Welcome to E-XFL.COM

Understanding Embedded - FPGAs (Field Programmable Gate Array)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details	
Product Status	Active
Number of LABs/CLBs	-
Number of Logic Elements/Cells	-
Total RAM Bits	276480
Number of I/O	119
Number of Gates	1500000
Voltage - Supply	1.425V ~ 1.575V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 85°C (TJ)
Package / Case	256-LBGA
Supplier Device Package	256-FPBGA (17x17)
Purchase URL	https://www.e-xfl.com/product-detail/microchip-technology/p1afs1500-2fgg256

Global Buffers with No Programmable Delays

The CLKBUF and CLKBUF_LVPECL/LVDS macros are composite macros that include an I/O macro driving a global buffer, hardwired together (Figure 2-20).

The CLKINT macro provides a global buffer function driven by the FPGA core.

The CLKBUF, CLKBUF_LVPECL/LVDS, and CLKINT macros are pass-through clock sources and do not use the PLL or provide any programmable delay functionality.

Many specific CLKBUF macros support the wide variety of single-ended and differential I/O standards supported by Fusion devices. The available CLKBUF macros are described in the *IGLOO, ProASIC3, SmartFusion and Fusion Macro Library Guide*.

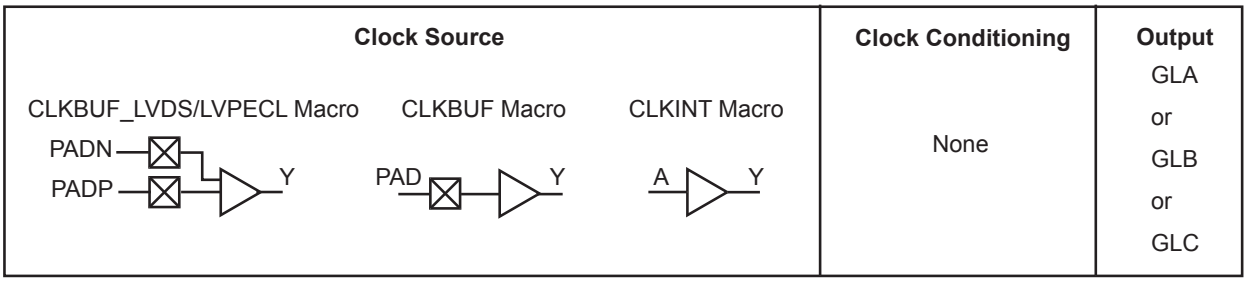


Figure 2-20 • **Global Buffers with No Programmable Delay**

Global Buffers with Programmable Delay

The CLKDLY macro is a pass-through clock source that does not use the PLL, but provides the ability to delay the clock input using a programmable delay (Figure 2-21 on page 2-25). The CLKDLY macro takes the selected clock input and adds a user-defined delay element. This macro generates an output clock phase shift from the input clock.

The CLKDLY macro can be driven by an INBUF macro to create a composite macro, where the I/O macro drives the global buffer (with programmable delay) using a hardwired connection. In this case, the I/O must be placed in one of the dedicated global I/O locations.

Many specific INBUF macros support the wide variety of single-ended and differential I/O standards supported by the Fusion family. The available INBUF macros are described in the *IGLOO, ProASIC3, SmartFusion and Fusion Macro Library Guide.*

The CLKDLY macro can be driven directly from the FPGA core.

The CLKDLY macro can also be driven from an I/O that is routed through the FPGA regular routing fabric. In this case, users must instantiate a special macro, PLLINT, to differentiate from the hardwired I/O connection described earlier.

The visual CLKDLY configuration in the SmartGen part of the Libero SoC and Designer tools allows the user to select the desired amount of delay and configures the delay elements appropriately. SmartGen also allows the user to select the input clock source. SmartGen will automatically instantiate the special macro, PLLINT, when needed.

Program Operation

A Program operation is initiated by asserting the PROGRAM signal on the interface. Program operations save the contents of the Page Buffer to the FB Array. Due to the technologies inherent in the FB, the total programming (including erase) time per page of the eNVM is 6.8 ms. While the FB is writing the data to the array, the BUSY signal will be asserted.

During a Program operation, the sector and page addresses on ADDR are compared with the stored address for the page (and sector) in the Page Buffer. If there is a mismatch between the two addresses, the Program operation will be aborted and an error will be reported on the STATUS output.

It is possible to write the Page Buffer to a different page in memory. When asserting the PROGRAM pin, if OVERWRITEPAGE is asserted as well, the FB will write the contents of the Page Buffer to the sector and page designated on the ADDR inputs if the destination page is not Overwrite Protected.

A Program operation can be utilized to either modify the contents of the page in the flash memory block or change the protections for the page. Setting the OVERWRITEPROTECT bit on the interface while asserting the PROGRAM pin will put the page addressed into Overwrite Protect Mode. Overwrite Protect Mode safeguards a page from being inadvertently overwritten during subsequent Program or Erase operations.

Program operations that result in a STATUS value of '01' do not modify the addressed page. For all other values of STATUS, the addressed page is modified. Program errors include the following:

1. Attempting to program a page that is Overwrite Protected (STATUS = '01')
2. Attempting to program a page that is not in the Page Buffer when the Page Buffer has entered Page Loss Protection Mode (STATUS = '01')
3. Attempting to perform a program with OVERWRITEPAGE set when the page addressed has been Overwrite Protected (STATUS = '01')
4. The Write Count of the page programmed exceeding the Write Threshold defined in the part specification (STATUS = '11')
5. The ECC Logic determining that there is an uncorrectable error within the programmed page (STATUS = '10')
6. Attempting to program a page that is **not** in the Page Buffer when OVERWRITEPAGE is not set and the page in the Page Buffer is modified (STATUS = '01')
7. Attempting to program the page in the Page Buffer when the Page Buffer is **not** modified

The waveform for a Program operation is shown in Figure 2-36.

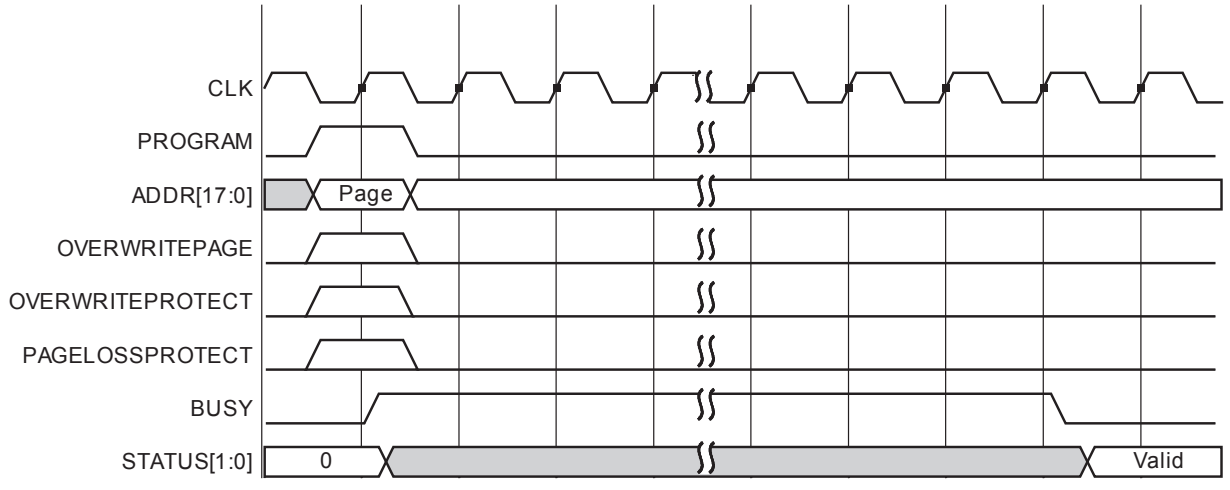


Figure 2-36 • **FB Program Waveform**

Note: OVERWRITEPAGE is only sampled when the PROGRAM or ERASEPAGE pins are asserted. OVERWRITEPAGE is ignored in all other operations.

The AEMPTY flag is asserted when the difference between the write address and the read address is less than a predefined value. In the example above, a value of 200 for AEVAL means that the AEMPTY flag will be asserted when a read causes the difference between the write address and the read address to drop to 200. It will stay asserted until that difference rises above 200. Note that the FIFO can be configured with different read and write widths; in this case, the AFVAL setting is based on the number of write data entries and the AEVAL setting is based on the number of read data entries. For aspect ratios of 512×9 and 256×18, only 4,096 bits can be addressed by the 12 bits of AFVAL and AEVAL. The number of words must be multiplied by 8 and 16, instead of 9 and 18. The SmartGen tool automatically uses the proper values. To avoid halfwords being written or read, which could happen if different read and write aspect ratios are specified, the FIFO will assert FULL or EMPTY as soon as at least a minimum of one word cannot be written or read. For example, if a two-bit word is written and a four-bit word is being read, the FIFO will remain in the empty state when the first word is written. This occurs even if the FIFO is not completely empty, because in this case, a complete word cannot be read. The same is applicable in the full state. If a four-bit word is written and a two-bit word is read, the FIFO is full and one word is read. The FULL flag will remain asserted because a complete word cannot be written at this point.

Channel Input Offset Error

Channel Offset error is measured as the input voltage that causes the transition from zero to a count of one. An Ideal Prescaler will have offset equal to ½ of LSB voltage. Offset error is a positive or negative when the first transition point is higher or lower than ideal. Offset error is expressed in LSB or input voltage.

Total Channel Error

Total Channel Error is defined as the total error measured compared to the ideal value. Total Channel Error is the sum of gain error and offset error combined. Figure 2-68 shows how Total Channel Error is measured.

Total Channel Error is defined as the difference between the actual ADC output and ideal ADC output. In the example shown in Figure 2-68, the Total Channel Error would be a negative number.

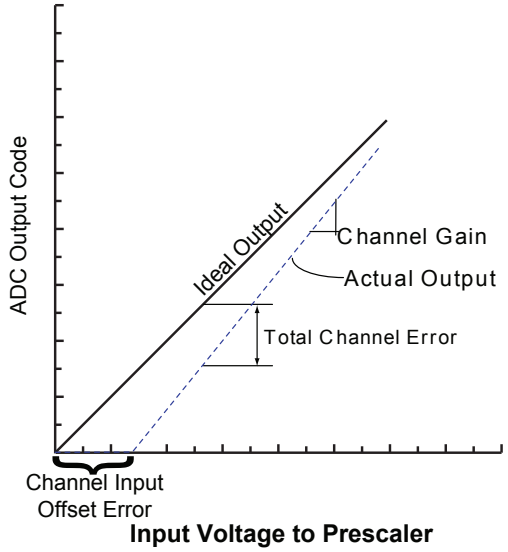


Figure 2-68 • **Total Channel Error Example**

Terminology

Resolution

Resolution defines the smallest temperature change Fusion Temperature Monitor can resolve. For ADC configured as 8-bit mode, each LSB represents 4°C, and 1°C per LSB for 10-bit mode. With 12-bit mode, the Temperature Monitor can still only resolve 1°C due to Temperature Monitor design.

Offset

The Fusion Temperature Monitor has a systematic offset (Table 2-49 on page 2-117), excluding error due to board resistance and ideality factor of the external diode. Microsemi provides an IP block (CalibIP) that is required in order to mitigate the systematic temperature offset. For further details on CalibIP, refer to the "*Temperature*, *Voltage, and Current Calibration in Fusion FPGAs" chapter of the Fusion FPGA Fabric User Guide*.

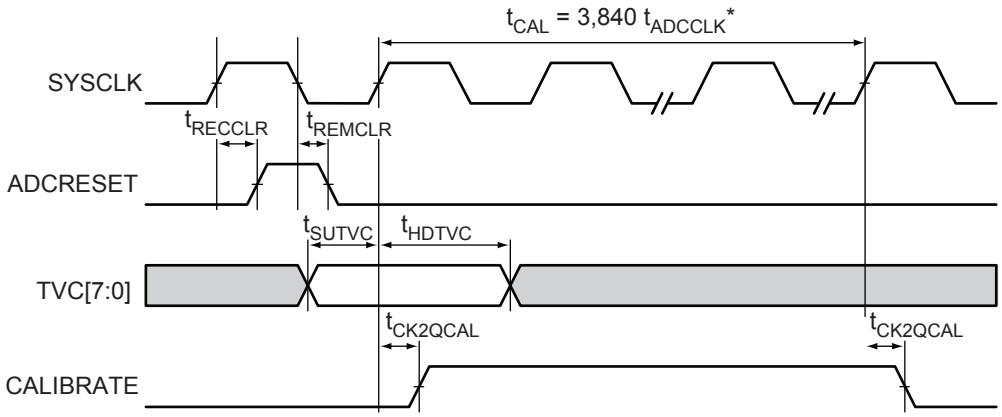
The optimal setting for the system running at 66 MHz with an ADC for 10-bit mode chosen is shown in Table 2-47:

Table 2-47 • **Optimal Setting at 66 MHz in 10-Bit Mode**

TVC[7:0]	= 1	= 0x01
STC[7:0]	= 3	= 0x03
MODE[3:0]	= b'0100	= 0x4*

Note: No power-down after every conversion is chosen in this case; however, if the application is power-sensitive, the MODE[2] can be set to '0', as described above, and it will not affect any performance.

Timing Diagrams



Note: **Refer to EQ 15 on page 2-107 for the calculation on the period of ADCCLK, t_{ADCCLK}.*

Figure 2-89 • **Power-Up Calibration Status Signal Timing Diagram**

Analog Quad ACM Description

Table 2-56 maps out the ACM space associated with configuration of the Analog Quads within the Analog Block. Table 2-56 shows the byte assignment within each quad and the function of each bit within each byte. Subsequent tables will explain each bit setting and how it corresponds to a particular configuration. After 3.3 V and 1.5 V are applied to Fusion, Analog Quad configuration registers are loaded with default settings until the initialization and configuration state machine changes them to user-defined settings.

***Table 2-56* • Analog Quad ACM Byte Assignment**

Byte	Bit	Signal (Bx)	Function	Default Setting
Byte 0 (AV)	0	B0[0]	Scaling factor control – prescaler	Highest voltage range
	1	B0[1]		
	2	B0[2]		
	3	B0[3]	Analog MUX select	Prescaler
	4	B0[4]	Current monitor switch	Off
	5	B0[5]	Direct analog input switch	Off
	6	B0[6]	Selects V-pad polarity	Positive
	7	B0[7]	Prescaler op amp mode	Power-down
Byte 1 (AC)	0	B1[0]	Scaling factor control – prescaler	Highest voltage range
	1	B1[1]		
	2	B1[2]		
	3	B1[3]	Analog MUX select	Prescaler
	4	B1[4]		
	5	B1[5]	Direct analog input switch	Off
	6	B1[6]	Selects C-pad polarity	Positive
	7	B1[7]	Prescaler op amp mode	Power-down
Byte 2 (AG)	0	B2[0]	Internal chip temperature monitor *	Off
	1	B2[1]	Spare	–
	2	B2[2]	Current drive control	Lowest current
	3	B2[3]		
	4	B2[4]	Spare	–
	5	B2[5]	Spare	–
	6	B2[6]	Selects G-pad polarity	Positive
	7	B2[7]	Selects low/high drive	Low drive
Byte 3 (AT)	0	B3[0]	Scaling factor control – prescaler	Highest voltage range
	1	B3[1]		
	2	B3[2]		
	3	B3[3]	Analog MUX select	Prescaler
	4	B3[4]		
	5	B3[5]	Direct analog input switch	Off
	6	B3[6]	–	–
	7	B3[7]	Prescaler op amp mode	Power-down

*Note: *For the internal temperature monitor to function, Bit 0 of Byte 2 for all 10 Quads must be set.*

User I/Os

Introduction

Fusion devices feature a flexible I/O structure, supporting a range of mixed voltages (1.5 V, 1.8 V, 2.5 V, and 3.3 V) through a bank-selectable voltage. Table 2-68, Table 2-69, Table 2-70, and Table 2-71 on page 2-135 show the voltages and the compatible I/O standards. I/Os provide programmable slew rates, drive strengths, weak pull-up, and weak pull-down circuits. 3.3 V PCI and 3.3 V PCI-X are 5 V–tolerant. See the "5 V Input Tolerance" section on page 2-144 for possible implementations of 5 V tolerance.

All I/Os are in a known state during power-up, and any power-up sequence is allowed without current impact. Refer to the "I/O Power-Up and Supply Voltage Thresholds for Power-On Reset (Commercial and Industrial)" section on page 3-5 for more information. In low power standby or sleep mode (VCC is OFF, VCC33A is ON, VCCI is ON) or when the resource is not used, digital inputs are tristated, digital outputs are tristated, and digital bibufs (input/output) are tristated.

I/O Tile

The Fusion I/O tile provides a flexible, programmable structure for implementing a large number of I/O standards. In addition, the registers available in the I/O tile in selected I/O banks can be used to support high-performance register inputs and outputs, with register enable if desired (Figure 2-99 on page 2-133). The registers can also be used to support the JESD-79C DDR standard within the I/O structure (see the "Double Data Rate (DDR) Support" section on page 2-139 for more information).

As depicted in Figure 2-100 on page 2-138, all I/O registers share one CLR port. The output register and output enable register share one CLK port. Refer to the "I/O Registers" section on page 2-138 for more information.

I/O Banks and I/O Standards Compatibility

The digital I/Os are grouped into I/O voltage banks. There are three digital I/O banks on the AFS090 and AFS250 devices and four digital I/O banks on the AFS600 and AFS1500 devices. Figure 2-113 on page 2-158 and Figure 2-114 on page 2-159 show the bank configuration by device. The north side of the I/O in the AFS600 and AFS1500 devices comprises two banks of Pro I/Os. The Pro I/Os support a wide number of voltage-referenced I/O standards in addition to the multitude of single-ended and differential I/O standards common throughout all Microsemi digital I/Os. Each I/O voltage bank has dedicated I/O supply and ground voltages (VCCI/GNDQ for input buffers and VCCI/GND for output buffers). Because of these dedicated supplies, only I/Os with compatible standards can be assigned to the same I/O voltage bank. Table 2-69 and Table 2-70 on page 2-134 show the required voltage compatibility values for each of these voltages.

For more information about I/O and global assignments to I/O banks, refer to the specific pin table of the device in the "Package Pin Assignments" on page 4-1 and the "User I/O Naming Convention" section on page 2-158.

Each Pro I/O bank is divided into minibanks. Any user I/O in a VREF minibank (a minibank is the region of scope of a VREF pin) can be configured as a VREF pin (Figure 2-99 on page 2-133). Only one VREF pin is needed to control the entire VREF minibank. The location and scope of the VREF minibanks can be determined by the I/O name. For details, see the "User I/O Naming Convention" section on page 2-158.

Table 2-70 on page 2-134 shows the I/O standards supported by Fusion devices and the corresponding voltage levels.

I/O standards are compatible if the following are true:

- Their VCCI values are identical.
- If both of the standards need a VREF, their VREF values must be identical (Pro I/O only).

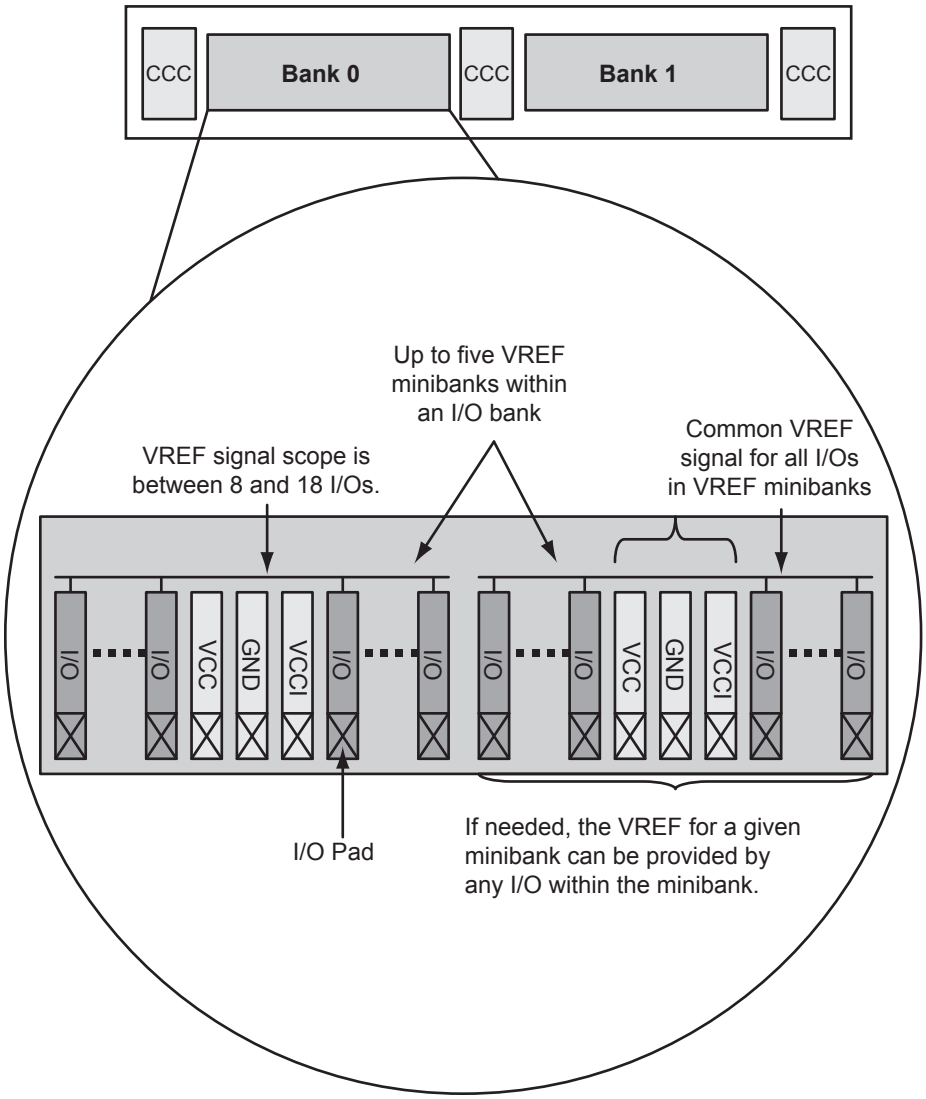


Figure 2-99 • **Fusion Pro I/O Bank Detail Showing VREF Minibanks (north side ofAFS600 and AFS1500)**

Table 2-67 • **I/O Standards Supported by Bank Type**

I/O Bank	Single-Ended I/O Standards	Differential I/O Standards	Voltage-Referenced	Hot-Swap
Standard I/O	LVTTL/LVCMOS 3.3 V, LVCMOS 2.5 V / 1.8 V / 1.5 V, LVCMOS 2.5/5.0 V	–	–	Yes
Advanced I/O	LVTTL/LVCMOS 3.3 V, LVCMOS 2.5 V / 1.8 V / 1.5 V, LVCMOS 2.5/5.0 V, 3.3 V PCI / 3.3 V PCI-X	LVPECL and LVDS	–	–
Pro I/O	LVTTL/LVCMOS 3.3 V, LVCMOS 2.5 V / 1.8 V / 1.5 V, LVCMOS 2.5/5.0 V, 3.3 V PCI / 3.3 V PCI-X	LVPECL and LVDS	GTL+ 2.5 V / 3.3 V, GTL 2.5 V / 3.3 V, HSTL Class I and II, SSTL2 Class I and II, SSTL3 Class I and II	Yes

At the system level, the skew circuit can be used in applications where transmission activities on bidirectional data lines need to be coordinated. This circuit, when selected, provides a timing margin that can prevent bus contention and subsequent data loss or transmitter overstress due to transmitter-to-transmitter current shorts. Figure 2-110 presents an example of the skew circuit implementation in a bidirectional communication system. Figure 2-111 shows how bus contention is created, and Figure 2-112 on page 2-151 shows how it can be avoided with the skew circuit.

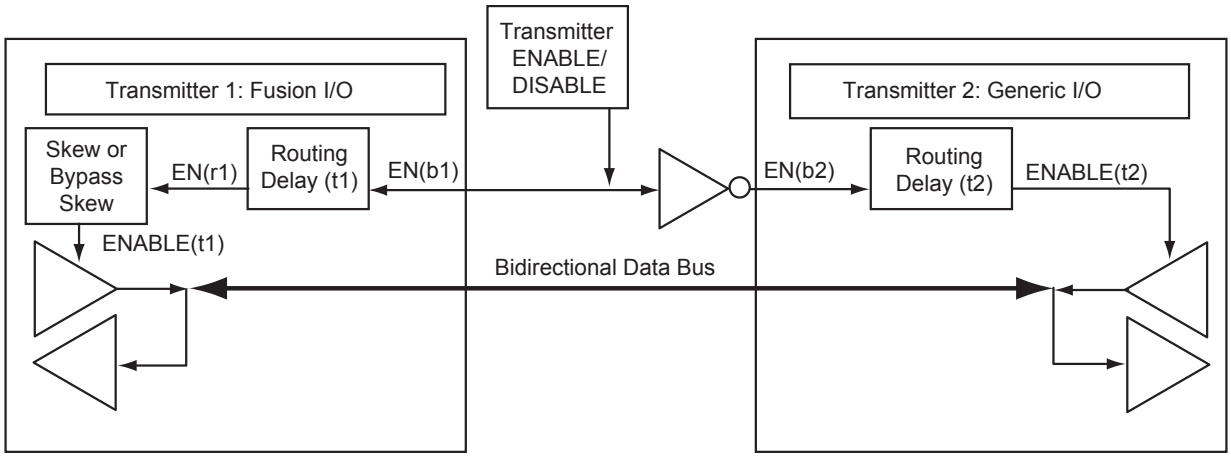


Figure 2-110 • **Example of Implementation of Skew Circuits in Bidirectional Transmission Systems Using Fusion Devices**

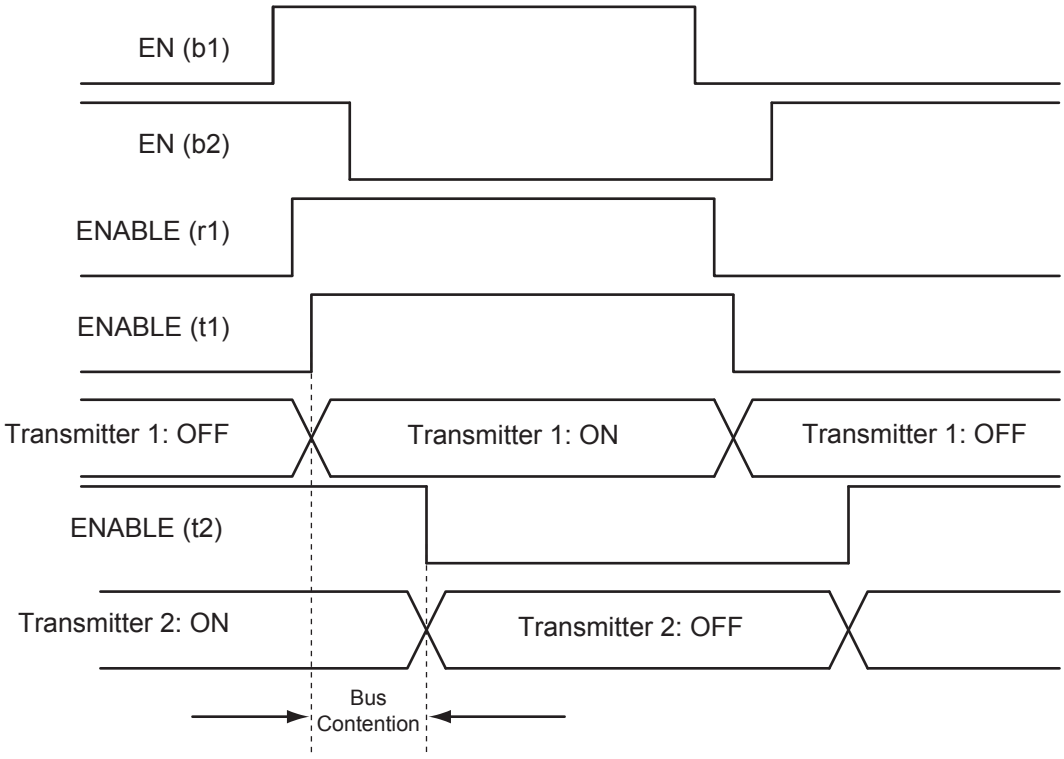


Figure 2-111 • **Timing Diagram (bypasses skew circuit)**

Table 2-93 • Summary of I/O Timing Characteristics – Software Default Settings
Commercial Temperature Range Conditions: T_J = 70°C, Worst-Case VCC = 1.425 V,
Worst-Case VCCI = I/O Standard Dependent
Applicable to Advanced I/Os

I/O Standard	Drive Strength (mA)	Slew Rate	Capacitive Load (pF)	External Resistor (Ohm)	t_{DOUT}	t_{DP}	t_{DIN}	t_{PY}	t_{EOUT}	t_{ZL}	t_{ZH}	t_{LZ}	t_{HZ}	t_{ZLS}	t_{ZHS}	Units
3.3 V LVTTL/ 3.3 V LVCMOS	12 mA	High	35 pF	–	0.49	2.64	0.03	0.90	0.32	2.69	2.11	2.40	2.68	4.36	3.78	ns
2.5 V LVCMOS	12 mA	High	35 pF	–	0.49	2.66	0.03	0.98	0.32	2.71	2.56	2.47	2.57	4.38	4.23	ns
1.8 V LVCMOS	12 mA	High	35 pF	–	0.49	2.64	0.03	0.91	0.32	2.69	2.27	2.76	3.05	4.36	3.94	ns
1.5 V LVCMOS	12 mA	High	35 pF	–	0.49	3.05	0.03	1.07	0.32	3.10	2.67	2.95	3.14	4.77	4.34	ns
3.3 V PCI	Per PCI spec	High	10 pF	25 [2]	0.49	2.00	0.03	0.65	0.32	2.04	1.46	2.40	2.68	3.71	3.13	ns
3.3 V PCI-X	Per PCI-X spec	High	10 pF	25 [2]	0.49	2.00	0.03	0.62	0.32	2.04	1.46	2.40	2.68	3.71	3.13	ns
LVDS	24 mA	High	–	–	0.49	1.37	0.03	1.20	N/A	N/A	N/A	N/A	N/A	N/A	N/A	ns
LVPECL	24 mA	High	–	–	0.49	1.34	0.03	1.05	N/A	N/A	N/A	N/A	N/A	N/A	N/A	ns

Notes:

1. *For specific junction temperature and voltage-supply levels, refer to Table 3-6 on page 3-7 for derating values.*
2. *Resistance is used to measure I/O propagation delays as defined in PCI specifications. See Figure 2-123 on page 2-197 for connectivity. This resistor is not required during normal operation.*

Table 2-94 • Summary of I/O Timing Characteristics – Software Default Settings
Commercial Temperature Range Conditions: T_J = 70°C, Worst-Case VCC = 1.425 V,
Worst-Case VCCI = I/O Standard Dependent
Applicable to Standard I/Os

I/O Standard	Drive Strength (mA)	Slew Rate	Capacitive Load (pF)	External Resistor (Ohm)	t_{DOUT}	t_{DP}	t_{DIN}	t_{PY}	t_{EOUT}	t_{ZL}	t_{ZH}	t_{LZ}	t_{HZ}	Units
3.3 V LVTTL/ 3.3 V LVCMOS	8 mA	High	35 pF	–	0.49	3.29	0.03	0.75	0.32	3.36	2.80	1.79	2.01	ns
2.5 V LVCMOS	8 mA	High	35pF	–	0.49	3.56	0.03	0.96	0.32	3.40	3.56	1.78	1.91	ns
1.8 V LVCMOS	4 mA	High	35pF	–	0.49	4.74	0.03	0.90	0.32	4.02	4.74	1.80	1.85	ns
1.5 V LVCMOS	2 mA	High	35pF	–	0.49	5.71	0.03	1.06	0.32	4.71	5.71	1.83	1.83	ns

Note: For specific junction temperature and voltage-supply levels, refer to Table 3-6 on page 3-7 for derating values.

***Table 2-98* • I/O Short Currents IOSH/IOSL (continued)**

	Drive Strength	IOSH (mA)*	IOSL (mA)*
2.5 V LVCMOS	2 mA	16	18
	4 mA	16	18
	6 mA	32	37
	8 mA	32	37
	12 mA	65	74
	16 mA	83	87
	24 mA	169	124
1.8 V LVCMOS	2 mA	9	11
	4 mA	17	22
	6 mA	35	44
	8 mA	45	51
	12 mA	91	74
	16 mA	91	74
1.5 V LVCMOS	2 mA	13	16
	4 mA	25	33
	6 mA	32	39
	8 mA	66	55
	12 mA	66	55
3.3 V PCI/PCI-X	Per PCI/PCI-X specification	103	109
Applicable to Standard I/O Banks			
3.3 V LVTTL / 3.3 V LVCMOS	2 mA	25	27
	4 mA	25	27
	6 mA	51	54
	8 mA	51	54
2.5 V LVCMOS	2 mA	16	18
	4 mA	16	18
	6 mA	32	37
	8 mA	32	37
1.8 V LVCMOS	2 mA	9	11
	4 mA	17	22
1.5 V LVCMOS	2 mA	13	16

*Note: *T_J = 100°C*

The length of time an I/O can withstand IOSH/IOSL events depends on the junction temperature. The reliability data below is based on a 3.3 V, 36 mA I/O setting, which is the worst case for this type of analysis.

For example, at 100°C, the short current condition would have to be sustained for more than six months to cause a reliability concern. The I/O design does not contain any short circuit protection, but such protection would only be needed in extremely prolonged stress conditions.

Table 2-113 • 2.5 V LVCMOS High Slew
Commercial Temperature Range Conditions: T_J = 70°C, Worst-Case VCC = 1.425 V, Worst-Case VCCI = 2.3 V
Applicable to Pro I/Os

Drive Strength	Speed Grade	t_{DOUT}	t_{DP}	t_{DIN}	t_{PY}	t_{PYS}	t_{EOUT}	t_{ZL}	t_{ZH}	t_{LZ}	t_{HZ}	t_{ZLS}	t_{ZHS}	Units
4 mA	Std.	0.60	8.82	0.04	1.51	1.66	0.43	8.13	8.82	2.72	2.29	10.37	11.05	ns
	–1	0.51	7.50	0.04	1.29	1.41	0.36	6.92	7.50	2.31	1.95	8.82	9.40	ns
	–2	0.45	6.58	0.03	1.13	1.24	0.32	6.07	6.58	2.03	1.71	7.74	8.25	ns
8 mA	Std.	0.60	5.27	0.04	1.51	1.66	0.43	5.27	5.27	3.10	3.03	7.50	7.51	ns
	–1	0.51	4.48	0.04	1.29	1.41	0.36	4.48	4.48	2.64	2.58	6.38	6.38	ns
	–2	0.45	3.94	0.03	1.13	1.24	0.32	3.93	3.94	2.32	2.26	5.60	5.61	ns
12 mA	Std.	0.66	3.74	0.04	1.51	1.66	0.43	3.81	3.49	3.37	3.49	6.05	5.73	ns
	–1	0.56	3.18	0.04	1.29	1.41	0.36	3.24	2.97	2.86	2.97	5.15	4.87	ns
	–2	0.49	2.80	0.03	1.13	1.24	0.32	2.85	2.61	2.51	2.61	4.52	4.28	ns
16 mA	Std.	0.66	3.53	0.04	1.51	1.66	0.43	3.59	3.12	3.42	3.62	5.83	5.35	ns
	–1	0.56	3.00	0.04	1.29	1.41	0.36	3.06	2.65	2.91	3.08	4.96	4.55	ns
	–2	0.49	2.63	0.03	1.13	1.24	0.32	2.68	2.33	2.56	2.71	4.35	4.00	ns
24 mA	Std.	0.66	3.26	0.04	1.51	1.66	0.43	3.32	2.48	3.49	4.11	5.56	4.72	ns
	–1	0.56	2.77	0.04	1.29	1.41	0.36	2.83	2.11	2.97	3.49	4.73	4.01	ns
	–2	0.49	2.44	0.03	1.13	1.24	0.32	2.48	1.85	2.61	3.07	4.15	3.52	ns

Note: For the derating values at specific junction temperature and voltage supply levels, refer to Table 3-7 on page 3-9.

Table 2-122 • 1.8 V LVCMOS Low Slew
Commercial Temperature Range Conditions: T_J = 70°C, Worst-Case VCC = 1.425 V, Worst-Case VCCI = 1.7 V
Applicable to Advanced I/Os

Drive Strength	Speed Grade	t_{DOUT}	t_{DP}	t_{DIN}	t_{PY}	t_{EOUT}	t_{ZL}	t_{ZH}	t_{LZ}	t_{HZ}	t_{ZLS}	t_{ZHS}	Units
2 mA	Std.	0.66	15.53	0.04	1.31	0.43	14.11	15.53	2.78	1.60	16.35	17.77	ns
	–1	0.56	13.21	0.04	1.11	0.36	12.01	13.21	2.36	1.36	13.91	15.11	ns
	–2[2]	0.49	11.60	0.03	0.98	0.32	10.54	11.60	2.07	1.19	12.21	13.27	ns
4 mA	Std.	0.66	10.48	0.04	1.31	0.43	10.41	10.48	3.23	2.73	12.65	12.71	ns
	–1	0.56	8.91	0.04	1.11	0.36	8.86	8.91	2.75	2.33	10.76	10.81	ns
	–2	0.49	7.82	0.03	0.98	0.32	7.77	7.82	2.41	2.04	9.44	9.49	ns
8 mA	Std.	0.66	8.05	0.04	1.31	0.43	8.20	7.84	3.54	3.27	10.43	10.08	ns
	–1	0.56	6.85	0.04	1.11	0.36	6.97	6.67	3.01	2.78	8.88	8.57	ns
	–2	0.49	6.01	0.03	0.98	0.32	6.12	5.86	2.64	2.44	7.79	7.53	ns
12 mA	Std.	0.66	7.50	0.04	1.31	0.43	7.64	7.30	3.61	3.41	9.88	9.53	ns
	–1	0.56	6.38	0.04	1.11	0.36	6.50	6.21	3.07	2.90	8.40	8.11	ns
	–2	0.49	5.60	0.03	0.98	0.32	5.71	5.45	2.69	2.55	7.38	7.12	ns
16 mA	Std.	0.66	7.29	0.04	1.31	0.43	7.23	7.29	3.71	3.95	9.47	9.53	ns
	–1	0.56	6.20	0.04	1.11	0.36	6.15	6.20	3.15	3.36	8.06	8.11	ns
	–2	0.49	5.45	0.03	0.98	0.32	5.40	5.45	2.77	2.95	7.07	7.12	ns

Note: For the derating values at specific junction temperature and voltage supply levels, refer to Table 3-7 on page 3-9.

1.5 V LVCMOS (JESD8-11)

Low-Voltage CMOS for 1.5 V is an extension of the LVCMOS standard (JESD8-5) used for general-purpose 1.5 V applications. It uses a 1.5 V input buffer and push-pull output buffer.

Table 2-126 • Minimum and Maximum DC Input and Output Levels

1.5 V LVCMOS	VIL		VIH		VOL	VOH	IOL	IOH	IOSL	IOSH	IIL[1]	IIH[2]
Drive Strength	Min. V	Max. V	Min. V	Max. V	Max. V	Min. V	mA	mA	Max. mA[3]	Max. mA[3]	µA[4]	µA[4]
Applicable to Pro I/O Banks												
2 mA	–0.3	0.35 * VCCI	0.65 * VCCI	3.6	0.25 * VCCI	0.75 * VCCI	2	2	16	13	10	10
4 mA	–0.3	0.35 * VCCI	0.65 * VCCI	3.6	0.25 * VCCI	0.75 * VCCI	4	4	33	25	10	10
6 mA	–0.3	0.35 * VCCI	0.65 * VCCI	3.6	0.25 * VCCI	0.75 * VCCI	6	6	39	32	10	10
8 mA	–0.3	0.35 * VCCI	0.65 * VCCI	3.6	0.25 * VCCI	0.75 * VCCI	8	8	55	66	10	10
12 mA	–0.3	0.35 * VCCI	0.65 * VCCI	3.6	0.25 * VCCI	0.75 * VCCI	12	12	55	66	10	10
Applicable to Advanced I/O Banks												
2 mA	–0.3	0.35 * VCCI	0.65 * VCCI	1.575	0.25 * VCCI	0.75 * VCCI	2	2	16	13	10	10
4 mA	–0.3	0.35 * VCCI	0.65 * VCCI	1.575	0.25 * VCCI	0.75 * VCCI	4	4	33	25	10	10
6 mA	–0.3	0.35 * VCCI	0.65 * VCCI	1.575	0.25 * VCCI	0.75 * VCCI	6	6	39	32	10	10
8 mA	–0.3	0.35 * VCCI	0.65 * VCCI	1.575	0.25 * VCCI	0.75 * VCCI	8	8	55	66	10	10
12 mA	–0.3	0.35 * VCCI	0.65 * VCCI	1.575	0.25 * VCCI	0.75 * VCCI	12	12	55	66	10	10
Applicable to Pro I/O Banks												
2 mA	–0.3	0.35 * VCCI	0.65 * VCCI	3.6	0.25 * VCCI	0.75 * VCCI	2	2	16	13	10	10

Notes:

1. *IIL is the input leakage current per I/O pin over recommended operation conditions where –0.3 V < VIN < VIL.*
2. *I_{IH} is the input leakage current per I/O pin over recommended operating conditions VIH < VIN < VCCI. Input current is larger when operating outside recommended ranges.*
3. *Currents are measured at high temperature (100°C junction temperature) and maximum voltage.*
4. *Currents are measured at 85°C junction temperature.*
5. *Software default selection highlighted in gray.*

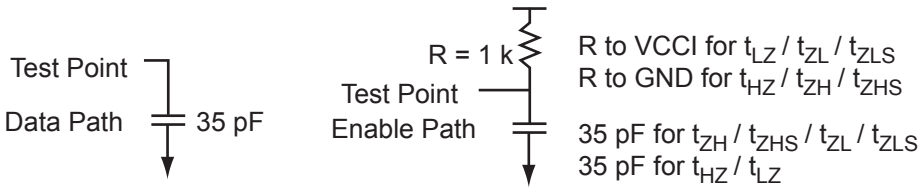


Figure 2-122 • AC Loading

Table 2-127 • AC Waveforms, Measuring Points, and Capacitive Loads

Input Low (V)	Input High (V)	Measuring Point* (V)	VREF (typ.) (V)	C_{LOAD} (pF)
0	1.5	0.75	–	35

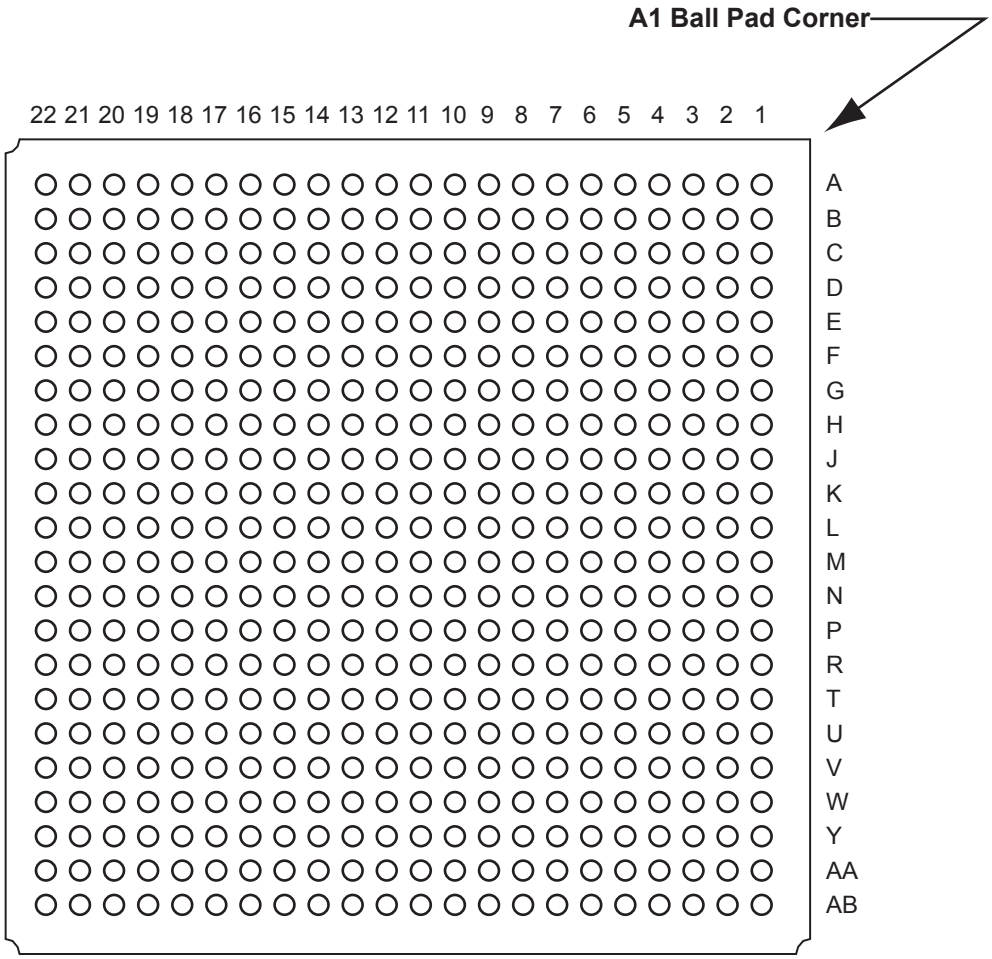
*Note: *Measuring point = Vtrip. See Table 2-90 on page 2-166 for a complete table of trip points.*

QN180		
Pin Number	AFS090 Function	AFS250 Function
A1	GNDQ	GNDQ
A2	VCCIB3	VCCIB3
A3	GAB2/IO52NDB3V0	IO74NDB3V0
A4	GFA2/IO51NDB3V0	IO71NDB3V0
A5	GFC2/IO50NDB3V0	IO69NPB3V0
A6	VCCIB3	VCCIB3
A7	GFA1/IO47PPB3V0	GFB1/IO67PPB3V0
A8	GEB0/IO45NDB3V0	NC
A9	XTAL1	XTAL1
A10	GNDOSC	GNDOSC
A11	GEC2/IO43PPB3V0	GEA1/IO61PPB3V0
A12	IO43NPB3V0	GEA0/IO61NPB3V0
A13	NC	VCCIB3
A14	GNDNVM	GNDNVM
A15	PCAP	PCAP
A16	VCC33PMP	VCC33PMP
A17	NC	NC
A18	AV0	AV0
A19	AG0	AG0
A20	ATRTN0	ATRTN0
A21	AG1	AG1
A22	AC1	AC1
A23	AV2	AV2
A24	AT2	AT2
A25	AT3	AT3
A26	AC3	AC3
A27	AV4	AV4
A28	AC4	AC4
A29	AT4	AT4
A30	NC	AG5
A31	NC	AV5
A32	ADCGNDREF	ADCGNDREF
A33	VCC33A	VCC33A
A34	GNDA	GNDA
A35	PTBASE	PTBASE
A36	VCCNVM	VCCNVM

QN180		
Pin Number	AFS090 Function	AFS250 Function
A37	VPUMP	VPUMP
A38	TDI	TDI
A39	TDO	TDO
A40	VJTAG	VJTAG
A41	GDB1/IO39PPB1V0	GDA1/IO54PPB1V0
A42	GDC1/IO38PDB1V0	GDB1/IO53PDB1V0
A43	VCC	VCC
A44	GCB0/IO35NPB1V0	GCB0/IO48NPB1V0
A45	GCC1/IO34PDB1V0	GCC1/IO47PDB1V0
A46	VCCIB1	VCCIB1
A47	GBC2/IO32PPB1V0	GBB2/IO41PPB1V0
A48	VCCIB1	VCCIB1
A49	NC	NC
A50	GBA0/IO29RSB0V0	GBB1/IO37RSB0V0
A51	VCCIB0	VCCIB0
A52	GBB0/IO27RSB0V0	GBC0/IO34RSB0V0
A53	GBC1/IO26RSB0V0	IO33RSB0V0
A54	IO24RSB0V0	IO29RSB0V0
A55	IO21RSB0V0	IO26RSB0V0
A56	VCCIB0	VCCIB0
A57	IO15RSB0V0	IO21RSB0V0
A58	IO10RSB0V0	IO13RSB0V0
A59	IO07RSB0V0	IO10RSB0V0
A60	GAC0/IO04RSB0V0	IO06RSB0V0
A61	GAB1/IO03RSB0V0	GAC1/IO05RSB0V0
A62	VCC	VCC
A63	GAA1/IO01RSB0V0	GAB0/IO02RSB0V0
A64	NC	NC
B1	VCOMPLA	VCOMPLA
B2	GAA2/IO52PDB3V0	GAC2/IO74PDB3V0
B3	GAC2/IO51PDB3V0	GFA2/IO71PDB3V0
B4	GFB2/IO50PDB3V0	GFB2/IO70PSB3V0
B5	VCC	VCC
B6	GFC0/IO49NDB3V0	GFC0/IO68NDB3V0
B7	GEB1/IO45PDB3V0	NC
B8	VCCOSC	VCCOSC

FG256				
Pin Number	**AFS090 Function**	**AFS250 Function**	**AFS600 Function**	**AFS1500 Function**
R5	AV0	AV0	AV2	AV2
R6	AT0	AT0	AT2	AT2
R7	AV1	AV1	AV3	AV3
R8	AT3	AT3	AT5	AT5
R9	AV4	AV4	AV6	AV6
R10	NC	AT5	AT7	AT7
R11	NC	AV5	AV7	AV7
R12	NC	NC	AT9	AT9
R13	NC	NC	AG9	AG9
R14	NC	NC	AC9	AC9
R15	PUB	PUB	PUB	PUB
R16	VCCIB1	VCCIB1	VCCIB2	VCCIB2
T1	GND	GND	GND	GND
T2	NCAP	NCAP	NCAP	NCAP
T3	VCC33N	VCC33N	VCC33N	VCC33N
T4	NC	NC	ATRTN0	ATRTN0
T5	AT1	AT1	AT3	AT3
T6	ATRTN0	ATRTN0	ATRTN1	ATRTN1
T7	AT2	AT2	AT4	AT4
T8	ATRTN1	ATRTN1	ATRTN2	ATRTN2
T9	AT4	AT4	AT6	AT6
T10	ATRTN2	ATRTN2	ATRTN3	ATRTN3
T11	NC	NC	AT8	AT8
T12	NC	NC	ATRTN4	ATRTN4
T13	GNDA	GNDA	GNDA	GNDA
T14	VCC33A	VCC33A	VCC33A	VCC33A
T15	VAREF	VAREF	VAREF	VAREF
T16	GND	GND	GND	GND

FG484



Note

For Package Manufacturing and Environmental information, visit the Resource Center at http://www.microsemi.com/soc/products/solutions/package/default.aspx.

FG676	
Pin Number	AFS1500 Function
R21	IO72NDB2V0
R22	IO72PDB2V0
R23	GND
R24	IO71PDB2V0
R25	VCCIB2
R26	IO67NDB2V0
T1	GND
T2	NC
T3	GFA1/IO105PDB4V0
T4	GFA0/IO105NDB4V0
T5	IO101NDB4V0
T6	IO96PDB4V0
T7	IO96NDB4V0
T8	IO99NDB4V0
T9	IO97NDB4V0
T10	VCCIB4
T11	VCC
T12	GND
T13	VCC
T14	GND
T15	VCC
T16	GND
T17	VCCIB2
T18	IO83NDB2V0
T19	IO78NDB2V0
T20	GDA1/IO81PDB2V0
T21	GDB1/IO80PDB2V0
T22	IO73NDB2V0
T23	IO73PDB2V0
T24	IO71NDB2V0
T25	NC
T26	GND
U1	NC
U2	NC
U3	IO102PDB4V0
U4	IO102NDB4V0

FG676	
Pin Number	AFS1500 Function
U5	VCCIB4
U6	IO91PDB4V0
U7	IO91NDB4V0
U8	IO92PDB4V0
U9	GND
U10	GND
U11	VCC33A
U12	GNDA
U13	VCC33A
U14	GNDA
U15	VCC33A
U16	GNDA
U17	VCC
U18	GND
U19	IO74NDB2V0
U20	GDA0/IO81NDB2V0
U21	GDB0/IO80NDB2V0
U22	VCCIB2
U23	IO75NDB2V0
U24	IO75PDB2V0
U25	NC
U26	NC
V1	NC
V2	VCCIB4
V3	IO100PPB4V0
V4	GND
V5	IO95PDB4V0
V6	IO95NDB4V0
V7	VCCIB4
V8	IO92NDB4V0
V9	GNDNVM
V10	GNDA
V11	NC
V12	AV4
V13	NC
V14	AV5

FG676	
Pin Number	AFS1500 Function
V15	AC5
V16	NC
V17	GNDA
V18	IO77PPB2V0
V19	IO74PDB2V0
V20	VCCIB2
V21	IO82NDB2V0
V22	GDA2/IO82PDB2V0
V23	GND
V24	GDC1/IO79PDB2V0
V25	VCCIB2
V26	NC
W1	GND
W2	IO94PPB4V0
W3	IO98PDB4V0
W4	IO98NDB4V0
W5	GEC1/IO90PDB4V0
W6	GEC0/IO90NDB4V0
W7	GND
W8	VCCNVM
W9	VCCIB4
W10	VCC15A
W11	GNDA
W12	AC4
W13	VCC33A
W14	GNDA
W15	AG5
W16	GNDA
W17	PUB
W18	VCCIB2
W19	TDI
W20	GND
W21	IO84NDB2V0
W22	GDC2/IO84PDB2V0
W23	IO77NPB2V0
W24	GDC0/IO79NDB2V0

FG676	
Pin Number	**AFS1500 Function**
W25	NC
W26	GND
Y1	NC
Y2	NC
Y3	GEB1/IO89PDB4V0
Y4	GEB0/IO89NDB4V0
Y5	VCCIB4
Y6	GEA1/IO88PDB4V0
Y7	GEA0/IO88NDB4V0
Y8	GND
Y9	VCC33PMP
Y10	NC
Y11	VCC33A
Y12	AG4
Y13	AT4
Y14	ATRTN2
Y15	AT5
Y16	VCC33A
Y17	NC
Y18	VCC33A
Y19	GND
Y20	TMS
Y21	VJTAG
Y22	VCCIB2
Y23	TRST
Y24	TDO
Y25	NC
Y26	NC