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Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details

Product Status	Active
Number of LABs/CLBs	-
Number of Logic Elements/Cells	-
Total RAM Bits	36864
Number of I/O	114
Number of Gates	250000
Voltage - Supply	1.425V ~ 1.575V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 85°C (TJ)
Package / Case	256-LBGA
Supplier Device Package	256-FPBGA (17x17)
Purchase URL	https://www.e-xfl.com/product-detail/microchip-technology/afs250-fg256

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Fusion Device Architecture Overview



Figure 1 • Fusion Device Architecture Overview (AFS600)

Package I/Os: Single-/Double-Ended (Analog)

Fusion Devices	AFS090	AFS250	AFS600	AFS1500
ARM Cortex-M1 Devices		M1AFS250	M1AFS600	M1AFS1500
Pigeon Point Devices			P1AFS600 ¹	P1AFS1500 ¹
MicroBlade Devices		U1AFS250 ²	U1AFS600 ²	U1AFS1500 ²
QN108 ³	37/9 (16)			
QN180 ³	60/16 (20)	65/15 (24)		
PQ208 ⁴		93/26 (24)	95/46 (40)	
FG256	75/22 (20)	114/37 (24)	119/58 (40)	119/58 (40)
FG484			172/86 (40)	223/109 (40)
FG676				252/126 (40)
Notes:	•	1		•

1. Pigeon Point devices are only offered in FG484 and FG256.

2. MicroBlade devices are only offered in FG256.

3. Package not available.

4. Fusion devices in the same package are pin compatible with the exception of the PQ208 package (AFS250 and AFS600).



Global Resources (VersaNets)

Fusion devices offer powerful and flexible control of circuit timing through the use of analog circuitry. Each chip has six CCCs. The west CCC also contains a PLL core. In the two larger devices (AFS600 and AFS1500), the west and the east CCCs each contain a PLL. The PLLs include delay lines, a phase shifter (0°, 90°, 180°, 270°), and clock multipliers/dividers. Each CCC has all the circuitry needed for the selection and interconnection of inputs to the VersaNet global network. The east and west CCCs each have access to three VersaNet global lines on each side of the chip (six lines total). The CCCs at the four corners each have access to three quadrant global lines on each quadrant of the chip.

Advantages of the VersaNet Approach

One of the architectural benefits of Fusion is the set of powerful and low-delay VersaNet global networks. Fusion offers six chip (main) global networks that are distributed from the center of the FPGA array (Figure 2-11). In addition, Fusion devices have three regional globals (quadrant globals) in each of the four chip quadrants. Each core VersaTile has access to nine global networks on the device. Each of these networks contains spines and ribs that reach all VersaTiles in all quadrants (Figure 2-12 on page 2-12). This flexible VersaNet global network architecture allows users to map up to 180 different internal/external clocks in a Fusion device. Details on the VersaNet networks are given in Table 2-4 on page 2-12. The flexibility of the Fusion VersaNet global network allows the designer to address several design requirements. User applications that are clock-resource-intensive can easily route external or gated internal clocks using VersaNet global routing networks. Designers can also drastically reduce delay penalties and minimize resource usage by mapping critical, high-fanout nets to the VersaNet global network.



Figure 2-11 • Overview of Fusion VersaNet Global Network

VersaNet Timing Characteristics

Global clock delays include the central rib delay, the spine delay, and the row delay. Delays do not include I/O input buffer clock delays, as these are dependent upon I/O standard, and the clock may be driven and conditioned internally by the CCC module. Table 2-5, Table 2-6, Table 2-7, and Table 2-8 on page 2-17 present minimum and maximum global clock delays within the device Minimum and maximum delays are measured with minimum and maximum loading, respectively.

Timing Characteristics

 Table 2-5 • AFS1500 Global Resource Timing

 Commercial Temperature Range Conditions: T_J = 70°C, Worst-Case VCC = 1.425 V

Parameter	Description		-2		-1		Std.	
Farameter	Description	Min. ¹	Max. ²	Min. ¹	Max. ²	Min. ¹	Max. ²	Units
t _{RCKL}	Input Low Delay for Global Clock	1.53	1.75	1.74	1.99	2.05	2.34	ns
t _{RCKH}	Input High Delay for Global Clock	1.53	1.79	1.75	2.04	2.05	2.40	ns
t _{RCKMPWH}	Minimum Pulse Width High for Global Clock							ns
t _{RCKMPWL}	Minimum Pulse Width Low for Global Clock							ns
t _{RCKSW}	Maximum Skew for Global Clock		0.26		0.29		0.34	ns

Notes:

1. Value reflects minimum load. The delay is measured from the CCC output to the clock pin of a sequential element located in a lightly loaded row (single element is connected to the global net).

2. Value reflects maximum load. The delay is measured on the clock pin of the farthest sequential element located in a fully loaded row (all available flip-flops are connected to the global net in the row).

3. For the derating values at specific junction temperature and voltage supply levels, refer to Table 3-7 on page 3-9.

Table 2-6 • AFS600 Global Resource Timing

Commercial Temperature Range Conditions: T_J = 70°C, Worst-Case VCC = 1.425 V

Parameter	Description		-2		-1		Std.	
Falailletei	Description	Min. ¹	Max. ²	Min. ¹	Max. ²	Min. ¹	td. U Max. ² U 2.00 2.06 0 0.36 0	Units
t _{RCKL}	Input Low Delay for Global Clock	1.27	1.49	1.44	1.70	1.69	2.00	ns
t _{RCKH}	Input High Delay for Global Clock	1.26	1.54	1.44	1.75	1.69	2.06	ns
t _{RCKMPWH}	Minimum Pulse Width High for Global Clock							ns
t _{RCKMPWL}	Minimum Pulse Width Low for Global Clock							ns
t _{RCKSW}	Maximum Skew for Global Clock		0.27		0.31		0.36	ns

Notes:

1. Value reflects minimum load. The delay is measured from the CCC output to the clock pin of a sequential element located in a lightly loaded row (single element is connected to the global net).

2. Value reflects maximum load. The delay is measured on the clock pin of the farthest sequential element located in a fully loaded row (all available flip-flops are connected to the global net in the row).

3. For the derating values at specific junction temperature and voltage supply levels, refer to Table 3-7 on page 3-9.



DINA and DINB

These are the input data signals, and they are nine bits wide. Not all nine bits are valid in all configurations. When a data width less than nine is specified, unused high-order signals must be grounded (Table 2-29).

DOUTA and DOUTB

These are the nine-bit output data signals. Not all nine bits are valid in all configurations. As with DINA and DINB, high-order bits may not be used (Table 2-29). The output data on unused pins is undefined.

Table 2-29 • Unused/Used Input and Output Data Pins for Various Supported Bus Widths

D×W	DINx/DOUTx						
	Unused	Used					
4k×1	[8:1]	[0]					
2k×2	[8:2]	[1:0]					
1k×4	[8:4]	[3:0]					
512×9	None	[8:0]					

Note: The "x" in DINx and DOUTx implies A or B.

ESTOP, FSTOP

ESTOP is used to stop the FIFO read counter from further counting once the FIFO is empty (i.e., the EMPTY flag goes High). A High on this signal inhibits the counting.

FSTOP is used to stop the FIFO write counter from further counting once the FIFO is full (i.e., the FULL flag goes High). A High on this signal inhibits the counting.

For more information on these signals, refer to the "ESTOP and FSTOP Usage" section on page 2-70.

FULL, EMPTY

When the FIFO is full and no more data can be written, the FULL flag asserts High. The FULL flag is synchronous to WCLK to inhibit writing immediately upon detection of a full condition and to prevent overflows. Since the write address is compared to a resynchronized (and thus time-delayed) version of the read address, the FULL flag will remain asserted until two WCLK active edges after a read operation eliminates the full condition.

When the FIFO is empty and no more data can be read, the EMPTY flag asserts High. The EMPTY flag is synchronous to RCLK to inhibit reading immediately upon detection of an empty condition and to prevent underflows. Since the read address is compared to a resynchronized (and thus time-delayed) version of the write address, the EMPTY flag will remain asserted until two RCLK active edges after a write operation removes the empty condition.

For more information on these signals, refer to the "FIFO Flag Usage Considerations" section on page 2-70.

AFULL, AEMPTY

These are programmable flags and will be asserted on the threshold specified by AFVAL and AEVAL, respectively.

When the number of words stored in the FIFO reaches the amount specified by AEVAL while reading, the AEMPTY output will go High. Likewise, when the number of words stored in the FIFO reaches the amount specified by AFVAL while writing, the AFULL output will go High.

AFVAL, AEVAL

The AEVAL and AFVAL pins are used to specify the almost-empty and almost-full threshold values, respectively. They are 12-bit signals. For more information on these signals, refer to "FIFO Flag Usage Considerations" section.

ESTOP and FSTOP Usage

The ESTOP pin is used to stop the read counter from counting any further once the FIFO is empty (i.e., the EMPTY flag goes High). Likewise, the FSTOP pin is used to stop the write counter from counting any further once the FIFO is full (i.e., the FULL flag goes High).

The FIFO counters in the Fusion device start the count at 0, reach the maximum depth for the configuration (e.g., 511 for a 512×9 configuration), and then restart at 0. An example application for the ESTOP, where the read counter keeps counting, would be writing to the FIFO once and reading the same content over and over without doing another write.

FIFO Flag Usage Considerations

The AEVAL and AFVAL pins are used to specify the 12-bit AEMPTY and AFULL threshold values, respectively. The FIFO contains separate 12-bit write address (WADDR) and read address (RADDR) counters. WADDR is incremented every time a write operation is performed, and RADDR is incremented every time a read operation is performed. Whenever the difference between WADDR and RADDR is greater than or equal to AFVAL, the AFULL output is asserted. Likewise, whenever the difference between WADDR and RADDR is less than or equal to AEVAL, the AEMPTY output is asserted. To handle different read and write aspect ratios, AFVAL and AEVAL are expressed in terms of total data bits instead of total data words. When users specify AFVAL and AEVAL in terms of read or write words, the SmartGen tool translates them into bit addresses and configures these signals automatically. SmartGen configures the AFULL flag to assert when the write address exceeds the read address by at least a predefined value. In a 2k×8 FIFO, for example, a value of 1,500 for AFVAL means that the AFULL flag will be asserted after a write when the difference between the write address and the read address reaches 1,500 (there have been at least 1500 more writes than reads). It will stay asserted until the difference between the write and read addresses drops below 1,500.

Table 2-36 describes each pin in the Analog Block. Each function within the Analog Block will be explained in detail in the following sections.

Table 2-36 • Analog Block Pin Description

Signal Name	Number of Bits	Direction	Function	Location of Details
VAREF	1	Input/Output	Voltage reference for ADC	ADC
ADCGNDREF	1	Input	External ground reference	ADC
MODE[3:0]	4	Input	ADC operating mode	ADC
SYSCLK	1	Input	External system clock	
TVC[7:0]	8	Input	Clock divide control	ADC
STC[7:0]	8	Input	Sample time control	ADC
ADCSTART	1	Input	Start of conversion	ADC
PWRDWN	1	Input	ADC comparator power-down if 1. When asserted, the ADC will stop functioning, and the digital portion of the analog block will continue operating. This may result in invalid status flags from the analog block. Therefore, Microsemi does not recommend asserting the PWRDWN pin.	ADC
ADCRESET	1	Input	ADC resets and disables Analog Quad – active high	ADC
BUSY	1	Output	1 – Running conversion	ADC
CALIBRATE	1	Output	1 – Power-up calibration	ADC
DATAVALID	1	Output	1 – Valid conversion result	ADC
RESULT[11:0]	12	Output	Conversion result	ADC
TMSTBINT	1	Input	Internal temp. monitor strobe	ADC
SAMPLE	1	Output	 1 – An analog signal is actively being sampled (stays high during signal acquisition only) 0 – No analog signal is being sampled 	ADC
VAREFSEL	1	Input	0 = Output internal voltage reference (2.56 V) to VAREF 1 = Input external voltage reference	ADC
	_		from VAREF and ADCGNDREF	
CHNUMBER[4:0]	5	Input	Analog input channel select	Input multiplexer
ACMCLK	1	Input	ACM clock	ACM
ACMWEN	1	Input	ACM write enable – active high	ACM
ACMRESET	1	Input	ACM reset – active low	ACM
ACMWDATA[7:0]	8	Input	ACM write data	ACM
ACMRDATA[7:0]	8	Output	ACM read data	ACM
ACMADDR[7:0]	8	Input	ACM address	ACM
CMSTB0 to CMSTB9	10	Input	Current monitor strobe – 1 per quad, active high	Analog Quad



Figure 2-90 • Input Setup Time

Standard Conversion



Notes:

1. Refer to EQ 20 on page 2-109 for the calculation on the sample time, t_{SAMPLE} .

2. See EQ 23 on page 2-109 for calculation of the conversion time, t_{CONV} .

3. Minimum time to issue an ADCSTART after DATAVALID is 1 SYSCLK period

Figure 2-91 • Standard Conversion Status Signal Timing Diagram

Table 2-68 • I/O Bank Support by Device

I/O Bank	AFS090	AFS250	AFS600	AFS1500
Standard I/O	Ν	Ν	_	-
Advanced I/O	E, W	E, W	E, W	E, W
Pro I/O	-	_	Ν	Ν
Analog Quad	S	S	S	S

Note: E = *East side of the device*

W = West side of the device

N = North side of the device

S = South side of the device

Table 2-69 • Fusion VCCI Voltages and Compatible Standards

VCCI (typical)	Compatible Standards
3.3 V	LVTTL/LVCMOS 3.3, PCI 3.3, SSTL3 (Class I and II),* GTL+ 3.3, GTL 3.3,* LVPECL
2.5 V	LVCMOS 2.5, LVCMOS 2.5/5.0, SSTL2 (Class I and II),* GTL+ 2.5,* GTL 2.5,* LVDS, BLVDS, M-LVDS
1.8 V	LVCMOS 1.8
1.5 V	LVCMOS 1.5, HSTL (Class I),* HSTL (Class II)*

Note: *I/O standard supported by Pro I/O banks.

Table 2-70 • Fusion VREF Voltages and Compatible Standards*

VREF (typical)	Compatible Standards
1.5 V	SSTL3 (Class I and II)
1.25 V	SSTL2 (Class I and II)
1.0 V	GTL+ 2.5, GTL+ 3.3
0.8 V	GTL 2.5, GTL 3.3
0.75 V	HSTL (Class I), HSTL (Class II)

Note: *I/O standards supported by Pro I/O banks.



Table 2-77 • Comparison Table for 5 V–Compliant Receiver Scheme

Scheme	Board Components	Speed	Current Limitations
1	Two resistors	Low to high ¹	Limited by transmitter's drive strength
2	Resistor and Zener 3.3 V	Medium	Limited by transmitter's drive strength
3	Bus switch	High	N/A
4	Minimum resistor value ² R = 47 Ω at T _J = 70°C	Medium	Maximum diode current at 100% duty cycle, signal constantly at '1'
	R = 150 Ω at T_J = 85°C		52.7 mA at $T_J = 70^{\circ}$ C / 10-year lifetime 16.5 mA at $T_J = 85^{\circ}$ C / 10-year lifetime
	$R = 420 \Omega a (1_{\rm J} = 100 C)$		5.9 mA at $T_J = 100^{\circ}$ C / 10-year lifetime
			For duty cycles other than 100%, the currents can be increased by a factor = 1 / (duty cycle).
			Example: 20% duty cycle at 70°C
			Maximum current = (1 / 0.2) * 52.7 mA = 5 * 52.7 mA = 263.5 mA

Notes:

1. Speed and current consumption increase as the board resistance values decrease.

2. Resistor values ensure I/O diode long-term reliability.



I/O Standard	Input/Output Supply Voltage (VCCI_TYP)	Input Reference Voltage (VREF_TYP)	Board Termination Voltage (VTT_TYP)
LVTTL/LVCMOS 3.3 V	3.30 V	-	-
LVCMOS 2.5 V	2.50 V	-	-
LVCMOS 2.5 V / 5.0 V Input	2.50 V	-	-
LVCMOS 1.8 V	1.80 V	-	-
LVCMOS 1.5 V	1.50 V	-	-
PCI 3.3 V	3.30 V	-	-
PCI-X 3.3 V	3.30 V	-	-
GTL+ 3.3 V	3.30 V	1.00 V	1.50 V
GTL+ 2.5 V	2.50 V	1.00 V	1.50 V
GTL 3.3 V	3.30 V	0.80 V	1.20 V
GTL 2.5 V	2.50 V	0.80 V	1.20 V
HSTL Class I	1.50 V	0.75 V	0.75 V
HSTL Class II	1.50 V	0.75 V	0.75 V
SSTL3 Class I	3.30 V	1.50 V	1.50 V
SSTL3 Class II	3.30 V	1.50 V	1.50 V
SSTL2 Class I	2.50 V	1.25 V	1.25 V
SSTL2 Class II	2.50 V	1.25 V	1.25 V
LVDS, BLVDS, M-LVDS	2.50 V	-	-
LVPECL	3.30 V	_	-

Table 2-83 • Fusion Pro I/O Supported Standards and Corresponding VREF and VTT Voltages

Overview of I/O Performance Summary of I/O DC Input and Output Levels – Default I/O Software Settings

Table 2-86 • Summary of Maximum and Minimum DC Input and Output Levels Applicable to Commercial and Industrial Conditions Applicable to Pro I/Os

	VIL		VIH		VOL	VOH	IOL	IOH		
I/O Standard	Drive Strength	Slew Rate	Min. V	Max. V	Min. V	Max. V	Max. V	Min. V	mA	mA
3.3 V LVTTL / 3.3 V LVCMOS	12 mA	High	-0.3	0.8	2	3.6	0.4	2.4	12	12
2.5 V LVCMOS	12 mA	High	-0.3	0.7	1.7	3.6	0.7	1.7	12	12
1.8 V LVCMOS	12 mA	High	-0.3	0.35 * VCCI	0.65 * VCCI	3.6	0.45	VCCI - 0.45	12	12
1.5 V LVCMOS	12 mA	High	-0.3	0.35 * VCCI	0.65 * VCCI	3.6	0.25 * VCCI	0.75 * VCCI	12	12
3.3 V PCI		•	•		Per PCI Spec	ification				
3.3 V PCI-X					Per PCI-X Spe	cification				
3.3 V GTL	20 mA ²	High	-0.3	VREF-0.05	VREF + 0.05	3.6	0.4	-	20	20
2.5 V GTL	20 mA ²	High	-0.3	VREF-0.05	VREF + 0.05	3.6	0.4	-	20	20
3.3 V GTL+	35 mA	High	-0.3	VREF – 0.1	VREF + 0.1	3.6	0.6	-	35	35
2.5 V GTL+	33 mA	High	-0.3	VREF – 0.1	VREF + 0.1	3.6	0.6	-	33	33
HSTL (I)	8 mA	High	-0.3	VREF – 0.1	VREF + 0.1	3.6	0.4	VCCI – 0.4	8	8
HSTL (II)	15 mA ²	High	-0.3	VREF – 0.1	VREF + 0.1	3.6	0.4	VCCI – 0.4	15	15
SSTL2 (I)	15 mA	High	-0.3	VREF – 0.2	VREF + 0.2	3.6	0.54	VCCI-0.62	15	15
SSTL2 (II)	18 mA	High	-0.3	VREF – 0.2	VREF + 0.2	3.6	0.35	VCCI-0.43	18	18
SSTL3 (I)	14 mA	High	-0.3	VREF – 0.2	VREF + 0.2	3.6	0.7	VCCI – 1.1	14	14
SSTL3 (II)	21 mA	High	-0.3	VREF – 0.2	VREF + 0.2	3.6	0.5	VCCI – 0.9	21	21

Notes:

1. Currents are measured at 85°C junction temperature.

2. Output drive strength is below JEDEC specification.

3. Output slew rate can be extracted by the IBIS models.

Table 2-87 • Summary of Maximum and Minimum DC Input and Output Levels Applicable to Commercial and Industrial Conditions Applicable to Advanced I/Os

			VIL VIH		VOL	VOH	IOL	ЮН		
I/O Standard	Drive Strength	Slew Rate	Min. V	Max. V	Min. V	Max. V	Max. V	Min. V	mA	mA
3.3 V LVTTL / 3.3 V LVCMOS	12 mA	High	-0.3	0.8	2	3.6	0.4	2.4	12	12
2.5 V LVCMOS	12 mA	High	-0.3	0.7	1.7	2.7	0.7	1.7	12	12
1.8 V LVCMOS	12 mA	High	-0.3	0.35 * VCCI	0.65 * VCCI	1.9	0.45	VCCI-0.45	12	12
1.5 V LVCMOS	12 mA	High	-0.3	0.35 * VCCI	0.65 * VCCI	1.575	0.25 * VCCI	0.75 * VCCI	12	12
3.3 V PCI		Per PCI specifications								
3.3 V PCI-X				Р	er PCI-X spec	cificatior	าร			

Note: Currents are measured at 85°C junction temperature.

Summary of I/O Timing Characteristics – Default I/O Software Settings

Table 2-90 • Summary of AC Measuring Points Applicable to All I/O Bank Types

Standard	Input Reference Voltage (VREF_TYP)	Board Termination Voltage (VTT_REF)	Measuring Trip Point (Vtrip)
3.3 V LVTTL / 3.3 V LVCMOS	_	-	1.4 V
2.5 V LVCMOS	_	-	1.2 V
1.8 V LVCMOS	-	-	0.90 V
1.5 V LVCMOS	_	-	0.75 V
3.3 V PCI	_	_	0.285 * VCCI (RR) 0.615 * VCCI (FF))
3.3 V PCI-X	_	_	0.285 * VCCI (RR) 0.615 * VCCI (FF)
3.3 V GTL	0.8 V	1.2 V	VREF
2.5 V GTL	0.8 V	1.2 V	VREF
3.3 V GTL+	1.0 V	1.5 V	VREF
2.5 V GTL+	1.0 V	1.5 V	VREF
HSTL (I)	0.75 V	0.75 V	VREF
HSTL (II)	0.75 V	0.75 V	VREF
SSTL2 (I)	1.25 V	1.25 V	VREF
SSTL2 (II)	1.25 V	1.25 V	VREF
SSTL3 (I)	1.5 V	1.485 V	VREF
SSTL3 (II)	1.5 V	1.485 V	VREF
LVDS	-	-	Cross point
LVPECL	-	-	Cross point

Table 2-91 • I/O AC Parameter Definitions

Parameter	Definition
t _{DP}	Data to Pad delay through the Output Buffer
t _{PY}	Pad to Data delay through the Input Buffer with Schmitt trigger disabled
t _{DOUT}	Data to Output Buffer delay through the I/O interface
t _{EOUT}	Enable to Output Buffer Tristate Control delay through the I/O interface
t _{DIN}	Input Buffer to Data delay through the I/O interface
t _{PYS}	Pad to Data delay through the Input Buffer with Schmitt trigger enabled
t _{HZ}	Enable to Pad delay through the Output Buffer—High to Z
t _{ZH}	Enable to Pad delay through the Output Buffer—Z to High
t _{LZ}	Enable to Pad delay through the Output Buffer—Low to Z
t _{ZL}	Enable to Pad delay through the Output Buffer—Z to Low
t _{ZHS}	Enable to Pad delay through the Output Buffer with delayed enable—Z to High
t _{ZLS}	Enable to Pad delay through the Output Buffer with delayed enable—Z to Low



Single-Ended I/O Characteristics

3.3 V LVTTL / 3.3 V LVCMOS

Low-Voltage Transistor–Transistor Logic is a general-purpose standard (EIA/JESD) for 3.3 V applications. It uses an LVTTL input buffer and push-pull output buffer. The 3.3 V LVCMOS standard is supported as part of the 3.3 V LVTTL support.

Table 2-102	• Minimum	and Maximum	DC Input	and Output	l evels
10016 2-102	• Willing the second		DO inpui	and Output	. LEVEIJ

3.3 V LVTTL / 3.3 V LVCMOS	v	IL	v	ін	VOL	νон	IOL	ЮН	IOSL	IOSH	IIL ¹	IIH ²
Drive Strength	Min. V	Max. V	Min. V	Max. V	Max. V	Min. V	mA	mA	Max. mA ³	Max. mA ³	μA ⁴	μA ⁴
Applicable to Pro I/O Banks												
4 mA	-0.3	0.8	2	3.6	0.4	2.4	4	4	27	25	10	10
8 mA	-0.3	0.8	2	3.6	0.4	2.4	8	8	54	51	10	10
12 mA	-0.3	0.8	2	3.6	0.4	2.4	12	12	109	103	10	10
16 mA	-0.3	0.8	2	3.6	0.4	2.4	16	16	127	132	10	10
24 mA	-0.3	0.8	2	3.6	0.4	2.4	24	24	181	268	10	10
Applicable to A	dvanced	I/O Bank	s									
2 mA	-0.3	0.8	2	3.6	0.4	2.4	2	2	27	25	10	10
4 mA	-0.3	0.8	2	3.6	0.4	2.4	4	4	27	25	10	10
6 mA	-0.3	0.8	2	3.6	0.4	2.4	6	6	54	51	10	10
8 mA	-0.3	0.8	2	3.6	0.4	2.4	8	8	54	51	10	10
12 mA	-0.3	0.8	2	3.6	0.4	2.4	12	12	109	103	10	10
16 mA	-0.3	0.8	2	3.6	0.4	2.4	16	16	127	132	10	10
24 mA	-0.3	0.8	2	3.6	0.4	2.4	24	24	181	268	10	10
Applicable to S	tandard I	/O Banks										
2 mA	-0.3	0.8	2	3.6	0.4	2.4	2	2	27	25	10	10
4 mA	-0.3	0.8	2	3.6	0.4	2.4	4	4	27	25	10	10
6 mA	-0.3	0.8	2	3.6	0.4	2.4	6	6	54	51	10	10
8 mA	-0.3	0.8	2	3.6	0.4	2.4	8	8	54	51	10	10

Notes:

1. IIL is the input leakage current per I/O pin over recommended operation conditions where –0.3 V < VIN < VIL.

2. IIH is the input leakage current per I/O pin over recommended operating conditions VIH < VIN < VCCI. Input current is larger when operating outside recommended ranges.

3. Currents are measured at high temperature (100°C junction temperature) and maximum voltage.

4. Currents are measured at 85°C junction temperature.

5. Software default selection highlighted in gray.



Figure 2-119 • AC Loading

Table 2-114 • 2.5 V LVCMOS Low Slew

Commercial Temperature Range Conditions: $T_J = 70^{\circ}$ C, Worst-Case VCC = 1.425 V, Worst-Case VCCI = 2.3 V Applicable to Advanced I/Os

Drive Strength	Speed Grade	toour	top	toin	tev	teour	tzı	t≂⊔	tı z	tu-z	tzı e	tzue	Units
4 mA	Std.	0.66	11.40	0.04	1.31	0.43	11.22	- <u>г</u> н 11.40	2.68	2.20	13.45	13.63	ns
	-1	0.56	9.69	0.04	1.11	0.36	9.54	9.69	2.28	1.88	11.44	11.60	ns
	-2	0.49	8.51	0.03	0.98	0.32	8.38	8.51	2.00	1.65	10.05	10.18	ns
8 mA	Std.	0.66	7.96	0.04	1.31	0.43	8.11	7.81	3.05	2.89	10.34	10.05	ns
	-1	0.56	6.77	0.04	1.11	0.36	6.90	6.65	2.59	2.46	8.80	8.55	ns
	-2	0.49	5.94	0.03	0.98	0.32	6.05	5.84	2.28	2.16	7.72	7.50	ns
12 mA	Std.	0.66	6.18	0.04	1.31	0.43	6.29	5.92	3.30	3.32	8.53	8.15	ns
	-1	0.56	5.26	0.04	1.11	0.36	5.35	5.03	2.81	2.83	7.26	6.94	ns
	-2	0.49	4.61	0.03	0.98	0.32	4.70	4.42	2.47	2.48	6.37	6.09	ns
16 mA	Std.	0.66	6.18	0.04	1.31	0.43	6.29	5.92	3.30	3.32	8.53	8.15	ns
	-1	0.56	5.26	0.04	1.11	0.36	5.35	5.03	2.81	2.83	7.26	6.94	ns
	-2	0.49	4.61	0.03	0.98	0.32	4.70	4.42	2.47	2.48	6.37	6.09	ns
24 mA	Std.	0.66	6.18	0.04	1.31	0.43	6.29	5.92	3.30	3.32	8.53	8.15	ns
	-1	0.56	5.26	0.04	1.11	0.36	5.35	5.03	2.81	2.83	7.26	6.94	ns
	-2	0.49	4.61	0.03	0.98	0.32	4.70	4.42	2.47	2.48	6.37	6.09	ns

Note: For the derating values at specific junction temperature and voltage supply levels, refer to Table 3-7 on page 3-9.

Table 2-122 • 1.8 V LVCMOS Low Slew

Commercial Temperature Range Conditions: $T_J = 70^{\circ}$ C, Worst-Case VCC = 1.425 V, Worst-Case VCCI = 1.7 V Applicable to Advanced I/Os

Drive Strength	Speed Grade	t _{DOUT}	t _{DP}	t _{DIN}	t _{PY}	t _{EOUT}	t _{ZL}	t _{zH}	t _{LZ}	t _{HZ}	t _{ZLS}	t _{zHS}	Units
2 mA	Std.	0.66	15.53	0.04	1.31	0.43	14.11	15.53	2.78	1.60	16.35	17.77	ns
	-1	0.56	13.21	0.04	1.11	0.36	12.01	13.21	2.36	1.36	13.91	15.11	ns
	-2 ²	0.49	11.60	0.03	0.98	0.32	10.54	11.60	2.07	1.19	12.21	13.27	ns
4 mA	Std.	0.66	10.48	0.04	1.31	0.43	10.41	10.48	3.23	2.73	12.65	12.71	ns
	-1	0.56	8.91	0.04	1.11	0.36	8.86	8.91	2.75	2.33	10.76	10.81	ns
	-2	0.49	7.82	0.03	0.98	0.32	7.77	7.82	2.41	2.04	9.44	9.49	ns
8 mA	Std.	0.66	8.05	0.04	1.31	0.43	8.20	7.84	3.54	3.27	10.43	10.08	ns
	-1	0.56	6.85	0.04	1.11	0.36	6.97	6.67	3.01	2.78	8.88	8.57	ns
	-2	0.49	6.01	0.03	0.98	0.32	6.12	5.86	2.64	2.44	7.79	7.53	ns
12 mA	Std.	0.66	7.50	0.04	1.31	0.43	7.64	7.30	3.61	3.41	9.88	9.53	ns
	-1	0.56	6.38	0.04	1.11	0.36	6.50	6.21	3.07	2.90	8.40	8.11	ns
	-2	0.49	5.60	0.03	0.98	0.32	5.71	5.45	2.69	2.55	7.38	7.12	ns
16 mA	Std.	0.66	7.29	0.04	1.31	0.43	7.23	7.29	3.71	3.95	9.47	9.53	ns
	-1	0.56	6.20	0.04	1.11	0.36	6.15	6.20	3.15	3.36	8.06	8.11	ns
	-2	0.49	5.45	0.03	0.98	0.32	5.40	5.45	2.77	2.95	7.07	7.12	ns

Note: For the derating values at specific junction temperature and voltage supply levels, refer to Table 3-7 on page 3-9.



Output DDR



Figure 2-144 • Output DDR Timing Model

Table 2-181 • Parameter Definitions

Parameter Name	Parameter Definition	Measuring Nodes (From, To)
t _{DDROCLKQ}	Clock-to-Out	B, E
t _{DDROCLR2Q}	Asynchronous Clear-to-Out	C, E
t _{DDROREMCLR}	Clear Removal	С, В
t _{DDRORECCLR}	Clear Recovery	С, В
t _{DDROSUD1}	Data Setup Data_F	A, B
t _{DDROSUD2}	Data Setup Data_R	D, B
t _{DDROHD1}	Data Hold Data_F	А, В
t _{DDROHD2}	Data Hold Data_R	D, B



TMS Test Mode Select

The TMS pin controls the use of the IEEE1532 boundary scan pins (TCK, TDI, TDO, TRST). There is an internal weak pull-up resistor on the TMS pin.

TRST Boundary Scan Reset Pin

The TRST pin functions as an active low input to asynchronously initialize (or reset) the boundary scan circuitry. There is an internal weak pull-up resistor on the TRST pin. If JTAG is not used, an external pull-down resistor could be included to ensure the TAP is held in reset mode. The resistor values must be chosen from Table 2-183 and must satisfy the parallel resistance value requirement. The values in Table 2-183 correspond to the resistor recommended when a single device is used and to the equivalent parallel resistor when multiple devices are connected via a JTAG chain.

In critical applications, an upset in the JTAG circuit could allow entering an undesired JTAG state. In such cases, Microsemi recommends tying off TRST to GND through a resistor placed close to the FPGA pin. Note that to operate at all VJTAG voltages, 500 Ω to 1 k Ω will satisfy the requirements.

Special Function Pins

NC No Connect

This pin is not connected to circuitry within the device. These pins can be driven to any voltage or can be left floating with no effect on the operation of the device.

DC Don't Connect

This pin should not be connected to any signals on the PCB. These pins should be left unconnected.

NCAP Negative Capacitor

Negative Capacitor is where the negative terminal of the charge pump capacitor is connected. A capacitor, with a 2.2 μ F recommended value, is required to connect between PCAP and NCAP.

PCAP Positive Capacitor

Positive Capacitor is where the positive terminal of the charge pump capacitor is connected. A capacitor, with a 2.2 μ F recommended value, is required to connect between PCAP and NCAP.

PUB Push Button

Push button is the connection for the external momentary switch used to turn on the 1.5 V voltage regulator and can be floating if not used.

PTBASE Pass Transistor Base

Pass Transistor Base is the control signal of the voltage regulator. This pin should be connected to the base of the external pass transistor used with the 1.5 V internal voltage regulator and can be floating if not used.

PTEM Pass Transistor Emitter

Pass Transistor Emitter is the feedback input of the voltage regulator.

This pin should be connected to the emitter of the external pass transistor used with the 1.5 V internal voltage regulator and can be floating if not used.

XTAL1 Crystal Oscillator Circuit Input

Input to crystal oscillator circuit. Pin for connecting external crystal, ceramic resonator, RC network, or external clock input. When using an external crystal or ceramic oscillator, external capacitors are also recommended (Please refer to the crystal oscillator manufacturer for proper capacitor value).

If using external RC network or clock input, XTAL1 should be used and XTAL2 left unconnected. In the case where the Crystal Oscillator block is not used, the XTAL1 pin should be connected to GND and the XTAL2 pin should be left floating.

Parameter	Description	Conditions	Temp.	Min	Тур	Мах	Unit					
ICC ¹	1.5 V quiescent current	Operational standby ⁴ ,	T _J = 25°C		4.8	10	Unit mA mA					
		VCC = 1.575 V	T _J = 85°C		8.2	30	mA					
			T _J = 100°C		15	50	mA					
		Standby mode ⁵ or Sleep mode ⁶ , VCC = 0 V			0	0	μA					
ICC33 ²	3.3 V analog supplies	Operational standby ⁴ ,	T _J = 25°C		9.8	13	mA					
	current	VCC33 = 3.63 V	T _J = 85°C		9.8	14	mA					
			T _J = 100°C		10.8	10 11 14 m/ 15 m/ 2 m/ 2 m/ 2 m/ 3.0 m/ 3.1 m/ 6 m/ 18 µ/4						
		Operational standby, only	T _J = 25°C		0.29	2	mA					
		Analog Quad and –3.3 V output ON, VCC33 = 3.63 V	T _J = 85°C		0.31	2	mA mA mA mA					
			T _J = 100°C		0.45	2	mA					
		Standby mode ⁵ , VCC33 = 3.63V	T _J = 25°C		2.9	3.0	mA					
			T _J = 85°C		2.9	3.1	mA					
			T _J = 100°C		3.5	6	mA					
		Sleep mode ⁶ , VCC33 = 3.63 V	T _J = 25°C		19	18	μA					
			T _J = 85°C		19	20	μA					
			T _J = 100°C		24	25	μA					
ICCI ³	I/O quiescent current	Operational standby ⁶ ,	T _J = 25°C		266	437	μΑ					
		VCCIX = 3.63 V	T _J = 85°C		266	437	μΑ					
			T _J = 100°C		266	437	μA					
IJTAG	JTAG I/O quiescent current	Operational standby ⁴ ,	T _J = 25°C		80	100	μA					
		VJIAG = 3.63 V	T _J = 85°C		80	100	μA					
			T _J = 100°C		80	100	μA					
		Standby mode ⁵ or Sleep mode ⁶ , VJTAG = 0 V			0	0	μA					

Table 3-10 • AFS250 Quiescent Supply Cu	urrent Characteristics
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Notes:

1. ICC is the 1.5 V power supplies, ICC, ICCPLL, ICC15A, ICCNVM.

2. ICC33A includes ICC33A, ICC33PMP, and ICCOSC.

3. ICCI includes all ICCI0, ICCI1, and ICCI2.

4. Operational standby is when the Fusion device is powered up, all blocks are used, no I/O is toggling, Voltage Regulator is loaded with 200 mA, VCC33PMP is ON, XTAL is ON, and ADC is ON.

5. XTAL is configured as high gain, VCC = VJTAG = VPUMP = 0 V.

6. Sleep Mode, VCC = VJTA G = VPUMP = 0 V.

FG256								
Pin Number	AFS090 Function	AFS250 Function	AFS600 Function	AFS1500 Function				
E13	VCCIB1	VCCIB1	VCCIB2	VCCIB2				
E14	GCC2/IO33NDB1V0	IO42NDB1V0	IO32NDB2V0	IO46NDB2V0				
E15	GCB2/IO33PDB1V0	GBC2/IO42PDB1V0	GBC2/IO32PDB2V0	GBC2/IO46PDB2V0				
E16	GND	GND	GND	GND				
F1	NC	NC	IO79NDB4V0	IO111NDB4V0				
F2	NC	NC	IO79PDB4V0	IO111PDB4V0				
F3	GFB1/IO48PPB3V0	IO72NDB3V0	IO76NDB4V0	IO112NDB4V0				
F4	GFC0/IO49NDB3V0	IO72PDB3V0	IO76PDB4V0	IO112PDB4V0				
F5	NC	NC	IO82PSB4V0	IO120PSB4V0				
F6	GFC1/IO49PDB3V0	GAC2/IO74PPB3V0	GAC2/IO83PPB4V0	GAC2/IO123PPB4V0				
F7	NC	IO09RSB0V0	IO04PPB0V0	IO05PPB0V1				
F8	NC	IO19RSB0V0	IO08NDB0V1	IO11NDB0V1				
F9	NC	NC	IO20PDB1V0	IO27PDB1V1				
F10	NC	IO29RSB0V0	IO23NDB1V1	IO37NDB1V2				
F11	NC	IO43NDB1V0	IO36NDB2V0	IO50NDB2V0				
F12	NC	IO43PDB1V0	IO36PDB2V0	IO50PDB2V0				
F13	NC	IO44NDB1V0	IO39NDB2V0	IO59NDB2V0				
F14	NC	GCA2/IO44PDB1V0	GCA2/IO39PDB2V0	GCA2/IO59PDB2V0				
F15	GCC1/IO34PDB1V0	GCB2/IO45PDB1V0	GCB2/IO40PDB2V0	GCB2/IO60PDB2V0				
F16	GCC0/IO34NDB1V0	IO45NDB1V0	IO40NDB2V0	IO60NDB2V0				
G1	GEC0/IO46NPB3V0	IO70NPB3V0	IO74NPB4V0	IO109NPB4V0				
G2	VCCIB3	VCCIB3	VCCIB4	VCCIB4				
G3	GEC1/IO46PPB3V0	GFB2/IO70PPB3V0	GFB2/IO74PPB4V0	GFB2/IO109PPB4V0				
G4	GFA1/IO47PDB3V0	GFA2/IO71PDB3V0	GFA2/IO75PDB4V0	GFA2/IO110PDB4V0				
G5	GND	GND	GND	GND				
G6	GFA0/IO47NDB3V0	IO71NDB3V0	IO75NDB4V0	IO110NDB4V0				
G7	GND	GND	GND	GND				
G8	VCC	VCC	VCC	VCC				
G9	GND	GND	GND	GND				
G10	VCC	VCC	VCC	VCC				
G11	GDA1/IO37NDB1V0	GCC0/IO47NDB1V0	GCC0/IO43NDB2V0	GCC0/IO62NDB2V0				
G12	GND	GND	GND	GND				
G13	IO37PDB1V0	GCC1/IO47PDB1V0	GCC1/IO43PDB2V0	GCC1/IO62PDB2V0				
G14	GCB0/IO35NPB1V0	IO46NPB1V0	IO41NPB2V0	IO61NPB2V0				
G15	VCCIB1	VCCIB1	VCCIB2	VCCIB2				
G16	GCB1/IO35PPB1V0	GCC2/IO46PPB1V0	GCC2/IO41PPB2V0	GCC2/IO61PPB2V0				
H1	GEB1/IO45PDB3V0	GFC2/IO69PDB3V0	GFC2/IO73PDB4V0	GFC2/IO108PDB4V0				
H2	GEB0/IO45NDB3V0	IO69NDB3V0	IO73NDB4V0	IO108NDB4V0				

Revision	Changes	Page				
Revision 3 (continued)	The "RC Oscillator" section was revised to correct a sentence that did not differentiate accuracy for commercial and industrial temperature ranges, which is given in Table 2-9 • Electrical Characteristics of RC Oscillator (SAR 33722).	2-19				
	Figure 2-57 • FIFO Read and Figure 2-58 • FIFO Write are new (SAR 34840).	2-72				
	The first paragraph of the "Offset" section was removed; it was intended to be replaced by the paragraph following it (SAR 22647).					
	IOL and IOH values for 3.3 V GTL+ and 2.5 V GTL+ were corrected in Table 2-86 • Summary of Maximum and Minimum DC Input and Output Levels Applicable to Commercial and Industrial Conditions (SAR 39813).	2-164				
	The drive strength, IOL, and IOH for 3.3 V GTL and 2.5 V GTL were changed from 25 mA to 20 mA in the following tables (SAR 37373):					
	Table 2-86 Summary of Maximum and Minimum DC Input and Output Levels Applicable to Commercial and Industrial Conditions,	2-164				
	Table 2-92 • Summary of I/O Timing Characteristics – Software Default Settings	2-167				
	Table 2-96 • I/O Output Buffer Maximum Resistances 1					
	Table 2-138 • Minimum and Maximum DC Input and Output Levels					
	Table 2-141 • Minimum and Maximum DC Input and Output Levels	2-200				
	The following sentence was deleted from the "2.5 V LVCMOS" section (SAR 34800): "It uses a 5 V–tolerant input buffer and push-pull output buffer."					
	Corrected the inadvertent error in maximum values for LVPECL VIH and VIL and revised them to "3.6" in Table 2-171 • Minimum and Maximum DC Input and Output Levels, making these consistent with Table 3-1 • Absolute Maximum Ratings, and Table 3-4 • Overshoot and Undershoot Limits 1 (SAR 37687).					
	The maximum frequency for global clock parameter was removed from Table 2-5 • AFS1500 Global Resource Timing through Table 2-8 • AFS090 Global Resource Timing because a frequency on the global is only an indication of what the global network can do. There are other limiters such as the SRAM, I/Os, and PLL. SmartTime software should be used to determine the design frequency (SAR 36955).	2-16 to 2-17				
Revision 2 (March 2012)	The phrase "without debug" was removed from the "Soft ARM Cortex-M1 Fusion Devices (M1)" section (SAR 21390).	I				
	The "In-System Programming (ISP) and Security" section, "Security" section, "Flash Advantages" section, and "Security" section were revised to clarify that although no existing security measures can give an absolute guarantee, Microsemi FPGAs implement the best security available in the industry (SAR 34679).	l, 1-2, 2-228				
	The Y security option and Licensed DPA Logo was added to the "Product Ordering Codes" section. The trademarked Licensed DPA Logo identifies that a product is covered by a DPA counter-measures license from Cryptography Research (SAR 34721).	III				
	The "Specifying I/O States During Programming" section is new (SAR 34693).	1-9				
	The following information was added before Figure 2-17 • XTLOSC Macro:					
	In the case where the Crystal Oscillator block is not used, the XTAL1 pin should be connected to GND and the XTAL2 pin should be left floating (SAR 24119).					
	Table 2-12 • Fusion CCC/PLL Specification was updated. A note was added indicating that when the CCC/PLL core is generated by Microsemi core generator software, not all delay values of the specified delay increments are available (SAR 34814).	2-28				