Microchip Technology - AFS600-2FGG484 Datasheet

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Understanding Embedded - FPGAs (Field Programmable Gate Array)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications,

Details	
Product Status	Active
Number of LABs/CLBs	-
Number of Logic Elements/Cells	-
Total RAM Bits	110592
Number of I/O	172
Number of Gates	600000
Voltage - Supply	1.425V ~ 1.575V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 85°C (TJ)
Package / Case	484-BGA
Supplier Device Package	484-FPBGA (23x23)
Purchase URL	https://www.e-xfl.com/product-detail/microchip-technology/afs600-2fgg484

Related Documents

Datasheet

Core8051

www.microsemi.com/soc/ipdocs/Core8051_DS.pdf

Application Notes

Fusion FlashROM

http://www.microsemi.com/soc/documents/Fusion_FROM_AN.pdf

Fusion SRAM/FIFO Blocks

http://www.microsemi.com/soc/documents/Fusion_RAM_FIFO_AN.pdf

Using DDR in Fusion Devices

http://www.microsemi.com/index.php?option=com_docman&task=doc_download&gid=129938

Fusion Security

http://www.microsemi.com/soc/documents/Fusion_Security_AN.pdf

Using Fusion RAM as Multipliers

http://www.microsemi.com/index.php?option=com_docman&task=doc_download&gid=129940

Handbook

Cortex-M1 Handbook

www.microsemi.com/soc/documents/CortexM1_HB.pdf

User Guides

Designer User Guide

http://www.microsemi.com/soc/documents/designer_UG.pdf

Fusion FPGA Fabric User Guide

http://www.microsemi.com/index.php?option=com_docman&task=doc_download&gid=130817

IGLOO, ProASIC3, SmartFusion and Fusion Macro Library Guide

http://www.microsemi.com/soc/documents/pa3_libguide_ug.pdf

SmartGen, FlashROM, Flash Memory System Builder, and Analog System Builder User Guide

http://www.microsemi.com/soc/documents/genguide_ug.pdf

White Papers

Fusion Technology

http://www.microsemi.com/soc/documents/Fusion_Tech_WP.pdf

Real-Time Counter System

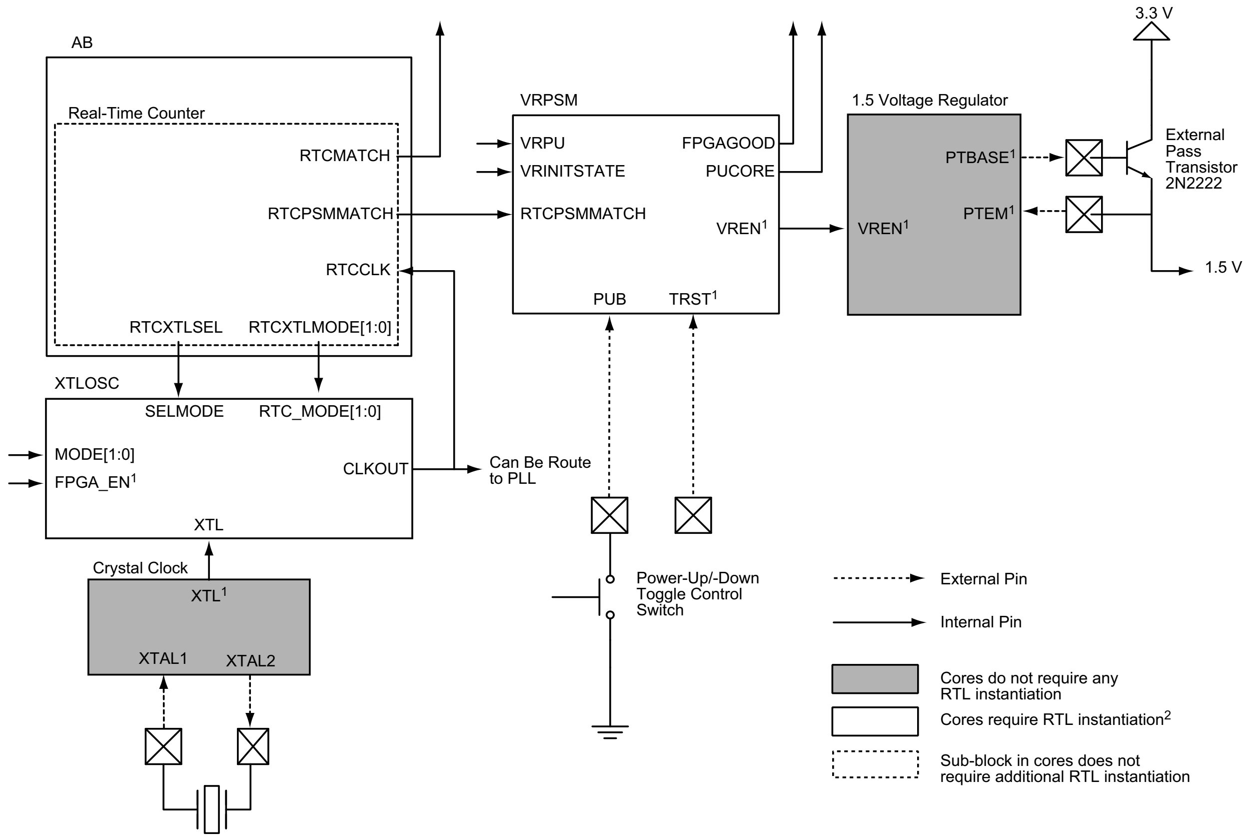
The RTC system enables Fusion devices to support standby and sleep modes of operation to reduce power consumption in many applications.

- Sleep mode, typical 10 µA
- Standby mode (RTC running), typical 3 mA with 20 MHz

The RTC system is composed of five cores:

- RTC sub-block inside Analog Block (AB)
- Voltage Regulator and Power System Monitor (VRPSM)
- Crystal oscillator (XTLOSC); refer to the "Crystal Oscillator" section in the Fusion Clock Resources chapter of the *Fusion FPGA Fabric User Guide* for more detail.
- Crystal clock; does not require instantiation in RTL
- 1.5 V voltage regulator; does not require instantiation in RTL

All cores are powered by 3.3 V supplies, so the RTC system is operational without a 1.5 V supply during standby mode. Figure 2-27 shows their connection.



Notes:

1. *Signals are hardwired internally and do not exist in the macro core.*
2. *User is only required to instantiate the VRPSM macro if the user wishes to specify PUPO behavior of the voltage regulator to be different from the default, or employ user logic to shut the voltage regulator off.*

***Figure 2-27* • Real-Time Counter System (not all the signals are shown for the AB macro)**

Table 2-25 • **Flash Memory Block Timing (continued)**
Commercial Temperature Range Conditions: T_J = 70°C, Worst-Case VCC = 1.425 V

Parameter	Description	–2	–1	Std.	Units
$t_{SUPGLOSSPRO}$	Page Loss Protect Setup Time for the Control Logic	1.69	1.93	2.27	ns
$t_{HDPGLOSSPRO}$	Page Loss Protect Hold Time for the Control Logic	0.00	0.00	0.00	ns
$t_{SUPGSTAT}$	Page Status Setup Time for the Control Logic	2.49	2.83	3.33	ns
$t_{HDPGSTAT}$	Page Status Hold Time for the Control Logic	0.00	0.00	0.00	ns
$t_{SUOVERWRPG}$	Over Write Page Setup Time for the Control Logic	1.88	2.14	2.52	ns
$t_{HDOVERWRPG}$	Over Write Page Hold Time for the Control Logic	0.00	0.00	0.00	ns
$t_{SULOCKREQUEST}$	Lock Request Setup Time for the Control Logic	0.87	0.99	1.16	ns
$t_{HDLOCKREQUEST}$	Lock Request Hold Time for the Control Logic	0.00	0.00	0.00	ns
$t_{RECARNVM}$	Reset Recovery Time	0.94	1.07	1.25	ns
$t_{REMARNVM}$	Reset Removal Time	0.00	0.00	0.00	ns
$t_{MPWARNVM}$	Asynchronous Reset Minimum Pulse Width for the Control Logic	10.00	12.50	12.50	ns
$t_{MPWCLKNVM}$	Clock Minimum Pulse Width for the Control Logic	4.00	5.00	5.00	ns
$t_{FMAXCLKNVM}$	Maximum Frequency for Clock for the Control Logic – for AFS1500/AFS600	80.00	80.00	80.00	MHz
	Maximum Frequency for Clock for the Control Logic – for AFS250/AFS090	100.00	80.00	80.00	MHz

FlashROM

Fusion devices have 1 kbit of on-chip nonvolatile flash memory that can be read from the FPGA core fabric. The FlashROM is arranged in eight banks of 128 bits during programming. The 128 bits in each bank are addressable as 16 bytes during the read-back of the FlashROM from the FPGA core (Figure 2-45).

The FlashROM can only be programmed via the IEEE 1532 JTAG port. It cannot be programmed directly from the FPGA core. When programming, each of the eight 128-bit banks can be selectively reprogrammed. The FlashROM can only be reprogrammed on a bank boundary. Programming involves an automatic, on-chip bank erase prior to reprogramming the bank. The FlashROM supports a synchronous read and can be read on byte boundaries. The upper three bits of the FlashROM address from the FPGA core define the bank that is being accessed. The lower four bits of the FlashROM address from the FPGA core define which of the 16 bytes in the bank is being accessed.

The maximum FlashROM access clock is given in Table 2-26 on page 2-54. Figure 2-46 shows the timing behavior of the FlashROM access cycle—the address has to be set up on the rising edge of the clock for DOUT to be valid on the next falling edge of the clock.

If the address is unchanged for two cycles:

- D0 becomes invalid t_{CK2Q} ns after the second rising edge of the clock.
- D0 becomes valid again t_{CK2Q} ns after the second falling edge.

If the address unchanged for three cycles:

- D0 becomes invalid t_{CK2Q} ns after the second rising edge of the clock.
- D0 becomes valid again t_{CK2Q} ns after the second falling edge.
- D0 becomes invalid t_{CK2Q} ns after the third rising edge of the clock.
- D0 becomes valid again t_{CK2Q} ns after the third falling edge.

The third part of the Analog Quad is called the Gate Driver Block, and its output pin is named AG. This section is used to drive an external FET. There are two modes available: a High Current Drive mode and a Current Source Control mode. Both negative and positive voltage polarities are available, and in the current source control mode, four different current levels are available.

The fourth section of the Analog Quad is called the Temperature Monitor Block, and its input pin name is AT. This block is similar to the Voltage Monitor Block, except that it has an additional function: it can be used to monitor the temperature of an external diode-connected transistor. It has a modified prescaler and is limited to positive voltages only.

The Analog Quad can be configured during design time by Libero SoC; however, the ACM can be used to change the parameters of any of these I/Os during runtime. This type of change is referred to as a context switch. The Analog Quad is a modular structure that is replicated to generate the analog I/O resources. Each Fusion device supports between 5 and 10 Analog Quads.

The analog pads are numbered to clearly identify both the type of pad (voltage, current, gate driver, or temperature pad) and its corresponding Analog Quad (AV0, AC0, AG0, AT0, AV1, …, AC9, AG9, and AT9). There are three types of input pads (AVx, ACx, and ATx) and one type of analog output pad (AGx). Since there can be up to 10 Analog Quads on a device, there can be a maximum of 30 analog input pads and 10 analog output pads.

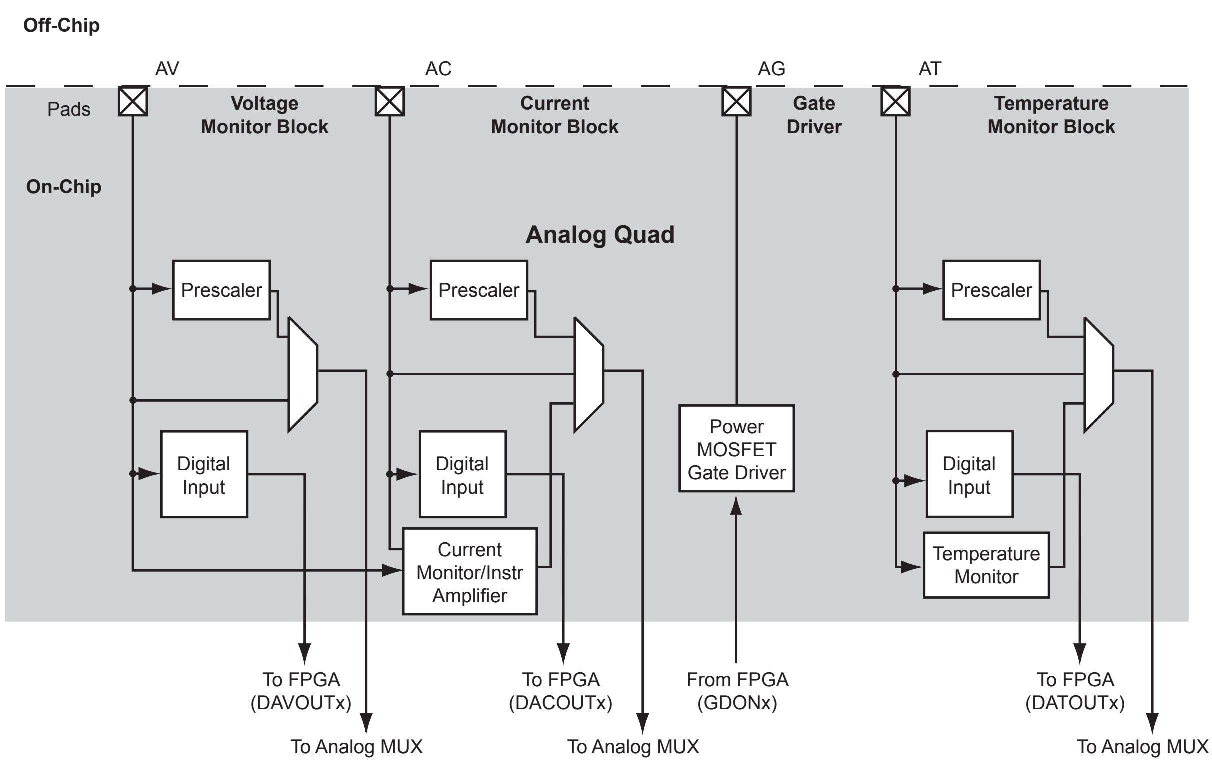


Figure 2-65 • **Analog Quad**

To initiate a current measurement, the appropriate Current Monitor Strobe (CMSTB) signal on the AB macro must be asserted low for at least t_{CMSLO} in order to discharge the previous measurement. Then CMSTB must be asserted high for at least t_{CMSET} prior to asserting the ADCSTART signal. The CMSTB must remain high until after the SAMPLE signal is de-asserted by the AB macro. Note that the minimum sample time cannot be less than t_{CMSHI}. Figure 2-71 shows the timing diagram of CMSTB in relationship with the ADC control signals.

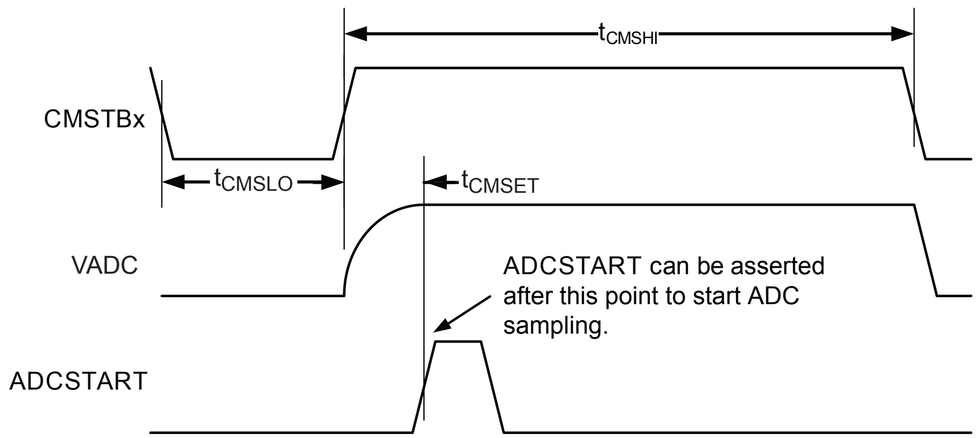


Figure 2-71 **• Timing Diagram for Current Monitor Strobe**

Figure 2-72 illustrates positive current monitor operation. The differential voltage between AV and AC goes into the 10× amplifier and is then converted by the ADC. For example, a current of 1.5 A is drawn from a 10 V supply and is measured by the voltage drop across a 0.050 Ω sense resistor, The voltage drop is amplified by ten times by the amplifier and then measured by the ADC. The 1.5 A current creates a differential voltage across the sense resistor of 75 mV. This becomes 750 mV after amplification. Thus, the ADC measures a current of 1.5 A as 750 mV. Using an ADC with 8-bit resolution and VAREF of 2.56 V, the ADC result is decimal 75. EQ 3 shows how to compute the current from the ADC result.

$$|I| = (ADC \times V_{AREF}) / (10 \times 2^N \times R_{sense})$$

EQ 3

where

I is the current flowing through the sense resistor

ADC is the result from the ADC

VAREF is the Reference voltage

N is the number of bits

Rsense is the resistance of the sense resistor

Fusion uses a remote diode as a temperature sensor. The Fusion Temperature Monitor uses a differential input; the AT pin and ATRTN (AT Return) pin are the differential inputs to the Temperature Monitor. There is one Temperature Monitor in each Quad. A simplified block diagram is shown in Figure 2-77.

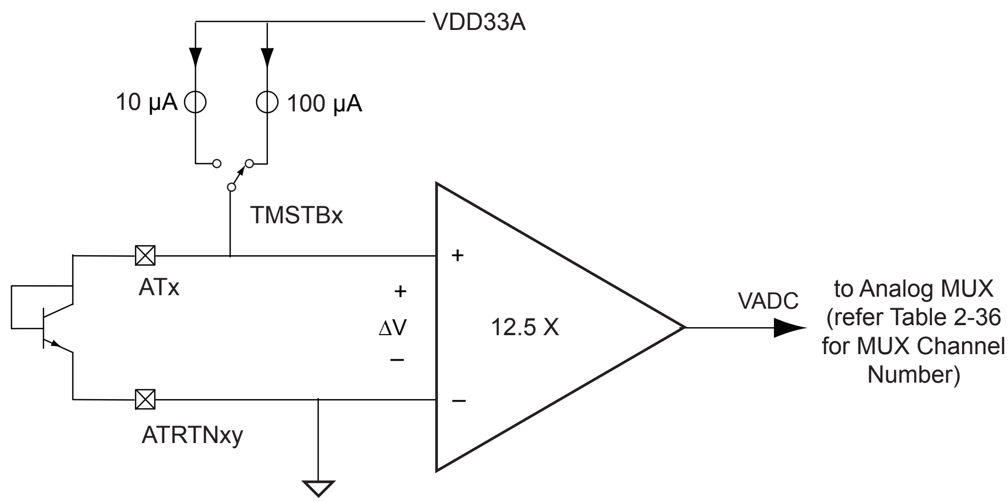


***Figure 2-77* • Block Diagram for Temperature Monitor Circuit**

The Fusion approach to measuring temperature is forcing two different currents through the diode with a ratio of 10:1. The switch that controls the different currents is controlled by the Temperature Monitor Strobe signal, TMSTB. Setting TMSTB to '1' will initiate a Temperature reading. The TMSTB should remain '1' until the ADC finishes sampling the voltage from the Temperature Monitor. The minimum sample time for the Temperature Monitor cannot be less than the minimum strobe high time minus the setup time. Figure 2-78 shows the timing diagram.

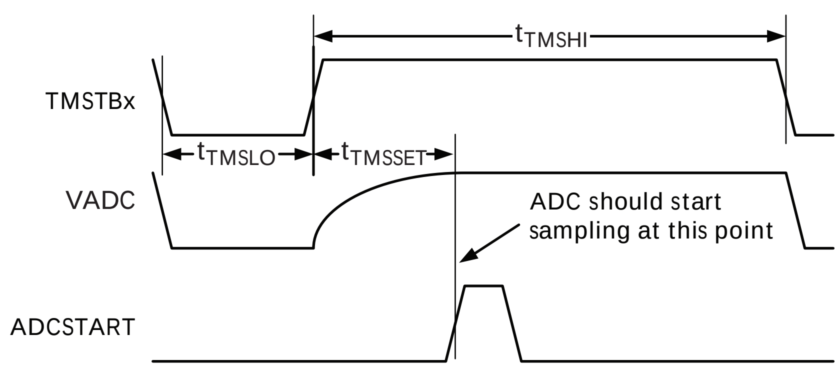


***Figure 2-78* • Timing Diagram for the Temperature Monitor Strobe Signal**

Note: When the IEEE 1149.1 Boundary Scan EXTEST instruction is executed, the AG pad drive strength ceases and becomes a 1 µA sink into the Fusion device.

ADC Description

The Fusion ADC is a 12-bit SAR ADC. It offers a wide variety of features for different use models. Figure 2-80 shows a block diagram of the Fusion ADC.

- Configurable resolution: 8-bit, 10-bit, and 12-bit mode
- DNL: 0.6 LSB for 10-bit mode
- INL: 0.4 LSB for 10-bit mode
- No missing code
- Internal VAREF = 2.56 V
- Maximum Sample Rate = 600 Ksps
- Power-up calibration and dynamic calibration after every sample to compensate for temperature drift over time

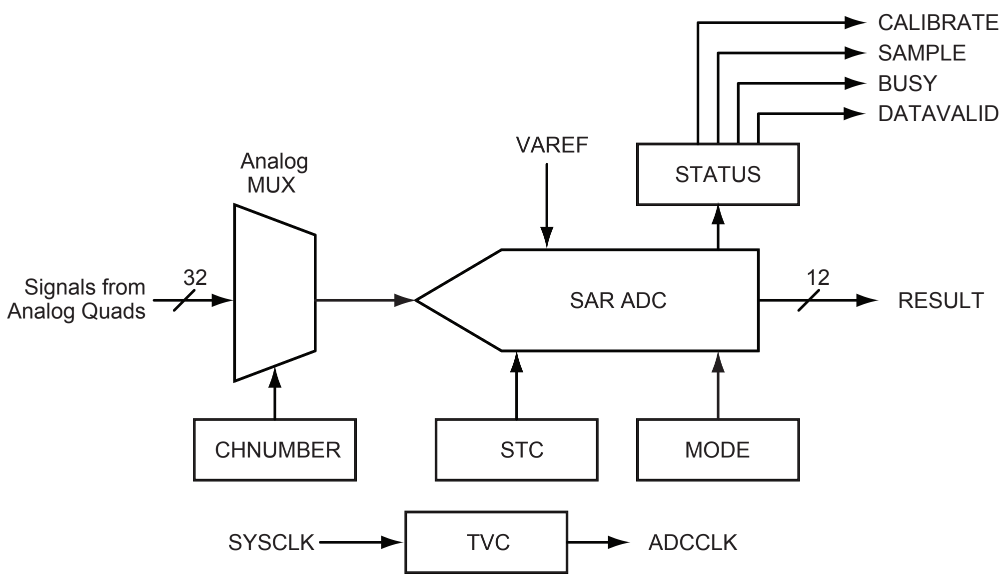


Figure 2-80 **• ADC Simplified Block Diagram**

ADC Theory of Operation

An analog-to-digital converter is used to capture discrete samples of a continuous analog voltage and provide a discrete binary representation of the signal. Analog-to-digital converters are generally characterized in three ways:

- Input voltage range
- Resolution
- Bandwidth or conversion rate

The input voltage range of an ADC is determined by its reference voltage (VREF). Fusion devices include an internal 2.56 V reference, or the user can supply an external reference of up to 3.3 V. The following examples use the internal 2.56 V reference, so the full-scale input range of the ADC is 0 to 2.56 V.

The resolution (LSB) of the ADC is a function of the number of binary bits in the converter. The ADC approximates the value of the input voltage using 2n steps, where n is the number of bits in the converter. Each step therefore represents VREF÷ 2n volts. In the case of the Fusion ADC configured for 12-bit operation, the LSB is 2.56 V / 4096 = 0.625 mV.

Finally, bandwidth is an indication of the maximum number of conversions the ADC can perform each second. The bandwidth of an ADC is constrained by its architecture and several key performance characteristics.

Integrated Voltage Reference

The Fusion device has an integrated on-chip 2.56 V reference voltage for the ADC. The value of this reference voltage was chosen to make the prescaling and postscaling factors for the prescaler blocks change in a binary fashion. However, if desired, an external reference voltage of up to 3.3 V can be connected between the VAREF and ADCGNDREF pins. The VAREFSEL control pin is used to select the reference voltage.

Table 2-42 • VAREF Bit Function

Name	Bit	Function
VAREF	0	Reference voltage selection 0 – Internal voltage reference selected. VAREF pin outputs 2.56 V. 1 – Input external voltage reference from VAREF and ADCGNDREF

ADC Clock

The speed of the ADC depends on its internal clock, ADCCLK, which is not accessible to users. The ADCCLK is derived from SYSCLK. Input signal TVC[7:0], Time Divider Control, determines the speed of the ADCCLK in relationship to SYSCLK, based on EQ 15.

$$t_{ADCCLK} = 4 \times (1 + TVC) \times t_{SYSCLK}$$

EQ 15

TVC: Time Divider Control (0–255)

t_{ADCCLK} is the period of ADCCLK, and must be between 0.5 MHz and 10 MHz

t_{SYSCLK} is the period of SYSCLK

Table 2-43 • TVC Bits Function

Name	Bits	Function
TVC	[7:0]	SYSCLK divider control

The frequency of ADCCLK, f_{ADCCLK}, must be within 0.5 Hz to 10 MHz.

The inputs to the ADC are synchronized to SYSCLK. A conversion is initiated by asserting the ADCSTART signal on a rising edge of SYSCLK. Figure 2-90 on page 2-112 and Figure 2-91 on page 2-112 show the timing diagram for the ADC.

Acquisition Time or Sample Time Control

Acquisition time (t_{SAMPLE}) specifies how long an analog input signal has to charge the internal capacitor array. Figure 2-88 shows a simplified internal input sampling mechanism of a SAR ADC.

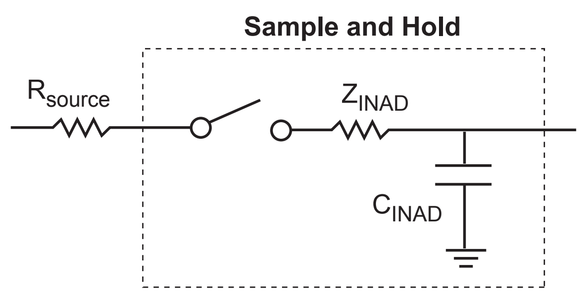


Figure 2-88 • Simplified Sample and Hold Circuitry

The internal impedance (Z_{INAD}), external source resistance (R_{SOURCE}), and sample capacitor (C_{INAD}) form a simple RC network. As a result, the accuracy of the ADC can be affected if the ADC is given insufficient time to charge the capacitor. To resolve this problem, you can either reduce the source resistance or increase the sampling time by changing the acquisition time using the STC signal.

Analog System Characteristics

***Table 2-49* • Analog Channel Specifications**
Commercial Temperature Range Conditions, T_J = 85°C (unless noted otherwise),
Typical: VCC33A = 3.3 V, VCC = 1.5 V

Parameter	Description	Condition	Min.	Typ.	Max.	Units
Voltage Monitor Using Analog Pads AV, AC and AT (using prescaler)						
	Input Voltage (Prescaler)	Refer to Table 3-2 on page 3-3				
VINAP	Uncalibrated Gain and Offset Errors	Refer to Table 2-51 on page 2-122				
	Calibrated Gain and Offset Errors	Refer to Table 2-52 on page 2-123				
	Bandwidth1				100	KHz
	Input Resistance	Refer to Table 3-3 on page 3-4				
	Scaling Factor	Prescaler modes (Table 2-57 on page 2-130)				
	Sample Time		10			µs
Current Monitor Using Analog Pads AV and AC						
VRSM[1]	Maximum Differential Input Voltage				VAREF / 10	mV
	Resolution	Refer to "Current Monitor" section				
	Common Mode Range				– 10.5 to +12	V
CMRR	Common Mode Rejection Ratio	DC – 1 KHz		60		dB
		1 KHz - 10 KHz		50		dB
		> 10 KHz		30		dB
t_{CMSHI}	Strobe High time		ADC conv. time		200	µs
t_{CMSHI}	Strobe Low time		5			µs
t_{CMSHI}	Settling time		0.02			µs
	Accuracy	Input differential voltage > 50 mV			–2 –(0.05 x VRSM) to +2 + (0.05 x VRSM)	mV

Notes:

1. *VRSM is the maximum voltage drop across the current sense resistor.*
2. *Analog inputs used as digital inputs can tolerate the same voltage limits as the corresponding analog pad. There is no reliability concern on digital inputs as long as VIND does not exceed these limits.*
3. *VIND is limited to VCC33A + 0.2 to allow reaching 10 MHz input frequency.*
4. *An averaging of 1,024 samples (LPF setting in Analog System Builder) is required and the maximum capacitance allowed across the AT pins is 500 pF.*
5. *The temperature offset is a fixed positive value.*
6. *The high current mode has a maximum power limit of 20 mW. Appropriate current limit resistors must be used, based on voltage on the pad.*
7. *When using SmartGen Analog System Builder, CalibIP is required to obtain specified offset. For further details on CalibIP, refer to the "Temperature, Voltage, and Current Calibration in Fusion FPGAs" chapter of the Fusion FPGA Fabric User Guide.*

Similarly,

Min. Output Voltage = (Max. Negative input offset) + (Input Voltage x Max. Negative Channel Gain) = (–88 mV) + (5 V x 0.96) = **4.712 V**

Calculating Accuracy for a Calibrated Analog Channel

Formula

For a given prescaler range, EQ 31 gives the output voltage.

Output Voltage = Channel Error in V + Input Voltage

EQ 31

where

Channel Error in V = Total Channel Error in LSBs x Equivalent voltage per LSB

Example

Input Voltage = 5 V

Chosen Prescaler range = 8 V range

Refer to Table 2-52 on page 2-123.

Max. Output Voltage = Max. Positive Channel Error in V + Input Voltage

Max. Positive Channel Error in V = (6 LSB) × (8 mV per LSB in 10-bit mode) = 48 mV

Max. Output Voltage = 48 mV + 5 V = **5.048 V**

Similarly,

Min. Output Voltage = Max. Negative Channel Error in V + Input Voltage = (–48 mV) + 5 V = **4.952 V**

Calculating LSBs from a Given Error Budget

Formula

For a given prescaler range,

LSB count = ± (Input Voltage × Required % error) / (Equivalent voltage per LSB)

Example

Input Voltage = 3.3 V

Required error margin= 1%

Refer to Table 2-52 on page 2-123.

Equivalent voltage per LSB = 16 mV for a 16V prescaler, with ADC in 10-bit mode

LSB Count = ± (5.0 V × 1%) / (0.016)

LSB Count = **± 3.125**

Equivalent voltage per LSB = **8 mV** for an 8 V prescaler, with ADC in 10-bit mode

LSB Count = ± (5.0 V × 1%) / (0.008)

LSB Count = **± 6.25**

The 8 V prescaler satisfies the calculated LSB count accuracy requirement (see Table 2-52 on page 2-123).

User I/O Naming Convention

Due to the comprehensive and flexible nature of Fusion device user I/Os, a naming scheme is used to show the details of the I/O (Figure 2-113 on page 2-158 and Figure 2-114 on page 2-159). The name identifies to which I/O bank it belongs, as well as the pairing and pin polarity for differential I/Os.

I/O Nomenclature = Gmn/IOuxwByVz

Gmn is only used for I/Os that also have CCC access—i.e., global pins.

G = Global

m = Global pin location associated with each CCC on the device: A (northwest corner), B (northeast corner), C (east middle), D (southeast corner), E (southwest corner), and F (west middle).

n = Global input MUX and pin number of the associated Global location m, either A0, A1, A2, B0, B1, B2, C0, C1, or C2. Figure 2-22 on page 2-25 shows the three input pins per clock source MUX at CCC location m.

u = I/O pair number in the bank, starting at 00 from the northwest I/O bank and proceeding in a clockwise direction.

x = P (Positive) or N (Negative) for differential pairs, or R (Regular – single-ended) for the I/Os that support single-ended and voltage-referenced I/O standards only. U (Positive-LVDS only) or V (Negative-LVDS only) restrict the I/O differential pair from being selected as an LVPECL pair.

w = D (Differential Pair), P (Pair), or S (Single-Ended). D (Differential Pair) if both members of the pair are bonded out to adjacent pins or are separated only by one GND or NC pin; P (Pair) if both members of the pair are bonded out but do not meet the adjacency requirement; or S (Single-Ended) if the I/O pair is not bonded out. For Differential (D) pairs, adjacency for ball grid packages means only vertical or horizontal. Diagonal adjacency does not meet the requirements for a true differential pair.

B = Bank

y = Bank number (0–3). The Bank number starts at 0 from the northwest I/O bank and proceeds in a clockwise direction.

V = Reference voltage

z = Minibank number

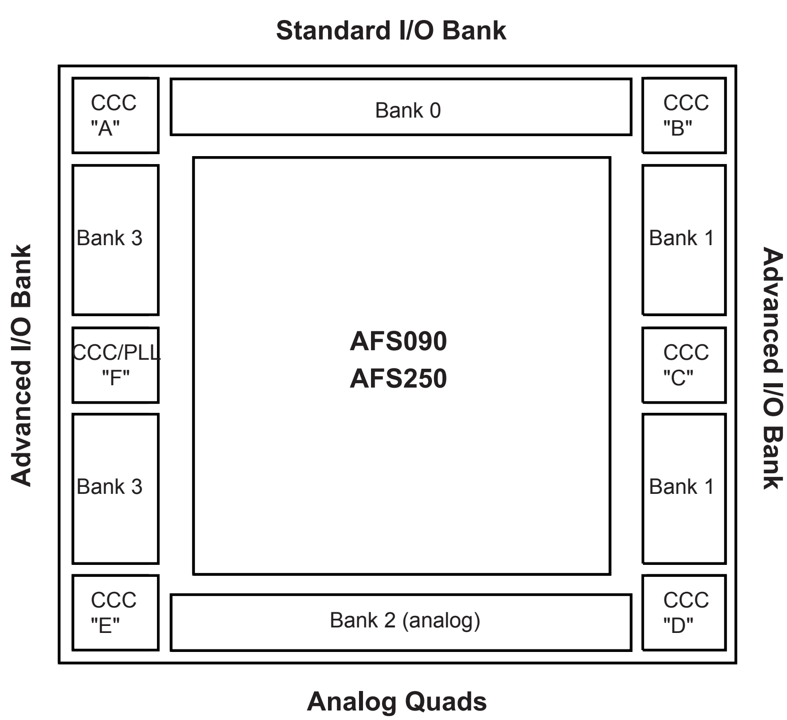


***Figure 2-113* • Naming Conventions of Fusion Devices with Three Digital I/O Banks**

***Table 2-93* • Summary of I/O Timing Characteristics – Software Default Settings**
Commercial Temperature Range Conditions: T_J = 70°C, Worst-Case VCC = 1.425 V,
Worst-Case VCCI = I/O Standard Dependent
Applicable to Advanced I/Os

I/O Standard	Drive Strength (mA)	Slew Rate	Capacitive Load (pF)	External Resistor (Ohm)	t_{DOUT}	t_{DP}	t_{DIN}	t_{PY}	t_{EOUT}	t_{ZL}	t_{ZH}	t_{LZ}	t_{HZ}	t_{ZLS}	t_{ZHS}	Units
3.3 V LVTTL/ 3.3 V LVCMOS	12 mA	High	35 pF	–	0.49	2.64	0.03	0.90	0.32	2.69	2.11	2.40	2.68	4.36	3.78	ns
2.5 V LVCMOS	12 mA	High	35 pF	–	0.49	2.66	0.03	0.98	0.32	2.71	2.56	2.47	2.57	4.38	4.23	ns
1.8 V LVCMOS	12 mA	High	35 pF	–	0.49	2.64	0.03	0.91	0.32	2.69	2.27	2.76	3.05	4.36	3.94	ns
1.5 V LVCMOS	12 mA	High	35 pF	–	0.49	3.05	0.03	1.07	0.32	3.10	2.67	2.95	3.14	4.77	4.34	ns
3.3 V PCI	Per PCI spec	High	10 pF	25 [2]	0.49	2.00	0.03	0.65	0.32	2.04	1.46	2.40	2.68	3.71	3.13	ns
3.3 V PCI-X	Per PCI-X spec	High	10 pF	25 [2]	0.49	2.00	0.03	0.62	0.32	2.04	1.46	2.40	2.68	3.71	3.13	ns
LVDS	24 mA	High	–	–	0.49	1.37	0.03	1.20	N/A	N/A	N/A	N/A	N/A	N/A	N/A	ns
LVPECL	24 mA	High	–	–	0.49	1.34	0.03	1.05	N/A	N/A	N/A	N/A	N/A	N/A	N/A	ns

Notes:

1. *For specific junction temperature and voltage-supply levels, refer to Table 3-6 on page 3-7 for derating values.*
2. *Resistance is used to measure I/O propagation delays as defined in PCI specifications. See Figure 2-123 on page 2-197 for connectivity. This resistor is not required during normal operation.*

***Table 2-94* • Summary of I/O Timing Characteristics – Software Default Settings**
Commercial Temperature Range Conditions: T_J = 70°C, Worst-Case VCC = 1.425 V,
Worst-Case VCCI = I/O Standard Dependent
Applicable to Standard I/Os

I/O Standard	Drive Strength (mA)	Slew Rate	Capacitive Load (pF)	External Resistor (Ohm)	t_{DOUT}	t_{DP}	t_{DIN}	t_{PY}	t_{EOUT}	t_{ZL}	t_{ZH}	t_{LZ}	t_{HZ}	Units
3.3 V LVTTL/ 3.3 V LVCMOS	8 mA	High	35 pF	–	0.49	3.29	0.03	0.75	0.32	3.36	2.80	1.79	2.01	ns
2.5 V LVCMOS	8 mA	High	35pF	–	0.49	3.56	0.03	0.96	0.32	3.40	3.56	1.78	1.91	ns
1.8 V LVCMOS	4 mA	High	35pF	–	0.49	4.74	0.03	0.90	0.32	4.02	4.74	1.80	1.85	ns
1.5 V LVCMOS	2 mA	High	35pF	–	0.49	5.71	0.03	1.06	0.32	4.71	5.71	1.83	1.83	ns

Note: For specific junction temperature and voltage-supply levels, refer to Table 3-6 on page 3-7 for derating values.

connected to the internal core logic I/O tile and the input, output, and control ports of an I/O buffer to capture and load data into the register to control or observe the logic state of each I/O.

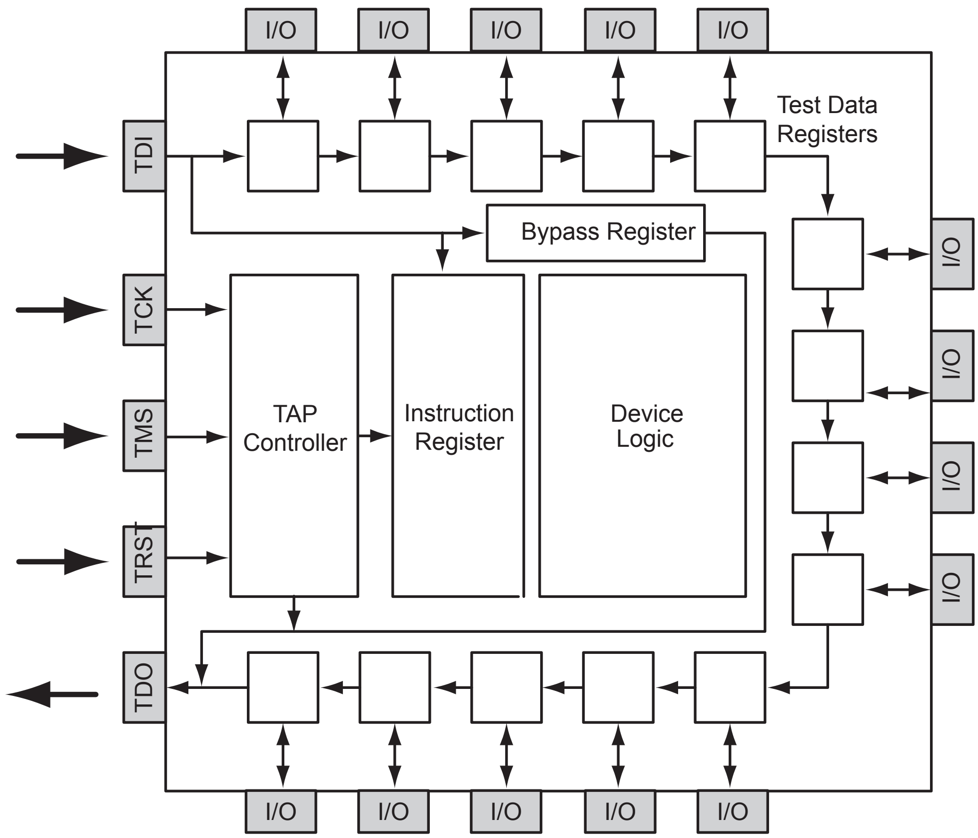


Figure 2-146 **• Boundary Scan Chain in Fusion**

Table 2-185 **• Boundary Scan Opcodes**

	Hex Opcode
EXTEST	00
HIGHZ	07
USERCODE	0E
SAMPLE/PRELOAD	01
IDCODE	0F
CLAMP	05
BYPASS	FF

The 1.76 W power is less than the required 3.00 W. The design therefore requires a heat sink, or the airflow where the device is mounted should be increased. The design's total junction-to-air thermal resistance requirement can be estimated by EQ 7:

$$\theta_{ja(total)} = \frac{T_J - T_A}{P} = \frac{100°C - 70°C}{3.00\ W} = 10.00°C/W$$

EQ 7

Determining the heat sink's thermal performance proceeds as follows:

$$\theta_{JA(TOTAL)} = \theta_{JC} + \theta_{CS} + \theta_{SA}$$

EQ 8

where

θ_{JA}	=	0.37°C/W
	=	Thermal resistance of the interface material between the case and the heat sink, usually provided by the thermal interface manufacturer
θ_{SA}	=	Thermal resistance of the heat sink in °C/W

$$\theta_{SA} = \theta_{JA(TOTAL)} - \theta_{JC} - \theta_{CS}$$

EQ 9

$$\theta_{SA} = 13.33°C/W - 8.28°C/W - 0.37°C/W = 5.01°C/W$$

A heat sink with a thermal resistance of 5.01°C/W or better should be used. Thermal resistance of heat sinks is a function of airflow. The heat sink performance can be significantly improved with increased airflow.

Carefully estimating thermal resistance is important in the long-term reliability of an Microsemi FPGA. Design engineers should always correlate the power consumption of the device with the maximum allowable power dissipation of the package selected for that device.

Note: The junction-to-air and junction-to-board thermal resistances are based on JEDEC standard (JESD-51) and assumptions made in building the model. It may not be realized in actual application and therefore should be used with a degree of caution. Junction-to-case thermal resistance assumes that all power is dissipated through the case.

Temperature and Voltage Derating Factors

***Table 3-7 •* Temperature and Voltage Derating Factors for Timing Delays (normalized to T_J = 70°C, Worst-Case VCC = 1.425 V)**

Array Voltage VCC (V)	Junction Temperature (°C)					
	–40°C	**0°C**	**25°C**	**70°C**	**85°C**	**100°C**
1.425	0.88	0.93	0.95	1.00	1.02	1.05
1.500	0.83	0.88	0.90	0.95	0.96	0.99
1.575	0.80	0.85	0.87	0.91	0.93	0.96

Table 3-9 • AFS600 Quiescent Supply Current Characteristics (continued)

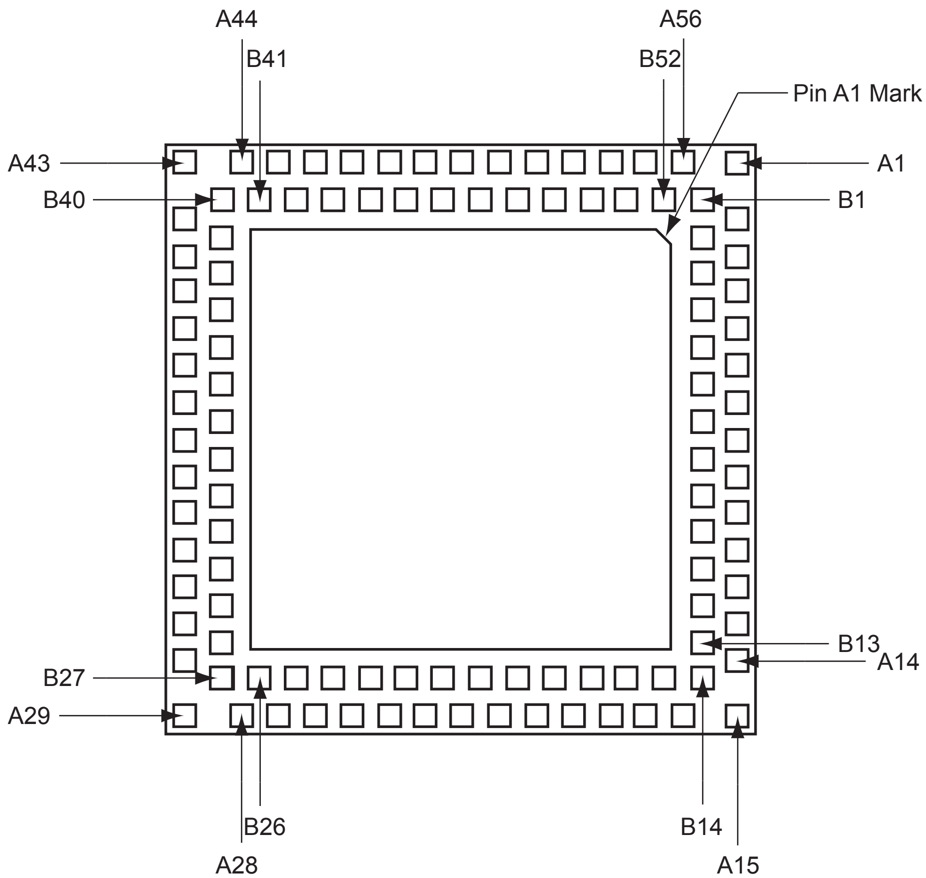
Parameter	Description	Conditions	Temp.	Min	Typ	Max	Unit
IPP	Programming supply current	Non-programming mode, VPUMP = 3.63 V	T_J = 25°C		36	80	µA
			T_J = 85°C		36	80	µA
			T_J = 100°C		36	80	µA
		Standby mode[5] or Sleep mode[6], VPUMP = 0 V			0	0	µA
ICCNVM	Embedded NVM current	Reset asserted, VCCNVM = 1.575 V	T_J = 25°C		22	80	µA
			T_J = 85°C		24	80	µA
			T_J = 100°C		25	80	µA
ICCPLL	1.5 V PLL quiescent current	Operational standby, VCCPLL = 1.575 V	T_J = 25°C		130	200	µA
			T_J = 85°C		130	200	µA
			T_J = 100°C		130	200	µA

Notes:

1. *ICC is the 1.5 V power supplies, ICC and ICC15A.*
2. *ICC33A includes ICC33A, ICC33PMP, and ICCOSC.*
3. *ICCI includes all ICCI0, ICCI1, ICCI2, and ICCI4.*
4. *Operational standby is when the Fusion device is powered up, all blocks are used, no I/O is toggling, Voltage Regulator is loaded with 200 mA, VCC33PMP is ON, XTAL is ON, and ADC is ON.*
5. *XTAL is configured as high gain, VCC = VJTAG = VPUMP = 0 V.*
6. *Sleep Mode, VCC = VJTAG = VPUMP = 0 V.*

4 – Package Pin Assignments

QN108

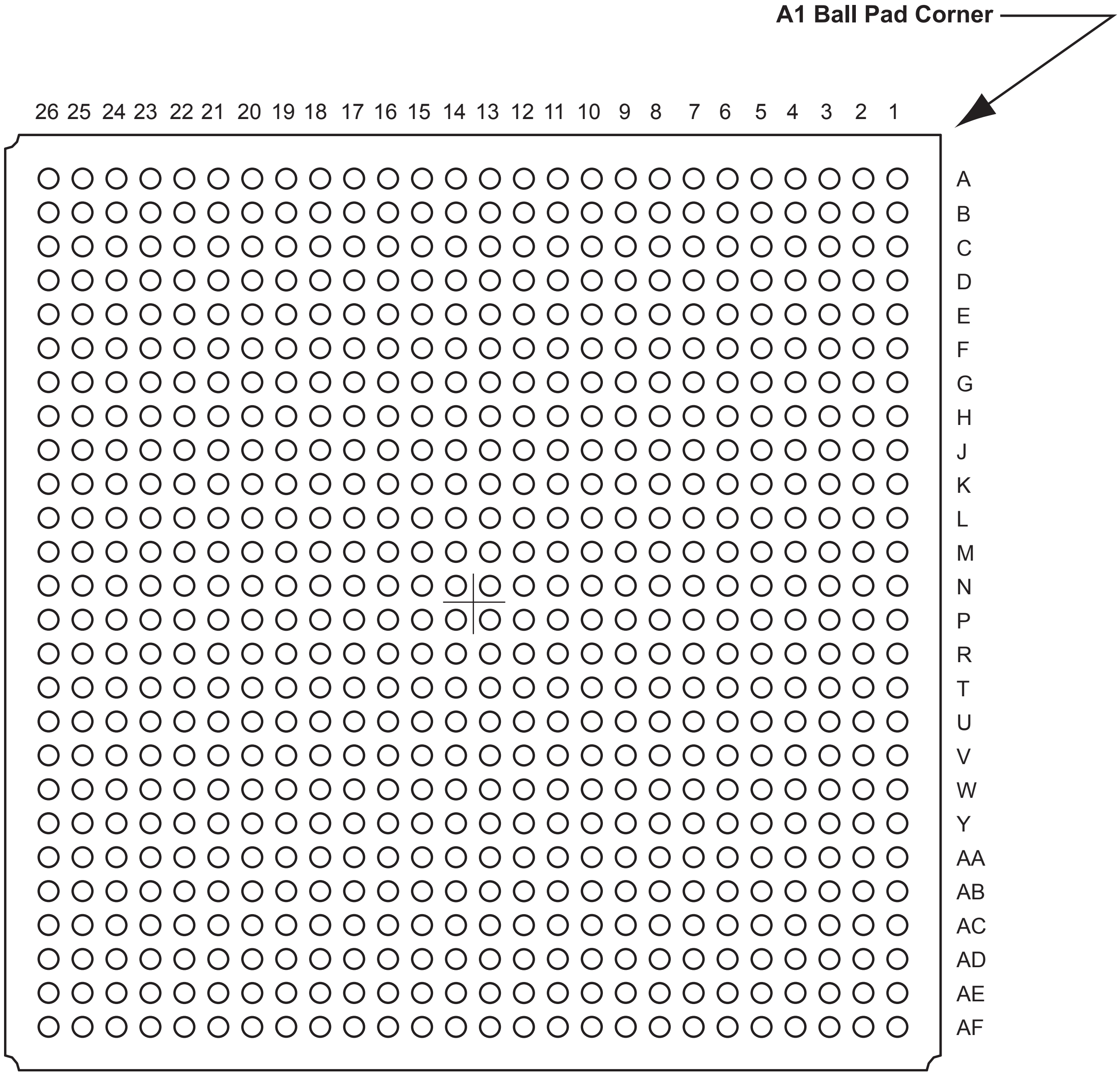


Note: The die attach paddle center of the package is tied to ground (GND).

Note

For Package Manufacturing and Environmental information, visit the Resource Center at http://www.microsemi.com/soc/products/solutions/package/default.aspx.

FG676



Note

For Package Manufacturing and Environmental information, visit the Resource Center at http://www.microsemi.com/soc/products/solutions/package/default.aspx.

Revision	Changes	Page
Revision 3 (continued)	The "RC Oscillator" section was revised to correct a sentence that did not differentiate accuracy for commercial and industrial temperature ranges, which is given in Table 2-9 • Electrical Characteristics of RC Oscillator (SAR 33722).	2-19
	Figure 2-57 • FIFO Read and Figure 2-58 • FIFO Write are new (SAR 34840).	2-72
	The first paragraph of the "Offset" section was removed; it was intended to be replaced by the paragraph following it (SAR 22647).	2-95
	IOL and IOH values for 3.3 V GTL+ and 2.5 V GTL+ were corrected in Table 2-86 • Summary of Maximum and Minimum DC Input and Output Levels Applicable to Commercial and Industrial Conditions (SAR 39813).	2-164
	The drive strength, IOL, and IOH for 3.3 V GTL and 2.5 V GTL were changed from 25 mA to 20 mA in the following tables (SAR 37373): Table 2-86 • Summary of Maximum and Minimum DC Input and Output Levels Applicable to Commercial and Industrial Conditions, Table 2-92 • Summary of I/O Timing Characteristics – Software Default Settings Table 2-96 • I/O Output Buffer Maximum Resistances 1 Table 2-138 • Minimum and Maximum DC Input and Output Levels Table 2-141 • Minimum and Maximum DC Input and Output Levels	 2-164 2-167 2-169 2-199 2-200
	The following sentence was deleted from the "2.5 V LVCMOS" section (SAR 34800): "It uses a 5 V–tolerant input buffer and push-pull output buffer."	2-181
	Corrected the inadvertent error in maximum values for LVPECL VIH and VIL and revised them to "3.6" in Table 2-171 • Minimum and Maximum DC Input and Output Levels, making these consistent with Table 3-1 • Absolute Maximum Ratings, and Table 3-4 • Overshoot and Undershoot Limits 1 (SAR 37687).	2-211
	The maximum frequency for global clock parameter was removed from Table 2-5 • AFS1500 Global Resource Timing through Table 2-8 • AFS090 Global Resource Timing because a frequency on the global is only an indication of what the global network can do. There are other limiters such as the SRAM, I/Os, and PLL. SmartTime software should be used to determine the design frequency (SAR 36955).	2-16 to 2-17
Revision 2 (March 2012)	The phrase "without debug" was removed from the "Soft ARM Cortex-M1 Fusion Devices (M1)" section (SAR 21390).	I
	The "In-System Programming (ISP) and Security" section, "Security" section, "Flash Advantages" section, and "Security" section were revised to clarify that although no existing security measures can give an absolute guarantee, Microsemi FPGAs implement the best security available in the industry (SAR 34679).	I, 1-2, 2-228
	The Y security option and Licensed DPA Logo was added to the "Product Ordering Codes" section. The trademarked Licensed DPA Logo identifies that a product is covered by a DPA counter-measures license from Cryptography Research (SAR 34721).	III
	The "Specifying I/O States During Programming" section is new (SAR 34693).	1-9
	The following information was added before Figure 2-17 • XTLOSC Macro: In the case where the Crystal Oscillator block is not used, the XTAL1 pin should be connected to GND and the XTAL2 pin should be left floating (SAR 24119).	2-20
	Table 2-12 • Fusion CCC/PLL Specification was updated. A note was added indicating that when the CCC/PLL core is generated by Microsemi core generator software, not all delay values of the specified delay increments are available (SAR 34814).	2-28

Revision	Changes	Page
Advance v1.5 (continued)	This bullet was added to the "Integrated A/D Converter (ADC) and Analog I/O" section: ADC Accuracy is Better than 1%	I
	In the "Integrated Analog Blocks and Analog I/Os" section, ±4 LSB was changed to 0.72. The following sentence was deleted: The input range for voltage signals is from –12 V to +12 V with full-scale output values from 0.125 V to 16 V. In addition, 2°C was changed to 3°C: "One analog input in each quad can be connected to an external temperature monitor diode and achieves detection accuracy of ±3ºC." The following sentence was deleted: The input range for voltage signals is from –12 V to +12 V with full-scale output values from 0.125 V to 16 V.	1-4
	The title of the datasheet changed from Actel Programmable System Chips to Actel Fusion Mixed Signal FPGAs. In addition, all instances of programmable system chip were changed to mixed signal FPGA.	N/A
Advance v1.4 (July 2008)	In Table 3-8 · Quiescent Supply Current Characteristics (IDDQ)1, footnote references were updated for I_{DC2} and I_{DC3}. Footnote 3 and 4 were updated and footnote 5 is new.	3-11
Advance v1.3 (July 2008)	The "ADC Description" section was significantly updated. Please review carefully.	2-102
Advance v1.2 (May 2008)	Table 2-25 • Flash Memory Block Timing was significantly updated.	2-55
	The "V_{AREF} Analog Reference Voltage" pin description section was significantly update. Please review it carefully.	2-226
	Table 2-45 • ADC Interface Timing was significantly updated.	2-110
	Table 2-56 • Direct Analog Input Switch Control Truth Table—AV (x = 0), AC (x = 1), and AT (x = 3) was significantly updated.	2-131
	The following sentence was deleted from the "Voltage Monitor" section: The Analog Quad inputs are tolerant up to 12 V + 10%.	2-86
	The "180-Pin QFN" figure was updated. D1 to D4 are new and the figure was changed to bottom view. The note below the figure is new.	3-3
Advance v1.1 (May 2008)	The following text was incorrect and therefore deleted: **VCC33A** **Analog Power Filter** Analog power pin for the analog power supply low-pass filter. An external 100 pF capacitor should be connected between this pin and ground. There is still a description of V_{CC33A} on page 2-224.	2-204

Revision	Changes	Page
Advance v1.0 (January 2008)	All Timing Characteristics tables were updated. For the Differential I/O Standards, the Standard I/O support tables are new.	N/A
	Table 2-3 • Array Coordinates was updated to change the max x and y values	2-9
	Table 2-12 • Fusion CCC/PLL Specification was updated.	2-31
	A note was added to Table 2-16 · RTC ACM Memory Map.	2-37
	A reference to the Peripheral's User's Guide was added to the "Voltage Regulator Power Supply Monitor (VRPSM)" section.	2-42
	In Table 2-25 • Flash Memory Block Timing, the commercial conditions were updated.	2-55
	In Table 2-26 • FlashROM Access Time, the commercial conditions were missing and have been added below the title of the table.	2-58
	In Table 2-36 • Analog Block Pin Description, the function description was updated for the ADCRESET.	2-82
	In the "Voltage Monitor" section, the following sentence originally had ± 10% and it was changed to +10%. The Analog Quad inputs are tolerant up to 12 V + 10%. In addition, this statement was deleted from the datasheet: Each I/O will draw power when connected to power (3 mA at 3 V).	2-86
	The "Terminology" section is new.	2-88
	The "Current Monitor" section was significantly updated. Figure 2-72 • Timing Diagram for Current Monitor Strobe to Figure 2-74 • Negative Current Monitor and Table 2-37 • Recommended Resistor for Different Current Range Measurement are new.	2-90
	The "ADC Description" section was updated to add the "Terminology" section.	2-93
	In the "Gate Driver" section, 25 mA was changed to 20 mA and 1.5 MHz was changed to 1.3 MHz. In addition, the following sentence was deleted: The maximum AG pad switching frequency is 1.25 MHz.	2-94
	The "Temperature Monitor" section was updated to rewrite most of the text and add Figure 2-78, Figure 2-79, and Table 2-38 • Temperature Data Format.	2-96
	In Table 2-38 • Temperature Data Format, the temperature K column was changed for 85°C from 538 to 358.	2-98
	In Table 2-45 • ADC Interface Timing, "Typical-Case" was changed to "Worst-Case."	2-110
	The "ADC Interface Timing" section is new.	2-110
	Table 2-46 • Analog Channel Specifications was updated.	2-118
	The "V_{CC15A} Analog Power Supply (1.5 V)" section was updated.	2-224
	The "$V_{CCPLA/B}$ PLL Supply Voltage" section is new.	2-225
	In "V_{CCNVM} Flash Memory Block Power Supply (1.5 V)" section, supply was changed to supply input.	2-224
	The "$V_{CCPLA/B}$ PLL Supply Voltage" pin description was updated to include the following statement: Actel recommends tying VCCPLX to VCC and using proper filtering circuits to decouple V_{CC} noise from PLL.	2-225
	The "$V_{COMPLA/B}$ Ground for West and East PLL" section was updated.	2-225