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Understanding [Embedded - FPGAs \(Field Programmable Gate Array\)](#)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details

Product Status	Active
Number of LABs/CLBs	-
Number of Logic Elements/Cells	-
Total RAM Bits	110592
Number of I/O	172
Number of Gates	600000
Voltage - Supply	1.425V ~ 1.575V
Mounting Type	Surface Mount
Operating Temperature	-40°C ~ 100°C (TJ)
Package / Case	484-BGA
Supplier Device Package	484-FPBGA (23x23)
Purchase URL	https://www.e-xfl.com/product-detail/microchip-technology/p1afs600-2fg484i

The system application, Level 3, is the larger user application that utilizes one or more applets. Designing at the highest level of abstraction supported by the Fusion technology stack, the application can be easily created in FPGA gates by importing and configuring multiple applets.

In fact, in some cases an entire FPGA system design can be created without any HDL coding.

An optional MCU enables a combination of software and HDL-based design methodologies. The MCU can be on-chip or off-chip as system requirements dictate. System portioning is very flexible, allowing the MCU to reside above the applets or to absorb applets, or applets and backbone, if desired.

The Fusion technology stack enables a very flexible design environment. Users can engage in design across a continuum of abstraction from very low to very high.

Core Architecture

VersaTile

Based upon successful ProASIC3/E logic architecture, Fusion devices provide granularity comparable to gate arrays. The Fusion device core consists of a sea-of-VersaTiles architecture.

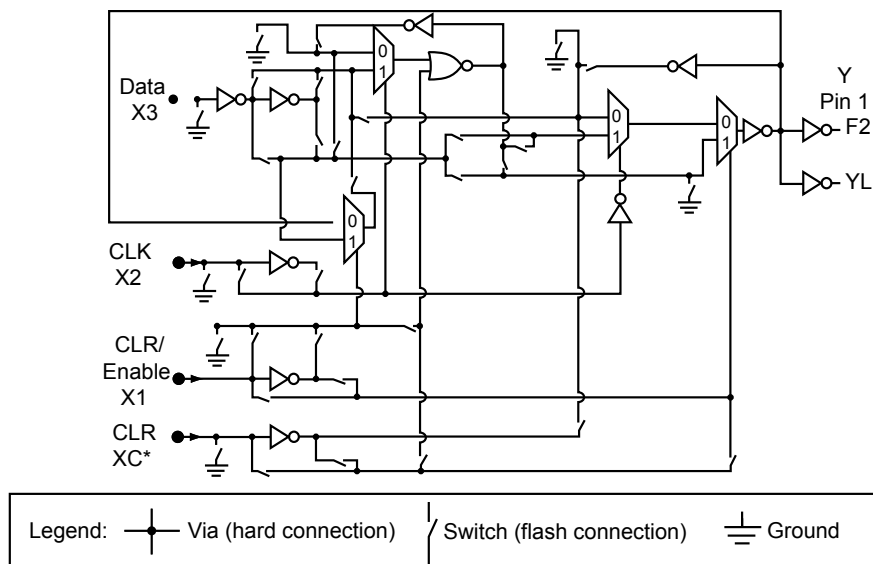
As illustrated in [Figure 2-2](#), there are four inputs in a logic VersaTile cell, and each VersaTile can be configured using the appropriate flash switch connections:

- Any 3-input logic function
- Latch with clear or set
- D-flip-flop with clear or set
- Enable D-flip-flop with clear or set (on a 4th input)

VersaTiles can flexibly map the logic and sequential gates of a design. The inputs of the VersaTile can be inverted (allowing bubble pushing), and the output of the tile can connect to high-speed, very-long-line routing resources. VersaTiles and larger functions are connected with any of the four levels of routing hierarchy.

When the VersaTile is used as an enable D-flip-flop, the SET/CLR signal is supported by a fourth input, which can only be routed to the core cell over the VersaNet (global) network.

The output of the VersaTile is F2 when the connection is to the ultra-fast local lines, or YL when the connection is to the efficient long-line or very-long-line resources ([Figure 2-2](#)).



Note: *This input can only be connected to the global clock distribution network.

Figure 2-2 • Fusion Core VersaTile

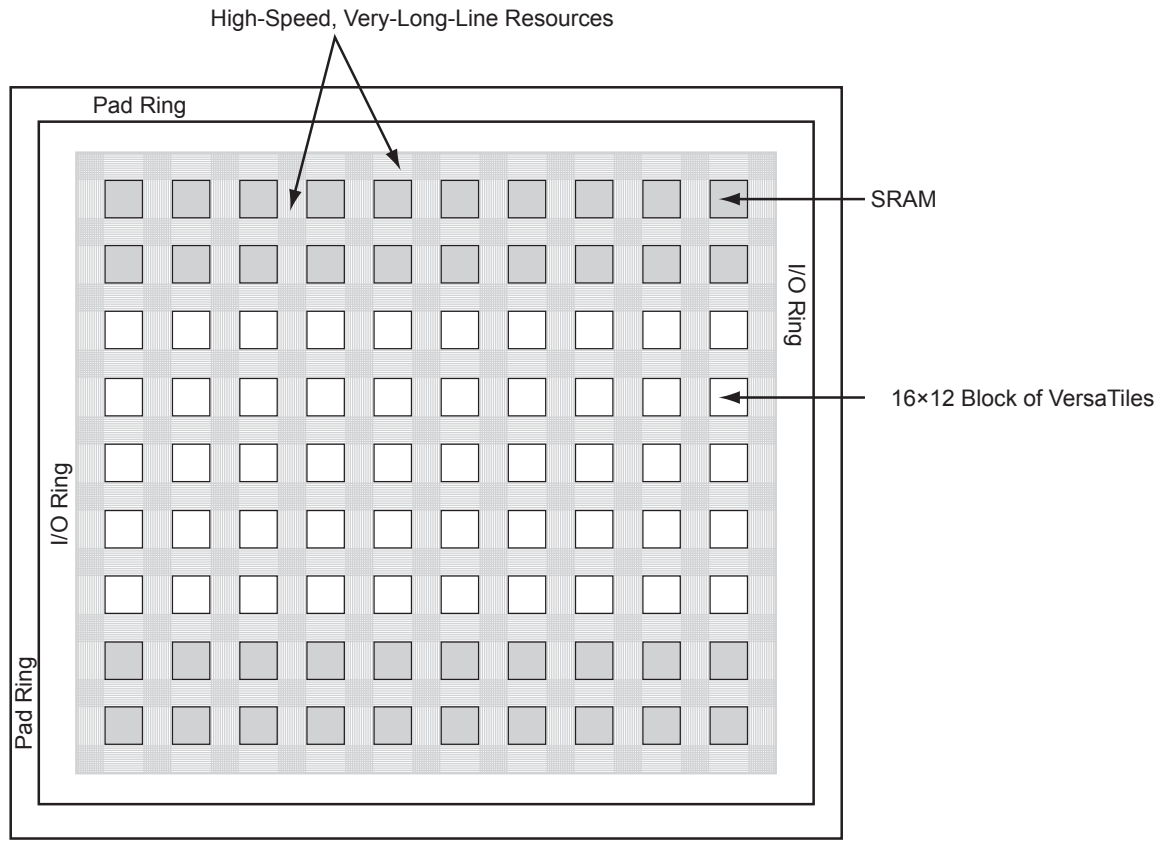
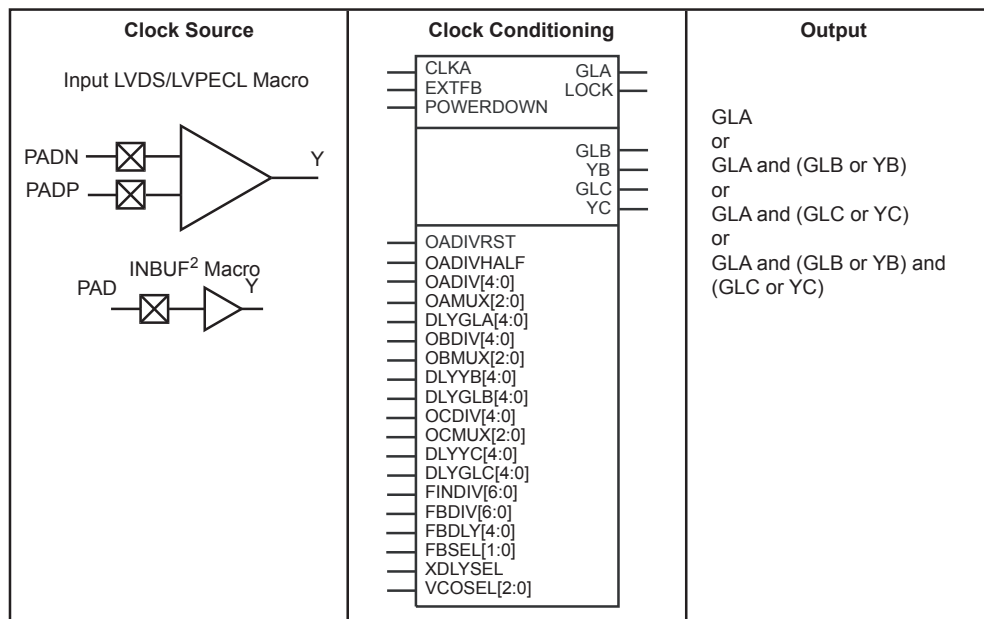


Figure 2-10 • Very-Long-Line Resources



Notes:

1. Visit the [Microsemi SoC Products Group website](#) for application notes concerning dynamic PLL reconfiguration. Refer to the "PLL Macro" section on page 2-27 for signal descriptions.
2. Many specific INBUF macros support the wide variety of single-ended and differential I/O standards for the Fusion family.
3. Refer to the [IGLOO](#), [ProASIC3](#), [SmartFusion](#) and [Fusion Macro Library Guide](#) for more information.

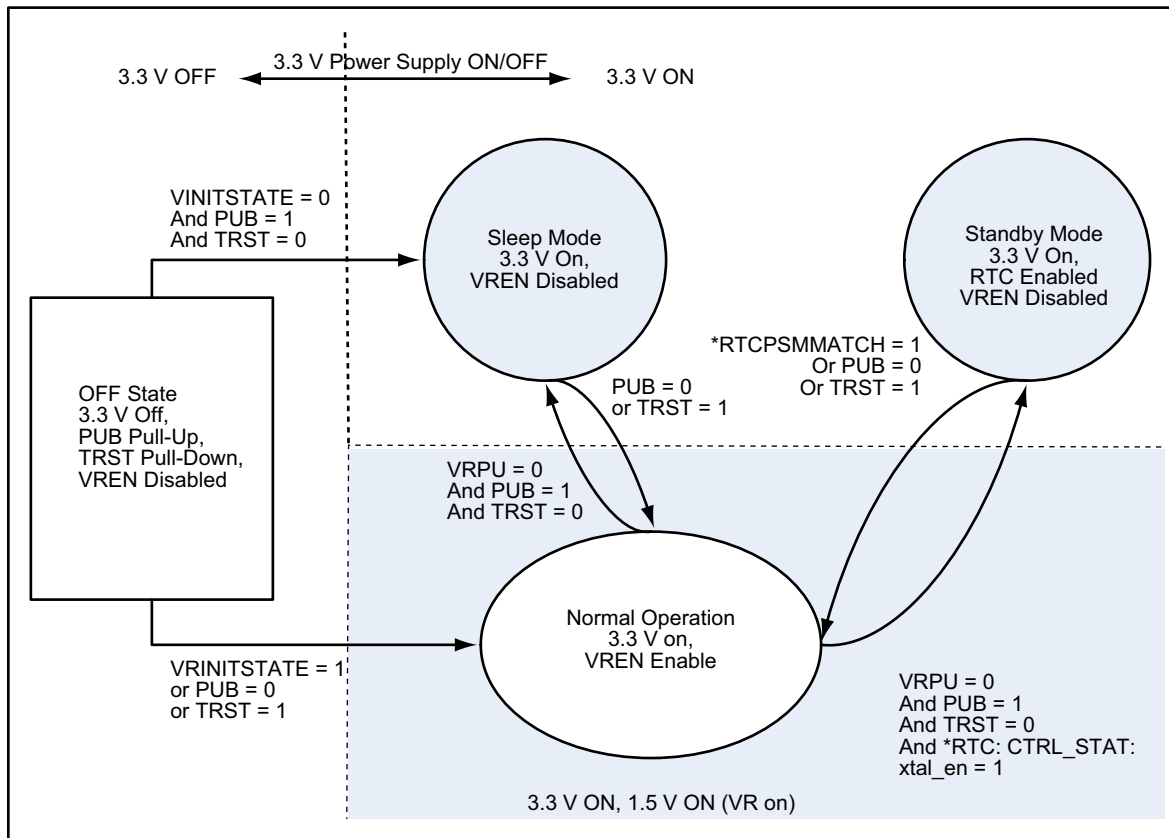
Figure 2-19 • Fusion CCC Options: Global Buffers with the PLL Macro

Table 2-11 • Available Selections of I/O Standards within CLKBUF and CLKBUF_LVDS/LVPECL Macros

CLKBUF Macros
CLKBUF_LVCMOS5
CLKBUF_LVCMOS33 ¹
CLKBUF_LVCMOS18
CLKBUF_LVCMOS15
CLKBUF_PCI
CLKBUF_LVDS ²
CLKBUF_LVPECL

Notes:

1. This is the default macro. For more details, refer to the [IGLOO](#), [ProASIC3](#), [SmartFusion](#) and [Fusion Macro Library Guide](#).
2. The B-LVDS and M-LVDS standards are supported with CLKBUF_LVDS.



Note: * To enter and exit standby mode without any external stimulus on PUB or TRST, the *vr_en_mat* in the CTRL_STAT register must also be set to 1, so that RTCPSMMATCH will assert when a match occurs; hence the device exits standby mode.

Figure 2-31 • State Diagram for All Different Power Modes

When TRST is 1 or PUB is 0, the 1.5 V voltage regulator is always ON, putting the Fusion device in normal operation at all times. Therefore, when the JTAG port is not in reset, the Fusion device cannot enter sleep mode or standby mode.

To enter standby mode, the Fusion device must first power-up into normal operation. The RTC is enabled through the RTC Control/Status Register described in the ["Real-Time Counter \(part of AB macro\)" section on page 2-33](#). A match value corresponding to the wake-up time is loaded into the Match Register. The 1.5 V voltage regulator is disabled by setting VRPU to 0 to allow the Fusion device to enter standby mode, when the 1.5 V supply is off but the RTC remains on.

1.5 V Voltage Regulator

The 1.5 V voltage regulator uses an external pass transistor to generate 1.5 V from a 3.3 V supply. The base of the pass transistor is tied to PTBASE, the collector is tied to 3.3 V, and an emitter is tied to PTBASE and the 1.5 V supplies of the Fusion device. [Figure 2-27 on page 2-31](#) shows the hook-up of the 1.5 V voltage regulator to an external pass transistor.

Microsemi recommends using a PN2222A or 2N2222A transistor. The gain of such a transistor is approximately 25, with a maximum base current of 20 mA. The maximum current that can be supported is 0.5 A. Transistors with different gain can also be used for different current requirements.

Table 2-18 • Electrical Characteristics
VCC33A = 3.3 V

Symbol	Parameter	Condition		Min	Typical	Max	Units
VOUT	Output Voltage	Tj = 25°C		1.425	1.5	1.575	V
ICC33A	Operation Current	Tj = 25°C	ILOAD = 1 mA ILOAD = 100 mA ILOAD = 0.5 A		11 11 30		mA mA mA
ΔVOUT	Load Regulation	Tj = 25°C	ILOAD = 1 mA to 0.5 A		90		mV
ΔVOUT	Line Regulation	Tj = 25°C	VCC33A = 2.97 V to 3.63 V ILOAD = 1 mA		10.6		mV/V
			VCC33A = 2.97 V to 3.63 V ILOAD = 100 mA		12.1		mV/V
			VCC33A = 2.97 V to 3.63 V ILOAD = 500 mA		10.6		mV/V
	Dropout Voltage*	Tj = 25°C	ILOAD = 1 mA		0.63		V
			ILOAD = 100 mA		0.84		V
			ILOAD = 0.5 A		1.35		V
IPTBASE	PTBase Current	Tj = 25°C	ILOAD = 1 mA ILOAD = 100 mA ILOAD = 0.5 A		48 736 12	20	μA μA mA

Note: *Data collected with 2N2222A.

Program Operation

A Program operation is initiated by asserting the PROGRAM signal on the interface. Program operations save the contents of the Page Buffer to the FB Array. Due to the technologies inherent in the FB, the total programming (including erase) time per page of the eNVM is 6.8 ms. While the FB is writing the data to the array, the BUSY signal will be asserted.

During a Program operation, the sector and page addresses on ADDR are compared with the stored address for the page (and sector) in the Page Buffer. If there is a mismatch between the two addresses, the Program operation will be aborted and an error will be reported on the STATUS output.

It is possible to write the Page Buffer to a different page in memory. When asserting the PROGRAM pin, if OVERWRITEPAGE is asserted as well, the FB will write the contents of the Page Buffer to the sector and page designated on the ADDR inputs if the destination page is not Overwrite Protected.

A Program operation can be utilized to either modify the contents of the page in the flash memory block or change the protections for the page. Setting the OVERWRITEPROTECT bit on the interface while asserting the PROGRAM pin will put the page addressed into Overwrite Protect Mode. Overwrite Protect Mode safeguards a page from being inadvertently overwritten during subsequent Program or Erase operations.

Program operations that result in a STATUS value of '01' do not modify the addressed page. For all other values of STATUS, the addressed page is modified. Program errors include the following:

1. Attempting to program a page that is Overwrite Protected (STATUS = '01')
2. Attempting to program a page that is not in the Page Buffer when the Page Buffer has entered Page Loss Protection Mode (STATUS = '01')
3. Attempting to perform a program with OVERWRITEPAGE set when the page addressed has been Overwrite Protected (STATUS = '01')
4. The Write Count of the page programmed exceeding the Write Threshold defined in the part specification (STATUS = '11')
5. The ECC Logic determining that there is an uncorrectable error within the programmed page (STATUS = '10')
6. Attempting to program a page that is **not** in the Page Buffer when OVERWRITEPAGE is not set and the page in the Page Buffer is modified (STATUS = '01')
7. Attempting to program the page in the Page Buffer when the Page Buffer is **not** modified

The waveform for a Program operation is shown in [Figure 2-36](#).

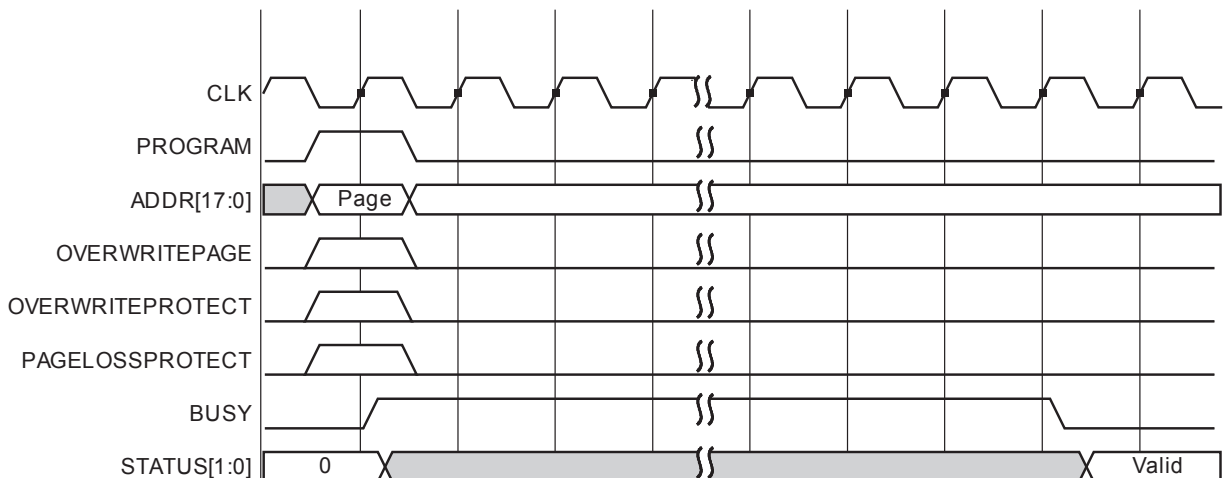


Figure 2-36 • FB Program Waveform

Note: OVERWRITEPAGE is only sampled when the PROGRAM or ERASEPAGE pins are asserted. OVERWRITEPAGE is ignored in all other operations.

Table 2-25 • Flash Memory Block Timing (continued)
Commercial Temperature Range Conditions: $T_J = 70^{\circ}\text{C}$, Worst-Case $V_{CC} = 1.425\text{ V}$

Parameter	Description	–2	–1	Std.	Units
$t_{\text{SUPGLOSSPRO}}$	Page Loss Protect Setup Time for the Control Logic	1.69	1.93	2.27	ns
$t_{\text{HDPGLOSSPRO}}$	Page Loss Protect Hold Time for the Control Logic	0.00	0.00	0.00	ns
t_{SUPGSTAT}	Page Status Setup Time for the Control Logic	2.49	2.83	3.33	ns
t_{HDPGSTAT}	Page Status Hold Time for the Control Logic	0.00	0.00	0.00	ns
$t_{\text{SUOVERWRPG}}$	Over Write Page Setup Time for the Control Logic	1.88	2.14	2.52	ns
$t_{\text{HDOVERWRPG}}$	Over Write Page Hold Time for the Control Logic	0.00	0.00	0.00	ns
$t_{\text{SULOCKREQUEST}}$	Lock Request Setup Time for the Control Logic	0.87	0.99	1.16	ns
$t_{\text{HDLOCKREQUEST}}$	Lock Request Hold Time for the Control Logic	0.00	0.00	0.00	ns
t_{REARNVM}	Reset Recovery Time	0.94	1.07	1.25	ns
t_{REARNVM}	Reset Removal Time	0.00	0.00	0.00	ns
t_{MPWARNVM}	Asynchronous Reset Minimum Pulse Width for the Control Logic	10.00	12.50	12.50	ns
$t_{\text{MPWCLKNVM}}$	Clock Minimum Pulse Width for the Control Logic	4.00	5.00	5.00	ns
$t_{\text{FMAXCLKNVM}}$	Maximum Frequency for Clock for the Control Logic – for AFS1500/AFS600	80.00	80.00	80.00	MHz
	Maximum Frequency for Clock for the Control Logic – for AFS250/AFS090	100.00	80.00	80.00	MHz

FlashROM

Fusion devices have 1 kbit of on-chip nonvolatile flash memory that can be read from the FPGA core fabric. The FlashROM is arranged in eight banks of 128 bits during programming. The 128 bits in each bank are addressable as 16 bytes during the read-back of the FlashROM from the FPGA core ([Figure 2-45](#)).

The FlashROM can only be programmed via the IEEE 1532 JTAG port. It cannot be programmed directly from the FPGA core. When programming, each of the eight 128-bit banks can be selectively reprogrammed. The FlashROM can only be reprogrammed on a bank boundary. Programming involves an automatic, on-chip bank erase prior to reprogramming the bank. The FlashROM supports a synchronous read and can be read on byte boundaries. The upper three bits of the FlashROM address from the FPGA core define the bank that is being accessed. The lower four bits of the FlashROM address from the FPGA core define which of the 16 bytes in the bank is being accessed.

The maximum FlashROM access clock is given in [Table 2-26 on page 2-54](#). [Figure 2-46](#) shows the timing behavior of the FlashROM access cycle—the address has to be set up on the rising edge of the clock for DOUT to be valid on the next falling edge of the clock.

If the address is unchanged for two cycles:

- D0 becomes invalid t_{CK2Q} ns after the second rising edge of the clock.
- D0 becomes valid again t_{CK2Q} ns after the second falling edge.

If the address is unchanged for three cycles:

- D0 becomes invalid t_{CK2Q} ns after the second rising edge of the clock.
- D0 becomes valid again t_{CK2Q} ns after the second falling edge.
- D0 becomes invalid t_{CK2Q} ns after the third rising edge of the clock.
- D0 becomes valid again t_{CK2Q} ns after the third falling edge.

ESTOP, FSTOP

ESTOP is used to stop the FIFO read counter from further counting once the FIFO is empty (i.e., the EMPTY flag goes High). A High on this signal inhibits the counting.

FSTOP is used to stop the FIFO write counter from further counting once the FIFO is full (i.e., the FULL flag goes High). A High on this signal inhibits the counting.

For more information on these signals, refer to the ["ESTOP and FSTOP Usage" section on page 2-70](#).

FULL, EMPTY

When the FIFO is full and no more data can be written, the FULL flag asserts High. The FULL flag is synchronous to WCLK to inhibit writing immediately upon detection of a full condition and to prevent overflows. Since the write address is compared to a resynchronized (and thus time-delayed) version of the read address, the FULL flag will remain asserted until two WCLK active edges after a read operation eliminates the full condition.

When the FIFO is empty and no more data can be read, the EMPTY flag asserts High. The EMPTY flag is synchronous to RCLK to inhibit reading immediately upon detection of an empty condition and to prevent underflows. Since the read address is compared to a resynchronized (and thus time-delayed) version of the write address, the EMPTY flag will remain asserted until two RCLK active edges after a write operation removes the empty condition.

For more information on these signals, refer to the ["FIFO Flag Usage Considerations" section on page 2-70](#).

AFULL, AEMPTY

These are programmable flags and will be asserted on the threshold specified by AFVAL and AEVAL, respectively.

When the number of words stored in the FIFO reaches the amount specified by AEVAL while reading, the AEMPTY output will go High. Likewise, when the number of words stored in the FIFO reaches the amount specified by AFVAL while writing, the AFULL output will go High.

AFVAL, AEVAL

The AEVAL and AFVAL pins are used to specify the almost-empty and almost-full threshold values, respectively. They are 12-bit signals. For more information on these signals, refer to ["FIFO Flag Usage Considerations" section](#).

ESTOP and FSTOP Usage

The ESTOP pin is used to stop the read counter from counting any further once the FIFO is empty (i.e., the EMPTY flag goes High). Likewise, the FSTOP pin is used to stop the write counter from counting any further once the FIFO is full (i.e., the FULL flag goes High).

The FIFO counters in the Fusion device start the count at 0, reach the maximum depth for the configuration (e.g., 511 for a 512×9 configuration), and then restart at 0. An example application for the ESTOP, where the read counter keeps counting, would be writing to the FIFO once and reading the same content over and over without doing another write.

FIFO Flag Usage Considerations

The AEVAL and AFVAL pins are used to specify the 12-bit AEMPTY and AFULL threshold values, respectively. The FIFO contains separate 12-bit write address (WADDR) and read address (RADDR) counters. WADDR is incremented every time a write operation is performed, and RADDR is incremented every time a read operation is performed. Whenever the difference between WADDR and RADDR is greater than or equal to AFVAL, the AFULL output is asserted. Likewise, whenever the difference between WADDR and RADDR is less than or equal to AEVAL, the AEMPTY output is asserted. To handle different read and write aspect ratios, AFVAL and AEVAL are expressed in terms of total data bits instead of total data words. When users specify AFVAL and AEVAL in terms of read or write words, the SmartGen tool translates them into bit addresses and configures these signals automatically. SmartGen configures the AFULL flag to assert when the write address exceeds the read address by at least a predefined value. In a 2k×8 FIFO, for example, a value of 1,500 for AFVAL means that the AFULL flag will be asserted after a write when the difference between the write address and the read address reaches 1,500 (there have been at least 1500 more writes than reads). It will stay asserted until the difference between the write and read addresses drops below 1,500.

Table 2-36 describes each pin in the Analog Block. Each function within the Analog Block will be explained in detail in the following sections.

Table 2-36 • Analog Block Pin Description

Signal Name	Number of Bits	Direction	Function	Location of Details
VAREF	1	Input/Output	Voltage reference for ADC	ADC
ADCGNDREF	1	Input	External ground reference	ADC
MODE[3:0]	4	Input	ADC operating mode	ADC
SYSCLK	1	Input	External system clock	
TVC[7:0]	8	Input	Clock divide control	ADC
STC[7:0]	8	Input	Sample time control	ADC
ADCSTART	1	Input	Start of conversion	ADC
PWRDWN	1	Input	ADC comparator power-down if 1. When asserted, the ADC will stop functioning, and the digital portion of the analog block will continue operating. This may result in invalid status flags from the analog block. Therefore, Microsemi does not recommend asserting the PWRDWN pin.	ADC
ADCRESET	1	Input	ADC resets and disables Analog Quad – active high	ADC
BUSY	1	Output	1 – Running conversion	ADC
CALIBRATE	1	Output	1 – Power-up calibration	ADC
DATAVALID	1	Output	1 – Valid conversion result	ADC
RESULT[11:0]	12	Output	Conversion result	ADC
TMSTBINT	1	Input	Internal temp. monitor strobe	ADC
SAMPLE	1	Output	1 – An analog signal is actively being sampled (stays high during signal acquisition only) 0 – No analog signal is being sampled	ADC
VAREFSEL	1	Input	0 = Output internal voltage reference (2.56 V) to VAREF 1 = Input external voltage reference from VAREF and ADCGNDREF	ADC
CHNUMBER[4:0]	5	Input	Analog input channel select	Input multiplexer
ACMCLK	1	Input	ACM clock	ACM
ACMWEN	1	Input	ACM write enable – active high	ACM
ACMRESET	1	Input	ACM reset – active low	ACM
ACMWDATA[7:0]	8	Input	ACM write data	ACM
ACMRDATA[7:0]	8	Output	ACM read data	ACM
ACMADDR[7:0]	8	Input	ACM address	ACM
CMSTB0 to CMSTB9	10	Input	Current monitor strobe – 1 per quad, active high	Analog Quad

Gate Driver

The Fusion Analog Quad includes a Gate Driver connected to the Quad's AG pin (Figure 2-74). Designed to work with external p- or n-channel MOSFETs, the Gate driver is a configurable current sink or source and requires an external pull-up or pull-down resistor. The AG supports 4 selectable gate drive levels: 1 μA , 3 μA , 10 μA , and 30 μA (Figure 2-75 on page 2-91). The AG also supports a High Current Drive mode in which it can sink 20 mA; in this mode the switching rate is approximately 1.3 MHz with 100 ns turn-on time and 600 ns turn-off time. Modeled on an open-drain-style output, it does not output a voltage level without an appropriate pull-up or pull-down resistor. If 1 V is forced on the drain, the current sinking/sourcing will exceed the ability of the transistor, and the device could be damaged.

The AG pad is turned on via the corresponding GDONx pin in the Analog Block macro, where x is the number of the corresponding Analog Quad for the AG pad to be enabled (GDON0 to GDON9).

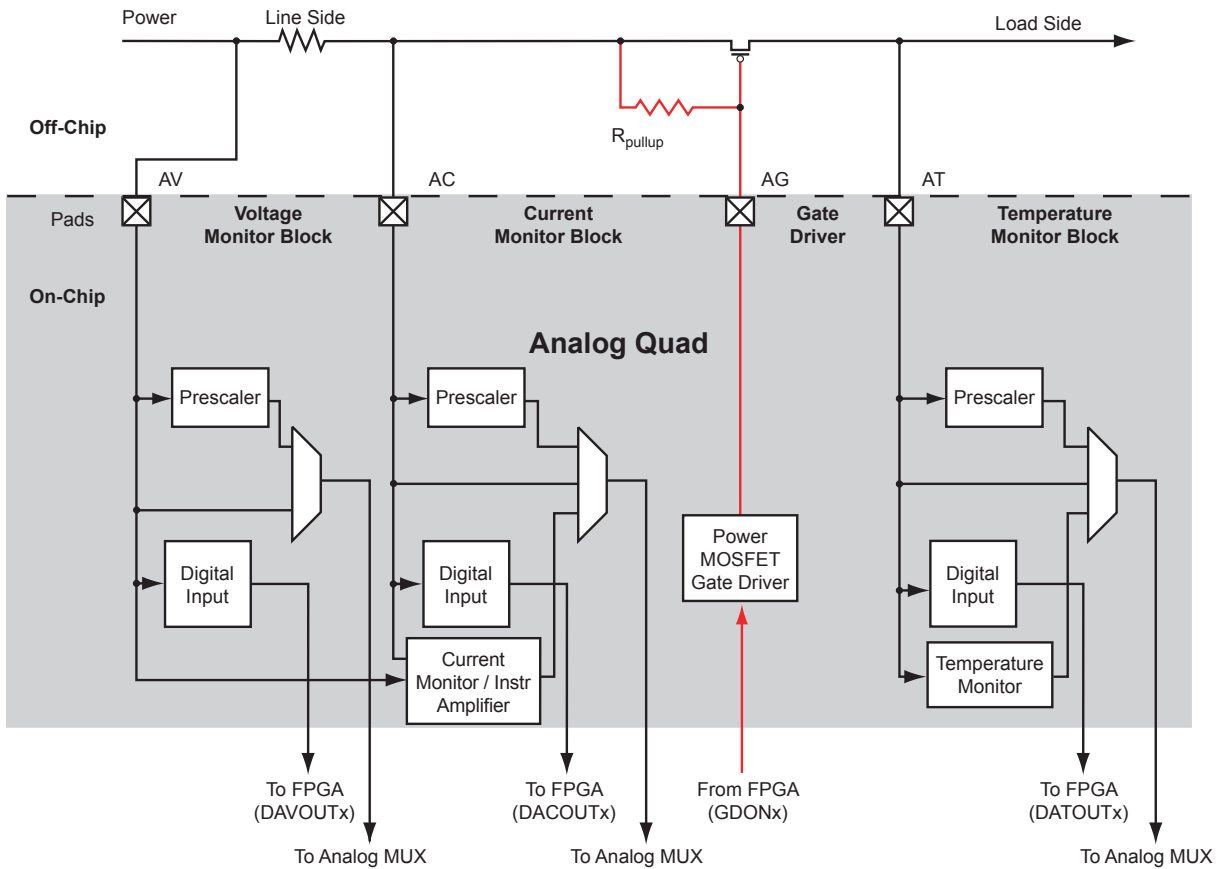


Figure 2-74 • Gate Driver

The gate-to-source voltage (V_{gs}) of the external MOSFET is limited to the programmable drive current times the external pull-up or pull-down resistor value (EQ 5).

$$V_{gs} \leq I_g \times (R_{pullup} \text{ or } R_{pulldown})$$

EQ 5

Hot-Swap Support

Hot-swapping (also called hot plugging) is the operation of hot insertion or hot removal of a card in (or from) a powered-up system. The levels of hot-swap support and examples of related applications are described in [Table 2-74](#). The I/Os also need to be configured in hot insertion mode if hot plugging compliance is required.

Table 2-74 • Levels of Hot-Swap Support

Hot Swapping Level	Description	Power Applied to Device	Bus State	Card Ground Connection	Device Circuitry Connected to Bus Pins	Example of Application with Cards that Contain Fusion Devices	Compliance of Fusion Devices
1	Cold-swap	No	—	—	—	System and card with Microsemi FPGA chip are powered down, then card gets plugged into system, then power supplies are turned on for system but not for FPGA on card.	Compliant I/Os can but do not have to be set to hot insertion mode.
2	Hot-swap while reset	Yes	Held in reset state	Must be made and maintained for 1 ms before, during, and after insertion/removal	—	In PCI hot plug specification, reset control circuitry isolates the card busses until the card supplies are at their nominal operating levels and stable.	Compliant I/Os can but do not have to be set to hot insertion mode.
3	Hot-swap while bus idle	Yes	Held idle (no ongoing I/O processes during insertion/removal)	Same as Level 2	Must remain glitch-free during power-up or power-down	Board bus shared with card bus is "frozen," and there is no toggling activity on bus. It is critical that the logic states set on the bus signal do not get disturbed during card insertion/removal.	Compliant with cards with two levels of staging. I/Os have to be set to hot insertion mode.
4	Hot-swap on an active bus	Yes	Bus may have active I/O processes ongoing, but device being inserted or removed must be idle.	Same as Level 2	Same as Level 3	There is activity on the system bus, and it is critical that the logic states set on the bus signal do not get disturbed during card insertion/removal.	Compliant with cards with two levels of staging. I/Os have to be set to hot insertion mode.

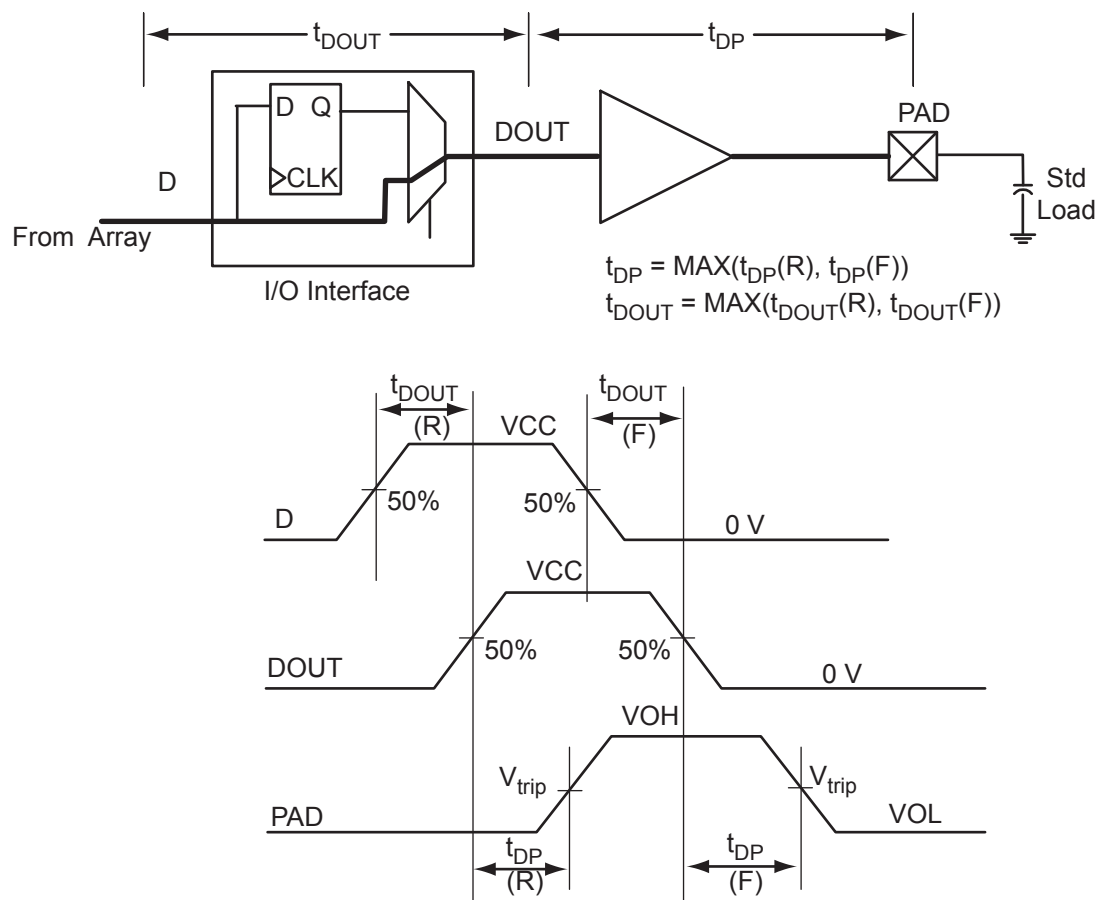


Figure 2-117 • Output Buffer Model and Delays (example)

Table 2-98 • I/O Short Currents IOSH/IOSL

	Drive Strength	IOSH (mA)*	IOSL (mA)*
Applicable to Pro I/O Banks			
3.3 V LVTTTL / 3.3 V LVCMOS	4 mA	25	27
	8 mA	51	54
	12 mA	103	109
	16 mA	132	127
	24 mA	268	181
2.5 V LVCMOS	4 mA	16	18
	8 mA	32	37
	12 mA	65	74
	16 mA	83	87
	24 mA	169	124
1.8 V LVCMOS	2 mA	9	11
	4 mA	17	22
	6 mA	35	44
	8 mA	45	51
	12 mA	91	74
	16 mA	91	74
1.5 V LVCMOS	2 mA	13	16
	4 mA	25	33
	6 mA	32	39
	8 mA	66	55
	12 mA	66	55
Applicable to Advanced I/O Banks			
3.3 V LVTTTL / 3.3 V LVCMOS	2 mA	25	27
	4 mA	25	27
	6 mA	51	54
	8 mA	51	54
	12 mA	103	109
	16 mA	132	127
	24 mA	268	181
3.3 V LVCMOS	2 mA	25	27
	4 mA	25	27
	6 mA	51	54
	8 mA	51	54
	12 mA	103	109
	16 mA	132	127
	24 mA	268	181

Note: * $T_J = 100^{\circ}\text{C}$

Table 2-106 • 3.3 V LVTTTL / 3.3 V LVCMOS Low Slew
Commercial Temperature Range Conditions: $T_J = 70^\circ\text{C}$, Worst-Case $V_{CC} = 1.425\text{ V}$,
Worst-Case $V_{CCI} = 3.0\text{ V}$
Applicable to Advanced I/Os

Drive Strength	Speed Grade	t_{DOUT}	t_{DP}	t_{DIN}	t_{PY}	t_{EOUT}	t_{ZL}	t_{ZH}	t_{LZ}	t_{HZ}	t_{ZLS}	t_{ZHS}	Units
4 mA	Std.	0.66	10.26	0.04	1.20	0.43	10.45	8.90	2.64	2.46	12.68	11.13	ns
	–1	0.56	8.72	0.04	1.02	0.36	8.89	7.57	2.25	2.09	10.79	9.47	ns
	–2	0.49	7.66	0.03	0.90	0.32	7.80	6.64	1.98	1.83	9.47	8.31	ns
8 mA	Std.	0.66	7.27	0.04	1.20	0.43	7.41	6.28	2.98	3.04	9.65	8.52	ns
	–1	0.56	6.19	0.04	1.02	0.36	6.30	5.35	2.54	2.59	8.20	7.25	ns
	–2	0.49	5.43	0.03	0.90	0.32	5.53	4.69	2.23	2.27	7.20	6.36	ns
12 mA	Std.	0.66	5.58	0.04	1.20	0.43	5.68	4.87	3.21	3.42	7.92	7.11	ns
	–1	0.56	4.75	0.04	1.02	0.36	4.84	4.14	2.73	2.91	6.74	6.05	ns
	–2	0.49	4.17	0.03	0.90	0.32	4.24	3.64	2.39	2.55	5.91	5.31	ns
16 mA	Std.	0.66	5.21	0.04	1.20	0.43	5.30	4.56	3.26	3.51	7.54	6.80	ns
	–1	0.56	4.43	0.04	1.02	0.36	4.51	3.88	2.77	2.99	6.41	5.79	ns
	–2	0.49	3.89	0.03	0.90	0.32	3.96	3.41	2.43	2.62	5.63	5.08	ns
24 mA	Std.	0.66	4.85	0.04	1.20	0.43	4.94	4.54	3.32	3.88	7.18	6.78	ns
	–1	0.56	4.13	0.04	1.02	0.36	4.20	3.87	2.82	3.30	6.10	5.77	ns
	–2	0.49	3.62	0.03	0.90	0.32	3.69	3.39	2.48	2.90	5.36	5.06	ns

Note: For the derating values at specific junction temperature and voltage supply levels, refer to [Table 3-7 on page 3-9](#).

Table 2-107 • 3.3 V LVTTTL / 3.3 V LVCMOS High Slew

Commercial Temperature Range Conditions: $T_J = 70^\circ\text{C}$, Worst-Case $V_{CC} = 1.425\text{ V}$,
Worst-Case $V_{CCI} = 3.0\text{ V}$
Applicable to Advanced I/Os

Drive Strength	Speed Grade	t_{DOUT}	t_{DP}	t_{DIN}	t_{PY}	t_{EOUT}	t_{ZL}	t_{ZH}	t_{LZ}	t_{HZ}	t_{ZLS}	t_{ZHS}	Units
4 mA	Std.	0.66	7.66	0.04	1.20	0.43	7.80	6.59	2.65	2.61	10.03	8.82	ns
	–1	0.56	6.51	0.04	1.02	0.36	6.63	5.60	2.25	2.22	8.54	7.51	ns
	–2	0.49	5.72	0.03	0.90	0.32	5.82	4.92	1.98	1.95	7.49	6.59	ns
8 mA	Std.	0.66	4.91	0.04	1.20	0.43	5.00	4.07	2.99	3.20	7.23	6.31	ns
	–1	0.56	4.17	0.04	1.02	0.36	4.25	3.46	2.54	2.73	6.15	5.36	ns
	–2	0.49	3.66	0.03	0.90	0.32	3.73	3.04	2.23	2.39	5.40	4.71	ns
12 mA	Std.	0.66	3.53	0.04	1.20	0.43	3.60	2.82	3.21	3.58	5.83	5.06	ns
	–1	0.56	3.00	0.04	1.02	0.36	3.06	2.40	2.73	3.05	4.96	4.30	ns
	–2	0.49	2.64	0.03	0.90	0.32	2.69	2.11	2.40	2.68	4.36	3.78	ns
16 mA	Std.	0.66	3.33	0.04	1.20	0.43	3.39	2.56	3.26	3.68	5.63	4.80	ns
	–1	0.56	2.83	0.04	1.02	0.36	2.89	2.18	2.77	3.13	4.79	4.08	ns
	–2	0.49	2.49	0.03	0.90	0.32	2.53	1.91	2.44	2.75	4.20	3.58	ns
24 mA	Std.	0.66	3.08	0.04	1.20	0.43	3.13	2.12	3.32	4.06	5.37	4.35	ns
	–1	0.56	2.62	0.04	1.02	0.36	2.66	1.80	2.83	3.45	4.57	3.70	ns
	–2	0.49	2.30	0.03	0.90	0.32	2.34	1.58	2.48	3.03	4.01	3.25	ns

Note: For the derating values at specific junction temperature and voltage supply levels, refer to [Table 3-7 on page 3-9](#).

Table 2-108 • 3.3 V LVTTTL / 3.3 V LVCMOS Low Slew

Commercial Temperature Range Conditions: $T_J = 70^\circ\text{C}$, Worst-Case $V_{CC} = 1.425\text{ V}$,
Worst-Case $V_{CCI} = 3.0\text{ V}$
Applicable to Standard I/Os

Drive Strength	Speed Grade	t_{DOUT}	t_{DP}	t_{DIN}	t_{PY}	t_{EOUT}	t_{ZL}	t_{ZH}	t_{LZ}	t_{HZ}	Units
2 mA	Std.	0.66	9.46	0.04	1.00	0.43	9.64	8.54	2.07	2.04	ns
	–1	0.56	8.05	0.04	0.85	0.36	8.20	7.27	1.76	1.73	ns
	–2	0.49	7.07	0.03	0.75	0.32	7.20	6.38	1.55	1.52	ns
4 mA	Std.	0.66	9.46	0.04	1.00	0.43	9.64	8.54	2.07	2.04	ns
	–1	0.56	8.05	0.04	0.85	0.36	8.20	7.27	1.76	1.73	ns
	–2	0.49	7.07	0.03	0.75	0.32	7.20	6.38	1.55	1.52	ns
6 mA	Std.	0.66	6.57	0.04	1.00	0.43	6.69	5.98	2.40	2.57	ns
	–1	0.56	5.59	0.04	0.85	0.36	5.69	5.09	2.04	2.19	ns
	–2	0.49	4.91	0.03	0.75	0.32	5.00	4.47	1.79	1.92	ns
8 mA	Std.	0.66	6.57	0.04	1.00	0.43	6.69	5.98	2.40	2.57	ns
	–1	0.56	5.59	0.04	0.85	0.36	5.69	5.09	2.04	2.19	ns
	–2	0.49	4.91	0.03	0.75	0.32	5.00	4.47	1.79	1.92	ns

Note: For the derating values at specific junction temperature and voltage supply levels, refer to [Table 3-7 on page 3-9](#).

Table 2-169 • AC Waveforms, Measuring Points, and Capacitive Loads

Input Low (V)	Input High (V)	Measuring Point* (V)	VREF (typ.) (V)
1.075	1.325	Cross point	—

Note: *Measuring point = V_{trip} . See [Table 2-90 on page 2-166](#) for a complete table of trip points.

Timing Characteristics

Table 2-170 • LVDS

Commercial Temperature Range Conditions: $T_J = 70^\circ\text{C}$, Worst-Case $V_{CC} = 1.425\text{ V}$,
Worst-Case $V_{CCI} = 2.3\text{ V}$
Applicable to Pro I/Os

Speed Grade	t_{DOUT}	t_{DP}	t_{DIN}	t_{PY}	Units
Std.	0.66	2.10	0.04	1.82	ns
–1	0.56	1.79	0.04	1.55	ns
–2	0.49	1.57	0.03	1.36	ns

Note: For the derating values at specific junction temperature and voltage supply levels, refer to [Table 3-7 on page 3-9](#).

BLVDS/M-LVDS

Bus LVDS (BLVDS) and Multipoint LVDS (M-LVDS) specifications extend the existing LVDS standard to high-performance multipoint bus applications. Multidrop and multipoint bus configurations can contain any combination of drivers, receivers, and transceivers. Microsemi LVDS drivers provide the higher drive current required by BLVDS and M-LVDS to accommodate the loading. The driver requires series terminations for better signal quality and to control voltage swing. Termination is also required at both ends of the bus, since the driver can be located anywhere on the bus. These configurations can be implemented using TRIBUF_LVDS and BIBUF_LVDS macros along with appropriate terminations. Multipoint designs using Microsemi LVDS macros can achieve up to 200 MHz with a maximum of 20 loads. A sample application is given in [Figure 2-135](#). The input and output buffer delays are available in the LVDS section in [Table 2-171](#).

Example: For a bus consisting of 20 equidistant loads, the following terminations provide the required differential voltage, in worst-case industrial operating conditions at the farthest receiver: $R_S = 60\ \Omega$ and $R_T = 70\ \Omega$, given $Z_0 = 50\ \Omega$ (2") and $Z_{stub} = 50\ \Omega$ (~1.5").

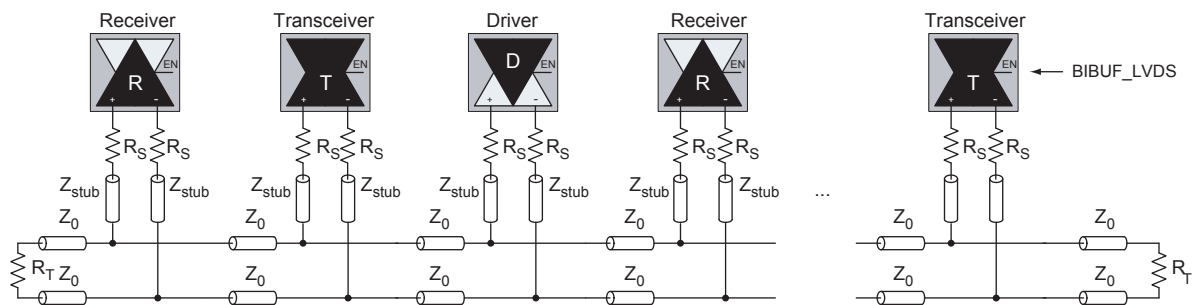

Figure 2-135 • BLVDS/M-LVDS Multipoint Application Using LVDS I/O Buffers

Table 2-175 • Parameter Definitions and Measuring Nodes

Parameter Name	Parameter Definition	Measuring Nodes (from, to)*
t_{OCLKQ}	Clock-to-Q of the Output Data Register	HH, DOUT
t_{OSUD}	Data Setup Time for the Output Data Register	FF, HH
t_{OHD}	Data Hold Time for the Output Data Register	FF, HH
t_{OSUE}	Enable Setup Time for the Output Data Register	GG, HH
t_{OHE}	Enable Hold Time for the Output Data Register	GG, HH
t_{OCLR2Q}	Asynchronous Clear-to-Q of the Output Data Register	LL, DOUT
$t_{OREMCLR}$	Asynchronous Clear Removal Time for the Output Data Register	LL, HH
$t_{ORECCLR}$	Asynchronous Clear Recovery Time for the Output Data Register	LL, HH
t_{OECLKQ}	Clock-to-Q of the Output Enable Register	HH, EOUT
t_{OESUD}	Data Setup Time for the Output Enable Register	JJ, HH
t_{OEHD}	Data Hold Time for the Output Enable Register	JJ, HH
t_{OESUE}	Enable Setup Time for the Output Enable Register	KK, HH
t_{OEHE}	Enable Hold Time for the Output Enable Register	KK, HH
$t_{OECLR2Q}$	Asynchronous Clear-to-Q of the Output Enable Register	II, EOUT
$t_{OEREMCLR}$	Asynchronous Clear Removal Time for the Output Enable Register	II, HH
$t_{OERECCLR}$	Asynchronous Clear Recovery Time for the Output Enable Register	II, HH
t_{ICLKQ}	Clock-to-Q of the Input Data Register	AA, EE
t_{ISUD}	Data Setup Time for the Input Data Register	CC, AA
t_{IHD}	Data Hold Time for the Input Data Register	CC, AA
t_{ISUE}	Enable Setup Time for the Input Data Register	BB, AA
t_{IHE}	Enable Hold Time for the Input Data Register	BB, AA
t_{ICLR2Q}	Asynchronous Clear-to-Q of the Input Data Register	DD, EE
$t_{IREMCLR}$	Asynchronous Clear Removal Time for the Input Data Register	DD, AA
$t_{IRECCLR}$	Asynchronous Clear Recovery Time for the Input Data Register	DD, AA

Note: *See Figure 2-138 on page 2-214 for more information.

3 – DC and Power Characteristics

General Specifications

Operating Conditions

Stresses beyond those listed in [Table 3-1](#) may cause permanent damage to the device.

Exposure to absolute maximum rated conditions for extended periods may affect device reliability. Devices should not be operated outside the recommended operating ranges specified in [Table 3-2](#) on [page 3-3](#).

Table 3-1 • Absolute Maximum Ratings

Symbol	Parameter	Commercial	Industrial	Units
VCC	DC core supply voltage	–0.3 to 1.65	–0.3 to 1.65	V
VJTAG	JTAG DC voltage	–0.3 to 3.75	–0.3 to 3.75	V
VPUMP	Programming voltage	–0.3 to 3.75	–0.3 to 3.75	V
VCCPLL	Analog power supply (PLL)	–0.3 to 1.65	–0.3 to 1.65	V
VCCI	DC I/O output buffer supply voltage	–0.3 to 3.75	–0.3 to 3.75	V
VI	I/O input voltage ¹	–0.3 V to 3.6 V (when I/O hot insertion mode is enabled) –0.3 V to (VCCI + 1 V) or 3.6 V, whichever voltage is lower (when I/O hot-insertion mode is disabled)		V
VCC33A	+3.3 V power supply	–0.3 to 3.75 ²	–0.3 to 3.75 ²	V
VCC33PMP	+3.3 V power supply	–0.3 to 3.75 ²	–0.3 to 3.75 ²	V
VAREF	Voltage reference for ADC	–0.3 to 3.75	–0.3 to 3.75	V
VCC15A	Digital power supply for the analog system	–0.3 to 1.65	–0.3 to 1.65	V
VCCNVM	Embedded flash power supply	–0.3 to 1.65	–0.3 to 1.65	V
VCCOSC	Oscillator power supply	–0.3 to 3.75	–0.3 to 3.75	V

Notes:

1. The device should be operated within the limits specified by the datasheet. During transitions, the input signal may undershoot or overshoot according to the limits shown in [Table 3-4](#) on [page 3-4](#).
2. Analog data not valid beyond 3.65 V.
3. The high current mode has a maximum power limit of 20 mW. Appropriate current limit resistors must be used, based on voltage on the pad.
4. For flash programming and retention maximum limits, refer to [Table 3-5](#) on [page 3-5](#). For recommended operating limits refer to [Table 3-2](#) on [page 3-3](#).

Table 3-3 • Input Resistance of Analog Pads

Pads	Pad Configuration	Prescaler Range	Input Resistance to Ground
AV, AC	Analog Input (direct input to ADC)	–	2 k Ω (typical)
		–	> 10 M Ω
	Analog Input (positive prescaler)	+16 V to +2 V	1 M Ω (typical)
		+1 V to +0.125 V	> 10 M Ω
	Analog Input (negative prescaler)	–16 V to –2 V	1 M Ω (typical)
		–1 V to –0.125 V	> 10 M Ω
	Digital input	+16 V to +2 V	1 M Ω (typical)
	Current monitor	+16 V to +2 V	1 M Ω (typical)
		–16 V to –2 V	1 M Ω (typical)
AT	Analog Input (direct input to ADC)	–	1 M Ω (typical)
	Analog Input (positive prescaler)	+16 V, +4 V	1 M Ω (typical)
	Digital input	+16 V, +4 V	1 M Ω (typical)
	Temperature monitor	+16 V, +4 V	> 10 M Ω

Table 3-4 • Overshoot and Undershoot Limits ¹

VCCI	Average VCCI–GND Overshoot or Undershoot Duration as a Percentage of Clock Cycle ²	Maximum Overshoot/Undershoot ²
2.7 V or less	10%	1.4 V
	5%	1.49 V
3.0 V	10%	1.1 V
	5%	1.19 V
3.3 V	10%	0.79 V
	5%	0.88 V
3.6 V	10%	0.45 V
	5%	0.54 V

Notes:

1. Based on reliability requirements at a junction temperature of 85°C.
2. The duration is allowed at one cycle out of six clock cycle. If the overshoot/undershoot occurs at one out of two cycles, the maximum overshoot/undershoot has to be reduced by 0.15 V.

5 – Datasheet Information

List of Changes

The following table lists critical changes that were made in each revision of the Fusion datasheet.

Revision	Changes	Page
Revision 6 (March 2014)	Note added for the discontinuance of QN108 and QN180 packages to the "Package I/Os: Single-/Double-Ended (Analog)" table and the "Temperature Grade Offerings" table (SAR 55113, PDN 1306).	II and IV
	Updated details about page programming time in the "Program Operation" section (SAR 49291).	2-46
	ADC_START changed to ADCSTART in the "ADC Operation" section (SAR 44104).	2-104
Revision 5 (January 2014)	Calibrated offset values (AFS090, AFS250) of the external temperature monitor in Table 2-49 • Analog Channel Specifications have been updated (SAR 51464).	2-117
	Specifications for the internal temperature monitor in Table 2-49 • Analog Channel Specifications have been updated (SAR 50870).	2-117
Revision 4 (January 2013)	The "Product Ordering Codes" section has been updated to mention "Y" as "Blank" mentioning "Device Does Not Include License to Implement IP Based on the Cryptography Research, Inc. (CRI) Patent Portfolio" (SAR 43177).	III
	The note in Table 2-12 • Fusion CCC/PLL Specification referring the reader to SmartGen was revised to refer instead to the online help associated with the core (SAR 42563).	2-28
	Table 2-49 • Analog Channel Specifications was modified to update the uncalibrated offset values (AFS250) of the external and internal temperature monitors (SAR 43134).	2-117
	In Table 2-57 • Prescaler Control Truth Table—AV (x = 0), AC (x = 1), and AT (x = 3), changed the column heading from 'Full-Scale Voltage' to 'Full Scale Voltage in 10-Bit Mode', and added and updated Notes as required (SAR 20812).	2-130
	The values for the Speed Grade (-1 and Std.) for FDDRIMAX (Table 2-180 • Input DDR Propagation Delays) and values for the Speed Grade (-2 and Std.) for FDDOMAX (Table 2-182 • Output DDR Propagation Delays) had been inadvertently interchanged. This has been rectified (SAR 38514).	2-220, 2-222
	Added description about what happens if a user connects VAREF to an external 3.3 V on their board to the "VAREF Analog Reference Voltage" section (SAR 35188).	2-225
	Added a note to Table 3-2 • Recommended Operating Conditions ¹ (SAR 43429): The programming temperature range supported is T _{ambient} = 0°C to 85°C.	3-3
	Added the Package Thermal details for AFS600-PQ208 and AFS250-PQ208 to Table 3-6 • Package Thermal Resistance (SAR 37816). Deleted the Die Size column from the table (SAR 43503).	3-7
	Libero Integrated Design Environment (IDE) was changed to Libero System-on-Chip (SoC) throughout the document (SAR 42495). Live at Power-Up (LAPU) has been replaced with 'Instant On'.	NA
Revision 3 (August 2012)	Microblade U1AFS250 and U1AFS1500 devices were added to the product tables.	I – IV
	A sentence pertaining to the analog I/Os was added to the "Specifying I/O States During Programming" section (SAR 34831).	1-9