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### What is "[Embedded - Microcontrollers](#)"?

"[Embedded - Microcontrollers](#)" refer to small, integrated circuits designed to perform specific tasks within larger systems. These microcontrollers are essentially compact computers on a single chip, containing a processor core, memory, and programmable input/output peripherals. They are called "embedded" because they are embedded within electronic devices to control various functions, rather than serving as standalone computers. Microcontrollers are crucial in modern electronics, providing the intelligence and control needed for a wide range of applications.

### Applications of "[Embedded - Microcontrollers](#)"

#### Details

Product Status	Active
Core Processor	AVR
Core Size	8/16-Bit
Speed	32MHz
Connectivity	EBI/EMI, I <sup>2</sup> C, IrDA, SPI, UART/USART
Peripherals	Brown-out Detect/Reset, DMA, POR, PWM, WDT
Number of I/O	78
Program Memory Size	64KB (32K x 16)
Program Memory Type	FLASH
EEPROM Size	2K x 8
RAM Size	4K x 8
Voltage - Supply (Vcc/Vdd)	1.6V ~ 3.6V
Data Converters	A/D 16x12b; D/A 4x12b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	100-TQFP
Supplier Device Package	100-TQFP (14x14)
Purchase URL	<a href="https://www.e-xfl.com/product-detail/microchip-technology/atxmega64a1-aur">https://www.e-xfl.com/product-detail/microchip-technology/atxmega64a1-aur</a>

### 8.3.4 Production Signature Row

The production signature row is a separate memory section for factory programmed data. It contains calibration data for functions such as oscillators and analog modules. Some of the calibration values will be automatically loaded to the corresponding module or peripheral unit during reset. Other values must be loaded from the signature row and written to the corresponding peripheral registers from software. For details on calibration conditions, refer to “Electrical Characteristics” on page 76.

The production signature row also contains an ID that identifies each microcontroller device type and a serial number for each manufactured device. The serial number consists of the production lot number, wafer number, and wafer coordinates for the device. The device ID for the available devices is shown in Table 8-1.

The production signature row cannot be written or erased, but it can be read from application software and external programmers.

**Table 8-1. Device ID bytes.**

Device	Device ID bytes		
	Byte 2	Byte 1	Byte 0
ATxmega64A1	4E	96	1E
ATxmega128A1	4C	97	1E

### 8.3.5 User Signature Row

The user signature row is a separate memory section that is fully accessible (read and write) from application software and external programmers. It is one flash page in size, and is meant for static user parameter storage, such as calibration data, custom serial number, identification numbers, random number seeds, etc. This section is not erased by chip erase commands that erase the flash, and requires a dedicated erase command. This ensures parameter storage during multiple program/erase operations and on-chip debug sessions.

## 8.4 Fuses and Lock bits

The fuses are used to configure important system functions, and can only be written from an external programmer. The application software can read the fuses. The fuses are used to configure reset sources such as brownout detector and watchdog, startup configuration, JTAG enable, and JTAG user ID.

The lock bits are used to set protection levels for the different flash sections (that is, if read and/or write access should be blocked). Lock bits can be written by external programmers and application software, but only to stricter protection levels. Chip erase is the only way to erase the lock bits. To ensure that flash contents are protected even during chip erase, the lock bits are erased after the rest of the flash memory has been erased.

An unprogrammed fuse or lock bit will have the value one, while a programmed fuse or lock bit will have the value zero. Both fuses and lock bits are reprogrammable like the flash program memory.

## 8.5 Data Memory

The data memory contains the I/O memory, internal SRAM, optionally memory mapped EEPROM, and external memory if available. The data memory is organized as one continuous memory section, see Figure 8-2 on page 15. To simplify development, I/O Memory, EEPROM and SRAM will always have the same start addresses for all Atmel AVR XMEGA devices. The address space for External Memory will always start at the end of Internal SRAM and end at address 0xFFFFF.

**Table 8-3. Number of Bytes and Pages in the EEPROM.**

Device	EEPROM	Page Size	E2BYTE	E2PAGE	No of pages
	Size	bytes			
ATxmega64A1	2 KB	32	ADDR[4:0]	ADDR[10:5]	64
ATxmega128A1	2 KB	32	ADDR[4:0]	ADDR[10:5]	64

#### 8.14.1 I/O Memory

All peripherals and modules are addressable through I/O memory locations in the data memory space. All I/O memory locations can be accessed by the Load (LD/LDS/LDD) and Store (ST/STS/STD) instructions, transferring data between the 32 general purpose registers in the CPU and the I/O Memory.

The IN and OUT instructions can address I/O memory locations in the range 0x00 - 0x3F directly.

I/O registers within the address range 0x00 - 0x1F are directly bit-accessible using the SBI and CBI instructions. The value of single bits can be checked by using the SBIS and SBIC instructions on these registers.

The I/O memory address for all peripherals and modules in XMEGA A1 is shown in the “Peripheral Module Address Map” on page 62.

## 14. Interrupts and Programmable Multilevel Interrupt Controller

### 14.1 Features

- Short and predictable interrupt response time
- Separate interrupt configuration and vector address for each interrupt
- Programmable multilevel interrupt controller
  - Interrupt prioritizing according to level and vector address
  - Three selectable interrupt levels for all interrupts: low, medium and high
  - Selectable, round-robin priority scheme within low-level interrupts
  - Non-maskable interrupts for critical functions
- Interrupt vectors optionally placed in the application section or the boot loader section

### 14.2 Overview

Interrupts signal a change of state in peripherals, and this can be used to alter program execution. Peripherals can have one or more interrupts, and all are individually enabled and configured. When an interrupt is enabled and configured, it will generate an interrupt request when the interrupt condition is present. The programmable multilevel interrupt controller (PMIC) controls the handling and prioritizing of interrupt requests. When an interrupt request is acknowledged by the PMIC, the program counter is set to point to the interrupt vector, and the interrupt handler can be executed.

All peripherals can select between three different priority levels for their interrupts: low, medium, and high. Interrupts are prioritized according to their level and their interrupt vector address. Medium-level interrupts will interrupt low-level interrupt handlers. High-level interrupts will interrupt both medium- and low-level interrupt handlers. Within each level, the interrupt priority is decided from the interrupt vector address, where the lowest interrupt vector address has the highest interrupt priority. Low-level interrupts have an optional round-robin scheduling scheme to ensure that all interrupts are serviced within a certain amount of time.

Non-maskable interrupts (NMI) are also supported, and can be used for system critical functions.

### 14.3 Interrupt vectors

The interrupt vector is the sum of the peripheral's base interrupt address and the offset address for specific interrupts in each peripheral. The base addresses for the Atmel AVR XMEGA A1U devices are shown in Table 14-1. Offset addresses for each interrupt available in the peripheral are described for each peripheral in the XMEGA AU manual. For peripherals or modules that have only one interrupt, the interrupt vector is shown in Table 14-1. The program address is the word address.

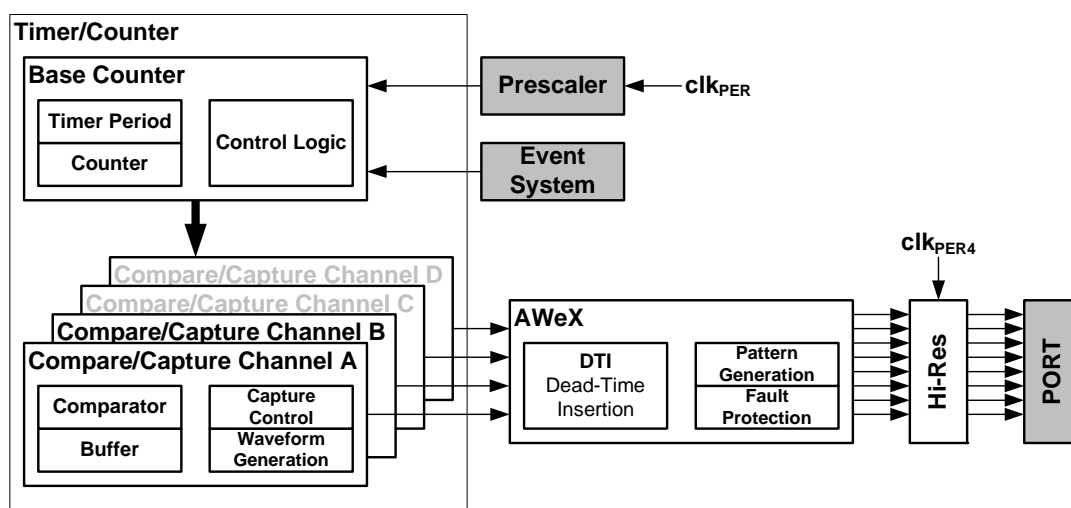
**Table 14-1. Reset and Interrupt vectors**

Program Address (Base Address)	Source	Interrupt Description
0x000	RESET	
0x002	OSCF_INT_vect	Crystal Oscillator Failure Interrupt vector (NMI)
0x004	PORTC_INT_base	Port C Interrupt base
0x008	PORTR_INT_base	Port R Interrupt base
0x00C	DMA_INT_base	DMA Controller Interrupt base
0x014	RTC_INT_base	Real Time Counter Interrupt base
0x018	TWIC_INT_base	Two-Wire Interface on Port C Interrupt base

Program Address (Base Address)	Source	Interrupt Description
0x0D8	TCF0_INT_base	Timer/Counter 0 on port F Interrupt base
0x0E4	TCF1_INT_base	Timer/Counter 1 on port F Interrupt base
0x0EC	SPIF_INT_vector	SPI on port F Interrupt base
0x0EE	USARTF0_INT_base	USART 0 on port F Interrupt base
0x0F4	USARTF1_INT_base	USART 1 on port F Interrupt base

The high-resolution (hi-res) extension can be used to increase the waveform output resolution by four or eight times by using an internal clock source running up to four times faster than the peripheral clock. See “Hi-Res - High Resolution Extension” on page 39 for more details.

**Figure 16-1. Overview of a Timer/Counter and closely related peripherals**



PORTC, PORTD, PORTE and PORTF each has one Timer/Counter 0 and one Timer/Counter1. Notation of these Timer/Counters are TCC0 (Time/Counter C0), TCC1, TCD0, TCD1, TCE0, TCE1, TCF0, and TCF1, respectively.

## 22. USART

### 22.1 Features

- Eight identical USART peripherals
- Full-duplex operation
- Asynchronous or synchronous operation
  - Synchronous clock rates up to 1/2 of the device clock frequency
  - Asynchronous clock rates up to 1/8 of the device clock frequency
- Supports serial frames with 5, 6, 7, 8, or 9 data bits and 1 or 2 stop bits
- Fractional baud rate generator
  - Can generate desired baud rate from any system clock frequency
  - No need for external oscillator with certain frequencies
- Built-in error detection and correction schemes
  - Odd or even parity generation and parity check
  - Data overrun and framing error detection
  - Noise filtering includes false start bit detection and digital low-pass filter
- Separate interrupts for
  - Transmit complete
  - Transmit data register empty
  - Receive complete
- Multiprocessor communication mode
  - Addressing scheme to address a specific devices on a multidevice bus
  - Enable unaddressed devices to automatically ignore all frames
- Master SPI mode
  - Double buffered operation
  - Operation up to 1/2 of the peripheral clock frequency
- IRCOM module for IrDA compliant pulse modulation/demodulation

### 22.2 Overview

The universal synchronous and asynchronous serial receiver and transmitter (USART) is a fast and flexible serial communication module. The USART supports full-duplex communication and asynchronous and synchronous operation. The USART can be configured to operate in SPI master mode and used for SPI communication.

Communication is frame based, and the frame format can be customized to support a wide range of standards. The USART is buffered in both directions, enabling continued data transmission without any delay between frames. Separate interrupts for receive and transmit complete enable fully interrupt driven communication. Frame error and buffer overflow are detected in hardware and indicated with separate status flags. Even or odd parity generation and parity check can also be enabled.

The clock generator includes a fractional baud rate generator that is able to generate a wide range of USART baud rates from any system clock frequencies. This removes the need to use an external crystal oscillator with a specific frequency to achieve a required baud rate. It also supports external clock input in synchronous slave operation.

When the USART is set in master SPI mode, all USART-specific logic is disabled, leaving the transmit and receive buffers, shift registers, and baud rate generator enabled. Pin control and interrupt generation are identical in both modes. The registers are used in both modes, but their functionality differs for some control settings.

An IRCOM module can be enabled for one USART to support IrDA 1.4 physical compliant pulse modulation and demodulation for baud rates up to 115.2Kbps.

PORTC, PORTD, PORTE, and PORTF each has two USARTs. Notation of these peripherals are USARTC0, USARTC1, USARTE0, USARTE1, USARTE0, USARTE1, USARTE0 and USARTE1.

## 23. IRCOM - IR Communication Module

### 23.1 Features

- Pulse modulation/demodulation for infrared communication
- IrDA compatible for baud rates up to 115.2Kbps
- Selectable pulse modulation scheme
  - 3/16 of the baud rate period
  - Fixed pulse period, 8-bit programmable
  - Pulse modulation disabled
- Built-in filtering
- Can be connected to and used by any USART

### 23.2 Overview

Atmel AVR XMEGA devices contain an infrared communication module (IRCOM) that is IrDA compatible for baud rates up to 115.2Kbps. It can be connected to any USART to enable infrared pulse encoding/decoding for that USART.



## 25. EBI – External Bus Interface

### 25.1 Features

- Supports SRAM up to:
  - 512KB using 3-port EBI configuration
  - 16MB using 3-port EBI configuration
- Supports SDRAM up to:
  - 128Mb using 3-port EBI configuration
- Four software configurable chip selects
- Software configurable wait state insertion
- Can run from the 2x peripheral clock frequency for fast access

### 25.2 Overview

The External Bus Interface (EBI) is used to connect external peripherals and memory for access through the data memory space. When the EBI is enabled, data address space outside the internal SRAM becomes available using dedicated EBI pins.

The EBI can interface external SRAM, SDRAM, and peripherals, such as LCD displays and other memory mapped devices.

The address space for the external memory is selectable from 256 bytes (8-bit) up to 16MB (24-bit). Various multiplexing modes for address and data lines can be selected for optimal use of pins when more or fewer pins are available for the EBI. The complete memory will be mapped into one linear data address space continuing from the end of the internal SRAM.

The EBI has four chip selects, each with separate configuration. Each can be configured for SRAM, SRAM low pin count (LPC), or SDRAM.

The EBI is clocked from the fast, 2x peripheral clock, running up to two times faster than the CPU.

Four-bit and eight-bit SDRAM are supported, and SDRAM configurations, such as CAS latency and refresh rate, are configurable in software.

## 27. DAC - 12-bit Digital to Analog Converter

### 27.1 Features

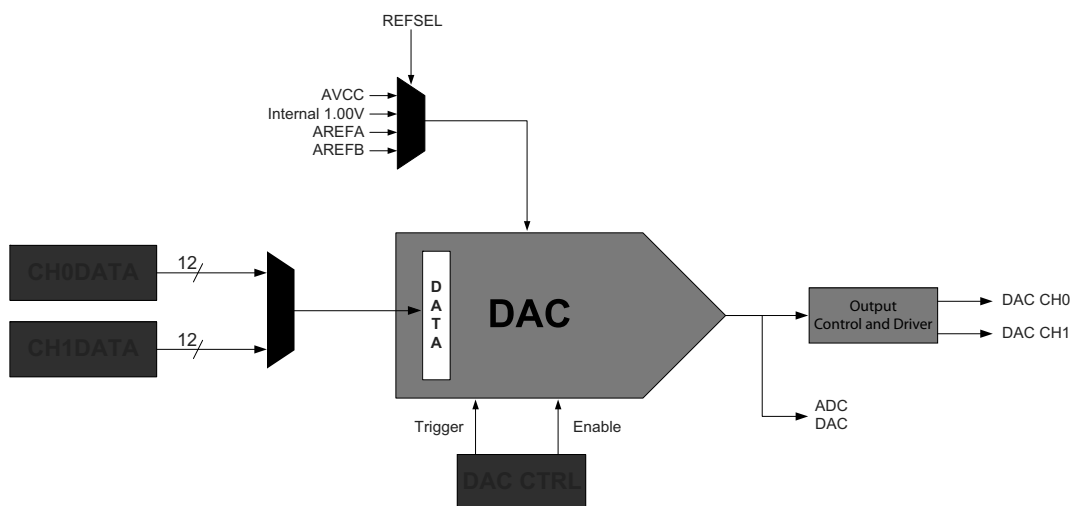
- 12-bit resolution
- Two independent, continuous-drive output channels
- Up to one million samples per second conversion rate
- Built-in calibration that removes:
  - Offset error
  - Gain error
- Multiple conversion trigger sources
  - On new available data
  - Events from the event system
- High drive capabilities and support for
  - Resistive loads
  - Capacitive loads
  - Combined resistive and capacitive loads
- Internal and external reference options
- DAC output available as input to analog comparator and ADC
- Low-power mode, with reduced drive strength
- Optional DMA transfer of data

### 27.2 Overview

The XMEGA A1 devices features two 12-bit, 1 Msps DACs with built-in offset and gain calibration, see Figure 27-1 on page 50.

A DAC converts a digital value into an analog signal. The DAC may use an internal 1.0V voltage as the upper limit for conversion, but it is also possible to use the supply voltage or any applied voltage in-between. The external reference input is shared with the ADC reference input.

Figure 27-1. DAC overview



## 30.2 Alternate Pin Functions

The tables below show the primary/default function for each pin on a port in the first column, the pin number in the second column, and then all alternate pin functions in the remaining columns. The head row shows what peripheral that enable and use the alternate pin functions.

**Table 30-1. Port A - Alternate functions.**

PORT A	PIN #	INTERRUPT	ADCA POS	ADCA NEG	ADCA GAINPOS	ADCA GAINNEG	ACA POS	ACA NEG	ACA OUT	DACA	REFA
GND	93										
AVCC	94										
PA0	95	SYNC	ADC0	ADC0	ADC0		AC0	AC0			AREF
PA1	96	SYNC	ADC1	ADC1	ADC1		AC1	AC1			
PA2	97	SYNC/ASYNC	ADC2	ADC2	ADC2		AC2			DAC0	
PA3	98	SYNC	ADC3	ADC3	ADC3		AC3	AC3		DAC1	
PA4	99	SYNC	ADC4		ADC4	ADC4	AC4				
PA5	100	SYNC	ADC5		ADC5	ADC5	AC5	AC5			
PA6	1	SYNC	ADC6		ADC6	ADC6	AC6				
PA7	2	SYNC	ADC7		ADC7	ADC7		AC7	AC0OUT		

**Table 30-2. Port B - Alternate functions.**

PORT B	PIN #	INTERRUPT	ADCB POS	ADCB NEG	ADCB GAINPOS	ADCB GAINNEG	ACB POS	ACB NEG	ACB OUT	DACB	REFB	JTAG
GND	3											
AVCC	4											
PB0	5	SYNC	ADC0	ADC0	ADC0		AC0	AC0			AREF	
PB1	6	SYNC	ADC1	ADC1	ADC1		AC1	AC1				
PB2	7	SYNC/ASYNC	ADC2	ADC2	ADC2		AC2			DAC0		
PB3	8	SYNC	ADC3	ADC3	ADC3		AC3	AC3		DAC1		
PB4	9	SYNC	ADC4		ADC4	ADC4	AC4					TMS
PB5	10	SYNC	ADC5		ADC5	ADC5	AC5	AC5				TDI
PB6	11	SYNC	ADC6		ADC6	ADC6	AC6					TCK
PB7	12	SYNC	ADC7		ADC7	ADC7		AC7	AC0OUT			TDO

**Table 30-3. Port C - Alternate functions.**

PORT C	PIN #	INTERRUPT	TCC0	AWEXC	TCC1	USARTC0	USARTC1	SPIC	TWIC	CLOCKOUT	EVENTOUT
GND	13										
VCC	14										
PC0	15	SYNC	OC0A	OC0ALS					SDA		
PC1	16	SYNC	OC0B	OC0AHS		XCK0			SCL		
PC2	17	SYNC/ASYNC	OC0C	OC0BLS		RXD0					

Mnemonics	Operands	Description	Operation			Flags	#Clocks
RCALL	k	Relative Call Subroutine	PC	←	PC + k + 1	None	2 / 3 <sup>(1)</sup>
ICALL		Indirect Call to (Z)	PC(15:0) PC(21:16)	← ←	Z, 0	None	2 / 3 <sup>(1)</sup>
EICALL		Extended Indirect Call to (Z)	PC(15:0) PC(21:16)	← ←	Z, EIND	None	3 <sup>(1)</sup>
CALL	k	call Subroutine	PC	←	k	None	3 / 4 <sup>(1)</sup>
RET		Subroutine Return	PC	←	STACK	None	4 / 5 <sup>(1)</sup>
RETI		Interrupt Return	PC	←	STACK	I	4 / 5 <sup>(1)</sup>
CPSE	Rd,Rr	Compare, Skip if Equal	if (Rd = Rr) PC	←	PC + 2 or 3	None	1 / 2 / 3
CP	Rd,Rr	Compare	Rd - Rr			Z,C,N,V,S,H	1
CPC	Rd,Rr	Compare with Carry	Rd - Rr - C			Z,C,N,V,S,H	1
CPI	Rd,K	Compare with Immediate	Rd - K			Z,C,N,V,S,H	1
SBRC	Rr, b	Skip if Bit in Register Cleared	if (Rr(b) = 0) PC	←	PC + 2 or 3	None	1 / 2 / 3
SBRs	Rr, b	Skip if Bit in Register Set	if (Rr(b) = 1) PC	←	PC + 2 or 3	None	1 / 2 / 3
SBIC	A, b	Skip if Bit in I/O Register Cleared	if (I/O(A,b) = 0) PC	←	PC + 2 or 3	None	2 / 3 / 4
SBIS	A, b	Skip if Bit in I/O Register Set	If (I/O(A,b) = 1) PC	←	PC + 2 or 3	None	2 / 3 / 4
BRBS	s, k	Branch if Status Flag Set	if (SREG(s) = 1) then PC	←	PC + k + 1	None	1 / 2
BRBC	s, k	Branch if Status Flag Cleared	if (SREG(s) = 0) then PC	←	PC + k + 1	None	1 / 2
BREQ	k	Branch if Equal	if (Z = 1) then PC	←	PC + k + 1	None	1 / 2
BRNE	k	Branch if Not Equal	if (Z = 0) then PC	←	PC + k + 1	None	1 / 2
BRCS	k	Branch if Carry Set	if (C = 1) then PC	←	PC + k + 1	None	1 / 2
BRCC	k	Branch if Carry Cleared	if (C = 0) then PC	←	PC + k + 1	None	1 / 2
BRSH	k	Branch if Same or Higher	if (C = 0) then PC	←	PC + k + 1	None	1 / 2
BRLO	k	Branch if Lower	if (C = 1) then PC	←	PC + k + 1	None	1 / 2
BRMI	k	Branch if Minus	if (N = 1) then PC	←	PC + k + 1	None	1 / 2
BRPL	k	Branch if Plus	if (N = 0) then PC	←	PC + k + 1	None	1 / 2
BRGE	k	Branch if Greater or Equal, Signed	if (N ⊕ V = 0) then PC	←	PC + k + 1	None	1 / 2
BRLT	k	Branch if Less Than, Signed	if (N ⊕ V = 1) then PC	←	PC + k + 1	None	1 / 2
BRHS	k	Branch if Half Carry Flag Set	if (H = 1) then PC	←	PC + k + 1	None	1 / 2
BRHC	k	Branch if Half Carry Flag Cleared	if (H = 0) then PC	←	PC + k + 1	None	1 / 2
BRTS	k	Branch if T Flag Set	if (T = 1) then PC	←	PC + k + 1	None	1 / 2
BRTC	k	Branch if T Flag Cleared	if (T = 0) then PC	←	PC + k + 1	None	1 / 2
BRVS	k	Branch if Overflow Flag is Set	if (V = 1) then PC	←	PC + k + 1	None	1 / 2
BRVC	k	Branch if Overflow Flag is Cleared	if (V = 0) then PC	←	PC + k + 1	None	1 / 2
BRIE	k	Branch if Interrupt Enabled	if (I = 1) then PC	←	PC + k + 1	None	1 / 2
BRID	k	Branch if Interrupt Disabled	if (I = 0) then PC	←	PC + k + 1	None	1 / 2
Data Transfer Instructions							
MOV	Rd, Rr	Copy Register	Rd	←	Rr	None	1
MOVW	Rd, Rr	Copy Register Pair	Rd+1:Rd	←	Rr+1:Rr	None	1

Mnemonics	Operands	Description	Operation			Flags	#Clocks
SPM	Z+	Store Program Memory and Post-Increment by 2	(RAMPZ:Z) Z	← ←	R1:R0, Z + 2	None	-
IN	Rd, A	In From I/O Location	Rd	←	I/O(A)	None	1
OUT	A, Rr	Out To I/O Location	I/O(A)	←	Rr	None	1
PUSH	Rr	Push Register on Stack	STACK	←	Rr	None	1 <sup>(1)</sup>
POP	Rd	Pop Register from Stack	Rd	←	STACK	None	2 <sup>(1)</sup>
Bit and Bit-test Instructions							
LSL	Rd	Logical Shift Left	Rd(n+1) Rd(0) C	← ← ←	Rd(n), 0, Rd(7)	Z,C,N,V,H	1
LSR	Rd	Logical Shift Right	Rd(n) Rd(7) C	← ← ←	Rd(n+1), 0, Rd(0)	Z,C,N,V	1
ROL	Rd	Rotate Left Through Carry	Rd(0) Rd(n+1) C	← ← ←	C, Rd(n), Rd(7)	Z,C,N,V,H	1
ROR	Rd	Rotate Right Through Carry	Rd(7) Rd(n) C	← ← ←	C, Rd(n+1), Rd(0)	Z,C,N,V	1
ASR	Rd	Arithmetic Shift Right	Rd(n)	←	Rd(n+1), n=0..6	Z,C,N,V	1
SWAP	Rd	Swap Nibbles	Rd(3..0)	↔	Rd(7..4)	None	1
BSET	s	Flag Set	SREG(s)	←	1	SREG(s)	1
BCLR	s	Flag Clear	SREG(s)	←	0	SREG(s)	1
SBI	A, b	Set Bit in I/O Register	I/O(A, b)	←	1	None	1
CBI	A, b	Clear Bit in I/O Register	I/O(A, b)	←	0	None	1
BST	Rr, b	Bit Store from Register to T	T	←	Rr(b)	T	1
BLD	Rd, b	Bit load from T to Register	Rd(b)	←	T	None	1
SEC		Set Carry	C	←	1	C	1
CLC		Clear Carry	C	←	0	C	1
SEN		Set Negative Flag	N	←	1	N	1
CLN		Clear Negative Flag	N	←	0	N	1
SEZ		Set Zero Flag	Z	←	1	Z	1
CLZ		Clear Zero Flag	Z	←	0	Z	1
SEI		Global Interrupt Enable	I	←	1	I	1
CLI		Global Interrupt Disable	I	←	0	I	1
SES		Set Signed Test Flag	S	←	1	S	1
CLS		Clear Signed Test Flag	S	←	0	S	1
SEV		Set Two's Complement Overflow	V	←	1	V	1
CLV		Clear Two's Complement Overflow	V	←	0	V	1
SET		Set T in SREG	T	←	1	T	1
CLT		Clear T in SREG	T	←	0	T	1
SEH		Set Half Carry Flag in SREG	H	←	1	H	1
CLH		Clear Half Carry Flag in SREG	H	←	0	H	1

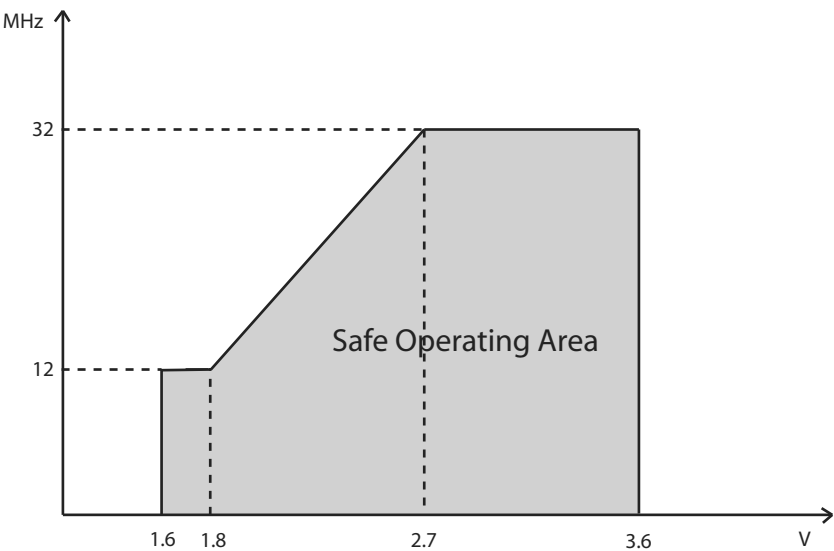
34.3 Speed

Table 34-2. Operating voltage and frequency.

Symbol	Parameter	Condition	Min	Typ	Max	Units
Clk <sub>CPU</sub>	CPU clock frequency	V <sub>CC</sub> = 1.6V	0		12	MHz
		V <sub>CC</sub> = 1.8V	0		12	
		V <sub>CC</sub> = 2.7V	0		32	
		V <sub>CC</sub> = 3.6V	0		32	

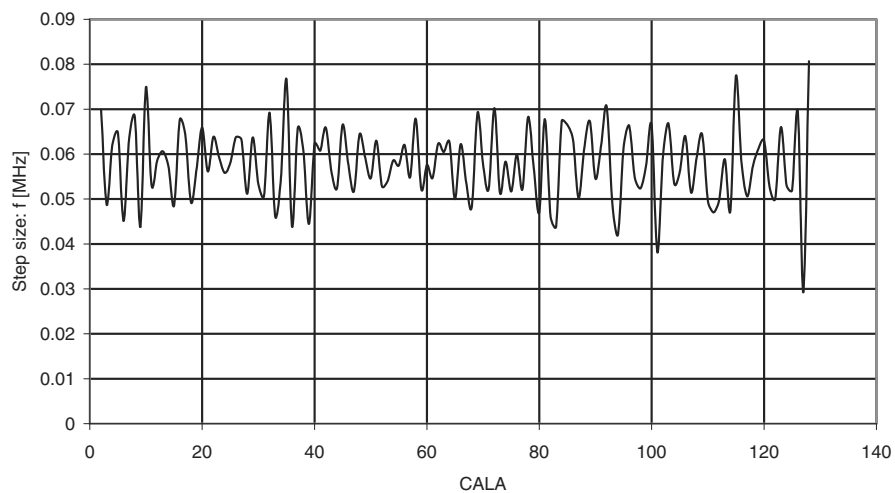
The maximum CPU clock frequency of the XMEGA A1 devices is depending on V<sub>CC</sub>. As shown in Figure 34-1 on page 75 the Frequency vs. V<sub>CC</sub> curve is linear between 1.8V < V<sub>CC</sub> < 2.7V.

Figure 34-1. Maximum Frequency vs. Vcc



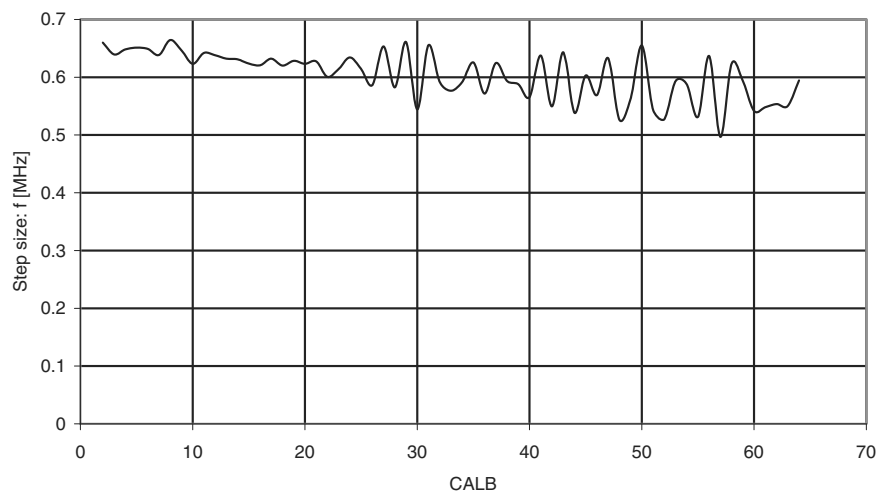
**Figure 35-25. Internal 32 MHz Oscillator CalA Calibration Step Size**

$T = -40 \text{ to } 85^\circ\text{C}$ ,  $V_{CC} = 3\text{V}$



**Figure 35-26. Internal 32 MHz Oscillator CalB Calibration Step Size**

$T = -40 \text{ to } 85^\circ\text{C}$ ,  $V_{CC} = 3\text{V}$



- TWI START condition at bus timeout will cause transaction to be dropped
- TWI Data Interrupt Flag erroneously read as set
- WDR instruction inside closed window will not issue reset

### 1. Bandgap voltage input for the ACs cannot be changed when used for both ACs simultaneously

If the Bandgap voltage is selected as input for one Analog Comparator (AC) and then selected/deselected as input for another AC, the first comparator will be affected for up to 1  $\mu$ s and could potentially give a wrong comparison result.

#### Problem fix/Workaround

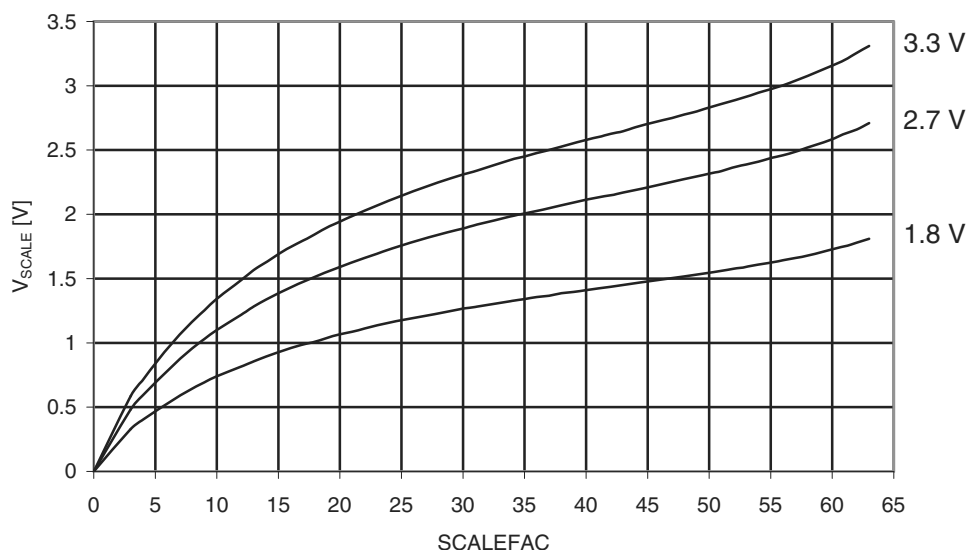
If the Bandgap is required for both ACs simultaneously, configure the input selection for both ACs before enabling any of them.

### 2. VCC voltage scaler for AC is non-linear

The 6-bit VCC voltage scaler in the Analog Comparators is non-linear.

**Figure 36-1. Analog Comparator Voltage Scaler vs. Scalefac**

$T = 25^{\circ}\text{C}$



#### Problem fix/Workaround

Use external voltage input for the analog comparator if accurate voltage levels are needed

### 3. The ADC has up to $\pm 2$ LSB inaccuracy

The ADC will have up to  $\pm 2$  LSB inaccuracy, visible as a saw-tooth pattern on the input voltage/ output value transfer function of the ADC. The inaccuracy increases with increasing voltage reference reaching  $\pm 2$  LSB with 3V reference.



## 18. DAC has up to $\pm 10$ LSB noise in Sampled Mode

If the DAC is running in Sample and Hold (S/H) mode and conversion for one channel is done at maximum rate (i.e. the DAC is always busy doing conversion for this channel), this will block refresh signals to the second channel.

### Problem fix/Workaround

When using the DAC in S/H mode, ensure that none of the channels is running at maximum conversion rate, or ensure that the conversion rate of both channels is high enough to not require refresh.

## 19. Conversion lost on DAC channel B in event triggered mode

If during dual channel operation channel 1 is set in auto triggered conversion mode, channel 1 conversions are occasionally lost. This means that not all data-values written to the Channel 1 data register are converted.

### Problem fix/Workaround

Keep the DAC conversion interval in the range 000-001 (1 and 3 CLK), and limit the Peripheral clock frequency so the conversion interval never is shorter than 1.5  $\mu$ s.

## 20. Both DFLLs and both oscillators have to be enabled for one to work

In order to use the automatic runtime calibration for the 2 MHz or the 32 MHz internal oscillators, the DFLL for both oscillators and both oscillators have to be enabled for one to work.

### Problem fix/Workaround

Enable both DFLLs and both oscillators when using automatic runtime calibration for either of the internal oscillators.

## 21. Access error when multiple bus masters are accessing SDRAM

If one bus master (CPU and DMA channels) is using the EBI to access an SDRAM in burst mode and another bus master is accessing the same row number in a different BANK of the SDRAM in the cycle directly after the burst access is complete, the access for the second bus master will fail.

### Problem fix/Workaround

Do not put stack pointer in SDRAM and use DMA Controller in 1 byte burst mode if CPU and DMA Controller are required to access SDRAM at the same time.

## 22. EEPROM page buffer always written when NVM DATA0 is written

If the EEPROM is memory mapped, writing to NVM DATA0 will corrupt data in the EEPROM page buffer.

### Problem fix/Workaround

Before writing to NVM DATA0, for example when doing software CRC or flash page buffer write, check if EEPROM page buffer active loading flag (EELoad) is set. Do not write NVM DATA0 when EELoad is set.

## 23. Pending full asynchronous pin change interrupts will not wake the device

### 37. TWI, a general address call will match independent of the R/W-bit value

When the TWI is in Slave mode and a general address call is issued on the bus, the TWI Slave will get an address match regardless of the received R/W bit.

#### Problem fix/Workaround

Use software to check the R/W-bit on general call address match.

### 38. TWI Transmit collision flag not cleared on repeated start

The TWI transmit collision flag should be automatically cleared on start and repeated start, but is only cleared on start.

#### Problem fix/Workaround

Clear the flag in software after address interrupt.

### 39. Clearing TWI Stop Interrupt Flag may lock the bus

If software clears the STOP Interrupt Flag (APIF) on the same Peripheral Clock cycle as the hardware sets this flag due to a new address received, CLKHOLD is not cleared and the SCL line is not released. This will lock the bus.

#### Problem fix/Workaround

Check if the bus state is IDLE. If this is the case, it is safe to clear APIF. If the bus state is not IDLE, wait for the SCL pin to be low before clearing APIF.

Code:

```
/* Only clear the interrupt flag if within a "safe zone". */
while ( /* Bus not IDLE: */
        ((COMMS_TWI.MASTER.STATUS & TWI_MASTER_BUSSTATE_gm) !=
         TWI_MASTER_BUSSTATE_IDLE_gc) &&
        /* SCL not held by slave: */
        !(COMMS_TWI.SLAVE.STATUS & TWI_SLAVE_CLKHOLD_bm)
      )
{
    /* Ensure that the SCL line is low */
    if ( !(COMMS_PORT.IN & PIN1_bm) )
        if ( !(COMMS_PORT.IN & PIN1_bm) )
            break;
}
/* Check for an pending address match interrupt */
if ( !(COMMS_TWI.SLAVE.STATUS & TWI_SLAVE_CLKHOLD_bm) )
{
    /* Safely clear interrupt flag */
    COMMS_TWI.SLAVE.STATUS |= (uint8_t)TWI_SLAVE_APIF_bm;
}
```

## 12. Sampled BOD in Active mode will cause noise when bandgap is used as reference

Using the BOD in sampled mode when the device is running in Active or Idle mode will add noise on the bandgap reference for ADC and DAC.

### Problem fix/Workaround

If the bandgap is used as reference for either the ADC or the DAC, the BOD must not be set in sampled mode.

## 13. Default setting for SDRAM refresh period too low

If the SDRAM refresh period is set to a value less than 0x20, the SDRAM content may be corrupted when accessing through On-Chip Debug sessions.

### Problem fix/Workaround

The SDRAM refresh period (REFRESHH/L) should not be set to a value less than 0x20.

## 14. Flash Power Reduction Mode can not be enabled when entering sleep mode

If Flash Power Reduction Mode is enabled when entering Power-save or Extended Standby sleep mode, the device will only wake up on every fourth wake-up request.

If Flash Power Reduction Mode is enabled when entering Idle sleep mode, the wake-up time will vary with up to 16 CPU clock cycles.

### Problem fix/Workaround

Disable Flash Power Reduction mode before entering sleep mode.

## 15. JTAG enable does not override Analog Comparator B output

When JTAG is enabled this will not override the Analog Comparator B (ACB) output, AC0OUT on pin 7 if this is enabled.

### Problem fix/Workaround

AC0OUT for ACB should not be enabled when JTAG is used. Use only analog comparator output for ACA when JTAG is used, or use the PDI as debug interface.

## 16. Bandgap measurement with the ADC is non-functional when $V_{CC}$ is below 2.7V

The ADC cannot be used to do bandgap measurements when  $V_{CC}$  is below 2.7V.

### Problem fix/Workaround

If internal voltages must be measured when  $V_{CC}$  is below 2.7V, measure the internal 1.00V reference instead of the bandgap.

### 37.13 8067C – 06/2008

1. Updated the Front page and “Features” on page 1.
2. Updated the “DC Characteristics” on page 73.
3. Updated Figure 3-1 on page 6.
4. Added “Flash and EEPROM Page Size” on page 15.
5. Updated Table 33-6 on page 72 with new data: Gain Error, Offset Error and Signal -to-Noise Ratio (SNR).
6. Updated Errata “ATxmega64A1 and ATxmega128A1 rev. G” on page 107.

### 37.14 8067B – 05/2008

1. Updated “Pinout/Block Diagram” on page 3 and “Pinout and Pin Functions” on page 55.
2. Added XMEGA A1 Block Diagram, Figure 3-1 on page 6.
3. Updated “Overview” on page 5 included the XMEGA A1 explanation text on page 6.
4. Updated AVR CPU “Features” on page 8.
5. Updated Event System block diagram, Figure 10-1 on page 20.
6. Updated “Interrupts and Programmable Multilevel Interrupt Controller” on page 29.
7. Updated “AC - Analog Comparator” on page 52.
8. Updated “Alternate Pin Function Description” on page 55.
9. Updated “Alternate Pin Functions” on page 58.
10. Updated “Typical Characteristics” on page 82.
11. Updated “Ordering Information” on page 2.
12. Updated “Overview” on page 5.
13. Updated Figure 6-1 on page 8.
14. Inserted a new Figure 16-1 on page 37.
15. Updated Speed grades in “Speed” on page 75.
16. Added a new ATxmega384A1 device in “Features” on page 1, updated “Ordering Information” on page 2 and “Memories” on page 12.
17. Replaced the Figure 3-1 on page 6 by a new XMEGA A1 detailed block diagram.
18. Inserted Errata “ATxmega64A1 and ATxmega128A1 rev. G” on page 107.

### 37.15 8067A – 02/2008

1. Initial revision.

