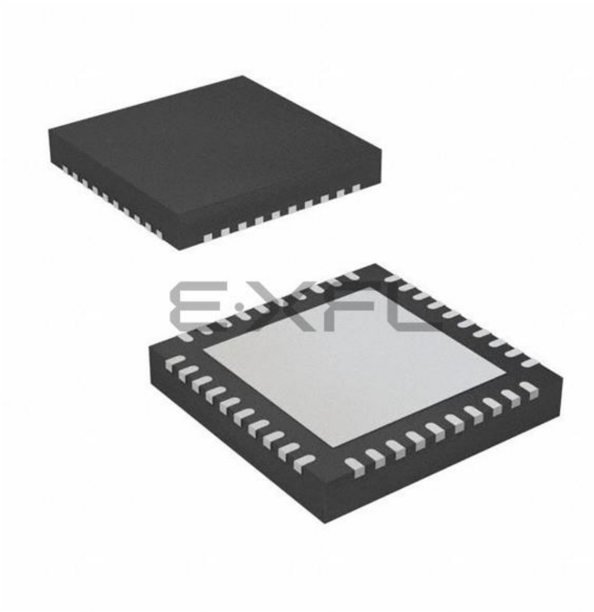
Silicon Labs - C8051F511-IM Datasheet



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What is "Embedded - Microcontrollers"?

"Embedded - Microcontrollers" refer to small, integrated circuits designed to perform specific tasks within larger systems. These microcontrollers are essentially compact computers on a single chip, containing a processor core, memory, and programmable input/output peripherals. They are called "embedded" because they are embedded within electronic devices to control various functions, rather than serving as standalone computers. Microcontrollers are crucial in modern electronics, providing the intelligence and control needed for a wide range of applications.

Applications of "Embedded - Microcontrollers"

Details	
Product Status	Active
Core Processor	8051
Core Size	8-Bit
Speed	50MHz
Connectivity	EBI/EMI, SMBus (2-Wire/I²C), SPI, UART/USART
Peripherals	POR, PWM, Temp Sensor, WDT
Number of I/O	33
Program Memory Size	32KB (32K x 8)
Program Memory Type	FLASH
EEPROM Size	-
RAM Size	4.25K x 8
Voltage - Supply (Vcc/Vdd)	1.8V ~ 5.25V
Data Converters	A/D 32x12b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 125°C (TA)
Mounting Type	Surface Mount
Package / Case	40-VFQFN Exposed Pad
Supplier Device Package	40-QFN (6x6)
Purchase URL	https://www.e-xfl.com/product-detail/silicon-labs/c8051f511-im

2. Ordering Information

The following features are common to all devices in this family:

- 50 MHz system clock and 50 MIPS throughput (peak)
- 4352 bytes of RAM (256 internal bytes and 4096 XRAM bytes)
- SMBus/I^{2}C, Enhanced SPI, Enhanced UART
- Four Timers
- Six Programmable Counter Array channels
- Internal 24 MHz oscillator
- Internal Voltage Regulator
- 12-bit, 200 ksps ADC
- Internal Voltage Reference and Temperature Sensor
- Two Analog Comparators

Table 2.1 shows the feature that differentiate the devices in this family.

6.2. Output Code Formatting

The registers ADC0H and ADC0L contain the high and low bytes of the output conversion code. When the repeat count is set to 1, conversion codes are represented in 12-bit unsigned integer format and the output conversion code is updated after each conversion. Inputs are measured from 0 to V_{REF} x 4095/4096. Data can be right-justified or left-justified, depending on the setting of the AD0LJST bit (ADC0CN.2). Unused bits in the ADC0H and ADC0L registers are set to 0. Example codes are shown below for both right-justified and left-justified data.

Input Voltage	Right-Justified ADC0H:ADC0L (AD0LJST = 0)	Left-Justified ADC0H:ADC0L (AD0LJST = 1)
VREF x 4095/4096	0x0FFF	0xFFF0
VREF x 2048/4096	0x0800	0x8000
VREF x 2047/4096	0x07FF	0x7FF0
0	0x0000	0x0000

When the ADC0 Repeat Count is greater than 1, the output conversion code represents the accumulated result of the conversions performed and is updated after the last conversion in the series is finished. Sets of 4, 8, or 16 consecutive samples can be accumulated and represented in unsigned integer format. The repeat count can be selected using the AD0RPT bits in the ADC0CF register. The value must be right-justified (AD0LJST = 0), and unused bits in the ADC0H and ADC0L registers are set to 0. The following example shows right-justified codes for repeat counts greater than 1. Notice that accumulating 2^n samples is equivalent to left-shifting by *n* bit positions when all samples returned from the ADC have the same value.

Input Voltage	Repeat Count = 4	Repeat Count = 8	Repeat Count = 16
V_{REF} x 4095/4096	0x3FFC	0x7FF8	0xFFF0
V_{REF} x 2048/4096	0x2000	0x4000	0x8000
V_{REF} x 2047/4096	0x1FFC	0x3FF8	0x7FF0
0	0x0000	0x0000	0x0000

6.2.1. Settling Time Requirements

A minimum tracking time is required before an accurate conversion is performed. This tracking time is determined by any series impedance, including the AMUX0 resistance, the ADC0 sampling capacitance, and the accuracy required for the conversion.

Figure 6.5 shows the equivalent ADC0 input circuit. The required ADC0 settling time for a given settling accuracy (SA) may be approximated by Equation 6.1. When measuring the Temperature Sensor output, use the settling time specified in Table 5.10 on page 50. When measuring V_{DD} with respect to GND, R_{TOTAL} reduces to R_{MUX}. See Table 5.9 for ADC0 minimum settling time requirements as well as the mux impedance and sampling capacitor values.

$$t = \ln\left(\frac{2^n}{SA}\right) \times R_{TOTAL}C_{SAMPLE}$$

Equation 6.1. ADC0 Settling Time Requirements

Where:
SA is the settling accuracy, given as a fraction of an LSB (for example, 0.25 to settle within 1/4 LSB)
t is the required settling time in seconds
R_{TOTAL} is the sum of the AMUX0 resistance and any external source resistance.
n is the ADC resolution in bits (10).

Note that false rising edges and falling edges can be detected when the comparator is first powered on or if changes are made to the hysteresis or response time control bits. Therefore, it is recommended that the rising-edge and falling-edge flags be explicitly cleared to logic 0 a short time after the comparator is enabled or its mode bits have been changed.

SFR Definition 9.1. CPT0CN: Comparator0 Control

Bit	7	6	5	4	3	2	1	0
Name	CP0EN	CP0OUT	CP0RIF	CP0FIF	CP0HYP[1:0]		CP0HYN[1:0]	
Type	R/W	R	R/W	R/W	R/W		R/W	
Reset	0	0	0	0	0	0	0	0

SFR Address = 0x9A; SFR Page = 0x00

Bit	Name	Function
7	CP0EN	**Comparator0 Enable Bit.** 0: Comparator0 Disabled. 1: Comparator0 Enabled.
6	CP0OUT	**Comparator0 Output State Flag.** 0: Voltage on CP0+ < CP0–. 1: Voltage on CP0+ > CP0–.
5	CP0RIF	**Comparator0 Rising-Edge Flag. Must be cleared by software.** 0: No Comparator0 Rising Edge has occurred since this flag was last cleared. 1: Comparator0 Rising Edge has occurred.
4	CP0FIF	**Comparator0 Falling-Edge Flag. Must be cleared by software.** 0: No Comparator0 Falling-Edge has occurred since this flag was last cleared. 1: Comparator0 Falling-Edge has occurred.
3:2	CP0HYP[1:0]	**Comparator0 Positive Hysteresis Control Bits.** 00: Positive Hysteresis Disabled. 01: Positive Hysteresis = 5 mV. 10: Positive Hysteresis = 10 mV. 11: Positive Hysteresis = 20 mV.
1:0	CP0HYN[1:0]	**Comparator0 Negative Hysteresis Control Bits.** 00: Negative Hysteresis Disabled. 01: Negative Hysteresis = 5 mV. 10: Negative Hysteresis = 10 mV. 11: Negative Hysteresis = 20 mV.

SFR Definition 13.3. SFRNEXT: SFR Next

Bit	7	6	5	4	3	2	1	0
Name	SFRNEXT[7:0]							
Type	R/W							
Reset	0	0	0	0	0	0	0	0

SFR Address = 0x85; SFR Page = All Pages

Bit	Name	Function
7:0	SFRNEXT[7:0]	**SFR Page Bits.** This is the value that will go to the SFR Page register upon a return from interrupt. Write: Sets the SFR Page contained in the second byte of the SFR Stack. This will cause the SFRPAGE SFR to have this SFR page value upon a return from interrupt. Read: Returns the value of the SFR page contained in the second byte of the SFR stack. SFR page context is retained upon interrupts/return from interrupts in a 3 byte SFR Page Stack: SFRPAGE is the first entry, SFRNEXT is the second, and SFRLAST is the third entry. The SFR stack bytes may be used alter the context in the SFR Page Stack, and will not cause the stack to “push” or “pop”. Only interrupts and return from interrupts cause pushes and pops of the SFR Page Stack.

18.4.2. Non-multiplexed Configuration

In Non-multiplexed mode, the Data Bus and the Address Bus pins are not shared. An example of a Non-multiplexed Configuration is shown in Figure 18.2. See Section “18.6.1. Non-Multiplexed Mode” on page 158 for more information about Non-multiplexed operation.

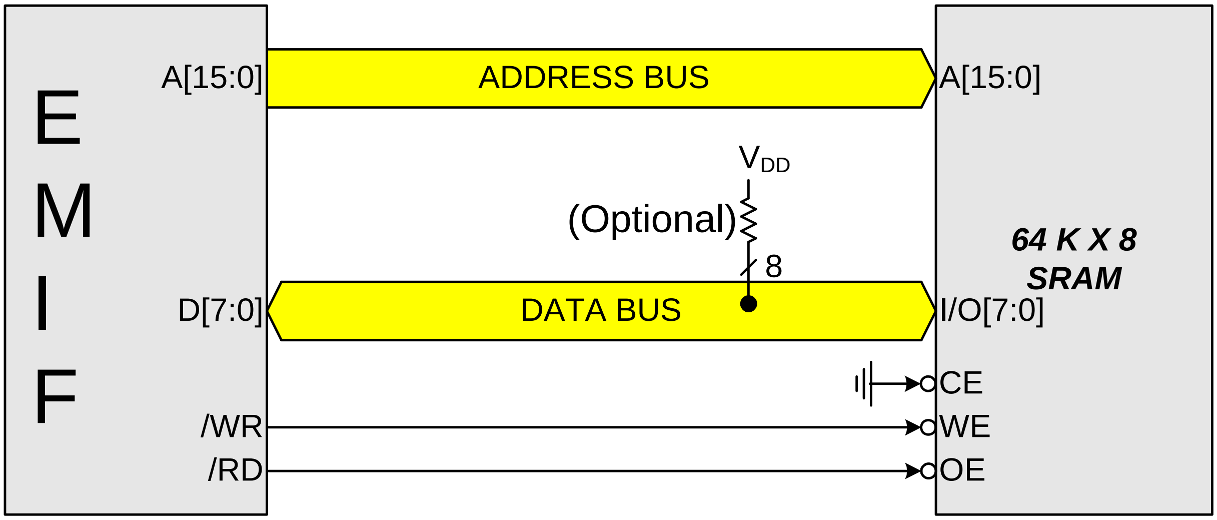


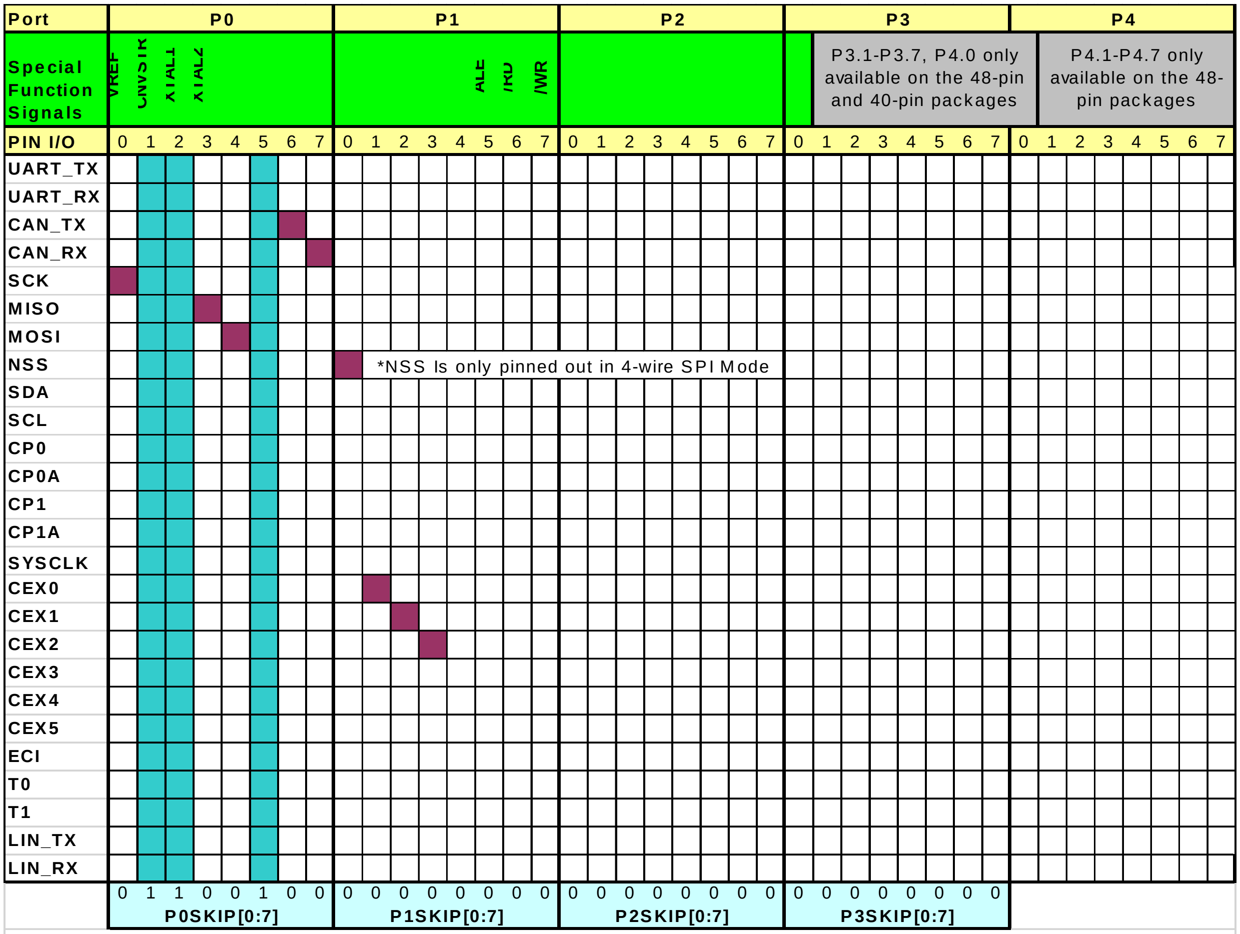
Figure 18.2. Non-multiplexed Configuration Example

SFR Definition 19.5. CLKMUL: Clock Multiplier

Bit	7	6	5	4	3	2	1	0
Name	MULEN	MULINIT	MULRDY	MULDIV[2:0]			MULSEL[1:0]	
Type	R/W	R/W	R	R/W			R/W	
Reset	0	0	0	0	0	0	0	0

SFR Address = 0x97; SFR Page = 0x0F;

<table>
<tr><th>Bit</th><th>Name</th><th colspan="3">Function</th></tr>
<tr><td>7</td><td>MULEN</td><td colspan="3"><b>Clock Multiplier Enable.</b>
0: Clock Multiplier disabled.
1: Clock Multiplier enabled.</td></tr>
<tr><td>6</td><td>MULINIT</td><td colspan="3"><b>Clock Multiplier Initialize.</b>
This bit is 0 when the Clock Multiplier is enabled. Once enabled, writing a 1 to this bit will initialize the Clock Multiplier. The MULRDY bit reads 1 when the Clock Multiplier is stabilized.</td></tr>
<tr><td>5</td><td>MULRDY</td><td colspan="3"><b>Clock Multiplier Ready.</b>
0: Clock Multiplier is not ready.
1: Clock Multiplier is ready (PLL is locked).</td></tr>
<tr><td>4:2</td><td>MULDIV[2:0]</td><td colspan="3"><b>Clock Multiplier Output Scaling Factor.</b>
000: Clock Multiplier Output scaled by a factor of 1.
001: Clock Multiplier Output scaled by a factor of 1.
010: Clock Multiplier Output scaled by a factor of 1.
011: Clock Multiplier Output scaled by a factor of 2/3*.
100: Clock Multiplier Output scaled by a factor of 2/4 (1/2).
101: Clock Multiplier Output scaled by a factor of 2/5*.
110: Clock Multiplier Output scaled by a factor of 2/6 (1/3).
111: Clock Multiplier Output scaled by a factor of 2/7*.
<b>*Note:</b> The Clock Multiplier output duty cycle is not 50% for these settings.</td></tr>
<tr><td rowspan="6">1:0</td><td rowspan="6">MULSEL[1:0]</td><td colspan="3"><b>Clock Multiplier Input Select.</b>
These bits select the clock supplied to the Clock Multiplier</td></tr>
<tr><td><b>MULSEL[1:0]</b></td><td><b>Selected Input Clock</b></td><td><b>Clock Multiplier Output for MULDIV[2:0] = 000b</b></td></tr>
<tr><td>00</td><td>Internal Oscillator</td><td>Internal Oscillator x 2</td></tr>
<tr><td>01</td><td>External Oscillator</td><td>External Oscillator x 2</td></tr>
<tr><td>10</td><td>Internal Oscillator</td><td>Internal Oscillator x 4</td></tr>
<tr><td>11</td><td>External Oscillator</td><td>External Oscillator x 4</td></tr>
<tr><td colspan="5"><b>Notes:</b>The maximum system clock is 50 MHz, and so the Clock Multiplier output should be scaled accordingly.
If Internal Oscillator x 2 or External Oscillator x 2 is selected using the MULSEL bits, MULDIV[2:0] is ignored.</td></tr>
</table>



Port	P0								P1							
Special Function Signals	VREF	CNVSTR	XTAL1	XTAL2										ALE	/RD	/WR
PIN I/O	0	1	2	3	4	5	6	7	0	1	2	3	4	5	6	7
UART_TX		🟦	🟦			🟦										
UART_RX		🟦	🟦			🟦										
CAN_TX		🟦	🟦			🟦	🟪									
CAN_RX		🟦	🟦			🟦		🟪								
SCK	🟪	🟦	🟦			🟦										
MISO		🟦	🟦	🟪		🟦										
MOSI		🟦	🟦		🟪	🟦										
NSS		🟦	🟦			🟦			🟪	*NSS Is only pinned out in 4-wire SPI Mode						
SDA		🟦	🟦			🟦										
SCL		🟦	🟦			🟦										
CP0		🟦	🟦			🟦										
CP0A		🟦	🟦			🟦										
CP1		🟦	🟦			🟦										
CP1A		🟦	🟦			🟦										
SYSCLK		🟦	🟦			🟦										
CEX0		🟦	🟦			🟦				🟪						
CEX1		🟦	🟦			🟦					🟪					
CEX2		🟦	🟦			🟦						🟪				
CEX3		🟦	🟦			🟦										
CEX4		🟦	🟦			🟦										
CEX5		🟦	🟦			🟦										
ECI		🟦	🟦			🟦										
T0		🟦	🟦			🟦										
T1		🟦	🟦			🟦										
LIN_TX		🟦	🟦			🟦										
LIN_RX		🟦	🟦			🟦										
	0	1	1	0	0	1	0	0	0	0	0	0	0	0	0	0
	P0SKIP[0:7]								P1SKIP[0:7]							

Port	P2								P3								P4							
Special Function Signals										P3.1-P3.7, P4.0 only available on the 48-pin and 40-pin packages								P4.1-P4.7 only available on the 48-pin packages						
PIN I/O	0	1	2	3	4	5	6	7	0	1	2	3	4	5	6	7	0	1	2	3	4	5	6	7
UART_TX																								
UART_RX																								
CAN_TX																								
CAN_RX																								
SCK																								
MISO																								
MOSI																								
NSS																								
SDA																								
SCL																								
CP0																								
CP0A																								
CP1																								
CP1A																								
SYSCLK																								
CEX0																								
CEX1																								
CEX2																								
CEX3																								
CEX4																								
CEX5																								
ECI																								
T0																								
T1																								
LIN_TX																								
LIN_RX																								
	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0								
	P2SKIP[0:7]								P3SKIP[0:7]															

Figure 20.4. Crossbar Priority Decoder in Example Configuration

20.4. Port I/O Initialization

Port I/O initialization consists of the following steps:

1. Select the input mode (analog or digital) for all Port pins, using the Port Input Mode register (PnMDIN).
2. Select the output mode (open-drain or push-pull) for all Port pins, using the Port Output Mode register (PnMDOUT).
3. Select any pins to be skipped by the I/O Crossbar using the Port Skip registers (PnSKIP).
4. Assign Port pins to desired peripherals.
5. Enable the Crossbar (XBARE = 1).

All Port pins must be configured as either analog or digital inputs. Port 4 on the C8051F500/1/4/5 and C8051F508/9-F510/1 is a digital-only Port. Any pins to be used as Comparator or ADC inputs should be configured as an analog inputs. When a pin is configured as an analog input, its weak pullup, digital driver, and digital receiver are disabled. This process saves power and reduces noise on the analog input. Pins configured as digital inputs may still be used by analog peripherals; however this practice is not recommended.

Additionally, all analog input pins should be configured to be skipped by the Crossbar (accomplished by setting the associated bits in PnSKIP). Port input mode is set in the PnMDIN register, where a 1 indicates a digital input, and a 0 indicates an analog input. All pins default to digital inputs on reset. See SFR Definition

SFR Definition 20.3. XBR2: Port I/O Crossbar Register 1

Bit	**7**	**6**	**5**	**4**	**3**	**2**	**1**	**0**
Name	WEAKPUD	XBARE	Reserved					LIN0E
Type	R/W	R/W	R/W	R/W	R/W	R	R/W	R/W
Reset	0	0	0	0	0	0	0	0

SFR Address = 0xC7; SFR Page = 0x0F

Bit	**Name**	**Function**
7	WEAKPUD	**Port I/O Weak Pullup Disable.** 0: Weak Pullups enabled (except for Ports whose I/O are configured for analog mode). 1: Weak Pullups disabled.
6	XBARE	**Crossbar Enable.** 0: Crossbar disabled. 1: Crossbar enabled.
5:1	Reserved	Always Write to 00000b.
0	LIN0E	**LIN I/O Output Enable.** 0: LIN I/O unavailable at Port pin. 1: LIN_TX, LIN_RX routed to Port pins.

20.6. Special Function Registers for Accessing and Configuring Port I/O

All Port I/O are accessed through corresponding special function registers (SFRs) that are both byte addressable and bit addressable, except for P4 which is only byte addressable. When writing to a Port, the value written to the SFR is latched to maintain the output data value at each pin. When reading, the logic levels of the Port's input pins are returned regardless of the XBRn settings (i.e., even when the pin is assigned to another signal by the Crossbar, the Port register can always read its corresponding Port I/O pin). The exception to this is the execution of the read-modify-write instructions that target a Port Latch register as the destination. The read-modify-write instructions when operating on a Port SFR are the following: ANL, ORL, XRL, JBC, CPL, INC, DEC, DJNZ and MOV, CLR or SETB, when the destination is an individual bit in a Port SFR. For these instructions, the value of the latch register (not the pin) is read, modified, and written back to the SFR.

Ports 0–3 have a corresponding PnSKIP register which allows its individual Port pins to be assigned to digital functions or skipped by the Crossbar. All Port pins used for analog functions, GPIO, or dedicated digital functions such as the EMIF should have their PnSKIP bit set to 1.

The Port input mode of the I/O pins is defined using the Port Input Mode registers (PnMDIN). Each Port cell can be configured for analog or digital I/O. This selection is required even for the digital resources selected in the XBRn registers, and is not automatic. The only exception to this is P4, which can only be used for digital I/O.

The output driver characteristics of the I/O pins are defined using the Port Output Mode registers (PnMDOUT). Each Port Output driver can be configured as either open drain or push-pull. This selection is required even for the digital resources selected in the XBRn registers, and is not automatic. The only exception to this is the SMBus (SDA, SCL) pins, which are configured as open-drain regardless of the PnMDOUT settings.

SFR Definition 20.12. P0: Port 0

Bit	**7**	**6**	**5**	**4**	**3**	**2**	**1**	**0**
Name	P0[7:0]							
Type	R/W							
Reset	1	1	1	1	1	1	1	1

SFR Address = 0x80; SFR Page = All Pages; Bit-Addressable

Bit	Name	Description	Write	Read
7:0	P0[7:0]	**Port 0 Data.** Sets the Port latch logic value or reads the Port pin logic state in Port cells configured for digital I/O.	0: Set output latch to logic LOW. 1: Set output latch to logic HIGH.	0: P0.n Port pin is logic LOW. 1: P0.n Port pin is logic HIGH.

SFR Definition 20.21. P2MDIN: Port 2 Input Mode

Bit	7	6	5	4	3	2	1	0
Name	P2MDIN[7:0]							
Type	R/W							
Reset	1	1	1	1	1	1	1	1

SFR Address = 0xF3; SFR Page = 0x0F

Bit	Name	Function
7:0	P2MDIN[7:0]	**Analog Configuration Bits for P2.7–P2.0 (respectively).** Port pins configured for analog mode have their weak pull-up and digital receiver disabled. For analog mode, the pin also needs to be configured for open-drain mode in the P2MDOUT register. 0: Corresponding P2.n pin is configured for analog mode. 1: Corresponding P2.n pin is not configured for analog mode.

SFR Definition 20.22. P2MDOUT: Port 2 Output Mode

Bit	7	6	5	4	3	2	1	0
Name	P2MDOUT[7:0]							
Type	R/W							
Reset	0	0	0	0	0	0	0	0

SFR Address = 0xA6; SFR Page = 0x0F

Bit	Name	Function
7:0	P2MDOUT[7:0]	**Output Configuration Bits for P2.7–P2.0 (respectively).** These bits are ignored if the corresponding bit in register P2MDIN is logic 0. 0: Corresponding P2.n Output is open-drain. 1: Corresponding P2.n Output is push-pull.

21.1. Software Interface with the LIN Controller

The selection of the mode (Master or Slave) and the automatic baud rate feature are done though the LIN0 Control Mode (LIN0CF) register. The other LIN registers are accessed indirectly through the two SFRs LIN0 Address (LIN0ADR) and LIN0 Data (LIN0DAT). The LIN0ADR register selects which LIN register is targeted by reads/writes of the LIN0DAT register. The full list of indirectly-accessible LIN registers is given in Table 21.4 on page 210.

21.2. LIN Interface Setup and Operation

The hardware based LIN controller allows for the implementation of both Master and Slave nodes with minimal firmware overhead and complete control of the interface status while allowing for interrupt and polled mode operation.

The first step to use the controller is to define the basic characteristics of the node:

Mode—Master or Slave

Baud Rate—Either defined manually or using the autobaud feature (slave mode only)

Checksum Type—Select between classic or enhanced checksum, both of which are implemented in hardware.

21.2.1. Mode Definition

Following the LIN specification, the controller implements in hardware both the Slave and Master operating modes. The mode is configured using the MODE bit (LIN0CF.6).

21.2.2. Baud Rate Options: Manual or Autobaud

The LIN controller can be selected to have its baud rate calculated manually or automatically. A master node must always have its baud rate set manually, but slave nodes can choose between a manual or automatic setup. The configuration is selected using the ABAUD bit (LIN0CF.5).

Both the manual and automatic baud rate configurations require additional setup. The following sections explain the different options available and their relation with the baud rate, along with the steps necessary to achieve the required baud rate.

21.2.3. Baud Rate Calculations: Manual Mode

The baud rate used by the LIN controller is a function of the System Clock (SYSCLK) and the LIN timing registers according to the following equation:

$$\text{baud\_rate} = \frac{\text{SYSCLK}}{2^{(\text{prescaler} + 1)} \times \text{divider} \times (\text{multiplier} + 1)}$$

The prescaler, divider and multiplier factors are part of the LIN0DIV and LIN0MUL registers and can assume values in the following range:

Table 21.1. Baud Rate Calculation Variable Ranges

Factor	Range
prescaler	0…3
multiplier	0…31
divider	200…511

Important Note: The minimum system clock (SYSCLK) to operate the LIN controller is 8 MHz.

1. Check the DONE bit (LIN0ST.0) and the ERROR bit (LIN0ST.2).
2. If performing a master receive operation and the transfer was successful, read the received data from the data buffer.
3. If the transfer was not successful, check the error register to determine the kind of error. Further error handling has to be done by the application.
4. Set the RSTINT (LIN0CTRL.3) and RSTERR bits (LIN0CTRL.2) to reset the interrupt request and the error flags.

21.4. LIN Slave Mode Operation

When the device is configured for slave mode operation, it must wait for a command from a master node. Access from the firmware to the data buffer and ID registers of the LIN controller is only possible when a data request is pending (DTREQ bit (LIN0ST.4) is 1) and also when the LIN bus is not active (ACTIVE bit (LIN0ST.7) is set to 0).

The LIN controller in slave mode detects the header of the message frame sent by the LIN master. If slave synchronization is enabled (autobaud), the slave synchronizes its internal bit time to the master bit time.

The LIN controller configured for slave mode will generated an interrupt in one of three situations:

1. After the reception of the IDENTIFIER FIELD
2. When an error is detected
3. When the message transfer is completed.

The application should perform the following steps when an interrupt is detected:

1. Check the status of the DTREQ bit (LIN0ST.4). This bit is set when the IDENTIFIER FIELD has been received.
2. If DTREQ (LIN0ST.4) is set, read the identifier from LIN0ID and process it. If DTREQ (LIN0ST.4) is not set, continue to step 7.
3. Set the TXRX bit (LIN0CTRL.5) to 1 if the current frame is a transmit operation for the slave and set to 0 if the current frame is a receive operation for the slave.
4. Load the data length into LIN0SIZE.
5. For a slave transmit operation, load the data to transmit into the data buffer.
6. Set the DTACK bit (LIN0CTRL.4). Continue to step 10.
7. If DTREQ (LIN0ST.4) is not set, check the DONE bit (LIN0ST.0). The transmission was successful if the DONE bit is set.
8. If the transmission was successful and the current frame was a receive operation for the slave, load the received data bytes from the data buffer.
9. If the transmission was not successful, check LIN0ERR to determine the nature of the error. Further error handling has to be done by the application.
10. Set the RSTINT (LIN0CTRL.3) and RSTERR bits (LIN0CTRL.2) to reset the interrupt request and the error flags.

In addition to these steps, the application should be aware of the following:

1. If the current frame is a transmit operation for the slave, steps 1 through 5 must be completed during the IN-FRAME RESPONSE SPACE. If it is not completed in time, a timeout will be detected by the master.
2. If the current frame is a receive operation for the slave, steps 1 through 5 have to be finished until the reception of the first byte after the IDENTIFIER FIELD. Otherwise, the internal receive buffer of the LIN controller will be overwritten and a timeout error will be detected in the LIN controller.
3. The LIN controller does not directly support LIN Version 1.3 Extended Frames. If the application detects an unknown identifier (e.g. extended identifier), it has to write a 1 to the STOP bit (LIN0CTRL.7) instead

21.7.2. LIN Indirect Access SFR Registers Definitions

Table 21.4 lists the 15 indirect registers used to configured and communicate with the LIN controller.

Table 21.4. LIN Registers (Indirectly Addressable)

Name	Address	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
LIN0DT1	0x00	DATA1[7:0]							
LIN0DT2	0x01	DATA2[7:0]							
LIN0DT3	0x02	DATA3[7:0]							
LIN0DT4	0x03	DATA4[7:0]							
LIN0DT5	0x04	DATA5[7:0]							
LIN0DT6	0x05	DATA67:0]							
LIN0DT7	0x06	DATA7[7:0]							
LIN0DT8	0x07	DATA8[7:0]							
LIN0CTRL	0x08	STOP(s)	SLEEP(s)	TXRX	DTACK(s)	RSTINT	RSTERR	WUPREQ	STREQ(m)
LIN0ST	0x09	ACTIVE	IDLTOUT	ABORT(s)	DTREQ(s)	LININT	ERROR	WAKEUP	DONE
LIN0ERR	0x0A				SYNCH(s)	PRTY(s)	TOUT	CHK	BITERR
LIN0SIZE	0x0B	ENHCHK				LINSIZE[3:0]			
LIN0DIV	0x0C	DIVLSB[7:0]							
LIN0MUL	0x0D	PRESCL[1:0]		LINMUL[4:0]					DIV9
LIN0ID	0x0E			ID5	ID4	ID3	ID2	ID1	ID0

Note: These registers are used in both master and slave mode. The register bits marked with (m) are accessible only in Master mode while the register bits marked with (s) are accessible only in slave mode. All other registers are accessible in both modes.

LIN Register Definition 21.6. LIN0ST: LIN0 Status Register

Bit	7	6	5	4	3	2	1	0
Name	ACTIVE	IDLTOUT	ABORT	DTREQ	LININT	ERROR	WAKEUP	DONE
Type	R	R	R	R	R	R	R	R
Reset	0	0	0	0	0	0	0	0

Indirect Address = 0x09

Bit	Name	Function
7	ACTIVE	**LIN Active Indicator Bit.** 0: No transmission activity detected on the LIN bus. 1: Transmission activity detected on the LIN bus.
6	IDLT	**Bus Idle Timeout Bit. (slave mode only)** 0: The bus has not been idle for four seconds. 1: No bus activity has been detected for four seconds, but the bus is not yet in Sleep mode.
5	ABORT	**Aborted Transmission Bit. (slave mode only)** 0: The current transmission has not been interrupted or stopped. This bit is reset to 0 after receiving a SYNCH BREAK that does not interrupt a pending transmission. 1: New SYNCH BREAK detected before the end of the last transmission or the STOP bit (LIN0CTRL.7) has been set.
4	DTREQ	**Data Request Bit. (slave mode only)** 0: Data identifier has not been received. 1: Data identifier has been received.
3	LININT	**Interrupt Request Bit.** 0: An interrupt is not pending. This bit is cleared by setting RSTINT (LIN0CTRL.3) 1: There is a pending LIN0 interrupt.
2	ERROR	**Communication Error Bit.** 0: No error has been detected. This bit is cleared by setting RSTERR (LIN0CTRL.2) 1: An error has been detected.
1	WAKEUP	**Wakeup Bit.** 0: A wakeup signal is not being transmitted and has not been received. 1: A wakeup signal is being transmitted or has been received
0	DONE	**Transmission Complete Bit.** 0: A transmission is not in progress or has not been started. This bit is cleared at the start of a transmission. 1: The current transmission is complete.

LIN Register Definition 21.7. LIN0ERR: LIN0 Error Register

Bit	7	6	5	4	3	2	1	0
Name				SYNCH	PRTY	TOUT	CHK	BITERR
Type	R	R	R	R	R	R	R	R
Reset	0	0	0	0	0	0	0	0

Indirect Address = 0x0A

Bit	Name	Function
7:5	Unused	Read = 000b; Write = Don’t Care
4	SYNCH	**Synchronization Error Bit (slave mode only).** 0: No error with the SYNCH FIELD has been detected. 1: Edges of the SYNCH FIELD are outside of the maximum tolerance.
3	PRTY	**Parity Error Bit (slave mode only).** 0: No parity error has been detected. 1: A parity error has been detected.
2	TOUT	**Timeout Error Bit.** 0: A timeout error has not been detected. 1: A timeout error has been detected. This error is detected whenever one of the following conditions is met: • The master is expecting data from a slave and the slave does not respond. • The slave is expecting data but no data is transmitted on the bus. • A frame is not finished within the maximum frame length. • The application does not set the DTACK bit (LIN0CTRL.4) or STOP bit (LIN0CTRL.7) until the end of the reception of the first byte after the identifier.
1	CHK	**Checksum Error Bit.** 0: Checksum error has not been detected. 1: Checksum error has been detected.
0	BITERR	**Bit Transmission Error Bit.** 0: No error in transmission has been detected. 1: The bit value monitored during transmission is different than the bit value sent.

23.1. Supporting Documents

It is assumed the reader is familiar with or has access to the following supporting documents:

1. The I^{2}C-Bus and How to Use It (including specifications), Philips Semiconductor.
2. The I^{2}C-Bus Specification—Version 2.0, Philips Semiconductor.
3. System Management Bus Specification—Version 1.1, SBS Implementers Forum.

23.2. SMBus Configuration

Figure 23.2 shows a typical SMBus configuration. The SMBus specification allows any recessive voltage between 3.0 V and 5.0 V; different devices on the bus may operate at different voltage levels. The bi-directional SCL (serial clock) and SDA (serial data) lines must be connected to a positive power supply voltage through a pullup resistor or similar circuit. Every device connected to the bus must have an open-drain or open-collector output for both the SCL and SDA lines, so that both are pulled high (recessive state) when the bus is free. The maximum number of devices on the bus is limited only by the requirement that the rise and fall times on the bus not exceed 300 ns and 1000 ns, respectively.

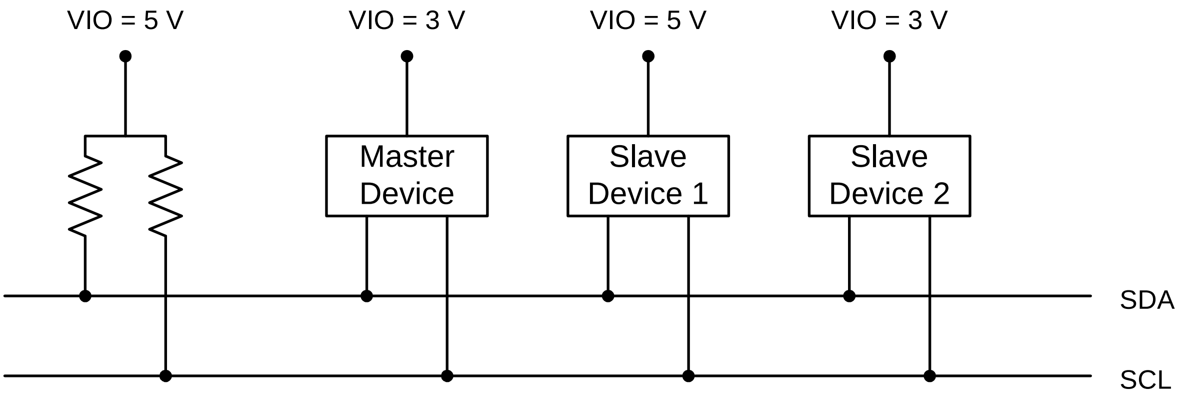


Figure 23.2. Typical SMBus Configuration

23.3. SMBus Operation

Two types of data transfers are possible: data transfers from a master transmitter to an addressed slave receiver (WRITE), and data transfers from an addressed slave transmitter to a master receiver (READ). The master device initiates both types of data transfers and provides the serial clock pulses on SCL. The SMBus interface may operate as a master or a slave, and multiple master devices on the same bus are supported. If two or more masters attempt to initiate a data transfer simultaneously, an arbitration scheme is employed with a single master always winning the arbitration. It is not necessary to specify one device as the Master in a system; any device who transmits a START and a slave address becomes the master for the duration of that transfer.

A typical SMBus transaction consists of a START condition followed by an address byte (Bits7–1: 7-bit slave address; Bit0: R/W direction bit), one or more bytes of data, and a STOP condition. Bytes that are received (by a master or slave) are acknowledged (ACK) with a low SDA during a high SCL (see Figure 23.3). If the receiving device does not ACK, the transmitting device will read a NACK (not acknowledge), which is a high SDA during a high SCL.

The direction bit (R/W) occupies the least-significant bit position of the address byte. The direction bit is set to logic 1 to indicate a "READ" operation and cleared to logic 0 to indicate a "WRITE" operation.

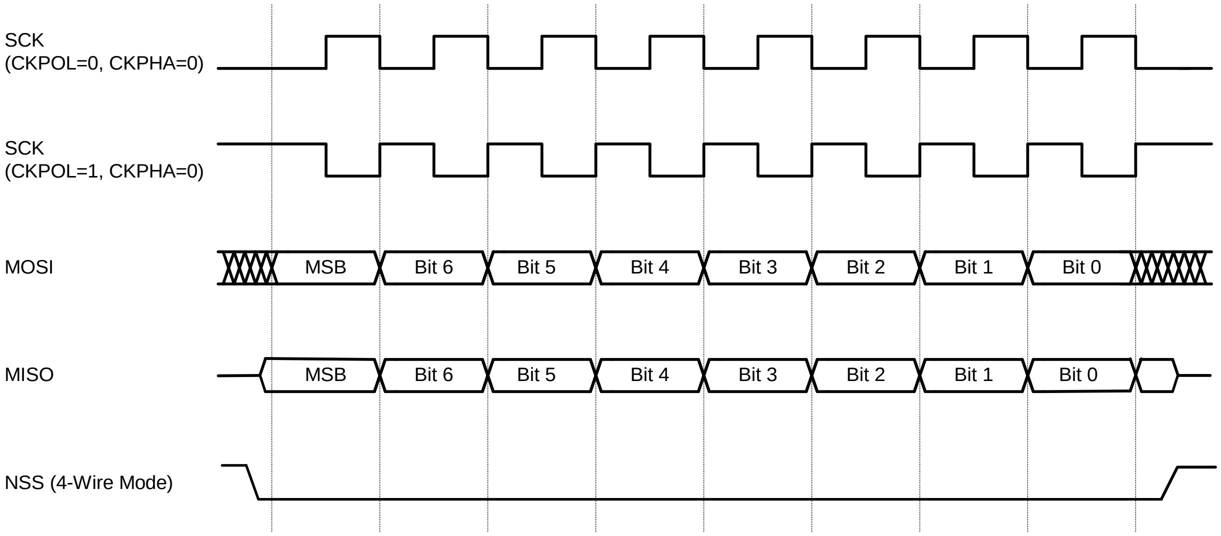


Figure 25.6. Slave Mode Data/Clock Timing (CKPHA = 0)

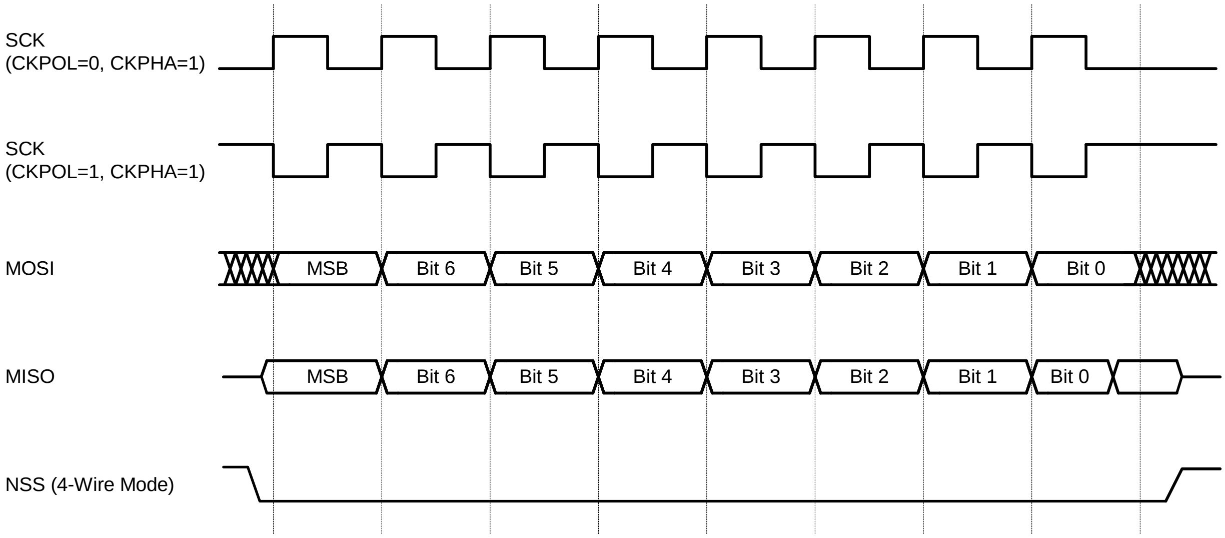


Figure 25.7. Slave Mode Data/Clock Timing (CKPHA = 1)

25.6. SPI Special Function Registers

SPI0 is accessed and controlled through four special function registers in the system controller: SPI0CN Control Register, SPI0DAT Data Register, SPI0CFG Configuration Register, and SPI0CKR Clock Rate Register. The four special function registers related to the operation of the SPI0 Bus are described in the following figures.

SFR Definition 25.1. SPI0CFG: SPI0 Configuration

Bit	**7**	**6**	**5**	**4**	**3**	**2**	**1**	**0**
Name	SPIBSY	MSTEN	CKPHA	CKPOL	SLVSEL	NSSIN	SRMT	RXBMT
Type	R	R/W	R/W	R/W	R	R	R	R
Reset	0	0	0	0	0	1	1	1

SFR Address = 0xA1; SFR Page = 0x00

Bit	Name	Function
7	SPIBSY	**SPI Busy.** This bit is set to logic 1 when a SPI transfer is in progress (master or slave mode).
6	MSTEN	**Master Mode Enable.** 0: Disable master mode. Operate in slave mode. 1: Enable master mode. Operate as a master.
5	CKPHA	**SPI0 Clock Phase.** 0: Data centered on first edge of SCK period.* 1: Data centered on second edge of SCK period.*
4	CKPOL	**SPI0 Clock Polarity.** 0: SCK line low in idle state. 1: SCK line high in idle state.
3	SLVSEL	**Slave Selected Flag.** This bit is set to logic 1 whenever the NSS pin is low indicating SPI0 is the selected slave. It is cleared to logic 0 when NSS is high (slave not selected). This bit does not indicate the instantaneous value at the NSS pin, but rather a de-glitched version of the pin input.
2	NSSIN	**NSS Instantaneous Pin Input.** This bit mimics the instantaneous value that is present on the NSS port pin at the time that the register is read. This input is not de-glitched.
1	SRMT	**Shift Register Empty (valid in slave mode only).** This bit will be set to logic 1 when all data has been transferred in/out of the shift register, and there is no new information available to read from the transmit buffer or write to the receive buffer. It returns to logic 0 when a data byte is transferred to the shift register from the transmit buffer or by a transition on SCK. SRMT = 1 when in Master Mode.
0	RXBMT	**Receive Buffer Empty (valid in slave mode only).** This bit will be set to logic 1 when the receive buffer has been read and contains no new information. If there is new information available in the receive buffer that has not been read, this bit will return to logic 0. RXBMT = 1 when in Master Mode.

Note: In slave mode, data on MOSI is sampled in the center of each data bit. In master mode, data on MISO is sampled one SYSCLK before the end of each data bit, to provide maximum settling time for the slave device. See Table 25.1 for timing parameters.

T3MH	T3XCLK	TMR3H Clock Source
0	0	SYSCLK/12
0	1	External Clock/8
1	X	SYSCLK

T3ML	T3XCLK	TMR3L Clock Source
0	0	SYSCLK/12
0	1	External Clock/8
1	X	SYSCLK

The TF3H bit is set when TMR3H overflows from 0xFF to 0x00; the TF3L bit is set when TMR3L overflows from 0xFF to 0x00. When Timer 3 interrupts are enabled, an interrupt is generated each time TMR3H overflows. If Timer 3 interrupts are enabled and TF3LEN (TMR3CN.5) is set, an interrupt is generated each time either TMR3L or TMR3H overflows. When TF3LEN is enabled, software must check the TF3H and TF3L flags to determine the source of the Timer 3 interrupt. The TF3H and TF3L interrupt flags are not cleared by hardware and must be manually cleared by software.

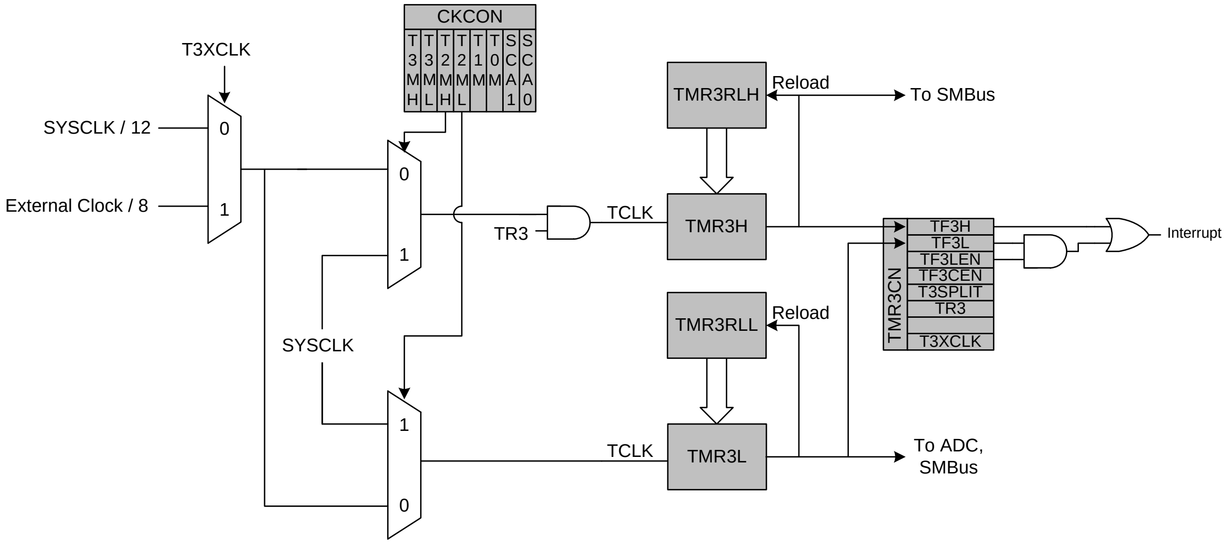


Figure 26.8. Timer 3 8-Bit Mode Block Diagram

26.3.3. External Oscillator Capture Mode

Capture Mode allows the external oscillator to be measured against the system clock. Timer 3 can be clocked from the system clock, or the system clock divided by 12, depending on the T3ML (CKCON.6), and T3XCLK bits. When a capture event is generated, the contents of Timer 3 (TMR3H:TMR3L) are loaded into the Timer 3 reload registers (TMR3RLH:TMR3RLL) and the TF3H flag is set. A capture event is generated by the falling edge of the clock source being measured, which is the external oscillator/8. By recording the difference between two successive timer capture values, the external oscillator frequency can be determined with respect to the Timer 3 clock. The Timer 3 clock should be much faster than the capture clock to achieve an accurate reading. Timer 3 should be in 16-bit auto-reload mode when using Capture Mode.

If the SYSCLK is 24 MHz and the difference between two successive captures is 5861, then the external clock frequency is as follows:

24 MHz/(5861/8) = 0.032754 MHz or 32.754 kHz

This mode allows software to determine the external oscillator frequency when an RC network or capacitor is used to generate the clock source.

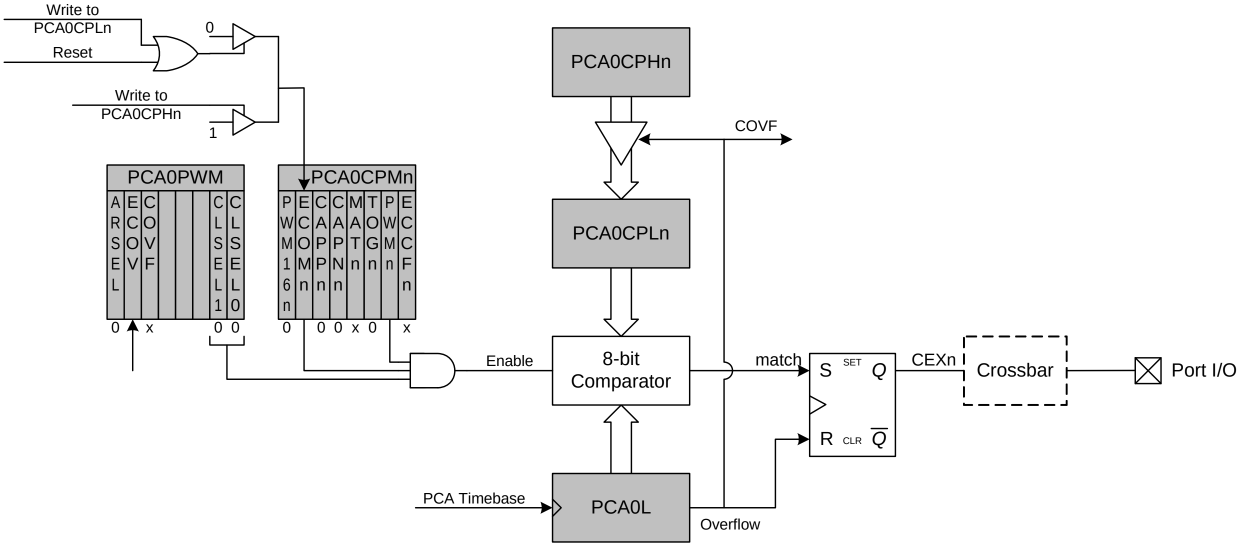


Figure 27.8. PCA 8-Bit PWM Mode Diagram

27.3.5.2. 9/10/11-bit Pulse Width Modulator Mode

The duty cycle of the PWM output signal in 9/10/11-bit PWM mode should be varied by writing to an “Auto-Reload” Register, which is dual-mapped into the PCA0CPHn and PCA0CPLn register locations. The data written to define the duty cycle should be right-justified in the registers. The auto-reload registers are accessed (read or written) when the bit ARSEL in PCA0PWM is set to 1. The capture/compare registers are accessed when ARSEL is set to 0.

When the least-significant N bits of the PCA0 counter match the value in the associated module’s capture/compare register (PCA0CPn), the output on CEXn is asserted high. When the counter overflows from the Nth bit, CEXn is asserted low (see Figure 27.9). Upon an overflow from the Nth bit, the COVF flag is set, and the value stored in the module’s auto-reload register is loaded into the capture/compare register. The value of N is determined by the CLSEL bits in register PCA0PWM.

The 9, 10 or 11-bit PWM mode is selected by setting the ECOMn and PWMn bits in the PCA0CPMn register, and setting the CLSEL bits in register PCA0PWM to the desired cycle length (other than 8-bits). If the MATn bit is set to 1, the CCFn flag for the module will be set each time a comparator match (rising edge) occurs. The COVF flag in PCA0PWM can be used to detect the overflow (falling edge), which will occur every 512 (9-bit), 1024 (10-bit) or 2048 (11-bit) PCA clock cycles. The duty cycle for 9/10/11-Bit PWM Mode is given in Equation 27.2, where N is the number of bits in the PWM cycle.

Important Note About PCA0CPHn and PCA0CPLn Registers: When writing a 16-bit value to the PCA0CPn registers, the low byte should always be written first. Writing to PCA0CPLn clears the ECOMn bit to 0; writing to PCA0CPHn sets ECOMn to 1.

$$\text{Duty Cycle} = \frac{(2^N - \text{PCA0CPn})}{2^N}$$

Equation 27.3. 9, 10, and 11-Bit PWM Duty Cycle

A 0% duty cycle may be generated by clearing the ECOMn bit to 0.