSCCON register becoming set). The FSCM then resumes monitoring the peripheral clock.

The primary clock source may never become ready during start-up. In this case, operation is clocked by the INTRC oscillator. The OSCCON register will remain in its Reset state until a power-managed mode is entered.

26.5.3 FSCM INTERRUPTS IN POWER-MANAGED MODES

By entering a power-managed mode, the clock multiplexer selects the clock source selected by the OSCCON register. FSCM of the power-managed clock source resumes in the power-managed mode.

If an oscillator failure occurs during power-managed operation, the subsequent events depend on whether or not the oscillator failure interrupt is enabled. If enabled (OSCFIF = 1), code execution will be clocked by the INTRC multiplexer. An automatic transition back to the failed clock source will not occur.

If the interrupt is disabled, subsequent interrupts while in Idle mode will cause the CPU to begin executing instructions while being clocked by the INTRC source.

26.5.4 POR OR WAKE-UP FROM SLEEP

The FSCM is designed to detect oscillator failure at any point after the device has exited Power-on Reset (POR) or low-power Sleep mode. When the primary device clock is either the EC or INTRC modes, monitoring can begin immediately following these events.

For HS or HSPLL modes, the situation is somewhat different. Since the oscillator may require a start-up time considerably longer than the FSCM sample clock time, a false clock failure may be detected. To prevent this, the internal oscillator block is automatically configured as the device clock and functions until the primary clock is stable (the OST and PLL timers have timed out). This is identical to Two-Speed Start-up mode. Once the primary clock is stable, the INTRC returns to its role as the FSCM source.

Note:	The same logic that prevents false						
	oscillator failure interrupts on POR, or						
	wake-up from Sleep, will also prevent the						
	detection of the oscillator's failure to start						
	at all following these events. This can be						
	avoided by monitoring the OSTS bit and						
	using a timing routine to determine if the						
	oscillator is taking too long to start. Even						
	so, no oscillator failure interrupt will be						
	flagged.						

As noted in Section 26.4.1 "Special Considerations for Using Two-Speed Start-up", it is also possible to select another clock configuration and enter an alternate power-managed mode while waiting for the primary clock to become stable. When the new power-managed mode is selected, the primary clock is disabled.

26.6 Program Verification and Code Protection

For all devices in the PIC18F46J11 family of devices, the on-chip program memory space is treated as a single block. Code protection for this block is controlled by one Configuration bit, CP0. This bit inhibits external reads and writes to the program memory space. It has no direct effect in normal execution mode.

26.6.1 CONFIGURATION REGISTER PROTECTION

The Configuration registers are protected against untoward changes or reads in two ways. The primary protection is the write-once feature of the Configuration bits, which prevents reconfiguration once the bit has been programmed during a power cycle. To safeguard against unpredictable events, Configuration bit changes resulting from individual cell level disruptions (such as ESD events) will cause a parity error and trigger a device Reset. This is seen by the user as a Configuration Mismatch (CM) Reset.

The data for the Configuration registers is derived from the FCW in program memory. When the CP0 bit is set, the source data for device configuration is also protected as a consequence.

BCF	Bit Clear f				BN			
Syntax:	BCF f, b	BCF f, b {,a}						
Operands:		Operands						
	0 ≤ b ≤ 7 a ∈ [0,1]	$\begin{array}{l} 0 \leq b \leq 7 \\ a \in [0,1] \end{array}$						
Operation:	Status Aff							
Status Affected:	None				Encodina			
Encoding:	1001	bbba	ffff	ffff	Descriptio			
Description:	Bit 'b' in reg	gister 'f' i	s cleared.	•	2 2 0 0 0 m m			
	lf 'a' is '0', t lf 'a' is '1', t GPR bank	he Acces he BSR i (default).	ss Bank is is used to	selected. select the				
	If 'a' is '0' a set is enabl in Indexed mode wher Section 27 Bit-Oriente Literal Offs	If 'a' is '0' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever f ≤ 95 (5Fh). See Section 27.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode" for details.						
Words:	1				If Jump:			
Cycles:	1				D			
Q Cycle Activity:								
Q1	Q2	Q3	3	Q4				
Decode	Read register 'f'	Proce Data	ess a reg	Write gister 'f'	op If No Jun			
Example:	BCF F	LAG_RE	G, 7,	0	D			
Before Instruc FLAG_F	ction REG = C7h							
After Instructi	on -47				Example:			
FLAG_F	(EG = 4/N				Befo			
					After			

BN		Branch if N	Branch if Negative						
Syntax:		BN n	BN n						
Oper	ands:	-128 ≤ n ≤ 1	127						
Oper	ation:	if Negative (PC) + 2 + 2	bit is '1', 2n \rightarrow PC						
Statu	s Affected:	None							
Enco	ding:	1110	0110 nnr	ın nnnn					
Desc	ription:	If the Negat program wi	tive bit is '1', th Il branch.	nen the					
		The 2's con added to th have incren instruction, PC + 2 + 2r two-cycle in	The 2's complement number '2n' is added to the PC. Since the PC will have incremented to fetch the next instruction, the new address will be PC + 2 + 2n. This instruction is then a two-cycle instruction.						
Word	ls:	1	1						
Cycle	es:	1(2)	1(2)						
Q Cycle Activity: If Jump:				.					
	Q1	Q2	Q3	Q4					
	Decode	Read literal 'n'	Process Data	Write to PC					
	No	No	No	No					
	operation	operation	operation	operation					
lf No	o Jump:								
	Q1	Q2	Q3	Q4					
	Decode	Read literal	Process	No					
		n	Data	operation					
Example:		HERE	BN Jump						
Before Instructio PC After Instruction If Negative PC If Negative PC		tion = ad on = 1; /e = 1; /e = 0; /e = 0;	dress (HERE) dress (Jump) dress (HERE	+ 2)					

29.2 DC Characteristics: Power-Down and Supply Current PIC18F46J11 Family (Industrial) (Continued)

PIC18LFXXJ11 Family		Standard Operating Conditions (unless otherwise stated)Operating temperature $-40^{\circ}C \le TA \le +85^{\circ}C$ for industrial					
PIC18FXXJ11 Family		Standard Operating Conditions (unless otherwise stated)Operating temperature $-40^{\circ}C \le TA \le +85^{\circ}C$ for industrial					
Param No.	Device	Тур	Max	Units	Conditions		
	Supply Current (IDD) ⁽²⁾						
	PIC18LFXXJ11	1.9	3.6	mA	-40°C		
		2.0	3.8	mA	+25°C	VDD = 2.0V, VDDCORF = 2.0V	
		2.0	3.8	mA	+85°C	VBBOOKE 2.0V	
	PIC18LFXXJ11	2.8	4.8	mA	-40°C		
		2.8	4.8	mA	+25°C	VDD = 2.5V, VDDCORF = 2.5V	
		2.8	4.9	mA	+85°C		Fosc = 8 MHz, RC_RUN
	PIC18FXXJ11	2.3	4.2	mA	-40°C	VDD = 2.15V, VDDCORE = 10 μF	mode, Internal RC Oscillator
		2.3	4.2	mA	+25°C		
		2.4	4.5	mA	+85°C	Capacitor	
	PIC18FXXJ11	2.8	5.1	mA	-40°C	VDD = 3.3V,	
		2.8	5.1	mA	+25°C	VDDCORE = 10 μF	
		2.8	5.4	mA	+85°C	Capacitor	
	PIC18LFXXJ11	1.9	9.4	μA	-40°C	V/00 = 2 0V/	
		2.3	9.4	μA	+25°C	VDD = 2.0V, VDDCORE = 2.0V	
		4.5	17.2	μA	+85°C		
	PIC18LFXXJ11	2.4	10.5	μA	-40°C	Vpp = 2.5V	
		2.8	10.5	μA	+25°C	- VDDCORE = 2.5V	
		5.4	19.5	μA	+85°C		mode, Internal RC Oscillator,
	PIC18FXXJ11	33.3	75	μA	-40°C	VDD = 2.15V, VDDCORE = $10 \mu F$	INTSRC = 0
		43.8	75	μA	+25°C		
		55.3	92	μA	+85°C	Capacitor	
	PIC18FXXJ11	36.1	82	μA	-40°C	VDD = 3.3V,	
		44.5	82	μA	+25°C	VDDCORE = $10 \mu\text{F}$	
		56.3	105	μA	+85°C	Capacitor	

Note 1: The power-down current in Sleep mode does not depend on the oscillator type. Power-down current is measured with the part in Sleep mode, with all I/O pins in high-impedance state and tied to VDD or Vss and all features that add delta current disabled (such as WDT, Timer1 oscillator, BOR, etc.).

2: The supply current is mainly a function of operating voltage, frequency and mode. Other factors, such as I/O pin loading and switching rate, oscillator type and circuit, internal code execution pattern and temperature, also have an impact on the current consumption. All features that add delta current are disabled (WDT, etc.). The test conditions for all IDD measurements in active operation mode are:

OSC1 = external square wave, from rail-to-rail; all I/O pins tri-stated, pulled to VDD/Vss;

MCLR = VDD; WDT disabled unless otherwise specified.

3: Low-Power Timer1 with standard, low-cost 32 kHz crystals have an operating temperature range of -10°C to +70°C. Extended temperature crystals are available at a much higher cost.

PRODUCT IDENTIFICATION SYSTEM

To order or obtain information, e.g., on pricing or delivery, refer to the factory or the listed sales office.

PART NO.	<u>× /xx xxx</u>	Examples:
Device	Temperature Package Pattern Range	 a) PIC18F46J11-I/PT 301 = Industrial temp., TQFP package, QTP pattern #301. b) PIC18F46J11T-I/PT = Tape and reel, Industrial temp., TQFP package.
Device ⁽¹⁾	PIC18F24J11 PIC18F25J11 PIC18F45J11 PIC18F45J11 PIC18F46J11 PIC18LF24J11 PIC18LF25J11 PIC18LF26J11 PIC18LF4J11 PIC18LF45J11 PIC18LF46J11	
Temperature Range	I = -40° C to $+85^{\circ}$ C (Industrial)	Note 1: F = Standard Voltage Range LF = Extended Voltage Range 2: T = In tape and reel
Package	SP = Skinny PDIP SS = SSOP SO = SOIC ML = QFN PT = TQFP (Thin Quad Flatpack)	
Pattern	QTP, SQTP, Code or Special Requirements (blank otherwise)	