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Understanding Embedded - FPGAs (Field Programmable Gate Array)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details	
Product Status	Active
Number of LABs/CLBs	-
Number of Logic Elements/Cells	-
Total RAM Bits	276480
Number of I/O	119
Number of Gates	1500000
Voltage - Supply	1.425V ~ 1.575V
Mounting Type	Surface Mount
Operating Temperature	-40°C ~ 100°C (TJ)
Package / Case	256-LBGA
Supplier Device Package	256-FPBGA (17x17)
Purchase URL	https://www.e-xfl.com/product-detail/microchip-technology/u1afs1500-fg256i

Advanced Architecture

The proprietary Fusion architecture provides granularity comparable to standard-cell ASICs. The Fusion device consists of several distinct and programmable architectural features, including the following (Figure 1-1 on page 1-5):

- Embedded memories
  - Flash memory blocks
  - FlashROM
  - SRAM and FIFO
- Clocking resources
  - PLL and CCC
  - RC oscillator
  - Crystal oscillator
  - No-Glitch MUX (NGMUX)
- Digital I/Os with advanced I/O standards
- FPGA VersaTiles
- Analog components
  - ADC
  - Analog I/Os supporting voltage, current, and temperature monitoring
  - 1.5 V on-board voltage regulator
  - Real-time counter

The FPGA core consists of a sea of VersaTiles. Each VersaTile can be configured as a three-input logic lookup table (LUT) equivalent or a D-flip-flop or latch (with or without enable) by programming the appropriate flash switch interconnections. This versatility allows efficient use of the FPGA fabric. The VersaTile capability is unique to the Microsemi families of flash-based FPGAs. VersaTiles and larger functions are connected with any of the four levels of routing hierarchy. Flash switches are distributed throughout the device to provide nonvolatile, reconfigurable interconnect programming. Maximum core utilization is possible for virtually any design.

In addition, extensive on-chip programming circuitry allows for rapid (3.3 V) single-voltage programming of Fusion devices via an IEEE 1532 JTAG interface.

Unprecedented Integration

Integrated Analog Blocks and Analog I/Os

Fusion devices offer robust and flexible analog mixed signal capability in addition to the high-performance flash FPGA fabric and flash memory block. The many built-in analog peripherals include a configurable 32:1 input analog MUX, up to 10 independent MOSFET gate driver outputs, and a configurable ADC. The ADC supports 8-, 10-, and 12-bit modes of operation with a cumulative sample rate up to 600 k samples per second (Ksps), differential nonlinearity (DNL) < 1.0 LSB, and Total Unadjusted Error (TUE) of 0.72 LSB in 10-bit mode. The TUE is used for characterization of the conversion error and includes errors from all sources, such as offset and linearity. Internal bandgap circuitry offers 1% voltage reference accuracy with the flexibility of utilizing an external reference voltage. The ADC channel sampling sequence and sampling rate are programmable and implemented in the FPGA logic using Designer and Libero SoC software tool support.

Two channels of the 32-channel ADCMUX are dedicated. Channel 0 is connected internally to VCC and can be used to monitor core power supply. Channel 31 is connected to an internal temperature diode which can be used to monitor device temperature. The 30 remaining channels can be connected to external analog signals. The exact number of I/Os available for external connection signals is device-dependent (refer to the "Fusion Family" table on page I for details).

Embedded Memories

Flash Memory Blocks

The flash memory available in each Fusion device is composed of one to four flash blocks, each 2 Mbits in density. Each block operates independently with a dedicated flash controller and interface. Fusion flash memory blocks combine fast access times (60 ns random access and 10 ns access in Read-Ahead mode) with a configurable 8-, 16-, or 32-bit datapath, enabling high-speed flash operation without wait states. The memory block is organized in pages and sectors. Each page has 128 bytes, with 33 pages comprising one sector and 64 sectors per block. The flash block can support multiple partitions. The only constraint on size is that partition boundaries must coincide with page boundaries. The flexibility and granularity enable many use models and allow added granularity in programming updates.

Fusion devices support two methods of external access to the flash memory blocks. The first method is a serial interface that features a built-in JTAG-compliant port, which allows in-system programmability during user or monitor/test modes. This serial interface supports programming of an AES-encrypted stream. Data protected with security measures can be passed through the JTAG interface, decrypted, and then programmed in the flash block. The second method is a soft parallel interface.

FPGA logic or an on-chip soft microprocessor can access flash memory through the parallel interface. Since the flash parallel interface is implemented in the FPGA fabric, it can potentially be customized to meet special user requirements. For more information, refer to the *CoreCFI Handbook.* The flash memory parallel interface provides configurable byte-wide (×8), word-wide (×16), or dual-word-wide (×32) data-port options. Through the programmable flash parallel interface, the on-chip and off-chip memories can be cascaded for wider or deeper configurations.

The flash memory has built-in security. The user can configure either the entire flash block or the small blocks to protect against unintentional or intrusive attempts to change or destroy the storage contents. Each on-chip flash memory block has a dedicated controller, enabling each block to operate independently.

The flash block logic consists of the following sub-blocks:

- Flash block – Contains all stored data. The flash block contains 64 sectors and each sector contains 33 pages of data.
- Page Buffer – Contains the contents of the current page being modified. A page contains 8 blocks of data.
- Block Buffer – Contains the contents of the last block accessed. A block contains 128 data bits.
- ECC Logic – The flash memory stores error correction information with each block to perform single-bit error correction and double-bit error detection on all data blocks.

User Nonvolatile FlashROM

In addition to the flash blocks, Fusion devices have 1 Kbit of user-accessible, nonvolatile FlashROM on-chip. The FlashROM is organized as 8×128-bit pages. The FlashROM can be used in diverse system applications:

- Internet protocol addressing (wireless or fixed)
- System calibration settings
- Device serialization and/or inventory control
- Subscription-based business models (for example, set-top boxes)
- Secure key storage for communications algorithms protected by security
- Asset management/tracking
- Date stamping
- Version management

The FlashROM is written using the standard IEEE 1532 JTAG programming interface. Pages can be individually programmed (erased and written). On-chip AES decryption can be used selectively over public networks to load data such as security keys stored in the FlashROM for a user design.

The FlashROM can be programmed (erased and written) via the JTAG programming interface, and its contents can be read back either through the JTAG programming interface or via direct FPGA core addressing.

RAM4K9 Description

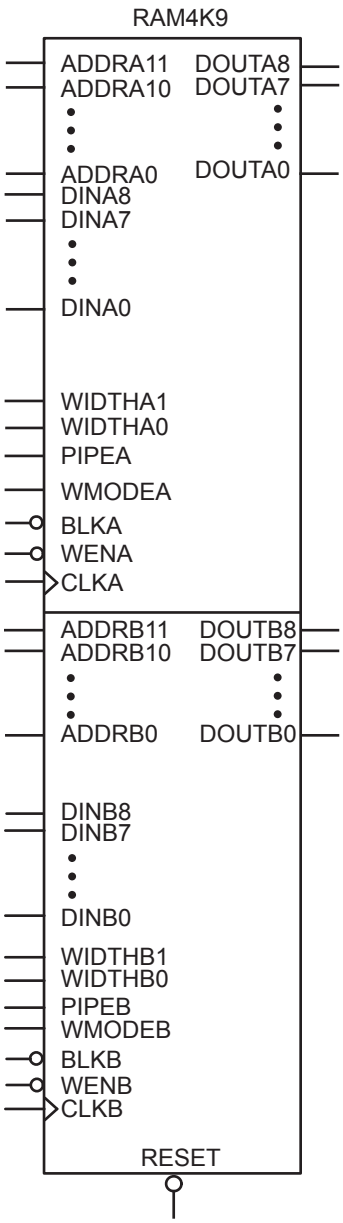

Figure 2-48 • **RAM4K9**

Modes of Operation

There are two read modes and one write mode:

- Read Nonpipelined (synchronous—1 clock edge): In the standard read mode, new data is driven onto the RD bus in the same clock cycle following RA and REN valid. The read address is registered on the read port clock active edge, and data appears at RD after the RAM access time. Setting PIPE to OFF enables this mode.
- Read Pipelined (synchronous—2 clock edges): The pipelined mode incurs an additional clock delay from the address to the data but enables operation at a much higher frequency. The read address is registered on the read port active clock edge, and the read data is registered and appears at RD after the second read clock edge. Setting PIPE to ON enables this mode.
- Write (synchronous—1 clock edge): On the write clock active edge, the write data is written into the SRAM at the write address when WEN is High. The setup times of the write address, write enables, and write data are minimal with respect to the write clock. Write and read transfers are described with timing requirements in the "SRAM Characteristics" section on page 2-63 and the "FIFO Characteristics" section on page 2-72.

RAM Initialization

Each SRAM block can be individually initialized on power-up by means of the JTAG port using the UJTAG mechanism (refer to the "JTAG IEEE 1532" section on page 2-229 and the *Fusion SRAM/FIFO Blocks* application note). The shift register for a target block can be selected and loaded with the proper bit configuration to enable serial loading. The 4,608 bits of data can be loaded in a single operation.

Table 2-36 describes each pin in the Analog Block. Each function within the Analog Block will be explained in detail in the following sections.

***Table 2-36* • Analog Block Pin Description**

Signal Name	Number of Bits	Direction	Function	Location of Details
VAREF	1	Input/Output	Voltage reference for ADC	ADC
ADCGNDREF	1	Input	External ground reference	ADC
MODE[3:0]	4	Input	ADC operating mode	ADC
SYSCLK	1	Input	External system clock	
TVC[7:0]	8	Input	Clock divide control	ADC
STC[7:0]	8	Input	Sample time control	ADC
ADCSTART	1	Input	Start of conversion	ADC
PWRDWN	1	Input	ADC comparator power-down if 1. When asserted, the ADC will stop functioning, and the digital portion of the analog block will continue operating. This may result in invalid status flags from the analog block. Therefore, Microsemi does not recommend asserting the PWRDWN pin.	ADC
ADCRESET	1	Input	ADC resets and disables Analog Quad – active high	ADC
BUSY	1	Output	1 – Running conversion	ADC
CALIBRATE	1	Output	1 – Power-up calibration	ADC
DATAVALID	1	Output	1 – Valid conversion result	ADC
RESULT[11:0]	12	Output	Conversion result	ADC
TMSTBINT	1	Input	Internal temp. monitor strobe	ADC
SAMPLE	1	Output	1 – An analog signal is actively being sampled (stays high during signal acquisition only) 0 – No analog signal is being sampled	ADC
VAREFSEL	1	Input	0 = Output internal voltage reference (2.56 V) to VAREF 1 = Input external voltage reference from VAREF and ADCGNDREF	ADC
CHNUMBER[4:0]	5	Input	Analog input channel select	Input multiplexer
ACMCLK	1	Input	ACM clock	ACM
ACMWEN	1	Input	ACM write enable – active high	ACM
ACMRESET	1	Input	ACM reset – active low	ACM
ACMWDATA[7:0]	8	Input	ACM write data	ACM
ACMRDATA[7:0]	8	Output	ACM read data	ACM
ACMADDR[7:0]	8	Input	ACM address	ACM
CMSTB0 to CMSTB9	10	Input	Current monitor strobe – 1 per quad, active high	Analog Quad

***Table 2-36* • Analog Block Pin Description (continued)**

Signal Name	Number of Bits	Direction	Function	Location of Details
GDON0 to GDON9	10	Input	Control to power MOS – 1 per quad	Analog Quad
TMSTB0 to TMSTB9	10	Input	Temperature monitor strobe – 1 per quad; active high	Analog Quad
DAVOUT0, DACOUT0, DATOUT0 to DAVOUT9, DACOUT9, DATOUT9	30	Output	Digital outputs – 3 per quad	Analog Quad
DENAV0, DENAC0, DENAT0 to DENAV9, DENAC9, DENAT9	30	Input	Digital input enables – 3 per quad	Analog Quad
AV0	1	Input	Analog Quad 0	Analog Quad
AC0	1	Input		Analog Quad
AG0	1	Output		Analog Quad
AT0	1	Input		Analog Quad
ATRETURN01	1	Input	Temperature monitor return shared by Analog Quads 0 and 1	Analog Quad
AV1	1	Input	Analog Quad 1	Analog Quad
AC1	1	Input		Analog Quad
AG1	1	Output		Analog Quad
AT1	1	Input		Analog Quad
AV2	1	Input	Analog Quad 2	Analog Quad
AC2	1	Input		Analog Quad
AG2	1	Output		Analog Quad
AT2	1	Input		Analog Quad
ATRETURN23	1	Input	Temperature monitor return shared by Analog Quads 2 and 3	Analog Quad
AV3	1	Input	Analog Quad 3	Analog Quad
AC3	1	Input		Analog Quad
AG3	1	Output		Analog Quad
AT3	1	Input		Analog Quad
AV4	1	Input	Analog Quad 4	Analog Quad
AC4	1	Input		Analog Quad
AG4	1	Output		Analog Quad
AT4	1	Input		Analog Quad
ATRETURN45	1	Input	Temperature monitor return shared by Analog Quads 4 and 5	Analog Quad
AV5	1	Input	Analog Quad 5	Analog Quad
AC5	1	Input		Analog Quad
AG5	1	Output		Analog Quad
AT5	1	Input		Analog Quad
AV6	1	Input	Analog Quad 6	Analog Quad
AC6	1	Input		Analog Quad

Channel Input Offset Error

Channel Offset error is measured as the input voltage that causes the transition from zero to a count of one. An Ideal Prescaler will have offset equal to ½ of LSB voltage. Offset error is a positive or negative when the first transition point is higher or lower than ideal. Offset error is expressed in LSB or input voltage.

Total Channel Error

Total Channel Error is defined as the total error measured compared to the ideal value. Total Channel Error is the sum of gain error and offset error combined. Figure 2-68 shows how Total Channel Error is measured.

Total Channel Error is defined as the difference between the actual ADC output and ideal ADC output. In the example shown in Figure 2-68, the Total Channel Error would be a negative number.

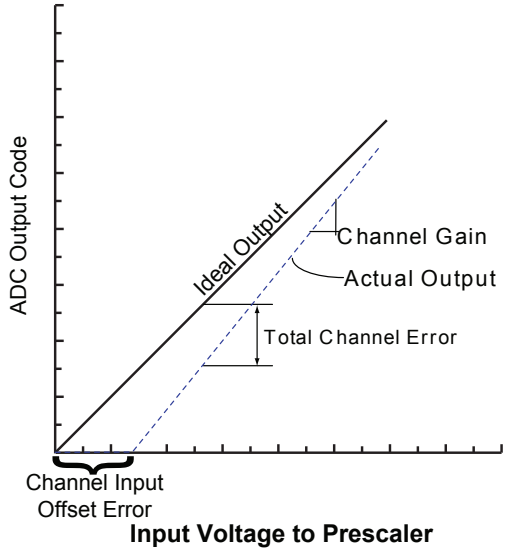

Figure 2-68 • **Total Channel Error Example**

Current Monitor

The Fusion Analog Quad is an excellent element for voltage- and current-monitoring applications. In addition to supporting the same functionality offered by the AV pad, the AC pad can be configured to monitor current across an external sense resistor (Figure 2-70). To support this current monitor function, a differential amplifier with 10x gain passes the amplified voltage drop between the AV and AC pads to the ADC. The amplifier enables the user to use very small resistor values, thereby limiting any impact on the circuit. This function of the AC pad does not limit AV pad operation. The AV pad can still be configured for use as a direct voltage input or scaled through the AV prescaler independently of it's use as an input to the AC pad's differential amplifier.

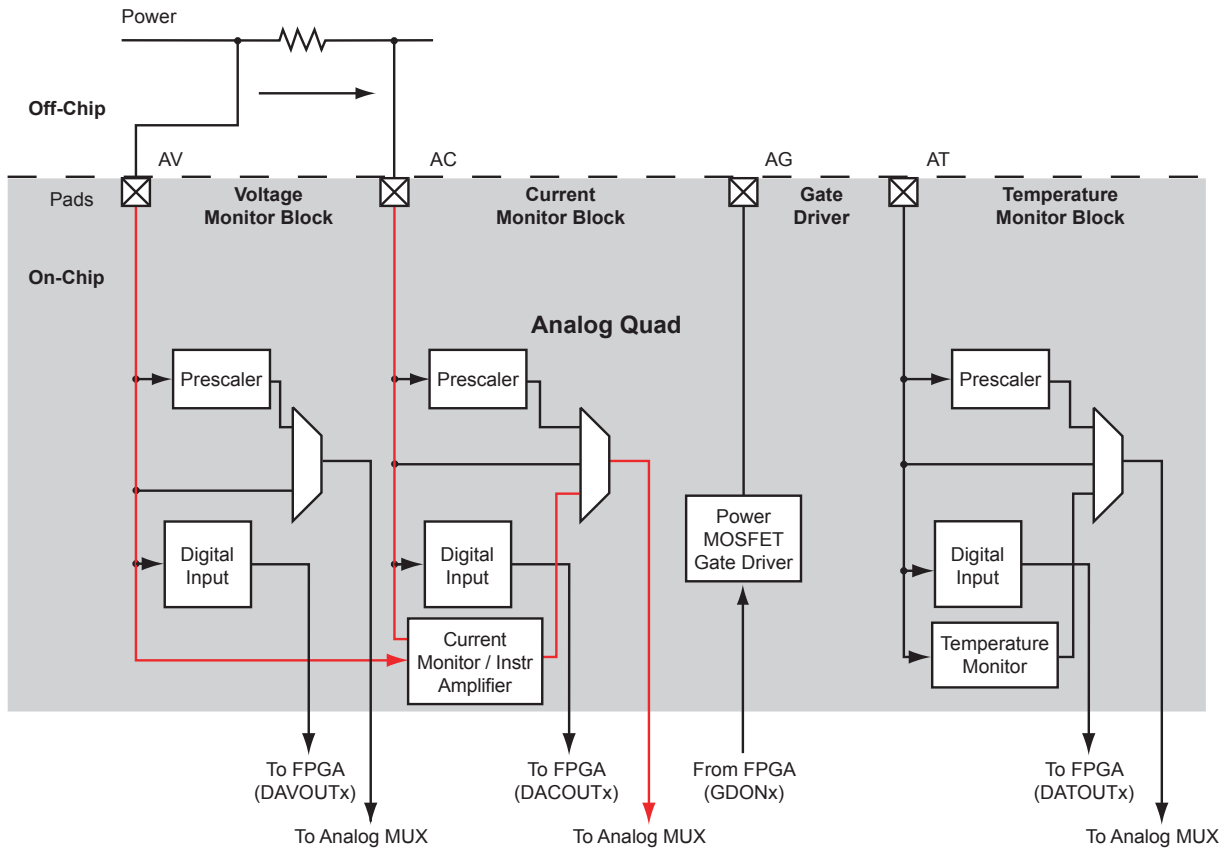

Figure 2-70 • **Analog Quad Current Monitor Configuration**

Analog-to-Digital Converter Block

At the heart of the Fusion analog system is a programmable Successive Approximation Register (SAR) ADC. The ADC can support 8-, 10-, or 12-bit modes of operation. In 12-bit mode, the ADC can resolve 500 ksps. All results are MSB-justified in the ADC. The input to the ADC is a large 32:1 analog input multiplexer. A simplified block diagram of the Analog Quads, analog input multiplexer, and ADC is shown in Figure 2-79. The ADC offers multiple self-calibrating modes to ensure consistent high performance both at power-up and during runtime.

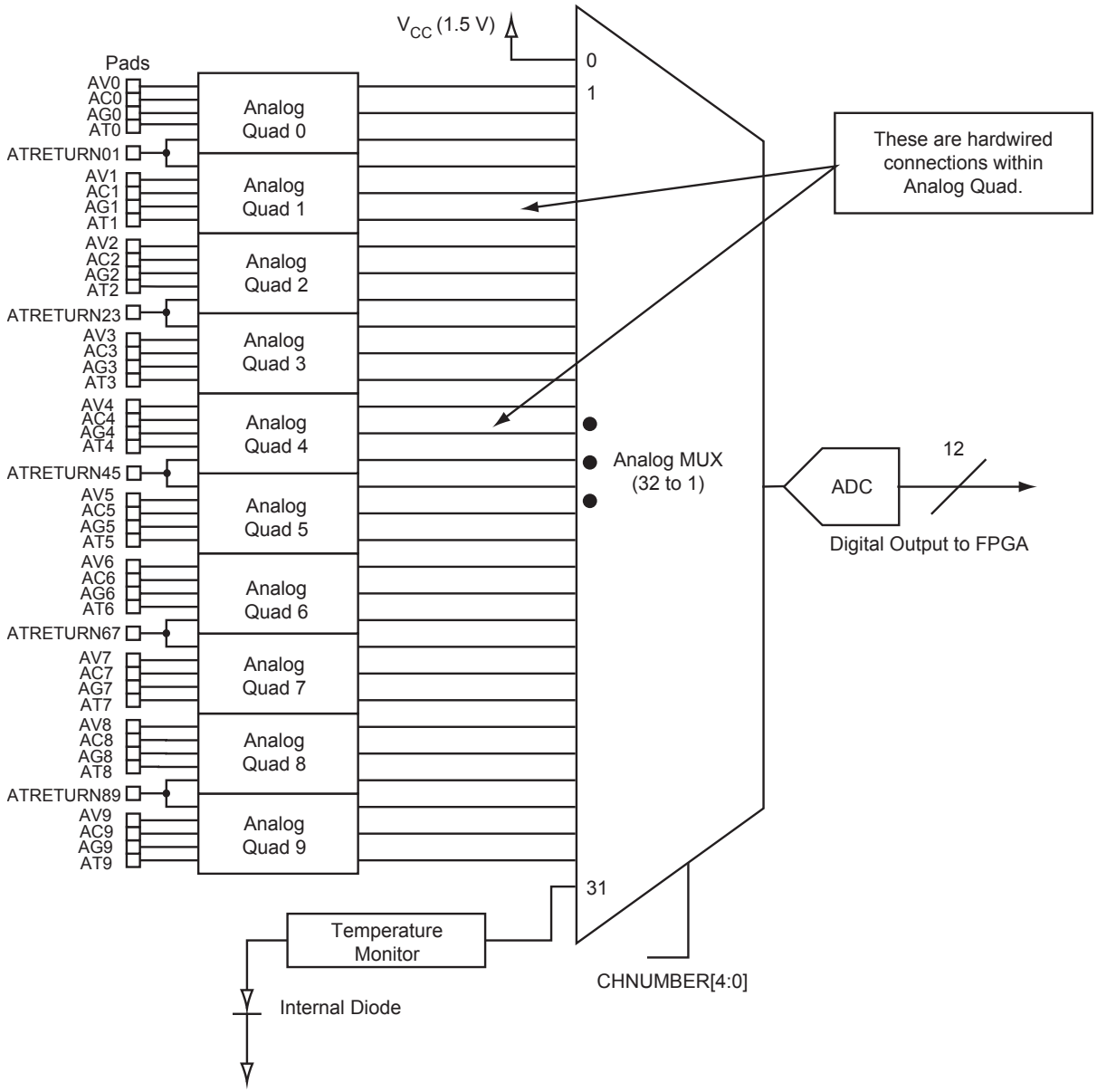

Figure 2-79 • **ADC Block Diagram**

ADC Terminology

Conversion Time

Conversion time is the interval between the release of the hold state (imposed by the input circuitry of a track-and-hold) and the instant at which the voltage on the sampling capacitor settles to within one LSB of a new input value.

DNL – Differential Non-Linearity

For an ideal ADC, the analog-input levels that trigger any two successive output codes should differ by one LSB (DNL = 0). Any deviation from one LSB in defined as DNL (Figure 2-83).

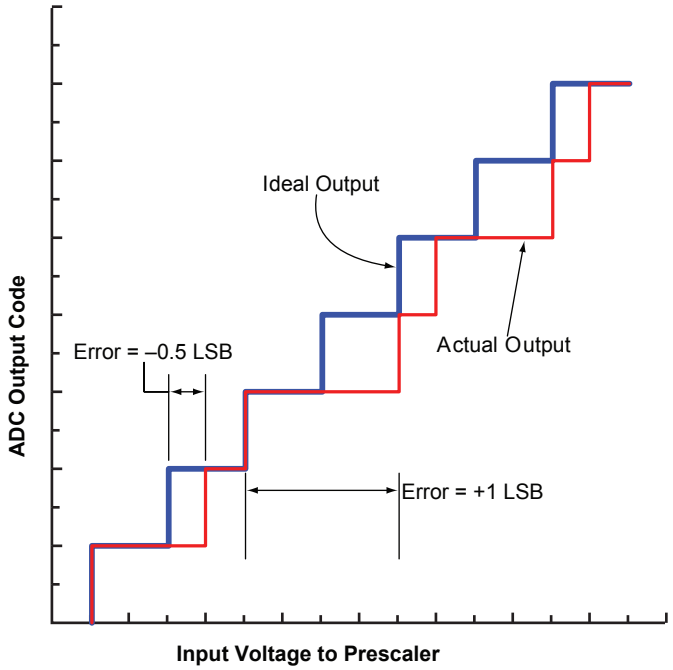

Figure 2-83 • **Differential Non-Linearity (DNL)**

ENOB – Effective Number of Bits

ENOB specifies the dynamic performance of an ADC at a specific input frequency and sampling rate. An ideal ADC's error consists only of quantization of noise. As the input frequency increases, the overall noise (particularly in the distortion components) also increases, thereby reducing the ENOB and SINAD (also see "Signal-to-Noise and Distortion Ratio (SINAD)".) ENOB for a full-scale, sinusoidal input waveform is computed using EQ 12.

$$ENOB = \frac{SINAD - 1.76}{6.02}$$

EQ 12

FS Error – Full-Scale Error

Full-scale error is the difference between the actual value that triggers that transition to full-scale and the ideal analog full-scale transition value. Full-scale error equals offset error plus gain error.

Analog System Characteristics

***Table 2-49* • Analog Channel Specifications**
Commercial Temperature Range Conditions, T_J = 85°C (unless noted otherwise),
Typical: VCC33A = 3.3 V, VCC = 1.5 V

Parameter	Description	Condition	Min.	Typ.	Max.	Units
Voltage Monitor Using Analog Pads AV, AC and AT (using prescaler)						
	Input Voltage (Prescaler)	Refer to Table 3-2 on page 3-3				
VINAP	Uncalibrated Gain and Offset Errors	Refer to Table 2-51 on page 2-122				
	Calibrated Gain and Offset Errors	Refer to Table 2-52 on page 2-123				
	Bandwidth1				100	KHz
	Input Resistance	Refer to Table 3-3 on page 3-4				
	Scaling Factor	Prescaler modes (Table 2-57 on page 2-130)				
	Sample Time		10			µs
Current Monitor Using Analog Pads AV and AC						
VRSM[1]	Maximum Differential Input Voltage				VAREF / 10	mV
	Resolution	Refer to "Current Monitor" section				
	Common Mode Range				– 10.5 to +12	V
CMRR	Common Mode Rejection Ratio	DC – 1 KHz		60		dB
		1 KHz - 10 KHz		50		dB
		> 10 KHz		30		dB
t_{CMSHI}	Strobe High time		ADC conv. time		200	µs
t_{CMSHI}	Strobe Low time		5			µs
t_{CMSHI}	Settling time		0.02			µs
	Accuracy	Input differential voltage > 50 mV			–2 –(0.05 x VRSM) to +2 + (0.05 x VRSM)	mV

Notes:

1. *VRSM is the maximum voltage drop across the current sense resistor.*
2. *Analog inputs used as digital inputs can tolerate the same voltage limits as the corresponding analog pad. There is no reliability concern on digital inputs as long as VIND does not exceed these limits.*
3. *VIND is limited to VCC33A + 0.2 to allow reaching 10 MHz input frequency.*
4. *An averaging of 1,024 samples (LPF setting in Analog System Builder) is required and the maximum capacitance allowed across the AT pins is 500 pF.*
5. *The temperature offset is a fixed positive value.*
6. *The high current mode has a maximum power limit of 20 mW. Appropriate current limit resistors must be used, based on voltage on the pad.*
7. *When using SmartGen Analog System Builder, CalibIP is required to obtain specified offset. For further details on CalibIP, refer to the "Temperature, Voltage, and Current Calibration in Fusion FPGAs" chapter of the* Fusion FPGA Fabric User Guide.

Analog Configuration MUX

The ACM is the interface between the FPGA, the Analog Block configurations, and the real-time counter. Microsemi Libero SoC will generate IP that will load and configure the Analog Block via the ACM. However, users are not limited to using the Libero SoC IP. This section provides a detailed description of the ACM's register map, truth tables for proper configuration of the Analog Block and RTC, as well as timing waveforms so users can access and control the ACM directly from their designs.

The Analog Block contains four 8-bit latches per Analog Quad that are initialized through the ACM. These latches act as configuration bits for Analog Quads. The ACM block runs from the core voltage supply (1.5 V).

Access to the ACM is achieved via 8-bit address and data busses with enables. The pin list is provided in Table 2-36 on page 2-78. The ACM clock speed is limited to a maximum of 10 MHz, more than sufficient to handle the low-bandwidth requirements of configuring the Analog Block and the RTC (sub-block of the Analog Block).

Table 2-54 decodes the ACM address space and maps it to the corresponding Analog Quad and configuration byte for that quad.

Table 2-54 • **ACM Address Decode Table for Analog Quad**

ACMADDR [7:0] in Decimal	Name	Description	Associated Peripheral
0	–	–	Analog Quad
1	AQ0	Byte 0	Analog Quad
2	AQ0	Byte 1	Analog Quad
3	AQ0	Byte 2	Analog Quad
4	AQ0	Byte 3	Analog Quad
5	AQ1	Byte 0	Analog Quad
. . .	. . .	. . .	Analog Quad
36	AQ8	Byte 3	Analog Quad
37	AQ9	Byte 0	Analog Quad
38	AQ9	Byte 1	Analog Quad
39	AQ9	Byte 2	Analog Quad
40	AQ9	Byte 3	Analog Quad
41		Undefined	Analog Quad
. . .	. . .	Undefined	Analog Quad
63		Undefined	RTC
64	COUNTER0	Counter bits 7:0	RTC
65	COUNTER1	Counter bits 15:8	RTC
66	COUNTER2	Counter bits 23:16	RTC
67	COUNTER3	Counter bits 31:24	RTC
68	COUNTER4	Counter bits 39:32	RTC
72	MATCHREG0	Match register bits 7:0	RTC

***Table 2-68* • I/O Bank Support by Device**

I/O Bank	AFS090	AFS250	AFS600	AFS1500
Standard I/O	N	N	–	–
Advanced I/O	E, W	E, W	E, W	E, W
Pro I/O	–	–	N	N
Analog Quad	S	S	S	S

Note: E = East side of the device
W = West side of the device
N = North side of the device
S = South side of the device

***Table 2-69* • Fusion VCCI Voltages and Compatible Standards**

VCCI (typical)	Compatible Standards
3.3 V	LVTTL/LVCMOS 3.3, PCI 3.3, SSTL3 (Class I and II),* GTL+ 3.3, GTL 3.3,* LVPECL
2.5 V	LVCMOS 2.5, LVCMOS 2.5/5.0, SSTL2 (Class I and II),* GTL+ 2.5,* GTL 2.5,* LVDS, BLVDS, M-LVDS
1.8 V	LVCMOS 1.8
1.5 V	LVCMOS 1.5, HSTL (Class I),* HSTL (Class II)*

*Note: *I/O standard supported by Pro I/O banks.*

Table 2-70* • Fusion VREF Voltages and Compatible Standards

VREF (typical)	Compatible Standards
1.5 V	SSTL3 (Class I and II)
1.25 V	SSTL2 (Class I and II)
1.0 V	GTL+ 2.5, GTL+ 3.3
0.8 V	GTL 2.5, GTL 3.3
0.75 V	HSTL (Class I), HSTL (Class II)

*Note: *I/O standards supported by Pro I/O banks.*

User I/O Naming Convention

Due to the comprehensive and flexible nature of Fusion device user I/Os, a naming scheme is used to show the details of the I/O (Figure 2-113 on page 2-158 and Figure 2-114 on page 2-159). The name identifies to which I/O bank it belongs, as well as the pairing and pin polarity for differential I/Os.

I/O Nomenclature = Gmn/IOuxwByVz

Gmn is only used for I/Os that also have CCC access—i.e., global pins.

G = Global

m = Global pin location associated with each CCC on the device: A (northwest corner), B (northeast corner), C (east middle), D (southeast corner), E (southwest corner), and F (west middle).

n = Global input MUX and pin number of the associated Global location m, either A0, A1, A2, B0, B1, B2, C0, C1, or C2. Figure 2-22 on page 2-25 shows the three input pins per clock source MUX at CCC location m.

u = I/O pair number in the bank, starting at 00 from the northwest I/O bank and proceeding in a clockwise direction.

x = P (Positive) or N (Negative) for differential pairs, or R (Regular – single-ended) for the I/Os that support single-ended and voltage-referenced I/O standards only. U (Positive-LVDS only) or V (Negative-LVDS only) restrict the I/O differential pair from being selected as an LVPECL pair.

w = D (Differential Pair), P (Pair), or S (Single-Ended). D (Differential Pair) if both members of the pair are bonded out to adjacent pins or are separated only by one GND or NC pin; P (Pair) if both members of the pair are bonded out but do not meet the adjacency requirement; or S (Single-Ended) if the I/O pair is not bonded out. For Differential (D) pairs, adjacency for ball grid packages means only vertical or horizontal. Diagonal adjacency does not meet the requirements for a true differential pair.

B = Bank

y = Bank number (0–3). The Bank number starts at 0 from the northwest I/O bank and proceeds in a clockwise direction.

V = Reference voltage

z = Minibank number

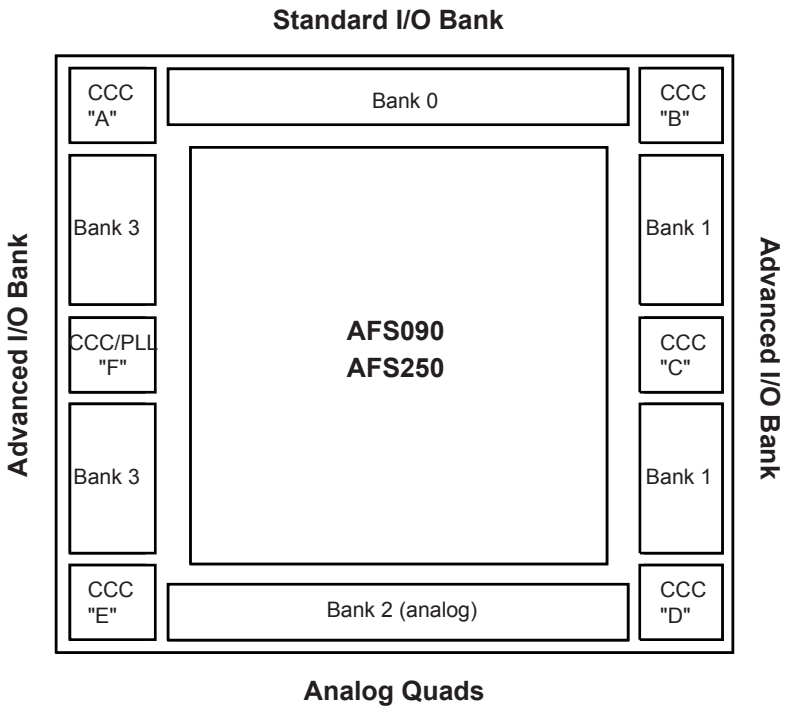

***Figure 2-113* • Naming Conventions of Fusion Devices with Three Digital I/O Banks**

Table 2-115 • 2.5 V LVCMOS High Slew
Commercial Temperature Range Conditions: T_J = 70°C, Worst-Case VCC = 1.425 V, Worst-Case VCCI = 2.3 V
Applicable to Advanced I/Os

Drive Strength	Speed Grade	t_{DOUT}	t_{DP}	t_{DIN}	t_{PY}	t_{EOUT}	t_{ZL}	t_{ZH}	t_{LZ}	t_{HZ}	t_{ZLS}	t_{ZHS}	Units
4 mA	Std.	0.66	8.66	0.04	1.31	0.43	7.83	8.66	2.68	2.30	10.07	10.90	ns
	–1	0.56	7.37	0.04	1.11	0.36	6.66	7.37	2.28	1.96	8.56	9.27	ns
	–2	0.49	6.47	0.03	0.98	0.32	5.85	6.47	2.00	1.72	7.52	8.14	ns
8 mA	Std.	0.66	5.17	0.04	1.31	0.43	5.04	5.17	3.05	3.00	7.27	7.40	ns
	–1	0.56	4.39	0.04	1.11	0.36	4.28	4.39	2.59	2.55	6.19	6.30	ns
	–2	0.49	3.86	0.03	0.98	0.32	3.76	3.86	2.28	2.24	5.43	5.53	ns
12 mA	Std.	0.66	3.56	0.04	1.31	0.43	3.63	3.43	3.30	3.44	5.86	5.67	ns
	–1	0.56	3.03	0.04	1.11	0.36	3.08	2.92	2.81	2.92	4.99	4.82	ns
	–2	0.49	2.66	0.03	0.98	0.32	2.71	2.56	2.47	2.57	4.38	4.23	ns
16 mA	Std.	0.66	3.35	0.04	1.31	0.43	3.41	3.06	3.36	3.55	5.65	5.30	ns
	–1	0.56	2.85	0.04	1.11	0.36	2.90	2.60	2.86	3.02	4.81	4.51	ns
	–2	0.49	2.50	0.03	0.98	0.32	2.55	2.29	2.51	2.65	4.22	3.96	ns
24 mA	Std.	0.66	3.56	0.04	1.31	0.43	3.63	3.43	3.30	3.44	5.86	5.67	ns
	–1	0.56	3.03	0.04	1.11	0.36	3.08	2.92	2.81	2.92	4.99	4.82	ns
	–2	0.49	2.66	0.03	0.98	0.32	2.71	2.56	2.47	2.57	4.38	4.23	ns

Note: For the derating values at specific junction temperature and voltage supply levels, refer to Table 3-7 on page 3-9.

Table 2-116 • 2.5 V LVCMOS Low Slew
Commercial Temperature Range Conditions: T_J = 70°C, Worst-Case VCC = 1.425 V, Worst-Case VCCI = 2.3 V
Applicable to Standard I/Os

Drive Strength	Speed Grade	t_{DOUT}	t_{DP}	t_{DIN}	t_{PY}	t_{EOUT}	t_{ZL}	t_{ZH}	t_{LZ}	t_{HZ}	Units
2 mA	Std.	0.66	11.00	0.04	1.29	0.43	10.37	11.00	2.03	1.83	ns
	–1	0.56	9.35	0.04	1.10	0.36	8.83	9.35	1.73	1.56	ns
	–2	0.49	8.21	0.03	0.96	0.32	7.75	8.21	1.52	1.37	ns
4 mA	Std.	0.66	11.00	0.04	1.29	0.43	10.37	11.00	2.03	1.83	ns
	–1	0.56	9.35	0.04	1.10	0.36	8.83	9.35	1.73	1.56	ns
	–2	0.49	8.21	0.03	0.96	0.32	7.75	8.21	1.52	1.37	ns
6 mA	Std.	0.66	7.50	0.04	1.29	0.43	7.36	7.50	2.39	2.46	ns
	–1	0.56	6.38	0.04	1.10	0.36	6.26	6.38	2.03	2.10	ns
	–2	0.49	5.60	0.03	0.96	0.32	5.49	5.60	1.78	1.84	ns
8 mA	Std.	0.66	7.50	0.04	1.29	0.43	7.36	7.50	2.39	2.46	ns
	–1	0.56	6.38	0.04	1.10	0.36	6.26	6.38	2.03	2.10	ns
	–2	0.49	5.60	0.03	0.96	0.32	5.49	5.60	1.78	1.84	ns

Note: For the derating values at specific junction temperature and voltage supply levels, refer to Table 3-7 on page 3-9.

Voltage Referenced I/O Characteristics

3.3 V GTL

Gunning Transceiver Logic is a high-speed bus standard (JESD8-3). It provides a differential amplifier input buffer and an open-drain output buffer. The VCCI pin should be connected to 3.3 V.

Table 2-138 • Minimum and Maximum DC Input and Output Levels

3.3 V GTL	VIL		VIH		VOL	VOH	IOL	IOH	IOSL	IOSH	IIL[1]	IIH[2]
Drive Strength	Min. V	Max. V	Min. V	Max. V	Max. V	Min. V	mA	mA	Max. mA[3]	Max. mA[3]	µA[4]	µA[4]
20 mA[3]	–0.3	VREF – 0.05	VREF + 0.05	3.6	0.4	–	20	20	181	268	10	10

Notes:

1. *IIL is the input leakage current per I/O pin over recommended operation conditions where –0.3 V < VIN < VIL.*
2. *IIH is the input leakage current per I/O pin over recommended operating conditions VIH < VIN < VCCI. Input current is larger when operating outside recommended ranges.*
3. *Currents are measured at high temperature (100°C junction temperature) and maximum voltage.*
4. *Currents are measured at 85°C junction temperature.*

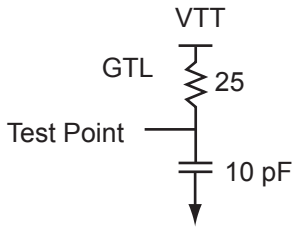

Figure 2-124 • AC Loading

Table 2-139 • AC Waveforms, Measuring Points, and Capacitive Loads

Input Low (V)	Input High (V)	Measuring Point* (V)	VREF (typ.) (V)	VTT (typ.) (V)	C_{LOAD} (pF)
VREF – 0.05	VREF + 0.05	0.8	0.8	1.2	10

*Note: *Measuring point = Vtrip. See Table 2-90 on page 2-166 for a complete table of trip points.*

Timing Characteristics

Table 2-140 • 3.3 V GTL
Commercial Temperature Range Conditions: T_J = 70°C, Worst-Case VCC = 1.425 V, Worst-Case VCCI = 3.0 V, VREF = 0.8 V

Speed Grade	t_{DOUT}	t_{DP}	t_{DIN}	t_{PY}	t_{EOUT}	t_{ZL}	t_{ZH}	t_{LZ}	t_{HZ}	t_{ZLS}	t_{ZHS}	Units
Std.	0.66	2.08	0.04	2.93	0.43	2.04	2.08			4.27	4.31	ns
–1	0.56	1.77	0.04	2.50	0.36	1.73	1.77			3.63	3.67	ns
–2	0.49	1.55	0.03	2.19	0.32	1.52	1.55			3.19	3.22	ns

Note: For the derating values at specific junction temperature and voltage supply levels, refer to Table 3-7 on page 3-9.

TMS **Test Mode Select**

The TMS pin controls the use of the IEEE1532 boundary scan pins (TCK, TDI, TDO, TRST). There is an internal weak pull-up resistor on the TMS pin.

TRST **Boundary Scan Reset Pin**

The TRST pin functions as an active low input to asynchronously initialize (or reset) the boundary scan circuitry. There is an internal weak pull-up resistor on the TRST pin. If JTAG is not used, an external pull-down resistor could be included to ensure the TAP is held in reset mode. The resistor values must be chosen from Table 2-183 and must satisfy the parallel resistance value requirement. The values in Table 2-183 correspond to the resistor recommended when a single device is used and to the equivalent parallel resistor when multiple devices are connected via a JTAG chain.

In critical applications, an upset in the JTAG circuit could allow entering an undesired JTAG state. In such cases, Microsemi recommends tying off TRST to GND through a resistor placed close to the FPGA pin.

Note that to operate at all VJTAG voltages, 500 Ω to 1 kΩ will satisfy the requirements.

Special Function Pins

NC **No Connect**

This pin is not connected to circuitry within the device. These pins can be driven to any voltage or can be left floating with no effect on the operation of the device.

DC **Don't Connect**

This pin should not be connected to any signals on the PCB. These pins should be left unconnected.

NCAP **Negative Capacitor**

Negative Capacitor is where the negative terminal of the charge pump capacitor is connected. A capacitor, with a 2.2 µF recommended value, is required to connect between PCAP and NCAP.

PCAP **Positive Capacitor**

Positive Capacitor is where the positive terminal of the charge pump capacitor is connected. A capacitor, with a 2.2 µF recommended value, is required to connect between PCAP and NCAP.

PUB **Push Button**

Push button is the connection for the external momentary switch used to turn on the 1.5 V voltage regulator and can be floating if not used.

PTBASE **Pass Transistor Base**

Pass Transistor Base is the control signal of the voltage regulator. This pin should be connected to the base of the external pass transistor used with the 1.5 V internal voltage regulator and can be floating if not used.

PTEM **Pass Transistor Emitter**

Pass Transistor Emitter is the feedback input of the voltage regulator.

This pin should be connected to the emitter of the external pass transistor used with the 1.5 V internal voltage regulator and can be floating if not used.

XTAL1 **Crystal Oscillator Circuit Input**

Input to crystal oscillator circuit. Pin for connecting external crystal, ceramic resonator, RC network, or external clock input. When using an external crystal or ceramic oscillator, external capacitors are also recommended (Please refer to the crystal oscillator manufacturer for proper capacitor value).

If using external RC network or clock input, XTAL1 should be used and XTAL2 left unconnected. In the case where the Crystal Oscillator block is not used, the XTAL1 pin should be connected to GND and the XTAL2 pin should be left floating.

$P_{S\text{-}CELL} = N_{S\text{-}CELL} * (PAC5 + (\alpha_1 / 2) * PAC6) * F_{CLK}$

$N_{S\text{-}CELL}$ is the number of VersaTiles used as sequential modules in the design. When a multi-tile sequential cell is used, it should be accounted for as 1.

α_1 is the toggle rate of VersaTile outputs—guidelines are provided in Table 3-16 on page 3-27.

F_{CLK} is the global clock signal frequency.

Standby Mode and Sleep Mode

$P_{S\text{-}CELL} = 0$ W

Combinatorial Cells Dynamic Contribution—$P_{C\text{-}CELL}$

Operating Mode

$P_{C\text{-}CELL} = N_{C\text{-}CELL} * (\alpha_1 / 2) * PAC7 * F_{CLK}$

$N_{C\text{-}CELL}$ is the number of VersaTiles used as combinatorial modules in the design.

α_1 is the toggle rate of VersaTile outputs—guidelines are provided in Table 3-16 on page 3-27.

F_{CLK} is the global clock signal frequency.

Standby Mode and Sleep Mode

$P_{C\text{-}CELL} = 0$ W

Routing Net Dynamic Contribution—P_{NET}

Operating Mode

$P_{NET} = (N_{S\text{-}CELL} + N_{C\text{-}CELL}) * (\alpha_1 / 2) * PAC8 * F_{CLK}$

$N_{S\text{-}CELL}$ is the number VersaTiles used as sequential modules in the design.

$N_{C\text{-}CELL}$ is the number of VersaTiles used as combinatorial modules in the design.

α_1 is the toggle rate of VersaTile outputs—guidelines are provided in Table 3-16 on page 3-27.

F_{CLK} is the global clock signal frequency.

Standby Mode and Sleep Mode

$P_{NET} = 0$ W

I/O Input Buffer Dynamic Contribution—P_{INPUTS}

Operating Mode

$P_{INPUTS} = N_{INPUTS} * (\alpha_2 / 2) * PAC9 * F_{CLK}$

N_{INPUTS} is the number of I/O input buffers used in the design.

α_2 is the I/O buffer toggle rate—guidelines are provided in Table 3-16 on page 3-27.

F_{CLK} is the global clock signal frequency.

Standby Mode and Sleep Mode

$P_{INPUTS} = 0$ W

I/O Output Buffer Dynamic Contribution—$P_{OUTPUTS}$

Operating Mode

$P_{OUTPUTS} = N_{OUTPUTS} * (\alpha_2 / 2) * \beta_1 * PAC10 * F_{CLK}$

$N_{OUTPUTS}$ is the number of I/O output buffers used in the design.

α_2 is the I/O buffer toggle rate—guidelines are provided in Table 3-16 on page 3-27.

β_1 is the I/O buffer enable rate—guidelines are provided in Table 3-17 on page 3-27.

F_{CLK} is the global clock signal frequency.

Standby Mode and Sleep Mode

$P_{OUTPUTS} = 0$ W

FG484		
Pin Number	AFS600 Function	AFS1500 Function
H13	GND	GND
H14	VCCIB1	VCCIB1
H15	GND	GND
H16	GND	GND
H17	NC	IO53NDB2V0
H18	IO38PDB2V0	IO57PDB2V0
H19	GCA2/IO39PDB2V0	GCA2/IO59PDB2V0
H20	VCCIB2	VCCIB2
H21	IO37NDB2V0	IO54NDB2V0
H22	IO37PDB2V0	IO54PDB2V0
J1	NC	IO112PPB4V0
J2	IO76NDB4V0	IO113NDB4V0
J3	GFB2/IO74PDB4V0	GFB2/IO109PDB4V0
J4	GFA2/IO75PDB4V0	GFA2/IO110PDB4V0
J5	NC	IO112NPB4V0
J6	NC	IO104PDB4V0
J7	NC	IO111PDB4V0
J8	VCCIB4	VCCIB4
J9	GND	GND
J10	VCC	VCC
J11	GND	GND
J12	VCC	VCC
J13	GND	GND
J14	VCC	VCC
J15	VCCIB2	VCCIB2
J16	GCB2/IO40PDB2V0	GCB2/IO60PDB2V0
J17	NC	IO58NDB2V0
J18	IO38NDB2V0	IO57NDB2V0
J19	IO39NDB2V0	IO59NDB2V0
J20	GCC2/IO41PDB2V0	GCC2/IO61PDB2V0
J21	NC	IO55PSB2V0
J22	IO42PDB2V0	IO56PDB2V0
K1	GFC2/IO73PDB4V0	GFC2/IO108PDB4V0
K2	GND	GND
K3	IO74NDB4V0	IO109NDB4V0

FG484		
Pin Number	AFS600 Function	AFS1500 Function
K4	IO75NDB4V0	IO110NDB4V0
K5	GND	GND
K6	NC	IO104NDB4V0
K7	NC	IO111NDB4V0
K8	GND	GND
K9	VCC	VCC
K10	GND	GND
K11	VCC	VCC
K12	GND	GND
K13	VCC	VCC
K14	GND	GND
K15	GND	GND
K16	IO40NDB2V0	IO60NDB2V0
K17	NC	IO58PDB2V0
K18	GND	GND
K19	NC	IO68NPB2V0
K20	IO41NDB2V0	IO61NDB2V0
K21	GND	GND
K22	IO42NDB2V0	IO56NDB2V0
L1	IO73NDB4V0	IO108NDB4V0
L2	VCCOSC	VCCOSC
L3	VCCIB4	VCCIB4
L4	XTAL2	XTAL2
L5	GFC1/IO72PDB4V0	GFC1/IO107PDB4V0
L6	VCCIB4	VCCIB4
L7	GFB1/IO71PDB4V0	GFB1/IO106PDB4V0
L8	VCCIB4	VCCIB4
L9	GND	GND
L10	VCC	VCC
L11	GND	GND
L12	VCC	VCC
L13	GND	GND
L14	VCC	VCC
L15	VCCIB2	VCCIB2
L16	IO48PDB2V0	IO70PDB2V0

Revision	Changes	Page
v2.0, Revision 1 (continued)	Table 3-6 • Package Thermal Resistance was updated to include new data.	3-7
	In EQ 4 to EQ 6, the junction temperature was changed from 110°C to 100°C.	3-8 to 3-8
	Table 3-8 • AFS1500 Quiescent Supply Current Characteristics through Table 3-11 • AFS090 Quiescent Supply Current Characteristics are new and have replaced the Quiescent Supply Current Characteristics (IDDQ) table.	3-10 to 3-16
	In Table 3-14 • Different Components Contributing to the Dynamic Power Consumption in Fusion Devices, the power supply for PAC9 and PAC10 were changed from VMV/VCC to VCCI.	3-22
	In Table 3-15 • Different Components Contributing to the Static Power Consumption in Fusion Devices, the power supply for PDC7 and PDC8 were changed from VMV/VCC to VCCI. PDC1 was updated from TBD to 18.	3-23
	The "QN108" table was updated to remove the duplicates of pins B12 and B34.	4-2
Preliminary v1.7 (October 2008)	The version number category was changed from Advance to Preliminary, which means the datasheet contains information based on simulation and/or initial characterization. The information is believed to be correct, but changes are possible.	
	For the VIL and VIH parameters, 0.30 * VCCI was changed to 0.35 * VCCI and 0.70 * VCCI was changed to 0.65 * VCCI in Table 2-126 • Minimum and Maximum DC Input and Output Levels.	2-193
	The version number category was changed from Advance to Preliminary, which means the datasheet contains information based on simulation and/or initial characterization. The information is believed to be correct, but changes are possible.	N/A
	The following updates were made to Table 2-141 • Minimum and Maximum DC Input and Output Levels: Temperature Digital Output 213 00 1111 1101 283 01 0001 1011 358 01 0110 0110 – only the digital output was updated. Temperature 358 remains in the temperature column.	2-200
	In Advance v1.2, the "VAREF Analog Reference Voltage" pin description was significantly updated but the change was not noted in the change table.	2-225
Advance v1.6 (August 2008)	The title of the datasheet changed from Actel Programmable System Chips to Actel Fusion Mixed Signal FPGAs. In addition, all instances of programmable system chip were changed to mixed signal FPGA.	N/A
	The references to the *Peripherals User's Guide* in the "No-Glitch MUX (NGMUX)" section and "Voltage Regulator Power Supply Monitor (VRPSM)" section were changed to *Fusion Handbook*.	2-32, 2-42
Advance v1.5 (July 2008)	The following bullet was updated from High-Voltage Input Tolerance: ±12 V to High-Voltage Input Tolerance: 10.5 V to 12 V.	I
	The following bullet was updated from Programmable 1, 3, 10, 30 µA and 25 mA Drive Strengths to Programmable 1, 3, 10, 30 µA and 20 mA Drive Strengths.	I