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### Understanding [Embedded - FPGAs \(Field Programmable Gate Array\)](#)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

### Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications,

#### Details

Product Status	Active
Number of LABs/CLBs	-
Number of Logic Elements/Cells	-
Total RAM Bits	276480
Number of I/O	252
Number of Gates	1500000
Voltage - Supply	1.425V ~ 1.575V
Mounting Type	Surface Mount
Operating Temperature	-40°C ~ 100°C (TJ)
Package / Case	676-BGA
Supplier Device Package	676-FBGA (27x27)
Purchase URL	<a href="https://www.e-xfl.com/product-detail/microchip-technology/m1afs1500-1fgg676i">https://www.e-xfl.com/product-detail/microchip-technology/m1afs1500-1fgg676i</a>

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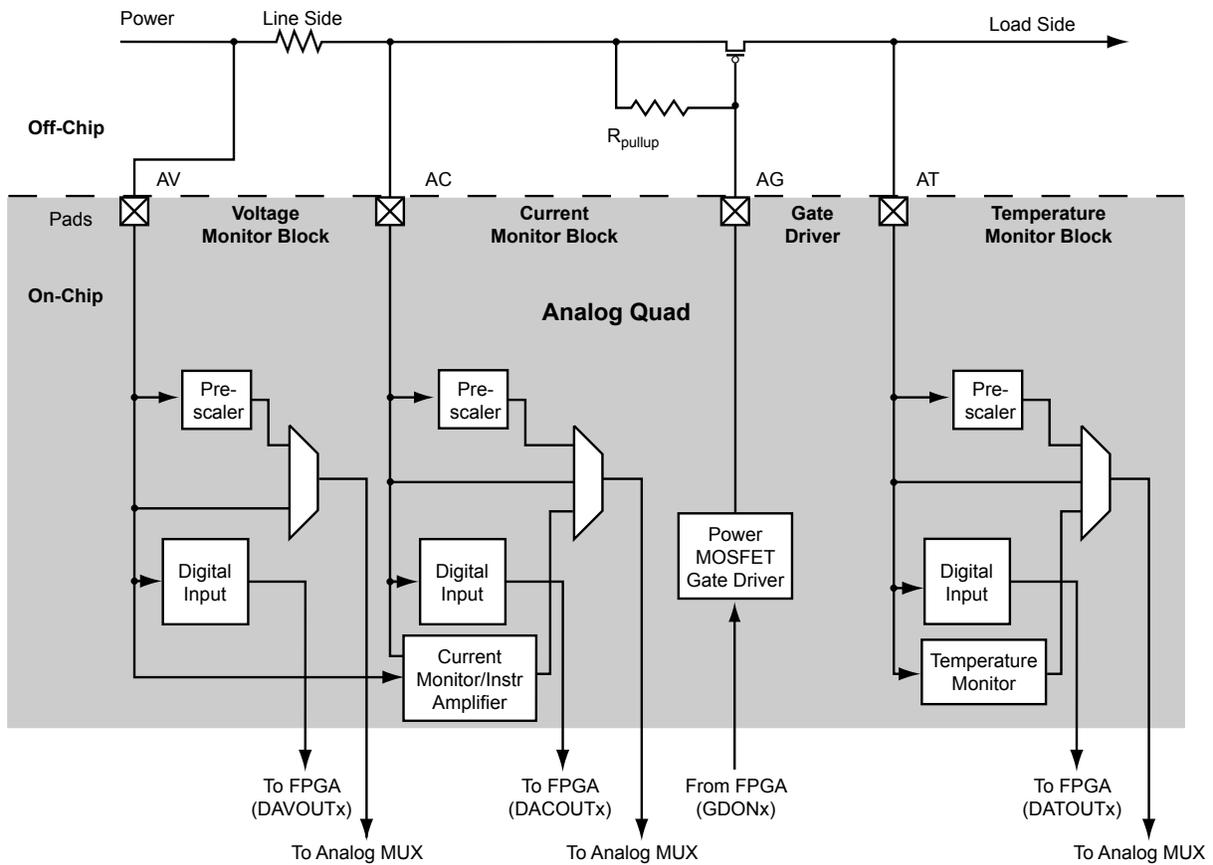
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With Fusion, Microsemi also introduces the Analog Quad I/O structure (Figure 1-1). Each quad consists of three analog inputs and one gate driver. Each quad can be configured in various built-in circuit combinations, such as three prescaler circuits, three digital input circuits, a current monitor circuit, or a temperature monitor circuit. Each prescaler has multiple scaling factors programmed by FPGA signals to support a large range of analog inputs with positive or negative polarity. When the current monitor circuit is selected, two adjacent analog inputs measure the voltage drop across a small external sense resistor. For more information, refer to the "Analog System Characteristics" section on page 2-117. Built-in operational amplifiers amplify small voltage signals for accurate current measurement. One analog input in each quad can be connected to an external temperature monitor diode. In addition to the external temperature monitor diode(s), a Fusion device can monitor an internal temperature diode using dedicated channel 31 of the ADCMUX.

Figure 1-1 on page 1-5 illustrates a typical use of the Analog Quad I/O structure. The Analog Quad shown is configured to monitor and control an external power supply. The AV pad measures the source of the power supply. The AC pad measures the voltage drop across an external sense resistor to calculate current. The AG MOSFET gate driver pad turns the external MOSFET on and off. The AT pad measures the load-side voltage level.



**Figure 1-1 • Analog Quad**

**Table 2-7 • AFS250 Global Resource Timing**  
**Commercial Temperature Range Conditions:  $T_J = 70^\circ\text{C}$ , Worst-Case  $V_{CC} = 1.425\text{ V}$**

Parameter	Description	-2		-1		Std.		Units
		Min. <sup>1</sup>	Max. <sup>2</sup>	Min. <sup>1</sup>	Max. <sup>2</sup>	Min. <sup>1</sup>	Max. <sup>2</sup>	
$t_{RCKL}$	Input Low Delay for Global Clock	0.89	1.12	1.02	1.27	1.20	1.50	ns
$t_{RCKH}$	Input High Delay for Global Clock	0.88	1.14	1.00	1.30	1.17	1.53	ns
$t_{RCKMPWH}$	Minimum Pulse Width High for Global Clock							ns
$t_{RCKMPWL}$	Minimum Pulse Width Low for Global Clock							ns
$t_{RCKSW}$	Maximum Skew for Global Clock		0.26		0.30		0.35	ns

**Notes:**

1. Value reflects minimum load. The delay is measured from the CCC output to the clock pin of a sequential element located in a lightly loaded row (single element is connected to the global net).
2. Value reflects maximum load. The delay is measured on the clock pin of the farthest sequential element located in a fully loaded row (all available flip-flops are connected to the global net in the row).
3. For the derating values at specific junction temperature and voltage supply levels, refer to [Table 3-7 on page 3-9](#).

**Table 2-8 • AFS090 Global Resource Timing**  
**Commercial Temperature Range Conditions:  $T_J = 70^\circ\text{C}$ , Worst-Case  $V_{CC} = 1.425\text{ V}$**

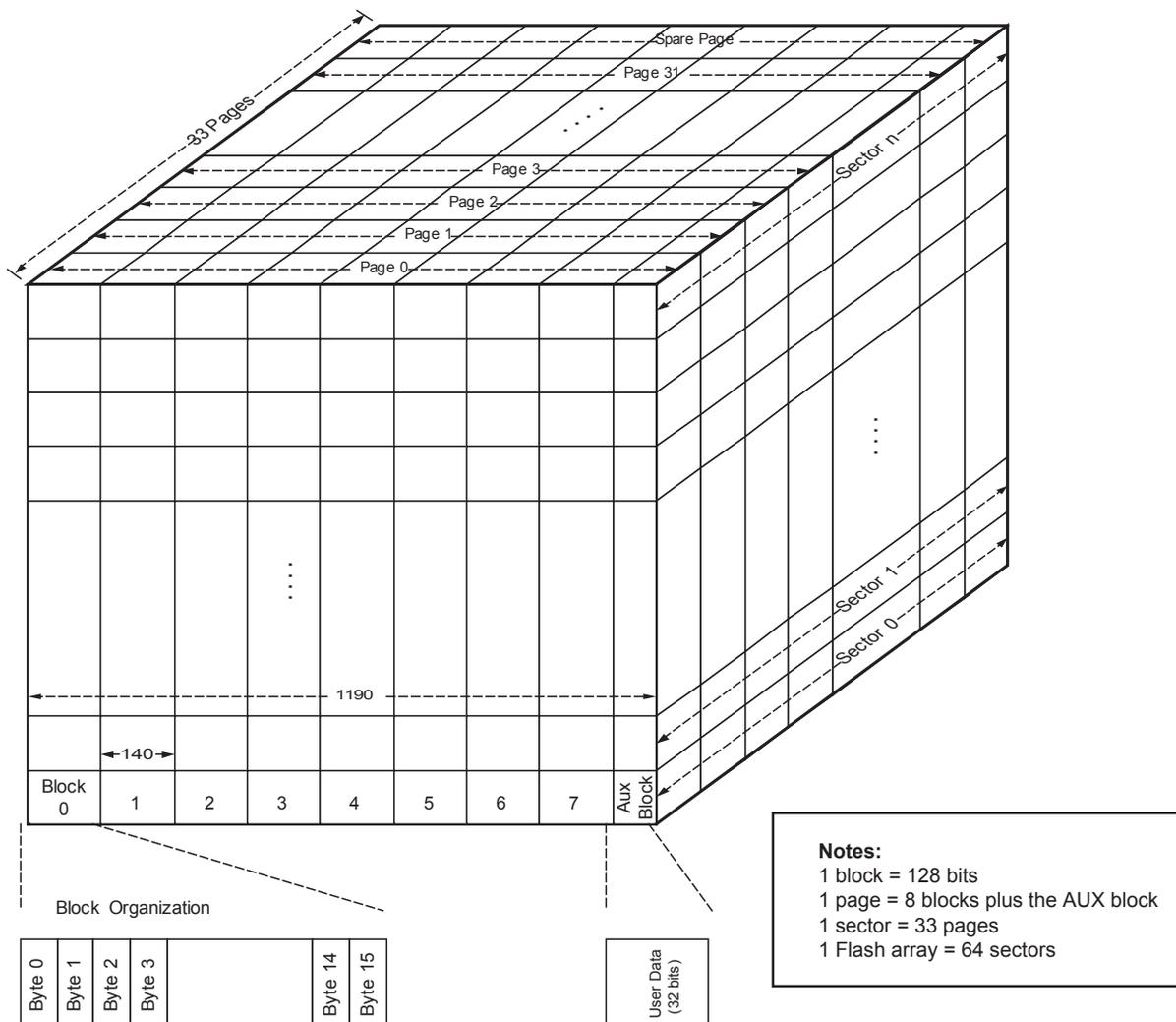
Parameter	Description	-2		-1		Std.		Units
		Min. <sup>1</sup>	Max. <sup>2</sup>	Min. <sup>1</sup>	Max. <sup>2</sup>	Min. <sup>1</sup>	Max. <sup>2</sup>	
$t_{RCKL}$	Input Low Delay for Global Clock	0.84	1.07	0.96	1.21	1.13	1.43	ns
$t_{RCKH}$	Input High Delay for Global Clock	0.83	1.10	0.95	1.25	1.12	1.47	ns
$t_{RCKMPWH}$	Minimum Pulse Width High for Global Clock							ns
$t_{RCKMPWL}$	Minimum Pulse Width Low for Global Clock							ns
$t_{RCKSW}$	Maximum Skew for Global Clock		0.27		0.30		0.36	ns

**Notes:**

1. Value reflects minimum load. The delay is measured from the CCC output to the clock pin of a sequential element located in a lightly loaded row (single element is connected to the global net).
2. Value reflects maximum load. The delay is measured on the clock pin of the farthest sequential element located in a fully loaded row (all available flip-flops are connected to the global net in the row).
3. For the derating values at specific junction temperature and voltage supply levels, refer to [Table 3-7 on page 3-9](#).

## Flash Memory Block Addressing

Figure 2-34 shows a graphical representation of the flash memory block.



**Figure 2-34 • Flash Memory Block Organization**

Each FB is partitioned into sectors, pages, blocks, and bytes. There are 64 sectors in an FB, and each sector contains 32 pages and 1 spare page. Each page contains 8 data blocks and 1 auxiliary block. Each data block contains 16 bytes of user data, and the auxiliary block contains 4 bytes of user data.

Addressing for the FB is shown in Table 2-20.

**Table 2-20 • FB Address Bit Allocation ADDR[17:0]**

17	12	11	7	6	4	3	0
Sector		Page		Block		Byte	

When the spare page of a sector is addressed (SPAREPAGE active), ADDR[11:7] are ignored.

When the Auxiliary block is addressed (AUXBLOCK active), ADDR[6:2] are ignored.

Note: The spare page of sector 0 is unavailable for any user data. Writes to this page will return an error, and reads will return all zeroes.

## FIFO4K18 Description

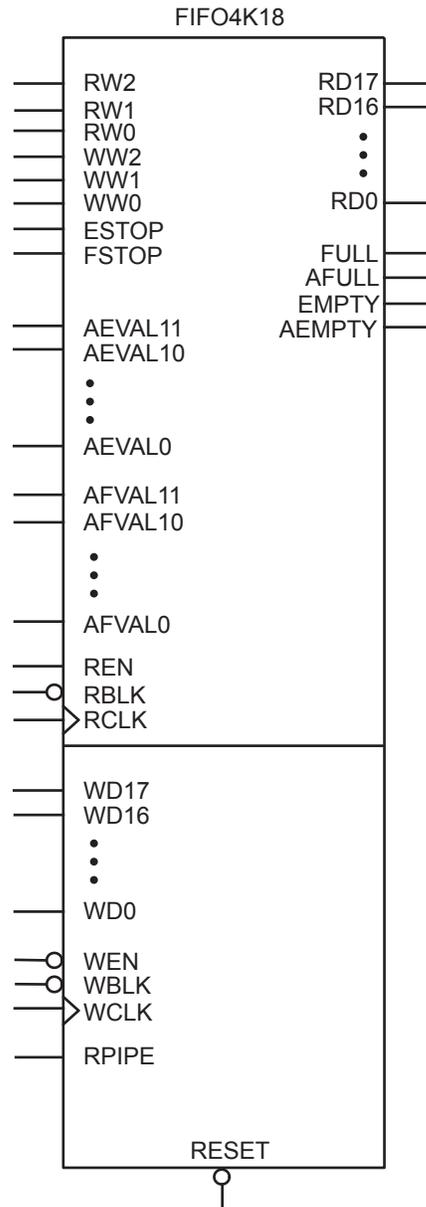
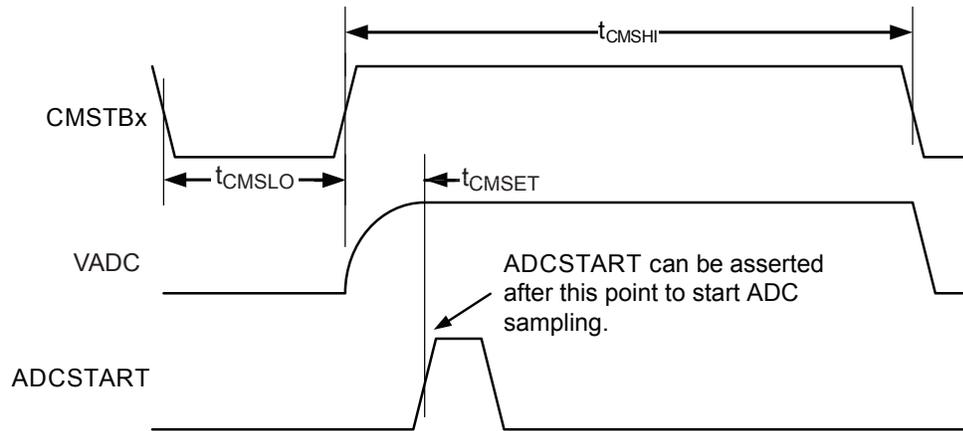


Figure 2-56 • FIFO4KX18

To initiate a current measurement, the appropriate Current Monitor Strobe (CMSTB) signal on the AB macro must be asserted low for at least  $t_{CMSLO}$  in order to discharge the previous measurement. Then CMSTB must be asserted high for at least  $t_{CMSET}$  prior to asserting the ADCSTART signal. The CMSTB must remain high until after the SAMPLE signal is de-asserted by the AB macro. Note that the minimum sample time cannot be less than  $t_{CMSHI}$ . Figure 2-71 shows the timing diagram of CMSTB in relationship with the ADC control signals.



**Figure 2-71 • Timing Diagram for Current Monitor Strobe**

Figure 2-72 illustrates positive current monitor operation. The differential voltage between AV and AC goes into the 10× amplifier and is then converted by the ADC. For example, a current of 1.5 A is drawn from a 10 V supply and is measured by the voltage drop across a 0.050 Ω sense resistor. The voltage drop is amplified by ten times by the amplifier and then measured by the ADC. The 1.5 A current creates a differential voltage across the sense resistor of 75 mV. This becomes 750 mV after amplification. Thus, the ADC measures a current of 1.5 A as 750 mV. Using an ADC with 8-bit resolution and VAREF of 2.56 V, the ADC result is decimal 75. EQ 3 shows how to compute the current from the ADC result.

$$I = (ADC \times V_{AREF}) / (10 \times 2^N \times R_{sense})$$

EQ 3

where

- I is the current flowing through the sense resistor
- ADC is the result from the ADC
- VAREF is the Reference voltage
- N is the number of bits
- R<sub>sense</sub> is the resistance of the sense resistor

EQ 16 through EQ 18 can be used to calculate the acquisition time required for a given input. The STC signal gives the number of sample periods in ADCCLK for the acquisition time of the desired signal. If the actual acquisition time is higher than the STC value, the settling time error can affect the accuracy of the ADC, because the sampling capacitor is only partially charged within the given sampling cycle. Example acquisition times are given in Table 2-44 and Table 2-45. When controlling the sample time for the ADC along with the use of the active bipolar prescaler, current monitor, or temperature monitor, the minimum sample time(s) for each must be obeyed. EQ 19 can be used to determine the appropriate value of STC.

You can calculate the minimum actual acquisition time by using EQ 16:

$$V_{OUT} = V_{IN}(1 - e^{-t/RC})$$

EQ 16

For 0.5 LSB gain error,  $V_{OUT}$  should be replaced with  $(V_{IN} - (0.5 \times \text{LSB Value}))$ :

$$(V_{IN} - 0.5 \times \text{LSB Value}) = V_{IN}(1 - e^{-t/RC})$$

EQ 17

where  $V_{IN}$  is the ADC reference voltage ( $V_{REF}$ )

Solving EQ 17:

$$t = RC \times \ln(V_{IN} / (0.5 \times \text{LSB Value}))$$

EQ 18

where  $R = Z_{INAD} + R_{SOURCE}$  and  $C = C_{INAD}$ .

Calculate the value of STC by using EQ 19.

$$t_{SAMPLE} = (2 + \text{STC}) \times (1 / \text{ADCCLK}) \text{ or } t_{SAMPLE} = (2 + \text{STC}) \times (\text{ADC Clock Period})$$

EQ 19

where ADCCLK = ADC clock frequency in MHz.

$t_{SAMPLE} = 0.449 \mu\text{s}$  from bit resolution in Table 2-44.

ADC Clock frequency = 10 MHz or a 100 ns period.

$\text{STC} = (t_{SAMPLE} / (1 / 10 \text{ MHz})) - 2 = 4.49 - 2 = 2.49$ .

You must round up to 3 to accommodate the minimum sample time.

**Table 2-44 • Acquisition Time Example with  $V_{AREF} = 2.56 \text{ V}$**

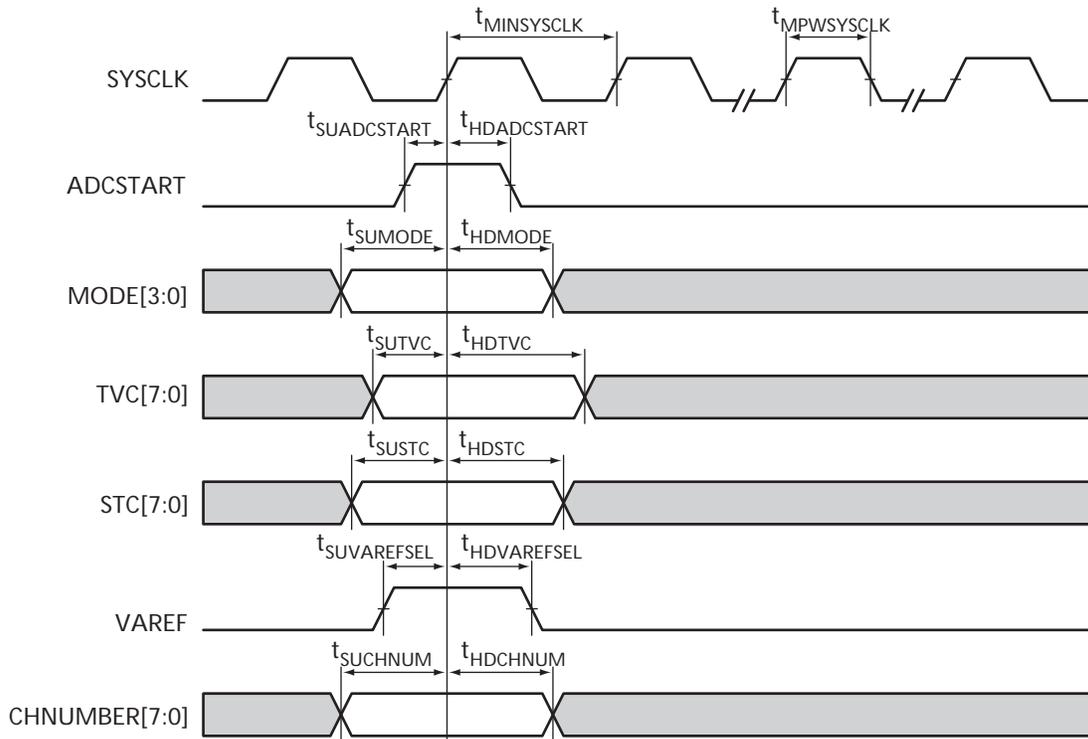
VIN = 2.56V, R = 4K (R <sub>SOURCE</sub> ~ 0), C = 18 pF		
Resolution	LSB Value (mV)	Min. Sample/Hold Time for 0.5 LSB (μs)
8	10	0.449
10	2.5	0.549
12	0.625	0.649

**Table 2-45 • Acquisition Time Example with  $V_{AREF} = 3.3 \text{ V}$**

VIN = 3.3V, R = 4K (R <sub>SOURCE</sub> ~ 0), C = 18 pF		
Resolution	LSB Value (mV)	Min. Sample/Hold time for 0.5 LSB (μs)
8	12.891	0.449
10	3.223	0.549
12	0.806	0.649

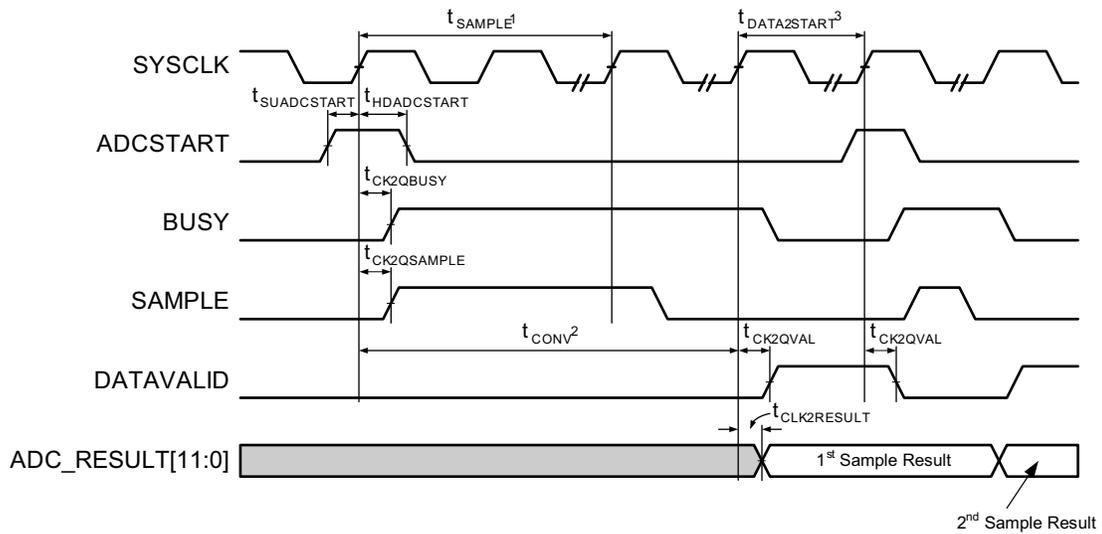
### Sample Phase

A conversion is performed in three phases. In the first phase, the analog input voltage is sampled on the input capacitor. This phase is called sample phase. During the sample phase, the output signals BUSY and SAMPLE change from '0' to '1', indicating the ADC is busy and sampling the analog signal. The sample time can be controlled by input signals STC[7:0]. The sample time can be calculated by EQ 20. When controlling the sample time for the ADC along with the use of Prescaler or Current Monitor or Temperature Monitor, the minimum sample time for each must be obeyed.



**Figure 2-90 • Input Setup Time**

**Standard Conversion**



**Notes:**

1. Refer to EQ 20 on page 2-109 for the calculation on the sample time,  $t_{SAMPLE}$ .
2. See EQ 23 on page 2-109 for calculation of the conversion time,  $t_{CONV}$ .
3. Minimum time to issue an ADCSTART after DATAVALID is 1 SYSCLK period

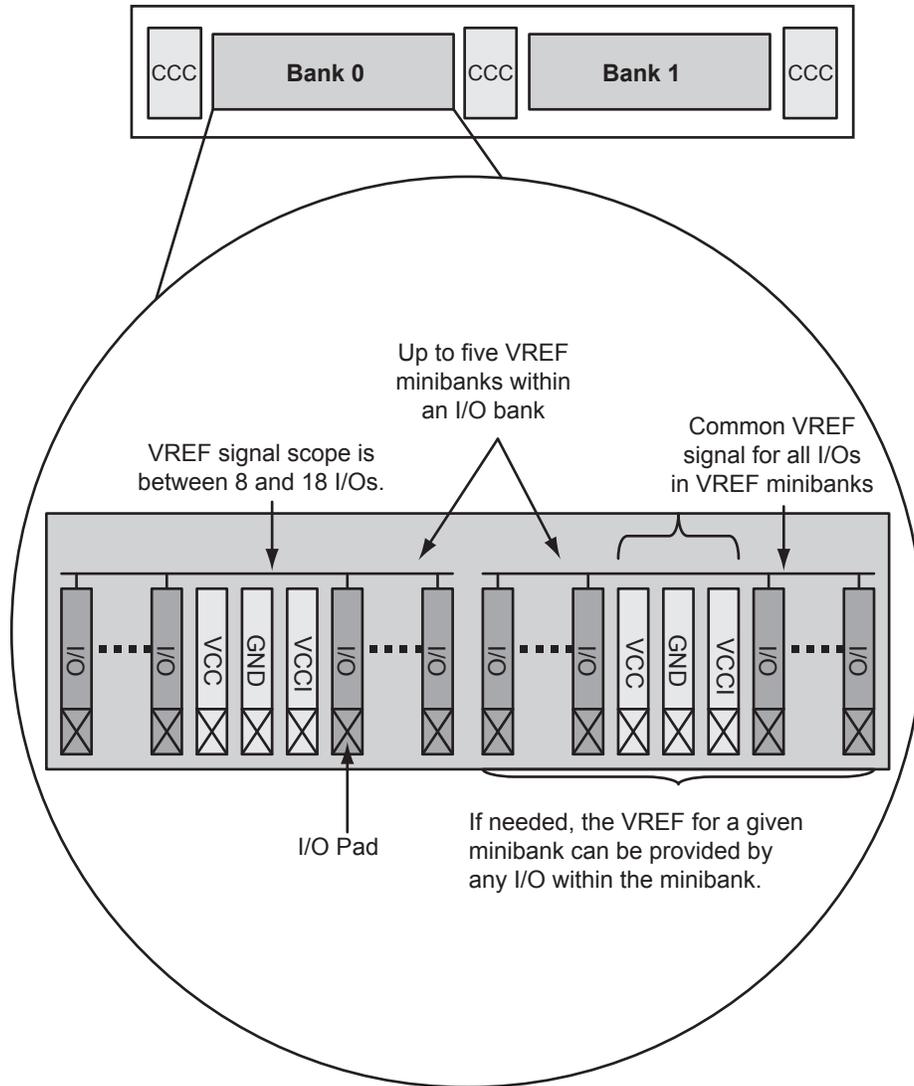
**Figure 2-91 • Standard Conversion Status Signal Timing Diagram**

**Table 2-52 • Calibrated Analog Channel Accuracy<sup>1,2,3</sup>**  
**Worst-Case Industrial Conditions, T<sub>J</sub> = 85°C**

		Condition	Total Channel Error (LSB)		
Analog Pad	Prescaler Range (V)	Input Voltage <sup>4</sup> (V)	Negative Max.	Median	Positive Max.
<b>Positive Range</b>			<b>ADC in 10-Bit Mode</b>		
AV, AC	16	0.300 to 12.0	-6	1	6
	8	0.250 to 8.00	-6	0	6
	4	0.200 to 4.00	-7	-1	7
	2	0.150 to 2.00	-7	0	7
	1	0.050 to 1.00	-6	-1	6
AT	16	0.300 to 16.0	-5	0	5
	4	0.100 to 4.00	-7	-1	7
<b>Negative Range</b>			<b>ADC in 10-Bit Mode</b>		
AV, AC	16	-0.400 to -10.5	-7	1	9
	8	-0.350 to -8.00	-7	-1	7
	4	-0.300 to -4.00	-7	-2	9
	2	-0.250 to -2.00	-7	-2	7
	1	-0.050 to -1.00	-16	-1	20

**Notes:**

1. Channel Accuracy includes prescaler and ADC accuracies. For 12-bit mode, multiply the LSB count by 4. For 8-bit mode, divide the LSB count by 4. Overall accuracy remains the same.
2. Requires enabling Analog Calibration using SmartGen Analog System Builder. For further details, refer to the "Temperature, Voltage, and Current Calibration in Fusion FPGAs" chapter of the [Fusion FPGA Fabric User Guide](#).
3. Calibrated with two-point calibration methodology, using 20% and 80% full-scale points.
4. The lower limit of the input voltage is determined by the prescaler input offset.



**Figure 2-99 • Fusion Pro I/O Bank Detail Showing VREF Minibanks (north side of AFS600 and AFS1500)**

**Table 2-67 • I/O Standards Supported by Bank Type**

I/O Bank	Single-Ended I/O Standards	Differential I/O Standards	Voltage-Referenced	Hot-Swap
Standard I/O	LVTTL/LVCMOS 3.3 V, LVCMOS 2.5 V / 1.8 V / 1.5 V, LVCMOS 2.5/5.0 V	–	–	Yes
Advanced I/O	LVTTL/LVCMOS 3.3 V, LVCMOS 2.5 V / 1.8 V / 1.5 V, LVCMOS 2.5/5.0 V, 3.3 V PCI / 3.3 V PCI-X	LVPECL and LVDS	–	–
Pro I/O	LVTTL/LVCMOS 3.3 V, LVCMOS 2.5 V / 1.8 V / 1.5 V, LVCMOS 2.5/5.0 V, 3.3 V PCI / 3.3 V PCI-X	LVPECL and LVDS	GTL+ 2.5 V / 3.3 V, GTL 2.5 V / 3.3 V, HSTL Class I and II, SSTL2 Class I and II, SSTL3 Class I and II	Yes

**Solution 3**

The board-level design must ensure that the reflected waveform at the pad does not exceed limits provided in Table 3-4 on page 3-4. This is a long-term reliability requirement.

This scheme will also work for a 3.3 V PCI/PCIX configuration, but the internal diode should not be used for clamping, and the voltage must be limited by the bus switch, as shown in Figure 2-105. Relying on the diode clamping would create an excessive pad DC voltage of 3.3 V + 0.7 V = 4 V.

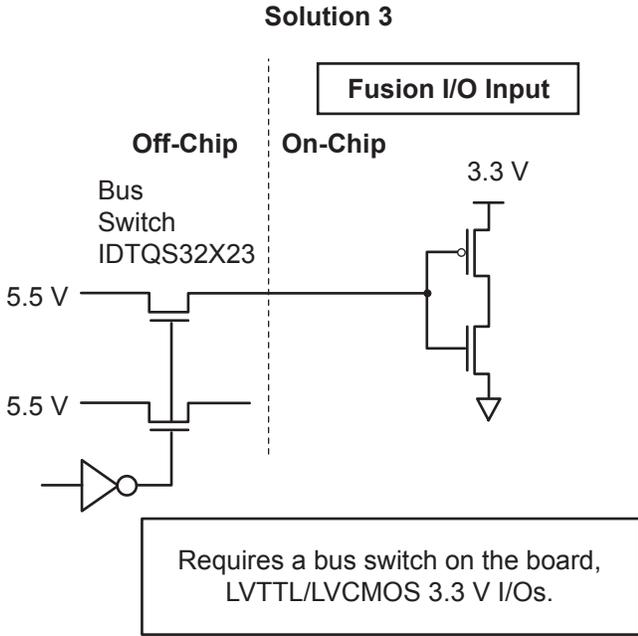


Figure 2-105 • Solution 3

**Solution 4**

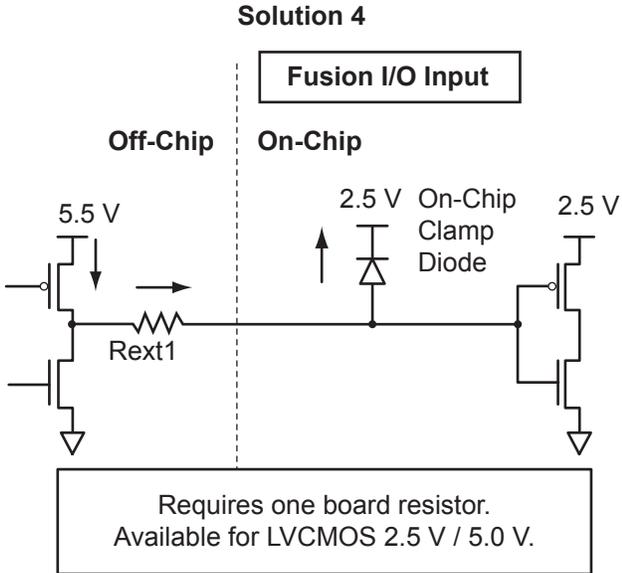


Figure 2-106 • Solution 4

**Table 2-107 • 3.3 V LVTTTL / 3.3 V LVCMOS High Slew**  
**Commercial Temperature Range Conditions:  $T_J = 70^\circ\text{C}$ , Worst-Case  $V_{CC} = 1.425\text{ V}$ ,**  
**Worst-Case  $V_{CCI} = 3.0\text{ V}$**   
**Applicable to Advanced I/Os**

Drive Strength	Speed Grade	$t_{DOUT}$	$t_{DP}$	$t_{DIN}$	$t_{PY}$	$t_{EOUT}$	$t_{ZL}$	$t_{ZH}$	$t_{LZ}$	$t_{HZ}$	$t_{ZLS}$	$t_{ZHS}$	Units
4 mA	Std.	0.66	7.66	0.04	1.20	0.43	7.80	6.59	2.65	2.61	10.03	8.82	ns
	-1	0.56	6.51	0.04	1.02	0.36	6.63	5.60	2.25	2.22	8.54	7.51	ns
	-2	0.49	5.72	0.03	0.90	0.32	5.82	4.92	1.98	1.95	7.49	6.59	ns
8 mA	Std.	0.66	4.91	0.04	1.20	0.43	5.00	4.07	2.99	3.20	7.23	6.31	ns
	-1	0.56	4.17	0.04	1.02	0.36	4.25	3.46	2.54	2.73	6.15	5.36	ns
	-2	0.49	3.66	0.03	0.90	0.32	3.73	3.04	2.23	2.39	5.40	4.71	ns
12 mA	Std.	0.66	3.53	0.04	1.20	0.43	3.60	2.82	3.21	3.58	5.83	5.06	ns
	-1	0.56	3.00	0.04	1.02	0.36	3.06	2.40	2.73	3.05	4.96	4.30	ns
	-2	0.49	2.64	0.03	0.90	0.32	2.69	2.11	2.40	2.68	4.36	3.78	ns
16 mA	Std.	0.66	3.33	0.04	1.20	0.43	3.39	2.56	3.26	3.68	5.63	4.80	ns
	-1	0.56	2.83	0.04	1.02	0.36	2.89	2.18	2.77	3.13	4.79	4.08	ns
	-2	0.49	2.49	0.03	0.90	0.32	2.53	1.91	2.44	2.75	4.20	3.58	ns
24 mA	Std.	0.66	3.08	0.04	1.20	0.43	3.13	2.12	3.32	4.06	5.37	4.35	ns
	-1	0.56	2.62	0.04	1.02	0.36	2.66	1.80	2.83	3.45	4.57	3.70	ns
	-2	0.49	2.30	0.03	0.90	0.32	2.34	1.58	2.48	3.03	4.01	3.25	ns

Note: For the derating values at specific junction temperature and voltage supply levels, refer to Table 3-7 on page 3-9.

**Table 2-108 • 3.3 V LVTTTL / 3.3 V LVCMOS Low Slew**  
**Commercial Temperature Range Conditions:  $T_J = 70^\circ\text{C}$ , Worst-Case  $V_{CC} = 1.425\text{ V}$ ,**  
**Worst-Case  $V_{CCI} = 3.0\text{ V}$**   
**Applicable to Standard I/Os**

Drive Strength	Speed Grade	$t_{DOUT}$	$t_{DP}$	$t_{DIN}$	$t_{PY}$	$t_{EOUT}$	$t_{ZL}$	$t_{ZH}$	$t_{LZ}$	$t_{HZ}$	Units
2 mA	Std.	0.66	9.46	0.04	1.00	0.43	9.64	8.54	2.07	2.04	ns
	-1	0.56	8.05	0.04	0.85	0.36	8.20	7.27	1.76	1.73	ns
	-2	0.49	7.07	0.03	0.75	0.32	7.20	6.38	1.55	1.52	ns
4 mA	Std.	0.66	9.46	0.04	1.00	0.43	9.64	8.54	2.07	2.04	ns
	-1	0.56	8.05	0.04	0.85	0.36	8.20	7.27	1.76	1.73	ns
	-2	0.49	7.07	0.03	0.75	0.32	7.20	6.38	1.55	1.52	ns
6 mA	Std.	0.66	6.57	0.04	1.00	0.43	6.69	5.98	2.40	2.57	ns
	-1	0.56	5.59	0.04	0.85	0.36	5.69	5.09	2.04	2.19	ns
	-2	0.49	4.91	0.03	0.75	0.32	5.00	4.47	1.79	1.92	ns
8 mA	Std.	0.66	6.57	0.04	1.00	0.43	6.69	5.98	2.40	2.57	ns
	-1	0.56	5.59	0.04	0.85	0.36	5.69	5.09	2.04	2.19	ns
	-2	0.49	4.91	0.03	0.75	0.32	5.00	4.47	1.79	1.92	ns

Note: For the derating values at specific junction temperature and voltage supply levels, refer to Table 3-7 on page 3-9.

**Table 2-121 • 1.8 V LVC MOS High Slew**  
**Commercial Temperature Range Conditions:  $T_J = 70^\circ\text{C}$ , Worst-Case  $V_{CC} = 1.425\text{ V}$ ,**  
**Worst-Case  $V_{CCI} = 1.7\text{ V}$**   
**Applicable to Pro I/Os**

Drive Strength	Speed Grade	$t_{DOUT}$	$t_{DP}$	$t_{DIN}$	$t_{PY}$	$t_{PYS}$	$t_{EOUT}$	$t_{ZL}$	$t_{ZH}$	$t_{LZ}$	$t_{HZ}$	$t_{ZLS}$	$t_{ZHS}$	Units
2 mA	Std.	0.66	12.10	0.04	1.45	1.91	0.43	9.59	12.10	2.78	1.64	11.83	14.34	ns
	-1	0.56	10.30	0.04	1.23	1.62	0.36	8.16	10.30	2.37	1.39	10.06	12.20	ns
	-2	0.49	9.04	0.03	1.08	1.42	0.32	7.16	9.04	2.08	1.22	8.83	10.71	ns
4 mA	Std.	0.66	7.05	0.04	1.45	1.91	0.43	6.20	7.05	3.25	2.86	8.44	9.29	ns
	-1	0.56	6.00	0.04	1.23	1.62	0.36	5.28	6.00	2.76	2.44	7.18	7.90	ns
	-2	0.49	5.27	0.03	1.08	1.42	0.32	4.63	5.27	2.43	2.14	6.30	6.94	ns
8 mA	Std.	0.66	4.52	0.04	1.45	1.91	0.43	4.47	4.52	3.57	3.47	6.70	6.76	ns
	-1	0.56	3.85	0.04	1.23	1.62	0.36	3.80	3.85	3.04	2.95	5.70	5.75	ns
	-2	0.49	3.38	0.03	1.08	1.42	0.32	3.33	3.38	2.66	2.59	5.00	5.05	ns
12 mA	Std.	0.66	4.12	0.04	1.45	1.91	0.43	4.20	3.99	3.63	3.62	6.43	6.23	ns
	-1	0.56	3.51	0.04	1.23	1.62	0.36	3.57	3.40	3.09	3.08	5.47	5.30	ns
	-2	0.49	3.08	0.03	1.08	1.42	0.32	3.14	2.98	2.71	2.71	4.81	4.65	ns
16 mA	Std.	0.66	3.80	0.04	1.45	1.91	0.43	3.87	3.09	3.73	4.24	6.10	5.32	ns
	-1	0.56	3.23	0.04	1.23	1.62	0.36	3.29	2.63	3.18	3.60	5.19	4.53	ns
	-2	0.49	2.83	0.03	1.08	1.42	0.32	2.89	2.31	2.79	3.16	4.56	3.98	ns

*Note:* For the derating values at specific junction temperature and voltage supply levels, refer to [Table 3-7 on page 3-9](#).

### 2.5 V GTL+

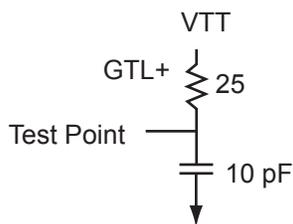
Gunning Transceiver Logic Plus is a high-speed bus standard (JESD8-3). It provides a differential amplifier input buffer and an open-drain output buffer. The VCCI pin should be connected to 2.5 V.

**Table 2-147 • Minimum and Maximum DC Input and Output Levels**

2.5 V GTL+	VIL		VIH		VOL	VOH	IOL	IOH	IOSL	IOSH	IIL <sup>1</sup>	I <sub>IH</sub> <sup>2</sup>
	Min. V	Max. V	Min. V	Max. V	Max. V	Min. V	mA	mA	Max. mA <sup>3</sup>	Max. mA <sup>3</sup>	μA <sup>4</sup>	μA <sup>4</sup>
33 mA	-0.3	VREF - 0.1	VREF + 0.1	3.6	0.6	-	33	33	124	169	10	10

**Notes:**

1. IIL is the input leakage current per I/O pin over recommended operation conditions where  $-0.3\text{ V} < V_{IN} < V_{IL}$ .
2. I<sub>IH</sub> is the input leakage current per I/O pin over recommended operating conditions  $V_{IH} < V_{IN} < V_{CCI}$ . Input current is larger when operating outside recommended ranges.
3. Currents are measured at high temperature (100°C junction temperature) and maximum voltage.
4. Currents are measured at 85°C junction temperature.



**Figure 2-127 • AC Loading**

**Table 2-148 • AC Waveforms, Measuring Points, and Capacitive Loads**

Input Low (V)	Input High (V)	Measuring Point* (V)	VREF (typ.) (V)	VTT (typ.) (V)	C <sub>LOAD</sub> (pF)
VREF - 0.1	VREF + 0.1	1.0	1.0	1.5	10

Note: \*Measuring point = V<sub>trip</sub>. See Table 2-90 on page 2-166 for a complete table of trip points.

### Timing Characteristics

**Table 2-149 • 2.5 V GTL+**

Commercial Temperature Range Conditions: T<sub>J</sub> = 70°C, Worst-Case VCC = 1.425 V, Worst-Case VCCI = 2.3 V, VREF = 1.0 V

Speed Grade	t <sub>DOUT</sub>	t <sub>DP</sub>	t <sub>DIN</sub>	t <sub>PY</sub>	t <sub>EOUT</sub>	t <sub>ZL</sub>	t <sub>ZH</sub>	t <sub>LZ</sub>	t <sub>HZ</sub>	t <sub>ZLS</sub>	t <sub>ZHS</sub>	Units
Std.	0.66	2.21	0.04	1.51	0.43	2.25	2.10			4.48	4.34	ns
-1	0.56	1.88	0.04	1.29	0.36	1.91	1.79			3.81	3.69	ns
-2	0.49	1.65	0.03	1.13	0.32	1.68	1.57			3.35	4.34	ns

Note: For the derating values at specific junction temperature and voltage supply levels, refer to Table 3-7 on page 3-9.

## Thermal Characteristics

### Introduction

The temperature variable in the Microsemi Designer software refers to the junction temperature, not the ambient, case, or board temperatures. This is an important distinction because dynamic and static power consumption will cause the chip's junction temperature to be higher than the ambient, case, or board temperatures. EQ 1 through EQ 3 give the relationship between thermal resistance, temperature gradient, and power.

$$\theta_{JA} = \frac{T_J - T_A}{P}$$

EQ 1

$$\theta_{JB} = \frac{T_J - T_B}{P}$$

EQ 2

$$\theta_{JC} = \frac{T_J - T_C}{P}$$

EQ 3

where

- $\theta_{JA}$  = Junction-to-air thermal resistance
- $\theta_{JB}$  = Junction-to-board thermal resistance
- $\theta_{JC}$  = Junction-to-case thermal resistance
- $T_J$  = Junction temperature
- $T_A$  = Ambient temperature
- $T_B$  = Board temperature (measured 1.0 mm away from the package edge)
- $T_C$  = Case temperature
- $P$  = Total power dissipated by the device

**Table 3-6 • Package Thermal Resistance**

Product	$\theta_{JA}$			$\theta_{JC}$	$\theta_{JB}$	Units
	Still Air	1.0 m/s	2.5 m/s			
AFS090-QN108	34.5	30.0	27.7	8.1	16.7	°C/W
AFS090-QN180	33.3	27.6	25.7	9.2	21.2	°C/W
AFS250-QN180	32.2	26.5	24.7	5.7	15.0	°C/W
AFS250-PQ208	42.1	38.4	37	20.5	36.3	°C/W
AFS600-PQ208	23.9	21.3	20.48	6.1	16.5	°C/W
AFS090-FG256	37.7	33.9	32.2	11.5	29.7	°C/W
AFS250-FG256	33.7	30.0	28.3	9.3	24.8	°C/W
AFS600-FG256	28.9	25.2	23.5	6.8	19.9	°C/W
AFS1500-FG256	23.3	19.6	18.0	4.3	14.2	°C/W
AFS600-FG484	21.8	18.2	16.7	7.7	16.8	°C/W
AFS1500-FG484	21.6	16.8	15.2	5.6	14.9	°C/W
AFS1500-FG676	TBD	TBD	TBD	TBD	TBD	°C/W

**Table 3-13 • Summary of I/O Output Buffer Power (per pin)—Default I/O Software Settings<sup>1</sup>**

	C <sub>LOAD</sub> (pF)	V <sub>CCI</sub> (V)	Static Power PDC8 (mW) <sup>2</sup>	Dynamic Power PAC10 (μW/MHz) <sup>3</sup>
<b>Applicable to Pro I/O Banks</b>				
<b>Single-Ended</b>				
3.3 V LVTTTL/LVCMOS	35	3.3	–	474.70
2.5 V LVCMOS	35	2.5	–	270.73
1.8 V LVCMOS	35	1.8	–	151.78
1.5 V LVCMOS (JESD8-11)	35	1.5	–	104.55
3.3 V PCI	10	3.3	–	204.61
3.3 V PCI-X	10	3.3	–	204.61
<b>Voltage-Referenced</b>				
3.3 V GTL	10	3.3	–	24.08
2.5 V GTL	10	2.5	–	13.52
3.3 V GTL+	10	3.3	–	24.10
2.5 V GTL+	10	2.5	–	13.54
HSTL (I)	20	1.5	7.08	26.22
HSTL (II)	20	1.5	13.88	27.22
SSTL2 (I)	30	2.5	16.69	105.56
SSTL2 (II)	30	2.5	25.91	116.60
SSTL3 (I)	30	3.3	26.02	114.87
SSTL3 (II)	30	3.3	42.21	131.76
<b>Differential</b>				
LVDS	–	2.5	7.70	89.62
LVPECL	–	3.3	19.42	168.02
<b>Applicable to Advanced I/O Banks</b>				
<b>Single-Ended</b>				
3.3 V LVTTTL / 3.3 V LVCMOS	35	3.3	–	468.67
2.5 V LVCMOS	35	2.5	–	267.48
1.8 V LVCMOS	35	1.8	–	149.46
1.5 V LVCMOS (JESD8-11)	35	1.5	–	103.12
3.3 V PCI	10	3.3	–	201.02
3.3 V PCI-X	10	3.3	–	201.02

**Notes:**

1. Dynamic power consumption is given for standard load and software-default drive strength and output slew.
2. PDC8 is the static power (where applicable) measured on V<sub>CCI</sub>.
3. PAC10 is the total dynamic power measured on V<sub>CC</sub> and V<sub>CCI</sub>.

## Static Power Consumption of Various Internal Resources

**Table 3-15 • Different Components Contributing to the Static Power Consumption in Fusion Devices**

Parameter	Definition	Power Supply		Device-Specific Static Contributions				Units
				AFS1500	AFS600	AFS250	AFS090	
PDC1	Core static power contribution in operating mode	VCC	1.5 V	18	7.5	4.50	3.00	mW
PDC2	Device static power contribution in standby mode	VCC33A	3.3 V	0.66				mW
PDC3	Device static power contribution in sleep mode	VCC33A	3.3 V	0.03				mW
PDC4	NVM static power contribution	VCC	1.5 V	1.19				mW
PDC5	Analog Block static power contribution of ADC	VCC33A	3.3 V	8.25				mW
PDC6	Analog Block static power contribution per Quad	VCC33A	3.3 V	3.3				mW
PDC7	Static contribution per input pin – standard dependent contribution	VCCI	See <a href="#">Table 3-12 on page 3-18</a>					
PDC8	Static contribution per input pin – standard dependent contribution	VCCI	See <a href="#">Table 3-13 on page 3-20</a>					
PDC9	Static contribution for PLL	VCC	1.5 V	2.55				mW

### Power Calculation Methodology

This section describes a simplified method to estimate power consumption of an application. For more accurate and detailed power estimations, use the SmartPower tool in the Libero SoC software.

The power calculation methodology described below uses the following variables:

- The number of PLLs as well as the number and the frequency of each output clock generated
- The number of combinatorial and sequential cells used in the design
- The internal clock frequencies
- The number and the standard of I/O pins used in the design
- The number of RAM blocks used in the design
- The number of NVM blocks used in the design
- The number of Analog Quads used in the design
- Toggle rates of I/O pins as well as VersaTiles—guidelines are provided in [Table 3-16 on page 3-27](#).
- Enable rates of output buffers—guidelines are provided for typical applications in [Table 3-17 on page 3-27](#).
- Read rate and write rate to the RAM—guidelines are provided for typical applications in [Table 3-17 on page 3-27](#).
- Read rate to the NVM blocks

The calculation should be repeated for each clock domain defined in the design.

### **Total Static Power Consumption— $P_{STAT}$**

Number of Quads used:  $N_{QUADS} = 4$

Number of NVM blocks available (AFS600):  $N_{NVM-BLOCKS} = 2$

Number of input pins used:  $N_{INPUTS} = 30$

Number of output pins used:  $N_{OUTPUTS} = 40$

#### **Operating Mode**

$$P_{STAT} = PDC1 + (N_{NVM-BLOCKS} * PDC4) + PDC5 + (N_{QUADS} * PDC6) + (N_{INPUTS} * PDC7) + (N_{OUTPUTS} * PDC8)$$

$$P_{STAT} = 7.50 \text{ mW} + (2 * 1.19 \text{ mW}) + 8.25 \text{ mW} + (4 * 3.30 \text{ mW}) + (30 * 0.00) + (40 * 0.00)$$

$$P_{STAT} = 31.33 \text{ mW}$$

#### **Standby Mode**

$$P_{STAT} = PDC2$$

$$P_{STAT} = 0.03 \text{ mW}$$

#### **Sleep Mode**

$$P_{STAT} = PDC3$$

$$P_{STAT} = 0.03 \text{ mW}$$

### **Total Power Consumption— $P_{TOTAL}$**

In operating mode, the total power consumption of the device is 174.39 mW:

$$P_{TOTAL} = P_{STAT} + P_{DYN}$$

$$P_{TOTAL} = 143.06 \text{ mW} + 31.33 \text{ mW}$$

$$P_{TOTAL} = 174.39 \text{ mW}$$

In standby mode, the total power consumption of the device is limited to 0.66 mW:

$$P_{TOTAL} = P_{STAT} + P_{DYN}$$

$$P_{TOTAL} = 0.03 \text{ mW} + 0.63 \text{ mW}$$

$$P_{TOTAL} = 0.66 \text{ mW}$$

In sleep mode, the total power consumption of the device drops as low as 0.03 mW:

$$P_{TOTAL} = P_{STAT} + P_{DYN}$$

$$P_{TOTAL} = 0.03 \text{ mW}$$

<b>FG256</b>				
<b>Pin Number</b>	<b>AFS090 Function</b>	<b>AFS250 Function</b>	<b>AFS600 Function</b>	<b>AFS1500 Function</b>
R5	AV0	AV0	AV2	AV2
R6	AT0	AT0	AT2	AT2
R7	AV1	AV1	AV3	AV3
R8	AT3	AT3	AT5	AT5
R9	AV4	AV4	AV6	AV6
R10	NC	AT5	AT7	AT7
R11	NC	AV5	AV7	AV7
R12	NC	NC	AT9	AT9
R13	NC	NC	AG9	AG9
R14	NC	NC	AC9	AC9
R15	PUB	PUB	PUB	PUB
R16	VCCIB1	VCCIB1	VCCIB2	VCCIB2
T1	GND	GND	GND	GND
T2	NCAP	NCAP	NCAP	NCAP
T3	VCC33N	VCC33N	VCC33N	VCC33N
T4	NC	NC	ATR TN0	ATR TN0
T5	AT1	AT1	AT3	AT3
T6	ATR TN0	ATR TN0	ATR TN1	ATR TN1
T7	AT2	AT2	AT4	AT4
T8	ATR TN1	ATR TN1	ATR TN2	ATR TN2
T9	AT4	AT4	AT6	AT6
T10	ATR TN2	ATR TN2	ATR TN3	ATR TN3
T11	NC	NC	AT8	AT8
T12	NC	NC	ATR TN4	ATR TN4
T13	GND A	GND A	GND A	GND A
T14	VCC33A	VCC33A	VCC33A	VCC33A
T15	VAREF	VAREF	VAREF	VAREF
T16	GND	GND	GND	GND

## 5 – Datasheet Information

### List of Changes

The following table lists critical changes that were made in each revision of the Fusion datasheet.

Revision	Changes	Page
Revision 6 (March 2014)	Note added for the discontinuance of QN108 and QN180 packages to the "Package I/Os: Single-/Double-Ended (Analog)" table and the "Temperature Grade Offerings" table (SAR 55113, PDN 1306).	II and IV
	Updated details about page programming time in the "Program Operation" section (SAR 49291).	2-46
	ADC_START changed to ADCSTART in the "ADC Operation" section (SAR 44104).	2-104
Revision 5 (January 2014)	Calibrated offset values (AFS090, AFS250) of the external temperature monitor in Table 2-49 • Analog Channel Specifications have been updated (SAR 51464).	2-117
	Specifications for the internal temperature monitor in Table 2-49 • Analog Channel Specifications have been updated (SAR 50870).	2-117
Revision 4 (January 2013)	The "Product Ordering Codes" section has been updated to mention "Y" as "Blank" mentioning "Device Does Not Include License to Implement IP Based on the Cryptography Research, Inc. (CRI) Patent Portfolio" (SAR 43177).	III
	The note in Table 2-12 • Fusion CCC/PLL Specification referring the reader to SmartGen was revised to refer instead to the online help associated with the core (SAR 42563).	2-28
	Table 2-49 • Analog Channel Specifications was modified to update the uncalibrated offset values (AFS250) of the external and internal temperature monitors (SAR 43134).	2-117
	In Table 2-57 • Prescaler Control Truth Table—AV (x = 0), AC (x = 1), and AT (x = 3), changed the column heading from 'Full-Scale Voltage' to 'Full Scale Voltage in 10-Bit Mode', and added and updated Notes as required (SAR 20812).	2-130
	The values for the Speed Grade (-1 and Std.) for FDDRIMAX (Table 2-180 • Input DDR Propagation Delays) and values for the Speed Grade (-2 and Std.) for FDDOMAX (Table 2-182 • Output DDR Propagation Delays) had been inadvertently interchanged. This has been rectified (SAR 38514).	2-220, 2-222
	Added description about what happens if a user connects VAREF to an external 3.3 V on their board to the "VAREF Analog Reference Voltage" section (SAR 35188).	2-225
	Added a note to Table 3-2 • Recommended Operating Conditions <sup>1</sup> (SAR 43429): The programming temperature range supported is T <sub>ambient</sub> = 0°C to 85°C.	3-3
	Added the Package Thermal details for AFS600-PQ208 and AFS250-PQ208 to Table 3-6 • Package Thermal Resistance (SAR 37816). Deleted the Die Size column from the table (SAR 43503).	3-7
	Libero Integrated Design Environment (IDE) was changed to Libero System-on-Chip (SoC) throughout the document (SAR 42495). Live at Power-Up (LAPU) has been replaced with 'Instant On'.	NA
Revision 3 (August 2012)	Microblade U1AFS250 and U1AFS1500 devices were added to the product tables.	I – IV
	A sentence pertaining to the analog I/Os was added to the "Specifying I/O States During Programming" section (SAR 34831).	1-9