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Understanding Embedded - FPGAs (Field Programmable Gate Array)

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Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details	
Product Status	Active
Number of LABs/CLBs	-
Number of Logic Elements/Cells	-
Total RAM Bits	276480
Number of I/O	119
Number of Gates	1500000
Voltage - Supply	1.425V ~ 1.575V
Mounting Type	Surface Mount
Operating Temperature	-40°C ~ 100°C (TJ)
Package / Case	256-LBGA
Supplier Device Package	256-FPBGA (17x17)
Purchase URL	https://www.e-xfl.com/product-detail/microchip-technology/m1afs1500-2fg256i

The NGMUX macro is simplified to show the two clock options that have been selected by the GLMUXCFG[1:0] bits. Figure 2-25 illustrates the NGMUX macro. During design, the two clock sources are connected to CLK0 and CLK1 and are controlled by GLMUXSEL[1:0] to determine which signal is to be passed through the MUX.

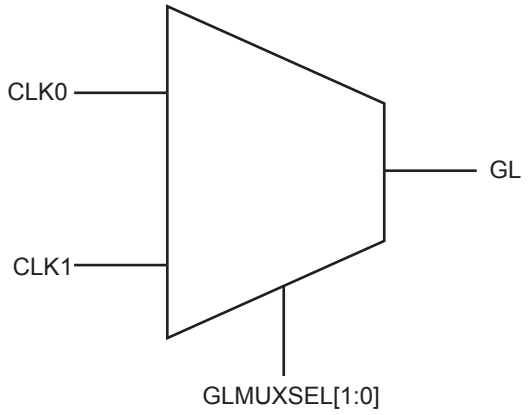

Figure 2-25 • **NGMUX Macro**

The sequence of switching between two clock sources (from CLK0 to CLK1) is as follows (Figure 2-26):

- GLMUXSEL[1:0] transitions to initiate a switch.
- GL drives one last complete CLK0 positive pulse (i.e., one rising edge followed by one falling edge).
- From that point, GL stays Low until the second rising edge of CLK1 occurs.
- At the second CLK1 rising edge, GL will begin to continuously deliver the CLK1 signal.
- Minimum t_{sw} = 0.05 ns at 25°C (typical conditions)

For examples of NGMUX operation, refer to the *Fusion FPGA Fabric User Guide*.

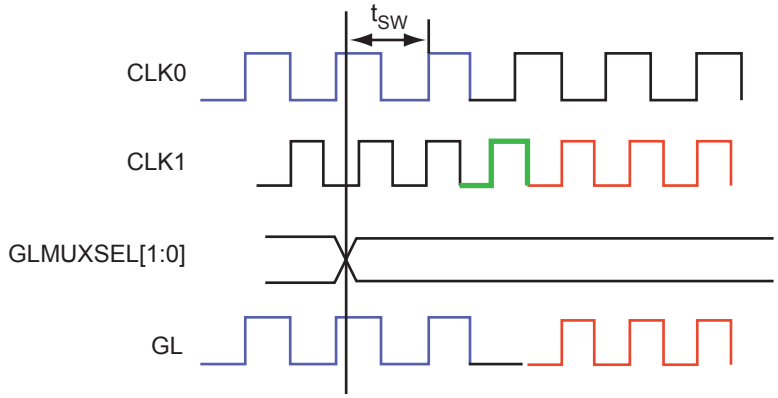

Figure 2-26 • **NGMUX Waveform**

RAM512X18 Description

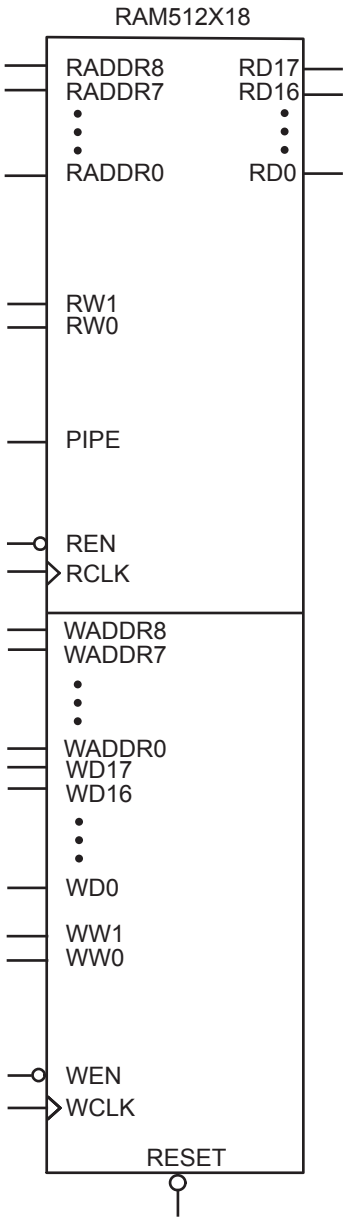

Figure 2-49 • **RAM512X18**

ESTOP, FSTOP

ESTOP is used to stop the FIFO read counter from further counting once the FIFO is empty (i.e., the EMPTY flag goes High). A High on this signal inhibits the counting.

FSTOP is used to stop the FIFO write counter from further counting once the FIFO is full (i.e., the FULL flag goes High). A High on this signal inhibits the counting.

For more information on these signals, refer to the "ESTOP and FSTOP Usage" section on page 2-70.

FULL, EMPTY

When the FIFO is full and no more data can be written, the FULL flag asserts High. The FULL flag is synchronous to WCLK to inhibit writing immediately upon detection of a full condition and to prevent overflows. Since the write address is compared to a resynchronized (and thus time-delayed) version of the read address, the FULL flag will remain asserted until two WCLK active edges after a read operation eliminates the full condition.

When the FIFO is empty and no more data can be read, the EMPTY flag asserts High. The EMPTY flag is synchronous to RCLK to inhibit reading immediately upon detection of an empty condition and to prevent underflows. Since the read address is compared to a resynchronized (and thus time-delayed) version of the write address, the EMPTY flag will remain asserted until two RCLK active edges after a write operation removes the empty condition.

For more information on these signals, refer to the "FIFO Flag Usage Considerations" section on page 2-70.

AFULL, AEMPTY

These are programmable flags and will be asserted on the threshold specified by AFVAL and AEVAL, respectively.

When the number of words stored in the FIFO reaches the amount specified by AEVAL while reading, the AEMPTY output will go High. Likewise, when the number of words stored in the FIFO reaches the amount specified by AFVAL while writing, the AFULL output will go High.

AFVAL, AEVAL

The AEVAL and AFVAL pins are used to specify the almost-empty and almost-full threshold values, respectively. They are 12-bit signals. For more information on these signals, refer to "FIFO Flag Usage Considerations" section.

ESTOP and FSTOP Usage

The ESTOP pin is used to stop the read counter from counting any further once the FIFO is empty (i.e., the EMPTY flag goes High). Likewise, the FSTOP pin is used to stop the write counter from counting any further once the FIFO is full (i.e., the FULL flag goes High).

The FIFO counters in the Fusion device start the count at 0, reach the maximum depth for the configuration (e.g., 511 for a 512×9 configuration), and then restart at 0. An example application for the ESTOP, where the read counter keeps counting, would be writing to the FIFO once and reading the same content over and over without doing another write.

FIFO Flag Usage Considerations

The AEVAL and AFVAL pins are used to specify the 12-bit AEMPTY and AFULL threshold values, respectively. The FIFO contains separate 12-bit write address (WADDR) and read address (RADDR) counters. WADDR is incremented every time a write operation is performed, and RADDR is incremented every time a read operation is performed. Whenever the difference between WADDR and RADDR is greater than or equal to AFVAL, the AFULL output is asserted. Likewise, whenever the difference between WADDR and RADDR is less than or equal to AEVAL, the AEMPTY output is asserted. To handle different read and write aspect ratios, AFVAL and AEVAL are expressed in terms of total data bits instead of total data words. When users specify AFVAL and AEVAL in terms of read or write words, the SmartGen tool translates them into bit addresses and configures these signals automatically. SmartGen configures the AFULL flag to assert when the write address exceeds the read address by at least a predefined value. In a 2k×8 FIFO, for example, a value of 1,500 for AFVAL means that the AFULL flag will be asserted after a write when the difference between the write address and the read address reaches 1,500 (there have been at least 1500 more writes than reads). It will stay asserted until the difference between the write and read addresses drops below 1,500.

Refer to Table 2-46 on page 2-109 and the "Acquisition Time or Sample Time Control" section on page 2-107

$$t_{sample} = (2 + STC) \times t_{ADCCLK}$$

EQ 20

STC: Sample Time Control value (0–255)

t_{SAMPLE} is the sample time

Table 2-46 • **STC Bits Function**

Name	Bits	Function
STC	[7:0]	Sample time control

Sample time is computed based on the period of ADCCLK.

Distribution Phase

The second phase is called the distribution phase. During distribution phase, the ADC computes the equivalent digital value from the value stored in the input capacitor. In this phase, the output signal SAMPLE goes back to '0', indicating the sample is completed; but the BUSY signal remains '1', indicating the ADC is still busy for distribution. The distribution time depends strictly on the number of bits. If the ADC is configured as a 10-bit ADC, then 10 ADCCLK cycles are needed. EQ 8 describes the distribution time.

$$t_{distrib} = N \times t_{ADCCLK}$$

EQ 21

N: Number of bits

Post-Calibration Phase

The last phase is the post-calibration phase. This is an optional phase. The post-calibration phase takes two ADCCLK cycles. The output BUSY signal will remain '1' until the post-calibration phase is completed. If the post-calibration phase is skipped, then the BUSY signal goes to '0' after distribution phase. As soon as BUSY signal goes to '0', the DATAVALID signal goes to '1', indicating the digital result is available on the RESULT output signals. DATAVAILD will remain '1' until the next ADCSTART is asserted. Microsemi recommends enabling post-calibration to compensate for drift and temperature-dependent effects. This ensures that the ADC remains consistent over time and with temperature. The post-calibration phase is enabled by bit 3 of the Mode register. EQ 9 describes the post-calibration time.

$$t_{post\text{-}cal} = MODE[3] \times (2 \times t_{ADCCLK})$$

EQ 22

MODE[3]: Bit 3 of the Mode register, described in Table 2-41 on page 2-106.

The calculation for the conversion time for the ADC is summarized in EQ 23.

$$t_{conv} = t_{sync\_read} + t_{sample} + t_{distrib} + t_{post\text{-}cal} + t_{sync\_write}$$

EQ 23

t_{conv}: conversion time

t_{sync_read}: maximum time for a signal to synchronize with SYSCLK. For calculation purposes, the worst case is a period of SYSCLK, t_{SYSCLK}.

t_{sample}: Sample time

$t_{distrib}$: Distribution time

$t_{post\text{-}cal}$: Post-calibration time

t_{sync_write}: Maximum time for a signal to synchronize with SYSCLK. For calculation purposes, the worst case is a period of SYSCLK, t_{SYSCLK}.

Typical Performance Characteristics

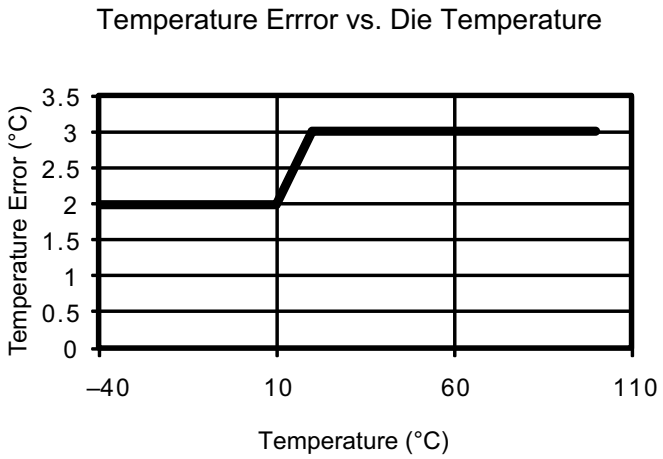

Figure 2-94 • **Temperature Error**

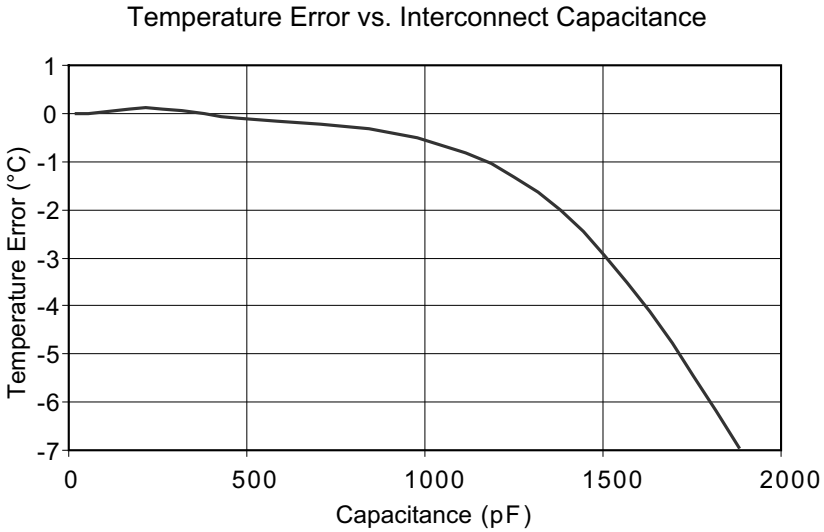

Figure 2-95 • **Effect of External Sensor Capacitance**

Table 2-50 • ADC Characteristics in Direct Input Mode (continued)
Commercial Temperature Range Conditions, T_J = 85°C (unless noted otherwise),
Typical: VCC33A = 3.3 V, VCC = 1.5 V

Parameter	Description	Condition	Min.	Typ.	Max.	Units
Dynamic Performance						
SNR	Signal-to-Noise Ratio	8-bit mode	48.0	49.5		dB
		10-bit mode	58.0	60.0		dB
		12-bit mode	62.9	64.5		dB
SINAD	Signal-to-Noise Distortion	8-bit mode	47.6	49.5		dB
		10-bit mode	57.4	59.8		dB
		12-bit mode	62.0	64.2		dB
THD	Total Harmonic Distortion	8-bit mode		–74.4	–63.0	dBc
		10-bit mode		–78.3	–63.0	dBc
		12-bit mode		–77.9	–64.4	dBc
ENOB	Effective Number of Bits	8-bit mode	7.6	7.9		bits
		10-bit mode	9.5	9.6		bits
		12-bit mode	10.0	10.4		bits
Conversion Rate						
	Conversion Time	8-bit mode	1.7			µs
		10-bit mode	1.8			µs
		12-bit mode	2			µs
	Sample Rate	8-bit mode			600	Ksps
		10-bit mode			550	Ksps
		12-bit mode			500	Ksps

Notes:

1. *Accuracy of the external reference is 2.56 V ± 4.6 mV.*
2. *Data is based on characterization.*
3. *The sample rate is time-shared among active analog inputs.*

Similarly,

Min. Output Voltage = (Max. Negative input offset) + (Input Voltage x Max. Negative Channel Gain)
= (–88 mV) + (5 V x 0.96) = **4.712 V**

Calculating Accuracy for a Calibrated Analog Channel

Formula

For a given prescaler range, EQ 31 gives the output voltage.

$$Output\ Voltage = Channel\ Error\ in\ V + Input\ Voltage \quad EQ\ 31$$

where

Channel Error in V = Total Channel Error in LSBs x Equivalent voltage per LSB

Example

Input Voltage = 5 V

Chosen Prescaler range = 8 V range

Refer to Table 2-52 on page 2-123.

Max. Output Voltage = Max. Positive Channel Error in V + Input Voltage

Max. Positive Channel Error in V = (6 LSB) × (8 mV per LSB in 10-bit mode) = 48 mV

Max. Output Voltage = 48 mV + 5 V = **5.048 V**

Similarly,

Min. Output Voltage = Max. Negative Channel Error in V + Input Voltage = (–48 mV) + 5 V = **4.952 V**

Calculating LSBs from a Given Error Budget

Formula

For a given prescaler range,

LSB count = ± (Input Voltage × Required % error) / (Equivalent voltage per LSB)

Example

Input Voltage = 3.3 V

Required error margin= 1%

Refer to Table 2-52 on page 2-123.

Equivalent voltage per LSB = 16 mV for a 16V prescaler, with ADC in 10-bit mode

LSB Count = ± (5.0 V × 1%) / (0.016)

LSB Count = **± 3.125**

Equivalent voltage per LSB = **8 mV** for an 8 V prescaler, with ADC in 10-bit mode

LSB Count = ± (5.0 V × 1%) / (0.008)

LSB Count = **± 6.25**

The 8 V prescaler satisfies the calculated LSB count accuracy requirement (see Table 2-52 on page 2-123).

Summary of I/O Timing Characteristics – Default I/O Software Settings

Table 2-90 • Summary of AC Measuring Points Applicable to All I/O Bank Types

Standard	Input Reference Voltage (VREF_TYP)	Board Termination Voltage (VTT_REF)	Measuring Trip Point (Vtrip)
3.3 V LVTTL / 3.3 V LVCMOS	–	–	1.4 V
2.5 V LVCMOS	–	–	1.2 V
1.8 V LVCMOS	–	–	0.90 V
1.5 V LVCMOS	–	–	0.75 V
3.3 V PCI	–	–	0.285 * VCCI (RR) 0.615 * VCCI (FF))
3.3 V PCI-X	–	–	0.285 * VCCI (RR) 0.615 * VCCI (FF)
3.3 V GTL	0.8 V	1.2 V	VREF
2.5 V GTL	0.8 V	1.2 V	VREF
3.3 V GTL+	1.0 V	1.5 V	VREF
2.5 V GTL+	1.0 V	1.5 V	VREF
HSTL (I)	0.75 V	0.75 V	VREF
HSTL (II)	0.75 V	0.75 V	VREF
SSTL2 (I)	1.25 V	1.25 V	VREF
SSTL2 (II)	1.25 V	1.25 V	VREF
SSTL3 (I)	1.5 V	1.485 V	VREF
SSTL3 (II)	1.5 V	1.485 V	VREF
LVDS	–	–	Cross point
LVPECL	–	–	Cross point

Table 2-91 • I/O AC Parameter Definitions

Parameter	Definition
t_{DP}	Data to Pad delay through the Output Buffer
t_{PY}	Pad to Data delay through the Input Buffer with Schmitt trigger disabled
t_{DOUT}	Data to Output Buffer delay through the I/O interface
t_{EOUT}	Enable to Output Buffer Tristate Control delay through the I/O interface
t_{DIN}	Input Buffer to Data delay through the I/O interface
t_{PYS}	Pad to Data delay through the Input Buffer with Schmitt trigger enabled
t_{HZ}	Enable to Pad delay through the Output Buffer—High to Z
t_{ZH}	Enable to Pad delay through the Output Buffer—Z to High
t_{LZ}	Enable to Pad delay through the Output Buffer—Low to Z
t_{ZL}	Enable to Pad delay through the Output Buffer—Z to Low
t_{ZHS}	Enable to Pad delay through the Output Buffer with delayed enable—Z to High
t_{ZLS}	Enable to Pad delay through the Output Buffer with delayed enable—Z to Low

Table 2-96 • I/O Output Buffer Maximum Resistances [1] (continued)

Standard	Drive Strength	$R_{PULL\text{-}DOWN}$ (ohms) [2]	$R_{PULL\text{-}UP}$ (ohms) [3]
HSTL (I)	8 mA	50	50
HSTL (II)	15 mA	25	25
SSTL2 (I)	17 mA	27	31
SSTL2 (II)	21 mA	13	15
SSTL3 (I)	16 mA	44	69
SSTL3 (II)	24 mA	18	32
Applicable to Advanced I/O Banks			
3.3 V LVTTL / 3.3 V LVCMOS	2 mA	100	300
	4 mA	100	300
	6 mA	50	150
	8 mA	50	150
	12 mA	25	75
	16 mA	17	50
	24 mA	11	33
2.5 V LVCMOS	2 mA	100	200
	4 mA	100	200
	6 mA	50	100
	8 mA	50	100
	12 mA	25	50
	16 mA	20	40
	24 mA	11	22
1.8 V LVCMOS	2 mA	200	225
	4 mA	100	112
	6 mA	50	56
	8 mA	50	56
	12 mA	20	22
	16 mA	20	22
1.5 V LVCMOS	2 mA	200	224
	4 mA	100	112
	6 mA	67	75
	8 mA	33	37
	12 mA	33	37
3.3 V PCI/PCI-X	Per PCI/PCI-X specification	25	75

Notes:

1. *These maximum values are provided for informational reasons only. Minimum output buffer resistance values depend on VCC, drive strength selection, temperature, and process. For board design considerations and detailed output buffer resistances, use the corresponding IBIS models located on the Microsemi SoC Products Group website: http://www.microsemi.com/soc/techdocs/models/ibis.html.*
2. $R_{(PULL\text{-}DOWN\text{-}MAX)} = VOLspec / I_{OLspec}$
3. $R_{(PULL\text{-}UP\text{-}MAX)} = (VCCImax – VOHspec) / IOHspec$

Table 2-130 • 1.5 V LVCMOS Low Slew
Commercial Temperature Range Conditions: T_J = 70°C, Worst-Case VCC = 1.425 V, Worst-Case VCCI = 1.4 V
Applicable to Advanced I/Os

Drive Strength	Speed Grade	t_{DOUT}	t_{DP}	t_{DIN}	t_{PY}	t_{EOUT}	t_{ZL}	t_{ZH}	t_{LZ}	t_{HZ}	t_{ZLS}	t_{ZHS}	Units
2 mA	Std.	0.66	12.78	0.04	1.31	0.43	12.81	12.78	3.40	2.64	15.05	15.02	ns
	–1	0.56	10.87	0.04	1.11	0.36	10.90	10.87	2.89	2.25	12.80	12.78	ns
	–2	0.49	9.55	0.03	0.98	0.32	9.57	9.55	2.54	1.97	11.24	11.22	ns
4 mA	Std.	0.66	10.01	0.04	1.31	0.43	10.19	9.55	3.75	3.27	12.43	11.78	ns
	–1	0.56	8.51	0.04	1.11	0.36	8.67	8.12	3.19	2.78	10.57	10.02	ns
	–2	0.49	7.47	0.03	0.98	0.32	7.61	7.13	2.80	2.44	9.28	8.80	ns
8 mA	Std.	0.66	9.33	0.04	1.31	0.43	9.51	8.89	3.83	3.43	11.74	11.13	ns
	–1	0.56	7.94	0.04	1.11	0.36	8.09	7.56	3.26	2.92	9.99	9.47	ns
	–2	0.49	6.97	0.03	0.98	0.32	7.10	6.64	2.86	2.56	8.77	8.31	ns
12 mA	Std.	0.66	8.91	0.04	1.31	0.43	9.07	8.89	3.95	4.05	11.31	11.13	ns
	–1	0.56	7.58	0.04	1.11	0.36	7.72	7.57	3.36	3.44	9.62	9.47	ns
	–2	0.49	6.65	0.03	0.98	0.32	6.78	6.64	2.95	3.02	8.45	8.31	ns

Note: For the derating values at specific junction temperature and voltage supply levels, refer to Table 3-7 on page 3-9.

Table 2-131 • 1.5 V LVCMOS High Slew
Commercial Temperature Range Conditions: T_J = 70°C, Worst-Case VCC = 1.425 V, Worst-Case VCCI = 1.4 V
Applicable to Advanced I/Os

Drive Strength	Speed Grade	t_{DOUT}	t_{DP}	t_{DIN}	t_{PY}	t_{EOUT}	t_{ZL}	t_{ZH}	t_{LZ}	t_{HZ}	t_{ZLS}	t_{ZHS}	Units
2 mA	Std.	0.66	8.36	0.04	1.44	0.43	6.82	8.36	3.39	2.77	9.06	10.60	ns
	–1	0.56	7.11	0.04	1.22	0.36	5.80	7.11	2.88	2.35	7.71	9.02	ns
	–2	0.49	6.24	0.03	1.07	0.32	5.10	6.24	2.53	2.06	6.76	7.91	ns
4 mA	Std.	0.66	5.31	0.04	1.44	0.43	4.85	5.31	3.74	3.40	7.09	7.55	ns
	–1	0.56	4.52	0.04	1.22	0.36	4.13	4.52	3.18	2.89	6.03	6.42	ns
	–2	0.49	3.97	0.03	1.07	0.32	3.62	3.97	2.79	2.54	5.29	5.64	ns
8 mA	Std.	0.66	4.67	0.04	1.44	0.43	4.55	4.67	3.82	3.56	6.78	6.90	ns
	–1	0.56	3.97	0.04	1.22	0.36	3.87	3.97	3.25	3.03	5.77	5.87	ns
	–2	0.49	3.49	0.03	1.07	0.32	3.40	3.49	2.85	2.66	5.07	5.16	ns
12 mA	Std.	0.66	4.08	0.04	1.44	0.43	4.15	3.58	3.94	4.20	6.39	5.81	ns
	–1	0.56	3.47	0.04	1.22	0.36	3.53	3.04	3.36	3.58	5.44	4.95	ns
	–2	0.49	3.05	0.03	1.07	0.32	3.10	2.67	2.95	3.14	4.77	4.34	ns

Note: For the derating values at specific junction temperature and voltage supply levels, refer to Table 3-7 on page 3-9.

***Table 2-174* • Parameter Definitions and Measuring Nodes**

Parameter Name	Parameter Definition	Measuring Nodes (from, to)*
t_{OCLKQ}	Clock-to-Q of the Output Data Register	H, DOUT
t_{OSUD}	Data Setup Time for the Output Data Register	F, H
t_{OHD}	Data Hold Time for the Output Data Register	F, H
t_{OSUE}	Enable Setup Time for the Output Data Register	G, H
t_{OHE}	Enable Hold Time for the Output Data Register	G, H
t_{OPRE2Q}	Asynchronous Preset-to-Q of the Output Data Register	L,DOUT
$t_{OREMPRE}$	Asynchronous Preset Removal Time for the Output Data Register	L, H
$t_{ORECPRE}$	Asynchronous Preset Recovery Time for the Output Data Register	L, H
t_{OECLKQ}	Clock-to-Q of the Output Enable Register	H, EOUT
t_{OESUD}	Data Setup Time for the Output Enable Register	J, H
t_{OEHD}	Data Hold Time for the Output Enable Register	J, H
t_{OESUE}	Enable Setup Time for the Output Enable Register	K, H
t_{OEHE}	Enable Hold Time for the Output Enable Register	K, H
$t_{OEPRE2Q}$	Asynchronous Preset-to-Q of the Output Enable Register	I, EOUT
$t_{OEREMPRE}$	Asynchronous Preset Removal Time for the Output Enable Register	I, H
$t_{OERECPRE}$	Asynchronous Preset Recovery Time for the Output Enable Register	I, H
t_{ICLKQ}	Clock-to-Q of the Input Data Register	A, E
t_{ISUD}	Data Setup Time for the Input Data Register	C, A
t_{IHD}	Data Hold Time for the Input Data Register	C, A
t_{ISUE}	Enable Setup Time for the Input Data Register	B, A
t_{IHE}	Enable Hold Time for the Input Data Register	B, A
t_{IPRE2Q}	Asynchronous Preset-to-Q of the Input Data Register	D, E
$t_{IREMPRE}$	Asynchronous Preset Removal Time for the Input Data Register	D, A
$t_{IRECPRE}$	Asynchronous Preset Recovery Time for the Input Data Register	D, A

*Note: *See Figure 2-137 on page 2-212 for more information.*

User-Defined Supply Pins

VREF **I/O Voltage Reference**

Reference voltage for I/O minibanks. Both AFS600 and AFS1500 (north bank only) support Microsemi Pro I/O. These I/O banks support voltage reference standard I/O. The VREF pins are configured by the user from regular I/Os, and any I/O in a bank, except JTAG I/Os, can be designated as the voltage reference I/O. Only certain I/O standards require a voltage reference—HSTL (I) and (II), SSTL2 (I) and (II), SSTL3 (I) and (II), and GTL/GTL+. One VREF pin can support the number of I/Os available in its minibank.

VAREF **Analog Reference Voltage**

The Fusion device can be configured to generate a 2.56 V internal reference voltage that can be used by the ADC. While using the internal reference, the reference voltage is output on the VAREF pin for use as a system reference. If a different reference voltage is required, it can be supplied by an external source and applied to this pin. The valid range of values that can be supplied to the ADC is 1.0 V to 3.3 V. When VAREF is internally generated by the Fusion device, a bypass capacitor must be connected from this pin to ground. The value of the bypass capacitor should be between 3.3 µF and 22 µF, which is based on the needs of the individual designs. The choice of the capacitor value has an impact on the settling time it takes the VAREF signal to reach the required specification of 2.56 V to initiate valid conversions by the ADC. If the lower capacitor value is chosen, the settling time required for VAREF to achieve 2.56 V will be shorter than when selecting the larger capacitor value. The above range of capacitor values supports the accuracy specification of the ADC, which is detailed in the datasheet. Designers choosing the smaller capacitor value will not obtain as much margin in the accuracy as that achieved with a larger capacitor value. Depending on the capacitor value selected in the Analog System Builder, a tool in Libero SoC, an automatic delay circuit will be generated using logic tiles available within the FPGA to ensure that VAREF has achieved the 2.56 V value. Microsemi recommends customers use 10 µF as the value of the bypass capacitor. Designers choosing to use an external VAREF need to ensure that a stable and clean VAREF source is supplied to the VAREF pin before initiating conversions by the ADC. Designers should also make sure that the ADCRESET signal is deasserted before initiating valid conversions.[2]

If the user connects VAREF to external 3.3 V on their board, the internal VAREF driving OpAmp tries to bring the pin down to the nominal 2.56 V until the device is programmed and up/functional. Under this scenario, it is recommended to connect an external 3.3 V supply through a ~1 KOhm resistor to limit current, along with placing a 10-100nF capacitor between VAREF and GNDA.

User Pins

I/O **User Input/Output**

The I/O pin functions as an input, output, tristate, or bidirectional buffer. Input and output signal levels are compatible with the I/O standard selected. Unused I/O pins are configured as inputs with pull-up resistors.

During programming, I/Os become tristated and weakly pulled up to VCCI. With the VCCI and VCC supplies continuously powered up, when the device transitions from programming to operating mode, the I/Os get instantly configured to the desired user configuration.

Unused I/Os are configured as follows:

- Output buffer is disabled (with tristate value of high impedance)
- Input buffer is disabled (with tristate value of high impedance)
- Weak pull-up is programmed

A*xy* **Analog Input/Output**

Analog I/O pin, where *x* is the analog pad type (C = current pad, G = Gate driver pad, T = Temperature pad, V = Voltage pad) and *y* is the Analog Quad number (0 to 9). There is a minimum 1 MΩ to ground on AV, AC, and AT. This pin can be left floating when it is unused.

2. *The ADC is functional with an external reference down to 1V, however to meet the performance parameters highlighted in the datasheet refer to the VAREF specification in Table 3-2 on page 3-3.*

Calculating Power Dissipation

Quiescent Supply Current

Table 3-8 • **AFS1500 Quiescent Supply Current Characteristics**

Parameter	Description	Conditions	Temp.	Min.	Typ.	Max.	Unit
ICC[1]	1.5 V quiescent current	Operational standby[4], VCC = 1.575 V	T_J = 25°C		20	40	mA
			T_J = 85°C		32	65	mA
			T_J = 100°C		59	120	mA
		Standby mode[5] or Sleep mode[6], VCC = 0 V			0	0	µA
ICC33[2]	3.3 V analog supplies current	Operational standby[4], VCC33 = 3.63 V	T_J = 25°C		9.8	13	mA
			T_J = 85°C		10.7	14	mA
			T_J = 100°C		10.8	15	mA
		Operational standby, only Analog Quad and –3.3 V output ON, VCC33 = 3.63 V	T_J = 25°C		0.31	2	mA
			T_J = 85°C		0.35	2	mA
			T_J = 100°C		0.45	2	mA
		Standby mode[5], VCC33 = 3.63 V	T_J = 25°C		2.9	3.6	mA
			T_J = 85°C		2.9	4	mA
			T_J = 100°C		3.3	6	mA
		Sleep mode[6], VCC33 = 3.63 V	T_J = 25°C		17	19	µA
			T_J = 85°C		18	20	µA
			T_J = 100°C		24	25	µA
ICCI[3]	I/O quiescent current	Operational standby[4], Standby mode, and Sleep Mode[6], VCCIx = 3.63 V	T_J = 25°C		417	649	µA
			T_J = 85°C		417	649	µA
			T_J = 100°C		417	649	µA

Notes:

1. *ICC is the 1.5 V power supplies, ICC and ICC15A.*
2. *ICC33A includes ICC33A, ICC33PMP, and ICCOSC.*
3. *ICCI includes all ICCI0, ICCI1, ICCI2, and ICCI4.*
4. *Operational standby is when the Fusion device is powered up, all blocks are used, no I/O is toggling, Voltage Regulator is loaded with 200 mA, VCC33PMP is ON, XTAL is ON, and ADC is ON.*
5. *XTAL is configured as high gain, VCC = VJTAG = VPUMP = 0 V.*
6. *Sleep Mode, VCC = VJTAG = VPUMP = 0 V.*

FG484		
Pin Number	AFS600 Function	AFS1500 Function
E9	NC	IO08PDB0V1
E10	GND	GND
E11	IO15NDB1V0	IO22NDB1V0
E12	IO15PDB1V0	IO22PDB1V0
E13	GND	GND
E14	NC	IO32PPB1V1
E15	NC	IO36NPB1V2
E16	VCCIB1	VCCIB1
E17	GND	GND
E18	NC	IO47NPB2V0
E19	IO33PDB2V0	IO49PDB2V0
E20	VCCIB2	VCCIB2
E21	IO32NDB2V0	IO46NDB2V0
E22	GBC2/IO32PDB2V0	GBC2/IO46PDB2V0
F1	IO80NDB4V0	IO118NDB4V0
F2	IO80PDB4V0	IO118PDB4V0
F3	NC	IO119NSB4V0
F4	IO84NDB4V0	IO124NDB4V0
F5	GND	GND
F6	VCOMPLA	VCOMPLA
F7	VCCPLA	VCCPLA
F8	VCCIB0	VCCIB0
F9	IO08NDB0V1	IO12NDB0V1
F10	IO08PDB0V1	IO12PDB0V1
F11	VCCIB0	VCCIB0
F12	VCCIB1	VCCIB1
F13	IO22NDB1V0	IO30NDB1V1
F14	IO22PDB1V0	IO30PDB1V1
F15	VCCIB1	VCCIB1
F16	NC	IO36PPB1V2
F17	NC	IO38NPB1V2
F18	GND	GND
F19	IO33NDB2V0	IO49NDB2V0
F20	IO34PDB2V0	IO50PDB2V0
F21	IO34NDB2V0	IO50NDB2V0

FG484		
Pin Number	AFS600 Function	AFS1500 Function
F22	IO35PDB2V0	IO51PDB2V0
G1	IO77PDB4V0	IO115PDB4V0
G2	GND	GND
G3	IO78NDB4V0	IO116NDB4V0
G4	IO78PDB4V0	IO116PDB4V0
G5	VCCIB4	VCCIB4
G6	NC	IO117PDB4V0
G7	VCCIB4	VCCIB4
G8	GND	GND
G9	IO04NDB0V0	IO06NDB0V1
G10	IO04PDB0V0	IO06PDB0V1
G11	IO12NDB0V1	IO16NDB0V2
G12	IO12PDB0V1	IO16PDB0V2
G13	NC	IO28NDB1V1
G14	NC	IO28PDB1V1
G15	GND	GND
G16	NC	IO38PPB1V2
G17	NC	IO53PDB2V0
G18	VCCIB2	VCCIB2
G19	IO36PDB2V0	IO52PDB2V0
G20	IO36NDB2V0	IO52NDB2V0
G21	GND	GND
G22	IO35NDB2V0	IO51NDB2V0
H1	IO77NDB4V0	IO115NDB4V0
H2	IO76PDB4V0	IO113PDB4V0
H3	VCCIB4	VCCIB4
H4	IO79NDB4V0	IO114NDB4V0
H5	IO79PDB4V0	IO114PDB4V0
H6	NC	IO117NDB4V0
H7	GND	GND
H8	VCC	VCC
H9	VCCIB0	VCCIB0
H10	GND	GND
H11	VCCIB0	VCCIB0
H12	VCCIB1	VCCIB1

FG484		
Pin Number	AFS600 Function	AFS1500 Function
H13	GND	GND
H14	VCCIB1	VCCIB1
H15	GND	GND
H16	GND	GND
H17	NC	IO53NDB2V0
H18	IO38PDB2V0	IO57PDB2V0
H19	GCA2/IO39PDB2V0	GCA2/IO59PDB2V0
H20	VCCIB2	VCCIB2
H21	IO37NDB2V0	IO54NDB2V0
H22	IO37PDB2V0	IO54PDB2V0
J1	NC	IO112PPB4V0
J2	IO76NDB4V0	IO113NDB4V0
J3	GFB2/IO74PDB4V0	GFB2/IO109PDB4V0
J4	GFA2/IO75PDB4V0	GFA2/IO110PDB4V0
J5	NC	IO112NPB4V0
J6	NC	IO104PDB4V0
J7	NC	IO111PDB4V0
J8	VCCIB4	VCCIB4
J9	GND	GND
J10	VCC	VCC
J11	GND	GND
J12	VCC	VCC
J13	GND	GND
J14	VCC	VCC
J15	VCCIB2	VCCIB2
J16	GCB2/IO40PDB2V0	GCB2/IO60PDB2V0
J17	NC	IO58NDB2V0
J18	IO38NDB2V0	IO57NDB2V0
J19	IO39NDB2V0	IO59NDB2V0
J20	GCC2/IO41PDB2V0	GCC2/IO61PDB2V0
J21	NC	IO55PSB2V0
J22	IO42PDB2V0	IO56PDB2V0
K1	GFC2/IO73PDB4V0	GFC2/IO108PDB4V0
K2	GND	GND
K3	IO74NDB4V0	IO109NDB4V0

FG484		
Pin Number	AFS600 Function	AFS1500 Function
K4	IO75NDB4V0	IO110NDB4V0
K5	GND	GND
K6	NC	IO104NDB4V0
K7	NC	IO111NDB4V0
K8	GND	GND
K9	VCC	VCC
K10	GND	GND
K11	VCC	VCC
K12	GND	GND
K13	VCC	VCC
K14	GND	GND
K15	GND	GND
K16	IO40NDB2V0	IO60NDB2V0
K17	NC	IO58PDB2V0
K18	GND	GND
K19	NC	IO68NPB2V0
K20	IO41NDB2V0	IO61NDB2V0
K21	GND	GND
K22	IO42NDB2V0	IO56NDB2V0
L1	IO73NDB4V0	IO108NDB4V0
L2	VCCOSC	VCCOSC
L3	VCCIB4	VCCIB4
L4	XTAL2	XTAL2
L5	GFC1/IO72PDB4V0	GFC1/IO107PDB4V0
L6	VCCIB4	VCCIB4
L7	GFB1/IO71PDB4V0	GFB1/IO106PDB4V0
L8	VCCIB4	VCCIB4
L9	GND	GND
L10	VCC	VCC
L11	GND	GND
L12	VCC	VCC
L13	GND	GND
L14	VCC	VCC
L15	VCCIB2	VCCIB2
L16	IO48PDB2V0	IO70PDB2V0

FG484		
Pin Number	AFS600 Function	AFS1500 Function
L17	VCCIB2	VCCIB2
L18	IO46PDB2V0	IO69PDB2V0
L19	GCA1/IO45PDB2V0	GCA1/IO64PDB2V0
L20	VCCIB2	VCCIB2
L21	GCC0/IO43NDB2V0	GCC0/IO62NDB2V0
L22	GCC1/IO43PDB2V0	GCC1/IO62PDB2V0
M1	NC	IO103PDB4V0
M2	XTAL1	XTAL1
M3	VCCIB4	VCCIB4
M4	GNDOSC	GNDOSC
M5	GFC0/IO72NDB4V0	GFC0/IO107NDB4V0
M6	VCCIB4	VCCIB4
M7	GFB0/IO71NDB4V0	GFB0/IO106NDB4V0
M8	VCCIB4	VCCIB4
M9	VCC	VCC
M10	GND	GND
M11	VCC	VCC
M12	GND	GND
M13	VCC	VCC
M14	GND	GND
M15	VCCIB2	VCCIB2
M16	IO48NDB2V0	IO70NDB2V0
M17	VCCIB2	VCCIB2
M18	IO46NDB2V0	IO69NDB2V0
M19	GCA0/IO45NDB2V0	GCA0/IO64NDB2V0
M20	VCCIB2	VCCIB2
M21	GCB0/IO44NDB2V0	GCB0/IO63NDB2V0
M22	GCB1/IO44PDB2V0	GCB1/IO63PDB2V0
N1	NC	IO103NDB4V0
N2	GND	GND
N3	IO68PDB4V0	IO101PDB4V0
N4	NC	IO100NPB4V0
N5	GND	GND
N6	NC	IO99PDB4V0
N7	NC	IO97PDB4V0

FG484		
Pin Number	AFS600 Function	AFS1500 Function
N8	GND	GND
N9	GND	GND
N10	VCC	VCC
N11	GND	GND
N12	VCC	VCC
N13	GND	GND
N14	VCC	VCC
N15	GND	GND
N16	GDB2/IO56PDB2V0	GDB2/IO83PDB2V0
N17	NC	IO78PDB2V0
N18	GND	GND
N19	IO47NDB2V0	IO72NDB2V0
N20	IO47PDB2V0	IO72PDB2V0
N21	GND	GND
N22	IO49PDB2V0	IO71PDB2V0
P1	GFA1/IO70PDB4V0	GFA1/IO105PDB4V0
P2	GFA0/IO70NDB4V0	GFA0/IO105NDB4V0
P3	IO68NDB4V0	IO101NDB4V0
P4	IO65PDB4V0	IO96PDB4V0
P5	IO65NDB4V0	IO96NDB4V0
P6	NC	IO99NDB4V0
P7	NC	IO97NDB4V0
P8	VCCIB4	VCCIB4
P9	VCC	VCC
P10	GND	GND
P11	VCC	VCC
P12	GND	GND
P13	VCC	VCC
P14	GND	GND
P15	VCCIB2	VCCIB2
P16	IO56NDB2V0	IO83NDB2V0
P17	NC	IO78NDB2V0
P18	GDA1/IO54PDB2V0	GDA1/IO81PDB2V0
P19	GDB1/IO53PDB2V0	GDB1/IO80PDB2V0
P20	IO51NDB2V0	IO73NDB2V0

FG676	
Pin Number	AFS1500 Function
R21	IO72NDB2V0
R22	IO72PDB2V0
R23	GND
R24	IO71PDB2V0
R25	VCCIB2
R26	IO67NDB2V0
T1	GND
T2	NC
T3	GFA1/IO105PDB4V0
T4	GFA0/IO105NDB4V0
T5	IO101NDB4V0
T6	IO96PDB4V0
T7	IO96NDB4V0
T8	IO99NDB4V0
T9	IO97NDB4V0
T10	VCCIB4
T11	VCC
T12	GND
T13	VCC
T14	GND
T15	VCC
T16	GND
T17	VCCIB2
T18	IO83NDB2V0
T19	IO78NDB2V0
T20	GDA1/IO81PDB2V0
T21	GDB1/IO80PDB2V0
T22	IO73NDB2V0
T23	IO73PDB2V0
T24	IO71NDB2V0
T25	NC
T26	GND
U1	NC
U2	NC
U3	IO102PDB4V0
U4	IO102NDB4V0

FG676	
Pin Number	AFS1500 Function
U5	VCCIB4
U6	IO91PDB4V0
U7	IO91NDB4V0
U8	IO92PDB4V0
U9	GND
U10	GND
U11	VCC33A
U12	GNDA
U13	VCC33A
U14	GNDA
U15	VCC33A
U16	GNDA
U17	VCC
U18	GND
U19	IO74NDB2V0
U20	GDA0/IO81NDB2V0
U21	GDB0/IO80NDB2V0
U22	VCCIB2
U23	IO75NDB2V0
U24	IO75PDB2V0
U25	NC
U26	NC
V1	NC
V2	VCCIB4
V3	IO100PPB4V0
V4	GND
V5	IO95PDB4V0
V6	IO95NDB4V0
V7	VCCIB4
V8	IO92NDB4V0
V9	GNDNVM
V10	GNDA
V11	NC
V12	AV4
V13	NC
V14	AV5

FG676	
Pin Number	AFS1500 Function
V15	AC5
V16	NC
V17	GNDA
V18	IO77PPB2V0
V19	IO74PDB2V0
V20	VCCIB2
V21	IO82NDB2V0
V22	GDA2/IO82PDB2V0
V23	GND
V24	GDC1/IO79PDB2V0
V25	VCCIB2
V26	NC
W1	GND
W2	IO94PPB4V0
W3	IO98PDB4V0
W4	IO98NDB4V0
W5	GEC1/IO90PDB4V0
W6	GEC0/IO90NDB4V0
W7	GND
W8	VCCNVM
W9	VCCIB4
W10	VCC15A
W11	GNDA
W12	AC4
W13	VCC33A
W14	GNDA
W15	AG5
W16	GNDA
W17	PUB
W18	VCCIB2
W19	TDI
W20	GND
W21	IO84NDB2V0
W22	GDC2/IO84PDB2V0
W23	IO77NPB2V0
W24	GDC0/IO79NDB2V0

Revision	Changes	Page
Revision 2 (continued)	The prescalar range for the 'Analog Input (direct input to ADC)" configurations was removed as inapplicable for direct inputs. The input resistance for direct inputs is covered in Table 2-50 • ADC Characteristics in Direct Input Mode (SAR 31201).	2-120
	The "Examples" for calibrating accuracy for ADC channels were revised and corrected to make them consistent with terminology in the associated tables (SARs 36791, 36773).	2-124
	A note was added to Table 2-56 • Analog Quad ACM Byte Assignment and the introductory text for Table 2-66 • Internal Temperature Monitor Control Truth Table, stating that for the internal temperature monitor to function, Bit 0 of Byte 2 for all 10 Quads must be set (SAR 34418).	2-129, 2-131
	t_{DOUT} was corrected to t_{DIN} in Figure 2-116 • Input Buffer Timing Model and Delays (example) (SAR 37115).	2-161
	The formulas in the table notes for Table 2-97 • I/O Weak Pull-Up/Pull-Down Resistances were corrected (SAR 34751).	2-171
	The AC Loading figures in the "Single-Ended I/O Characteristics" section were updated to match tables in the "Summary of I/O Timing Characteristics – Default I/O Software Settings" section (SAR 34877).	2-175
	The following notes were removed from Table 2-168 • Minimum and Maximum DC Input and Output Levels (SAR 34808): ±5% Differential input voltage = ±350 mV	2-209
	An incomplete, duplicate sentence was removed from the end of the "GNDAQ Ground (analog quiet)" pin description (SAR 30185).	2-223
	Information about configuration of unused I/Os was added to the "User Pins" section (SAR 32642).	2-225
	The following information was added to the pin description for "XTAL1 Crystal Oscillator Circuit Input" and "XTAL2 Crystal Oscillator Circuit Input" (SAR 24119).	2-227
	The input resistance to ground value in Table 3-3 • Input Resistance of Analog Pads for Analog Input (direct input to ADC), was corrected from 1 MΩ (typical) to 2 kΩ (typical) (SAR 34371).	3-4
	The Storage Temperature column in Table 3-5 • FPGA Programming, Storage, and Operating Limits stated Min. T_J twice for commercial and industrial product grades and has been corrected to Min. T_J and Max. T_J (SAR 29416).	3-5
	The reference to guidelines for global spines and VersaTile rows, given in the "Global Clock Dynamic Contribution—PCLOCK" section, was corrected to the "Spine Architecture" section of the Global Resources chapter in the *Fusion FPGA Fabric User's Guide* (SAR 34741).	3-24
	Package names used in the "Package Pin Assignments" section were revised to match standards given in *Package Mechanical Drawings* (SAR 36612).	4-1
July 2010	The versioning system for datasheets has been changed. Datasheets are assigned a revision number that increments each time the datasheet is revised. The "Fusion Device Status" table indicates the status for each device in the device family.	N/A

Revision	Changes	Page
Advance v0.6 (continued)	The "Analog-to-Digital Converter Block" section was updated with the following statement: "All results are MSB justified in the ADC."	2-99
	The information about the ADCSTART signal was updated in the "ADC Description" section.	2-102
	Table 2-46 · Analog Channel Specifications was updated.	2-118
	Table 2-47 · ADC Characteristics in Direct Input Mode was updated.	2-121
	Table 2-51 • ACM Address Decode Table for Analog Quad was updated.	2-127
	In Table 2-53 • Analog Quad ACM Byte Assignment, the Function and Default Setting for Bit 6 in Byte 3 was updated.	2-130
	The "Introduction" section was updated to include information about digital inputs, outputs, and bibufs.	2-133
	In Table 2-69 • Fusion Pro I/O Features, the programmable delay descriptions were updated for the following features: Single-ended receiver Voltage-referenced differential receiver LVDS/LVPECL differential receiver features	2-137
	The "User I/O Naming Convention" section was updated to include "V" and "z" descriptions	2-159
	The "VCC33PMP Analog Power Supply (3.3 V)" section was updated to include information about avoiding high current draw.	2-224
	The "VCCNVM Flash Memory Block Power Supply (1.5 V)" section was updated to include information about avoiding high current draw.	2-224
	The "VMVx I/O Supply Voltage (quiet)" section was updated to include this statement: VMV and VCCI must be connected to the same power supply and V_{CCI} pins within a given I/O bank.	2-185
	The "PUB Push Button" section was updated to include information about leaving the pin floating if it is not used.	2-228
	The "PTBASE Pass Transistor Base" section was updated to include information about leaving the pin floating if it is not used.	2-228
	The "PTEM Pass Transistor Emitter" section was updated to include information about leaving the pin floating if it is not used.	2-228
	The heading was incorrect in the "208-Pin PQFP" table. It should be AFS250 and not AFS090.	3-8

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