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"[Embedded - Microcontrollers](#)" refer to small, integrated circuits designed to perform specific tasks within larger systems. These microcontrollers are essentially compact computers on a single chip, containing a processor core, memory, and programmable input/output peripherals. They are called "embedded" because they are embedded within electronic devices to control various functions, rather than serving as standalone computers. Microcontrollers are crucial in modern electronics, providing the intelligence and control needed for a wide range of applications.

Applications of "[Embedded - Microcontrollers](#)"

Details

Product Status	Not For New Designs
Core Processor	ARM® Cortex®-M3
Core Size	32-Bit Single-Core
Speed	32MHz
Connectivity	I ² C, IrDA, SmartCard, SPI, UART/USART
Peripherals	Brown-out Detect/Reset, DMA, I ² S, LCD, POR, PWM, WDT
Number of I/O	53
Program Memory Size	32KB (32K x 8)
Program Memory Type	FLASH
EEPROM Size	-
RAM Size	4K x 8
Voltage - Supply (Vcc/Vdd)	1.85V ~ 3.8V
Data Converters	A/D 8x12b; D/A 1x12b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	64-TQFP
Supplier Device Package	64-TQFP (10x10)
Purchase URL	https://www.e-xfl.com/product-detail/silicon-labs/efm32tg842f32-qfp64t

1 Ordering Information

Table 1.1 (p. 2) shows the available EFM32TG842 devices.

Table 1.1. Ordering Information

Ordering Code	Flash (kB)	RAM (kB)	Max Speed (MHz)	Supply Voltage (V)	Temperature (°C)	Package
EFM32TG842F8-QFP64	8	2	32	1.98 - 3.8	-40 - 85	TQFP64
EFM32TG842F16-QFP64	16	4	32	1.98 - 3.8	-40 - 85	TQFP64
EFM32TG842F32-QFP64	32	4	32	1.98 - 3.8	-40 - 85	TQFP64

Visit **www.silabs.com** for information on global distributors and representatives.

2.1.19 Voltage Comparator (VCMP)

The Voltage Supply Comparator is used to monitor the supply voltage from software. An interrupt can be generated when the supply falls below or rises above a programmable threshold. Response time and thereby also the current consumption can be configured by altering the current supply to the comparator.

2.1.20 Analog to Digital Converter (ADC)

The ADC is a Successive Approximation Register (SAR) architecture, with a resolution of up to 12 bits at up to one million samples per second. The integrated input mux can select inputs from 8 external pins and 6 internal signals.

2.1.21 Digital to Analog Converter (DAC)

The Digital to Analog Converter (DAC) can convert a digital value to an analog output voltage. The DAC is fully differential rail-to-rail, with 12-bit resolution. It has one single ended output buffer connected to channel 0. The DAC may be used for a number of different applications such as sensor interfaces or sound output.

2.1.22 Operational Amplifier (OPAMP)

The EFM32TG842 features 3 Operational Amplifiers. The Operational Amplifier is a versatile general purpose amplifier with rail-to-rail differential input and rail-to-rail single ended output. The input can be set to pin, DAC or OPAMP, whereas the output can be pin, OPAMP or ADC. The current is programmable and the OPAMP has various internal configurations such as unity gain, programmable gain using internal resistors etc.

2.1.23 Low Energy Sensor Interface (LESENSE)

The Low Energy Sensor Interface (LESENSE[™]), is a highly configurable sensor interface with support for up to 8 individually configurable sensors. By controlling the analog comparators and DAC, LESENSE is capable of supporting a wide range of sensors and measurement schemes, and can for instance measure LC sensors, resistive sensors and capacitive sensors. LESENSE also includes a programmable FSM which enables simple processing of measurement results without CPU intervention. LESENSE is available in energy mode EM2, in addition to EM0 and EM1, making it ideal for sensor monitoring in applications with a strict energy budget.

2.1.24 Advanced Encryption Standard Accelerator (AES)

The AES accelerator performs AES encryption and decryption with 128-bit or 256-bit keys. Encrypting or decrypting one 128-bit data block takes 52 HFCORECLK cycles with 128-bit keys and 75 HFCORECLK cycles with 256-bit keys. The AES module is an AHB slave which enables efficient access to the data and key registers. All write accesses to the AES module must be 32-bit operations, i.e. 8- or 16-bit operations are not supported.

2.1.25 General Purpose Input/Output (GPIO)

In the EFM32TG842, there are 53 General Purpose Input/Output (GPIO) pins, which are divided into ports with up to 16 pins each. These pins can individually be configured as either an output or input. More advanced configurations like open-drain, filtering and drive strength can also be configured individually for the pins. The GPIO pins can also be overridden by peripheral pin connections, like Timer PWM outputs or USART communication, which can be routed to several locations on the device. The GPIO supports up to 16 asynchronous external pin interrupts, which enables interrupts from any pin on the device. Also, the input value of a pin can be routed through the Peripheral Reflex System to other peripherals.

3.4 Current Consumption

Table 3.3. Current Consumption

Symbol	Parameter	Condition	Min	Typ	Max	Unit
I _{EM0}	EM0 current. No prescaling. Running prime number calculation code from Flash. (Production test condition = 14 MHz)	32 MHz HFXO, all peripheral clocks disabled, V _{DD} = 3.0 V		157		μA/MHz
		28 MHz HFRCO, all peripheral clocks disabled, V _{DD} = 3.0 V		150	170	μA/MHz
		21 MHz HFRCO, all peripheral clocks disabled, V _{DD} = 3.0 V		153	172	μA/MHz
		14 MHz HFRCO, all peripheral clocks disabled, V _{DD} = 3.0 V		155	175	μA/MHz
		11 MHz HFRCO, all peripheral clocks disabled, V _{DD} = 3.0 V		157	178	μA/MHz
		6.6 MHz HFRCO, all peripheral clocks disabled, V _{DD} = 3.0 V		162	183	μA/MHz
		1.2 MHz HFRCO, all peripheral clocks disabled, V _{DD} = 3.0 V		200	240	μA/MHz
I _{EM1}	EM1 current (Production test condition = 14 MHz)	32 MHz HFXO, all peripheral clocks disabled, V _{DD} = 3.0 V		53		μA/MHz
		28 MHz HFRCO, all peripheral clocks disabled, V _{DD} = 3.0 V		51	57	μA/MHz
		21 MHz HFRCO, all peripheral clocks disabled, V _{DD} = 3.0 V		55	59	μA/MHz
		14 MHz HFRCO, all peripheral clocks disabled, V _{DD} = 3.0 V		56	61	μA/MHz
		11 MHz HFRCO, all peripheral clocks disabled, V _{DD} = 3.0 V		58	63	μA/MHz
		6.6 MHz HFRCO, all peripheral clocks disabled, V _{DD} = 3.0 V		63	68	μA/MHz
		1.2 MHz HFRCO. all peripheral clocks disabled, V _{DD} = 3.0 V		100	122	μA/MHz
I _{EM2}	EM2 current	EM2 current with RTC prescaled to 1 Hz, 32.768 kHz LFRCO, V _{DD} = 3.0 V, T _{AMB} =25°C		1.0	1.2	μA
		EM2 current with RTC prescaled to 1 Hz, 32.768 kHz LFRCO, V _{DD} = 3.0 V, T _{AMB} =85°C		2.4	5.0	μA
I _{EM3}	EM3 current	V _{DD} = 3.0 V, T _{AMB} =25°C		0.59	1.0	μA
		V _{DD} = 3.0 V, T _{AMB} =85°C		2.0	4.5	μA
I _{EM4}	EM4 current	V _{DD} = 3.0 V, T _{AMB} =25°C		0.02	0.055	μA
		V _{DD} = 3.0 V, T _{AMB} =85°C		0.25	0.70	μA

3.7 Flash

Table 3.6. Flash

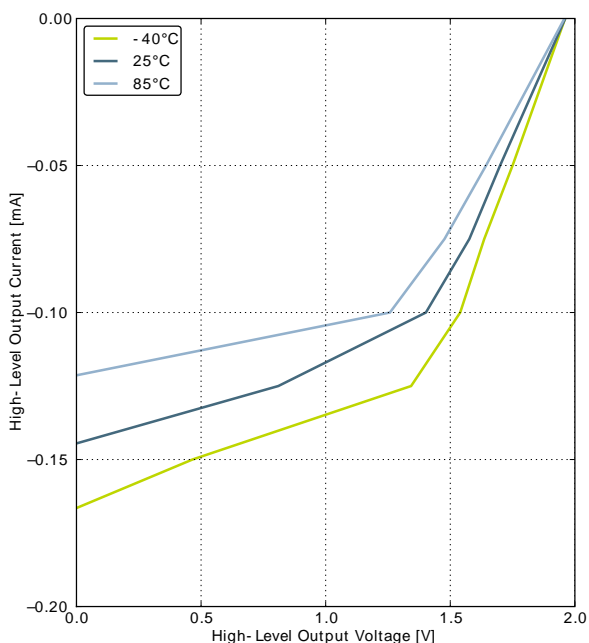
Symbol	Parameter	Condition	Min	Typ	Max	Unit
EC _{FLASH}	Flash erase cycles before failure		20000			cycles
RET _{FLASH}	Flash data retention	T _{AMB} <150°C	10000			h
		T _{AMB} <85°C	10			years
		T _{AMB} <70°C	20			years
t _{W_PROG}	Word (32-bit) programming time		20			µs
t _{P_ERASE}	Page erase time		20	20.4	20.8	ms
t _{D_ERASE}	Device erase time		40	40.8	41.6	ms
I _{ERASE}	Erase current				7 ¹	mA
I _{WRITE}	Write current				7 ¹	mA
V _{FLASH}	Supply voltage during flash erase and write		1.98		3.8	V

¹ Measured at 25°C

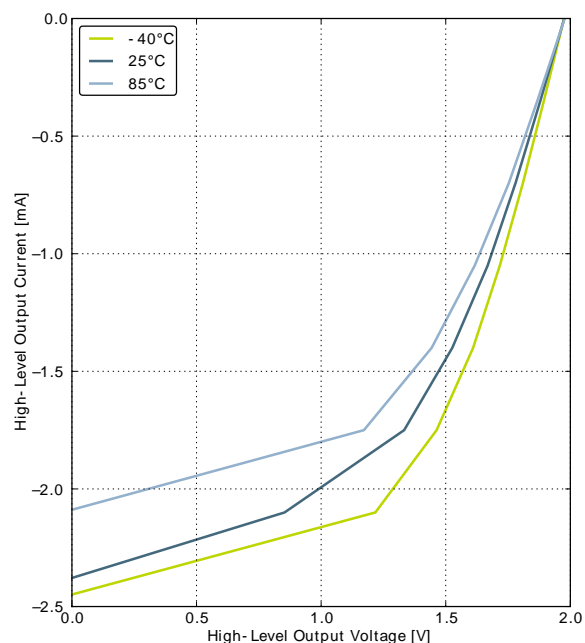
3.8 General Purpose Input Output

Table 3.7. GPIO

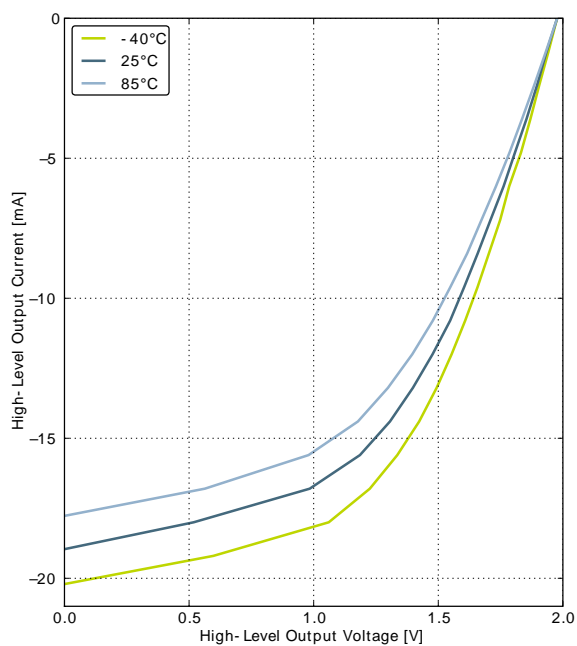
Symbol	Parameter	Condition	Min	Typ	Max	Unit
V _{IOIL}	Input low voltage				0.30V _{DD}	V
V _{IOIH}	Input high voltage		0.70V _{DD}			V
V _{IOOH}	Output high voltage (Production test condition = 3.0V, DRIVEMODE = STANDARD)	Sourcing 0.1 mA, V _{DD} =1.98 V, GPIO_Px_CTRL DRIVEMODE = LOWEST		0.80V _{DD}		V
		Sourcing 0.1 mA, V _{DD} =3.0 V, GPIO_Px_CTRL DRIVEMODE = LOWEST		0.90V _{DD}		V
		Sourcing 1 mA, V _{DD} =1.98 V, GPIO_Px_CTRL DRIVEMODE = LOW		0.85V _{DD}		V
		Sourcing 1 mA, V _{DD} =3.0 V, GPIO_Px_CTRL DRIVEMODE = LOW		0.90V _{DD}		V
		Sourcing 6 mA, V _{DD} =1.98 V, GPIO_Px_CTRL DRIVEMODE = STANDARD	0.75V _{DD}			V
		Sourcing 6 mA, V _{DD} =3.0 V, GPIO_Px_CTRL DRIVEMODE = STANDARD	0.85V _{DD}			V
		Sourcing 20 mA, V _{DD} =1.98 V, GPIO_Px_CTRL DRIVEMODE = HIGH	0.60V _{DD}			V

Figure 3.5. Typical High-Level Output Current, 2V Supply Voltage

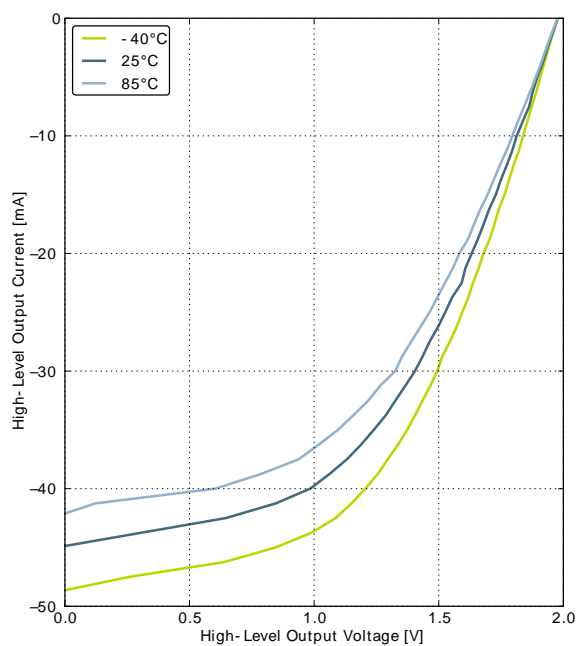
GPIO_Px_CTRL DRIVEMODE = LOWEST



GPIO_Px_CTRL DRIVEMODE = LOW

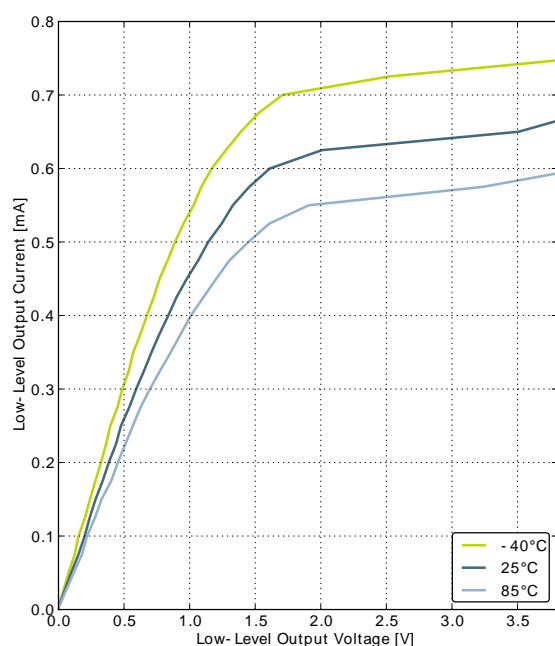


GPIO_Px_CTRL DRIVEMODE = STANDARD

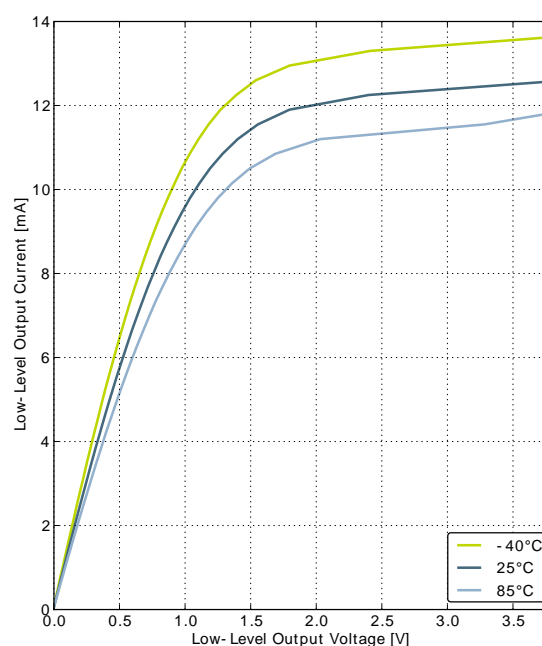


GPIO_Px_CTRL DRIVEMODE = HIGH

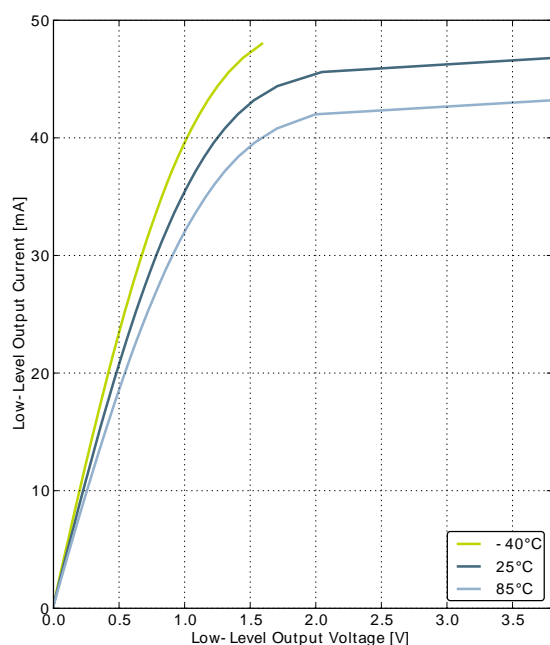
Figure 3.8. Typical Low-Level Output Current, 3.8V Supply Voltage



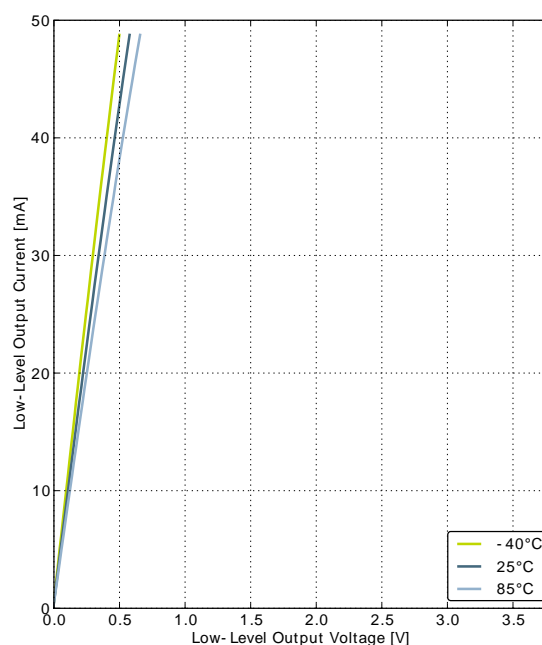
GPIO_Px_CTRL DRIVEMODE = LOWEST



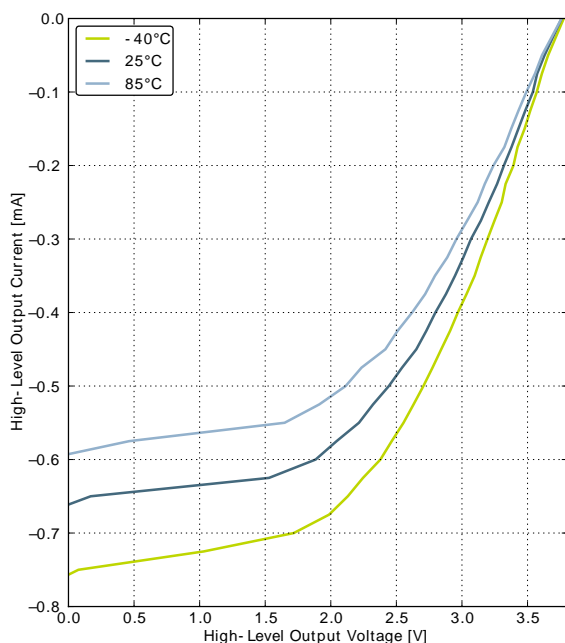
GPIO_Px_CTRL DRIVEMODE = LOW



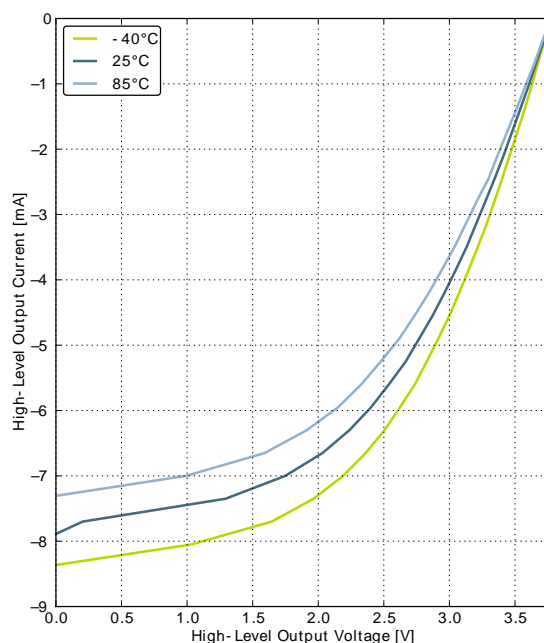
GPIO_Px_CTRL DRIVEMODE = STANDARD



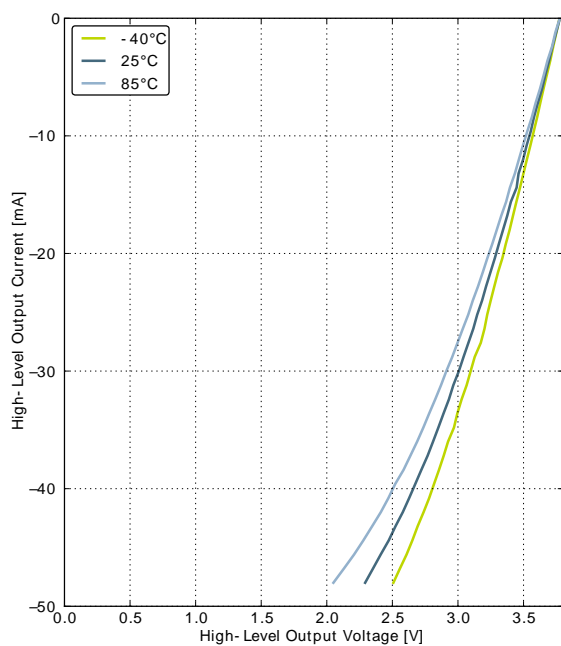
GPIO_Px_CTRL DRIVEMODE = HIGH

Figure 3.9. Typical High-Level Output Current, 3.8V Supply Voltage

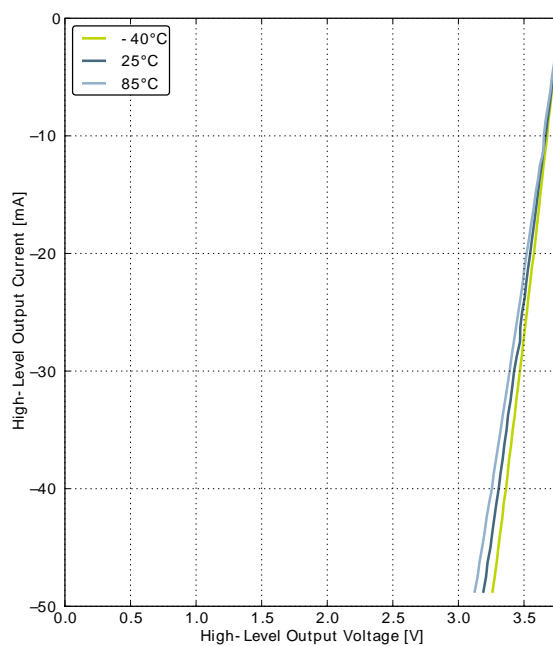
GPIO_Px_CTRL DRIVEMODE = LOWEST



GPIO_Px_CTRL DRIVEMODE = LOW



GPIO_Px_CTRL DRIVEMODE = STANDARD



GPIO_Px_CTRL DRIVEMODE = HIGH

Symbol	Parameter	Condition	Min	Typ	Max	Unit
		$f_{\text{HFRCO}} = 14 \text{ MHz}$		104	120	μA
		$f_{\text{HFRCO}} = 11 \text{ MHz}$		94	110	μA
		$f_{\text{HFRCO}} = 6.6 \text{ MHz}$		63	90	μA
		$f_{\text{HFRCO}} = 1.2 \text{ MHz}$		22	32	μA
TUNESTEP _{H-FRCO}	Frequency step for LSB change in TUNING value			0.3 ³		%

¹For devices with prod. rev. < 19, Typ = 7MHz and Min/Max values not applicable.

²For devices with prod. rev. < 19, Typ = 1MHz and Min/Max values not applicable.

³The TUNING field in the CMU_HFRCOCTRL register may be used to adjust the HFRCO frequency. There is enough adjustment range to ensure that the frequency bands above 7 MHz will always have some overlap across supply voltage and temperature. By using a stable frequency reference such as the LFXO or HFXO, a firmware calibration routine can vary the TUNING bits and the frequency band to maintain the HFRCO frequency at any arbitrary value between 7 MHz and 28 MHz across operating conditions.

Figure 3.11. Calibrated HFRCO 1 MHz Band Frequency vs Supply Voltage and Temperature

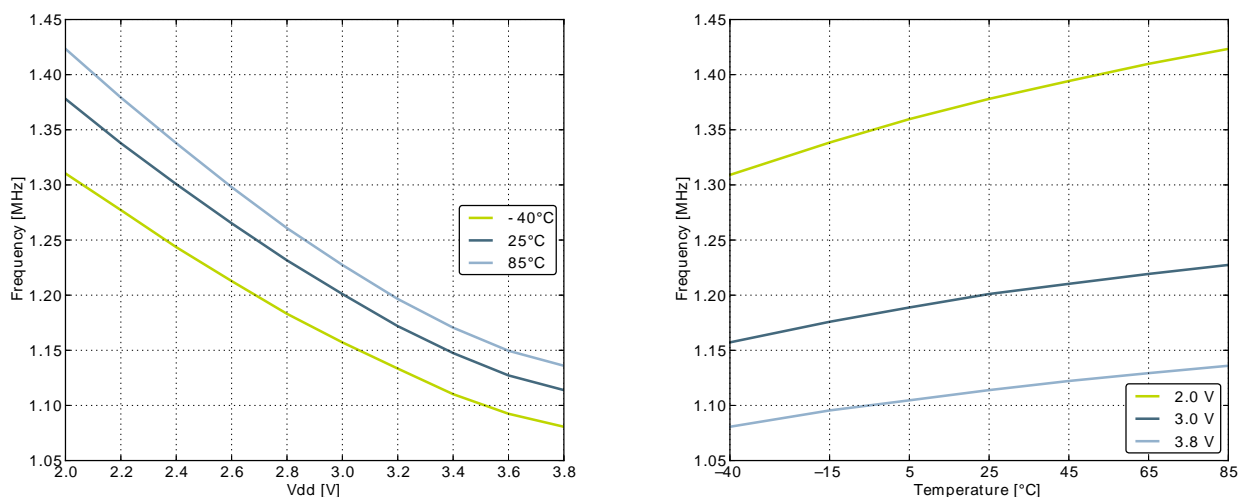


Figure 3.12. Calibrated HFRCO 7 MHz Band Frequency vs Supply Voltage and Temperature

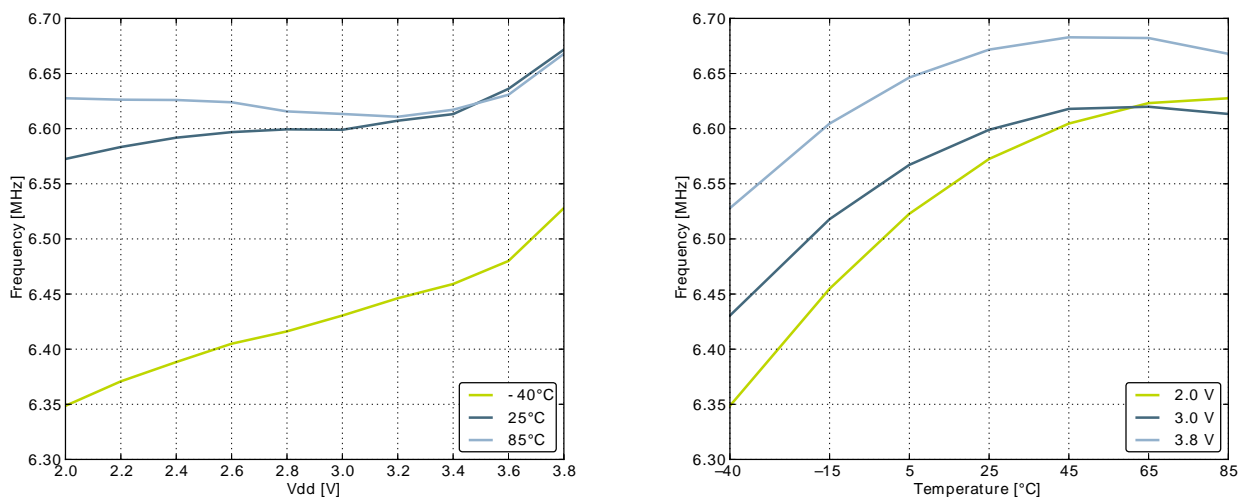
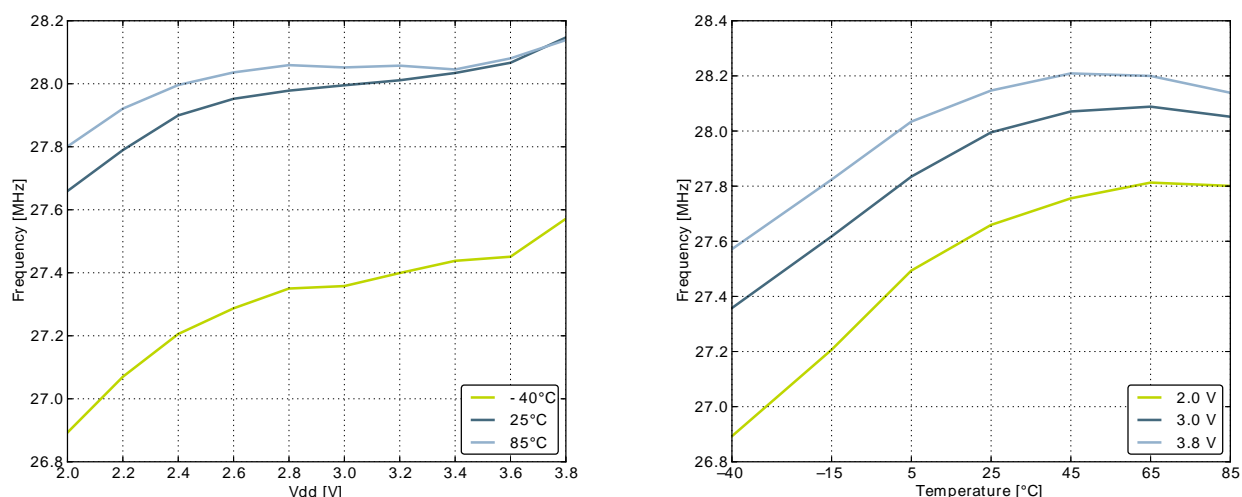


Figure 3.16. Calibrated HFRCO 28 MHz Band Frequency vs Supply Voltage and Temperature

3.9.5 AUXHFRCO

Table 3.12. AUXHFRCO

Symbol	Parameter	Condition	Min	Typ	Max	Unit
f_{AUXHFRCO}	Oscillation frequency, $V_{\text{DD}} = 3.0 \text{ V}$, $T_{\text{AMB}} = 25^\circ\text{C}$	28 MHz frequency band	27.16	28.0	28.84	MHz
		21 MHz frequency band	20.37	21.0	21.63	MHz
		14 MHz frequency band	13.58	14.0	14.42	MHz
		11 MHz frequency band	10.67	11.0	11.33	MHz
		7 MHz frequency band	6.40 ¹	6.60 ¹	6.80 ¹	MHz
		1 MHz frequency band	1.16 ²	1.20 ²	1.24 ²	MHz
$t_{\text{AUXHFRCO_settling}}$	Settling time after start-up	$f_{\text{AUXHFRCO}} = 14 \text{ MHz}$		0.6		Cycles
$\text{TUNESTEP}_{\text{AUXHFRCO}}$	Frequency step for LSB change in TUNING value			0.3 ³		%

¹For devices with prod. rev. < 19, Typ = 7MHz and Min/Max values not applicable.

²For devices with prod. rev. < 19, Typ = 1MHz and Min/Max values not applicable.

³The TUNING field in the CMU_AUXHFRCOCTRL register may be used to adjust the AUXHFRCO frequency. There is enough adjustment range to ensure that the frequency bands above 7 MHz will always have some overlap across supply voltage and temperature. By using a stable frequency reference such as the LFXO or HFXO, a firmware calibration routine can vary the TUNING bits and the frequency band to maintain the AUXHFRCO frequency at any arbitrary value between 7 MHz and 28 MHz across operating conditions.

3.9.6 ULFRCO

Table 3.13. ULFRCO

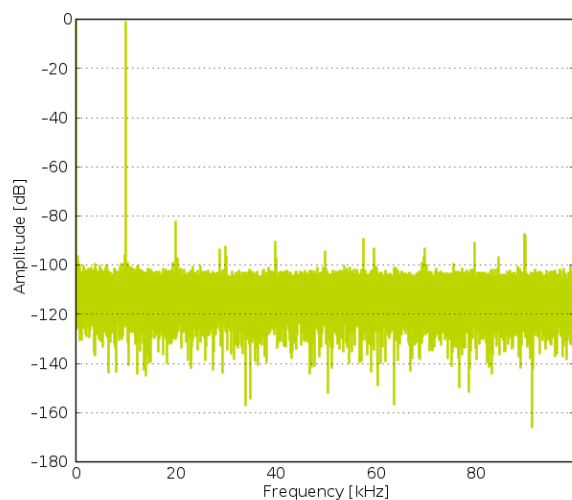
Symbol	Parameter	Condition	Min	Typ	Max	Unit
f_{ULFRCO}	Oscillation frequency	25°C, 3V	0.70		1.75	kHz
$\text{TC}_{\text{ULFRCO}}$	Temperature coefficient			0.05		%/°C
$\text{VC}_{\text{ULFRCO}}$	Supply voltage coefficient			-18.2		%/V

Symbol	Parameter	Condition	Min	Typ	Max	Unit
		1 MSamples/s, 12 bit, single ended, V_{DD} reference		73		dBc
		1 MSamples/s, 12 bit, differential, internal 1.25V reference		66		dBc
		1 MSamples/s, 12 bit, differential, internal 2.5V reference		77		dBc
		1 MSamples/s, 12 bit, differential, V_{DD} reference		76		dBc
		1 MSamples/s, 12 bit, differential, $2xV_{DD}$ reference		75		dBc
		1 MSamples/s, 12 bit, differential, 5V reference		69		dBc
		200 kSamples/s, 12 bit, single ended, internal 1.25V reference		75		dBc
		200 kSamples/s, 12 bit, single ended, internal 2.5V reference		75		dBc
		200 kSamples/s, 12 bit, single ended, V_{DD} reference	68	76		dBc
		200 kSamples/s, 12 bit, differential, internal 1.25V reference		79		dBc
		200 kSamples/s, 12 bit, differential, internal 2.5V reference		79		dBc
		200 kSamples/s, 12 bit, differential, 5V reference		78		dBc
		200 kSamples/s, 12 bit, differential, V_{DD} reference		79		dBc
		200 kSamples/s, 12 bit, differential, $2xV_{DD}$ reference		79		dBc
$V_{ADCOFFSET}$	Offset voltage	After calibration, single ended	-4	0.3	4	mV
		After calibration, differential		0.3		mV
$TGRAD_{ADCTH}$	Thermometer output gradient			-1.92		mV/°C
				-6.3		ADC Codes/°C
DNL_{ADC}	Differential non-linearity (DNL)	$V_{DD}=3.0\text{ V}$, external 2.5V reference	-1	± 0.7	4	LSB
INL_{ADC}	Integral non-linearity (INL), End point method	$V_{DD}=3.0\text{ V}$, external 2.5V reference		± 1.2	± 3	LSB
MC_{ADC}	No missing codes		11.999 ¹	12		bits
$GAIN_{ED}$	Gain error drift	1.25V reference		0.01 ²	0.033 ³	%/°C
		2.5V reference		0.01 ²	0.03 ³	%/°C
$OFFSET_{ED}$	Offset error drift	1.25V reference		0.2 ²	0.7 ³	LSB/°C
		2.5V reference		0.2 ²	0.62 ³	LSB/°C

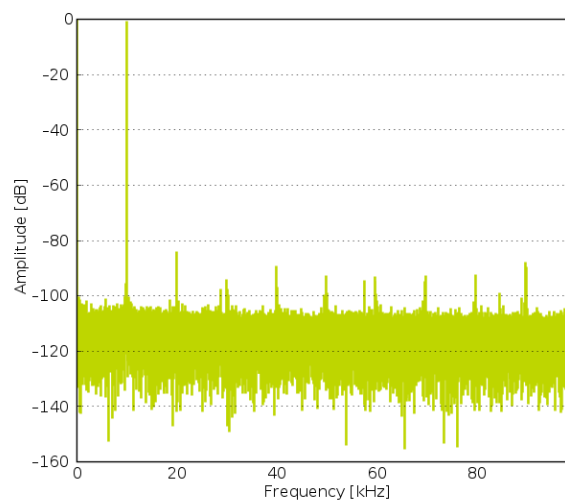
¹ On the average every ADC will have one missing code, most likely to appear around $2048 \pm n \cdot 512$ where n can be a value in the set {-3, -2, -1, 1, 2, 3}. There will be no missing code around 2048, and in spite of the missing code the ADC will be monotonic

3.10.1 Typical performance

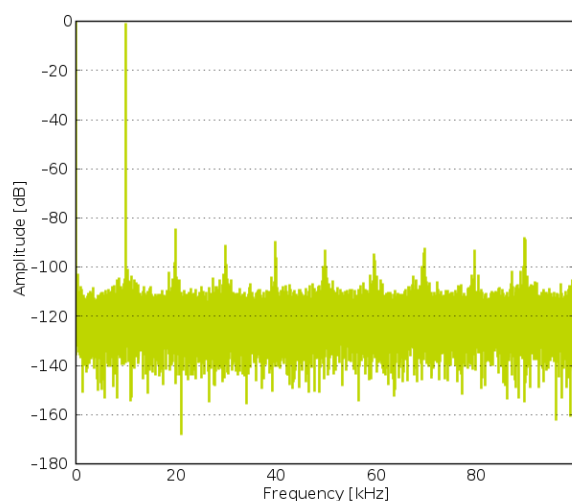
Figure 3.19. ADC Frequency Spectrum, $V_{dd} = 3V$, Temp = 25°C



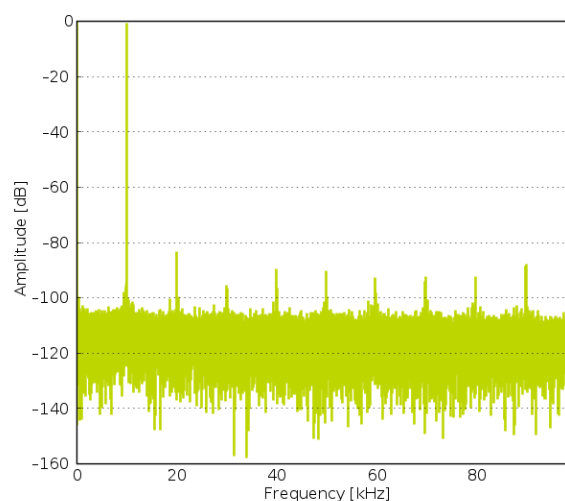
1.25V Reference



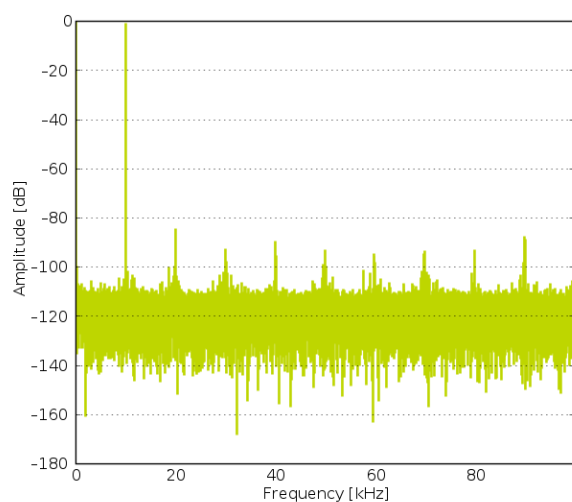
2.5V Reference



2XVDDVSS Reference

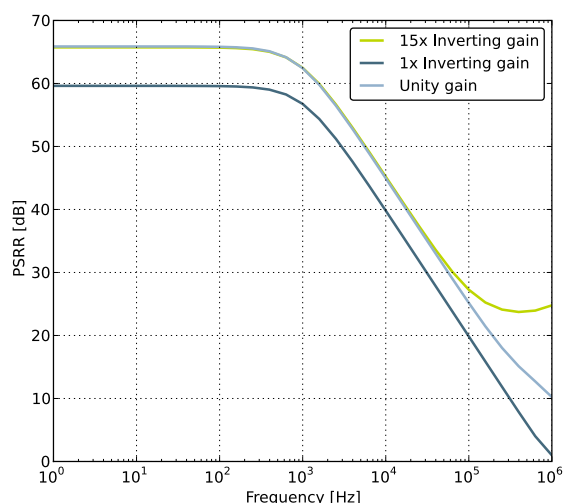
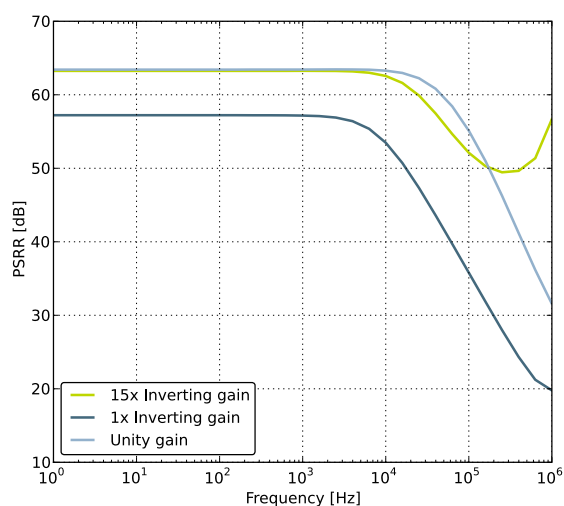
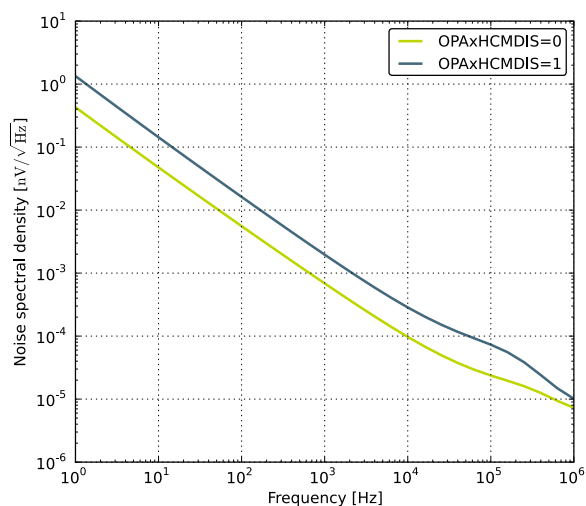


5VDIFF Reference



VDD Reference

Symbol	Parameter	Condition	Min	Typ	Max	Unit
GBW _{OPAMP}	Gain Bandwidth Product	OPA0/OPA1 BIASPROG=0xF, HALFBIAS=0x0		16.36		MHz
		OPA0/OPA1 BIASPROG=0x7, HALFBIAS=0x1		0.81		MHz
		OPA0/OPA1 BIASPROG=0x0, HALFBIAS=0x1		0.11		MHz
		OPA2 BIASPROG=0xF, HALFBIAS=0x0		2.11		MHz
		OPA2 BIASPROG=0x7, HALFBIAS=0x1		0.72		MHz
		OPA2 BIASPROG=0x0, HALFBIAS=0x1		0.09		MHz
PM _{OPAMP}	Phase Margin	BIASPROG=0xF, HALFBIAS=0x0, C _L =75 pF		64		°
		BIASPROG=0x7, HALFBIAS=0x1, C _L =75 pF		58		°
		BIASPROG=0x0, HALFBIAS=0x1, C _L =75 pF		58		°
R _{INPUT}	Input Resistance			100		Mohm
R _{LOAD}	Load Resistance	OPA0/OPA1	200			Ohm
		OPA2	2000			Ohm
I _{LOAD_DC}	Load Current	OPA0/OPA1			11	mA
		OPA2			1.5	mA
V _{INPUT}	Input Voltage	OPAxHCMDIS=0	V _{SS}		V _{DD}	V
		OPAxHCMDIS=1	V _{SS}		V _{DD} -1.2	V
V _{OUTPUT}	Output Voltage		V _{SS}		V _{DD}	V
V _{OFFSET}	Input Offset Voltage	Unity Gain, V _{SS} <V _{in} <V _{DD} , OPAxHCMDIS=0		6		mV
		Unity Gain, V _{SS} <V _{in} <V _{DD} -1.2, OPAxHCMDIS=1		1		mV
V _{OFFSET_DRIFT}	Input Offset Voltage Drift				0.02	mV/°C
SR _{OPAMP}	Slew Rate	OPA0/OPA1 BIASPROG=0xF, HALFBIAS=0x0		46.11		V/μs
		OPA0/OPA1 BIASPROG=0x7, HALFBIAS=0x1		1.21		V/μs
		OPA0/OPA1 BIASPROG=0x0, HALFBIAS=0x1		0.16		V/μs
		OPA2 BIASPROG=0xF, HALFBIAS=0x0		4.43		V/μs
		OPA2 BIASPROG=0x7, HALFBIAS=0x1		1.30		V/μs
		OPA2 BIASPROG=0x0, HALFBIAS=0x1		0.16		V/μs

Figure 3.25. OPAMP Positive Power Supply Rejection Ratio**Figure 3.26. OPAMP Negative Power Supply Rejection Ratio****Figure 3.27. OPAMP Voltage Noise Spectral Density (Unity Gain) $V_{out}=1V$** 

3.14 Voltage Comparator (VCMP)

Table 3.18. VCMP

Symbol	Parameter	Condition	Min	Typ	Max	Unit
V _{VCMPIN}	Input voltage range			V _{DD}		V
V _{VCMP_{CM}}	VCMP Common Mode voltage range			V _{DD}		V
I _{VCMP}	Active current	BIASPROG=0b0000 and HALFBIAS=1 in VCMPn_CTRL register		0.3	0.6	μA
		BIASPROG=0b1111 and HALFBIAS=0 in VCMPn_CTRL register. LPREF=0.		22	30	μA
t _{VCMPREF}	Startup time reference generator	NORMAL		10		μs
V _{VCMP_{OFFSET}}	Offset voltage	Single ended		10		mV
		Differential		10		mV
V _{VCMPHYST}	VCMP hysteresis			17		mV
t _{VCMPSTART}	Startup time				10	μs

The V_{DD} trigger level can be configured by setting the TRIGLEVEL field of the VCMP_CTRL register in accordance with the following equation:

VCMP Trigger Level as a Function of Level Setting

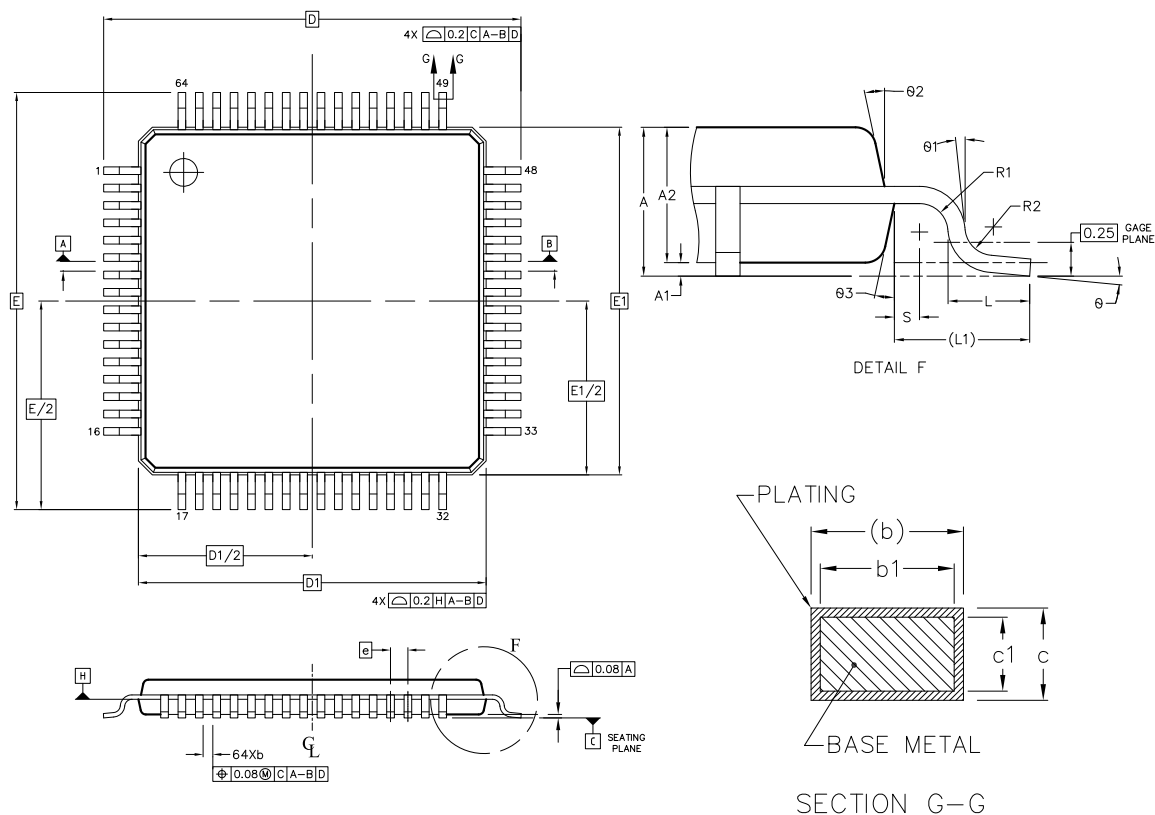
$$V_{DD \text{ Trigger Level}} = 1.667V + 0.034 \times \text{TRIGLEVEL} \quad (3.2)$$

QFP64 Pin# and Name		Pin Alternate Functionality / Description			
Pin #	Pin Name	Analog	Timers	Communication	Other
32	PD4	ADC0_CH4 OPAMP_P2		LEU0_TX #0	
33	PD5	ADC0_CH5 OPAMP_OUT2 #0		LEU0_RX #0	
34	PD6	ADC0_CH6 DAC0_P1 / OPAMP_P1	TIM1_CC0 #4 LETIM0_OUT0 #0 PCNT0_S0IN #3	US1_RX #2 I2C0_SDA #1	LES_ALTEX0 #0 ACMP0_O #2
35	PD7	ADC0_CH7 DAC0_N1 / OPAMP_N1	TIM1_CC1 #4 LETIM0_OUT1 #0 PCNT0_S1IN #3	US1_TX #2 I2C0_SCL #1	CMU_CLK0 #2 LES_ALTEX1 #0 ACMP1_O #2
36	PD8				CMU_CLK1 #1
37	PC6	ACMP0_CH6		I2C0_SDA #2	LES_CH6 #0
38	PC7	ACMP0_CH7		I2C0_SCL #2	LES_CH7 #0
39	VDD_DREG	Power supply for on-chip voltage regulator.			
40	DECOUPLE	Decouple output for on-chip voltage regulator. An external capacitance of size C _{DECOUPLE} is required at this pin.			
41	PE4	LCD_COM0		US0_CS #1	
42	PE5	LCD_COM1		US0_CLK #1	
43	PE6	LCD_COM2		US0_RX #1	
44	PE7	LCD_COM3		US0_TX #1	
45	PC12	ACMP1_CH4 DAC0_OUT1ALT #0/ OPAMP_OUT1ALT			CMU_CLK0 #1 LES_CH12 #0
46	PC13	ACMP1_CH5 DAC0_OUT1ALT #1/ OPAMP_OUT1ALT	TIM1_CC0 #0 TIM1_CC2 #4 PCNT0_S0IN #0		LES_CH13 #0
47	PC14	ACMP1_CH6 DAC0_OUT1ALT #2/ OPAMP_OUT1ALT	TIM1_CC1 #0 PCNT0_S1IN #0	US0_CS #3	LES_CH14 #0
48	PC15	ACMP1_CH7 DAC0_OUT1ALT #3/ OPAMP_OUT1ALT	TIM1_CC2 #0	US0_CLK #3	LES_CH15 #0 DBG_SWO #1
49	PF0		TIM0_CC0 #5 LETIM0_OUT0 #2	US1_CLK #2 LEU0_TX #3 I2C0_SDA #5	DBG_SWCLK #0/1
50	PF1		TIM0_CC1 #5 LETIM0_OUT1 #2	US1_CS #2 LEU0_RX #3 I2C0_SCL #5	DBG_SWDIO #0/1 GPIO_EM4WU3
51	PF2	LCD_SEG0	TIM0_CC2 #5	LEU0_TX #4	ACMP1_O #0 DBG_SWO #0 GPIO_EM4WU4
52	PF3	LCD_SEG1			PRS_CH0 #1
53	PF4	LCD_SEG2			PRS_CH1 #1
54	PF5	LCD_SEG3			PRS_CH2 #1
55	IOVDD_5	Digital IO power supply 5.			
56	VSS	Ground.			
57	PE8	LCD_SEG4			PRS_CH3 #1
58	PE9	LCD_SEG5			
59	PE10	LCD_SEG6	TIM1_CC0 #1	US0_TX #0	BOOT_TX
60	PE11	LCD_SEG7	TIM1_CC1 #1	US0_RX #0	LES_ALTEX5 #0 BOOT_RX

Alternate	LOCATION							
Functionality	0	1	2	3	4	5	6	Description
CMU_CLK1	PA1	PD8	PE12					Clock Management Unit, clock output number 1.
DAC0_N0 / OPAMP_N0	PC5							Operational Amplifier 0 external negative input.
DAC0_N1 / OPAMP_N1	PD7							Operational Amplifier 1 external negative input.
OPAMP_N2	PD3							Operational Amplifier 2 external negative input.
DAC0_OUT0 / OPAMP_OUT0	PB11							Digital to Analog Converter DAC0_OUT0 / OPAMP output channel number 0.
DAC0_OUT0ALT / OPAMP_OUT0ALT					PD0			Digital to Analog Converter DAC0_OUT0ALT / OPAMP alternative output for channel 0.
DAC0_OUT1ALT / OPAMP_OUT1ALT	PC12	PC13	PC14	PC15	PD1			Digital to Analog Converter DAC0_OUT1ALT / OPAMP alternative output for channel 1.
OPAMP_OUT2	PD5	PD0						Operational Amplifier 2 output.
DAC0_P0 / OPAMP_P0	PC4							Operational Amplifier 0 external positive input.
DAC0_P1 / OPAMP_P1	PD6							Operational Amplifier 1 external positive input.
OPAMP_P2	PD4							Operational Amplifier 2 external positive input.
DBG_SWCLK	PF0	PF0						Debug-interface Serial Wire clock input. Note that this function is enabled to pin out of reset, and has a built-in pull down.
DBG_SWDIO	PF1	PF1						Debug-interface Serial Wire data input / output. Note that this function is enabled to pin out of reset, and has a built-in pull up.
DBG_SWO	PF2	PC15						Debug-interface Serial Wire viewer Output. Note that this function is not enabled after reset, and must be enabled by software to be used.
GPIO_EM4WU0	PA0							Pin can be used to wake the system up from EM4
GPIO_EM4WU3	PF1							Pin can be used to wake the system up from EM4
GPIO_EM4WU4	PF2							Pin can be used to wake the system up from EM4
GPIO_EM4WU5	PE13							Pin can be used to wake the system up from EM4
HFX TAL_N	PB14							High Frequency Crystal negative pin. Also used as external optional clock input pin.
HFX TAL_P	PB13							High Frequency Crystal positive pin.
I2C0_SCL	PA1	PD7	PC7			PF1	PE13	I2C0 Serial Clock Line input / output.
I2C0_SDA	PA0	PD6	PC6			PF0	PE12	I2C0 Serial Data input / output.
LCD_BCAP_N	PA13							LCD voltage booster (optional), boost capacitor, negative pin. If using the LCD voltage booster, connect a 22 nF capacitor between LCD_BCAP_N and LCD_BCAP_P.
LCD_BCAP_P	PA12							LCD voltage booster (optional), boost capacitor, positive pin. If using the LCD voltage booster, connect a 22 nF capacitor between LCD_BCAP_N and LCD_BCAP_P.
LCD_BEXT	PA14							LCD voltage booster (optional), boost output. If using the LCD voltage booster, connect a 1 uF capacitor between this pin and VSS. An external LCD voltage may also be applied to this pin if the booster is not enabled. If AVDD is used directly as the LCD supply voltage, this pin may be left unconnected or used as a GPIO.
LCD_COM0	PE4							LCD driver common line number 0.
LCD_COM1	PE5							LCD driver common line number 1.

4.5 TQFP64 Package

Figure 4.3. TQFP64



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Note:

1. All dimensions & tolerancing confirm to ASME Y14.5M-1994.
2. The top package body size may be smaller than the bottom package body size.
3. Datum 'A,B', and 'B' to be determined at datum plane 'H'.
4. To be determined at seating place 'C'.
5. Dimension 'D1' and 'E1' do not include mold protrusions. Allowable protrusion is 0.25mm per side. 'D1' and 'E1' are maximum plastic body size dimension including mold mismatch. Dimension 'D1' and 'E1' shall be determined at datum plane 'H'.
6. Detail of Pin 1 indicatifier are option all but must be located within the zone indicated.
7. Dimension 'b' does not include dambar protrusion. Allowable dambar protrusion shall not cause the lead width to exceed the maximum 'b' dimension by more than 0.08 mm. Dambar can not be located on the lower radius or the foot. Minimum space between protrusion and an adjacent lead is 0.07 mm
8. Exact shape of each corner is optional.
9. These dimension apply to the flat section of the lead between 0.10 mm and 0.25 mm from the lead tip.
10. All dimensions are in millimeters.

Table 4.4. QFP64 (Dimensions in mm)

DIM	MIN	NOM	MAX	DIM	MIN	NOM	MAX
A	-	1.10	1.20	L1	-		
A1	0.05	-	0.15	R1	0.08	-	-
A2	0.95	1.00	1.05	R2	0.08	-	0.20

Added link to Environmental and Quality information.

Re-added missing DAC-data.

7.4 Revision 1.20

September 30th, 2013

Added I2C characterization data.

Corrected GPIO operating voltage from 1.8 V to 1.85 V.

Corrected the ADC gain and offset measurement reference voltage from 2.25 to 2.5V.

Corrected the ADC resolution from 12, 10 and 6 bit to 12, 8 and 6 bit.

Document changed status from "Preliminary".

Updated Environmental information.

Updated trademark, disclaimer and contact information.

Other minor corrections.

7.5 Revision 1.10

June 28th, 2013

Updated power requirements in the Power Management section.

Removed minimum load capacitance figure and table. Added reference to application note.

Other minor corrections.

7.6 Revision 1.00

September 11th, 2012

Updated the HFRCO 1 MHz band typical value to 1.2 MHz.

Updated the HFRCO 7 MHz band typical value to 6.6 MHz.

Added GPIO_EM4WU3, GPIO_EM4WU4 and GPIO_EM4WU5 pins and removed GPIO_EM4WU1 in the Alternate functionality overview table.

Other minor corrections.

7.7 Revision 0.96

May 4th, 2012

Corrected PCB footprint figures and tables.

7.8 Revision 0.95

February 27th, 2012

Corrected operating voltage from 1.8 V to 1.85 V.

Added rising POR level and corrected Thermometer output gradient in Electrical Characteristics section.

Updated Minimum Load Capacitance (C_{LFXOL}) Requirement For Safe Crystal Startup.

Added Gain error drift and Offset error drift to ADC table.

Added reference to errata document.

7.9 Revision 0.92

July 22nd, 2011

Updated current consumption numbers from latest device characterization data.

Updated OPAMP electrical characteristics.

Made ADC plots render properly in Adobe Reader.

Corrected number of DAC channels available.

7.10 Revision 0.90

June 30th, 2011

Initial preliminary release.

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