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Applications of "<u>Embedded - Microcontrollers</u>"

Details	
Product Status	Active
Core Processor	PIC
Core Size	8-Bit
Speed	48MHz
Connectivity	I ² C, SPI, UART/USART
Peripherals	Brown-out Detect/Reset, POR, PWM, WDT
Number of I/O	34
Program Memory Size	64KB (32K x 16)
Program Memory Type	FLASH
EEPROM Size	-
RAM Size	3.8K x 8
Voltage - Supply (Vcc/Vdd)	2V ~ 3.6V
Data Converters	A/D 13x10b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	44-VQFN Exposed Pad
Supplier Device Package	44-QFN (8x8)
Purchase URL	https://www.e-xfl.com/product-detail/microchip-technology/pic18lf46j11-i-ml

Pin Diagrams

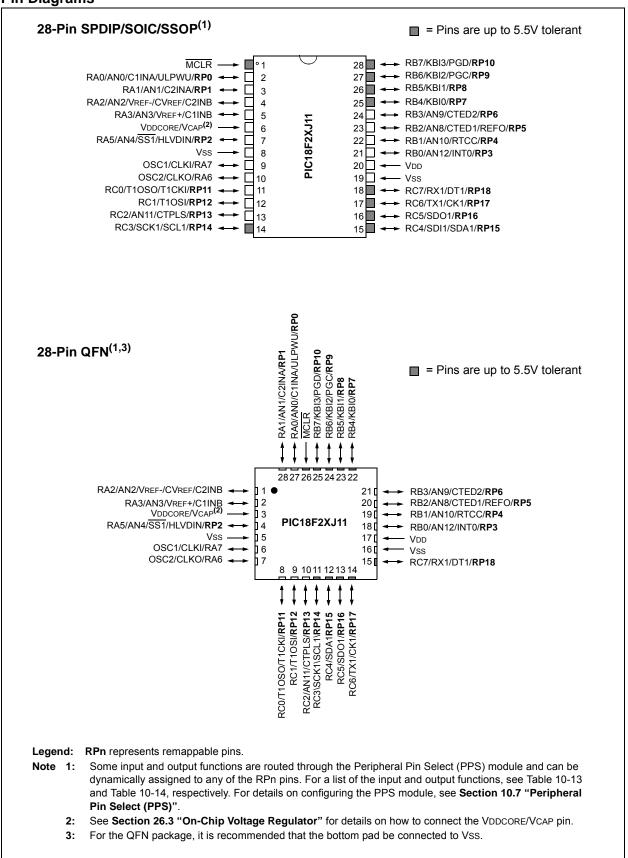


TABLE 1-3: PIC18F2XJ11 PINOUT I/O DESCRIPTIONS (CONTINUED)

	Pin Number				
Pin Name	28-SPDIP/ SSOP/ SOIC	28-QFN	Pin Type	Buffer Type	Description
					PORTA is a bidirectional I/O port.
RA0/AN0/C1INA/ULPWU/RP0 RA0 AN0 C1INA ULPWU RP0	2	27	I/O I I I I/O	DIG Analog Analog Analog DIG	Comparator 1 input A.
RA1/AN1/C2INA/RP1 RA1 AN1 C2INA RP1	3	28	I/O O I I/O	DIG Analog Analog DIG	,
RA2/AN2/VREF-/CVREF/C2INB RA2 AN2 VREF- CVREF C2INB	4	1	I/O I O I	DIG Analog Analog Analog Analog	A/D reference voltage (low) input. Comparator reference voltage output.
RA3/AN3/VREF+/C1INB RA3 AN3 VREF+ C1INB	5	2	I/O I I I	DIG Analog Analog Analog	A/D reference voltage (high) input
RA5/AN4/SS1/HLVDIN/ RP2 RA5 AN4 SS1 HLVDIN RP2	7	4	I/O I I I I/O	DIG Analog TTL Analog DIG	SPI slave select input. High/low-voltage detect input. Remappable peripheral pin 2.
RA6 ⁽¹⁾ RA7 ⁽¹⁾					See the OSC2/CLKO/RA6 pin. See the OSC1/CLKI/RA7 pin.

Legend: TTL = TTL compatible input

= Input

= Power DIG = Digital output

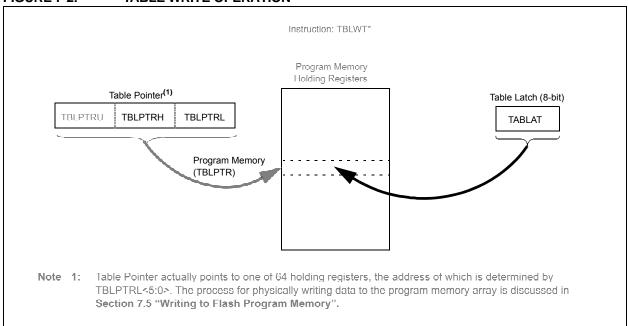
CMOS = CMOS compatible input or output ST = Schmitt Trigger input with CMOS levels Analog = Analog input

0 = Output

OD = Open-Drain (no P diode to VDD)

Note 1: RA7 and RA6 will be disabled if OSC1 and OSC2 are used for the clock function.

FIGURE 7-2: TABLE WRITE OPERATION



7.2 Control Registers

Several control registers are used in conjunction with the TBLRD and TBLWT instructions. Those are:

- EECON1 register
- · EECON2 register
- · TABLAT register
- · TBLPTR registers

7.2.1 EECON1 AND EECON2 REGISTERS

The EECON1 register (Register 7-1) is the control register for memory accesses. The EECON2 register is not a physical register; it is used exclusively in the memory write and erase sequences. Reading EECON2 will read all '0's.

The WPROG bit, when set, will allow programming two bytes per word on the execution of the WR command. If this bit is cleared, the WR command will result in programming on a block of 64 bytes.

The FREE bit, when set, will allow a program memory erase operation. When FREE is set, the erase operation is initiated on the next WR command. When FREE is clear, only writes are enabled.

The WREN bit, when set, will allow a write operation. On power-up, the WREN bit is clear. The WRERR bit is set in hardware when the WR bit is set, and cleared when the internal programming timer expires and the write operation is complete.

Note: During normal operation, the WRERR is read as '1'. This can indicate that a write operation was prematurely terminated by a Reset, or a write operation was attempted improperly.

The WR control bit initiates write operations. The bit cannot be cleared, only set, in software. It is cleared in hardware at the completion of the write operation.

10.6 PORTE, TRISE and LATE Registers

Note: PORTE is available only in 44-pin devices.

Depending on the particular PIC18F46J11 family device selected, PORTE is implemented in two different ways.

For 44-pin devices, PORTE is a 3-bit wide port. Three pins (RE0/AN5/PMRD, RE1/AN6/PMWR and RE2/AN7/PMCS) are individually configurable as inputs or outputs. These pins have Schmitt Trigger input buffers. When selected as analog inputs, these pins will read as '0's.

The corresponding Data Direction register is TRISE. Setting a TRISE bit (= 1) will make the corresponding PORTE pin an input (i.e., put the corresponding output driver in a high-impedance mode). Clearing a TRISE bit (= 0) will make the corresponding PORTE pin an output (i.e., put the contents of the output latch on the selected pin).

TRISE controls the direction of the RE pins, even when they are being used as analog inputs. The user must make sure to keep the pins configured as inputs when using them as analog inputs.

Note: On a POR, RE<2:0> are configured as analog inputs.

The Data Latch register (LATE) is also memory mapped. Read-modify-write operations on the LATE register read and write the latched output value for PORTE.

EXAMPLE 10-6: INITIALIZING PORTE

CLRF	LATE	;	Initialize LATE
		;	to clear output
		;	data latches
MOVLW	0xE0	;	Configure REx
MOVWF	ANCON0	;	for digital inputs
MOVLW	0x03	;	Value used to
		;	initialize data
		;	direction
MOVWF	TRISE	;	Set RE<0> as inputs
		;	RE<1> as outputs
		;	RE<2> as inputs

Each of the PORTE pins has a weak internal pull-up. A single control bit can turn on all the pull-ups. This is performed by setting bit, REPU (PORTE<6>). The weak pull-up is automatically turned off when the port pin is configured as an output. The pull-ups are disabled on a POR.

Note that the pull-ups can be used for any set of features, similar to the pull-ups found on PORTB.

11.2.4 BUFFERED PARALLEL SLAVE PORT MODE

Buffered Parallel Slave Port mode is functionally identical to the legacy PSP mode with one exception, the implementation of 4-level read and write buffers. Buffered PSP mode is enabled by setting the INCM bits in the PMMODEH register. If the INCM<1:0> bits are set to '11', the PMP module will act as the Buffered PSP.

When the Buffered mode is active, the PMDIN1L, PMDIN1H, PMDIN2L and PMDIN2H registers become the write buffers and the PMDOUT1L, PMDOUT1H, PMDOUT2L and PMDOUT2H registers become the read buffers. Buffers are numbered 0 through 3, starting with the lower byte of PMDIN1L to PMDIN2H as the read buffers and PMDOUT1L to PMDOUT2H as the write buffers.

11.2.4.1 READ FROM SLAVE PORT

For read operations, the bytes will be sent out sequentially, starting with Buffer 0 (PMDOUT1L<7:0>) and ending with Buffer 3 (PMDOUT2H<7:0>) for every read strobe. The module maintains an internal pointer to keep track of which buffer is to be read. Each buffer has a corresponding read status bit, OBxE, in the PMSTATL register. This bit is cleared when a buffer contains data that has not been written to the bus, and is set when data is written to the bus. If the current buffer location being read from is empty, a buffer underflow is generated, and the Buffer Overflow flag bit, OBUF, is set. If all four OBxE status bits are set, then the Output Buffer Empty flag (OBE) will also be set.

11.2.4.2 WRITE TO SLAVE PORT

For write operations, the data has to be stored sequentially, starting with Buffer 0 (PMDIN1L<7:0>) and ending with Buffer 3 (PMDIN2H<7:0>). As with read operations, the module maintains an internal pointer to the buffer that is to be written next.

The input buffers have their own write status bits, IBxF in the PMSTATH register. The bit is set when the buffer contains unread incoming data, and cleared when the data has been read. The flag bit is set on the write strobe. If a write occurs on a buffer when its associated IBxF bit is set, the Buffer Overflow flag, IBOV, is set; any incoming data in the buffer will be lost. If all four IBxF flags are set, the Input Buffer Full Flag (IBF) is set.

In Buffered Slave mode, the module can be configured to generate an interrupt on every read or write strobe (IRQM<1:0> = 01). It can be configured to generate an interrupt on a read from Read Buffer 3 or a write to Write Buffer 3, which is essentially an interrupt every fourth read or write strobe (RQM<1:0> = 11). When interrupting every fourth byte for input data, all input buffer registers should be read to clear the IBxF flags. If these flags are not cleared, then there is a risk of hitting an overflow condition.



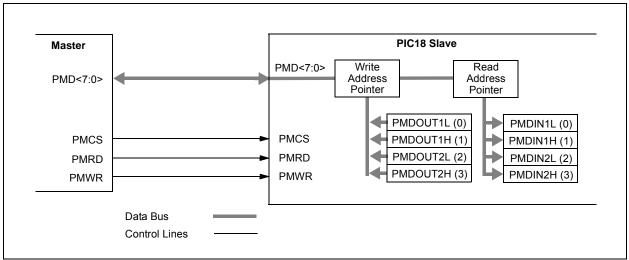


FIGURE 11-15: WRITE TIMING, 8-BIT DATA, PARTIALLY MULTIPLEXED ADDRESS

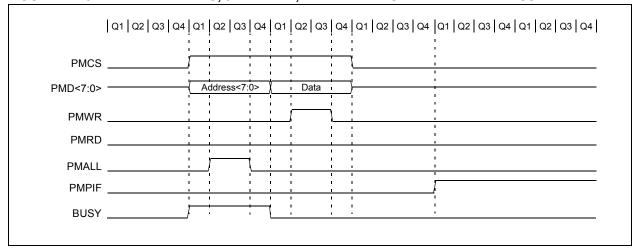


FIGURE 11-16: WRITE TIMING, 8-BIT DATA, WAIT STATES ENABLED, PARTIALLY MULTIPLEXED ADDRESS

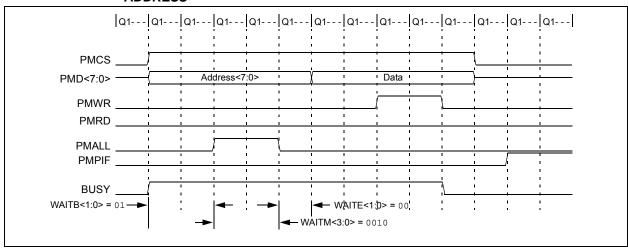
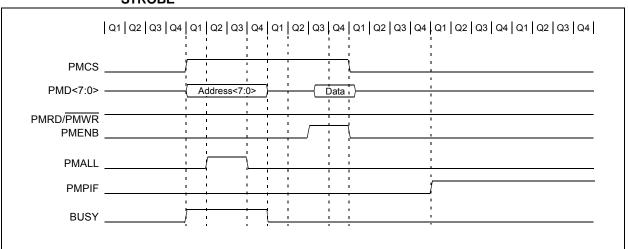


FIGURE 11-17: READ TIMING, 8-BIT DATA, PARTIALLY MULTIPLEXED ADDRESS, ENABLE STROBE



12.3 Prescaler

An 8-bit counter is available as a prescaler for the Timer0 module. The prescaler is not directly readable or writable. Its value is set by the PSA and T0PS<2:0> bits (T0CON<3:0>), which determine the prescaler assignment and prescale ratio.

Clearing the PSA bit assigns the prescaler to the Timer0 module. When it is assigned, prescale values from 1:2 through 1:256 in power-of-2 increments are selectable.

When assigned to the Timer0 module, all instructions writing to the TMR0 register (e.g., CLRF TMR0, MOVWF TMR0, BSF TMR0, etc.) clear the prescaler count.

Note: Writing to TMR0 when the prescaler is assigned to Timer0 will clear the prescaler count but will not change the prescaler assignment.

12.3.1 SWITCHING PRESCALER ASSIGNMENT

The prescaler assignment is fully under software control and can be changed "on-the-fly" during program execution.

12.4 Timer0 Interrupt

The TMR0 interrupt is generated when the TMR0 register overflows from FFh to 00h in 8-bit mode, or from FFFFh to 0000h in 16-bit mode. This overflow sets the TMR0IF flag bit. The interrupt can be masked by clearing the TMR0IE bit (INTCON<5>). Before re-enabling the interrupt, the TMR0IF bit must be cleared in software by the Interrupt Service Routine (ISR)

Since Timer0 is shut down in Sleep mode, the TMR0 interrupt cannot awaken the processor from Sleep.

TABLE 12-1: REGISTERS ASSOCIATED WITH TIMERO

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on Page:		
TMR0L	Timer0 Register Low Byte										
TMR0H	Timer0 Register High Byte										
INTCON	GIE/GIEH PEIE/GIEL TMR0IE INT0IE RBIE TMR0IF INT0IF RBIF							90			
T0CON	TMR00N	T08BIT	TOCS	T0SE	PSA	T0PS2	T0PS1	T0PS0	91		

Legend: — = unimplemented, read as '0'. Shaded cells are not used by Timer0.

13.2 **Timer1 Operation**

The Timer1 module is an 8-bit or 16-bit incrementing counter, which is accessed through TMR1H:TMR1L register pair.

When used with an internal clock source, the module is a timer and increments on every instruction cycle. When used with an external clock source, the module can be used as either a timer or counter and increments on every selected edge of the external source.

Timer1 is enabled by configuring the TMR1ON and TMR1GE bits in the T1CON and T1GCON registers, respectively.

When Timer1 is enabled, the RC1/T1OSI/RP12 and RC0/T1OSO/T1CKI/RP11 pins become inputs. This means the values of TRISC<1:0> are ignored and the pins are read as '0'.

13.3 **Clock Source Selection**

The TMR1CS<1:0> and T1OSCEN bits of the T1CON register are used to select the clock source for Timer1. Register 13-1 displays the clock source selections.

When switching clock sources and using the clock prescaler, write to TMR1L afterwards to reset the internal prescaler count to 0.

13.3.1 INTERNAL CLOCK SOURCE

When the internal clock source is selected, the TMR1H:TMR1L register pair will increment on multiples of Fosc as determined by the Timer1 prescaler.

13.3.2 **EXTERNAL CLOCK SOURCE**

When the external clock source is selected, the Timer1 module may work as a timer or a counter.

When enabled to count, Timer1 is incremented on the rising edge of the external clock input, T1CKI, or the capacitive sensing oscillator signal. Either of these external clock sources can be synchronized to the microcontroller system clock or they can run asynchronously.

When used as a timer with a clock oscillator, an external 32.768 kHz crystal can be used in conjunction with the dedicated internal oscillator circuit.

Note:

In Counter mode, a falling edge must be registered by the counter prior to the first incrementing rising edge after any one or more of the following conditions:

- Timer1 enabled after POR Reset
- Write to TMR1H or TMR1L
- · Timer1 is disabled
- Timer1 is disabled (TMR1ON = 0)

when T1CKI is high, then Timer1 is enabled (TMR1ON = 1) when T1CKI is low.

TABLE 13-1: TIMER1 CLOCK SOURCE SELECTION

TMR1CS1	TMR1CS0	T10SCEN	Clock Source
0	1	х	Clock Source (Fosc)
0	0	х	Instruction Clock (Fosc/4)
1	0	0	External Clock on T1CKI Pin
1	0	1	Oscillator Circuit on T1OSI/T1OSO Pin

13.4 Timer1 16-Bit Read/Write Mode

Timer1 can be configured for 16-bit reads and writes. When the RD16 control bit (T1CON<1>) is set, the address for TMR1H is mapped to a buffer register for the high byte of Timer1. A read from TMR1L loads the contents of the high byte of Timer1 into the Timer1 High Byte Buffer register. This provides the user with the ability to accurately read all 16 bits of Timer1 without having to determine whether a read of the high byte, followed by a read of the low byte, has become invalid due to a rollover between reads.

A write to the high byte of Timer1 must also take place through the TMR1H Buffer register. The Timer1 high byte is updated with the contents of TMR1H when a write occurs to TMR1L. This allows a user to write all 16 bits to both the high and low bytes of Timer1 at once.

The high byte of Timer1 is not directly readable or writable in this mode. All reads and writes must take place through the Timer1 High Byte Buffer register. Writes to TMR1H do not clear the Timer1 prescaler. The prescaler is only cleared on writes to TMR1L.

13.5 Timer1 Oscillator

An on-chip crystal oscillator circuit is incorporated between pins, T1OSI (input) and T1OSO (amplifier output). It is enabled by setting the Timer1 Oscillator Enable bit, T1OSCEN (T1CON<3>). The oscillator is a low-power circuit rated for 32 kHz crystals. It will continue to run during all power-managed modes. The circuit for a typical LP oscillator is depicted in Figure 13-2. Table 13-2 provides the capacitor selection for the Timer1 oscillator.

The user must provide a software time delay to ensure proper start-up of the Timer1 oscillator.

FIGURE 13-2: EXTERNAL COMPONENTS
FOR THE TIMER1 LP
OSCILLATOR

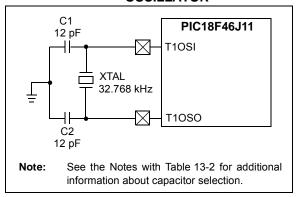


TABLE 13-2: CAPACITOR SELECTION FOR THE TIMER OSCILLATOR^(2,3,4,5)

Oscillator Type	Freq.	C1	C2
LP	32 kHz	12 pF ⁽¹⁾	12 pF ⁽¹⁾

- **Note 1:** Microchip suggests these values as a starting point in validating the oscillator circuit.
 - **2:** Higher capacitance increases the stability of the oscillator but also increases the start-up time.
 - **3:** Since each resonator/crystal has its own characteristics, the user should consult the resonator/crystal manufacturer for appropriate values of external components.
 - **4:** Capacitor values are for design guidance only. Values listed would be typical of a CL = 10 pF rated crystal, when LPT1OSC = 1.
 - 5: Incorrect capacitance value may result in a frequency not meeting the crystal manufacturer's tolerance specification.

The Timer1 crystal oscillator drive level is determined based on the LPT1OSC (CONFIG2L<4>) Configuration bit. The higher drive level mode, LPT1OSC = 1, is intended to drive a wide variety of 32.768 kHz crystals with a variety of load capacitance (CL) ratings.

The lower drive level mode is highly optimized for extremely low-power consumption. It is not intended to drive all types of 32.768 kHz crystals. In the low drive level mode, the crystal oscillator circuit may not work if excessively large discrete capacitors are placed on the T1OSI and T1OSO pins. This mode is only designed to work with discrete capacitances of approximately 3 pF-10 pF on each pin.

Crystal manufacturers usually specify a CL (load capacitance) rating for their crystals. This value is related to, but not necessarily the same as, the values that should be used for C1 and C2 in Figure 13-2. See the crystal manufacturer's applications' information for more details on how to select the optimum C1 and C2 for a given crystal. The optimum value depends in part on the amount of parasitic capacitance in the circuit, which is often unknown. Therefore, after values have been selected, it is highly recommended that thorough testing and validation of the oscillator be performed.

REGISTER 17-11: HOURS: HOURS VALUE REGISTER (ACCESS F98h, PTR 01b)(1)

U-0	U-0	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x
_	— — HRTEN1 H		HRTEN0	HRONE3	HRONE2	HRONE1	HRONE0
bit 7							bit 0

Legend:

R = Readable bit W = Writable bit U = Unimplemented bit, read as '0'

-n = Value at POR '1' = Bit is set '0' = Bit is cleared x = Bit is unknown

bit 7-6 Unimplemented: Read as '0'

bit 5-4 HRTEN<1:0>: Binary Coded Decimal Value of Hour's Tens Digit bits

Contains a value from 0 to 2.

bit 3-0 HRONE<3:0>: Binary Coded Decimal Value of Hour's Ones Digit bits

Contains a value from 0 to 9.

Note 1: A write to this register is only allowed when RTCWREN = 1.

REGISTER 17-12: MINUTES: MINUTES VALUE REGISTER (ACCESS F99h, PTR 00b)

	U-0	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x	
	— MINTEN2 M		MINTEN1	MINTEN0	MINONE3	MINONE2	MINONE1	MINONE0	
Ī	bit 7							bit 0	

Legend:

R = Readable bit W = Writable bit U = Unimplemented bit, read as '0'

-n = Value at POR '1' = Bit is set '0' = Bit is cleared x = Bit is unknown

bit 7 **Unimplemented:** Read as '0'

bit 6-4 MINTEN<2:0>: Binary Coded Decimal Value of Minute's Tens Digit bits

Contains a value from 0 to 5.

bit 3-0 MINONE<3:0>: Binary Coded Decimal Value of Minute's Ones Digit bits

Contains a value from 0 to 9.

REGISTER 17-13: SECONDS: SECONDS VALUE REGISTER (ACCESS F98h, PTR 00b)

U-0	R/W-x						
_	SECTEN2	SECTEN1	SECTEN0	SECONE3	SECONE2	SECONE1	SECONE0
bit 7							bit 0

Legend:

R = Readable bit W = Writable bit U = Unimplemented bit, read as '0'

-n = Value at POR '1' = Bit is set '0' = Bit is cleared x = Bit is unknown

bit 7 Unimplemented: Read as '0'

bit 6-4 SECTEN<2:0>: Binary Coded Decimal Value of Second's Tens Digit bits

Contains a value from 0 to 5.

bit 3-0 SECONE<3:0>: Binary Coded Decimal Value of Second's Ones Digit bits

Contains a value from 0 to 9.

19.2 Control Registers

Each MSSP module has three associated control registers. These include a status register (SSPxSTAT) and two control registers (SSPxCON1 and SSPxCON2). The use of these registers and their individual Configuration bits differ significantly depending on whether the MSSP module is operated in SPI or I²C mode.

Additional details are provided under the individual sections.

Note:

In devices with more than one MSSP module, it is very important to pay close attention to the SSPxCON register names. SSP1CON1 and SSP1CON2 control different operational aspects of the same module, while SSP1CON1 and SSP2CON1 control the same features for two different modules.

19.3 SPI Mode

The SPI mode allows 8 bits of data to be synchronously transmitted and received simultaneously. All four modes of SPI are supported.

When MSSP2 is used in SPI mode, it can optionally be configured to work with the SPI DMA submodule described in **Section 19.4** "SPI DMA Module".

To accomplish communication, typically three pins are used:

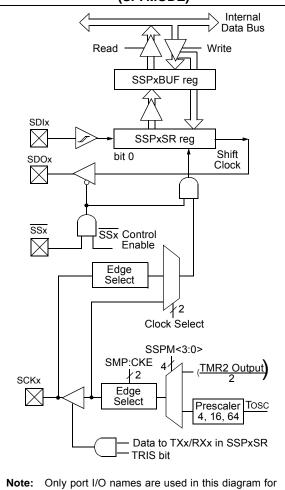
- Serial Data Out (SDOx) RC5/SDO1/RP16 or SDO2/Remappable
- Serial Data In (SDIx) RC4/SDI1/SDA1/RP15 or SDI2/Remappable
- Serial Clock (SCKx) RC3/SCK1/SCL1/RP14 or SCK2/Remappable

Additionally, a fourth pin may be used when in a Slave mode of operation:

 Slave Select (SSx) – RA5/AN4/SS1/ HLVDIN/RP2 or SS2/Remappable

Figure 19-1 depicts the block diagram of the MSSP module when operating in SPI mode.

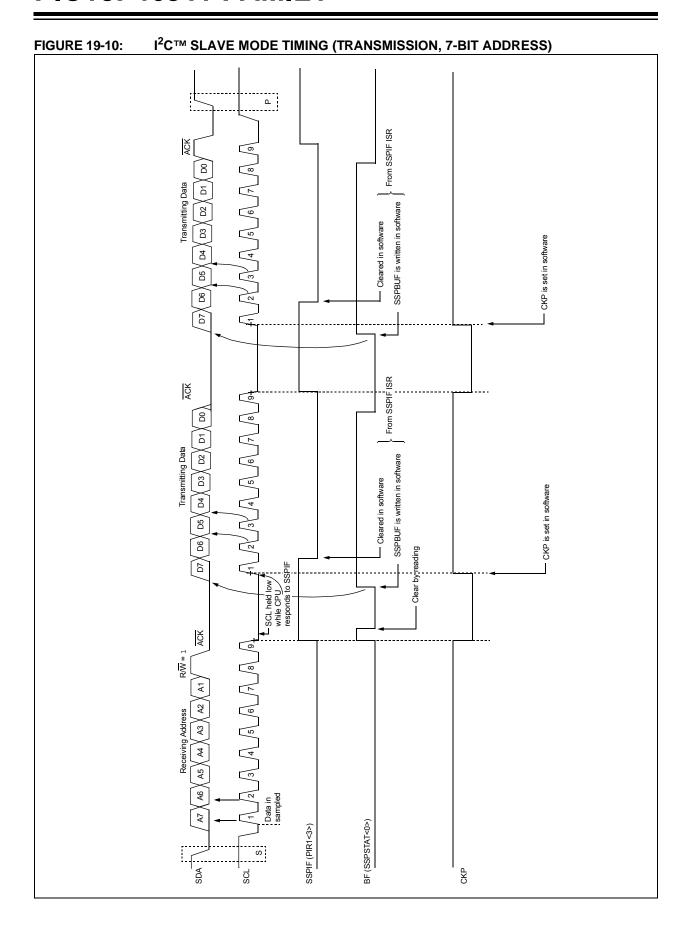
FIGURE 19-1: MSSPx BLOCK DIAGRAM (SPI MODE)



Note: Only port I/O names are used in this diagram for the sake of brevity. Refer to the text for a full list of multiplexed functions.

EXAMPLE 19-2: 512-BYTE SPI MASTER MODE Init AND TRANSFER (CONTINUED)

			;Somewhere else in our project, lets assume we have ;allocated some RAM for use as SPI receive and ;transmit buffers.
;	udata	0x500	
;DestBuf	res	0x200	;Let's reserve 0x500-0x6FF for use as our SPI
;			receive data buffer in this example
;SrcBuf	res	0x200	Lets reserve 0x700-0x8FF for use as our SPI
,			transmit data buffer in this example
PrepareTransfe	er:		
movlw	HIGH(Dest	Buf)	Get high byte of DestBuf address (0x05)
movwf	RXADDRH		;Load upper four bits of the RXADDR register
movlw	LOW(DestBu	ıf)	; Get low byte of the DestBuf address $(0x00)$
movwf	RXADDRL		;Load lower eight bits of the RXADDR register
movlw	HIGH(SrcBu	ıf)	;Get high byte of SrcBuf address (0x07)
movwf	TXADDRH		Load upper four bits of the TXADDR register
movlw	LOW(SrcBui	=)	;Get low byte of the SrcBuf address (0x00)
movwf	TXADDRL		;Load lower eight bits of the TXADDR register
movlw	0x01		;Lets move 0x200 (512) bytes in one DMA xfer
movwf	DMABCH		;Load the upper two bits of DMABC register
movlw	0xFF		Actual bytes transferred is (DMABC + 1), so
movwf	DMABCL		;we load 0x01FF into DMABC to xfer 0x200 bytes
BeginXfer:			
bsf	DMACON1, I	OMAEN	;The SPI DMA module will now begin transferring
			;the data taken from SrcBuf, and will store
			received bytes into DestBuf.
;Execute w	vhatever		CPU is now free to do whatever it wants to and the DMA operation will continue without intervention, until it completes.
			;When the transfer is complete, the SSP2IF flag in ;the PIR3 register will become set, and the DMAEN bit ;is automatically cleared by the hardware. ;The DestBuf (0x500-0x7FF) will contain the received ;data. To start another transfer, firmware will need ;to reinitialize RXADDR, TXADDR, DMABC and then ;set the DMAEN bit.



19.5.4 CLOCK STRETCHING

Both 7-Bit and 10-Bit Slave modes implement automatic clock stretching during a transmit sequence.

The SEN bit (SSPxCON2<0>) allows clock stretching to be enabled during receives. Setting SEN will cause the SCLx pin to be held low at the end of each data receive sequence.

19.5.4.1 Clock Stretching for 7-Bit Slave Receive Mode (SEN = 1)

In 7-Bit Slave Receive mode, on the falling edge of the ninth clock at the end of the ACK sequence, if the BF bit is set, the CKP bit in the SSPxCON1 register is automatically cleared, forcing the SCLx output to be held low. The CKP bit being cleared to '0' will assert the SCLx line low. The CKP bit must be set in the user's ISR before reception is allowed to continue. By holding the SCLx line low, the user has time to service the ISR and read the contents of the SSPxBUF before the master device can initiate another receive sequence. This will prevent buffer overruns from occurring (see Figure 19-15).

- Note 1: If the user reads the contents of the SSPxBUF before the falling edge of the ninth clock, thus clearing the BF bit, the CKP bit will not be cleared and clock stretching will not occur.
 - 2: The CKP bit can be set in software regardless of the state of the BF bit. The user should be careful to clear the BF bit in the ISR before the next receive sequence in order to prevent an overflow condition.

19.5.4.2 Clock Stretching for 10-Bit Slave Receive Mode (SEN = 1)

In 10-Bit Slave Receive mode, during the address sequence, clock stretching automatically takes place but CKP is not cleared. During this time, if the UA bit is set after the ninth clock, clock stretching is initiated. The UA bit is set after receiving the upper byte of the 10-bit address and following the receive of the second byte of the 10-bit address with the R/W bit cleared to '0'. The release of the clock line occurs upon updating SSPxADD. Clock stretching will occur on each data receive sequence as described in 7-bit mode.

Note: If the user polls the UA bit and clears it by updating the SSPxADD register before the falling edge of the ninth clock occurs, and if the user has not cleared the BF bit by reading the SSPxBUF register before that time, then the CKP bit will still NOT be asserted low. Clock stretching on the basis of the state of the BF bit only occurs during a data sequence, not an address sequence.

19.5.4.3 Clock Stretching for 7-Bit Slave Transmit Mode

The 7-Bit Slave Transmit mode implements clock stretching by clearing the CKP bit after the falling edge of the ninth clock if the BF bit is clear. This occurs regardless of the state of the SEN bit.

The user's Interrupt Service Routine (ISR) must set the CKP bit before transmission is allowed to continue. By holding the SCLx line low, the user has time to service the ISR and load the contents of the SSPxBUF before the master device can initiate another transmit sequence (see Figure 19-10).

- Note 1: If the user loads the contents of SSPxBUF, setting the BF bit before the falling edge of the ninth clock, the CKP bit will not be cleared and clock stretching will not occur.
 - **2:** The CKP bit can be set in software regardless of the state of the BF bit.

19.5.4.4 Clock Stretching for 10-Bit Slave Transmit Mode

In 10-Bit Slave Transmit mode, clock stretching is controlled during the first two address sequences by the state of the UA bit, just as it is in 10-Bit Slave Receive mode. The first two addresses are followed by a third address sequence, which contains the high-order bits of the 10-bit address and the R/\overline{W} bit set to '1'. After the third address sequence is performed, the UA bit is not set, the module is now configured in Transmit mode and clock stretching is controlled by the BF flag as in 7-Bit Slave Transmit mode (see Figure 19-13).

TABLE 20-3: BAUD RATES FOR ASYNCHRONOUS MODES (CONTINUED)

		SYNC = 0, BRGH = 0, BRG16 = 1												
BAUD RATE	Fosc = 40.000 MHz			Fosc = 20.000 MHz			Fosc = 10.000 MHz			Fosc = 8.000 MHz				
(K)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)		
0.3	0.300	0.00	8332	0.300	0.02	4165	0.300	0.02	2082	0.300	-0.04	1665		
1.2	1.200	0.02	2082	1.200	-0.03	1041	1.200	-0.03	520	1.201	-0.16	415		
2.4	2.402	0.06	1040	2.399	-0.03	520	2.404	0.16	259	2.403	-0.16	207		
9.6	9.615	0.16	259	9.615	0.16	129	9.615	0.16	64	9.615	-0.16	51		
19.2	19.231	0.16	129	19.231	0.16	64	19.531	1.73	31	19.230	-0.16	25		
57.6	58.140	0.94	42	56.818	-1.36	21	56.818	-1.36	10	55.555	3.55	8		
115.2	113.636	-1.36	21	113.636	-1.36	10	125.000	8.51	4	-	_	_		

	SYNC = 0, BRGH = 0, BRG16 = 1											
BAUD RATE	Fosc	c = 4.000	MHz	Fos	c = 2.000	MHz	Fosc = 1.000 MHz					
(K)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)			
0.3	0.300	0.04	832	0.300	-0.16	415	0.300	-0.16	207			
1.2	1.202	0.16	207	1.201	-0.16	103	1.201	-0.16	51			
2.4	2.404	0.16	103	2.403	-0.16	51	2.403	-0.16	25			
9.6	9.615	0.16	25	9.615	-0.16	12	_	_	_			
19.2	19.231	0.16	12	_	_	_	_	_	_			
57.6	62.500	8.51	3	_	_	_	_	_	_			
115.2	125.000	8.51	1	-	_	_	_	_	_			

	SYNC = 0, BRGH = 1, BRG16 = 1 or SYNC = 1, BRG16 = 1												
BAUD RATE (K)	Fosc	= 40.000) MHz	Fosc	= 20.000	0 MHz	Fosc	= 10.000) MHz	Fosc	Fosc = 8.000 MHz		
	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)	
0.3	0.300	0.00	33332	0.300	0.00	16665	0.300	0.00	8332	0.300	-0.01	6665	
1.2	1.200	0.00	8332	1.200	0.02	4165	1.200	0.02	2082	1.200	-0.04	1665	
2.4	2.400	0.02	4165	2.400	0.02	2082	2.402	0.06	1040	2.400	-0.04	832	
9.6	9.606	0.06	1040	9.596	-0.03	520	9.615	0.16	259	9.615	-0.16	207	
19.2	19.193	-0.03	520	19.231	0.16	259	19.231	0.16	129	19.230	-0.16	103	
57.6	57.803	0.35	172	57.471	-0.22	86	58.140	0.94	42	57.142	0.79	34	
115.2	114.943	-0.22	86	116.279	0.94	42	113.636	-1.36	21	117.647	-2.12	16	

	SYNC = 0, BRGH = 1, BRG16 = 1 or SYNC = 1, BRG16 = 1									
BAUD	Fos	c = 4.000	MHz	Fos	Fosc = 2.000 MHz			Fosc = 1.000 MHz		
RATE (K)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)	
0.3	0.300	0.01	3332	0.300	-0.04	1665	0.300	-0.04	832	
1.2	1.200	0.04	832	1.201	-0.16	415	1.201	-0.16	207	
2.4	2.404	0.16	415	2.403	-0.16	207	2.403	-0.16	103	
9.6	9.615	0.16	103	9.615	-0.16	51	9.615	-0.16	25	
19.2	19.231	0.16	51	19.230	-0.16	25	19.230	-0.16	12	
57.6	58.824	2.12	16	55.555	3.55	8	_	_	_	
115.2	111.111	-3.55	8	_	_	_	_	_	_	

REGISTER 25-2: CTMUCONL: CTMU CONTROL REGISTER LOW (ACCESS FB2h)

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-x	R/W-x
EDG2POL	EDG2SEL1	EDG2SEL0	EDG1POL	EDG1SEL1	EDG1SEL0	EDG2STAT	EDG1STAT
bit 7							bit 0

Legend:

R = Readable bit W = Writable bit U = Unimplemented bit, read as '0'

-n = Value at POR '1' = Bit is set '0' = Bit is cleared x = Bit is unknown

bit 7 EDG2POL: Edge 2 Polarity Select bit

1 = Edge 2 programmed for a positive edge response0 = Edge 2 programmed for a negative edge response

bit 6-5 EDG2SEL<1:0>: Edge 2 Source Select bits

11 = CTED1 pin 10 = CTED2 pin

01 = ECCP1 Special Event Trigger 00 = ECCP2 Special Event Trigger

bit 4 EDG1POL: Edge 1 Polarity Select bit

1 = Edge 1 programmed for a positive edge response 0 = Edge 1 programmed for a negative edge response

bit 3-2 EDG1SEL<1:0>: Edge 1 Source Select bits

11 = CTED1 pin 10 = CTED2 pin

01 = ECCP1 Special Event Trigger00 = ECCP2 Special Event Trigger

bit 1 EDG2STAT: Edge 2 Status bit

1 = Edge 2 event has occurred0 = Edge 2 event has not occurred

bit 0 EDG1STAT: Edge 1 Status bit

1 = Edge 1 event has occurred0 = Edge 1 event has not occurred

BTG Bit Toggle f Syntax: BTG f, b {,a} Operands: $0 \leq f \leq 255$ $0 \le b < 7$ $a \in [0,1]$ $(\overline{f < b >}) \rightarrow f < b >$ Operation: Status Affected: None Encoding: 0111 bbba ffff ffff Bit 'b' in data memory location 'f' is Description:

inverted.

If 'a' is '0', the Access Bank is selected. If 'a' is '1', the BSR is used to select the GPR bank (default).

If 'a' is '0' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever $f \le 95$ (5Fh). See Section 27.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode" for details.

Words: 1
Cycles: 1

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	Read	Process	Write
	register 'f'	Data	register 'f'

Example: BTG LATC, 4,0

Before Instruction:

LATC = 0111 0101 [75h]

After Instruction:

LATC = 0110 0101 [65h]

BOV Branch if Overflow

Syntax: BOV n

Operands: $-128 \le n \le 127$

Operation: if Overflow bit is '1', $(PC) + 2 + 2n \rightarrow PC$

Status Affected: None

Encoding: 1110 0100 nnnn nnnn

Description: If the Overflow bit is '1', then the

program will branch.

The 2's complement number '2n' is added to the PC. Since the PC will have incremented to fetch the next instruction, the new address will be PC + 2 + 2n. This instruction is then a

two-cycle instruction.

Words: 1 Cycles: 1(2)

Q Cycle Activity:

If Jump:

Q1	Q2	Q3	Q4
Decode	Read literal	Process	Write to PC
	ʻn'	Data	
No	No	No	No
operation	operation	operation	operation

If No Jump:

Q1	Q2	Q3	Q4
Decode	Read literal	Process	No
	ʻn'	Data	operation

Example: HERE BOV Jump

Before Instruction

PC = address (HERE)

After Instruction

If Overflow = 1;

PC = address (Jump)

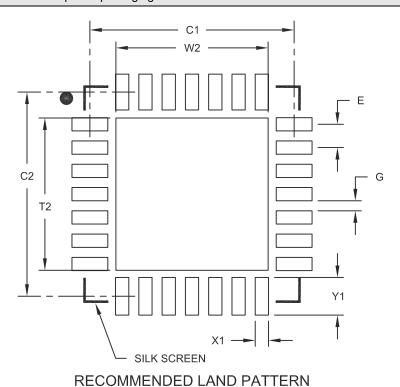
If Overflow = 0;

PC = address (HERE + 2)

DECFSZ	Decrement f, Skip if 0		DCF	SNZ	Decrement f, Skip if not 0					
Syntax:	DECFSZ	f {,d {,a}}		Synta	ax:	DCFSNZ	f {,d {,a}}			
Operands:	$0 \le f \le 255$ $d \in [0,1]$ $a \in [0,1]$			Oper	Operands:		$0 \le f \le 255$ $d \in [0,1]$ $a \in [0,1]$			
Operation:	(f) $-1 \rightarrow de$ skip if resul	-		Oper	ration:	(f) – 1 \rightarrow deskip if result				
Status Affected:	None			Statu	s Affected:	None				
Encoding: 0010 11da ffff ffff		Enco	oding:	0100 11da ffff ff						
Description:	decremente placed in W	ts of register 'ed. If 'd' is '0', /. If 'd' is '1', the self is	the result is ne result is	Desc	cription:	decremente placed in W	ts of register 'fed. If 'd' is '0', the /. If 'd' is '1', the k in register 'f'	the result is e result is		
	which is alr and a NOP i it a two-cyc	le instruction.					If the result is not '0', the next instruction which is already fetched is discarded and a NOP is executed instead, making it a two-cycle instruction. If 'a' is '0', the Access Bank is selected. If 'a' is '1', the BSR is used to select the GPR bank (default).			
	If 'a' is '1', to GPR bank	he BSR is use	d to select the							
	set is enabl in Indexed mode wher Section 27 Bit-Oriente		ction operates Addressing Fh). See riented and as in Indexed			set is enabl in Indexed mode wher Section 27 Bit-Oriente	nd the extended, this instruction that the control of the control	addressing Fh). See iented and s in Indexed		
Words:	1			Word	1-	1	set Mode" for	details.		
Cycles:	•	Note: 3 cycles if skip and followed by a 2-word instruction.			es:	1(2) Note: 3 cycles if skip and followed by a 2-word instruction.				
Q Cycle Activity:				0.0		Бу	a 2-word instri	action.		
Q1 Decede	Q2	Q3	Q4	Q C I	ycle Activity: Q1	Q2	Q3	Q4		
Decode	Read register 'f'	Process Data	Write to destination		Decode	Read	Process	Write to		
If skip:			·I	ı		register 'f'	Data	destination		
Q1	Q2	Q3	Q4	lf sk	•					
No	No	No	No		Q1	Q2	Q3	Q4		
operation If skip and followe	operation		operation		No operation	No operation	No operation	No operation		
Q1	Q2	Q3	Q4	If sk	ip and followe	•		operation.		
No	No	No	No No		Q1	Q2	Q3	Q4		
operation	operation	operation	operation		No	No	No	No		
No	No	No	No		operation	operation	operation	operation		
operation	operation	operation	operation		No operation	No operation	No operation	No operation		
Example:	HERE CONTINUE	DECFSZ GOTO	CNT, 1, 1 LOOP	Exan	nple:	ZERO	:	IP, 1, 0		
Before Instruct PC After Instructi CNT If CNT PC If CNT PC	= Address on = CNT - 7 = 0; = Address ≠ 0;	S (HERE) 1 S (CONTINUE S (HERE + 2			Before Instruc TEMP After Instructio TEMP If TEMP PC If TEMP PC	tion =	? TEMP - 1, 0; Address (2) 0; Address (1)			

28-Lead Plastic Quad Flat, No Lead Package (ML) – 6x6 mm Body [QFN] with 0.55 mm Contact Length

Note: For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



Units **MILLIMETERS Dimension Limits** MIN MOM MAX Contact Pitch 0.65 BSC Ε Optional Center Pad Width W2 4.25 Optional Center Pad Length T2 4.25 Contact Pad Spacing C1 5.70 Contact Pad Spacing C2 5.70 Contact Pad Width (X28) X1 0.37

Y1

G

0.20

Notes:

1. Dimensioning and tolerancing per ASME Y14.5M

Contact Pad Length (X28)

Distance Between Pads

BSC: Basic Dimension. Theoretically exact value shown without tolerances.

Microchip Technology Drawing No. C04-2105A

1.00

APPENDIX A: REVISION HISTORY Revision A (October 2008)

Original data sheet for the PIC18F46J11 family of devices.

Revision B (February 2009)

Changes to the Electrical Characteristics and minor edits throughout text.

Revision C (October 2009)

Removed "Preliminary" marking.

Revision D (March 2011)

Committed data sheet errata changes and minor corrections throughout text.

APPENDIX B: DEVICE DIFFERENCES

The differences between the devices listed in this data sheet are shown in Table B-1,

TABLE B-1: DEVICE DIFFERENCES BETWEEN PIC18F46J11 FAMILY MEMBERS

Features	PIC18F24J11	PIC18F25J11	PIC18F26J11	PIC18F44J11	PIC18F45J11	PIC18F46J11	
Program Memory	16K	32K	64K	16K	32K	64K	
Program Memory (Instructions)	8,192	16,384	32,768	8,192	16,384	32,768	
I/O Ports (Pins)		Ports A, B, C		Ports A, B, C, D, E			
10-Bit ADC Module	10 Input Channels			13 Input Channels			
Packages	28-Pin QFN,	SOIC, SSOP and SP	DIP (300 mil)	44-Pin QFN and TQFP			