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### What is "[Embedded - Microcontrollers](#)"?

"[Embedded - Microcontrollers](#)" refer to small, integrated circuits designed to perform specific tasks within larger systems. These microcontrollers are essentially compact computers on a single chip, containing a processor core, memory, and programmable input/output peripherals. They are called "embedded" because they are embedded within electronic devices to control various functions, rather than serving as standalone computers. Microcontrollers are crucial in modern electronics, providing the intelligence and control needed for a wide range of applications.

### Applications of "[Embedded - Microcontrollers](#)"

#### Details

Product Status	Active
Core Processor	8051
Core Size	8-Bit
Speed	50MHz
Connectivity	SMBus (2-Wire/I <sup>2</sup> C), SPI, UART/USART
Peripherals	POR, PWM, Temp Sensor, WDT
Number of I/O	25
Program Memory Size	32KB (32K x 8)
Program Memory Type	FLASH
EEPROM Size	-
RAM Size	4.25K x 8
Voltage - Supply (Vcc/Vdd)	1.8V ~ 5.25V
Data Converters	A/D 25x12b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 125°C (TA)
Mounting Type	Surface Mount
Package / Case	32-LQFP
Supplier Device Package	32-LQFP (7x7)
Purchase URL	<a href="https://www.e-xfl.com/product-detail/silicon-labs/c8051f507-iq">https://www.e-xfl.com/product-detail/silicon-labs/c8051f507-iq</a>

## 2. Ordering Information

The following features are common to all devices in this family:

- 50 MHz system clock and 50 MIPS throughput (peak)
- 4352 bytes of RAM (256 internal bytes and 4096 XRAM bytes)
- SMBus/I<sup>2</sup>C, Enhanced SPI, Enhanced UART
- Four Timers
- Six Programmable Counter Array channels
- Internal 24 MHz oscillator
- Internal Voltage Regulator
- 12-bit, 200 ksps ADC
- Internal Voltage Reference and Temperature Sensor
- Two Analog Comparators

Table 2.1 shows the feature that differentiate the devices in this family.

**Table 3.1. Pin Definitions for the C8051F50x/F51x(Continued)**

<b>Name</b>	<b>Pin 'F500/1/4/5 (48-pin)</b>	<b>Pin F508/9- F510/1 (40-pin)</b>	<b>Pin 'F502/3/6/7 (32-pin)</b>	<b>Type</b>	<b>Description</b>
P3.7	19	11	—	D I/O	Port 3.7.
P4.0	18	—	—	D I/O	Port 4.0. See SFR Definition 20.28 for a description.
P4.1	17	—	—	D I/O	Port 4.1.
P4.2	16	—	—	D I/O	Port 4.2.
P4.3	15	—	—	D I/O	Port 4.3.
P4.4	14	—	—	D I/O	Port 4.4.
P4.5	13	—	—	D I/O	Port 4.5.
P4.6	10	—	—	D I/O	Port 4.6.
P4.7	9	—	—	D I/O	Port 4.7.

**Table 5.2. Global Electrical Characteristics (Continued)**

–40 to +125 °C, 24 MHz system clock unless otherwise specified.

Parameter	Conditions	Min	Typ	Max	Units
$I_{DD}^4$	$V_{DD} = 2.6\text{ V}, F = 200\text{ kHz}$	—	130	—	$\mu\text{A}$
	$V_{DD} = 2.6\text{ V}, F = 1.5\text{ MHz}$	—	990	—	$\mu\text{A}$
	$V_{DD} = 2.6\text{ V}, F = 25\text{ MHz}$	—	14	21	$\text{mA}$
	$V_{DD} = 2.6\text{ V}, F = 50\text{ MHz}$	—	25	33	$\text{mA}$
$I_{DD}$ Supply Sensitivity <sup>4</sup>	$F = 25\text{ MHz}$	—	68	—	$\%/V$
	$F = 1\text{ MHz}$	—	73	—	$\%/V$
$I_{DD}$ Frequency Sensitivity <sup>4,5</sup>	$V_{DD} = 2.1\text{V}, F \leq 12.5\text{ MHz}, T = 25\text{ }^\circ\text{C}$	—	0.46	—	$\text{mA}/\text{MHz}$
	$V_{DD} = 2.1\text{V}, F > 12.5\text{ MHz}, T = 25\text{ }^\circ\text{C}$	—	0.36	—	$\text{mA}/\text{MHz}$
	$V_{DD} = 2.6\text{V}, F \leq 12.5\text{ MHz}, T = 25\text{ }^\circ\text{C}$	—	0.64	—	$\text{mA}/\text{MHz}$
	$V_{DD} = 2.6\text{V}, F > 12.5\text{ MHz}, T = 25\text{ }^\circ\text{C}$	—	0.47	—	$\text{mA}/\text{MHz}$
<b>Notes:</b>					
<ol style="list-style-type: none"> <li>Given in Table 5.4 on page 46.</li> <li><math>V_{IO}</math> should not be lower than the <math>V_{DD}</math> voltage.</li> <li>SYSCLK must be at least 32 kHz to enable debugging.</li> <li>Based on device characterization data; Not production tested. Does not include oscillator supply current.</li> <li><math>I_{DD}</math> can be estimated for frequencies <math>\leq 12.5\text{ MHz}</math> by simply multiplying the frequency of interest by the frequency sensitivity number for that range. When using these numbers to estimate <math>I_{DD}</math> for <math>&gt;12.5\text{ MHz}</math>, the estimate should be the current at 50 MHz minus the difference in current indicated by the frequency sensitivity number. For example: <math>V_{DD} = 2.6\text{ V}; F = 20\text{ MHz}, I_{DD} = 26\text{ mA} - (50\text{ MHz} - 20\text{ MHz}) * 0.48\text{ mA}/\text{MHz} = 11.6\text{ mA}</math>.</li> <li>Idle <math>I_{DD}</math> can be estimated for frequencies <math>\leq 1\text{ MHz}</math> by simply multiplying the frequency of interest by the frequency sensitivity number. When using these numbers to estimate Idle <math>I_{DD}</math> for <math>&gt;1\text{ MHz}</math>, the estimate should be the current at 50 MHz minus the difference in current indicated by the frequency sensitivity number. For example: <math>V_{DD} = 2.6\text{ V}; F = 5\text{ MHz}, \text{Idle } I_{DD} = 21\text{ mA} - (50\text{ MHz} - 5\text{ MHz}) * 0.41\text{ mA}/\text{MHz} = 2.6\text{ mA}</math>.</li> </ol>					

**Table 5.4. Reset Electrical Characteristics**

–40 to +125 °C unless otherwise specified.

Parameter	Conditions	Min	Typ	Max	Units
$\overline{\text{RST}}$ Output Low Voltage	$V_{\text{IO}} = 5 \text{ V}$ ; $\text{IOL} = 70 \mu\text{A}$	—	—	40	mV
$\overline{\text{RST}}$ Input High Voltage		$0.7 \times V_{\text{IO}}$	—	—	
$\overline{\text{RST}}$ Input Low Voltage		—	—	$0.3 \times V_{\text{IO}}$	
$\overline{\text{RST}}$ Input Pullup Current	$\overline{\text{RST}} = 0.0 \text{ V}$ , $V_{\text{IO}} = 5 \text{ V}$	—	47	115	$\mu\text{A}$
$V_{\text{DD}}$ RST Threshold ( $V_{\text{RST-LOW}}$ )		1.65	1.75	1.80	V
$V_{\text{DD}}$ RST Threshold ( $V_{\text{RST-HIGH}}$ )		2.25	2.30	2.45	V
Missing Clock Detector Timeout	Time from last system clock rising edge to reset initiation $V_{\text{DD}} = 2.1\text{V}$	200	370	600	$\mu\text{s}$
	$V_{\text{DD}} = 2.5\text{V}$	200	270	600	
Reset Time Delay	Delay between release of any reset source and code execution at location 0x0000	—	130	160	$\mu\text{s}$
Minimum $\overline{\text{RST}}$ Low Time to Generate a System Reset		6	—	—	$\mu\text{s}$
$V_{\text{DD}}$ Monitor Turn-on Time		—	60	100	$\mu\text{s}$
$V_{\text{DD}}$ Monitor Supply Current		—	1	2	$\mu\text{A}$

**Table 5.5. Flash Electrical Characteristics**

$V_{\text{DD}} = 1.8$  to  $2.75 \text{ V}$ , –40 to +125 °C unless otherwise specified.

Parameter	Conditions	Min	Typ	Max	Units
Flash Size	C8051F500/1/2/3/8/9	65536*			Bytes
	C8051F504/5/6/7-F510/1	32768			
Endurance		20 k	150 k	—	Erase/Write
Flash Retention	85 °C	10	—	—	Years
Erase Cycle Time	25 MHz System Clock	28	30	45	ms
Write Cycle Time	25 MHz System Clock	79	84	125	$\mu\text{s}$
$V_{\text{DD}}$	Write / Erase operations	$V_{\text{RST-HIGH}}^2$	—	—	V

1. On the 64K Flash devices, 1024 bytes at addresses 0xFC00 to 0xFFFF are reserved.  
 2. See Table 5.4 for the  $V_{\text{RST-HIGH}}$  specification.

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**Table 5.9. ADC0 Electrical Characteristics**

VDDA = 1.8 to 2.75 V, -40 to +125 °C, VREF = 1.5 V (REFSL=0) unless otherwise specified.

Parameter	Conditions	Min	Typ	Max	Units
<b>DC Accuracy</b>					
Resolution			12		bits
Integral Nonlinearity		—	±0.5	±3	LSB
Differential Nonlinearity	Guaranteed Monotonic	—	±0.5	±1	LSB
Offset Error <sup>1</sup>		-10	-1.8	10	LSB
Full Scale Error		-20	1.7	20	LSB
Offset Temperature Coefficient		—	-2	—	ppm/°C
<b>Dynamic performance (10 kHz sine-wave single-ended input, 1 dB below Full Scale, 200 ksps)</b>					
Signal-to-Noise Plus Distortion		63	66	—	dB
Total Harmonic Distortion	Up to the 5th harmonic	—	82	—	dB
Spurious-Free Dynamic Range		—	-84	—	dB
<b>Conversion Rate</b>					
SAR Conversion Clock		—	—	3.6	MHz
Conversion Time in SAR Clocks <sup>2</sup>		13	—	—	clocks
Track/Hold Acquisition Time <sup>3</sup>	VDDA ≥ 2.0 V VDDA < 2.0 V	1.5 3.5	—	—	µs
Throughput Rate <sup>4</sup>	VDDA ≥ 2.0 V	—	—	200	ksps
<b>Analog Inputs</b>					
ADC Input Voltage Range <sup>5</sup>	gain = 1.0 (default) gain = n	0 0	—	VREF VREF/n	V
Absolute Pin Voltage with Respect to GND		0	—	V <sub>IO</sub>	V
Sampling Capacitance		—	32	—	pF
Input Multiplexer Impedance		—	3	—	kΩ
<b>Power Specifications</b>					
Power Supply Current (VDDA supplied to ADC0)	Operating Mode, 200 ksps	—	1100	1500	µA
Burst Mode (Idle)		—	1100	1500	µA
Power-On Time		5	—	—	µs
Power Supply Rejection Ratio		—	-60	—	dB
<b>Notes:</b>					
<ol style="list-style-type: none"> <li>1. Represents one standard deviation from the mean. Offset and full-scale error can be removed through calibration.</li> <li>2. An additional 2 FCLK cycles are required to start and complete a conversion</li> <li>3. Additional tracking time may be required depending on the output impedance connected to the ADC input. See Section “6.2.1. Settling Time Requirements” on page 57.</li> <li>4. An increase in tracking time will decrease the ADC throughput.</li> <li>5. See Section “6.3. Selectable Gain” on page 58 for more information about the setting the gain.</li> </ol>					

Note that false rising edges and falling edges can be detected when the comparator is first powered on or if changes are made to the hysteresis or response time control bits. Therefore, it is recommended that the rising-edge and falling-edge flags be explicitly cleared to logic 0 a short time after the comparator is enabled or its mode bits have been changed.

## SFR Definition 9.1. CPT0CN: Comparator0 Control

Bit	7	6	5	4	3	2	1	0
Name	CP0EN	CP0OUT	CP0RIF	CP0FIF	CP0HYP[1:0]		CP0HYN[1:0]	
Type	R/W	R	R/W	R/W	R/W		R/W	
Reset	0	0	0	0	0	0	0	0

SFR Address = 0x9A; SFR Page = 0x00

Bit	Name	Function
7	CP0EN	<b>Comparator0 Enable Bit.</b> 0: Comparator0 Disabled. 1: Comparator0 Enabled.
6	CP0OUT	<b>Comparator0 Output State Flag.</b> 0: Voltage on CP0+ < CP0-. 1: Voltage on CP0+ > CP0-.
5	CP0RIF	<b>Comparator0 Rising-Edge Flag. Must be cleared by software.</b> 0: No Comparator0 Rising Edge has occurred since this flag was last cleared. 1: Comparator0 Rising Edge has occurred.
4	CP0FIF	<b>Comparator0 Falling-Edge Flag. Must be cleared by software.</b> 0: No Comparator0 Falling-Edge has occurred since this flag was last cleared. 1: Comparator0 Falling-Edge has occurred.
3:2	CP0HYP[1:0]	<b>Comparator0 Positive Hysteresis Control Bits.</b> 00: Positive Hysteresis Disabled. 01: Positive Hysteresis = 5 mV. 10: Positive Hysteresis = 10 mV. 11: Positive Hysteresis = 20 mV.
1:0	CP0HYN[1:0]	<b>Comparator0 Negative Hysteresis Control Bits.</b> 00: Negative Hysteresis Disabled. 01: Negative Hysteresis = 5 mV. 10: Negative Hysteresis = 10 mV. 11: Negative Hysteresis = 20 mV.

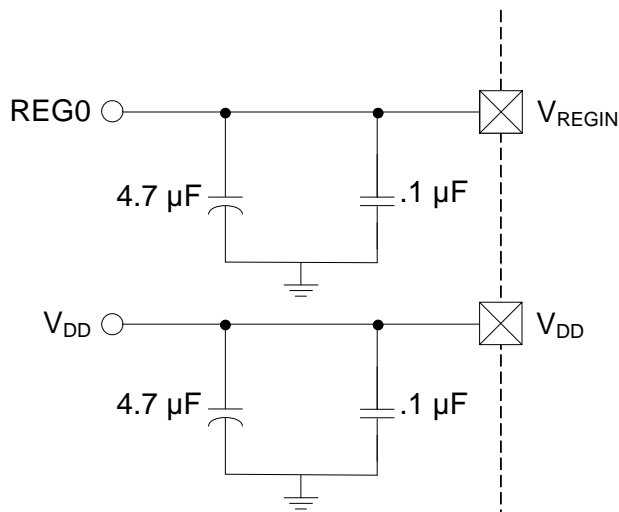
## 10. Voltage Regulator (REG0)

C8051F50x/F51x devices include an on-chip low dropout voltage regulator (REG0). The input to REG0 at the  $V_{\text{REGIN}}$  pin can be as high as 5.25 V. The output can be selected by software to 2.1 V or 2.6 V. When enabled, the output of REG0 appears on the  $V_{\text{DD}}$  pin, powers the microcontroller core, and can be used to power external devices. On reset, REG0 is enabled and can be disabled by software.

The Voltage regulator can generate an interrupt (if enabled by EREG0, EIE2.0) that is triggered whenever the  $V_{\text{REGIN}}$  input voltage drops below the dropout threshold voltage. This dropout interrupt has no pending flag and the recommended procedure to use it is as follows:

1. Wait enough time to ensure the  $V_{\text{REGIN}}$  input voltage is stable
2. Enable the dropout interrupt (EREG0, EIE2.0) and select the proper priority (PREG0, EIP2.0)
3. If triggered, inside the interrupt disable it (clear EREG0, EIE2.0), execute all procedures necessary to protect your application (put it in a safe mode and leave the interrupt now disabled).
4. In the main application, now running in the safe mode, regularly checks the DROPOUT bit (REG0CN.0). Once it is cleared by the regulator hardware the application can enable the interrupt again (EREG0, EIE1.6) and return to the normal mode operation.

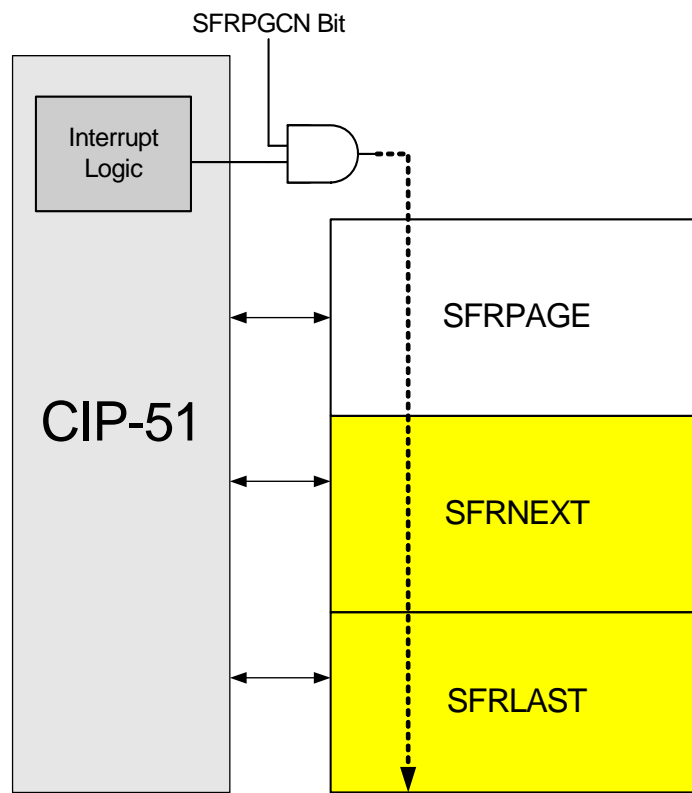
The input ( $V_{\text{REGIN}}$ ) and output ( $V_{\text{DD}}$ ) of the voltage regulator should both be bypassed with a large capacitor (4.7  $\mu\text{F}$  + 0.1  $\mu\text{F}$ ) to ground as shown in Figure 10.1 below. This capacitor will eliminate power spikes and provide any immediate power required by the microcontroller. The settling time associated with the voltage regulator is shown in Table X.



**Figure 10.1. External Capacitors for Voltage Regulator Input/Output—Regulator Enabled**

If the internal voltage regulator is not used, the  $V_{\text{REGIN}}$  input should be tied to  $V_{\text{DD}}$ , as shown in Figure 10.2.





**Figure 13.1. SFR Page Stack**

Automatic hardware switching of the SFR Page on interrupts may be enabled or disabled as desired using the SFR Automatic Page Control Enable Bit located in the SFR Page Control Register (SFR0CN). This function defaults to “enabled” upon reset. In this way, the autoswitching function will be enabled unless disabled in software.

A summary of the SFR locations (address and SFR page) are provided in Table 13.3 in the form of an SFR memory map. Each memory location in the map has an SFR page row, denoting the page in which that SFR resides. Certain SFRs are accessible from ALL SFR pages, and are denoted by the “(ALL PAGES)” designation. For example, the Port I/O registers P0, P1, P2, and P3 all have the “(ALL PAGES)” designation, indicating these SFRs are accessible from all SFR pages regardless of the SFRPAGE register value.

### 13.3. SFR Page Stack Example

The following is an example that shows the operation of the SFR Page Stack during interrupts. In this example, the SFR Control register is left in the default enabled state (i.e., SFRPGEN = 1), and the CIP-51 is executing in-line code that is writing values to SPI Data Register (SFR “SPI0DAT”, located at address 0xA3 on SFR Page 0x00). The device is also using the CAN peripheral (CAN0) and the Programmable Counter Array (PCA0) peripheral to generate a PWM output. The PCA is timing a critical control function in its interrupt service round so its associated ISR that is set to low priority. At this point, the SFR page is set to access the SPI0DAT SFR (SFRPAGE = 0x00). See Figure 13.2.

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## 14.2. Interrupt Register Descriptions

The SFRs used to enable the interrupt sources and set their priority level are described in this section. Refer to the data sheet section associated with a particular on-chip peripheral for information regarding valid interrupt conditions for the peripheral and the behavior of its interrupt-pending flag(s).

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## 18.5.3. Split Mode with Bank Select

When EMI0CF[3:2] are set to 10, the XRAM memory map is split into two areas, on-chip space and off-chip space.

- Effective addresses below the internal XRAM size boundary will access on-chip XRAM space.
- Effective addresses above the internal XRAM size boundary will access off-chip space.
- 8-bit MOVX operations use the contents of EMI0CN to determine whether the memory access is on-chip or off-chip. The upper 8-bits of the Address Bus A[15:8] are determined by EMI0CN, and the lower 8-bits of the Address Bus A[7:0] are determined by R0 or R1. All 16-bits of the Address Bus A[15:0] are driven in “Bank Select” mode.
- 16-bit MOVX operations use the contents of DPTR to determine whether the memory access is on-chip or off-chip, and the full 16-bits of the Address Bus A[15:0] are driven during the off-chip transaction.

## 18.5.4. External Only

When EMI0CF[3:2] are set to 11, all MOVX operations are directed to off-chip space. On-chip XRAM is not visible to the CPU. This mode is useful for accessing off-chip memory located between 0x0000 and the internal XRAM size boundary.

- 8-bit MOVX operations ignore the contents of EMI0CN. The upper Address bits A[15:8] are not driven (identical behavior to an off-chip access in “Split Mode without Bank Select” described above). This allows the user to manipulate the upper address bits at will by setting the Port state directly. The lower 8-bits of the effective address A[7:0] are determined by the contents of R0 or R1.
- 16-bit MOVX operations use the contents of DPTR to determine the effective address A[15:0]. The full 16-bits of the Address Bus A[15:0] are driven during the off-chip transaction.

## 18.6. Timing

The timing parameters of the External Memory Interface can be configured to enable connection to devices having different setup and hold time requirements. The Address Setup time, Address Hold time,  $\overline{RD}$  and  $\overline{WR}$  strobe widths, and in multiplexed mode, the width of the ALE pulse are all programmable in units of SYSCLK periods through EMI0TC, shown in SFR Definition 18.3, and EMI0CF[1:0].

The timing for an off-chip MOVX instruction can be calculated by adding 4 SYSCLK cycles to the timing parameters defined by the EMI0TC register. Assuming non-multiplexed operation, the minimum execution time for an off-chip XRAM operation is 5 SYSCLK cycles (1 SYSCLK for  $\overline{RD}$  or  $\overline{WR}$  pulse + 4 SYSCLKs). For multiplexed operations, the Address Latch Enable signal will require a minimum of 2 additional SYSCLK cycles. Therefore, the minimum execution time for an off-chip XRAM operation in multiplexed mode is 7 SYSCLK cycles (2 for  $\overline{ALE}$  + 1 for  $\overline{RD}$  or  $\overline{WR}$  + 4). The programmable setup and hold times default to the maximum delay settings after a reset. Table 18.3 lists the ac parameters for the External Memory Interface, and Figure 18.4 through Figure 18.9 show the timing diagrams for the different External Memory Interface modes and MOVX operations.

## 19. Oscillators and Clock Selection

C8051F50x/F51x devices include a programmable internal high-frequency oscillator, an external oscillator drive circuit, and a clock multiplier. The internal oscillator can be enabled/disabled and calibrated using the OSCICN, OSCICRS, and OSCIFIN registers, as shown in Figure 19.1. The system clock can be sourced by the external oscillator circuit or the internal oscillator. The clock multiplier can produce three possible base outputs which can be scaled by a programmable factor of 1, 2/3, 2/4 (or 1/2), 2/5, 2/6 (or 1/3), or 2/7: Internal Oscillator x 2, Internal Oscillator x 4, External Oscillator x 2, or External Oscillator x 4.

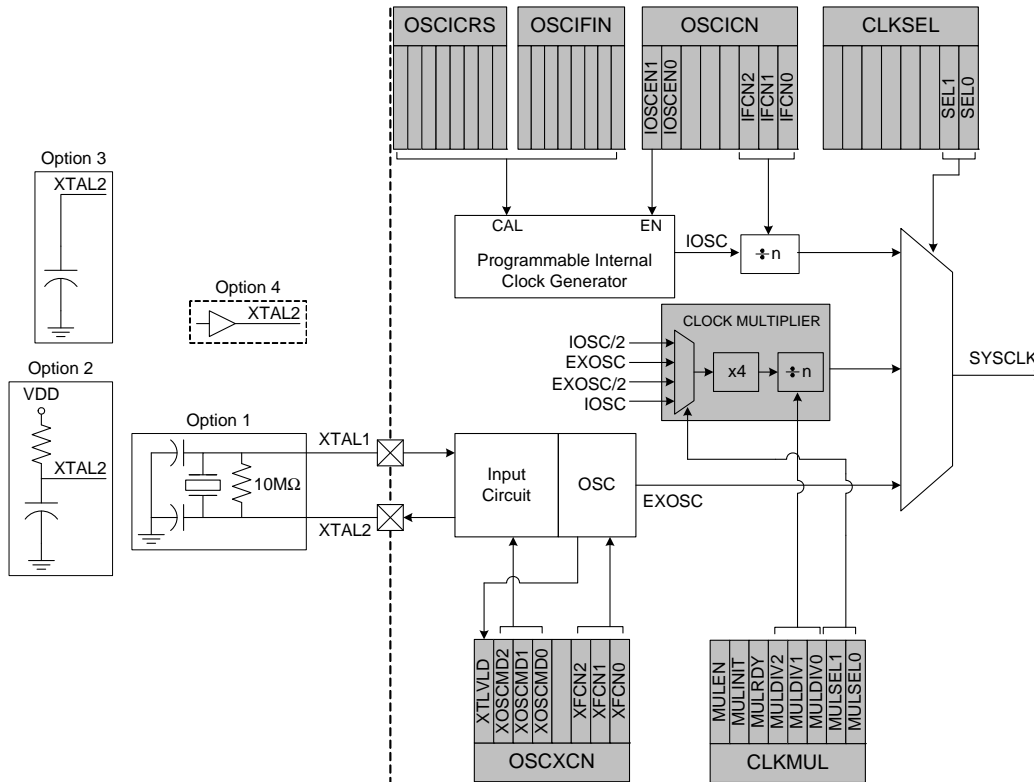


Figure 19.1. Oscillator Options

### 19.1. System Clock Selection

The CLKSL[1:0] bits in register CLKSEL select which oscillator source is used as the system clock. CLKSL[1:0] must be set to 01b for the system clock to run from the external oscillator; however the external oscillator may still clock certain peripherals (timers, PCA) when the internal oscillator is selected as the system clock. The system clock may be switched on-the-fly between the internal oscillator, external oscillator, and Clock Multiplier so long as the selected clock source is enabled and has settled.

The internal oscillator requires little start-up time and may be selected as the system clock immediately following the register write which enables the oscillator. The external RC and C modes also typically require no startup time.

External crystals and ceramic resonators however, typically require a start-up time before they are settled and ready for use. The Crystal Valid Flag (XTLVLD in register OSCXCN) is set to 1 by hardware when the external crystal or ceramic resonator is settled. **In crystal mode, to avoid reading a false XTLVLD, software should delay at least 1 ms between enabling the external oscillator and checking XTLVLD.**

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## SFR Definition 20.3. XBR2: Port I/O Crossbar Register 1

Bit	7	6	5	4	3	2	1	0
Name	WEAKPUD	XBARE	Reserved					LIN0E
Type	R/W	R/W	R/W	R/W	R/W	R	R/W	R/W
Reset	0	0	0	0	0	0	0	0

SFR Address = 0xC7; SFR Page = 0x0F

Bit	Name	Function
7	WEAKPUD	<b>Port I/O Weak Pullup Disable.</b> 0: Weak Pullups enabled (except for Ports whose I/O are configured for analog mode). 1: Weak Pullups disabled.
6	XBARE	<b>Crossbar Enable.</b> 0: Crossbar disabled. 1: Crossbar enabled.
5:1	Reserved	Always Write to 00000b.
0	LIN0E	<b>LIN I/O Output Enable.</b> 0: LIN I/O unavailable at Port pin. 1: LIN_TX, LIN_RX routed to Port pins.

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## SFR Definition 20.6. P1MASK: Port 1 Mask Register

Bit	7	6	5	4	3	2	1	0
Name	P1MASK[7:0]							
Type	R/W							
Reset	0	0	0	0	0	0	0	0

SFR Address = 0xF4; SFR Page = 0x00

Bit	Name	Function
7:0	P1MASK[7:0]	<b>Port 1 Mask Value.</b> Selects P1 pins to be compared to the corresponding bits in P1MAT. 0: P1.n pin logic value is ignored and cannot cause a Port Mismatch event. 1: P1.n pin logic value is compared to P1MAT.n.

## SFR Definition 20.7. P1MAT: Port 1 Match Register

Bit	7	6	5	4	3	2	1	0
Name	P1MAT[7:0]							
Type	R/W							
Reset	1	1	1	1	1	1	1	1

SFR Address = 0xF3; SFR Page = 0x00

Bit	Name	Function
7:0	P1MAT[7:0]	<b>Port 1 Match Value.</b> Match comparison value used on Port 1 for bits in P1MAT which are set to 1. 0: P1.n pin logic value is compared with logic LOW. 1: P1.n pin logic value is compared with logic HIGH.

## SFR Definition 20.19. P1SKIP: Port 1 Skip

<b>Bit</b>	7	6	5	4	3	2	1	0
<b>Name</b>	P1SKIP[7:0]							
<b>Type</b>	R/W							
<b>Reset</b>	0	0	0	0	0	0	0	0

SFR Address = 0xD5; SFR Page = 0x0F

Bit	Name	Function
7:0	P1SKIP[7:0]	<p><b>Port 1 Crossbar Skip Enable Bits.</b></p> <p>These bits select Port 1 pins to be skipped by the Crossbar Decoder. Port pins used for analog, special functions or GPIO should be skipped by the Crossbar.</p> <p>0: Corresponding P1.n pin is not skipped by the Crossbar.</p> <p>1: Corresponding P1.n pin is skipped by the Crossbar.</p>

## SFR Definition 20.20. P2: Port 2

<b>Bit</b>	7	6	5	4	3	2	1	0
<b>Name</b>	P2[7:0]							
<b>Type</b>	R/W							
<b>Reset</b>	1	1	1	1	1	1	1	1

SFR Address = 0xA0; SFR Page = All Pages; Bit-Addressable

Bit	Name	Description	Write	Read
7:0	P2[7:0]	<p><b>Port 2Data.</b></p> <p>Sets the Port latch logic value or reads the Port pin logic state in Port cells configured for digital I/O.</p>	<p>0: Set output latch to logic LOW.</p> <p>1: Set output latch to logic HIGH.</p>	<p>0: P2.n Port pin is logic LOW.</p> <p>1: P2.n Port pin is logic HIGH.</p>

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## 21.1. Software Interface with the LIN Controller

The selection of the mode (Master or Slave) and the automatic baud rate feature are done through the LIN0 Control Mode (LIN0CF) register. The other LIN registers are accessed indirectly through the two SFRs LIN0 Address (LIN0ADR) and LIN0 Data (LIN0DAT). The LIN0ADR register selects which LIN register is targeted by reads/writes of the LIN0DAT register. The full list of indirectly-accessible LIN registers is given in Table 21.4 on page 210.

## 21.2. LIN Interface Setup and Operation

The hardware based LIN controller allows for the implementation of both Master and Slave nodes with minimal firmware overhead and complete control of the interface status while allowing for interrupt and polled mode operation.

The first step to use the controller is to define the basic characteristics of the node:

Mode—Master or Slave

Baud Rate—Either defined manually or using the autobaud feature (slave mode only)

Checksum Type—Select between classic or enhanced checksum, both of which are implemented in hardware.

### 21.2.1. Mode Definition

Following the LIN specification, the controller implements in hardware both the Slave and Master operating modes. The mode is configured using the MODE bit (LIN0CF.6).

### 21.2.2. Baud Rate Options: Manual or Autobaud

The LIN controller can be selected to have its baud rate calculated manually or automatically. A master node must always have its baud rate set manually, but slave nodes can choose between a manual or automatic setup. The configuration is selected using the ABAUD bit (LIN0CF.5).

Both the manual and automatic baud rate configurations require additional setup. The following sections explain the different options available and their relation with the baud rate, along with the steps necessary to achieve the required baud rate.

### 21.2.3. Baud Rate Calculations: Manual Mode

The baud rate used by the LIN controller is a function of the System Clock (SYSCLK) and the LIN timing registers according to the following equation:

$$\text{baud\_rate} = \frac{\text{SYSCLK}}{2^{(\text{prescaler} + 1)} \times \text{divider} \times (\text{multiplier} + 1)}$$

The prescaler, divider and multiplier factors are part of the LIN0DIV and LIN0MUL registers and can assume values in the following range:

**Table 21.1. Baud Rate Calculation Variable Ranges**

Factor	Range
prescaler	0...3
multiplier	0...31
divider	200...511

**Important Note:** The minimum system clock (SYSCLK) to operate the LIN controller is 8 MHz.



# C8051F50x/F51x

## LIN Register Definition 21.7. LIN0ERR: LIN0 Error Register

Bit	7	6	5	4	3	2	1	0
Name				SYNCH	PRTY	TOUT	CHK	BITERR
Type	R	R	R	R	R	R	R	R
Reset	0	0	0	0	0	0	0	0

Indirect Address = 0x0A

Bit	Name	Function
7:5	Unused	Read = 000b; Write = Don't Care
4	SYNCH	<b>Synchronization Error Bit (slave mode only).</b> 0: No error with the SYNCH FIELD has been detected. 1: Edges of the SYNCH FIELD are outside of the maximum tolerance.
3	PRTY	<b>Parity Error Bit (slave mode only).</b> 0: No parity error has been detected. 1: A parity error has been detected.
2	TOUT	<b>Timeout Error Bit.</b> 0: A timeout error has not been detected. 1: A timeout error has been detected. This error is detected whenever one of the following conditions is met: <ul style="list-style-type: none"> <li>The master is expecting data from a slave and the slave does not respond.</li> <li>The slave is expecting data but no data is transmitted on the bus.</li> <li>A frame is not finished within the maximum frame length.</li> <li>The application does not set the DTACK bit (LIN0CTRL.4) or STOP bit (LIN0CTRL.7) until the end of the reception of the first byte after the identifier.</li> </ul>
1	CHK	<b>Checksum Error Bit.</b> 0: Checksum error has not been detected. 1: Checksum error has been detected.
0	BITERR	<b>Bit Transmission Error Bit.</b> 0: No error in transmission has been detected. 1: The bit value monitored during transmission is different than the bit value sent.

## SFR Definition 23.1. SMB0CF: SMBus Clock/Configuration

Bit	7	6	5	4	3	2	1	0
Name	ENSMB	INH	BUSY	EXTHOLD	SMBTOE	SMBFTE	SMBCS[1:0]	
Type	R/W	R/W	R	R/W	R/W	R/W	R/W	
Reset	0	0	0	0	0	0	0	0

SFR Address = 0xC1; SFR Page = 0x00

Bit	Name	Function
7	ENSMB	<b>SMBus Enable.</b> This bit enables the SMBus interface when set to 1. When enabled, the interface constantly monitors the SDA and SCL pins.
6	INH	<b>SMBus Slave Inhibit.</b> When this bit is set to logic 1, the SMBus does not generate an interrupt when slave events occur. This effectively removes the SMBus slave from the bus. Master Mode interrupts are not affected.
5	BUSY	<b>SMBus Busy Indicator.</b> This bit is set to logic 1 by hardware when a transfer is in progress. It is cleared to logic 0 when a STOP or free-timeout is sensed.
4	EXTHOLD	<b>SMBus Setup and Hold Time Extension Enable.</b> This bit controls the SDA setup and hold times according to Table 23.2. 0: SDA Extended Setup and Hold Times disabled. 1: SDA Extended Setup and Hold Times enabled.
3	SMBTOE	<b>SMBus SCL Timeout Detection Enable.</b> This bit enables SCL low timeout detection. If set to logic 1, the SMBus forces Timer 3 to reload while SCL is high and allows Timer 3 to count when SCL goes low. If Timer 3 is configured to Split Mode, only the High Byte of the timer is held in reload while SCL is high. Timer 3 should be programmed to generate interrupts at 25 ms, and the Timer 3 interrupt service routine should reset SMBus communication.
2	SMBFTE	<b>SMBus Free Timeout Detection Enable.</b> When this bit is set to logic 1, the bus will be considered free if SCL and SDA remain high for more than 10 SMBus clock source periods.
1:0	SMBCS[1:0]	<b>SMBus Clock Source Selection.</b> These two bits select the SMBus clock source, which is used to generate the SMBus bit rate. The selected device should be configured according to Equation 23.1. 00: Timer 0 Overflow 01: Timer 1 Overflow 10: Timer 2 High Byte Overflow 11: Timer 2 Low Byte Overflow

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## 25.2. SPI0 Master Mode Operation

A SPI master device initiates all data transfers on a SPI bus. SPI0 is placed in master mode by setting the Master Enable flag (MSTEN, SPI0CN.6). Writing a byte of data to the SPI0 data register (SPI0DAT) when in master mode writes to the transmit buffer. If the SPI shift register is empty, the byte in the transmit buffer is moved to the shift register, and a data transfer begins. The SPI0 master immediately shifts out the data serially on the MOSI line while providing the serial clock on SCK. The SPIF (SPI0CN.7) flag is set to logic 1 at the end of the transfer. If interrupts are enabled, an interrupt request is generated when the SPIF flag is set. While the SPI0 master transfers data to a slave on the MOSI line, the addressed SPI slave device simultaneously transfers the contents of its shift register to the SPI master on the MISO line in a full-duplex operation. Therefore, the SPIF flag serves as both a transmit-complete and receive-data-ready flag. The data byte received from the slave is transferred MSB-first into the master's shift register. When a byte is fully shifted into the register, it is moved to the receive buffer where it can be read by the processor by reading SPI0DAT.

When configured as a master, SPI0 can operate in one of three different modes: multi-master mode, 3-wire single-master mode, and 4-wire single-master mode. The default, multi-master mode is active when NSSMD1 (SPI0CN.3) = 0 and NSSMD0 (SPI0CN.2) = 1. In this mode, NSS is an input to the device, and is used to disable the master SPI0 when another master is accessing the bus. When NSS is pulled low in this mode, MSTEN (SPI0CN.6) and SPIEN (SPI0CN.0) are set to 0 to disable the SPI master device, and a Mode Fault is generated (MODF, SPI0CN.5 = 1). Mode Fault will generate an interrupt if enabled. SPI0 must be manually re-enabled in software under these circumstances. In multi-master systems, devices will typically default to being slave devices while they are not acting as the system master device. In multi-master mode, slave devices can be addressed individually (if needed) using general-purpose I/O pins. Figure 25.2 shows a connection diagram between two master devices in multiple-master mode.

3-wire single-master mode is active when NSSMD1 (SPI0CN.3) = 0 and NSSMD0 (SPI0CN.2) = 0. In this mode, NSS is not used, and is not mapped to an external port pin through the crossbar. Any slave devices that must be addressed in this mode should be selected using general-purpose I/O pins. Figure 25.3 shows a connection diagram between a master device in 3-wire master mode and a slave device.

4-wire single-master mode is active when NSSMD1 (SPI0CN.3) = 1. In this mode, NSS is configured as an output pin, and can be used as a slave-select signal for a single SPI device. In this mode, the output value of NSS is controlled (in software) with the bit NSSMD0 (SPI0CN.2). Additional slave devices can be addressed using general-purpose I/O pins. Figure 25.4 shows a connection diagram for a master device in 4-wire master mode and two slave devices.

## SFR Definition 26.9. TMR2RLL: Timer 2 Reload Register Low Byte

<b>Bit</b>	7	6	5	4	3	2	1	0
<b>Name</b>	TMR2RLL[7:0]							
<b>Type</b>	R/W							
<b>Reset</b>	0	0	0	0	0	0	0	0

SFR Address = 0xCA; SFR Page = 0x00

Bit	Name	Function
7:0	TMR2RLL[7:0]	<b>Timer 2 Reload Register Low Byte.</b> TMR2RLL holds the low byte of the reload value for Timer 2.

## SFR Definition 26.10. TMR2RLH: Timer 2 Reload Register High Byte

<b>Bit</b>	7	6	5	4	3	2	1	0
<b>Name</b>	TMR2RLH[7:0]							
<b>Type</b>	R/W							
<b>Reset</b>	0	0	0	0	0	0	0	0

SFR Address = 0xCB; SFR Page = 0x00

Bit	Name	Function
7:0	TMR2RLH[7:0]	<b>Timer 2 Reload Register High Byte.</b> TMR2RLH holds the high byte of the reload value for Timer 2.

## 28. C2 Interface

C8051F50x/F51x devices include an on-chip Silicon Labs 2-Wire (C2) debug interface to allow Flash programming and in-system debugging with the production part installed in the end application. The C2 interface uses a clock signal (C2CK) and a bi-directional C2 data signal (C2D) to transfer information between the device and a host system. See the C2 Interface Specification for details on the C2 protocol.

### 28.1. C2 Interface Registers

The following describes the C2 registers necessary to perform Flash programming through the C2 interface. All C2 registers are accessed through the C2 interface as described in the C2 Interface Specification.

#### C2 Register Definition 28.1. C2ADD: C2 Address

Bit	7	6	5	4	3	2	1	0
Name	C2ADD[7:0]							
Type	R/W							
Reset	0	0	0	0	0	0	0	0

Bit	Name	Function										
7:0	C2ADD[7:0]	<p><b>C2 Address.</b> The C2ADD register is accessed via the C2 interface to select the target Data register for C2 Data Read and Data Write commands.</p> <table border="1"> <thead> <tr> <th>Address</th> <th>Description</th> </tr> </thead> <tbody> <tr> <td>0x00</td> <td>Selects the Device ID register for Data Read instructions</td> </tr> <tr> <td>0x01</td> <td>Selects the Revision ID register for Data Read instructions</td> </tr> <tr> <td>0x02</td> <td>Selects the C2 Flash Programming Control register for Data Read/Write instructions</td> </tr> <tr> <td>0xB4</td> <td>Selects the C2 Flash Programming Data register for Data Read/Write instructions</td> </tr> </tbody> </table>	Address	Description	0x00	Selects the Device ID register for Data Read instructions	0x01	Selects the Revision ID register for Data Read instructions	0x02	Selects the C2 Flash Programming Control register for Data Read/Write instructions	0xB4	Selects the C2 Flash Programming Data register for Data Read/Write instructions
Address	Description											
0x00	Selects the Device ID register for Data Read instructions											
0x01	Selects the Revision ID register for Data Read instructions											
0x02	Selects the C2 Flash Programming Control register for Data Read/Write instructions											
0xB4	Selects the C2 Flash Programming Data register for Data Read/Write instructions											