Welcome to E-XFL.COM

Understanding Embedded - FPGAs (Field Programmable Gate Array)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details	
Product Status	Active
Number of LABs/CLBs	-
Number of Logic Elements/Cells	-
Total RAM Bits	36864
Number of I/O	114
Number of Gates	250000
Voltage - Supply	1.425V ~ 1.575V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 85°C (TJ)
Package / Case	256-LBGA
Supplier Device Package	256-FPBGA (17x17)
Purchase URL	https://www.e-xfl.com/product-detail/microchip-technology/afs250-2fgg256

Table 2-7 • AFS250 Global Resource Timing
Commercial Temperature Range Conditions: T_J = 70°C, Worst-Case VCC = 1.425 V

Parameter	Description	–2		–1		Std.		Units
		Min.[1]	Max.[2]	Min.[1]	Max.[2]	Min.[1]	Max.[2]	
t_{RCKL}	Input Low Delay for Global Clock	0.89	1.12	1.02	1.27	1.20	1.50	ns
t_{RCKH}	Input High Delay for Global Clock	0.88	1.14	1.00	1.30	1.17	1.53	ns
$t_{RCKMPWH}$	Minimum Pulse Width High for Global Clock							ns
$t_{RCKMPWL}$	Minimum Pulse Width Low for Global Clock							ns
t_{RCKSW}	Maximum Skew for Global Clock		0.26		0.30		0.35	ns

Notes:

1. *Value reflects minimum load. The delay is measured from the CCC output to the clock pin of a sequential element located in a lightly loaded row (single element is connected to the global net).*
2. *Value reflects maximum load. The delay is measured on the clock pin of the farthest sequential element located in a fully loaded row (all available flip-flops are connected to the global net in the row).*
3. *For the derating values at specific junction temperature and voltage supply levels, refer to Table 3-7 on page 3-9.*

Table 2-8 • AFS090 Global Resource Timing
Commercial Temperature Range Conditions: T_J = 70°C, Worst-Case VCC = 1.425 V

Parameter	Description	–2		–1		Std.		Units
		Min.[1]	Max.[2]	Min.[1]	Max.[2]	Min.[1]	Max.[2]	
t_{RCKL}	Input Low Delay for Global Clock	0.84	1.07	0.96	1.21	1.13	1.43	ns
t_{RCKH}	Input High Delay for Global Clock	0.83	1.10	0.95	1.25	1.12	1.47	ns
$t_{RCKMPWH}$	Minimum Pulse Width High for Global Clock							ns
$t_{RCKMPWL}$	Minimum Pulse Width Low for Global Clock							ns
t_{RCKSW}	Maximum Skew for Global Clock		0.27		0.30		0.36	ns

Notes:

1. *Value reflects minimum load. The delay is measured from the CCC output to the clock pin of a sequential element located in a lightly loaded row (single element is connected to the global net).*
2. *Value reflects maximum load. The delay is measured on the clock pin of the farthest sequential element located in a fully loaded row (all available flip-flops are connected to the global net in the row).*
3. *For the derating values at specific junction temperature and voltage supply levels, refer to Table 3-7 on page 3-9.*

PLL Macro

The PLL functionality of the clock conditioning block is supported by the PLL macro. Note that the PLL macro reference clock uses the CLKA input of the CCC block, which is only accessible from the global A[2:0] package pins. Refer to Figure 2-22 on page 2-25 for more information.

The PLL macro provides five derived clocks (three independent) from a single reference clock. The PLL feedback loop can be driven either internally or externally. The PLL macro also provides power-down input and lock output signals. During power-up, POWERDOWN should be asserted Low until VCC is up. See Figure 2-19 on page 2-23 for more information.

Inputs:

- CLKA: selected clock input
- POWERDOWN (active low): disables PLLs. The default state is power-down on (active low).

Outputs:

- LOCK (active high): indicates that PLL output has locked on the input reference signal
- GLA, GLB, GLC: outputs to respective global networks
- YB, YC: allows output from the CCC to be routed back to the FPGA core

As previously described, the PLL allows up to five flexible and independently configurable clock outputs. Figure 2-23 on page 2-26 illustrates the various clock output options and delay elements.

As illustrated, the PLL supports three distinct output frequencies from a given input clock. Two of these (GLB and GLC) can be routed to the B and C global networks, respectively, and/or routed to the device core (YB and YC).

There are five delay elements to support phase control on all five outputs (GLA, GLB, GLC, YB, and YC).

There is also a delay element in the feedback loop that can be used to advance the clock relative to the reference clock.

The PLL macro reference clock can be driven by an INBUF macro to create a composite macro, where the I/O macro drives the global buffer (with programmable delay) using a hardwired connection. In this case, the I/O must be placed in one of the dedicated global I/O locations.

The PLL macro reference clock can be driven directly from the FPGA core.

The PLL macro reference clock can also be driven from an I/O routed through the FPGA regular routing fabric. In this case, users must instantiate a special macro, PLLINT, to differentiate it from the hardwired I/O connection described earlier.

The visual PLL configuration in SmartGen, available with the Libero SoC and Designer tools, will derive the necessary internal divider ratios based on the input frequency and desired output frequencies selected by the user. SmartGen allows the user to select the various delays and phase shift values necessary to adjust the phases between the reference clock (CLKA) and the derived clocks (GLA, GLB, GLC, YB, and YC). SmartGen also allows the user to select where the input clock is coming from. SmartGen automatically instantiates the special macro, PLLINT, when needed.

Flash Memory Block Addressing

Figure 2-34 shows a graphical representation of the flash memory block.

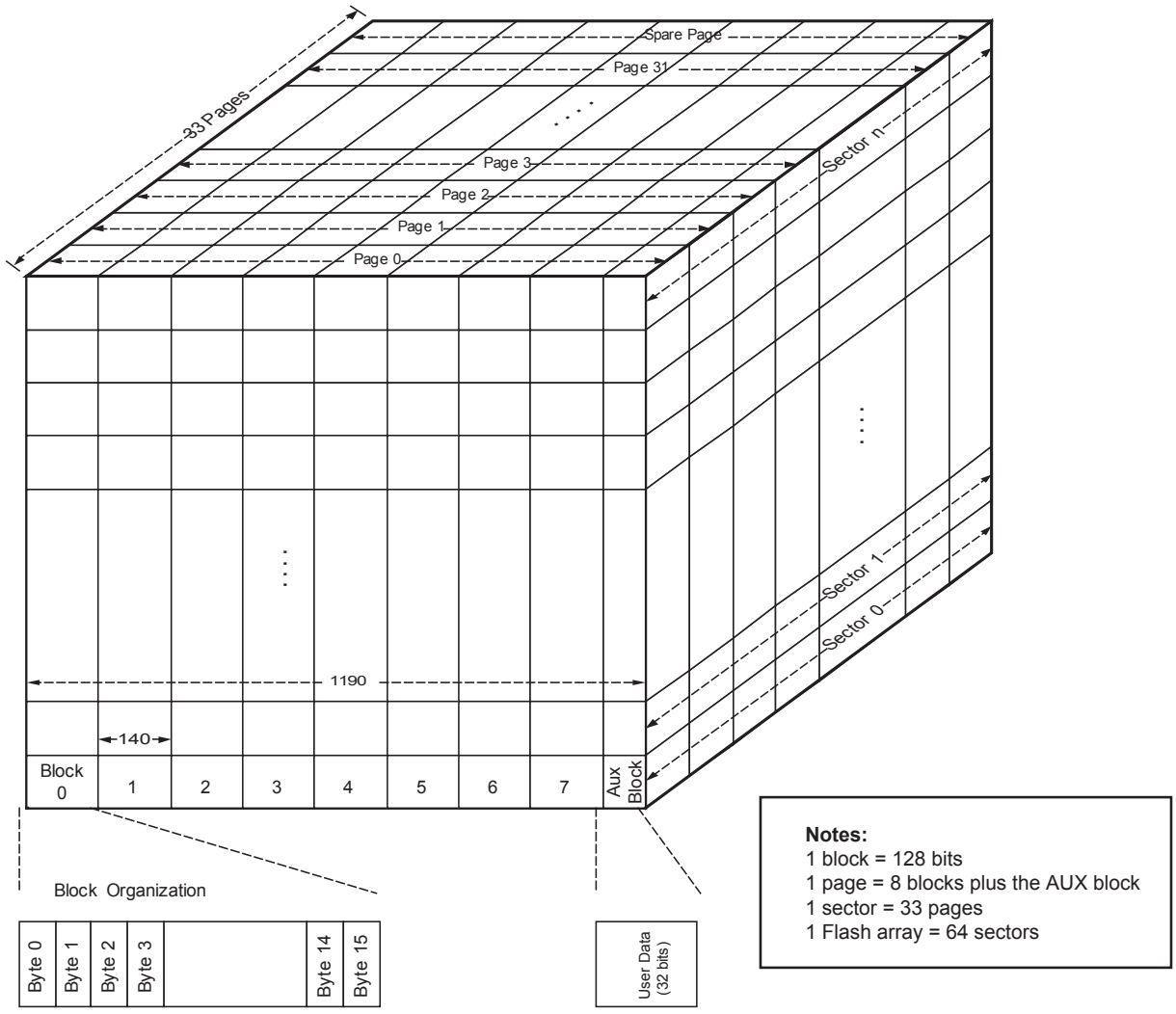

Figure 2-34 • **Flash Memory Block Organization**

Each FB is partitioned into sectors, pages, blocks, and bytes. There are 64 sectors in an FB, and each sector contains 32 pages and 1 spare page. Each page contains 8 data blocks and 1 auxiliary block. Each data block contains 16 bytes of user data, and the auxiliary block contains 4 bytes of user data.

Addressing for the FB is shown in Table 2-20.

Table 2-20 • **FB Address Bit Allocation ADDR[17:0]**

17	12	11	7	6	4	3	0
Sector		Page		Block		Byte	

When the spare page of a sector is addressed (SPAREPAGE active), ADDR[11:7] are ignored.

When the Auxiliary block is addressed (AUXBLOCK active), ADDR[6:2] are ignored.

Note: The spare page of sector 0 is unavailable for any user data. Writes to this page will return an error, and reads will return all zeroes.

Data operations are performed in widths of 1 to 4 bytes. A write to a location in a page that is not already in the Page Buffer will cause the page to be read from the FB Array and stored in the Page Buffer. The block that was addressed during the write will be put into the Block Buffer, and the data written by WD will overwrite the data in the Block Buffer. After the data is written to the Block Buffer, the Block Buffer is then written to the Page Buffer to keep both buffers in sync. Subsequent writes to the same block will overwrite the Block Buffer and the Page Buffer. A write to another block in the page will cause the addressed block to be loaded from the Page Buffer, and the write will be performed as described previously.

The data width can be selected dynamically via the DATAWIDTH input bus. The truth table for the data width settings is detailed in Table 2-21. The minimum resolvable address is one 8-bit byte. For data widths greater than 8 bits, the corresponding address bits are ignored—when DATAWIDTH = 0 (2 bytes), ADDR[0] is ignored, and when DATAWIDTH = '10' or '11' (4 bytes), ADDR[1:0] are ignored. Data pins are LSB-oriented and unused WD data pins must be grounded.

Table 2-21 • **Data Width Settings**

DATAWIDTH[1:0]	Data Width
00	1 byte [7:0]
01	2 byte [15:0]
10, 11	4 bytes [31:0]

Flash Memory Block Protection

Page Loss Protection

When the PAGELOSSPROTECT pin is set to logic 1, it prevents writes to any page other than the current page in the Page Buffer until the page is either discarded or programmed.

A write to another page while the current page is Page Loss Protected will return a STATUS of '11'.

Overwrite Protection

Any page that is Overwrite Protected will result in the STATUS being set to '01' when an attempt is made to either write, program, or erase it. To set the Overwrite Protection state for a page, set the OVERWRITEPROTECT pin when a Program operation is undertaken. To clear the Overwrite Protect state for a given page, an Unprotect Page operation must be performed on the page, and then the page must be programmed with the OVERWRITEPROTECT pin cleared to save the new page.

LOCKREQUEST

The LOCKREQUEST signal is used to give the user interface control over simultaneous access of the FB from both the User and JTAG interfaces. When LOCKREQUEST is asserted, the JTAG interface will hold off any access attempts until LOCKREQUEST is deasserted.

Flash Memory Block Operations

FB Operation Priority

The FB provides for priority of operations when multiple actions are requested simultaneously. Table 2-22 shows the priority order (priority 0 is the highest).

Table 2-22 • **FB Operation Priority**

Operation	Priority
System Initialization	0
FB Reset	1
Read	2
Write	3
Erase Page	4
Program	5
Unprotect Page	6
Discard Page	7

Conversely, when writing 4-bit values and reading 9-bit values, the ninth bit of a read operation will be undefined. The RAM blocks employ little-endian byte order for read and write operations.

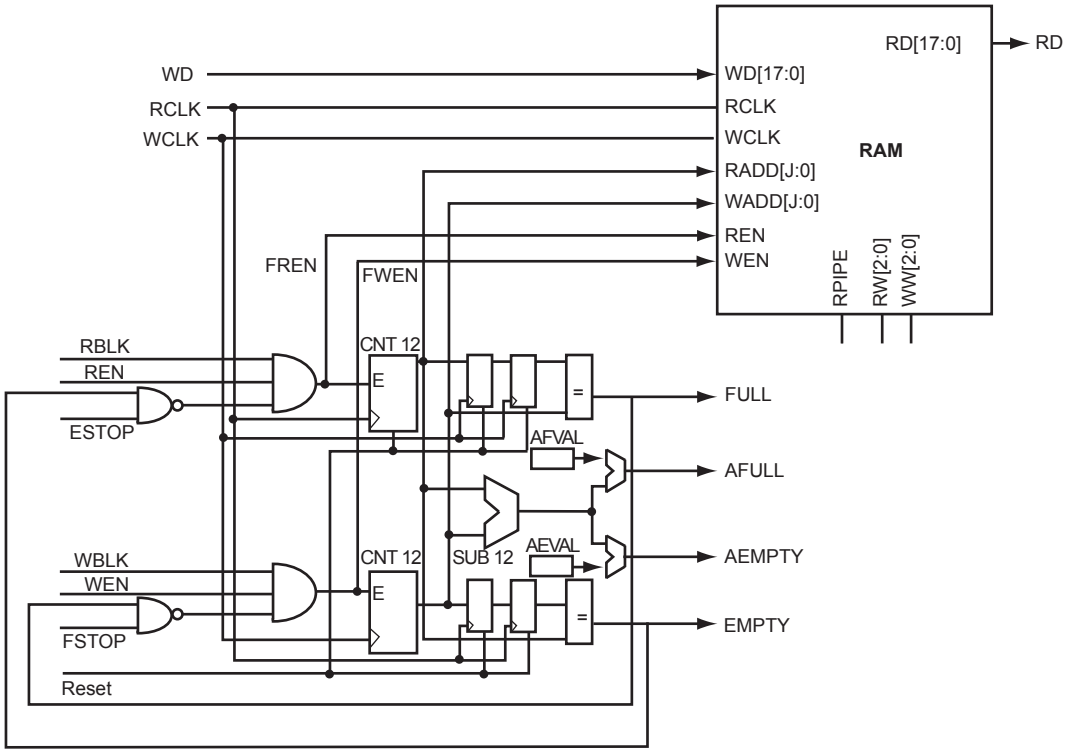

Figure 2-47 • **Fusion RAM Block with Embedded FIFO Controller**

Table 2-36 • **Analog Block Pin Description (continued)**

Signal Name	Number of Bits	Direction	Function	Location of Details
AG6	1	Output		Analog Quad
AT6	1	Input		Analog Quad
ATRETURN67	1	Input	Temperature monitor return shared by Analog Quads 6 and 7	Analog Quad
AV7	1	Input	Analog Quad 7	Analog Quad
AC7	1	Input		Analog Quad
AG7	1	Output		Analog Quad
AT7	1	Input		Analog Quad
AV8	1	Input	Analog Quad 8	Analog Quad
AC8	1	Input		Analog Quad
AG8	1	Output		Analog Quad
AT8	1	Input		Analog Quad
ATRETURN89	1	Input	Temperature monitor return shared by Analog Quads 8 and 9	Analog Quad
AV9	1	Input	Analog Quad 9	Analog Quad
AC9	1	Input		Analog Quad
AG9	1	Output		Analog Quad
AT9	1	Input		Analog Quad
RTCMATCH	1	Output	MATCH	RTC
RTCPSMMATCH	1	Output	MATCH connected to VRPSM	RTC
RTCXTLMODE[1:0]	2	Output	Drives XTLOSC RTCMODE[1:0] pins	RTC
RTCXTLSEL	1	Output	Drives XTLOSC MODESEL pin	RTC
RTCCLK	1	Input	RTC clock input	RTC

Analog Quad

With the Fusion family, Microsemi introduces the Analog Quad, shown in Figure 2-65 on page 2-81, as the basic analog I/O structure. The Analog Quad is a four-channel system used to precondition a set of analog signals before sending it to the ADC for conversion into a digital signal. To maximize the usefulness of the Analog Quad, the analog input signals can also be configured as LVTTL digital input signals. The Analog Quad is divided into four sections.

The first section is called the Voltage Monitor Block, and its input pin is named AV. It contains a two-channel analog multiplexer that allows an incoming analog signal to be routed directly to the ADC or allows the signal to be routed to a prescaler circuit before being sent to the ADC. The prescaler can be configured to accept analog signals between –12 V and 0 or between 0 and +12 V. The prescaler circuit scales the voltage applied to the ADC input pad such that it is compatible with the ADC input voltage range. The AV pin can also be used as a digital input pin.

The second section of the Analog Quad is called the Current Monitor Block. Its input pin is named AC. The Current Monitor Block contains all the same functions as the Voltage Monitor Block with one addition, which is a current monitoring function. A small external current sensing resistor (typically less than 1 Ω) is connected between the AV and AC pins and is in series with a power source. The Current Monitor Block contains a current monitor circuit that converts the current through the external resistor to a voltage that can then be read using the ADC.

Gate Driver

The Fusion Analog Quad includes a Gate Driver connected to the Quad's AG pin (Figure 2-74). Designed to work with external p- or n-channel MOSFETs, the Gate driver is a configurable current sink or source and requires an external pull-up or pull-down resistor. The AG supports 4 selectable gate drive levels: 1 µA, 3 µA, 10 µA, and 30 µA (Figure 2-75 on page 2-91). The AG also supports a High Current Drive mode in which it can sink 20 mA; in this mode the switching rate is approximately 1.3 MHz with 100 ns turn-on time and 600 ns turn-off time. Modeled on an open-drain-style output, it does not output a voltage level without an appropriate pull-up or pull-down resistor. If 1 V is forced on the drain, the current sinking/sourcing will exceed the ability of the transistor, and the device could be damaged.

The AG pad is turned on via the corresponding GDON*x* pin in the Analog Block macro, where *x* is the number of the corresponding Analog Quad for the AG pad to be enabled (GDON0 to GDON9).

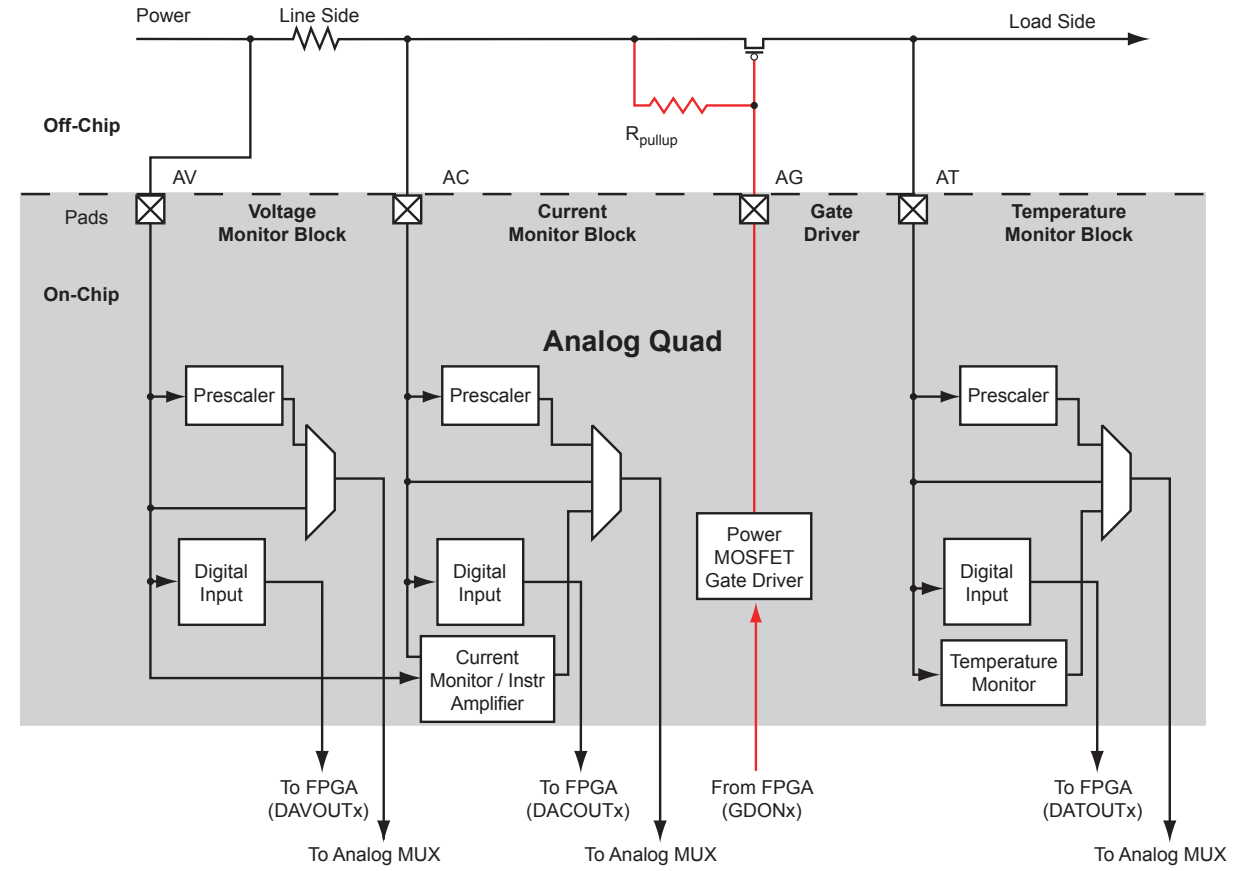

Figure 2-74 • **Gate Driver**

The gate-to-source voltage (V_{gs}) of the external MOSFET is limited to the programmable drive current times the external pull-up or pull-down resistor value (EQ 5).

$$V_{gs} \le I_g \times (R_{pullup} \text{ or } R_{pulldown})$$

EQ 5

Offset Error

Offset error indicates how well the actual transfer function matches the ideal transfer function at a single point. For an ideal ADC, the first transition occurs at 0.5 LSB above zero. The offset voltage is measured by applying an analog input such that the ADC outputs all zeroes and increases until the first transition occurs (Figure 2-86).

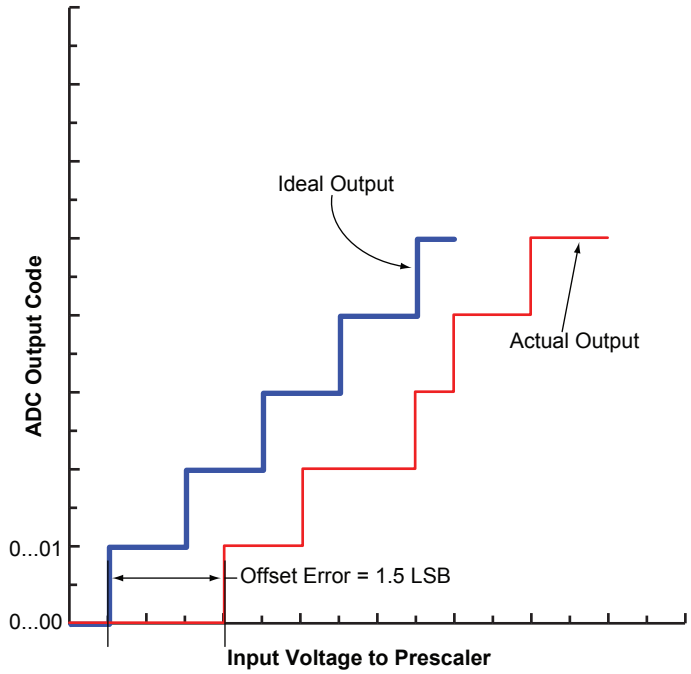

Figure 2-86 • **Offset Error**

Resolution

ADC resolution is the number of bits used to represent an analog input signal. To more accurately replicate the analog signal, resolution needs to be increased.

Sampling Rate

Sampling rate or sample frequency, specified in samples per second (sps), is the rate at which an ADC acquires (samples) the analog input.

SNR – Signal-to-Noise Ratio

SNR is the ratio of the amplitude of the desired signal to the amplitude of the noise signals at a given point in time. For a waveform perfectly reconstructed from digital samples, the theoretical maximum SNR (EQ 14) is the ratio of the full-scale analog input (RMS value) to the RMS quantization error (residual error). The ideal, theoretical minimum ADC noise is caused by quantization error only and results directly from the ADC's resolution (N bits):

$$SNR_{dB[MAX]} = 6.02_{dB} \times N + 1.76_{dB}$$

EQ 14

SINAD – Signal-to-Noise and Distortion

SINAD is the ratio of the rms amplitude to the mean value of the root-sum-square of the all other spectral components, including harmonics, but excluding DC. SINAD is a good indication of the overall dynamic performance of an ADC because it includes all components which make up noise and distortion.

Total Harmonic Distortion

THD measures the distortion content of a signal, and is specified in decibels relative to the carrier (dBc). THD is the ratio of the RMS sum of the selected harmonics of the input signal to the fundamental itself. Only harmonics within the Nyquist limit are included in the measurement.

***Table 2-68* • I/O Bank Support by Device**

I/O Bank	AFS090	AFS250	AFS600	AFS1500
Standard I/O	N	N	–	–
Advanced I/O	E, W	E, W	E, W	E, W
Pro I/O	–	–	N	N
Analog Quad	S	S	S	S

Note: E = East side of the device
W = West side of the device
N = North side of the device
S = South side of the device

***Table 2-69* • Fusion VCCI Voltages and Compatible Standards**

VCCI (typical)	Compatible Standards
3.3 V	LVTTL/LVCMOS 3.3, PCI 3.3, SSTL3 (Class I and II),* GTL+ 3.3, GTL 3.3,* LVPECL
2.5 V	LVCMOS 2.5, LVCMOS 2.5/5.0, SSTL2 (Class I and II),* GTL+ 2.5,* GTL 2.5,* LVDS, BLVDS, M-LVDS
1.8 V	LVCMOS 1.8
1.5 V	LVCMOS 1.5, HSTL (Class I),* HSTL (Class II)*

*Note: *I/O standard supported by Pro I/O banks.*

Table 2-70* • Fusion VREF Voltages and Compatible Standards

VREF (typical)	Compatible Standards
1.5 V	SSTL3 (Class I and II)
1.25 V	SSTL2 (Class I and II)
1.0 V	GTL+ 2.5, GTL+ 3.3
0.8 V	GTL 2.5, GTL 3.3
0.75 V	HSTL (Class I), HSTL (Class II)

*Note: *I/O standards supported by Pro I/O banks.*

***Table 2-77* • Comparison Table for 5 V–Compliant Receiver Scheme**

Scheme	Board Components	Speed	Current Limitations
1	Two resistors	Low to high[1]	Limited by transmitter's drive strength
2	Resistor and Zener 3.3 V	Medium	Limited by transmitter's drive strength
3	Bus switch	High	N/A
4	Minimum resistor value[2] R = 47 Ω at T_J = 70°C R = 150 Ω at T_J = 85°C R = 420 Ω at T_J = 100°C	Medium	Maximum diode current at 100% duty cycle, signal constantly at '1' 52.7 mA at T_J =70°C / 10-year lifetime 16.5 mA at T_J = 85°C / 10-year lifetime 5.9 mA at T_J = 100°C / 10-year lifetime For duty cycles other than 100%, the currents can be increased by a factor = 1 / (duty cycle). Example: 20% duty cycle at 70°C Maximum current = (1 / 0.2) * 52.7 mA = 5 * 52.7 mA = 263.5 mA

Notes:

1. *Speed and current consumption increase as the board resistance values decrease.*
2. *Resistor values ensure I/O diode long-term reliability.*

***Table 2-82* • Advanced I/O Default Attributes**

I/O Standards	SLEW (output only)	OUT_DRIVE (output only)	SKEW (tribuf and bibuf only)	RES_PULL	OUT_LOAD (output only)	COMBINE_REGISTER
LVTTL/LVCMOS 3.3 V	Refer to the following tables for more information: Table 2-78 on page 2-152 Table 2-79 on page 2-152 Table 2-80 on page 2-152	Refer to the following tables for more information: Table 2-78 on page 2-152 Table 2-79 on page 2-152 Table 2-80 on page 2-152	Off	None	35 pF	–
LVCMOS 2.5 V			Off	None	35 pF	–
LVCMOS 2.5/5.0 V			Off	None	35 pF	–
LVCMOS 1.8 V			Off	None	35 pF	–
LVCMOS 1.5 V			Off	None	35 pF	–
PCI (3.3 V)			Off	None	10 pF	–
PCI-X (3.3 V)			Off	None	10 pF	–
LVDS, BLVDS, M-LVDS			Off	None	–	–
LVPECL			Off	None	–	–

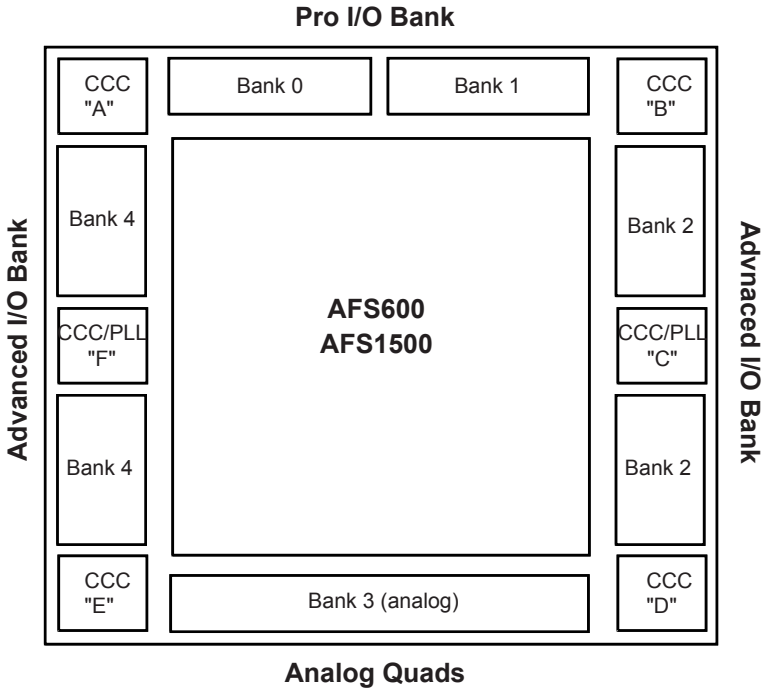

Figure 2-114 • **Naming Conventions of Fusion Devices with Four I/O Banks**

***Table 2-98* • I/O Short Currents IOSH/IOSL (continued)**

	Drive Strength	IOSH (mA)*	IOSL (mA)*
2.5 V LVCMOS	2 mA	16	18
	4 mA	16	18
	6 mA	32	37
	8 mA	32	37
	12 mA	65	74
	16 mA	83	87
	24 mA	169	124
1.8 V LVCMOS	2 mA	9	11
	4 mA	17	22
	6 mA	35	44
	8 mA	45	51
	12 mA	91	74
	16 mA	91	74
1.5 V LVCMOS	2 mA	13	16
	4 mA	25	33
	6 mA	32	39
	8 mA	66	55
	12 mA	66	55
3.3 V PCI/PCI-X	Per PCI/PCI-X specification	103	109
Applicable to Standard I/O Banks			
3.3 V LVTTL / 3.3 V LVCMOS	2 mA	25	27
	4 mA	25	27
	6 mA	51	54
	8 mA	51	54
2.5 V LVCMOS	2 mA	16	18
	4 mA	16	18
	6 mA	32	37
	8 mA	32	37
1.8 V LVCMOS	2 mA	9	11
	4 mA	17	22
1.5 V LVCMOS	2 mA	13	16

*Note: *T_J = 100°C*

The length of time an I/O can withstand IOSH/IOSL events depends on the junction temperature. The reliability data below is based on a 3.3 V, 36 mA I/O setting, which is the worst case for this type of analysis.

For example, at 100°C, the short current condition would have to be sustained for more than six months to cause a reliability concern. The I/O design does not contain any short circuit protection, but such protection would only be needed in extremely prolonged stress conditions.

Table 3-3 • **Input Resistance of Analog Pads**

Pads	Pad Configuration	Prescaler Range	Input Resistance to Ground
AV, AC	Analog Input (direct input to ADC)	–	2 kΩ (typical)
		–	> 10 MΩ
	Analog Input (positive prescaler)	+16 V to +2 V	1 MΩ (typical)
		+1 V to +0.125 V	> 10 MΩ
	Analog Input (negative prescaler)	–16 V to –2 V	1 MΩ (typical)
		–1 V to –0.125 V	> 10 MΩ
	Digital input	+16 V to +2 V	1 MΩ (typical)
	Current monitor	+16 V to +2 V	1 MΩ (typical)
		–16 V to –2 V	1 MΩ (typical)
AT	Analog Input (direct input to ADC)	–	1 MΩ (typical)
	Analog Input (positive prescaler)	+16 V, +4 V	1 MΩ (typical)
	Digital input	+16 V, +4 V	1 MΩ (typical)
	Temperature monitor	+16 V, +4 V	> 10 MΩ

Table 3-4 • **Overshoot and Undershoot Limits** [1]

VCCI	Average VCCI–GND Overshoot or Undershoot Duration as a Percentage of Clock Cycle[2]	Maximum Overshoot/ Undershoot[2]
2.7 V or less	10%	1.4 V
	5%	1.49 V
3.0 V	10%	1.1 V
	5%	1.19 V
3.3 V	10%	0.79 V
	5%	0.88 V
3.6 V	10%	0.45 V
	5%	0.54 V

Notes:

1. *Based on reliability requirements at a junction temperature of 85°C.*
2. *The duration is allowed at one cycle out of six clock cycle. If the overshoot/undershoot occurs at one out of two cycles, the maximum overshoot/undershoot has to be reduced by 0.15 V.*

Theta-JA

Junction-to-ambient thermal resistance (θ_{JA}) is determined under standard conditions specified by JEDEC (JESD-51), but it has little relevance in actual performance of the product. It should be used with caution but is useful for comparing the thermal performance of one package to another.

A sample calculation showing the maximum power dissipation allowed for the AFS600-FG484 package under forced convection of 1.0 m/s and 75°C ambient temperature is as follows:

$$\text{Maximum Power Allowed} = \frac{T_{J(MAX)} - T_{A(MAX)}}{\theta_{JA}}$$

EQ 4

where

θ_{JA} = 19.00°C/W (taken from Table 3-6 on page 3-7).

T_A = 75.00°C

$$\text{Maximum Power Allowed} = \frac{100.00°C - 75.00°C}{19.00°C/W} = 1.3\ W$$

EQ 5

The power consumption of a device can be calculated using the Microsemi power calculator. The device's power consumption must be lower than the calculated maximum power dissipation by the package. If the power consumption is higher than the device's maximum allowable power dissipation, a heat sink can be attached on top of the case, or the airflow inside the system must be increased.

Theta-JB

Junction-to-board thermal resistance (θ_{JB}) measures the ability of the package to dissipate heat from the surface of the chip to the PCB. As defined by the JEDEC (JESD-51) standard, the thermal resistance from junction to board uses an isothermal ring cold plate zone concept. The ring cold plate is simply a means to generate an isothermal boundary condition at the perimeter. The cold plate is mounted on a JEDEC standard board with a minimum distance of 5.0 mm away from the package edge.

Theta-JC

Junction-to-case thermal resistance (θ_{JC}) measures the ability of a device to dissipate heat from the surface of the chip to the top or bottom surface of the package. It is applicable for packages used with external heat sinks. Constant temperature is applied to the surface in consideration and acts as a boundary condition. This only applies to situations where all or nearly all of the heat is dissipated through the surface in consideration.

Calculation for Heat Sink

For example, in a design implemented in an AFS600-FG484 package with 2.5 m/s airflow, the power consumption value using the power calculator is 3.00 W. The user-dependent T_a and T_j are given as follows:

T_J = 100.00°C

T_A = 70.00°C

From the datasheet:

θ_{JA} = 17.00°C/W

θ_{JC} = 8.28°C/W

$$P = \frac{T_J - T_A}{\theta_{JA}} = \frac{100°C - 70°C}{17.00\ W} = 1.76\ W$$

EQ 6

Table 3-8 • **AFS1500 Quiescent Supply Current Characteristics (continued)**

Parameter	Description	Conditions	Temp.	Min.	Typ.	Max.	Unit
IJTAG	JTAG I/O quiescent current	Operational standby[4], VJTAG = 3.63 V	T_J = 25°C		80	100	µA
			T_J = 85°C		80	100	µA
			T_J = 100°C		80	100	µA
		Standby mode[5] or Sleep mode[6], VJTAG = 0 V			0	0	µA
IPP	Programming supply current	Non-programming mode, VPUMP = 3.63 V	T_J = 25°C		39	80	µA
			T_J = 85°C		40	80	µA
			T_J = 100°C		40	80	µA
		Standby mode[5] or Sleep mode[6], VPUMP = 0 V			0	0	µA
ICCNVM	Embedded NVM current	Reset asserted, V_{CCNVM} = 1.575 V	T_J = 25°C		50	150	µA
			T_J =85°C		50	150	µA
			T_J = 100°C		50	150	µA
ICCPLL	1.5 V PLL quiescent current	Operational standby , VCCPLL = 1.575 V	T_J = 25°C		130	200	µA
			T_J = 85°C		130	200	µA
			T_J = 100°C		130	200	µA

Notes:

1. *ICC is the 1.5 V power supplies, ICC and ICC15A.*
2. *ICC33A includes ICC33A, ICC33PMP, and ICCOSC.*
3. *ICCI includes all ICCI0, ICCI1, ICCI2, and ICCI4.*
4. *Operational standby is when the Fusion device is powered up, all blocks are used, no I/O is toggling, Voltage Regulator is loaded with 200 mA, VCC33PMP is ON, XTAL is ON, and ADC is ON.*
5. *XTAL is configured as high gain, VCC = VJTAG = VPUMP = 0 V.*
6. *Sleep Mode, VCC = VJTAG = VPUMP = 0 V.*

FG256				
Pin Number	**AFS090 Function**	**AFS250 Function**	**AFS600 Function**	**AFS1500 Function**
M15	TRST	TRST	TRST	TRST
M16	GND	GND	GND	GND
N1	GEB2/IO42PDB3V0	GEB2/IO59PDB3V0	GEB2/IO59PDB4V0	GEB2/IO86PDB4V0
N2	GEA2/IO42NDB3V0	IO59NDB3V0	IO59NDB4V0	IO86NDB4V0
N3	NC	GEA2/IO58PPB3V0	GEA2/IO58PPB4V0	GEA2/IO85PPB4V0
N4	VCC33PMP	VCC33PMP	VCC33PMP	VCC33PMP
N5	VCC15A	VCC15A	VCC15A	VCC15A
N6	NC	NC	AG0	AG0
N7	AC1	AC1	AC3	AC3
N8	AG3	AG3	AG5	AG5
N9	AV3	AV3	AV5	AV5
N10	AG4	AG4	AG6	AG6
N11	NC	NC	AC8	AC8
N12	GNDA	GNDA	GNDA	GNDA
N13	VCC33A	VCC33A	VCC33A	VCC33A
N14	VCCNVM	VCCNVM	VCCNVM	VCCNVM
N15	TCK	TCK	TCK	TCK
N16	TDI	TDI	TDI	TDI
P1	VCCNVM	VCCNVM	VCCNVM	VCCNVM
P2	GNDNVM	GNDNVM	GNDNVM	GNDNVM
P3	GNDA	GNDA	GNDA	GNDA
P4	NC	NC	AC0	AC0
P5	NC	NC	AG1	AG1
P6	NC	NC	AV1	AV1
P7	AG0	AG0	AG2	AG2
P8	AG2	AG2	AG4	AG4
P9	GNDA	GNDA	GNDA	GNDA
P10	NC	AC5	AC7	AC7
P11	NC	NC	AV8	AV8
P12	NC	NC	AG8	AG8
P13	NC	NC	AV9	AV9
P14	ADCGNDREF	ADCGNDREF	ADCGNDREF	ADCGNDREF
P15	PTBASE	PTBASE	PTBASE	PTBASE
P16	GNDNVM	GNDNVM	GNDNVM	GNDNVM
R1	VCCIB3	VCCIB3	VCCIB4	VCCIB4
R2	PCAP	PCAP	PCAP	PCAP
R3	NC	NC	AT1	AT1
R4	NC	NC	AT0	AT0

FG256				
Pin Number	**AFS090 Function**	**AFS250 Function**	**AFS600 Function**	**AFS1500 Function**
R5	AV0	AV0	AV2	AV2
R6	AT0	AT0	AT2	AT2
R7	AV1	AV1	AV3	AV3
R8	AT3	AT3	AT5	AT5
R9	AV4	AV4	AV6	AV6
R10	NC	AT5	AT7	AT7
R11	NC	AV5	AV7	AV7
R12	NC	NC	AT9	AT9
R13	NC	NC	AG9	AG9
R14	NC	NC	AC9	AC9
R15	PUB	PUB	PUB	PUB
R16	VCCIB1	VCCIB1	VCCIB2	VCCIB2
T1	GND	GND	GND	GND
T2	NCAP	NCAP	NCAP	NCAP
T3	VCC33N	VCC33N	VCC33N	VCC33N
T4	NC	NC	ATRTN0	ATRTN0
T5	AT1	AT1	AT3	AT3
T6	ATRTN0	ATRTN0	ATRTN1	ATRTN1
T7	AT2	AT2	AT4	AT4
T8	ATRTN1	ATRTN1	ATRTN2	ATRTN2
T9	AT4	AT4	AT6	AT6
T10	ATRTN2	ATRTN2	ATRTN3	ATRTN3
T11	NC	NC	AT8	AT8
T12	NC	NC	ATRTN4	ATRTN4
T13	GNDA	GNDA	GNDA	GNDA
T14	VCC33A	VCC33A	VCC33A	VCC33A
T15	VAREF	VAREF	VAREF	VAREF
T16	GND	GND	GND	GND

FG484		
Pin Number	AFS600 Function	AFS1500 Function
H13	GND	GND
H14	VCCIB1	VCCIB1
H15	GND	GND
H16	GND	GND
H17	NC	IO53NDB2V0
H18	IO38PDB2V0	IO57PDB2V0
H19	GCA2/IO39PDB2V0	GCA2/IO59PDB2V0
H20	VCCIB2	VCCIB2
H21	IO37NDB2V0	IO54NDB2V0
H22	IO37PDB2V0	IO54PDB2V0
J1	NC	IO112PPB4V0
J2	IO76NDB4V0	IO113NDB4V0
J3	GFB2/IO74PDB4V0	GFB2/IO109PDB4V0
J4	GFA2/IO75PDB4V0	GFA2/IO110PDB4V0
J5	NC	IO112NPB4V0
J6	NC	IO104PDB4V0
J7	NC	IO111PDB4V0
J8	VCCIB4	VCCIB4
J9	GND	GND
J10	VCC	VCC
J11	GND	GND
J12	VCC	VCC
J13	GND	GND
J14	VCC	VCC
J15	VCCIB2	VCCIB2
J16	GCB2/IO40PDB2V0	GCB2/IO60PDB2V0
J17	NC	IO58NDB2V0
J18	IO38NDB2V0	IO57NDB2V0
J19	IO39NDB2V0	IO59NDB2V0
J20	GCC2/IO41PDB2V0	GCC2/IO61PDB2V0
J21	NC	IO55PSB2V0
J22	IO42PDB2V0	IO56PDB2V0
K1	GFC2/IO73PDB4V0	GFC2/IO108PDB4V0
K2	GND	GND
K3	IO74NDB4V0	IO109NDB4V0

FG484		
Pin Number	AFS600 Function	AFS1500 Function
K4	IO75NDB4V0	IO110NDB4V0
K5	GND	GND
K6	NC	IO104NDB4V0
K7	NC	IO111NDB4V0
K8	GND	GND
K9	VCC	VCC
K10	GND	GND
K11	VCC	VCC
K12	GND	GND
K13	VCC	VCC
K14	GND	GND
K15	GND	GND
K16	IO40NDB2V0	IO60NDB2V0
K17	NC	IO58PDB2V0
K18	GND	GND
K19	NC	IO68NPB2V0
K20	IO41NDB2V0	IO61NDB2V0
K21	GND	GND
K22	IO42NDB2V0	IO56NDB2V0
L1	IO73NDB4V0	IO108NDB4V0
L2	VCCOSC	VCCOSC
L3	VCCIB4	VCCIB4
L4	XTAL2	XTAL2
L5	GFC1/IO72PDB4V0	GFC1/IO107PDB4V0
L6	VCCIB4	VCCIB4
L7	GFB1/IO71PDB4V0	GFB1/IO106PDB4V0
L8	VCCIB4	VCCIB4
L9	GND	GND
L10	VCC	VCC
L11	GND	GND
L12	VCC	VCC
L13	GND	GND
L14	VCC	VCC
L15	VCCIB2	VCCIB2
L16	IO48PDB2V0	IO70PDB2V0

FG676	
Pin Number	AFS1500 Function
R21	IO72NDB2V0
R22	IO72PDB2V0
R23	GND
R24	IO71PDB2V0
R25	VCCIB2
R26	IO67NDB2V0
T1	GND
T2	NC
T3	GFA1/IO105PDB4V0
T4	GFA0/IO105NDB4V0
T5	IO101NDB4V0
T6	IO96PDB4V0
T7	IO96NDB4V0
T8	IO99NDB4V0
T9	IO97NDB4V0
T10	VCCIB4
T11	VCC
T12	GND
T13	VCC
T14	GND
T15	VCC
T16	GND
T17	VCCIB2
T18	IO83NDB2V0
T19	IO78NDB2V0
T20	GDA1/IO81PDB2V0
T21	GDB1/IO80PDB2V0
T22	IO73NDB2V0
T23	IO73PDB2V0
T24	IO71NDB2V0
T25	NC
T26	GND
U1	NC
U2	NC
U3	IO102PDB4V0
U4	IO102NDB4V0

FG676	
Pin Number	AFS1500 Function
U5	VCCIB4
U6	IO91PDB4V0
U7	IO91NDB4V0
U8	IO92PDB4V0
U9	GND
U10	GND
U11	VCC33A
U12	GNDA
U13	VCC33A
U14	GNDA
U15	VCC33A
U16	GNDA
U17	VCC
U18	GND
U19	IO74NDB2V0
U20	GDA0/IO81NDB2V0
U21	GDB0/IO80NDB2V0
U22	VCCIB2
U23	IO75NDB2V0
U24	IO75PDB2V0
U25	NC
U26	NC
V1	NC
V2	VCCIB4
V3	IO100PPB4V0
V4	GND
V5	IO95PDB4V0
V6	IO95NDB4V0
V7	VCCIB4
V8	IO92NDB4V0
V9	GNDNVM
V10	GNDA
V11	NC
V12	AV4
V13	NC
V14	AV5

FG676	
Pin Number	AFS1500 Function
V15	AC5
V16	NC
V17	GNDA
V18	IO77PPB2V0
V19	IO74PDB2V0
V20	VCCIB2
V21	IO82NDB2V0
V22	GDA2/IO82PDB2V0
V23	GND
V24	GDC1/IO79PDB2V0
V25	VCCIB2
V26	NC
W1	GND
W2	IO94PPB4V0
W3	IO98PDB4V0
W4	IO98NDB4V0
W5	GEC1/IO90PDB4V0
W6	GEC0/IO90NDB4V0
W7	GND
W8	VCCNVM
W9	VCCIB4
W10	VCC15A
W11	GNDA
W12	AC4
W13	VCC33A
W14	GNDA
W15	AG5
W16	GNDA
W17	PUB
W18	VCCIB2
W19	TDI
W20	GND
W21	IO84NDB2V0
W22	GDC2/IO84PDB2V0
W23	IO77NPB2V0
W24	GDC0/IO79NDB2V0