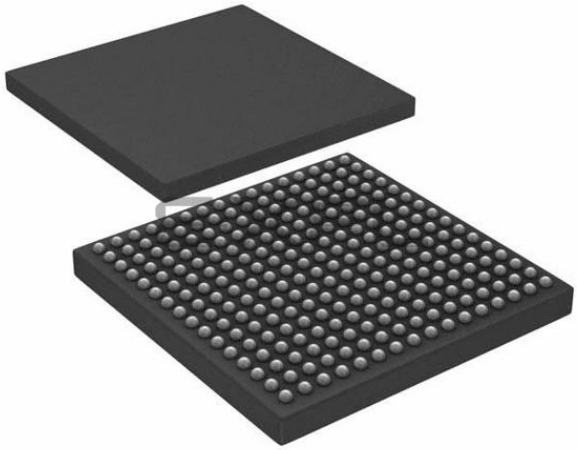

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Understanding Embedded - FPGAs (Field Programmable Gate Array)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details	
Product Status	Active
Number of LABs/CLBs	-
Number of Logic Elements/Cells	-
Total RAM Bits	276480
Number of I/O	119
Number of Gates	1500000
Voltage - Supply	1.425V ~ 1.575V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 85°C (TJ)
Package / Case	256-LBGA
Supplier Device Package	256-FPBGA (17x17)
Purchase URL	https://www.e-xfl.com/product-detail/microchip-technology/m1afs1500-2fg256

Timing Characteristics

Table 2-1 Combinatorial Cell Propagation Delays
Commercial Temperature Range Conditions: T_J = 70°C, Worst-Case VCC = 1.425 V

Combinatorial Cell	Equation	Parameter	–2	–1	Std.	Units
INV	Y = !A	t_{PD}	0.40	0.46	0.54	ns
AND2	Y = A • B	t_{PD}	0.47	0.54	0.63	ns
NAND2	Y = !(A • B)	t_{PD}	0.47	0.54	0.63	ns
OR2	Y = A + B	t_{PD}	0.49	0.55	0.65	ns
NOR2	Y = !(A + B)	t_{PD}	0.49	0.55	0.65	ns
XOR2	Y = A B	t_{PD}	0.74	0.84	0.99	ns
MAJ3	Y = MAJ(A, B, C)	t_{PD}	0.70	0.79	0.93	ns
XOR3	Y = A B C	t_{PD}	0.87	1.00	1.17	ns
MUX2	Y = A !S + B S	t_{PD}	0.51	0.58	0.68	ns
AND3	Y = A • B • C	t_{PD}	0.56	0.64	0.75	ns

Note: For the derating values at specific junction temperature and voltage supply levels, refer to Table 3-7 on page 3-9.

Sample VersaTile Specifications Sequential Module

The Fusion library offers a wide variety of sequential cells, including flip-flops and latches. Each has a data input and optional enable, clear, or preset. In this section, timing characteristics are presented for a representative sample from the library (Figure 2-5). For more details, refer to the IGLOO, ProASIC3, SmartFusion and Fusion Macro Library Guide.

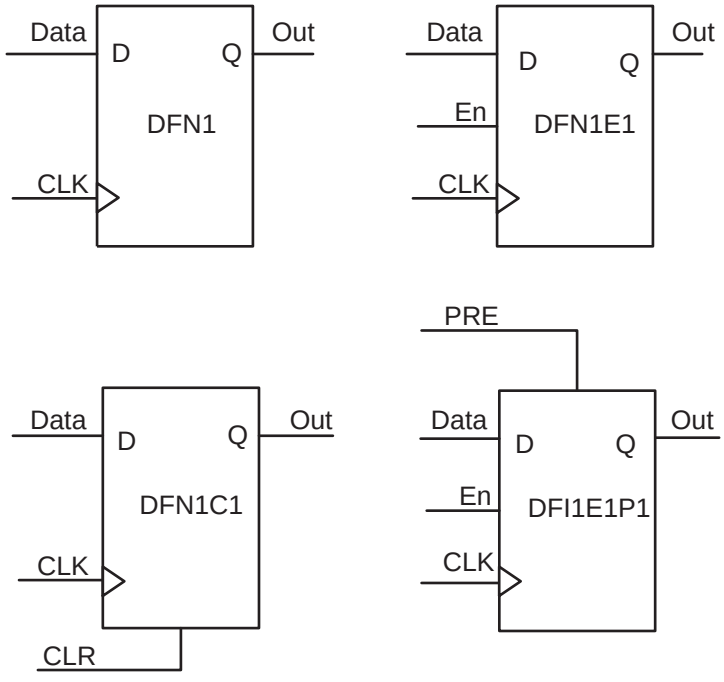


Figure 2-5 Sample of Sequential Cells

VersaNet Timing Characteristics

Global clock delays include the central rib delay, the spine delay, and the row delay. Delays do not include I/O input buffer clock delays, as these are dependent upon I/O standard, and the clock may be driven and conditioned internally by the CCC module. Table 2-5, Table 2-6, Table 2-7, and Table 2-8 on page 2-17 present minimum and maximum global clock delays within the device Minimum and maximum delays are measured with minimum and maximum loading, respectively.

Timing Characteristics

Table 2-5 AFS1500 Global Resource Timing
Commercial Temperature Range Conditions: T_J = 70°C, Worst-Case VCC = 1.425 V

Parameter	Description	–2		–1		Std.		Units
		Min.[1]	Max.[2]	Min.[1]	Max.[2]	Min.[1]	Max.[2]	
t_{RCKL}	Input Low Delay for Global Clock	1.53	1.75	1.74	1.99	2.05	2.34	ns
t_{RCKH}	Input High Delay for Global Clock	1.53	1.79	1.75	2.04	2.05	2.40	ns
$t_{RCKMPWH}$	Minimum Pulse Width High for Global Clock							ns
$t_{RCKMPWL}$	Minimum Pulse Width Low for Global Clock							ns
t_{RCKSW}	Maximum Skew for Global Clock		0.26		0.29		0.34	ns

Notes:

1. Value reflects minimum load. The delay is measured from the CCC output to the clock pin of a sequential element located in a lightly loaded row (single element is connected to the global net).
2. Value reflects maximum load. The delay is measured on the clock pin of the farthest sequential element located in a fully loaded row (all available flip-flops are connected to the global net in the row).
3. For the derating values at specific junction temperature and voltage supply levels, refer to Table 3-7 on page 3-9.

Table 2-6 AFS600 Global Resource Timing
Commercial Temperature Range Conditions: T_J = 70°C, Worst-Case VCC = 1.425 V

Parameter	Description	–2		–1		Std.		Units
		Min.[1]	Max.[2]	Min.[1]	Max.[2]	Min.[1]	Max.[2]	
t_{RCKL}	Input Low Delay for Global Clock	1.27	1.49	1.44	1.70	1.69	2.00	ns
t_{RCKH}	Input High Delay for Global Clock	1.26	1.54	1.44	1.75	1.69	2.06	ns
$t_{RCKMPWH}$	Minimum Pulse Width High for Global Clock							ns
$t_{RCKMPWL}$	Minimum Pulse Width Low for Global Clock							ns
t_{RCKSW}	Maximum Skew for Global Clock		0.27		0.31		0.36	ns

Notes:

1. Value reflects minimum load. The delay is measured from the CCC output to the clock pin of a sequential element located in a lightly loaded row (single element is connected to the global net).
2. Value reflects maximum load. The delay is measured on the clock pin of the farthest sequential element located in a fully loaded row (all available flip-flops are connected to the global net in the row).
3. For the derating values at specific junction temperature and voltage supply levels, refer to Table 3-7 on page 3-9.

Clocking Resources

The Fusion family has a robust collection of clocking peripherals, as shown in the block diagram in Figure 2-16. These on-chip resources enable the creation, manipulation, and distribution of many clock signals. The Fusion integrated RC oscillator produces a 100 MHz clock source with no external components. For systems requiring more precise clock signals, the Fusion family supports an on-chip crystal oscillator circuit. The integrated PLLs in each Fusion device can use the RC oscillator, crystal oscillator, or another on-chip clock signal as a source. These PLLs offer a variety of capabilities to modify the clock source (multiply, divide, synchronize, advance, or delay). Utilizing the CCC found in the popular ProASIC3 family, Fusion incorporates six CCC blocks. The CCCs allow access to Fusion global and local clock distribution nets, as described in the "Global Resources (VersaNets)" section on page 2-11.

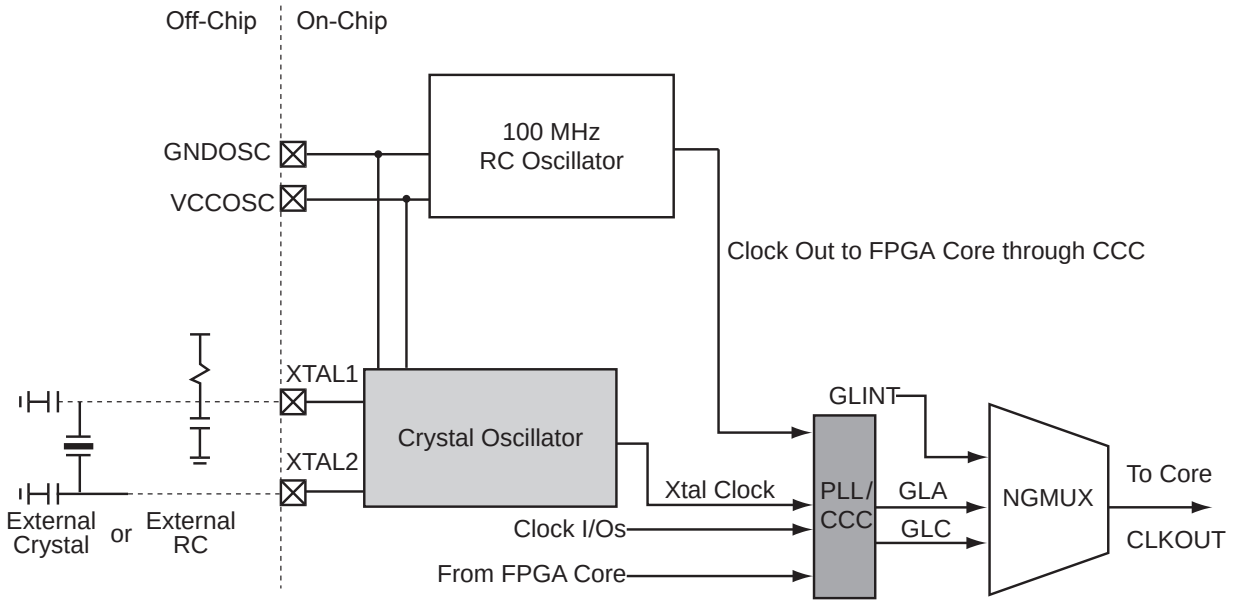


Figure 2-16 Fusion Clocking Options

Clock Conditioning Circuits

In Fusion devices, the CCCs are used to implement frequency division, frequency multiplication, phase shifting, and delay operations.

The CCCs are available in six chip locations each of the four chip corners and the middle of the east and west chip sides.

Each CCC can implement up to three independent global buffers (with or without programmable delay), or a PLL function (programmable frequency division/multiplication, phase shift, and delays) with up to three global outputs. Unused global outputs of a PLL can be used to implement independent global buffers, up to a maximum of three global outputs for a given CCC.

A global buffer can be placed in any of the three global locations (CLKA-GLA, CLKB-GLB, and CLKC-GLC) of a given CCC.

A PLL macro uses the CLKA CCC input to drive its reference clock. It uses the GLA and, optionally, the GLB and GLC global outputs to drive the global networks. A PLL macro can also drive the YB and YC regular core outputs. The GLB (or GLC) global output cannot be reused if the YB (or YC) output is used (Figure 2-19). Refer to the "PLL Macro" section on page 2-27 for more information.

Each global buffer, as well as the PLL reference clock, can be driven from one of the following:

- 3 dedicated single-ended I/Os using a hardwired connection
- 2 dedicated differential I/Os using a hardwired connection
- The FPGA core

The CCC block is fully configurable, either via flash configuration bits set in the programming bitstream or through an asynchronous interface. This asynchronous interface is dynamically accessible from inside the Fusion device to permit changes of parameters (such as divide ratios) during device operation. To increase the versatility and flexibility of the clock conditioning system, the CCC configuration is determined either by the user during the design process, with configuration data being stored in flash memory as part of the device programming procedure, or by writing data into a dedicated shift register during normal device operation. This latter mode allows the user to dynamically reconfigure the CCC without the need for core programming. The shift register is accessed through a simple serial interface. Refer to the "UJTAG Applications in Microsemi s Low-Power Flash Devices" chapter of the Fusion FPGA Fabric User Guide and the "CCC and PLL Characteristics" section on page 2-28 for more information.

FIFO4K18 Description

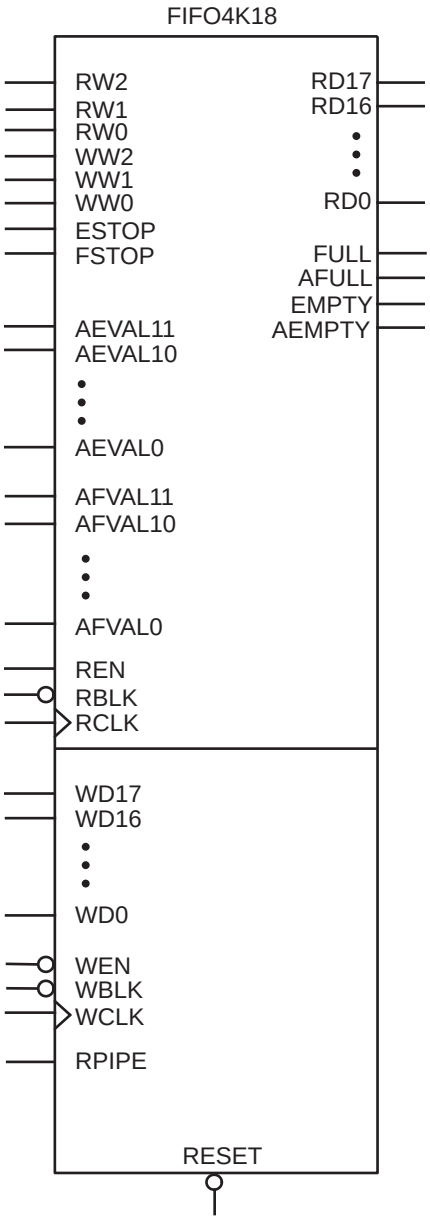


Figure 2-56 FIFO4KX18

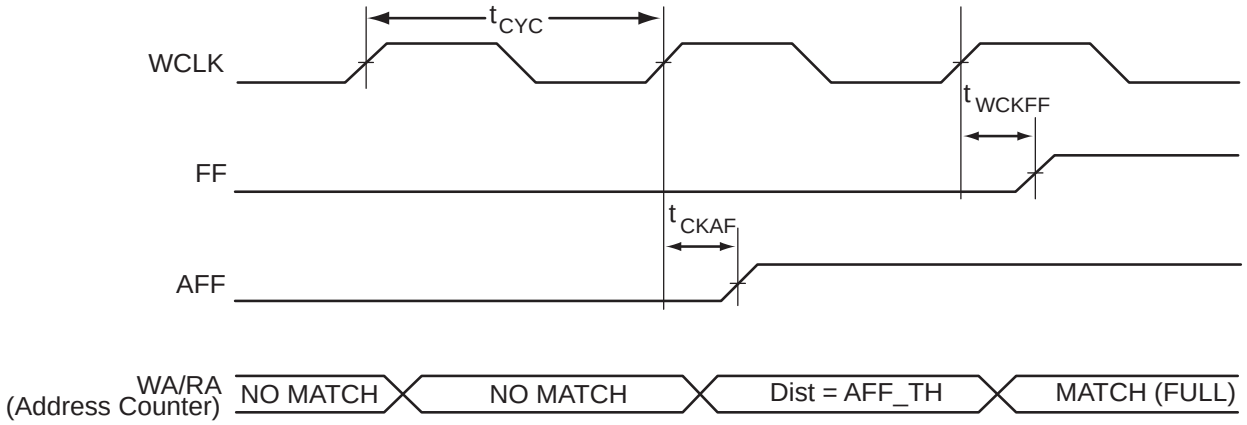


Figure 2-61 FIFO FULL and AFULL Flag Assertion

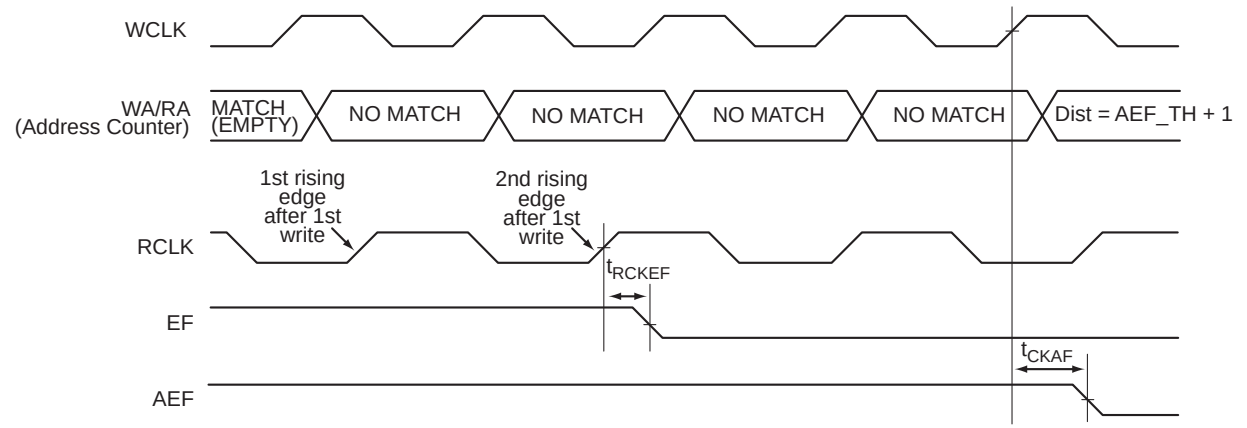


Figure 2-62 FIFO EMPTY Flag and AEMPTY Flag Deassertion

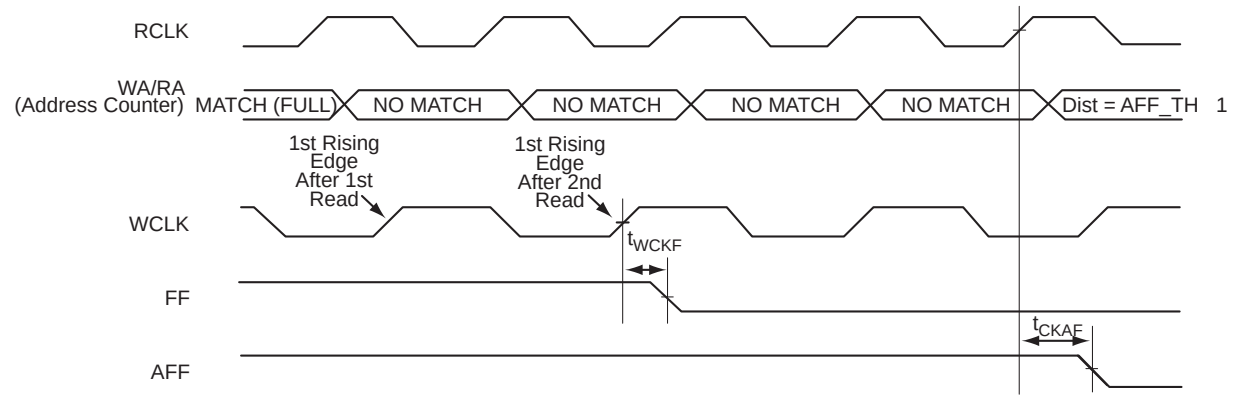


Figure 2-63 FIFO FULL Flag and AFULL Flag Deassertion

Table 2-36 describes each pin in the Analog Block. Each function within the Analog Block will be explained in detail in the following sections.

Table 2-36 Analog Block Pin Description

Signal Name	Number of Bits	Direction	Function	Location of Details
VAREF	1	Input/Output	Voltage reference for ADC	ADC
ADCGNDREF	1	Input	External ground reference	ADC
MODE[3:0]	4	Input	ADC operating mode	ADC
SYSCLK	1	Input	External system clock	
TVC[7:0]	8	Input	Clock divide control	ADC
STC[7:0]	8	Input	Sample time control	ADC
ADCSTART	1	Input	Start of conversion	ADC
PWRDWN	1	Input	ADC comparator power-down if 1. When asserted, the ADC will stop functioning, and the digital portion of the analog block will continue operating. This may result in invalid status flags from the analog block. Therefore, Microsemi does not recommend asserting the PWRDWN pin.	ADC
ADCRESET	1	Input	ADC resets and disables Analog Quad active high	ADC
BUSY	1	Output	1 Running conversion	ADC
CALIBRATE	1	Output	1 Power-up calibration	ADC
DATAVALID	1	Output	1 Valid conversion result	ADC
RESULT[11:0]	12	Output	Conversion result	ADC
TMSTBINT	1	Input	Internal temp. monitor strobe	ADC
SAMPLE	1	Output	1 An analog signal is actively being sampled (stays high during signal acquisition only) 0 No analog signal is being sampled	ADC
VAREFSEL	1	Input	0 = Output internal voltage reference (2.56 V) to VAREF 1 = Input external voltage reference from VAREF and ADCGNDREF	ADC
CHNUMBER[4:0]	5	Input	Analog input channel select	Input multiplexer
ACMCLK	1	Input	ACM clock	ACM
ACMWEN	1	Input	ACM write enable active high	ACM
ACMRESET	1	Input	ACM reset active low	ACM
ACMWDATA[7:0]	8	Input	ACM write data	ACM
ACMRDATA[7:0]	8	Output	ACM read data	ACM
ACMADDR[7:0]	8	Input	ACM address	ACM
CMSTB0 to CMSTB9	10	Input	Current monitor strobe 1 per quad, active high	Analog Quad

Typical scaling factors are given in Table 2-57 on page 2-130, and the gain error (which contributes to the minimum and maximum) is in Table 2-49 on page 2-117.

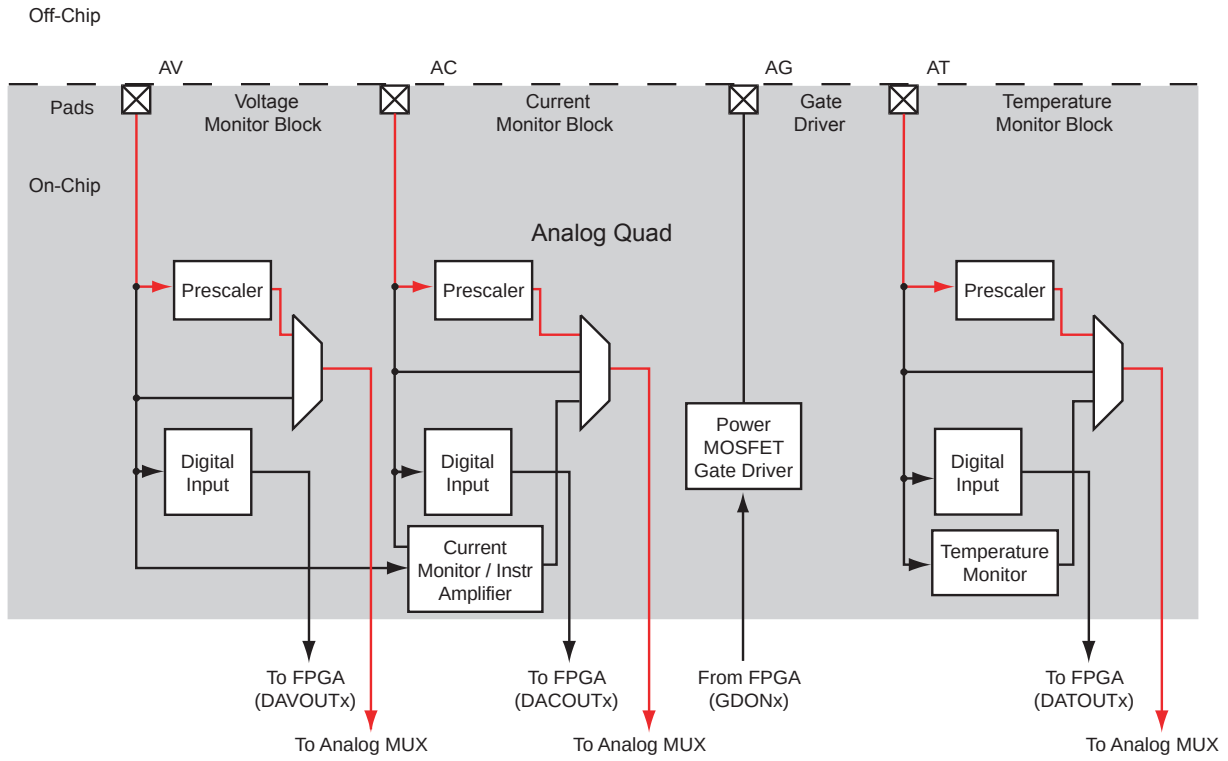


Figure 2-67 Analog Quad Prescaler Input Configuration

Terminology

BW Bandwidth

BW is a range of frequencies that a Channel can handle.

Channel

A channel is define as an analog input configured as one of the Prescaler range shown in Table 2-57 on page 2-130. The channel includes the Prescaler circuit and the ADC.

Channel Gain

Channel Gain is a measured of the deviation of the actual slope from the ideal slope. The slope is measured from the 20% and 80% point.

= ——————

EQ 1

Channel Gain Error

Channel Gain Error is a deviation from the ideal slope of the transfer function. The Prescaler Gain Error is expressed as the percent difference between the actual and ideal, as shown in EQ 2.

= u

EQ 2

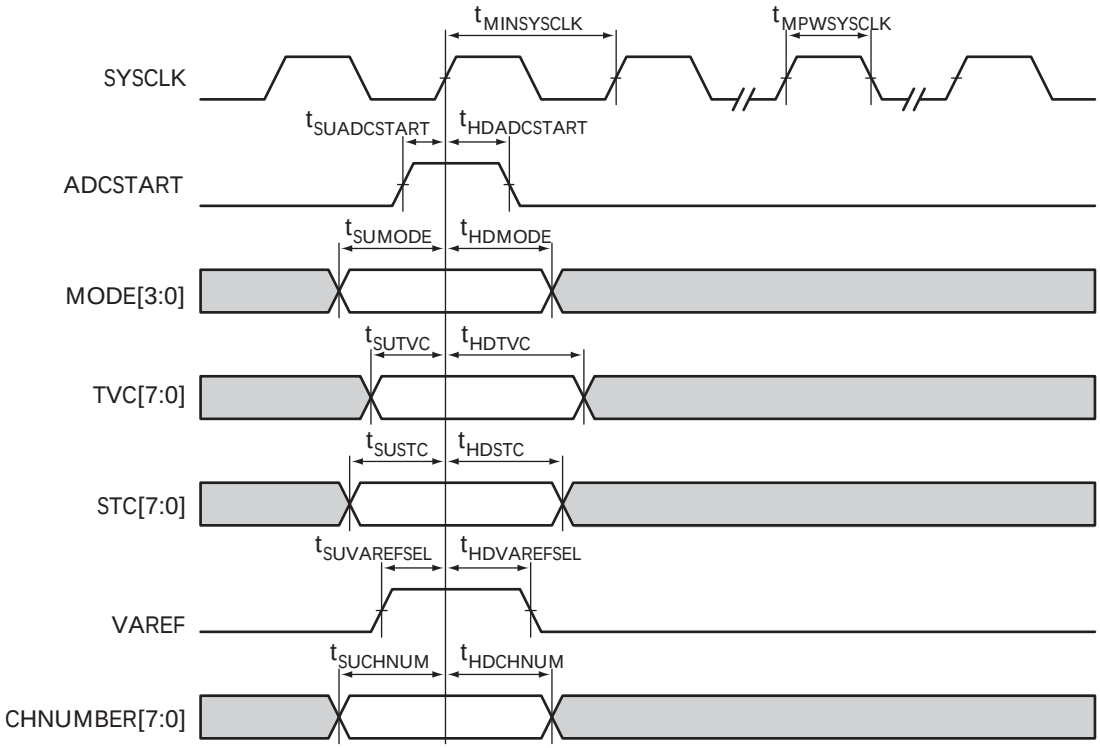
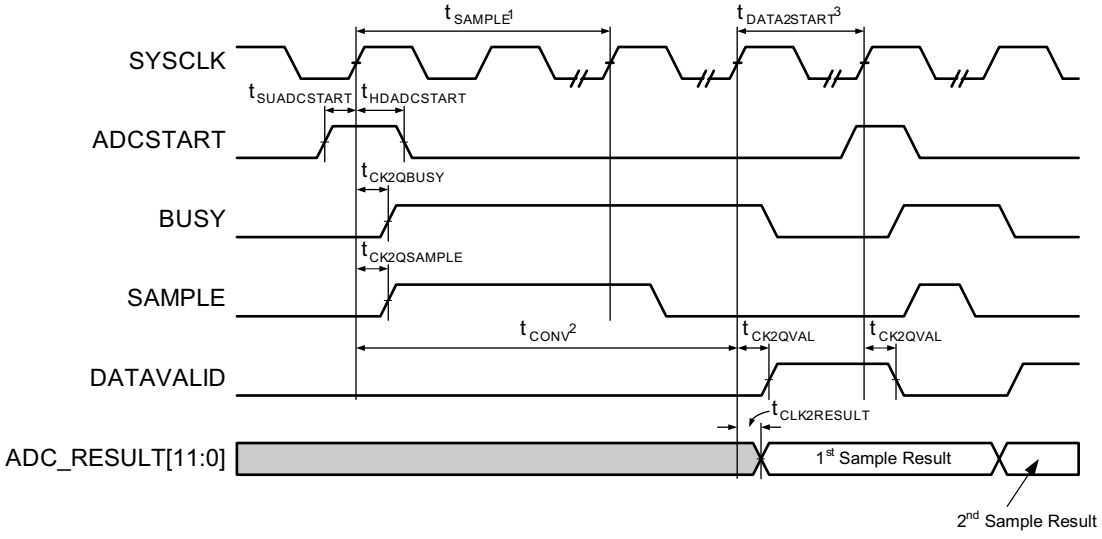


Figure 2-90 Input Setup Time

Standard Conversion



Notes:

1. Refer to EQ 20 on page 2-109 for the calculation on the sample time, t_{SAMPLE}.
2. See EQ 23 on page 2-109 for calculation of the conversion time, t_{CONV}.
3. Minimum time to issue an ADCSTART after DATAVALID is 1 SYSCLK period

Figure 2-91 Standard Conversion Status Signal Timing Diagram

Table 2-78 Fusion Standard I/O St andards—OUT_DRIVE Settings

I/O Standards	OUT_DRIVE (mA)					
	2	4	6	8	Slew	
LVTTL/LVCMOS 3.3 V	3	3	3	3	High	Low
LVCMOS 2.5 V	3	3	3	3	High	Low
LVCMOS 1.8 V	3	3			High	Low
LVCMOS 1.5 V	3				High	Low

Table 2-79 Fusion Advanced I/O Standards—SLEW and OUT_DRIVE Settings

I/O Standards	OUT_DRIVE (mA)							
	2	4	6	8	12	16	Slew	
LVTTL/LVCMOS 3.3 V	3	3	3	3	3	3	High	Low
LVCMOS 2.5 V	3	3	3	3	3		High	Low
LVCMOS 1.8 V	3	3	3	3			High	Low
LVCMOS 1.5 V	3	3					High	Low

Table 2-80 Fusion Pro I/O Standards— SLEW and OUT_DRIVE Settings

I/O Standards	OUT_DRIVE (mA)								
	2	4	6	8	12	16	24	Slew	
LVTTL/LVCMOS 3.3 V	3	3	3	3	3	3	3	High	Low
LVCMOS 2.5 V	3	3	3	3	3	3	3	High	Low
LVCMOS 2.5 V/5.0 V	3	3	3	3	3	3	3	High	Low
LVCMOS 1.8 V	3	3	3	3	3	3		High	Low
LVCMOS 1.5 V	3	3	3	3	3			High	Low

I/O Software Support

In the Fusion development software, default settings have been defined for the various I/O standards supported. Changes can be made to the default settings via the use of attributes; however, not all I/O attributes are applicable for all I/O standards. Table 2-84 and Table 2-85 list the valid I/O attributes that can be manipulated by the user for each I/O standard.

Single-ended I/O standards in Fusion support up to five different drive strengths.

Table 2-84 Fusion Standard and Advanced I/O Attributes vs. I/O Standard Applications

I/O Standards	SLEW (output only)	OUT_DRIVE (output only)	SKEW (all macros with OE)*	RES_PULL	OUT_LOAD (output only)	COMBINE_REGISTER
LVTTL/LVCMOS 3.3 V	3	3	3	3	3	3
LVCMOS 2.5 V	3	3	3	3	3	3
LVCMOS 2.5/5.0 V	3	3	3	3	3	3
LVCMOS 1.8 V	3	3	3	3	3	3
LVCMOS 1.5 V	3	3	3	3	3	3
PCI (3.3 V)			3		3	3
PCI-X (3.3 V)	3		3		3	3
LVDS, BLVDS, M-LVDS			3			3
LVPECL						3

Note: * This feature does not apply to the standard I/O banks, which are the north I/O banks of AFS090 and AFS250 devices

Timing Characteristics

Table 2-112 2.5 V LVCMOS Low Slew
Commercial Temperature Range Conditions: T_J = 70°C, Worst-Case VCC = 1.425 V,
Worst-Case VCCI = 2.3 V
Applicable to Pro I/Os

Drive Strength	Speed Grade	t_{DOUT}	t_{DP}	t_{DIN}	t_{PY}	t_{PYS}	t_{EOUT}	t_{ZL}	t_{ZH}	t_{LZ}	t_{HZ}	t_{ZLS}	t_{ZHS}	Units
4 mA	Std.	0.60	12.00	0.04	1.51	1.66	0.43	12.23	11.61	2.72	2.20	14.46	13.85	ns
	1	0.51	10.21	0.04	1.29	1.41	0.36	10.40	9.88	2.31	1.87	12.30	11.78	ns
	2	0.45	8.96	0.03	1.13	1.24	0.32	9.13	8.67	2.03	1.64	10.80	10.34	ns
8 mA	Std.	0.60	8.73	0.04	1.51	1.66	0.43	8.89	8.01	3.10	2.93	11.13	10.25	ns
	1	0.51	7.43	0.04	1.29	1.41	0.36	7.57	6.82	2.64	2.49	9.47	8.72	ns
	2	0.45	6.52	0.03	1.13	1.24	0.32	6.64	5.98	2.32	2.19	8.31	7.65	ns
12 mA	Std.	0.66	6.77	0.04	1.51	1.66	0.43	6.90	6.11	3.37	3.39	9.14	8.34	ns
	1	0.56	5.76	0.04	1.29	1.41	0.36	5.87	5.20	2.86	2.89	7.77	7.10	ns
	2	0.49	5.06	0.03	1.13	1.24	0.32	5.15	4.56	2.51	2.53	6.82	6.23	ns
16 mA	Std.	0.66	6.31	0.04	1.51	1.66	0.43	6.42	5.73	3.42	3.52	8.66	7.96	ns
	1	0.56	5.37	0.04	1.29	1.41	0.36	5.46	4.87	2.91	3.00	7.37	6.77	ns
	2	0.49	4.71	0.03	1.13	1.24	0.32	4.80	4.28	2.56	2.63	6.47	5.95	ns
24 mA	Std.	0.66	5.93	0.04	1.51	1.66	0.43	6.04	5.70	3.49	4.00	8.28	7.94	ns
	1	0.56	5.05	0.04	1.29	1.41	0.36	5.14	4.85	2.97	3.40	7.04	6.75	ns
	2	0.49	4.43	0.03	1.13	1.24	0.32	4.51	4.26	2.61	2.99	6.18	5.93	ns

Note: For the derating values at specific junction temperature and voltage supply levels, refer to Table 3-7 on page 3-9.

TMS Test Mode Select

The TMS pin controls the use of the IEEE1532 boundary scan pins (TCK, TDI, TDO, TRST). There is an internal weak pull-up resistor on the TMS pin.

TRST Boundary Scan Reset Pin

The TRST pin functions as an active low input to asynchronously initialize (or reset) the boundary scan circuitry. There is an internal weak pull-up resistor on the TRST pin. If JTAG is not used, an external pull-down resistor could be included to ensure the TAP is held in reset mode. The resistor values must be chosen from Table 2-183 and must satisfy the parallel resistance value requirement. The values in Table 2-183 correspond to the resistor recommended when a single device is used and to the equivalent parallel resistor when multiple devices are connected via a JTAG chain.

In critical applications, an upset in the JTAG circuit could allow entering an undesired JTAG state. In such cases, Microsemi recommends tying off TRST to GND through a resistor placed close to the FPGA pin.

Note that to operate at all VJTAG voltages, 500 : to 1 k : will satisfy the requirements.

Special Function Pins

NC No Connect

This pin is not connected to circuitry within the device. These pins can be driven to any voltage or can be left floating with no effect on the operation of the device.

DC Don't Connect

This pin should not be connected to any signals on the PCB. These pins should be left unconnected.

NCAP Negative Capacitor

Negative Capacitor is where the negative terminal of the charge pump capacitor is connected. A capacitor, with a 2.2 µF recommended value, is required to connect between PCAP and NCAP.

PCAP Positive Capacitor

Positive Capacitor is where the positive terminal of the charge pump capacitor is connected. A capacitor, with a 2.2 µF recommended value, is required to connect between PCAP and NCAP.

PUB Push Button

Push button is the connection for the external momentary switch used to turn on the 1.5 V voltage regulator and can be floating if not used.

PTBASE Pass Transistor Base

Pass Transistor Base is the control signal of the voltage regulator. This pin should be connected to the base of the external pass transistor used with the 1.5 V internal voltage regulator and can be floating if not used.

PTEM Pass Transistor Emitter

Pass Transistor Emitter is the feedback input of the voltage regulator.

This pin should be connected to the emitter of the external pass transistor used with the 1.5 V internal voltage regulator and can be floating if not used.

XTAL1 Crystal Oscillator Circuit Input

Input to crystal oscillator circuit. Pin for connecting external crystal, ceramic resonator, RC network, or external clock input. When using an external crystal or ceramic oscillator, external capacitors are also recommended (Please refer to the crystal oscillator manufacturer for proper capacitor value).

If using external RC network or clock input, XTAL1 should be used and XTAL2 left unconnected. In the case where the Crystal Oscillator block is not used, the XTAL1 pin should be connected to GND and the XTAL2 pin should be left floating.

Theta-JA

Junction-to-ambient thermal resistance (θ_{JA}) is determined under standard conditions specified by JEDEC (JESD-51), but it has little relevance in actual performance of the product. It should be used with caution but is useful for comparing the thermal performance of one package to another.

A sample calculation showing the maximum power dissipation allowed for the AFS600-FG484 package under forced convection of 1.0 m/s and 75°C ambient temperature is as follows:

$$\text{Maximum Power Allowed} = \frac{T_{J(MAX)} - T_{A(MAX)}}{\theta_{JA}}$$

EQ 4

where

θ_{JA} = 19.00°C/W (taken from Table 3-6 on page 3-7).

T_A = 75.00°C

$$\text{Maximum Power Allowed} = \frac{100.00°C - 75.00°C}{19.00°C/W} = 1.3\ W$$

EQ 5

The power consumption of a device can be calculated using the Microsemi power calculator. The device's power consumption must be lower than the calculated maximum power dissipation by the package. If the power consumption is higher than the device's maximum allowable power dissipation, a heat sink can be attached on top of the case, or the airflow inside the system must be increased.

Theta-JB

Junction-to-board thermal resistance (θ_{JB}) measures the ability of the package to dissipate heat from the surface of the chip to the PCB. As defined by the JEDEC (JESD-51) standard, the thermal resistance from junction to board uses an isothermal ring cold plate zone concept. The ring cold plate is simply a means to generate an isothermal boundary condition at the perimeter. The cold plate is mounted on a JEDEC standard board with a minimum distance of 5.0 mm away from the package edge.

Theta-JC

Junction-to-case thermal resistance (θ_{JC}) measures the ability of a device to dissipate heat from the surface of the chip to the top or bottom surface of the package. It is applicable for packages used with external heat sinks. Constant temperature is applied to the surface in consideration and acts as a boundary condition. This only applies to situations where all or nearly all of the heat is dissipated through the surface in consideration.

Calculation for Heat Sink

For example, in a design implemented in an AFS600-FG484 package with 2.5 m/s airflow, the power consumption value using the power calculator is 3.00 W. The user-dependent T_a and T_j are given as follows:

T_J = 100.00°C

T_A = 70.00°C

From the datasheet:

θ_{JA} = 17.00°C/W

θ_{JC} = 8.28°C/W

$$P = \frac{T_J - T_A}{\theta_{JA}} = \frac{100°C - 70°C}{17.00\ W} = 1.76\ W$$

EQ 6

Table 3-8 AFS1500 Quiescent Supply Current Characteristics (continued)

Parameter	Description	Conditions	Temp.	Min.	Typ.	Max.	Unit
IJTAG	JTAG I/O quiescent current	Operational standby[4], VJTAG = 3.63 V	T_J = 25°C		80	100	µA
			T_J = 85°C		80	100	µA
			T_J = 100°C		80	100	µA
		Standby mode[5] or Sleep mode[6], VJTAG = 0 V			0	0	µA
IPP	Programming supply current	Non-programming mode, VPUMP = 3.63 V	T_J = 25°C		39	80	µA
			T_J = 85°C		40	80	µA
			T_J = 100°C		40	80	µA
		Standby mode[5] or Sleep mode[6], VPUMP = 0 V			0	0	µA
ICCNVM	Embedded NVM current	Reset asserted, V_{CCNVM} = 1.575 V	T_J = 25°C		50	150	µA
			T_J =85°C		50	150	µA
			T_J = 100°C		50	150	µA
ICCPLL	1.5 V PLL quiescent current	Operational standby , VCCPLL = 1.575 V	T_J = 25°C		130	200	µA
			T_J = 85°C		130	200	µA
			T_J = 100°C		130	200	µA

Notes:

1. ICC is the 1.5 V power supplies, ICC and ICC15A.
2. ICC33A includes ICC33A, ICC33PMP, and ICCOSC.
3. ICCI includes all ICCI0, ICCI1, ICCI2, and ICCI4.
4. Operational standby is when the Fusion device is powered up, all blocks are used, no I/O is toggling, Voltage Regulator is loaded with 200 mA, VCC33PMP is ON, XTAL is ON, and ADC is ON.
5. XTAL is configured as high gain, VCC = VJTAG = VPUMP = 0 V.
6. Sleep Mode, VCC = VJTAG = VPUMP = 0 V.

Table 3-1 • **Absolute Maximum Ratings (continued)**

Symbol	Parameter	Commercial	Industrial	Units
AV, AC	Unpowered, ADC reset asserted or unconfigured	–11.0 to 12.6	–11.0 to 12.0	V
	Analog input (+16 V to +2 V prescaler range)	–0.4 to 12.6	–0.4 to 12.0	V
	Analog input (+1 V to +0.125 V prescaler range)	–0.4 to 3.75	–0.4 to 3.75	V
	Analog input (–16 V to –2 V prescaler range)	–11.0 to 0.4	–11.0 to 0.4	V
	Analog input (–1 V to –0.125 V prescaler range)	–3.75 to 0.4	–3.75 to 0.4	V
	Analog input (direct input to ADC)	–0.4 to 3.75	–0.4 to 3.75	V
	Digital input	–0.4 to 12.6	–0.4 to 12.0	V
AG	Unpowered, ADC reset asserted or unconfigured	–11.0 to 12.6	–11.0 to 12.0	V
	Low Current Mode (1 µA, 3 µA, 10 µA, 30 µA)	–0.4 to 12.6	–0.4 to 12.0	V
	Low Current Mode (–1 µA, –3 µA, –10 µA, –30 µA)	–11.0 to 0.4	–11.0 to 0.4	V
	High Current Mode [3]	–11.0 to 12.6	–11.0 to 12.0	V
AT	Unpowered, ADC reset asserted or unconfigured	–0.4 to 16.0	–0.4 to 15.0	V
	Analog input (+16 V, 4 V prescaler range)	–0.4 to 16.0	–0.4 to 15.0	V
	Analog input (direct input to ADC)	–0.4 to 3.75	–0.4 to 3.75	V
	Digital input	–0.4 to 16.0	–0.4 to 15.0	V
T_{STG} [4]	Storage temperature	–65 to +150		°C
T_J [4]	Junction temperature	+125		°C

Notes:

1. *The device should be operated within the limits specified by the datasheet. During transitions, the input signal may undershoot or overshoot according to the limits shown in Table 3-4 on page 3-4.*
2. *Analog data not valid beyond 3.65 V.*
3. *The high current mode has a maximum power limit of 20 mW. Appropriate current limit resistors must be used, based on voltage on the pad.*
4. *For flash programming and retention maximum limits, refer to Table 3-5 on page 3-5. For recommended operating limits refer to Table 3-2 on page 3-3.*