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### What is "[Embedded - Microcontrollers](#)"?

"[Embedded - Microcontrollers](#)" refer to small, integrated circuits designed to perform specific tasks within larger systems. These microcontrollers are essentially compact computers on a single chip, containing a processor core, memory, and programmable input/output peripherals. They are called "embedded" because they are embedded within electronic devices to control various functions, rather than serving as standalone computers. Microcontrollers are crucial in modern electronics, providing the intelligence and control needed for a wide range of applications.

### Applications of "[Embedded - Microcontrollers](#)"

#### Details

Product Status	Active
Core Processor	8051
Core Size	8-Bit
Speed	50MHz
Connectivity	EBI/EMI, SMBus (2-Wire/I <sup>2</sup> C), SPI, UART/USART
Peripherals	POR, PWM, Temp Sensor, WDT
Number of I/O	33
Program Memory Size	32KB (32K x 8)
Program Memory Type	FLASH
EEPROM Size	-
RAM Size	4.25K x 8
Voltage - Supply (Vcc/Vdd)	1.8V ~ 5.25V
Data Converters	A/D 32x12b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 125°C (TA)
Mounting Type	Surface Mount
Package / Case	40-VFQFN Exposed Pad
Supplier Device Package	40-QFN (6x6)
Purchase URL	<a href="https://www.e-xfl.com/product-detail/silicon-labs/c8051f511-imr">https://www.e-xfl.com/product-detail/silicon-labs/c8051f511-imr</a>

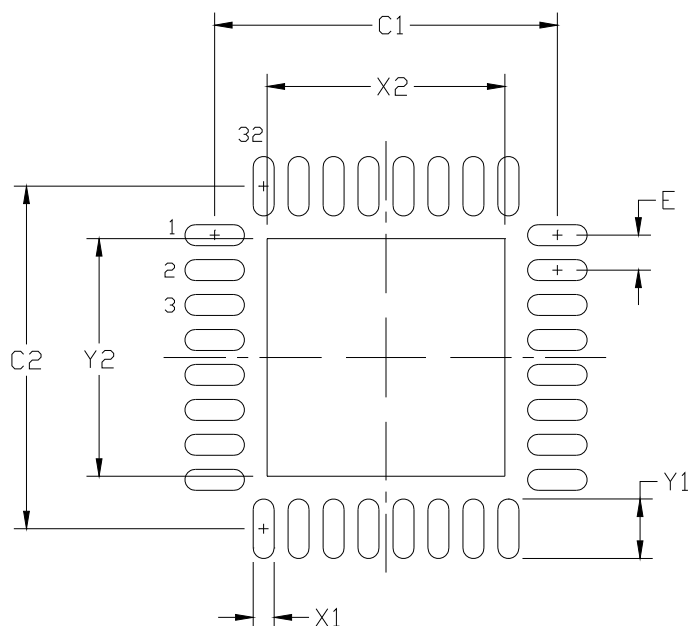


# C8051F50x/F51x

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12.2.1.3. Stack .....	99
<b>13. Special Function Registers.....</b>	<b>100</b>
13.1. SFR Paging .....	100
13.2. Interrupts and SFR Paging .....	100
13.3. SFR Page Stack Example .....	101
<b>14. Interrupts .....</b>	<b>117</b>
14.1. MCU Interrupt Sources and Vectors.....	117
14.1.1. Interrupt Priorities.....	118
14.1.2. Interrupt Latency .....	118
14.2. Interrupt Register Descriptions .....	120
14.3. External Interrupts INT0 and INT1.....	126
<b>15. Flash Memory.....</b>	<b>129</b>
15.1. Programming the Flash Memory .....	129
15.1.1. Flash Lock and Key Functions.....	129
15.1.2. Flash Erase Procedure .....	129
15.1.3. Flash Write Procedure .....	130
15.1.4. Flash Write Optimization.....	130
15.2. Non-volatile Data Storage .....	131
15.3. Security Options .....	131
15.4. Flash Write and Erase Guidelines.....	133
15.4.1. V <sub>DD</sub> Maintenance and the V <sub>DD</sub> monitor .....	133
15.4.2. PSWE Maintenance .....	133
15.4.3. System Clock .....	134
<b>16. Power Management Modes.....</b>	<b>138</b>
16.1. Idle Mode.....	138
16.2. Stop Mode .....	139
16.3. Suspend Mode .....	139
<b>17. Reset Sources .....</b>	<b>141</b>
17.1. Power-On Reset.....	142
17.2. Power-Fail Reset/VDD Monitor .....	142
17.3. External Reset.....	144
17.4. Missing Clock Detector Reset .....	144
17.5. Comparator0 Reset .....	145
17.6. PCA Watchdog Timer Reset .....	145
17.7. Flash Error Reset .....	145
17.8. Software Reset.....	145
<b>18. External Data Memory Interface and On-Chip XRAM.....</b>	<b>147</b>
18.1. Accessing XRAM.....	147
18.1.1. 16-Bit MOVX Example .....	147
18.1.2. 8-Bit MOVX Example .....	147
18.2. Configuring the External Memory Interface .....	148
18.3. Port Configuration.....	148
18.4. Multiplexed and Non-multiplexed Selection.....	153
18.4.1. Multiplexed Configuration.....	153
18.4.2. Non-multiplexed Configuration.....	154

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**Figure 4.10. QFN-32 Package Drawing**

**Table 4.10. QFN-32 Landing Diagram Dimensions**

Dimension	Min	Max	Dimension	Min	Max
C1	4.80	4.90	X2	3.20	3.40
C2	4.80	4.90	Y1	0.75	0.85
e	0.50 BSC		Y2	3.20	3.40
X1	0.20	0.30			

**Notes:**

**General**

1. All dimensions shown are in millimeters (mm) unless otherwise noted.
2. This Land Pattern Design is based on the IPC-7351 guidelines.

**Solder Mask Design**

3. All metal pads are to be non-solder mask defined (NSMD). Clearance between the solder mask and the metal pad is to be 60µm minimum, all the way around the pad.

**Stencil Design**

4. A stainless steel, laser-cut and electro-polished stencil with trapezoidal walls should be used to assure good solder paste release.
5. The stencil thickness should be 0.125mm (5 mils).
6. The ratio of stencil aperture to land pad size should be 1:1 for all perimeter pads.
7. A 3x3 array of 1.0 mm openings on a 1.20 mm pitch should be used for the center ground pad.

**Card Assembly**

8. A No-Clean, Type-3 solder paste is recommended.
9. The recommended card reflow profile is per the JEDEC/IPC J-STD-020 specification for Small Body Components.

# C8051F50x/F51x

**Table 5.2. Global Electrical Characteristics (Continued)**

–40 to +125 °C, 24 MHz system clock unless otherwise specified.

Parameter	Conditions	Min	Typ	Max	Units
<b>Digital Supply Current—CPU Inactive (Idle Mode, not fetching instructions from Flash)</b>					
$I_{DD}^4$	$V_{DD} = 2.1 \text{ V}$ , $F = 200 \text{ kHz}$	—	60	—	$\mu\text{A}$
	$V_{DD} = 2.1 \text{ V}$ , $F = 1.5 \text{ MHz}$	—	460	—	$\mu\text{A}$
	$V_{DD} = 2.1 \text{ V}$ , $F = 25 \text{ MHz}$	—	7.2	8.0	$\text{mA}$
	$V_{DD} = 2.1 \text{ V}$ , $F = 50 \text{ MHz}$	—	14	16	$\text{mA}$
$I_{DD}^4$	$V_{DD} = 2.6 \text{ V}$ , $F = 200 \text{ kHz}$	—	75	—	$\mu\text{A}$
	$V_{DD} = 2.6 \text{ V}$ , $F = 1.5 \text{ MHz}$	—	600	—	$\mu\text{A}$
	$V_{DD} = 2.6 \text{ V}$ , $F = 25 \text{ MHz}$	—	9.3	15	$\text{mA}$
	$V_{DD} = 2.6 \text{ V}$ , $F = 50 \text{ MHz}$	—	19	25	$\text{mA}$
$I_{DD}$ Supply Sensitivity <sup>4</sup>	$F = 25 \text{ MHz}$	—	57	—	$\%/V$
	$F = 1 \text{ MHz}$	—	56	—	
$I_{DD}$ Frequency Sensitivity <sup>4,6</sup>	$V_{DD} = 2.1 \text{ V}$ , $F \leq 12.5 \text{ MHz}$ , $T = 25^\circ\text{C}$	—	0.29	—	$\text{mA/MHz}$
	$V_{DD} = 2.1 \text{ V}$ , $F > 12.5 \text{ MHz}$ , $T = 25^\circ\text{C}$	—	0.29	—	
	$V_{DD} = 2.6 \text{ V}$ , $F \leq 12.5 \text{ MHz}$ , $T = 25^\circ\text{C}$	—	0.38	—	
	$V_{DD} = 2.6 \text{ V}$ , $F > 12.5 \text{ MHz}$ , $T = 25^\circ\text{C}$	—	0.38	—	
Digital Supply Current <sup>4</sup> (Stop or Suspend Mode)	Oscillator not running, $V_{DD}$ Monitor Disabled				$\mu\text{A}$
	Temp = 25 °C	—	2	—	
	Temp = 60 °C	—	10	—	
	Temp = 125 °C	—	120	—	

**Notes:**

1. Given in Table 5.4 on page 46.
2.  $V_{IO}$  should not be lower than the  $V_{DD}$  voltage.
3. SYSCLK must be at least 32 kHz to enable debugging.
4. Based on device characterization data; Not production tested. Does not include oscillator supply current.
5. IDD can be estimated for frequencies  $\leq 12.5 \text{ MHz}$  by simply multiplying the frequency of interest by the frequency sensitivity number for that range. When using these numbers to estimate  $I_{DD}$  for  $>12.5 \text{ MHz}$ , the estimate should be the current at 50 MHz minus the difference in current indicated by the frequency sensitivity number. For example:  $V_{DD} = 2.6 \text{ V}$ ;  $F = 20 \text{ MHz}$ ,  $I_{DD} = 26 \text{ mA} - (50 \text{ MHz} - 20 \text{ MHz}) \times 0.48 \text{ mA/MHz} = 11.6 \text{ mA}$ .
6. Idle IDD can be estimated for frequencies  $\leq 1 \text{ MHz}$  by simply multiplying the frequency of interest by the frequency sensitivity number for that range. When using these numbers to estimate Idle  $I_{DD}$  for  $>1 \text{ MHz}$ , the estimate should be the current at 50 MHz minus the difference in current indicated by the frequency sensitivity number.  
For example:  $V_{DD} = 2.6 \text{ V}$ ;  $F = 5 \text{ MHz}$ , Idle  $I_{DD} = 21 \text{ mA} - (50 \text{ MHz} - 5 \text{ MHz}) \times 0.41 \text{ mA/MHz} = 2.6 \text{ mA}$ .

# C8051F50x/F51x

**Table 5.6. Internal High-Frequency Oscillator Electrical Characteristics**

$V_{DD} = 1.8$  to  $2.75$  V,  $-40$  to  $+125$  °C unless otherwise specified; Using factory-calibrated settings.

Parameter	Conditions	Min	Typ	Max	Units
Oscillator Frequency	IFCN = 111b; VDD ≥ VREGMIN <sup>1</sup>	24 – 0.5%	24 <sup>2</sup>	24 + 0.5%	MHz
	IFCN = 111b; VDD < VREGMIN <sup>1</sup>	24 – 1.0%	24 <sup>2</sup>	24 + 1.0%	
Oscillator Supply Current (from V <sub>DD</sub> )	Internal Oscillator On OSCICN[7:6] = 11b	—	830	1300	μA
Internal Oscillator Suspend OSCICN[7:6] = 00b ZTCEN = 1	Temp = 25 °C	—	66	—	μA
	Temp = 85 °C	—	110	—	
	Temp = 125 °C	—	190	—	
Wake-up Time From Suspend	OSCICN[7:6] = 00b	—	1	—	μs
Power Supply Sensitivity	Constant Temperature	—	0.10	—	%/V
Temperature Sensitivity <sup>3</sup>	Constant Supply				
	TC <sub>1</sub>	—	5.0	—	ppm/°C
	TC <sub>2</sub>	—	–0.65	—	ppm/°C <sup>2</sup>
<div>1. VREGMIN is the minimum output of the voltage regulator for its low setting (REG0CN: REG0MD = 0b). See Table 5.8, “Voltage Regulator Electrical Characteristics,” on page 48.</div> <div>2. This is the average frequency across the operating temperature range.</div> <div>3. Use temperature coefficients TC<sub>1</sub> and TC<sub>2</sub> to calculate the new internal oscillator frequency using the following equation:</div> <div><math display="block">f(T) = f_0 \times (1 + TC_1 \times (T - T_0) + TC_2 \times (T - T_0)^2)</math></div> <div>where f<sub>0</sub> is the internal oscillator frequency at 25 °C and T<sub>0</sub> is 25 °C.</div>					

## SFR Definition 6.5. ADC0H: ADC0 Data Word MSB

Bit	7	6	5	4	3	2	1	0
Name	ADC0H[7:0]							
Type	R/W							
Reset	0	0	0	0	0	0	0	0

SFR Address = 0xBE; SFR Page = 0x00

Bit	Name	Function
7:0	ADC0H[7:0]	<b>ADC0 Data Word High-Order Bits.</b> For AD0LJST = 0 and AD0RPT as follows: 00: Bits 3–0 are the upper 4 bits of the 12-bit result. Bits 7–4 are 0000b. 01: Bits 4–0 are the upper 5 bits of the 14-bit result. Bits 7–5 are 000b. 10: Bits 5–0 are the upper 6 bits of the 15-bit result. Bits 7–6 are 00b. 11: Bits 7–0 are the upper 8 bits of the 16-bit result. For AD0LJST = 1 (AD0RPT must be 00): Bits 7–0 are the most-significant bits of the ADC0 12-bit result.

## SFR Definition 6.6. ADC0L: ADC0 Data Word LSB

Bit	7	6	5	4	3	2	1	0
Name	ADC0L[7:0]							
Type	R/W							
Reset	0	0	0	0	0	0	0	0

SFR Address = 0xBD; SFR Page = 0x00

Bit	Name	Function
7:0	ADC0L[7:0]	<b>ADC0 Data Word Low-Order Bits.</b> For AD0LJST = 0: Bits 7–0 are the lower 8 bits of the ADC0 Accumulated Result. For AD0LJST = 1 (AD0RPT must be '00'): Bits 7–4 are the lower 4 bits of the 12-bit result. Bits 3–0 are 0000b.

# C8051F50x/F51x

## SFR Definition 9.4. CPT1MD: Comparator1 Mode Selection

Bit	7	6	5	4	3	2	1	0
Name			CP1RIE	CP1FIE			CP1MD[1:0]	
Type	R	R	R/W	R/W	R	R	R/W	
Reset	0	0	0	0	0	0	1	0

SFR Address = 0x9E; SFR Page = 0x00

Bit	Name	Function
7:6	Unused	Read = 00b, Write = Don't Care.
5	CP1RIE	<b>Comparator1 Rising-Edge Interrupt Enable.</b> 0: Comparator1 Rising-edge interrupt disabled. 1: Comparator1 Rising-edge interrupt enabled.
4	CP1FIE	<b>Comparator1 Falling-Edge Interrupt Enable.</b> 0: Comparator1 Falling-edge interrupt disabled. 1: Comparator1 Falling-edge interrupt enabled.
3:2	Unused	Read = 00b, Write = don't care.
1:0	CP1MD[1:0]	<b>Comparator1 Mode Select.</b> These bits affect the response time and power consumption for Comparator1. 00: Mode 0 (Fastest Response Time, Highest Power Consumption) 01: Mode 1 10: Mode 2 11: Mode 3 (Slowest Response Time, Lowest Power Consumption)



## SFR Definition 9.6. CPT1MX: Comparator1 MUX Selection

Bit	7	6	5	4	3	2	1	0
Name	CMX1N[3:0]				CMX1P[3:0]			
Type	R/W				R/W			
Reset	0	1	1	1	0	1	1	1

SFR Address = 0x9F; SFR Page = 0x00

Bit	Name	Function
7:4	CMX1N[3:0]	<b>Comparator1 Negative Input MUX Selection.</b> 0000: P0.1 0001: P0.3 0010: P0.5 0011: P0.7 0100: P1.1 0101: P1.3 0110: P1.5 0111: P1.7 1000: P2.1 1001: P2.3 1010: P2.5 1011: P2.7 1100–1111: None
3:0	CMX1P[3:0]	<b>Comparator1 Positive Input MUX Selection.</b> 0000: P0.0 0001: P0.2 0010: P0.4 0011: P0.6 0100: P1.0 0101: P1.2 0110: P1.4 0111: P1.6 1000: P2.0 1001: P2.2 1010: P2.4 1011: P2.6 1100–1111: None

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## 11. CIP-51 Microcontroller

The MCU system controller core is the CIP-51 microcontroller. The CIP-51 is fully compatible with the MCS-51™ instruction set; standard 803x/805x assemblers and compilers can be used to develop software. The MCU family has a superset of all the peripherals included with a standard 8051. The CIP-51 also includes on-chip debug hardware (see description in Section 28), and interfaces directly with the analog and digital subsystems providing a complete data acquisition or control-system solution in a single integrated circuit.

The CIP-51 Microcontroller core implements the standard 8051 organization and peripherals as well as additional custom peripherals and functions to extend its capability (see Figure 11.1 for a block diagram). The CIP-51 includes the following features:

- Fully Compatible with MCS-51 Instruction Set
- 50 MIPS Peak Throughput with 50 MHz Clock
- 0 to 50 MHz Clock Frequency
- Extended Interrupt Handler
- Reset Input
- Power Management Modes
- On-chip Debug Logic
- Program and Data Memory Security

### 11.1. Performance

The CIP-51 employs a pipelined architecture that greatly increases its instruction throughput over the standard 8051 architecture. In a standard 8051, all instructions except for MUL and DIV take 12 or 24 system clock cycles to execute, and usually have a maximum system clock of 12 MHz. By contrast, the CIP-51 core executes 70% of its instructions in one or two system clock cycles, with no instructions taking more than eight system clock cycles.

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## 11.2. Instruction Set

The instruction set of the CIP-51 System Controller is fully compatible with the standard MCS-51™ instruction set. Standard 8051 development tools can be used to develop software for the CIP-51. All CIP-51 instructions are the binary and functional equivalent of their MCS-51™ counterparts, including opcodes, addressing modes and effect on PSW flags. However, instruction timing is different than that of the standard 8051.

### 11.2.1. Instruction and CPU Timing

In many 8051 implementations, a distinction is made between machine cycles and clock cycles, with machine cycles varying from 2 to 12 clock cycles in length. However, the CIP-51 implementation is based solely on clock cycle timing. All instruction timings are specified in terms of clock cycles.

Due to the pipelined architecture of the CIP-51, most instructions execute in the same number of clock cycles as there are program bytes in the instruction. Conditional branch instructions take one less clock cycle to complete when the branch is not taken as opposed to when the branch is taken. Table 11.1 is the CIP-51 Instruction Set Summary, which includes the mnemonic, number of bytes, and number of clock cycles for each instruction.

## SFR Definition 11.3. SP: Stack Pointer

Bit	7	6	5	4	3	2	1	0
Name	SP[7:0]							
Type	R/W							
Reset	0	0	0	0	0	1	1	1

SFR Address = 0x81; SFR Page = All Pages

Bit	Name	Function
7:0	SP[7:0]	<b>Stack Pointer.</b> The Stack Pointer holds the location of the top of the stack. The stack pointer is incremented before every PUSH operation. The SP register defaults to 0x07 after reset.

## SFR Definition 11.4. ACC: Accumulator

Bit	7	6	5	4	3	2	1	0
Name	ACC[7:0]							
Type	R/W							
Reset	0	0	0	0	0	0	0	0

SFR Address = 0xE0; SFR Page = All Pages; Bit-Addressable

Bit	Name	Function
7:0	ACC[7:0]	<b>Accumulator.</b> This register is the accumulator for arithmetic operations.

## SFR Definition 11.5. B: B Register

Bit	7	6	5	4	3	2	1	0
Name	B[7:0]							
Type	R/W							
Reset	0	0	0	0	0	0	0	0

SFR Address = 0xF0; SFR Page = All Pages; Bit-Addressable

Bit	Name	Function
7:0	B[7:0]	<b>B Register.</b> This register serves as a second accumulator for certain arithmetic operations.

## 11.4. Serial Number Special Function Registers (SFRs)

The C8051F50x/F51x devices include four SFRs, SN0 through SN3, that are pre-programmed during production with a unique, 32-bit serial number. The serial number provides a unique identification number for each device and can be read from the application firmware. If the serial number is not used in the application, these four registers can be used as general purpose SFRs.

### SFR Definition 11.7. SNn: Serial Number n

Bit	7	6	5	4	3	2	1	0
Name	SERNUMn[7:0]							
Type	R/W							
Reset	Varies—Unique 32-bit value							

SFR Addresses: SN0 = 0xF9; SN1 = 0xFA; SN2 = 0xFB; SN3 = 0xFC; SFR Page = 0x0F;

Bit	Name	Function
7:0	SERNUMn[7:0]	<b>Serial Number Bits.</b> The four serial number registers form a 32-bit serial number, with SN3 as the most significant byte and SN0 as the least significant byte.

# C8051F50x/F51x

**Table 13.1. Special Function Register (SFR) Memory Map for Pages 0x0 and 0xF**

Address	Page	0(8)	1(9)	2(A)	3(B)	4(C)	5(D)	6(E)	7(F)
F8	0	SPI0CN	PCA0L SN0	PCA0H SN1	PCA0CPL0 SN2	PCA0CPH0 SN3	PCACPL4	PCACPH4	VDM0CN
F0	0	B (All Pages)	P0MAT P0MDIN	P0MASK P1MDIN	P1MAT P2MDIN	P1MASK P3MDIN		EIP1 EIP1	EIP2 EIP2
E8	0	ADC0CN	PCA0CPL1	PCA0CPH1	PCA0CPL2	PCA0CPH2	PCA0CPL3	PCA0CPL3	RSTSRC
E0	0	ACC (All Pages)	XBR0	XBR1	CCH0CN	IT01CF		EIE1 (All Pages)	EIE2 (All Pages)
D8	0	PCA0CN	PCA0MD PCA0PWM	PCA0CPM0	PCA0CPM1	PCA0CPM2	PCA0CPM3	PCA0CPM4	PCA0CPM5
D0	0	PSW (All Pages)	REF0CN	LIN0DATA	LIN0ADDR	P0SKIP	P1SKIP	P2SKIP	P3SKIP
C8	0	TMR2CN	REG0CN LIN0CF	TMR2RLL	TMR2RLH	TMR2L	TMR2H	PCA0CPL5	PCA0CPH5
C0	0	SMB0CN	SMB0CF	SMB0DAT	ADC0GTL	ADC0GTH	ADC0LTL	ADC0LTH	XBR2
B8	0	IP (All Pages)		ADC0TK	ADC0MX	ADC0CF	ADC0L	ADC0H	
B0	0	P3 (All Pages)	P2MAT	P2MASK EMI0CF			P4 (All Pages)	FLSCL (All Pages)	FLKEY (All Pages)
A8	0	IE (All Pages)	SMOD0	EMI0CN EMI0TC	SBCON0	SBRLLO	SBRLH0	P3MAT P3MDOUT	P3MASK P4MDOUT
A0	0	P2 (All Pages)	SPI0CFG OSCICN	SPI0CKR OSCICRS	SPI0DAT	P0MDOUT	P1MDOUT	P2MDOUT	SFRPAGE (All Pages)
98	0	SCON0	SBUF0	CPT0CN	CPT0MD	CPT0MX	CPT1CN	CPT1MD OSCIFIN	CPT1MX OSCXCN
90	0	P1 (All Pages)	TMR3CN	TMR3RLL	TMR3RLH	TMR3L	TMR3H		CLKMUL
88	0	TCON (All Pages)	TMOD (All Pages)	TL0 (All Pages)	TL1 (All Pages)	TH0 (All Pages)	TH1 (All Pages)	CKCON (All Pages)	PSCTL CLKSEL
80	0	P0 (All Pages)	SP (All Pages)	DPL (All Pages)	DPH (All Pages)	SFR0CN	SFRNEXT (All Pages)	SFRLAST (All Pages)	PCON (All Pages)
		0(8)	1(9)	2(A)	3(B)	4(C)	5(D)	6(E)	7(F)

(bit addressable)

**SFR Definition 14.3. EIE1: Extended Interrupt Enable 1**

Bit	7	6	5	4	3	2	1	0
Name	ELIN0	ET3	ECP1	ECP0	EPCA0	EADC0	EWADC0	ESMB0
Type	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
Reset	0	0	0	0	0	0	0	0

SFR Address = 0xE6; SFR Page = All Pages

Bit	Name	Function
7	ELIN0	<b>Enable LIN0 Interrupt.</b> This bit sets the masking of the LIN0 interrupt. 0: Disable LIN0 interrupts. 1: Enable interrupt requests generated by the LIN0INT flag.
6	ET3	<b>Enable Timer 3 Interrupt.</b> This bit sets the masking of the Timer 3 interrupt. 0: Disable Timer 3 interrupts. 1: Enable interrupt requests generated by the TF3L or TF3H flags.
5	ECP1	<b>Enable Comparator1 (CP1) Interrupt.</b> This bit sets the masking of the CP1 interrupt. 0: Disable CP1 interrupts. 1: Enable interrupt requests generated by the CP1RIF or CP1FIF flags.
4	ECP0	<b>Enable Comparator0 (CP0) Interrupt.</b> This bit sets the masking of the CP0 interrupt. 0: Disable CP0 interrupts. 1: Enable interrupt requests generated by the CP0RIF or CP0FIF flags.
3	EPCA0	<b>Enable Programmable Counter Array (PCA0) Interrupt.</b> This bit sets the masking of the PCA0 interrupts. 0: Disable all PCA0 interrupts. 1: Enable interrupt requests generated by PCA0.
2	EADC0	<b>Enable ADC0 Conversion Complete Interrupt.</b> This bit sets the masking of the ADC0 Conversion Complete interrupt. 0: Disable ADC0 Conversion Complete interrupt. 1: Enable interrupt requests generated by the AD0INT flag.
1	EWADC0	<b>Enable Window Comparison ADC0 Interrupt.</b> This bit sets the masking of ADC0 Window Comparison interrupt. 0: Disable ADC0 Window Comparison interrupt. 1: Enable interrupt requests generated by ADC0 Window Compare flag (AD0WINT).
0	ESMB0	<b>Enable SMBus (SMB0) Interrupt.</b> This bit sets the masking of the SMB0 interrupt. 0: Disable all SMB0 interrupts. 1: Enable interrupt requests generated by SMB0.

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IE0 (TCON.1) and IE1 (TCON.3) serve as the interrupt-pending flags for the  $\overline{\text{INT0}}$  and  $\overline{\text{INT1}}$  external interrupts, respectively. If an  $\overline{\text{INT0}}$  or  $\overline{\text{INT1}}$  external interrupt is configured as edge-sensitive, the corresponding interrupt-pending flag is automatically cleared by the hardware when the CPU vectors to the ISR. When configured as level sensitive, the interrupt-pending flag remains logic 1 while the input is active as defined by the corresponding polarity bit (IN0PL or IN1PL); the flag remains logic 0 while the input is inactive. The external interrupt source must hold the input active until the interrupt request is recognized. It must then deactivate the interrupt request before execution of the ISR completes or another interrupt request will be generated.



## 18.5. Memory Mode Selection

The external data memory space can be configured in one of four modes, shown in Figure 18.3, based on the EMIF Mode bits in the EMI0CF register (SFR Definition 18.2). These modes are summarized below. More information about the different modes can be found in Section “18.6. Timing” on page 156.

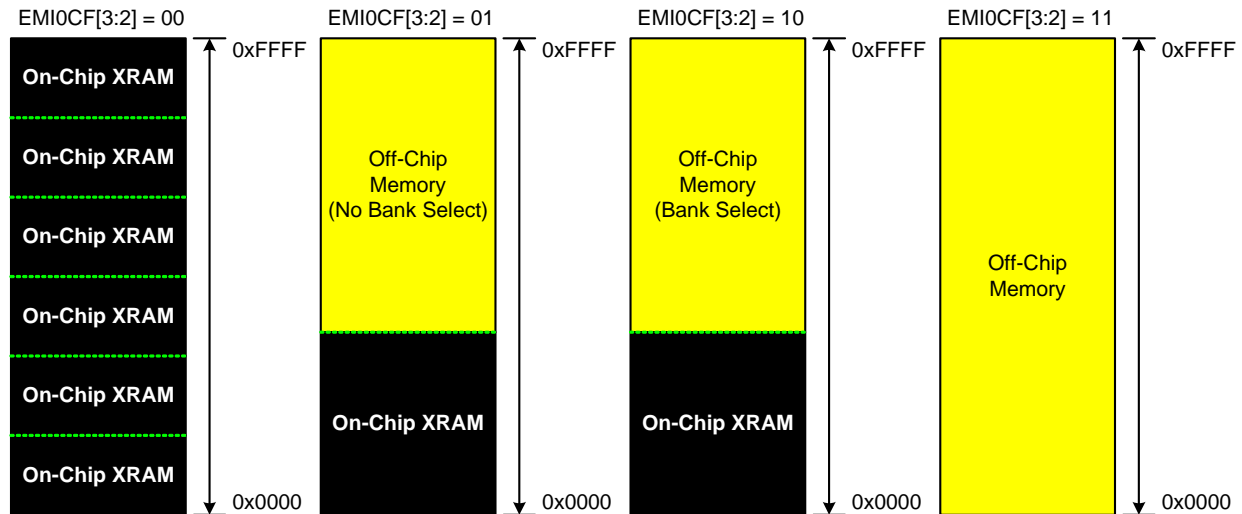


Figure 18.3. EMIF Operating Modes

### 18.5.1. Internal XRAM Only

When bits EMI0CF[3:2] are set to 00, all MOVX instructions will target the internal XRAM space on the device. Memory accesses to addresses beyond the populated space will wrap on 4 kB boundaries. As an example, the addresses 0x1000 and 0x2000 both evaluate to address 0x0000 in on-chip XRAM space.

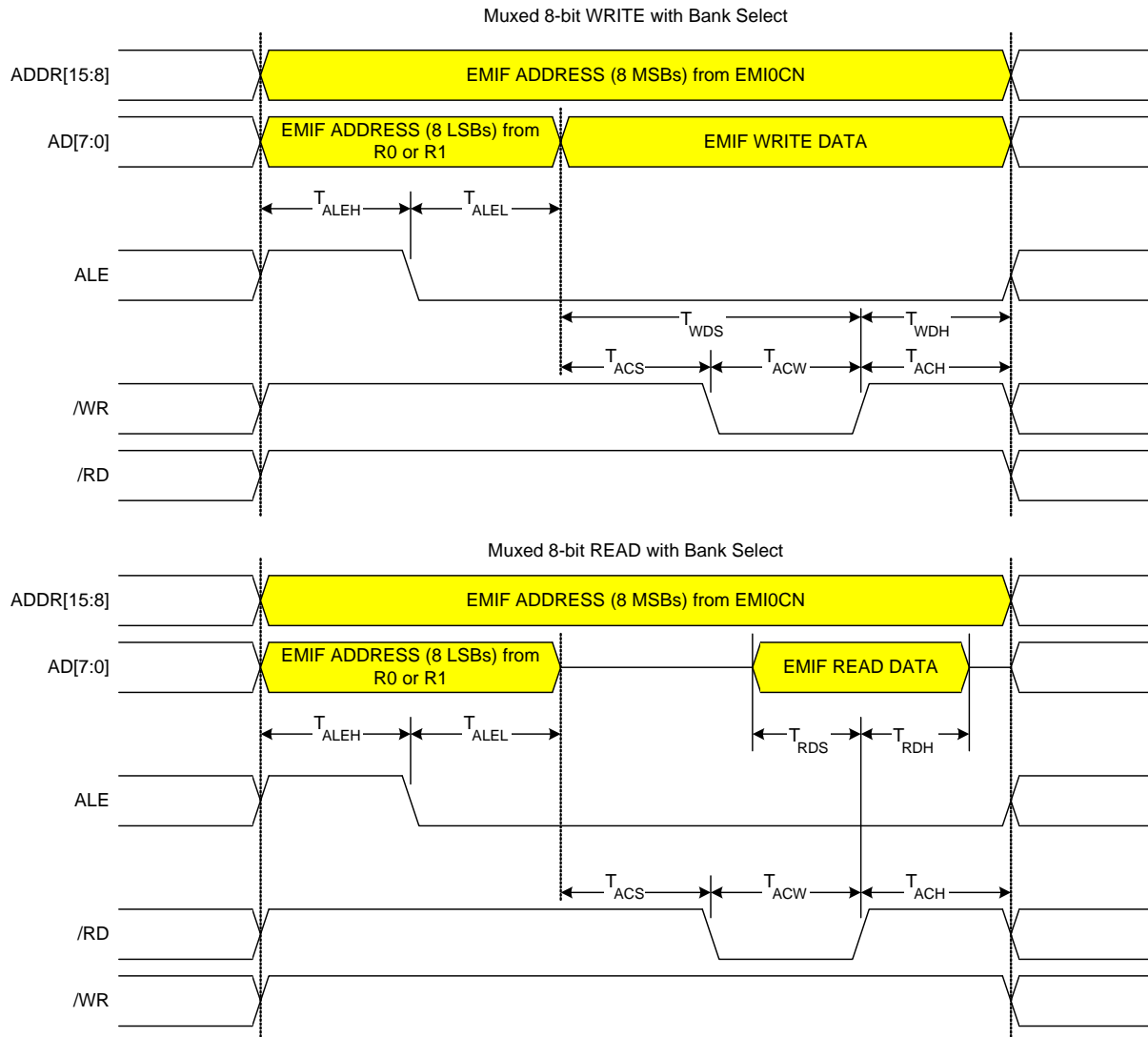
- 8-bit MOVX operations use the contents of EMI0CN to determine the high-byte of the effective address and R0 or R1 to determine the low-byte of the effective address.
- 16-bit MOVX operations use the contents of the 16-bit DPTR to determine the effective address.

### 18.5.2. Split Mode without Bank Select

When bit EMI0CF.[3:2] are set to 01, the XRAM memory map is split into two areas, on-chip space and off-chip space.

- Effective addresses below the internal XRAM size boundary will access on-chip XRAM space.
- Effective addresses above the internal XRAM size boundary will access off-chip space.
- 8-bit MOVX operations use the contents of EMI0CN to determine whether the memory access is on-chip or off-chip. However, in the “No Bank Select” mode, an 8-bit MOVX operation will not drive the upper 8-bits A[15:8] of the Address Bus during an off-chip access. This allows the user to manipulate the upper address bits at will by setting the Port state directly via the port latches. This behavior is in contrast with “Split Mode with Bank Select” described below. The lower 8-bits of the Address Bus A[7:0] are driven, determined by R0 or R1.
- 16-bit MOVX operations use the contents of DPTR to determine whether the memory access is on-chip or off-chip, and unlike 8-bit MOVX operations, the full 16-bits of the Address Bus A[15:0] are driven during the off-chip transaction.

## 18.6.2.3. 8-bit MOVX with Bank Select: EMI0CF[4:2] = 010



**Figure 18.9. Multiplexed 8-bit MOVX with Bank Select Timing**

## 20.2.3. Assigning Port I/O Pins to External Digital Event Capture Functions

External digital event capture functions can be used to trigger an interrupt or wake the device from a low power mode when a transition occurs on a digital I/O pin. The digital event capture functions do not require dedicated pins and will function on both GPIO pins (PnSKIP = 1) and pins in use by the Crossbar (PnSKIP = 0). External digital event capture functions cannot be used on pins configured for analog I/O. Table 20.3 shows all available external digital event capture functions.

**Table 20.3. Port I/O Assignment for External Digital Event Capture Functions**

Digital Function	Potentially Assignable Port Pins	SFR(s) used for Assignment
External Interrupt 0	P1.0–P1.7	IT01CF
External Interrupt 1	P1.0–P1.7	IT01CF
Port Match	P0.0–P3.7*	P0MASK, P0MAT P1MASK, P1MAT P2MASK, P2MAT P3MASK, P3MAT
<b>*Note:</b> P3.1–P3.7 are only available on the 48-pin packages.		

## 20.3. Priority Crossbar Decoder

The Priority Crossbar Decoder (Figure 20.3) assigns a priority to each I/O function, starting at the top with UART0. When a digital resource is selected, the least-significant unassigned Port pin is assigned to that resource excluding UART0, which is always assigned to pins P0.4 and P0.5, and excluding CAN0 which is always assigned to pins P0.6 and P0.7. If a Port pin is assigned, the Crossbar skips that pin when assigning the next selected resource. Additionally, the Crossbar will skip Port pins whose associated bits in the PnSKIP registers are set. The PnSKIP registers allow software to skip Port pins that are to be used for analog input, dedicated functions, or GPIO.

Because of the nature of Priority Crossbar Decoder, not all peripherals can be located on all port pins. Figure 20.3 maps peripherals to the potential port pins on which the peripheral I/O can appear.

**Important Note on Crossbar Configuration:** If a Port pin is claimed by a peripheral without use of the Crossbar, its corresponding PnSKIP bit should be set. This applies to P0.0 if VREF is used, P0.1 if the ADC is configured to use the external conversion start signal (CNVSTR), P0.3 and/or P0.2 if the external oscillator circuit is enabled, and any selected ADC or Comparator inputs. The Crossbar skips selected pins as if they were already assigned, and moves to the next unassigned pin.

## SFR Definition 20.23. P2SKIP: Port 2 Skip

Bit	7	6	5	4	3	2	1	0
Name	P2SKIP[7:0]							
Type	R/W							
Reset	0	0	0	0	0	0	0	0

SFR Address = 0xD6; SFR Page = 0x0F

Bit	Name	Function
7:0	P2SKIP[7:0]	<b>Port 2 Crossbar Skip Enable Bits.</b> These bits select Port 2 pins to be skipped by the Crossbar Decoder. Port pins used for analog, special functions or GPIO should be skipped by the Crossbar. 0: Corresponding P2.n pin is not skipped by the Crossbar. 1: Corresponding P2.n pin is skipped by the Crossbar.

## SFR Definition 20.24. P3: Port 3

Bit	7	6	5	4	3	2	1	0
Name	P3[7:0]							
Type	R/W							
Reset	1	1	1	1	1	1	1	1

SFR Address = 0xB0; SFR Page = All Pages; Bit-Addressable

Bit	Name	Description	Write	Read
7:0	P3[7:0]	<b>Port 3 Data.</b> Sets the Port latch logic value or reads the Port pin logic state in Port cells configured for digital I/O.	0: Set output latch to logic LOW. 1: Set output latch to logic HIGH.	0: P3.n Port pin is logic LOW. 1: P3.n Port pin is logic HIGH.
<b>Note:</b> Port P3.1–P3.7 are only available on the 48-pin and 40-pin packages.				

1. Check the DONE bit (LIN0ST.0) and the ERROR bit (LIN0ST.2).
2. If performing a master receive operation and the transfer was successful, read the received data from the data buffer.
3. If the transfer was not successful, check the error register to determine the kind of error. Further error handling has to be done by the application.
4. Set the RSTINT (LIN0CTRL.3) and RSTERR bits (LIN0CTRL.2) to reset the interrupt request and the error flags.

## 21.4. LIN Slave Mode Operation

When the device is configured for slave mode operation, it must wait for a command from a master node. Access from the firmware to the data buffer and ID registers of the LIN controller is only possible when a data request is pending (DTREQ bit (LIN0ST.4) is 1) and also when the LIN bus is not active (ACTIVE bit (LIN0ST.7) is set to 0).

The LIN controller in slave mode detects the header of the message frame sent by the LIN master. If slave synchronization is enabled (autobaud), the slave synchronizes its internal bit time to the master bit time.

The LIN controller configured for slave mode will generate an interrupt in one of three situations:

1. After the reception of the IDENTIFIER FIELD
2. When an error is detected
3. When the message transfer is completed.

The application should perform the following steps when an interrupt is detected:

1. Check the status of the DTREQ bit (LIN0ST.4). This bit is set when the IDENTIFIER FIELD has been received.
2. If DTREQ (LIN0ST.4) is set, read the identifier from LIN0ID and process it. If DTREQ (LIN0ST.4) is not set, continue to step 7.
3. Set the TXRX bit (LIN0CTRL.5) to 1 if the current frame is a transmit operation for the slave and set to 0 if the current frame is a receive operation for the slave.
4. Load the data length into LIN0SIZE.
5. For a slave transmit operation, load the data to transmit into the data buffer.
6. Set the DTACK bit (LIN0CTRL.4). Continue to step 10.
7. If DTREQ (LIN0ST.4) is not set, check the DONE bit (LIN0ST.0). The transmission was successful if the DONE bit is set.
8. If the transmission was successful and the current frame was a receive operation for the slave, load the received data bytes from the data buffer.
9. If the transmission was not successful, check LIN0ERR to determine the nature of the error. Further error handling has to be done by the application.
10. Set the RSTINT (LIN0CTRL.3) and RSTERR bits (LIN0CTRL.2) to reset the interrupt request and the error flags.

In addition to these steps, the application should be aware of the following:

1. If the current frame is a transmit operation for the slave, steps 1 through 5 must be completed during the IN-FRAME RESPONSE SPACE. If it is not completed in time, a timeout will be detected by the master.
2. If the current frame is a receive operation for the slave, steps 1 through 5 have to be finished until the reception of the first byte after the IDENTIFIER FIELD. Otherwise, the internal receive buffer of the LIN controller will be overwritten and a timeout error will be detected in the LIN controller.
3. The LIN controller does not directly support LIN Version 1.3 Extended Frames. If the application detects an unknown identifier (e.g. extended identifier), it has to write a 1 to the STOP bit (LIN0CTRL.7) instead