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Understanding Embedded - FPGAs (Field Programmable Gate Array)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details	
Product Status	Active
Number of LABs/CLBs	-
Number of Logic Elements/Cells	-
Total RAM Bits	36864
Number of I/O	114
Number of Gates	250000
Voltage - Supply	1.425V ~ 1.575V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 85°C (TJ)
Package / Case	256-LBGA
Supplier Device Package	256-FPBGA (17x17)
Purchase URL	https://www.e-xfl.com/product-detail/microchip-technology/afs250-1fgg256

Fusion Device Architecture Overview

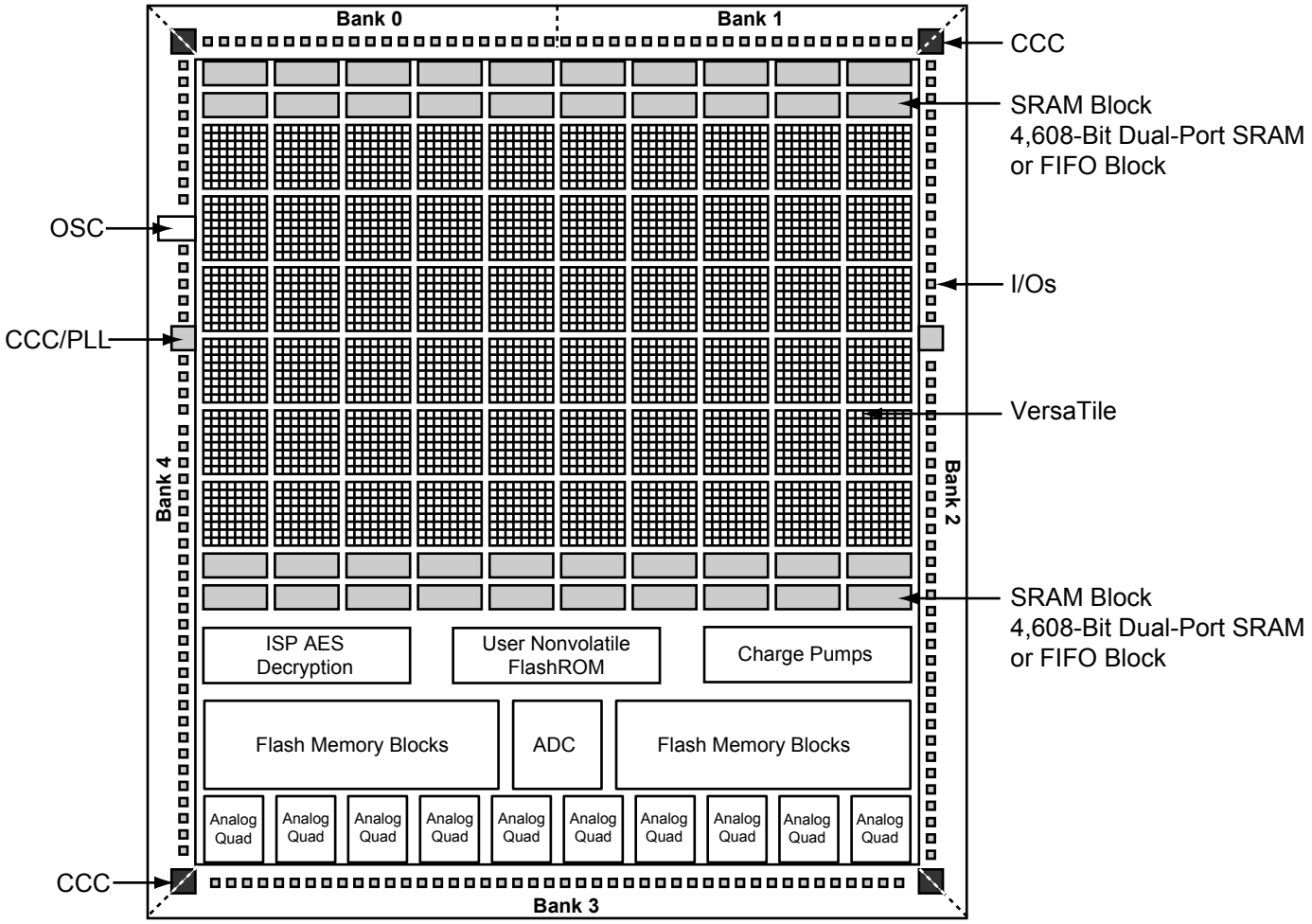


Figure 1 • **Fusion Device Architecture Overview (AFS600)**

Package I/Os: Single-/Double-Ended (Analog)

Fusion Devices	**AFS090**	**AFS250**	**AFS600**	**AFS1500**
ARM Cortex-M1 Devices		**M1AFS250**	**M1AFS600**	**M1AFS1500**
Pigeon Point Devices			**P1AFS600** [1]	**P1AFS1500** [1]
MicroBlade Devices		**U1AFS250** [2]	**U1AFS600** [2]	**U1AFS1500** [2]
QN108 [3]	37/9 (16)			
QN180 [3]	60/16 (20)	65/15 (24)		
PQ208 [4]		93/26 (24)	95/46 (40)	
FG256	75/22 (20)	114/37 (24)	119/58 (40)	119/58 (40)
FG484			172/86 (40)	223/109 (40)
FG676				252/126 (40)

Notes:

1. *Pigeon Point devices are only offered in FG484 and FG256.*
2. *MicroBlade devices are only offered in FG256.*
3. *Package not available.*
4. *Fusion devices in the same package are pin compatible with the exception of the PQ208 package (AFS250 and AFS600).*

VersaTile Characteristics

Sample VersaTile Specifications—Combinatorial Module

The Fusion library offers all combinations of LUT-3 combinatorial functions. In this section, timing characteristics are presented for a sample of the library (Figure 2-3). For more details, refer to the *IGLOO, ProASIC3, SmartFusion, and Fusion Macro Library Guide*.

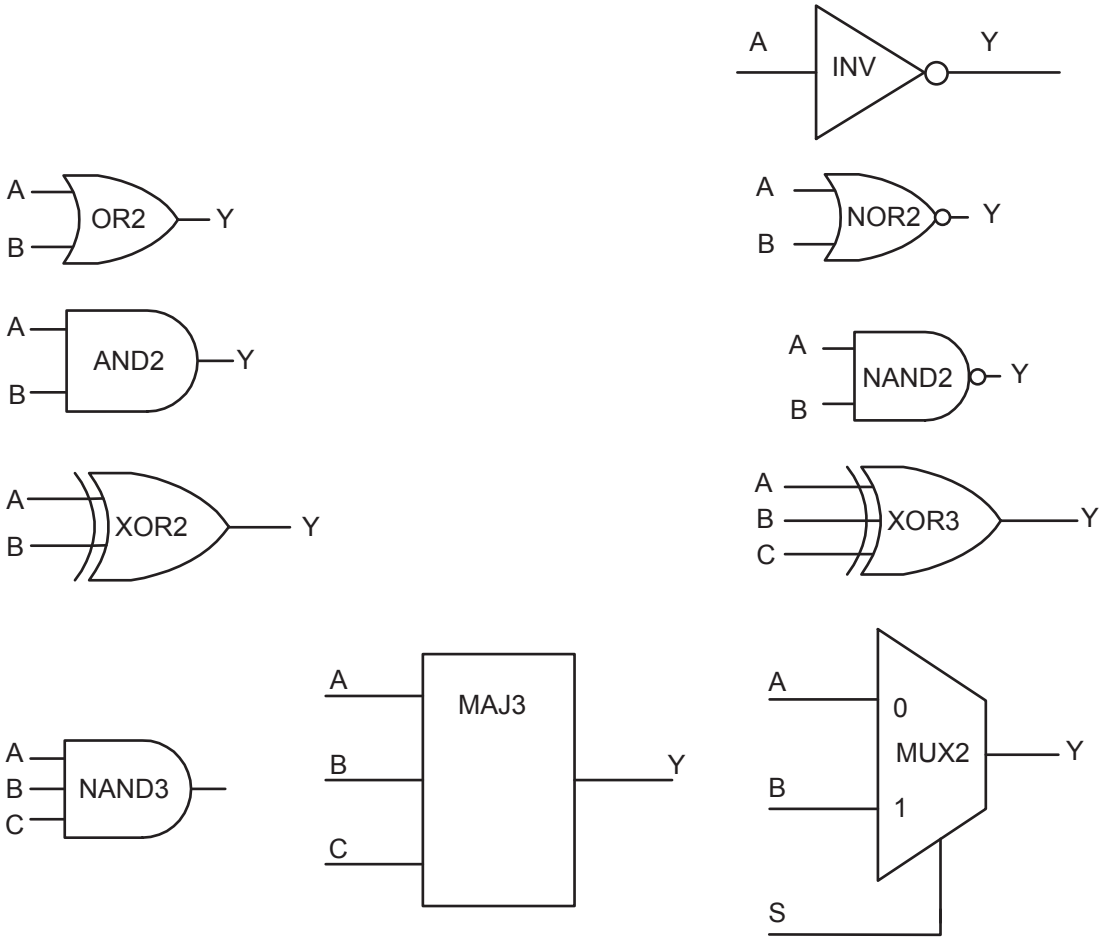


Figure 2-3 • **Sample of Combinatorial Cells**

Routing Architecture

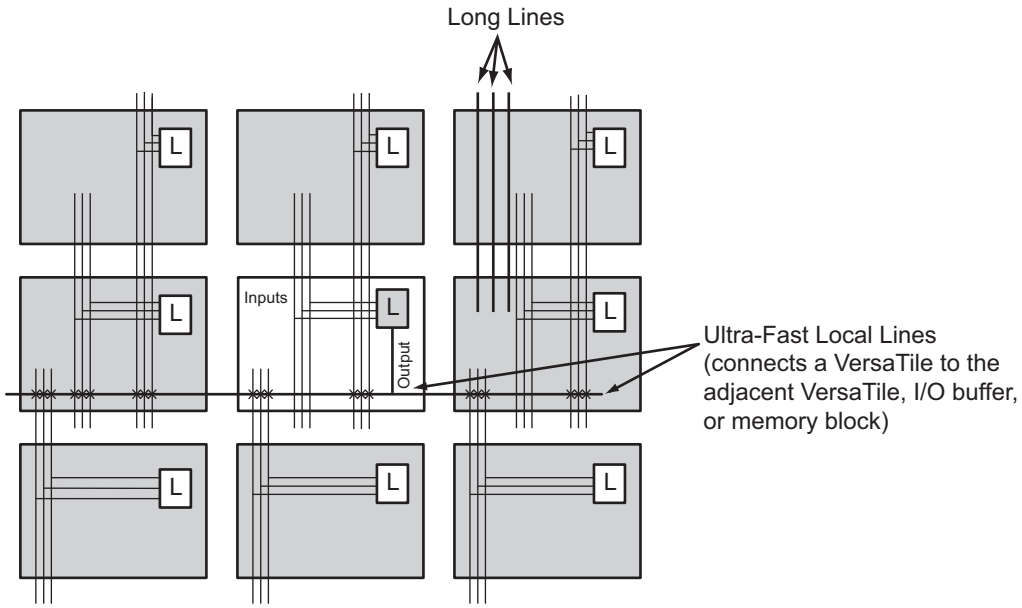
The routing structure of Fusion devices is designed to provide high performance through a flexible four-level hierarchy of routing resources: ultra-fast local resources; efficient long-line resources; high-speed very-long-line resources; and the high-performance VersaNet networks.

The ultra-fast local resources are dedicated lines that allow the output of each VersaTile to connect directly to every input of the eight surrounding VersaTiles (Figure 2-8). The exception to this is that the SET/CLR input of a VersaTile configured as a D-flip-flop is driven only by the VersaNet global network.

The efficient long-line resources provide routing for longer distances and higher-fanout connections. These resources vary in length (spanning one, two, or four VersaTiles), run both vertically and horizontally, and cover the entire Fusion device (Figure 2-9 on page 2-9). Each VersaTile can drive signals onto the efficient long-line resources, which can access every input of every VersaTile. Active buffers are inserted automatically by routing software to limit loading effects.

The high-speed very-long-line resources, which span the entire device with minimal delay, are used to route very long or high-fanout nets: length ±12 VersaTiles in the vertical direction and length ±16 in the horizontal direction from a given core VersaTile (Figure 2-10 on page 2-10). Very long lines in Fusion devices, like those in ProASIC3 devices, have been enhanced. This provides a significant performance boost for long-reach signals.

The high-performance VersaNet global networks are low-skew, high-fanout nets that are accessible from external pins or from internal logic (Figure 2-11 on page 2-11). These nets are typically used to distribute clocks, reset signals, and other high-fanout nets requiring minimum skew. The VersaNet networks are implemented as clock trees, and signals can be introduced at any junction. These can be employed hierarchically, with signals accessing every input on all VersaTiles.



Note: Input to the core cell for the D-flip-flop set and reset is only available via the VersaNet global network connection.

Figure 2-8 • **Ultra-Fast Local Lines Connected to the Eight Nearest Neighbors**

Table 2-7 • AFS250 Global Resource Timing
Commercial Temperature Range Conditions: T_J = 70°C, Worst-Case VCC = 1.425 V

Parameter	Description	–2		–1		Std.		Units
		Min.[1]	Max.[2]	Min.[1]	Max.[2]	Min.[1]	Max.[2]	
t_{RCKL}	Input Low Delay for Global Clock	0.89	1.12	1.02	1.27	1.20	1.50	ns
t_{RCKH}	Input High Delay for Global Clock	0.88	1.14	1.00	1.30	1.17	1.53	ns
$t_{RCKMPWH}$	Minimum Pulse Width High for Global Clock							ns
$t_{RCKMPWL}$	Minimum Pulse Width Low for Global Clock							ns
t_{RCKSW}	Maximum Skew for Global Clock		0.26		0.30		0.35	ns

Notes:

1. *Value reflects minimum load. The delay is measured from the CCC output to the clock pin of a sequential element located in a lightly loaded row (single element is connected to the global net).*
2. *Value reflects maximum load. The delay is measured on the clock pin of the farthest sequential element located in a fully loaded row (all available flip-flops are connected to the global net in the row).*
3. *For the derating values at specific junction temperature and voltage supply levels, refer to Table 3-7 on page 3-9.*

Table 2-8 • AFS090 Global Resource Timing
Commercial Temperature Range Conditions: T_J = 70°C, Worst-Case VCC = 1.425 V

Parameter	Description	–2		–1		Std.		Units
		Min.[1]	Max.[2]	Min.[1]	Max.[2]	Min.[1]	Max.[2]	
t_{RCKL}	Input Low Delay for Global Clock	0.84	1.07	0.96	1.21	1.13	1.43	ns
t_{RCKH}	Input High Delay for Global Clock	0.83	1.10	0.95	1.25	1.12	1.47	ns
$t_{RCKMPWH}$	Minimum Pulse Width High for Global Clock							ns
$t_{RCKMPWL}$	Minimum Pulse Width Low for Global Clock							ns
t_{RCKSW}	Maximum Skew for Global Clock		0.27		0.30		0.36	ns

Notes:

1. *Value reflects minimum load. The delay is measured from the CCC output to the clock pin of a sequential element located in a lightly loaded row (single element is connected to the global net).*
2. *Value reflects maximum load. The delay is measured on the clock pin of the farthest sequential element located in a fully loaded row (all available flip-flops are connected to the global net in the row).*
3. *For the derating values at specific junction temperature and voltage supply levels, refer to Table 3-7 on page 3-9.*

Embedded Memories

Fusion devices include four types of embedded memory: flash block, FlashROM, SRAM, and FIFO.

Flash Memory Block

Fusion is the first FPGA that offers a flash memory block (FB). Each FB block stores 2 Mbits of data. The flash memory block macro is illustrated in Figure 2-32. The port pin name and descriptions are detailed on Table 2-19 on page 2-40. All flash memory block signals are active high, except for CLK and active low RESET. All flash memory operations are synchronous to the rising edge of CLK.

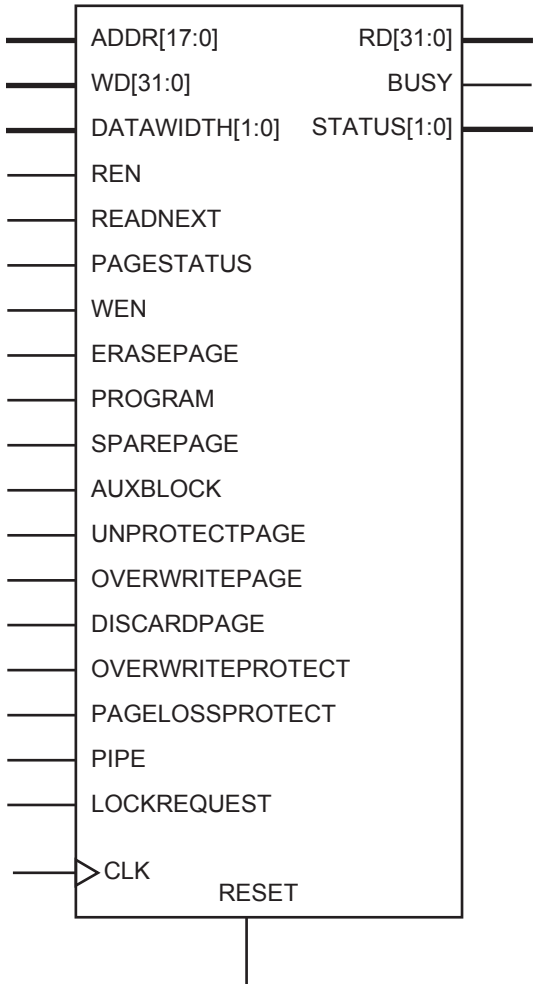


Figure 2-32 • **Flash Memory Block**

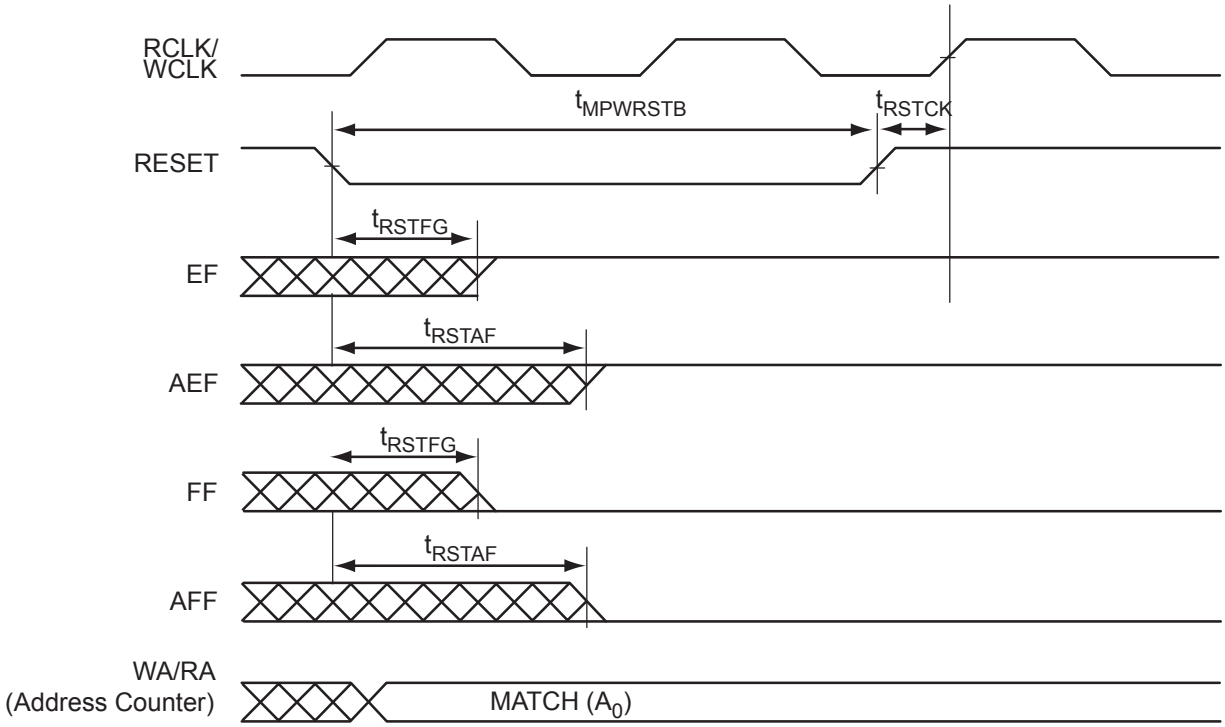


Figure 2-59 • **FIFO Reset**

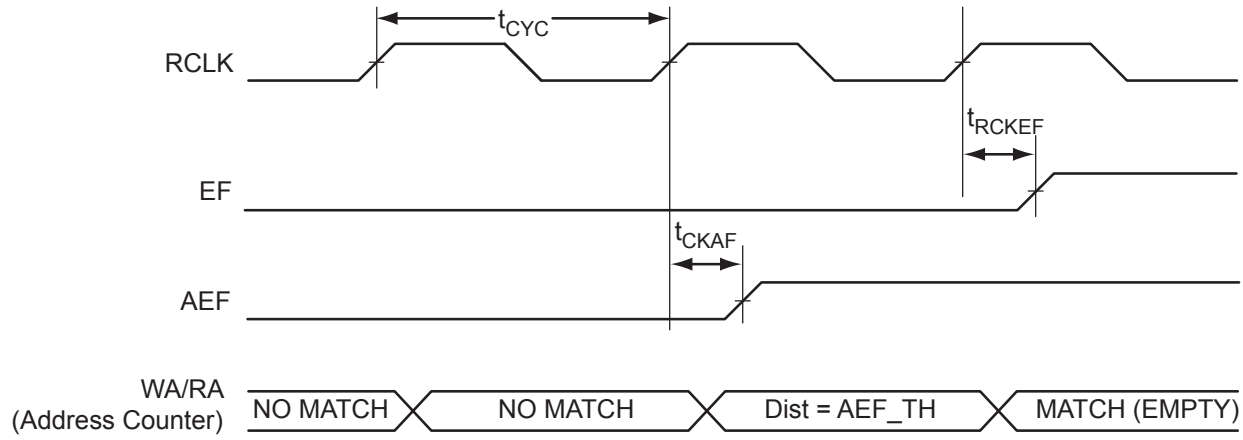


Figure 2-60 • **FIFO EMPTY Flag and AEMPTY Flag Assertion**

Timing Characteristics

Table 2-35 • FIFO
Commercial Temperature Range Conditions: T_J = 70°C, Worst-Case VCC = 1.425 V

Parameter	Description	–2	–1	Std.	Units
t_{ENS}	REN, WEN Setup time	1.34	1.52	1.79	ns
t_{ENH}	REN, WEN Hold time	0.00	0.00	0.00	ns
t_{BKS}	BLK Setup time	0.19	0.22	0.26	ns
t_{BKH}	BLK Hold time	0.00	0.00	0.00	ns
t_{DS}	Input data (WD) Setup time	0.18	0.21	0.25	ns
t_{DH}	Input data (WD) Hold time	0.00	0.00	0.00	ns
t_{CKQ1}	Clock High to New Data Valid on RD (flow-through)	2.17	2.47	2.90	ns
t_{CKQ2}	Clock High to New Data Valid on RD (pipelined)	0.94	1.07	1.26	ns
t_{RCKEF}	RCLK High to Empty Flag Valid	1.72	1.96	2.30	ns
t_{WCKFF}	WCLK High to Full Flag Valid	1.63	1.86	2.18	ns
t_{CKAF}	Clock High to Almost Empty/Full Flag Valid	6.19	7.05	8.29	ns
t_{RSTFG}	RESET Low to Empty/Full Flag Valid	1.69	1.93	2.27	ns
t_{RSTAF}	RESET Low to Almost-Empty/Full Flag Valid	6.13	6.98	8.20	ns
t_{RSTBQ}	RESET Low to Data out Low on RD (flow-through)	0.92	1.05	1.23	ns
	RESET Low to Data out Low on RD (pipelined)	0.92	1.05	1.23	ns
$t_{REMRSTB}$	RESET Removal	0.29	0.33	0.38	ns
$t_{RECRSTB}$	RESET Recovery	1.50	1.71	2.01	ns
$t_{MPWRSTB}$	RESET Minimum Pulse Width	0.21	0.24	0.29	ns
t_{CYC}	Clock Cycle time	3.23	3.68	4.32	ns
F_{MAX}	Maximum Frequency for FIFO	310	272	231	ns

Note: For the derating values at specific junction temperature and voltage supply levels, refer to Table 3-7 on page 3-9.

The rate at which the gate voltage of the external MOSFET slews is determined by the current, I_g, sourced or sunk by the AG pin and the gate-to-source capacitance, C_{GS}, of the external MOSFET. As an approximation, the slew rate is given by EQ 6.

$$dv/dt = I_g / C_{GS}$$

EQ 6

C_{GS} is not a fixed capacitance but, depending on the circuitry connected to its drain terminal, can vary significantly during the course of a turn-on or turn-off transient. Thus, EQ 6 on page 2-91 can only be used for a first-order estimate of the switching speed of the external MOSFET.

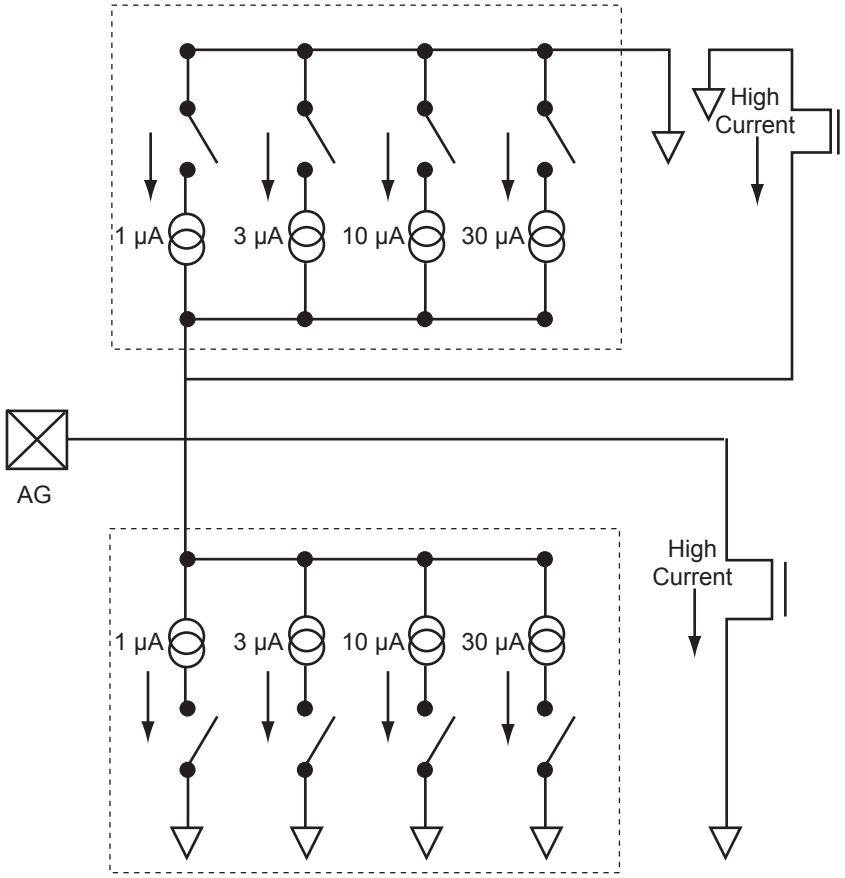


Figure 2-75 • **Gate Driver Example**

Analog-to-Digital Converter Block

At the heart of the Fusion analog system is a programmable Successive Approximation Register (SAR) ADC. The ADC can support 8-, 10-, or 12-bit modes of operation. In 12-bit mode, the ADC can resolve 500 ksps. All results are MSB-justified in the ADC. The input to the ADC is a large 32:1 analog input multiplexer. A simplified block diagram of the Analog Quads, analog input multiplexer, and ADC is shown in Figure 2-79. The ADC offers multiple self-calibrating modes to ensure consistent high performance both at power-up and during runtime.

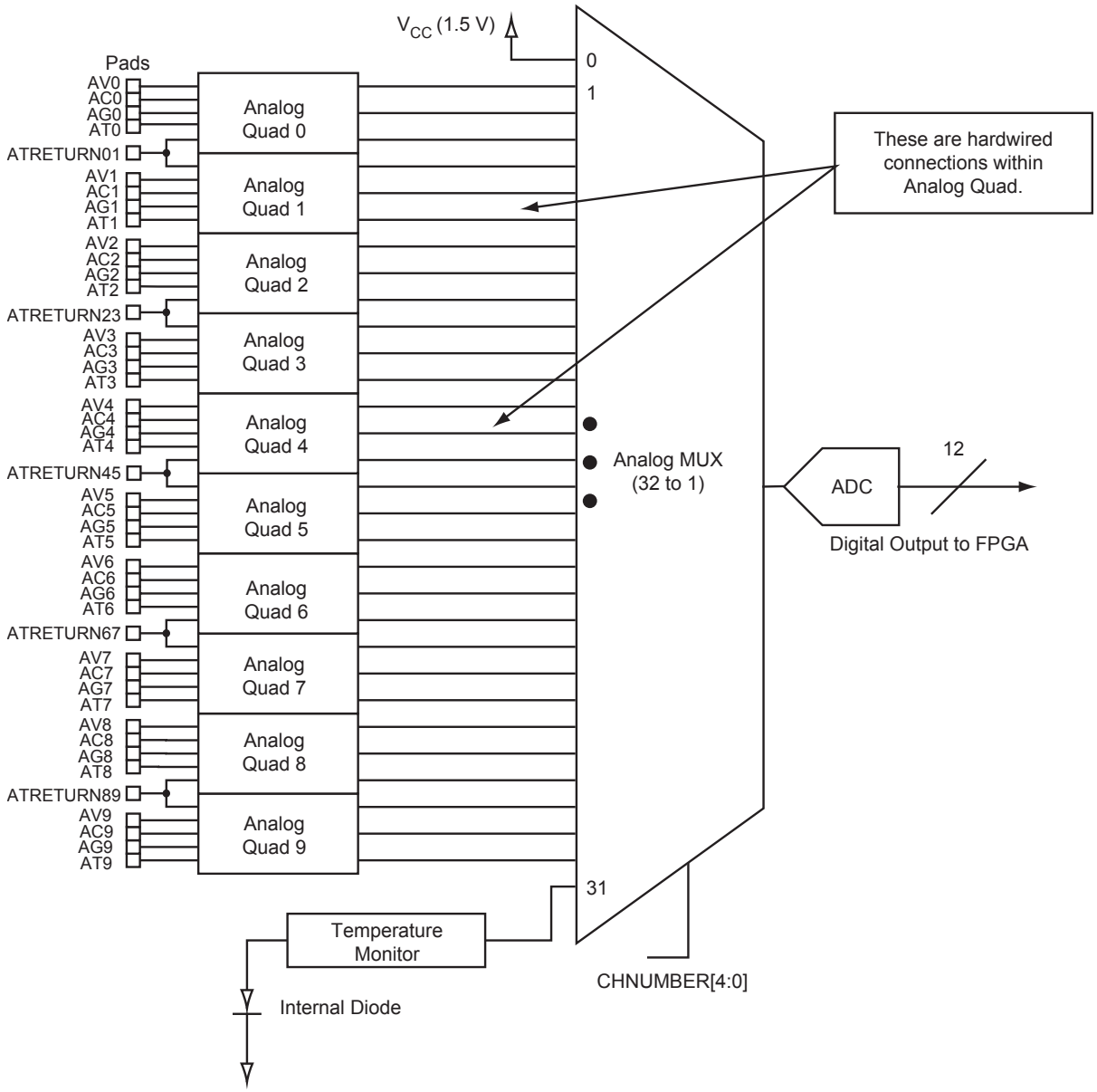


Figure 2-79 • **ADC Block Diagram**

Gain Error

The gain error of an ADC indicates how well the slope of an actual transfer function matches the slope of the ideal transfer function. Gain error is usually expressed in LSB or as a percent of full-scale (%FSR). Gain error is the full-scale error minus the offset error (Figure 2-84).

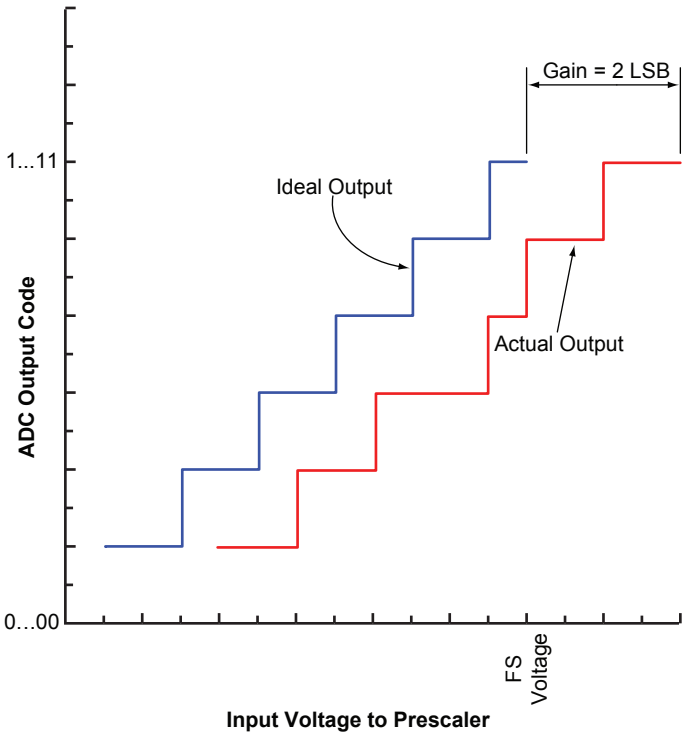


Figure 2-84 • **Gain Error**

Gain Error Drift

Gain-error drift is the variation in gain error due to a change in ambient temperature, typically expressed in ppm/°C.

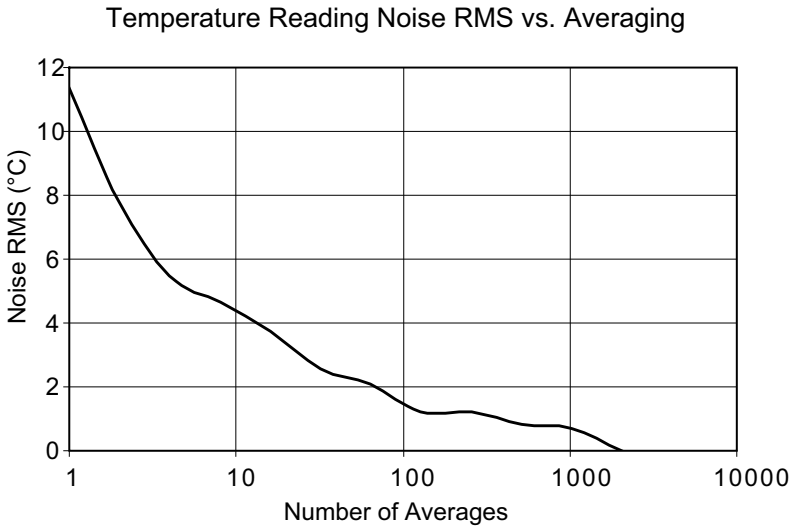


Figure 2-96 • **Temperature Reading Noise When Averaging is Used**

***Table 2-73* • Maximum I/O Frequency for Single-Ended, Voltage-Referenced, and Differential I/Os; All I/O Bank Types (maximum drive strength and high slew selected)**

Specification	Performance Up To
LVTTL/LVCMOS 3.3 V	200 MHz
LVCMOS 2.5 V	250 MHz
LVCMOS 1.8 V	200 MHz
LVCMOS 1.5 V	130 MHz
PCI	200 MHz
PCI-X	200 MHz
HSTL-I	300 MHz
HSTL-II	300 MHz
SSTL2-I	300 MHz
SSTL2-II	300 MHz
SSTL3-I	300 MHz
SSTL3-II	300 MHz
GTL+ 3.3 V	300 MHz
GTL+ 2.5 V	300 MHz
GTL 3.3 V	300 MHz
GTL 2.5 V	300 MHz
LVDS	350 MHz
LVPECL	300 MHz

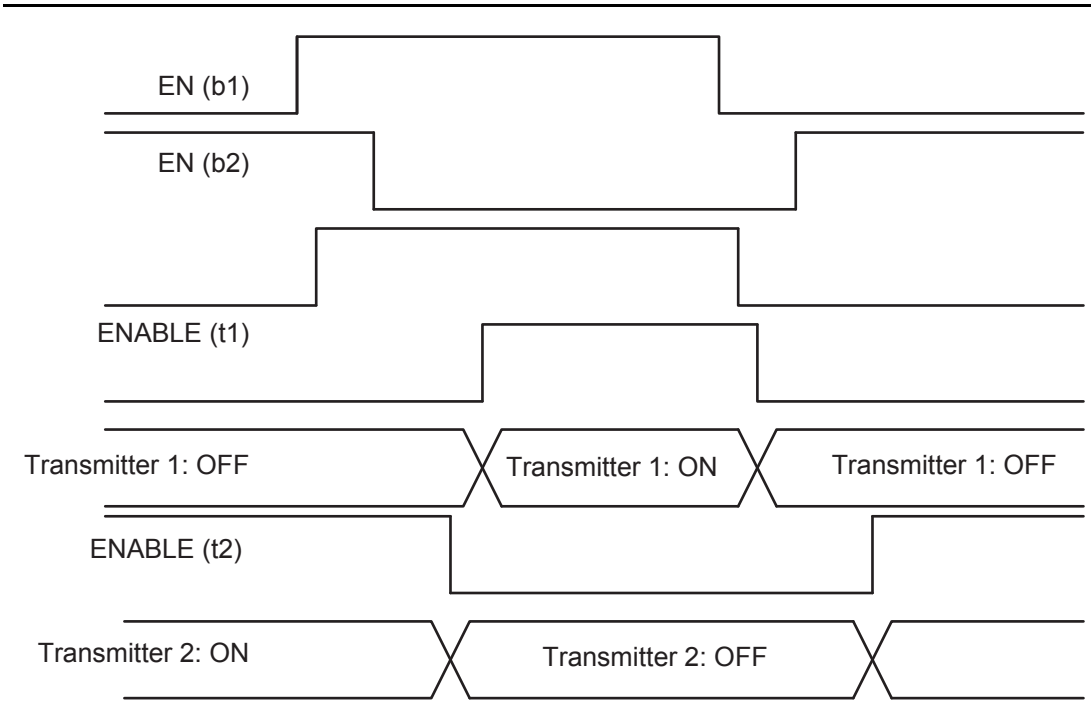


***Figure 2-112* • Timing Diagram (with skew circuit selected)**

Weak Pull-Up and Weak Pull-Down Resistors

Fusion devices support optional weak pull-up and pull-down resistors for each I/O pin. When the I/O is pulled up, it is connected to the VCCI of its corresponding I/O bank. When it is pulled down, it is connected to GND. Refer to Table 2-97 on page 2-171 for more information.

Slew Rate Control and Drive Strength

Fusion devices support output slew rate control: high and low. The high slew rate option is recommended to minimize the propagation delay. This high-speed option may introduce noise into the system if appropriate signal integrity measures are not adopted. Selecting a low slew rate reduces this kind of noise but adds some delays in the system. Low slew rate is recommended when bus transients are expected. Drive strength should also be selected according to the design requirements and noise immunity of the system.

The output slew rate and multiple drive strength controls are available in LVTTL/LVCMOS 3.3 V, LVCMOS 2.5 V, LVCMOS 2.5 V / 5.0 V input, LVCMOS 1.8 V, and LVCMOS 1.5 V. All other I/O standards have a high output slew rate by default.

For Fusion slew rate and drive strength specifications, refer to the appropriate I/O bank table:

- Fusion Standard I/O (Table 2-78 on page 2-152)
- Fusion Advanced I/O (Table 2-79 on page 2-152)
- Fusion Pro I/O (Table 2-80 on page 2-152)

Table 2-83 on page 2-155 lists the default values for the above selectable I/O attributes as well as those that are preset for each I/O standard.

Refer to Table 2-78, Table 2-79, and Table 2-80 on page 2-152 for SLEW and OUT_DRIVE settings. Table 2-81 on page 2-153 and Table 2-82 on page 2-154 list the I/O default attributes. Table 2-83 on page 2-155 lists the voltages for the supported I/O standards.

Table 2-99 • Short Current Event Duration before Failure

Temperature	Time Before Failure
–40°C	>20 years
0°C	>20 years
25°C	>20 years
70°C	5 years
85°C	2 years
100°C	6 months

Table 2-100 • Schmitt Trigger Input Hysteresis
Hysteresis Voltage Value (typ.) for Schmitt Mode Input Buffers

Input Buffer Configuration	Hysteresis Value (typ.)
3.3 V LVTTL/LVCMOS/PCI/PCI-X (Schmitt trigger mode)	240 mV
2.5 V LVCMOS (Schmitt trigger mode)	140 mV
1.8 V LVCMOS (Schmitt trigger mode)	80 mV
1.5 V LVCMOS (Schmitt trigger mode)	60 mV

Table 2-101 • I/O Input Rise Time, Fall Time, and Related I/O Reliability

Input Buffer	Input Rise/Fall Time (min.)	Input Rise/Fall Time (max.)	Reliability
LVTTL/LVCMOS (Schmitt trigger disabled)	No requirement	10 ns*	20 years (100°C)
LVTTL/LVCMOS (Schmitt trigger enabled)	No requirement	No requirement, but input noise voltage cannot exceed Schmitt hysteresis	20 years (100°C)
HSTL/SSTL/GTL	No requirement	10 ns*	10 years (100°C)
LVDS/BLVDS/M-LVDS/LVPECL	No requirement	10 ns*	10 years (100°C)

*Note: * The maximum input rise/fall time is related only to the noise induced into the input buffer trace. If the noise is low, the rise time and fall time of input buffers, when Schmitt trigger is disabled, can be increased beyond the maximum value. The longer the rise/fall times, the more susceptible the input signal is to the board noise. Microsemi recommends signal integrity evaluation/characterization of the system to ensure there is no excessive noise coupling into input signals.*

Table 2-103 • AC Waveforms, Measuring Points, and Capacitive Loads

Input Low (V)	Input High (V)	Measuring Point* (V)	VREF (typ.) (V)	C_{LOAD} (pF)
0	3.3	1.4	–	35

*Note: *Measuring point = Vtrip. See Table 2-90 on page 2-166 for a complete table of trip points.*

Timing Characteristics

Table 2-104 • 3.3 V LVTTL / 3.3 V LVCMOS Low Slew
Commercial Temperature Range Conditions: T_J = 70°C, Worst-Case VCC = 1.425 V,
Worst-Case VCCI = 3.0 V
Applicable to Pro I/Os

Drive Strength	Speed Grade	t_{DOUT}	t_{DP}	t_{DIN}	t_{PY}	t_{PYS}	t_{EOUT}	t_{ZL}	t_{ZH}	t_{LZ}	t_{HZ}	t_{ZLS}	t_{ZHS}	Units
4 mA	Std.	0.66	11.01	0.04	1.20	1.57	0.43	11.21	9.05	2.69	2.44	13.45	11.29	ns
	–1	0.56	9.36	0.04	1.02	1.33	0.36	9.54	7.70	2.29	2.08	11.44	9.60	ns
	–2	0.49	8.22	0.03	0.90	1.17	0.32	8.37	6.76	2.01	1.82	10.04	8.43	ns
8 mA	Std.	0.66	7.86	0.04	1.20	1.57	0.43	8.01	6.44	3.04	3.06	10.24	8.68	ns
	–1	0.56	6.69	0.04	1.02	1.33	0.36	6.81	5.48	2.58	2.61	8.71	7.38	ns
	–2	0.49	5.87	0.03	0.90	1.17	0.32	5.98	4.81	2.27	2.29	7.65	6.48	ns
12 mA	Std.	0.66	6.03	0.04	1.20	1.57	0.43	6.14	5.02	3.28	3.47	8.37	7.26	ns
	–1	0.56	5.13	0.04	1.02	1.33	0.36	5.22	4.27	2.79	2.95	7.12	6.17	ns
	–2	0.49	4.50	0.03	0.90	1.17	0.32	4.58	3.75	2.45	2.59	6.25	5.42	ns
16 mA	Std.	0.66	5.62	0.04	1.20	1.57	0.43	5.72	4.72	3.32	3.58	7.96	6.96	ns
	–1	0.56	4.78	0.04	1.02	1.33	0.36	4.87	4.02	2.83	3.04	6.77	5.92	ns
	–2	0.49	4.20	0.03	0.90	1.17	0.32	4.27	3.53	2.48	2.67	5.94	5.20	ns
24 mA	Std.	0.66	5.24	0.04	1.20	1.57	0.43	5.34	4.69	3.39	3.96	7.58	6.93	ns
	–1	0.56	4.46	0.04	1.02	1.33	0.36	4.54	3.99	2.88	3.37	6.44	5.89	ns
	–2	0.49	3.92	0.03	0.90	1.17	0.32	3.99	3.50	2.53	2.96	5.66	5.17	ns

Note: For the derating values at specific junction temperature and voltage supply levels, refer to Table 3-7 on page 3-9.

Differential I/O Characteristics

Configuration of the I/O modules as a differential pair is handled by the Microsemi Designer software when the user instantiates a differential I/O macro in the design.

Differential I/Os can also be used in conjunction with the embedded Input Register (InReg), Output Register (OutReg), Enable Register (EnReg), and Double Data Rate (DDR). However, there is no support for bidirectional I/Os or tristates with these standards.

LVDS

Low-Voltage Differential Signal (ANSI/TIA/EIA-644) is a high-speed differential I/O standard. It requires that one data bit be carried through two signal lines, so two pins are needed. It also requires external resistor termination.

The full implementation of the LVDS transmitter and receiver is shown in an example in Figure 2-134. The building blocks of the LVDS transmitter–receiver are one transmitter macro, one receiver macro, three board resistors at the transmitter end, and one resistor at the receiver end. The values for the three driver resistors are different from those used in the LVPECL implementation because the output standard specifications are different.

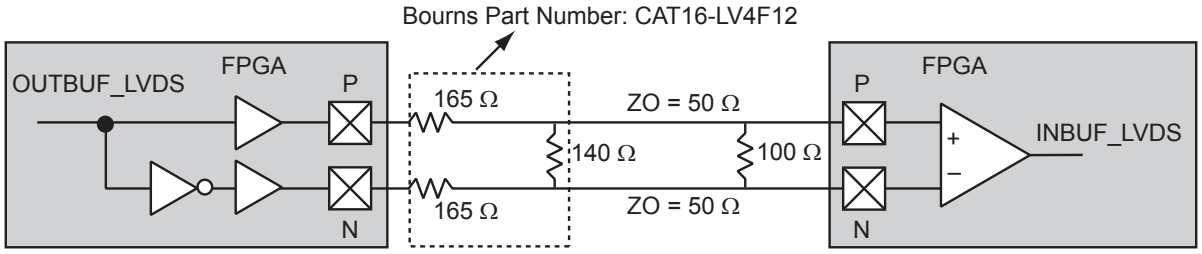


***Figure 2-134* • LVDS Circuit Diagram and Board-Level Implementation**

***Table 2-168* • Minimum and Maximum DC Input and Output Levels**

DC Parameter	Description	Min.	Typ.	Max.	Units
VCCI	Supply Voltage	2.375	2.5	2.625	V
VOL	Output Low Voltage	0.9	1.075	1.25	V
VOH	Input High Voltage	1.25	1.425	1.6	V
IOL [1]	Output Low Voltage	0.65	0.91	1.16	mA
IOH [1]	Output High Voltage	0.65	0.91	1.16	mA
VI	Input Voltage	0		2.925	V
IIL [2,3]	Input Low Voltage			10	µA
IIH [2,4]	Input High Voltage			10	µA
VODIFF	Differential Output Voltage	250	350	450	mV
VOCM	Output Common Mode Voltage	1.125	1.25	1.375	V
VICM	Input Common Mode Voltage	0.05	1.25	2.35	V
VIDIFF	Input Differential Voltage	100	350		mV

Notes:

1. *IOL/IOH defined by VODIFF/(Resistor Network)*
2. *Currents are measured at 85°C junction temperature.*
3. *IIL is the input leakage current per I/O pin over recommended operation conditions where –0.3 V < VIN < VIL.*
4. *IIH is the input leakage current per I/O pin over recommended operating conditions VIH < VIN < VCCI. Input current is larger when operating outside recommended ranges.*

FG676	
Pin Number	AFS1500 Function
G13	IO22NDB1V0
G14	IO22PDB1V0
G15	GND
G16	IO32PPB1V1
G17	IO36NPB1V2
G18	VCCIB1
G19	GND
G20	IO47NPB2V0
G21	IO49PDB2V0
G22	VCCIB2
G23	IO46NDB2V0
G24	GBC2/IO46PDB2V0
G25	IO48NPB2V0
G26	NC
H1	GND
H2	NC
H3	IO118NDB4V0
H4	IO118PDB4V0
H5	IO119NPB4V0
H6	IO124NDB4V0
H7	GND
H8	VCOMPLA
H9	VCCPLA
H10	VCCIB0
H11	IO12NDB0V1
H12	IO12PDB0V1
H13	VCCIB0
H14	VCCIB1
H15	IO30NDB1V1
H16	IO30PDB1V1
H17	VCCIB1
H18	IO36PPB1V2
H19	IO38NPB1V2
H20	GND
H21	IO49NDB2V0
H22	IO50PDB2V0

FG676	
Pin Number	AFS1500 Function
H23	IO50NDB2V0
H24	IO51PDB2V0
H25	NC
H26	GND
J1	NC
J2	VCCIB4
J3	IO115PDB4V0
J4	GND
J5	IO116NDB4V0
J6	IO116PDB4V0
J7	VCCIB4
J8	IO117PDB4V0
J9	VCCIB4
J10	GND
J11	IO06NDB0V1
J12	IO06PDB0V1
J13	IO16NDB0V2
J14	IO16PDB0V2
J15	IO28NDB1V1
J16	IO28PDB1V1
J17	GND
J18	IO38PPB1V2
J19	IO53PDB2V0
J20	VCCIB2
J21	IO52PDB2V0
J22	IO52NDB2V0
J23	GND
J24	IO51NDB2V0
J25	VCCIB2
J26	NC
K1	NC
K2	NC
K3	IO115NDB4V0
K4	IO113PDB4V0
K5	VCCIB4
K6	IO114NDB4V0

FG676	
Pin Number	AFS1500 Function
K7	IO114PDB4V0
K8	IO117NDB4V0
K9	GND
K10	VCC
K11	VCCIB0
K12	GND
K13	VCCIB0
K14	VCCIB1
K15	GND
K16	VCCIB1
K17	GND
K18	GND
K19	IO53NDB2V0
K20	IO57PDB2V0
K21	GCA2/IO59PDB2V0
K22	VCCIB2
K23	IO54NDB2V0
K24	IO54PDB2V0
K25	NC
K26	NC
L1	GND
L2	NC
L3	IO112PPB4V0
L4	IO113NDB4V0
L5	GFB2/IO109PDB4V0
L6	GFA2/IO110PDB4V0
L7	IO112NPB4V0
L8	IO104PDB4V0
L9	IO111PDB4V0
L10	VCCIB4
L11	GND
L12	VCC
L13	GND
L14	VCC
L15	GND
L16	VCC

FG676	
Pin Number	**AFS1500 Function**
L17	VCCIB2
L18	GCB2/IO60PDB2V0
L19	IO58NDB2V0
L20	IO57NDB2V0
L21	IO59NDB2V0
L22	GCC2/IO61PDB2V0
L23	IO55PPB2V0
L24	IO56PDB2V0
L25	IO55NPB2V0
L26	GND
M1	NC
M2	VCCIB4
M3	GFC2/IO108PDB4V0
M4	GND
M5	IO109NDB4V0
M6	IO110NDB4V0
M7	GND
M8	IO104NDB4V0
M9	IO111NDB4V0
M10	GND
M11	V**CC**
M12	GND
M13	VCC
M14	GND
M15	VCC
M16	GND
M17	GND
M18	IO60NDB2V0
M19	IO58PDB2V0
M20	GND
M21	IO68NPB2V0
M22	IO61NDB2V0
M23	GND
M24	IO56NDB2V0
M25	VCCIB2
M26	IO65PDB2V0

FG676	
Pin Number	**AFS1500 Function**
N1	NC
N2	NC
N3	IO108NDB4V0
N4	VCCOSC
N5	VCCIB4
N6	XTAL2
N7	GFC1/IO107PDB4V0
N8	VCCIB4
N9	GFB1/IO106PDB4V0
N10	VCCIB4
N11	GND
N12	VCC
N13	GND
N14	VCC
N15	GND
N16	VCC
N17	VCCIB2
N18	IO70PDB2V0
N19	VCCIB2
N20	IO69PDB2V0
N21	GCA1/IO64PDB2V0
N22	VCCIB2
N23	GCC0/IO62NDB2V0
N24	GCC1/IO62PDB2V0
N25	IO66PDB2V0
N26	IO65NDB2V0
P1	NC
P2	NC
P3	IO103PDB4V0
P4	XTAL1
P5	VCCIB4
P6	GNDOSC
P7	GFC0/IO107NDB4V0
P8	VCCIB4
P9	GFB0/IO106NDB4V0
P10	VCCIB4

FG676	
Pin Number	**AFS1500 Function**
P11	VCC
P12	GND
P13	VCC
P14	GND
P15	VCC
P16	GND
P17	VCCIB2
P18	IO70NDB2V0
P19	VCCIB2
P20	IO69NDB2V0
P21	GCA0/IO64NDB2V0
P22	VCCIB2
P23	GCB0/IO63NDB2V0
P24	GCB1/IO63PDB2V0
P25	IO66NDB2V0
P26	IO67PDB2V0
R1	NC
R2	VCCIB4
R3	IO103NDB4V0
R4	GND
R5	IO101PDB4V0
R6	IO100NPB4V0
R7	GND
R8	IO99PDB4V0
R9	IO97PDB4V0
R10	GND
R11	GND
R12	VCC
R13	GND
R14	VCC
R15	GND
R16	VCC
R17	GND
R18	GDB2/IO83PDB2V0
R19	IO78PDB2V0
R20	GND

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