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"[Embedded - Microcontrollers](#)" refer to small, integrated circuits designed to perform specific tasks within larger systems. These microcontrollers are essentially compact computers on a single chip, containing a processor core, memory, and programmable input/output peripherals. They are called "embedded" because they are embedded within electronic devices to control various functions, rather than serving as standalone computers. Microcontrollers are crucial in modern electronics, providing the intelligence and control needed for a wide range of applications.

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Details

Product Status	Active
Core Processor	PIC
Core Size	8-Bit
Speed	40MHz
Connectivity	I ² C, SPI, UART/USART
Peripherals	Brown-out Detect/Reset, HLVD, POR, PWM, WDT
Number of I/O	25
Program Memory Size	8KB (4K x 16)
Program Memory Type	FLASH
EEPROM Size	256 x 8
RAM Size	512 x 8
Voltage - Supply (Vcc/Vdd)	4.2V ~ 5.5V
Data Converters	A/D 10x10b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	28-SSOP (0.209", 5.30mm Width)
Supplier Device Package	28-SSOP
Purchase URL	https://www.e-xfl.com/product-detail/microchip-technology/pic18f2321-i-ss

PIC18F2221/2321/4221/4321 FAMILY

TABLE 3-2: CAPACITOR SELECTION FOR QUARTZ CRYSTALS

Osc Type	Crystal Freq	Typical Capacitor Values Tested:	
		C1	C2
LP	32 kHz	22 pF	22 pF
XT	1 MHz	22 pF	22 pF
	4 MHz	22 pF	22 pF
HS	4 MHz	22 pF	22 pF
	10 MHz	22 pF	22 pF
	20 MHz	22 pF	22 pF
	25 MHz	22 pF	22 pF

Capacitor values are for design guidance only.

Different capacitor values may be required to produce acceptable oscillator operation. The user should test the performance of the oscillator over the expected VDD and temperature range for the application. Refer to the following application notes for oscillator specific information:

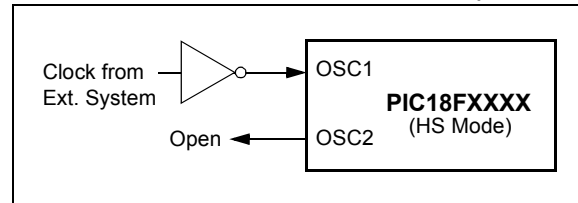
- AN588, "PIC® Microcontroller Oscillator Design Guide"
- AN826, "Crystal Oscillator Basics and Crystal Selection for *rfPIC*® and PIC® Devices"
- AN849, "Basic PIC® Oscillator Design"
- AN943, "Practical PIC® Oscillator Analysis and Design"
- AN949, "Making Your Oscillator Work"

See the notes following this table for additional information.

- Note 1:** Higher capacitance increases the stability of the oscillator but also increases the start-up time.
- 2: When operating below 3V VDD, or when using certain ceramic resonators at any voltage, it may be necessary to use the HS mode or switch to a crystal oscillator.
 - 3: Since each resonator/crystal has its own characteristics, the user should consult the resonator/crystal manufacturer for appropriate values of external components.
 - 4: Rs may be required to avoid overdriving crystals with low drive level specification.
 - 5: Always verify oscillator performance over the VDD and temperature range that is expected for the application.

An external clock source may also be connected to the OSC1 pin in the HS mode, as shown in Figure 3-2. When operated in this mode, parameters D033 and D043 apply.

FIGURE 3-2: EXTERNAL CLOCK INPUT OPERATION (HS OSC CONFIGURATION)

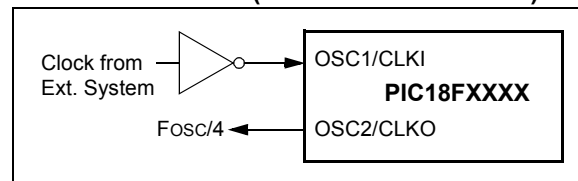


3.3 External Clock Input

The EC and ECIO Oscillator modes require an external clock source to be connected to the OSC1 pin. There is no oscillator start-up time required after a Power-on Reset or after an exit from Sleep mode.

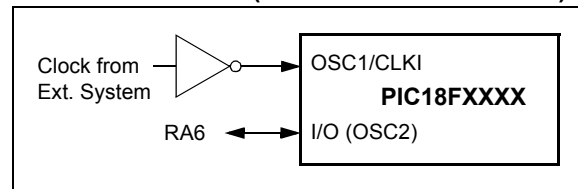
In the EC Oscillator mode, the oscillator frequency divided by 4 is available on the OSC2 pin. This signal may be used for test purposes or to synchronize other logic. Figure 3-3 shows the pin connections for the EC Oscillator mode.

FIGURE 3-3: EXTERNAL CLOCK INPUT OPERATION (EC CONFIGURATION)



The ECIO Oscillator mode functions like the EC mode, except that the OSC2 pin becomes an additional general purpose I/O pin. The I/O pin becomes bit 6 of PORTA (RA6). Figure 3-4 shows the pin connections for the ECIO Oscillator mode. When operated in this mode, parameters D033A and D043A apply.

FIGURE 3-4: EXTERNAL CLOCK INPUT OPERATION (ECIO CONFIGURATION)



PIC18F2221/2321/4221/4321 FAMILY

4.4.3 RC_IDLE MODE

In RC_IDLE mode, the CPU is disabled but the peripherals continue to be clocked from the internal oscillator block using the INTOSC multiplexer. This mode allows for controllable power conservation during Idle periods.

From RC_RUN, this mode is entered by setting the IDLEN bit and executing a `SLEEP` instruction. If the device is in another Run mode, first set IDLEN, then set the SCS1 bit and execute `SLEEP`. Although its value is ignored, it is recommended that SCS0 also be cleared; this is to maintain software compatibility with future devices. The INTOSC multiplexer may be used to select a higher clock frequency by modifying the IRCF bits before executing the `SLEEP` instruction. When the clock source is switched to the INTOSC multiplexer, the primary oscillator is shut down and the OSTS bit is cleared.

If the IRCF bits are set to any non-zero value, or the INTSRC bit is set, the INTOSC output is enabled. The IOFS bit becomes set, after the INTOSC output becomes stable, after an interval of TIOBST (parameter 39, Table 27-10). Clocks to the peripherals continue while the INTOSC source stabilizes. If the IRCF bits were previously at a non-zero value, or INTSRC was set before the `SLEEP` instruction was executed and the INTOSC source was already stable, the IOFS bit will remain set. If the IRCF bits and INTSRC are all clear, the INTOSC output will not be enabled, the IOFS bit will remain clear and there will be no indication of the current clock source.

When a wake event occurs, the peripherals continue to be clocked from the INTOSC multiplexer. After a delay of TcSD following the wake event, the CPU begins executing code being clocked by the INTOSC multiplexer. The IDLEN and SCS bits are not affected by the wake-up. The INTRC source will continue to run if either the WDT or the Fail-Safe Clock Monitor is enabled.

4.5 Exiting Idle and Sleep Modes

An exit from Sleep mode or any of the Idle modes is triggered by an interrupt, a Reset or a WDT time-out. This section discusses the triggers that cause exits from power-managed modes. The clocking subsystem actions are discussed in each of the power-managed modes (see **Section 4.2 “Run Modes”**, **Section 4.3 “Sleep Mode”** and **Section 4.4 “Idle Modes”**).

4.5.1 EXIT BY INTERRUPT

Any of the available interrupt sources can cause the device to exit from an Idle mode, or the Sleep mode to a Run mode. To enable this functionality, an interrupt source must be enabled by setting its enable bit in one of the INTCON or PIE registers. The exit sequence is initiated when the corresponding interrupt flag bit is set.

On all exits from Idle or Sleep modes by interrupt, code execution branches to the interrupt vector if the GIE/GIEH bit (INTCON<7>) is set. Otherwise, code execution continues or resumes without branching (see **Section 10.0 “Interrupts”**).

A fixed delay of interval TcSD following the wake event is required when leaving Sleep and Idle modes. This delay is required for the CPU to prepare for execution. Instruction execution resumes on the first clock cycle following this delay.

4.5.2 EXIT BY WDT TIME-OUT

A WDT time-out will cause different actions depending on which power-managed mode the device is in when the time-out occurs.

If the device is not executing code (all Idle modes and Sleep mode), the time-out will result in an exit from the power-managed mode (see **Section 4.2 “Run Modes”** and **Section 4.3 “Sleep Mode”**). If the device is executing code (all Run modes), the time-out will result in a WDT Reset (see **Section 24.2 “Watchdog Timer (WDT)”**).

The WDT timer and postscaler are cleared by executing a `SLEEP` or `CLRWDT` instruction, the loss of a currently selected clock source (if the Fail-Safe Clock Monitor is enabled) and modifying the IRCF bits in the OSCCON register if the internal oscillator block is the device clock source.

4.5.3 EXIT BY RESET

Normally, the device is held in Reset by the Oscillator Start-up Timer (OST) until the primary clock becomes ready. At that time, the OSTS bit is set and the device begins executing code. If the internal oscillator block is the new clock source, the IOFS bit is set instead.

The exit delay time from Reset to the start of code execution depends on both the clock sources before and after the wake-up and the type of oscillator if the new clock source is the primary clock. Exit delays are summarized in Table 4-2.

Code execution can begin before the primary clock becomes ready. If either the Two-Speed Start-up (see **Section 24.3 “Two-Speed Start-up”**) or Fail-Safe Clock Monitor (see **Section 24.4 “Fail-Safe Clock Monitor”**) is enabled, the device may begin execution as soon as the Reset source has cleared. Execution is clocked by the INTOSC multiplexer driven by the internal oscillator block. Execution is clocked by the internal oscillator block until either the primary clock becomes ready or a power-managed mode is entered before the primary clock becomes ready; the primary clock is then shut down.

PIC18F2221/2321/4221/4321 FAMILY

5.5 Device Reset Timers

PIC18F2221/2321/4221/4321 family devices incorporate three separate on-chip timers that help regulate the Power-on Reset process. Their main function is to ensure that the device clock is stable before code is executed. These timers are:

- Power-up Timer (PWRT)
- Oscillator Start-up Timer (OST)
- PLL Lock Time-out

5.5.1 POWER-UP TIMER (PWRT)

The Power-up Timer (PWRT) of the PIC18F2221/2321/4221/4321 family devices is an 11-bit counter which uses the INTRC source as the clock input. This yields an approximate time interval of $2048 \times 32 \mu\text{s} = 65.6 \text{ ms}$. While the PWRT is counting, the device is held in Reset.

The power-up time delay depends on the INTRC clock and will vary from chip to chip due to temperature and process variation. See DC parameter 33 for details.

The PWRT is enabled by clearing the $\overline{\text{PWRTE}}\text{N}$ Configuration bit.

5.5.2 OSCILLATOR START-UP TIMER (OST)

The Oscillator Start-up Timer (OST) provides a 1024 oscillator cycle (from OSC1 input) delay after the PWRT delay is over (parameter 33). This ensures that the crystal oscillator or resonator has started and stabilized.

The OST time-out is invoked only for XT, LP, HS and HSPLL modes and only on Power-on Reset, or on exit from most power-managed modes.

5.5.3 PLL LOCK TIME-OUT

With the PLL enabled in HSPLL mode, the time-out sequence following a Power-on Reset is slightly different from other oscillator modes. A separate timer is used to provide a fixed time-out that is sufficient for the PLL to lock to the main oscillator frequency. This PLL lock time-out (T_{PLL}) is typically 2 ms and follows the oscillator start-up time-out.

5.5.4 TIME-OUT SEQUENCE

On power-up, the time-out sequence is as follows:

1. After the POR pulse has cleared, PWRT time-out is invoked (if enabled).
2. Then, the OST is activated.

The total time-out will vary based on oscillator configuration and the status of the PWRT. Figure 5-3, Figure 5-4, Figure 5-5, Figure 5-6 and Figure 5-7 all depict time-out sequences on power-up, with the Power-up Timer enabled and the device operating in HS Oscillator mode. Figures 5-3 through 5-6 also apply to devices operating in XT or LP modes. For devices in RC mode and with the PWRT disabled, there will be no time-out at all.

Since the time-outs occur from the POR pulse, if $\overline{\text{MCLR}}$ is kept low long enough, all time-outs will expire. Bringing $\overline{\text{MCLR}}$ high will begin execution immediately (Figure 5-5). This is useful for testing purposes or to synchronize more than one PIC18FXXXX device operating in parallel.

TABLE 5-2: TIME-OUT IN VARIOUS SITUATIONS

Oscillator Configuration	Power-up ⁽²⁾ and Brown-out Reset		Exit from Power-Managed Mode
	$\overline{\text{PWRTE}}\text{N} = 0$	$\overline{\text{PWRTE}}\text{N} = 1$	
HSPLL	$66 \text{ ms}^{(1)} + 1024 \text{ TOSC} + 2 \text{ ms}^{(2)}$	$1024 \text{ TOSC} + 2 \text{ ms}^{(2)}$	$1024 \text{ TOSC} + 2 \text{ ms}^{(2)}$
HS, XT, LP	$66 \text{ ms}^{(1)} + 1024 \text{ TOSC}$	1024 TOSC	1024 TOSC
EC, ECIO	$66 \text{ ms}^{(1)}$	—	—
RC, RCIO	$66 \text{ ms}^{(1)}$	—	—
INTIO1, INTIO2	$66 \text{ ms}^{(1)}$	—	—

Note 1: 66 ms (65.5 ms) is the nominal Power-up Timer (PWRT) delay.

Note 2: 2 ms is the nominal time required for the PLL to lock.

PIC18F2221/2321/4221/4321 FAMILY

The PLUSW register can be used to implement a form of indexed addressing in the data memory space. By manipulating the value in the W register, users can reach addresses that are fixed offsets from pointer addresses. In some applications, this can be used to implement some powerful program control structure, such as software stacks, inside of data memory.

6.4.3.3 Operations by FSRs on FSRs

Indirect addressing operations that target other FSRs or virtual registers represent special cases. For example, using an FSR to point to one of the virtual registers will not result in successful operations. As a specific case, assume that FSR0H:FSR0L contains FE7h, the address of INDF1. Attempts to read the value of the INDF1 using INDF0 as an operand will return 00h. Attempts to write to INDF1 using INDF0 as the operand will result in a NOP.

On the other hand, using the virtual registers to write to an FSR pair may not occur as planned. In these cases, the value will be written to the FSR pair but without any incrementing or decrementing. Thus, writing to INDF2 or POSTDEC2 will write the same value to the FSR2H:FSR2L.

Since the FSRs are physical registers mapped in the SFR space, they can be manipulated through all direct operations. Users should proceed cautiously when working on these registers, particularly if their code uses indirect addressing.

Similarly, operations by indirect addressing are generally permitted on all other SFRs. Users should exercise the appropriate caution that they do not inadvertently change settings that might affect the operation of the device.

6.5 Data Memory and the Extended Instruction Set

Enabling the PIC18 extended instruction set (XINST Configuration bit = 1) significantly changes certain aspects of data memory and its addressing. Specifically, the use of the Access Bank for many of the core PIC18 instructions is different. This is due to the introduction of a new addressing mode for the data memory space.

What does not change is just as important. The size of the data memory space is unchanged, as well as its linear addressing. The SFR map remains the same. Core PIC18 instructions can still operate in both Direct and Indirect Addressing mode; inherent and literal instructions do not change at all. Indirect addressing with FSR0 and FSR1 also remain unchanged.

6.5.1 INDEXED ADDRESSING WITH LITERAL OFFSET

Enabling the PIC18 extended instruction set changes the behavior of indirect addressing using the FSR2 register pair within Access RAM. Under the proper conditions, instructions that use the Access Bank – that is, most bit-oriented and byte-oriented instructions – can invoke a form of indexed addressing using an offset specified in the instruction. This special addressing mode is known as Indexed Addressing with Literal Offset, or Indexed Literal Offset mode.

When using the extended instruction set, this addressing mode requires the following:

- The use of the Access Bank is forced ('a' = 0); and
- The file address argument is less than or equal to 5Fh.

Under these conditions, the file address of the instruction is not interpreted as the lower byte of an address (used with the BSR in direct addressing), or as an 8-bit address in the Access Bank. Instead, the value is interpreted as an offset value to an Address Pointer, specified by FSR2. The offset and the contents of FSR2 are added to obtain the target address of the operation.

6.5.2 INSTRUCTIONS AFFECTED BY INDEXED LITERAL OFFSET MODE

Any of the core PIC18 instructions that can use direct addressing are potentially affected by the Indexed Literal Offset Addressing mode. This includes all byte-oriented and bit-oriented instructions, or almost one-half of the standard PIC18 instruction set. Instructions that only use Inherent or Literal Addressing modes are unaffected.

Additionally, byte-oriented and bit-oriented instructions are not affected if they do not use the Access Bank (Access RAM bit is '1'), or include a file address of 60h or above. Instructions meeting these criteria will continue to execute as before. A comparison of the different possible addressing modes when the extended instruction set is enabled is shown in Figure 6-8.

Those who desire to use bit-oriented or byte-oriented instructions in the Indexed Literal Offset mode should note the changes to assembler syntax for this mode. This is described in more detail in **Section 24.2.1 "Extended Instruction Syntax"**.

PIC18F2221/2321/4221/4321 FAMILY

7.0 FLASH PROGRAM MEMORY

The Flash program memory is readable, writable and erasable during normal operation over the entire VDD range.

A read from program memory is executed on one byte at a time. A write to program memory is executed on blocks of 8 bytes at a time. Program memory is erased in blocks of 64 bytes at a time. A bulk erase operation may not be issued from user code.

Writing or erasing program memory will cease instruction fetches until the operation is complete. The program memory cannot be accessed during the write or erase, therefore, code cannot execute. An internal programming timer terminates program memory writes and erases.

A value written to program memory does not need to be a valid instruction. Executing a program memory location that forms an invalid instruction results in a NOP.

7.1 Table Reads and Table Writes

In order to read and write program memory, there are two operations that allow the processor to move bytes between the program memory space and the data RAM:

- Table Read (TBLRD)
- Table Write (TBLWT)

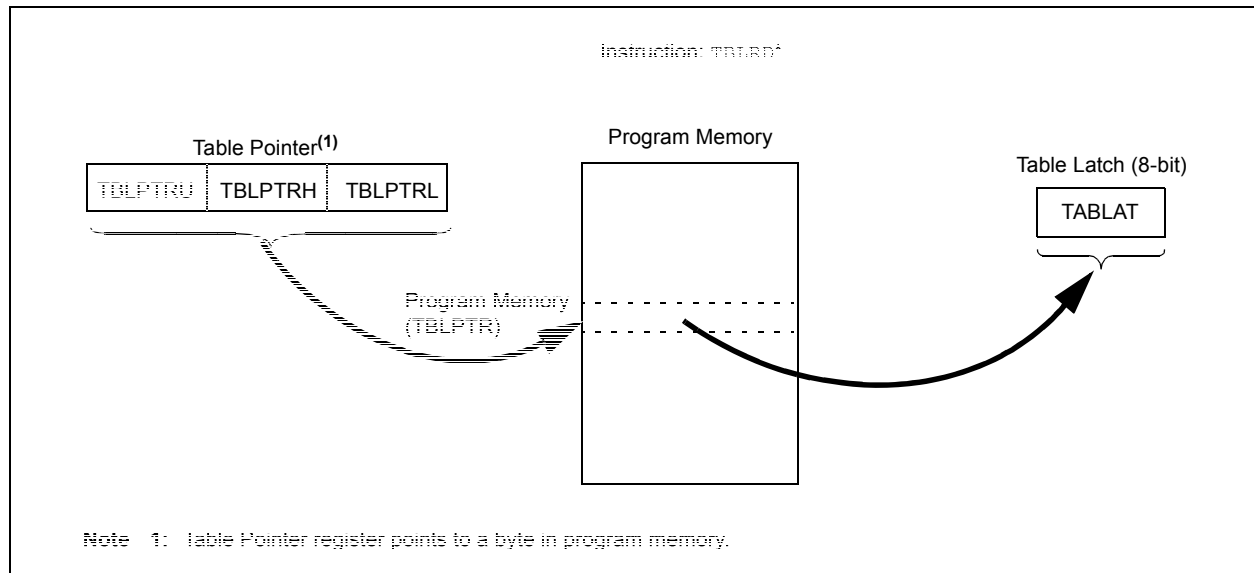
The program memory space is 16 bits wide, while the data RAM space is 8 bits wide. Table reads and table writes move data between these two memory spaces through an 8-bit register (TABLAT).

Table read operations retrieve data from program memory and place it into the data RAM space. Figure 7-1 shows the operation of a table read with program memory and data RAM.

Table write operations store data from the data memory space into holding registers in program memory. The procedure to write the contents of the holding registers into program memory is detailed in **Section 7.5 “Writing to Flash Program Memory”**. Figure 7-2 shows the operation of a table write with program memory and data RAM.

Table operations work with byte entities. A table block containing data, rather than program instructions, is not required to be word-aligned. Therefore, a table block can start and end at any byte address. If a table write is being used to write executable code into program memory, program instructions will need to be word-aligned.

FIGURE 7-1: TABLE READ OPERATION



PIC18F2221/2321/4221/4321 FAMILY

12.1 Timer0 Operation

Timer0 can operate as either a timer or a counter; the mode is selected with the T0CS bit (T0CON<5>). In Timer mode (T0CS = 0), the module increments on every clock by default unless a different prescaler value is selected (see **Section 12.3 “Prescaler”**). If the TMR0 register is written to, the increment is inhibited for the following two instruction cycles. The user can work around this by writing an adjusted value to the TMR0 register.

The Counter mode is selected by setting the T0CS bit (= 1). In this mode, Timer0 increments either on every rising or falling edge of pin RA4/T0CKI. The incrementing edge is determined by the Timer0 Source Edge Select bit, T0SE (T0CON<4>); clearing this bit selects the rising edge. Restrictions on the external clock input are discussed below.

An external clock source can be used to drive Timer0; however, it must meet certain requirements to ensure that the external clock can be synchronized with the

internal phase clock (Tosc). There is a delay between synchronization and the onset of incrementing the timer/counter.

12.2 Timer0 Reads and Writes in 16-Bit Mode

TMR0H is not the actual high byte of Timer0 in 16-bit mode; it is actually a buffered version of the real high byte of Timer0 which is not directly readable nor writable (refer to Figure 12-2). TMR0H is updated with the contents of the high byte of Timer0 during a read of TMR0L. This provides the ability to read all 16 bits of Timer0 without having to verify that the read of the high and low byte were valid, due to a rollover between successive reads of the high and low byte.

Similarly, a write to the high byte of Timer0 must also take place through the TMR0H Buffer register. The high byte is updated with the contents of TMR0H when a write occurs to TMR0L. This allows all 16 bits of Timer0 to be updated at once.

FIGURE 12-1: TIMER0 BLOCK DIAGRAM (8-BIT MODE)

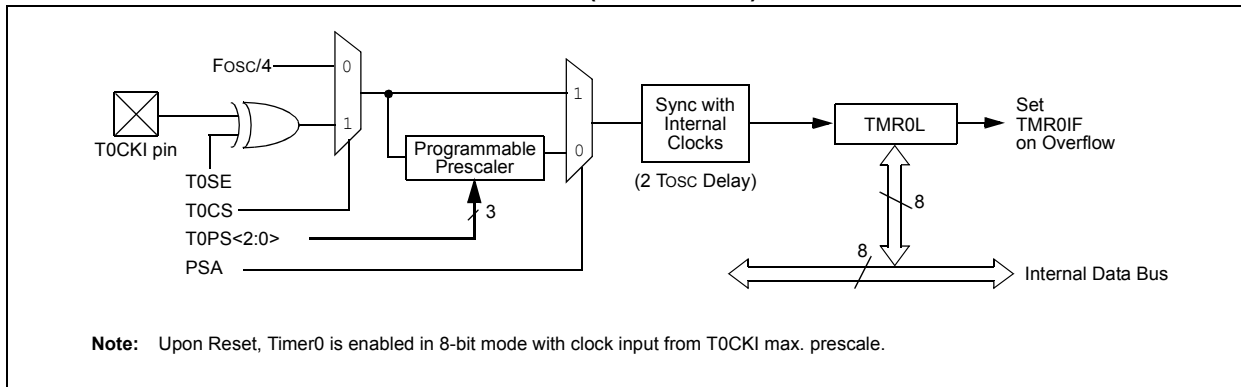
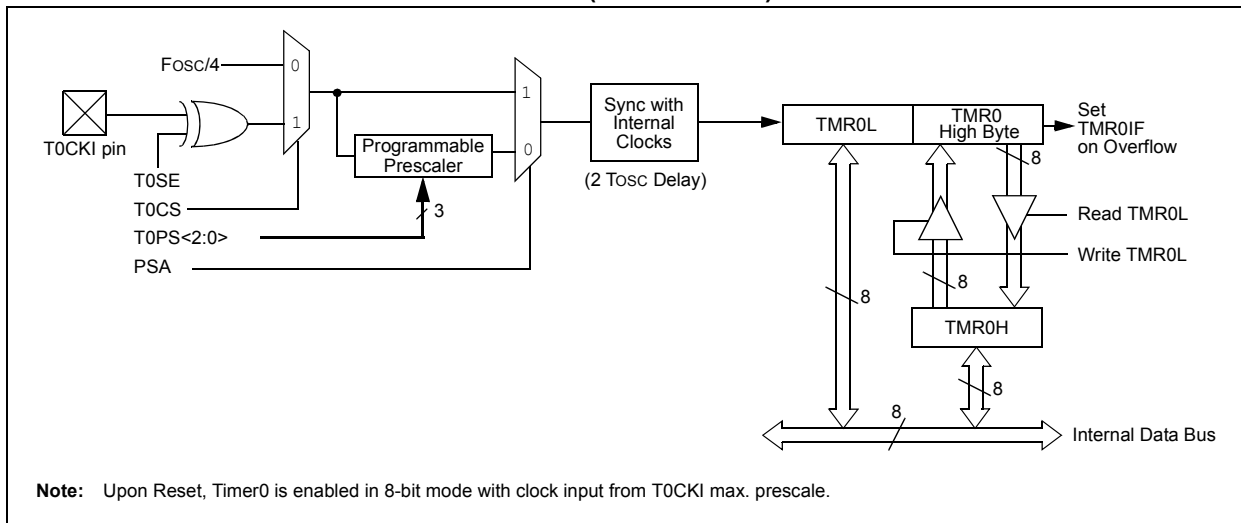


FIGURE 12-2: TIMER0 BLOCK DIAGRAM (16-BIT MODE)



PIC18F2221/2321/4221/4321 FAMILY

TABLE 16-3: REGISTERS ASSOCIATED WITH CAPTURE, COMPARE, TIMER1 AND TIMER3

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on page
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF	55
RCON	IPEN	SBOREN ⁽¹⁾	—	\overline{RI}	\overline{TO}	\overline{PD}	\overline{POR}	\overline{BOR}	54
PIR1	PSPIF ⁽²⁾	ADIF	RCIF	TXIF	SSPIF	CCP1IF	TMR2IF	TMR1IF	58
PIE1	PSPIE ⁽²⁾	ADIE	RCIE	TXIE	SSPIE	CCP1IE	TMR2IE	TMR1IE	58
IPR1	PSPIP ⁽²⁾	ADIP	RCIP	TXIP	SSPIP	CCP1IP	TMR2IP	TMR1IP	58
PIR2	OSCFIF	CMIF	—	EEIF	BCLIF	HLVDIF	TMR3IF	CCP2IF	58
PIE2	OSCFIE	CMIE	—	EEIE	BCLIE	HLVDIE	TMR3IE	CCP2IE	58
IPR2	OSCFIP	CMIP	—	EEIP	BCLIP	HLVDIP	TMR3IP	CCP2IP	58
TRISB	PORTB Data Direction Register								58
TRISC	PORTC Data Direction Register								58
TMR1L	Timer1 Register Low Byte								56
TMR1H	Timer1 Register High Byte								56
T1CON	RD16	T1RUN	T1CKPS1	T1CKPS0	T1OSCEN	$\overline{T1SYNC}$	TMR1CS	TMR1ON	56
TMR3H	Timer3 Register High Byte								57
TMR3L	Timer3 Register Low Byte								57
T3CON	RD16	T3CCP2	T3CKPS1	T3CKPS0	T3CCP1	$\overline{T3SYNC}$	TMR3CS	TMR3ON	57
CCPR1L	Capture/Compare/PWM Register 1 Low Byte								57
CCPR1H	Capture/Compare/PWM Register 1 High Byte								57
CCP1CON	P1M1 ⁽²⁾	P1M0 ⁽²⁾	DC1B1	DC1B0	CCP1M3	CCP1M2	CCP1M1	CCP1M0	57
CCPR2L	Capture/Compare/PWM Register 2 Low Byte								57
CCPR2H	Capture/Compare/PWM Register 2 High Byte								57
CCP2CON	—	—	DC2B1	DC2B0	CCP2M3	CCP2M2	CCP2M1	CCP2M0	57

Legend: — = unimplemented, read as '0'. Shaded cells are not used by Capture/Compare, Timer1 or Timer3.

Note 1: The SBOREN bit is only available when the BOREN<1:0> Configuration bits = 01; otherwise, it is disabled and reads as '0'. See **Section 5.4 “Brown-out Reset (BOR)”**.

2: These bits are unimplemented on 28-pin devices and read as '0'.

PIC18F2221/2321/4221/4321 FAMILY

18.4.6 MASTER MODE

Master mode is enabled by setting and clearing the appropriate SSPM bits in SSPCON1 and by setting the SSPEN bit. In Master mode, the SCL and SDA lines are manipulated by the MSSP hardware.

Master mode of operation is supported by interrupt generation on the detection of the Start and Stop conditions. The Stop (P) and Start (S) bits are cleared from a Reset or when the MSSP module is disabled. Control of the I²C bus may be taken when the P bit is set, or the bus is Idle, with both the S and P bits clear.

In Firmware Controlled Master mode, user code conducts all I²C bus operations based on Start and Stop bit conditions.

Once Master mode is enabled, the user has six options.

1. Assert a Start condition on SDA and SCL.
2. Assert a Repeated Start condition on SDA and SCL.
3. Write to the SSPBUF register initiating transmission of data/address.
4. Configure the I²C port to receive data.
5. Generate an Acknowledge condition at the end of a received byte of data.
6. Generate a Stop condition on SDA and SCL.

Note: The MSSP module, when configured in I²C Master mode, does not allow queueing of events. For instance, the user is not allowed to initiate a Start condition and immediately write the SSPBUF register to initiate transmission before the Start condition is complete. In this case, the SSPBUF will not be written to and the WCOL bit will be set, indicating that a write to the SSPBUF did not occur.

The following events will cause the MSSP Interrupt Flag bit, SSPIF, to be set (MSSP interrupt, if enabled):

- Start condition
- Stop condition
- Data transfer byte transmitted/received
- Acknowledge transmit
- Repeated Start

FIGURE 18-18: MSSP BLOCK DIAGRAM (I²C™ MASTER MODE)

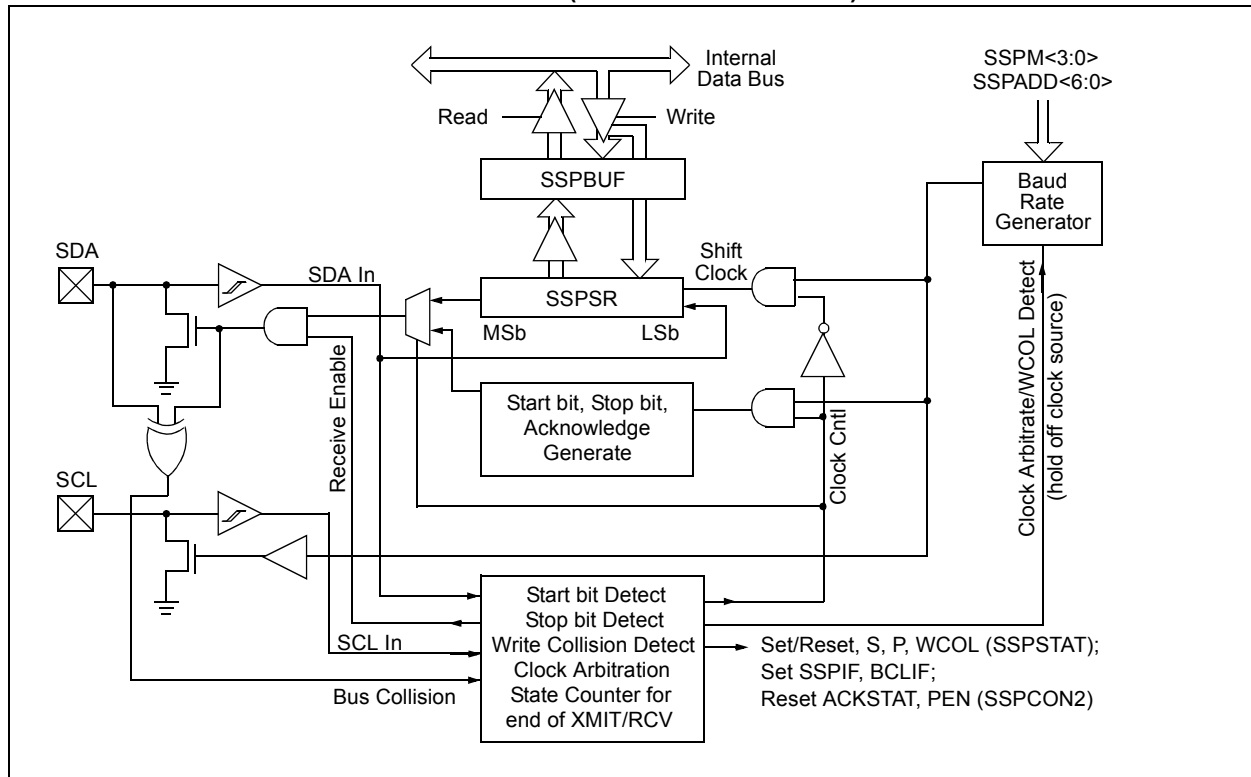
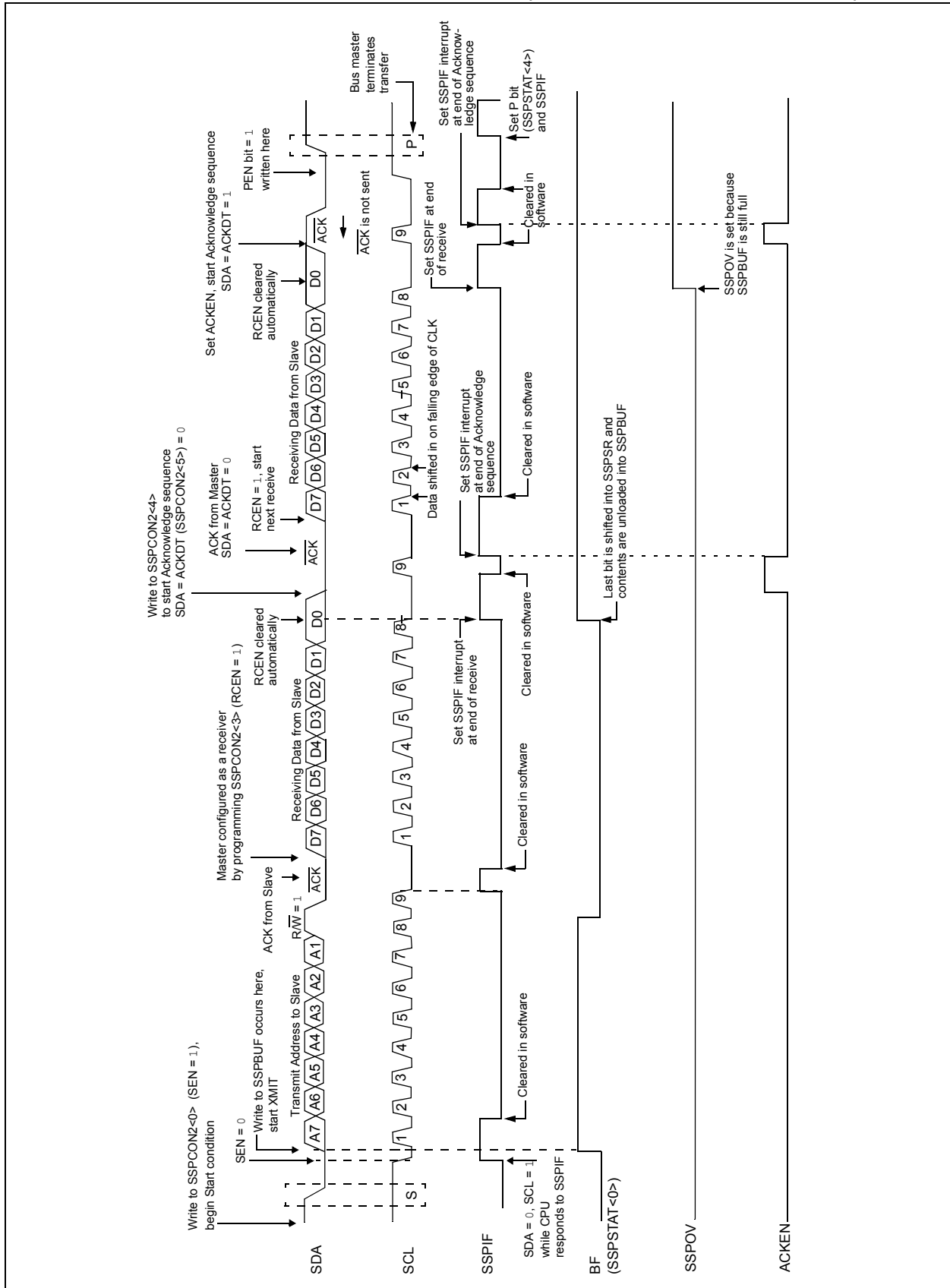


FIGURE 18-24: I²C™ MASTER MODE WAVEFORM (RECEPTION, 7-BIT ADDRESSING)



PIC18F2221/2321/4221/4321 FAMILY

TABLE 19-3: BAUD RATES FOR ASYNCHRONOUS MODES

BAUD RATE (K)	SYNC = 0, BRGH = 0, BRG16 = 0											
	Fosc = 40.000 MHz			Fosc = 20.000 MHz			Fosc = 10.000 MHz			Fosc = 8.000 MHz		
	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)
0.3	—	—	—	—	—	—	—	—	—	—	—	—
1.2	—	—	—	1.221	1.73	255	1.202	0.16	129	1.201	-0.16	103
2.4	2.441	1.73	255	2.404	0.16	129	2.404	0.16	64	2.403	-0.16	51
9.6	9.615	0.16	64	9.766	1.73	31	9.766	1.73	15	9.615	-0.16	12
19.2	19.531	1.73	31	19.531	1.73	15	19.531	1.73	7	—	—	—
57.6	56.818	-1.36	10	62.500	8.51	4	52.083	-9.58	2	—	—	—
115.2	125.000	8.51	4	104.167	-9.58	2	78.125	-32.18	1	—	—	—

BAUD RATE (K)	SYNC = 0, BRGH = 0, BRG16 = 0								
	Fosc = 4.000 MHz			Fosc = 2.000 MHz			Fosc = 1.000 MHz		
	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)
0.3	0.300	0.16	207	0.300	-0.16	103	0.300	-0.16	51
1.2	1.202	0.16	51	1.201	-0.16	25	1.201	-0.16	12
2.4	2.404	0.16	25	2.403	-0.16	12	—	—	—
9.6	8.929	-6.99	6	—	—	—	—	—	—
19.2	20.833	8.51	2	—	—	—	—	—	—
57.6	62.500	8.51	0	—	—	—	—	—	—
115.2	62.500	-45.75	0	—	—	—	—	—	—

BAUD RATE (K)	SYNC = 0, BRGH = 1, BRG16 = 0											
	Fosc = 40.000 MHz			Fosc = 20.000 MHz			Fosc = 10.000 MHz			Fosc = 8.000 MHz		
	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)
0.3	—	—	—	—	—	—	—	—	—	—	—	—
1.2	—	—	—	—	—	—	—	—	—	—	—	—
2.4	—	—	—	—	—	—	2.441	1.73	255	2.403	-0.16	207
9.6	9.766	1.73	255	9.615	0.16	129	9.615	0.16	64	9.615	-0.16	51
19.2	19.231	0.16	129	19.231	0.16	64	19.531	1.73	31	19.230	-0.16	25
57.6	58.140	0.94	42	56.818	-1.36	21	56.818	-1.36	10	55.555	3.55	8
115.2	113.636	-1.36	21	113.636	-1.36	10	125.000	8.51	4	—	—	—

BAUD RATE (K)	SYNC = 0, BRGH = 1, BRG16 = 0								
	Fosc = 4.000 MHz			Fosc = 2.000 MHz			Fosc = 1.000 MHz		
	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)
0.3	—	—	—	—	—	—	0.300	-0.16	207
1.2	1.202	0.16	207	1.201	-0.16	103	1.201	-0.16	51
2.4	2.404	0.16	103	2.403	-0.16	51	2.403	-0.16	25
9.6	9.615	0.16	25	9.615	-0.16	12	—	—	—
19.2	19.231	0.16	12	—	—	—	—	—	—
57.6	62.500	8.51	3	—	—	—	—	—	—
115.2	125.000	8.51	1	—	—	—	—	—	—

PIC18F2221/2321/4221/4321 FAMILY

TABLE 19-3: BAUD RATES FOR ASYNCHRONOUS MODES (CONTINUED)

BAUD RATE (K)	SYNC = 0, BRGH = 0, BRG16 = 1											
	Fosc = 40.000 MHz			Fosc = 20.000 MHz			Fosc = 10.000 MHz			Fosc = 8.000 MHz		
	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)
0.3	0.300	0.00	8332	0.300	0.02	4165	0.300	0.02	2082	0.300	-0.04	1665
1.2	1.200	0.02	2082	1.200	-0.03	1041	1.200	-0.03	520	1.201	-0.16	415
2.4	2.402	0.06	1040	2.399	-0.03	520	2.404	0.16	259	2.403	-0.16	207
9.6	9.615	0.16	259	9.615	0.16	129	9.615	0.16	64	9.615	-0.16	51
19.2	19.231	0.16	129	19.231	0.16	64	19.531	1.73	31	19.230	-0.16	25
57.6	58.140	0.94	42	56.818	-1.36	21	56.818	-1.36	10	55.555	3.55	8
115.2	113.636	-1.36	21	113.636	-1.36	10	125.000	8.51	4	—	—	—

BAUD RATE (K)	SYNC = 0, BRGH = 0, BRG16 = 1								
	Fosc = 4.000 MHz			Fosc = 2.000 MHz			Fosc = 1.000 MHz		
	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)
0.3	0.300	0.04	832	0.300	-0.16	415	0.300	-0.16	207
1.2	1.202	0.16	207	1.201	-0.16	103	1.201	-0.16	51
2.4	2.404	0.16	103	2.403	-0.16	51	2.403	-0.16	25
9.6	9.615	0.16	25	9.615	-0.16	12	—	—	—
19.2	19.231	0.16	12	—	—	—	—	—	—
57.6	62.500	8.51	3	—	—	—	—	—	—
115.2	125.000	8.51	1	—	—	—	—	—	—

BAUD RATE (K)	SYNC = 0, BRGH = 1, BRG16 = 1 or SYNC = 1, BRG16 = 1											
	Fosc = 40.000 MHz			Fosc = 20.000 MHz			Fosc = 10.000 MHz			Fosc = 8.000 MHz		
	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)
0.3	0.300	0.00	33332	0.300	0.00	16665	0.300	0.00	8332	0.300	-0.01	6665
1.2	1.200	0.00	8332	1.200	0.02	4165	1.200	0.02	2082	1.200	-0.04	1665
2.4	2.400	0.02	4165	2.400	0.02	2082	2.402	0.06	1040	2.400	-0.04	832
9.6	9.606	0.06	1040	9.596	-0.03	520	9.615	0.16	259	9.615	-0.16	207
19.2	19.193	-0.03	520	19.231	0.16	259	19.231	0.16	129	19.230	-0.16	103
57.6	57.803	0.35	172	57.471	-0.22	86	58.140	0.94	42	57.142	0.79	34
115.2	114.943	-0.22	86	116.279	0.94	42	113.636	-1.36	21	117.647	-2.12	16

BAUD RATE (K)	SYNC = 0, BRGH = 1, BRG16 = 1 or SYNC = 1, BRG16 = 1								
	Fosc = 4.000 MHz			Fosc = 2.000 MHz			Fosc = 1.000 MHz		
	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)
0.3	0.300	0.01	3332	0.300	-0.04	1665	0.300	-0.04	832
1.2	1.200	0.04	832	1.201	-0.16	415	1.201	-0.16	207
2.4	2.404	0.16	415	2.403	-0.16	207	2.403	-0.16	103
9.6	9.615	0.16	103	9.615	-0.16	51	9.615	-0.16	25
19.2	19.231	0.16	51	19.230	-0.16	25	19.230	-0.16	12
57.6	58.824	2.12	16	55.555	3.55	8	—	—	—
115.2	111.111	-3.55	8	—	—	—	—	—	—

19.2 EUSART Asynchronous Mode

The Asynchronous mode of operation is selected by clearing the SYNC bit (TXSTA<4>). In this mode, the EUSART uses standard Non-Return-to-Zero (NRZ) format (one Start bit, eight or nine data bits and one Stop bit). The most common data format is 8 bits. An on-chip dedicated 8-bit/16-bit Baud Rate Generator can be used to derive standard baud rate frequencies from the oscillator.

The EUSART transmits and receives the LSb first. The EUSART's transmitter and receiver are functionally independent but use the same data format and baud rate. The Baud Rate Generator produces a clock, either x16 or x64 of the bit shift rate depending on the BRGH and BRG16 bits (TXSTA<2> and BAUDCON<3>). Parity is not supported by the hardware but can be implemented in software and stored as the 9th data bit.

The TXCKP (BAUDCON<4>) and RXDTP (BAUDCON<5>) bits allow the TX and RX signals to be inverted (polarity reversed). Devices that buffer signals between TTL and RS-232 levels also invert the signal. Setting the TXCKP and RXDTP bits allows for the use of circuits that provide buffering without inverting the signal.

In Asynchronous mode, clock polarity is selected with the TXCKP bit (BAUDCON<4>). Setting TXCKP sets the Idle state on CK as high, while clearing the bit sets the Idle state as low. Data polarity is selected with the RXDTP bit (BAUDCON<5>). Setting RXDTP inverts data on RX, while clearing the bit has no effect on received data.

When operating in Asynchronous mode, the EUSART module consists of the following important elements:

- Baud Rate Generator
- Sampling Circuit
- Asynchronous Transmitter
- Asynchronous Receiver
- Auto-Wake-up on Break signal
- 12-bit Break Character Transmit
- Auto-Baud Rate Detection
- Pin State Polarity

19.2.1 EUSART ASYNCHRONOUS TRANSMITTER

The EUSART transmitter block diagram is shown in Figure 19-3. The heart of the transmitter is the Transmit (Serial) Shift Register (TSR). The Shift register obtains its data from the Read/Write Transmit Buffer register, TXREG. The TXREG register is loaded with data in software. The TSR register is not loaded until the Stop bit has been transmitted from the previous load. As soon as the Stop bit is transmitted, the TSR is loaded with new data from the TXREG register (if available).

Once the TXREG register transfers the data to the TSR register (occurs in one Tcy), the TXREG register is empty and the TXIF flag bit (PIR1<4>) is set. This interrupt can be enabled or disabled by setting or clearing the interrupt enable bit, TXIE (PIE1<4>). TXIF will be set regardless of the state of TXIE; it cannot be cleared in software. TXIF is also not cleared immediately upon loading TXREG, but becomes valid in the second instruction cycle following the load instruction. Polling TXIF immediately following a load of TXREG will return invalid results.

While TXIF indicates the status of the TXREG register, another bit, TRMT (TXSTA<1>), shows the status of the TSR register. TRMT is a read-only bit which is set when the TSR register is empty. No interrupt logic is tied to this bit so the user has to poll this bit in order to determine if the TSR register is empty.

The TXCKP bit (BAUDCON<4>) allows the TX signal to be inverted (polarity reversed). Devices that buffer signals from TTL to RS-232 levels also invert the signal (when TTL = 1, RS-232 = negative). Inverting the polarity of the TX pin data by setting the TXCKP bit allows for use of circuits that provide buffering without inverting the signal.

Note 1: The TSR register is not mapped in data memory so it is not available to the user.

2: Flag bit TXIF is set when enable bit TXEN is set.

To set up an Asynchronous Transmission:

1. Initialize the SPBRGH:SPBRG registers for the appropriate baud rate. Set or clear the BRGH and BRG16 bits, as required, to achieve the desired baud rate.
2. Enable the asynchronous serial port by clearing bit, SYNC, and setting bit, SPEN.
3. If the signal from the TX pin is to be inverted, set the TXCKP bit.
4. If interrupts are desired, set enable bit, TXIE.
5. If 9-bit transmission is desired, set transmit bit, TX9; can be used as address/data bit.
6. Enable the transmission by setting bit, TXEN, which will also set bit, TXIF.
7. If 9-bit transmission is selected, the ninth bit should be loaded in bit, TX9D.
8. Load data to the TXREG register (starts transmission).
9. If using interrupts, ensure that the GIE and PEIE bits in the INTCON register (INTCON<7:6>) are set.

PIC18F2221/2321/4221/4321 FAMILY

FIGURE 19-12: SYNCHRONOUS TRANSMISSION (THROUGH TXEN)

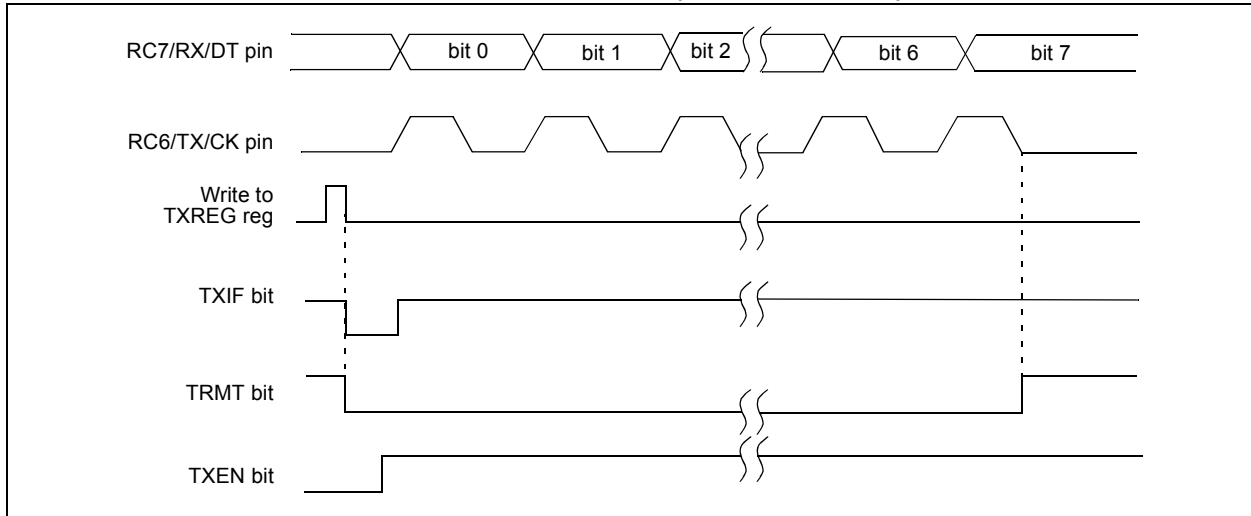


TABLE 19-7: REGISTERS ASSOCIATED WITH SYNCHRONOUS MASTER TRANSMISSION

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on page
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF	55
PIR1	PSPIF ⁽¹⁾	ADIF	RCIF	TXIF	SSPIF	CCP1IF	TMR2IF	TMR1IF	58
PIE1	PSPIE ⁽¹⁾	ADIE	RCIE	TXIE	SSPIE	CCP1IE	TMR2IE	TMR1IE	58
IPR1	PSPIP ⁽¹⁾	ADIP	RCIP	TXIP	SSPIP	CCP1IP	TMR2IP	TMR1IP	58
RCSTA	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D	57
TXREG	EUSART Transmit Register								57
TXSTA	CSRC	TX9	TXEN	SYNC	SEnDB	BRGH	TRMT	TX9D	57
BAUDCON	ABDOVF	RCIDL	RXDTP	TXCKP	BRG16	—	WUE	ABDEN	57
SPBRGH	EUSART Baud Rate Generator Register High Byte								57
SPBRG	EUSART Baud Rate Generator Register Low Byte								57

Legend: — = unimplemented, read as '0'. Shaded cells are not used for synchronous master transmission.

Note 1: These bits are unimplemented on 28-pin devices and read as '0'.

PIC18F2221/2321/4221/4321 FAMILY

BNC

Branch if Not Carry

Syntax:	BNC n				
Operands:	$-128 \leq n \leq 127$				
Operation:	If Carry bit is '0', $(PC) + 2 + 2n \rightarrow PC$				
Status Affected:	None				
Encoding:	<table border="1"><tr><td>1110</td><td>0011</td><td>nnnn</td><td>nnnn</td></tr></table>	1110	0011	nnnn	nnnn
1110	0011	nnnn	nnnn		
Description:	<p>If the Carry bit is '0', then the program will branch.</p> <p>The 2's complement number '2n' is added to the PC. Since the PC will have incremented to fetch the next instruction, the new address will be $PC + 2 + 2n$. This instruction is then a two-cycle instruction.</p>				
Words:	1				
Cycles:	1(2)				
Q Cycle Activity:					
If Jump:					

Q1	Q2	Q3	Q4
Decode	Read literal 'n'	Process Data	Write to PC
No operation	No operation	No operation	No operation

If No Jump:

Q1	Q2	Q3	Q4
Decode	Read literal 'n'	Process Data	No operation

Example: HERE BNC Jump

Before Instruction
PC = address (HERE)
After Instruction
If Carry = 0;
PC = address (Jump)
If Carry = 1;
PC = address (HERE + 2)

BNN

Branch if Not Negative

Syntax:	BNN n				
Operands:	$-128 \leq n \leq 127$				
Operation:	If Negative bit is '0', $(PC) + 2 + 2n \rightarrow PC$				
Status Affected:	None				
Encoding:	<table border="1"><tr><td>1110</td><td>0111</td><td>nnnn</td><td>nnnn</td></tr></table>	1110	0111	nnnn	nnnn
1110	0111	nnnn	nnnn		
Description:	<p>If the Negative bit is '0', then the program will branch.</p> <p>The 2's complement number '2n' is added to the PC. Since the PC will have incremented to fetch the next instruction, the new address will be $PC + 2 + 2n$. This instruction is then a two-cycle instruction.</p>				
Words:	1				
Cycles:	1(2)				
Q Cycle Activity:					
If Jump:					

Q1	Q2	Q3	Q4
Decode	Read literal 'n'	Process Data	Write to PC
No operation	No operation	No operation	No operation

If No Jump:

Q1	Q2	Q3	Q4
Decode	Read literal 'n'	Process Data	No operation

Example: HERE BNN Jump

Before Instruction
PC = address (HERE)
After Instruction
If Negative = 0;
PC = address (Jump)
If Negative = 1;
PC = address (HERE + 2)

PIC18F2221/2321/4221/4321 FAMILY

BRA Unconditional Branch

Syntax: BRA n

Operands: $-1024 \leq n \leq 1023$

Operation: $(PC) + 2 + 2n \rightarrow PC$

Status Affected: None

Encoding:

1101	0nnn	nnnn	nnnn
------	------	------	------

Description: Add the 2's complement number '2n' to the PC. Since the PC will have incremented to fetch the next instruction, the new address will be $PC + 2 + 2n$. This instruction is a two-cycle instruction.

Words: 1

Cycles: 2

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	Read literal 'n'	Process Data	Write to PC
No operation	No operation	No operation	No operation

Example: HERE BRA Jump

Before Instruction
PC = address (HERE)

After Instruction
PC = address (Jump)

BSF Bit Set f

Syntax: BSF f, b {,a}

Operands: $0 \leq f \leq 255$
 $0 \leq b \leq 7$
 $a \in [0, 1]$

Operation: $1 \rightarrow f < b >$

Status Affected: None

Encoding:

1000	bbba	ffff	ffff
------	------	------	------

Description: Bit 'b' in register 'f' is set.
If 'a' is '0', the Access Bank is selected.
If 'a' is '1', the BSR is used to select the GPR bank (default).
If 'a' is '0' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever $f \leq 95$ (5Fh). See **Section 25.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode"** for details.

Words: 1

Cycles: 1

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	Read register 'f'	Process Data	Write register 'f'

Example: BSF FLAG_REG, 7, 1

Before Instruction
FLAG_REG = 0Ah

After Instruction
FLAG_REG = 8Ah

PIC18F2221/2321/4221/4321 FAMILY

27.3 DC Characteristics: PIC18F2221/2321/4221/4321 (Industrial) PIC18LF2221/2321/4221/4321 (Industrial) (Continued)

DC CHARACTERISTICS			Standard Operating Conditions (unless otherwise stated) Operating temperature $-40^{\circ}\text{C} \leq T_A \leq +85^{\circ}\text{C}$ for industrial $-40^{\circ}\text{C} \leq T_A \leq +125^{\circ}\text{C}$ for extended			
Param No.	Symbol	Characteristic	Min	Max	Units	Conditions
D070	IPU IPURB	Weak Pull-up Current PORTB Weak Pull-up Current	50	400	μA	$V_{DD} = 5\text{V}$, $V_{PIN} = V_{SS}$
D080	VOL	Output Low Voltage I/O Ports	—	0.6	V	$I_{OL} = 8.5\text{ mA}$, $V_{DD} = 4.5\text{V}$, -40°C to $+85^{\circ}\text{C}$
D083		OSC2/CLKO (RC, RCIO, EC, ECIO modes)	—	0.6	V	$I_{OL} = 1.6\text{ mA}$, $V_{DD} = 4.5\text{V}$, -40°C to $+85^{\circ}\text{C}$
D090	VOH	Output High Voltage ⁽³⁾ I/O Ports	$V_{DD} - 0.7$	—	V	$I_{OH} = -3.0\text{ mA}$, $V_{DD} = 4.5\text{V}$, -40°C to $+85^{\circ}\text{C}$
D092		OSC2/CLKO (RC, RCIO, EC, ECIO modes)	$V_{DD} - 0.7$	—	V	$I_{OH} = -1.3\text{ mA}$, $V_{DD} = 4.5\text{V}$, -40°C to $+85^{\circ}\text{C}$
D100	COSC2	Capacitive Loading Specs on Output Pins OSC2 Pin	—	15	pF	In XT, HS and LP modes when external clock is used to drive OSC1
D101	Cio	All I/O Pins and OSC2 (in RC mode)	—	50	pF	Maximum that allows the AC Timing Specifications to be met
D102	Cb	SCL, SDA	—	400	pF	Maximum bus capacitance permitted by I ² C™ Specification

- Note 1:** In RC oscillator configuration, the OSC1/CLKI pin is a Schmitt Trigger input. It is not recommended that the PIC® device be driven with an external clock while in RC mode.
- 2:** The leakage current on the $\overline{\text{MCLR}}$ pin is strongly dependent on the applied voltage level. The specified levels represent normal operating conditions. Higher leakage current may be measured at different input voltages.
- 3:** Negative current is defined as current sourced by the pin.

PIC18F2221/2321/4221/4321 FAMILY

APPENDIX E: MIGRATION FROM MID-RANGE TO ENHANCED DEVICES

A detailed discussion of the differences between the mid-range MCU devices (i.e., PIC16CXXX) and the Enhanced devices (i.e., PIC18FXXX) is provided in AN716, “*Migrating Designs from PIC16C74A/74B to PIC18C442*”. The changes discussed, while device specific, are generally applicable to all mid-range to Enhanced device migrations.

This Application Note is available as Literature Number DS00716.

APPENDIX F: MIGRATION FROM HIGH-END TO ENHANCED DEVICES

A detailed discussion of the migration pathway and differences between the high-end MCU devices (i.e., PIC17CXXX) and the Enhanced devices (i.e., PIC18FXXX) is provided in AN726, “*PIC17CXXX to PIC18CXXX Migration*”.

This Application Note is available as Literature Number DS00726.

PIC18F2221/2321/4221/4321 FAMILY

DAW	298
DC Characteristics	347
Power-Down and Supply Current	337
Supply Voltage	336
DCFSNZ	299
DECF	298
DECFSZ	299
Development Support	329
Device Differences	386
Device Overview	9
Details on Individual Family Members	10
Features (table)	11
New Core Features	9
Other Special Features	10
Device Reset Timers	51
Oscillator Start-up Timer (OST)	51
PLL Lock Time-out	51
Power-up Timer (PWRT)	51
Time-out Sequence	51
Direct Addressing	74
E	
Effect on Standard PIC MCU Instructions	326
Effects of Power-Managed Modes on Various	
Clock Sources	38
Electrical Characteristics	333
Enhanced Capture/Compare/PWM (ECCP)	153
Associated Registers	166
Capture and Compare Modes	154
Capture Mode. <i>See</i> Capture (ECCP Module).	
Outputs and Configuration	154
Pin Configurations for ECCP1	154
PWM Mode. <i>See</i> PWM (ECCP Module).	
Standard PWM Mode	154
Timer Resources	154
Enhanced PWM Mode. <i>See</i> PWM (ECCP Module).	155
Enhanced Universal Synchronous Asynchronous Receiver	
Transmitter (EUSART). <i>See</i> EUSART.	
Equations	
A/D Acquisition Time	238
A/D Minimum Charging Time	238
Calculating the Minimum Required	
Acquisition Time	238
Errata	8
EUSART	
Asynchronous Mode	221
12-Bit Break Transmit and Receive	227
Associated Registers, Receive	225
Associated Registers, Transmit	223
Auto-Wake-up on Sync Break	226
Receiver	224
Setting up 9-Bit Mode with Address Detect	224
Transmitter	221
Baud Rate Generator	
Operation in Power-Managed Mode	215
Baud Rate Generator (BRG)	215
Associated Registers	216
Auto-Baud Rate Detect	219
Baud Rate Error, Calculating	216
Baud Rates, Asynchronous Modes	217
High Baud Rate Select (BRGH Bit)	215
Sampling	215

Synchronous Master Mode	228
Associated Registers, Receive	230
Associated Registers, Transmit	229
Reception	230
Transmission	228
Synchronous Slave Mode	231
Associated Registers, Receive	232
Associated Registers, Transmit	231
Reception	232
Transmission	231
Extended Instruction Set	
ADDFSR	322
ADDLNLK	322
and Using MPLAB Tools	328
CALLW	323
Considerations for Use	326
MOVSF	323
MOVSS	324
PUSHL	324
SUBFSR	325
SUBLNLK	325
Syntax	321
External Clock Input	30
F	
Fail-Safe Clock Monitor	259, 272
Exiting Operation	272
Interrupts in Power-Managed Modes	273
POR or Wake From Sleep	273
WDT During Oscillator Failure	272
Fast Register Stack	62
Firmware Instructions	279
Flash Program Memory	79
Associated Registers	87
Control Registers	80
EECON1 and EECON2	80
TABLAT (Table Latch) Register	82
TBLPTR (Table Pointer) Register	82
Erase Sequence	84
Erasing	84
Operation During Code-Protect	87
Reading	83
Table Pointer	
Boundaries	82
Boundaries Based on Operation	82
Operations with TBLRD and TBLWT (table)	82
Table Reads and Table Writes	79
Write Sequence	85
Writing	85
Protection Against Spurious Writes	87
Unexpected Termination	87
Write Verify	87
FSCM. <i>See</i> Fail-Safe Clock Monitor.	
G	
GOTO	300
H	
Hardware Multiplier	95
Introduction	95
Operation	95
Performance Comparison	95

PIC18F2221/2321/4221/4321 FAMILY

Software Simulator (MPLAB SIM)	330	TMR1L Register	133
Special Event Trigger. <i>See</i> Compare (CCP Mode).		Use as a Real-Time Clock	136
Special Event Trigger. <i>See</i> Compare (ECCP Module).		Timer2	139
Special Features of the CPU	259	Associated Registers	140
Special Function Registers	68	Interrupt	140
Map	68	Operation	139
SPI Mode (MSSP)		Output	140
Associated Registers	175	PR2 Register	150, 155
Bus Mode Compatibility	175	TMR2 to PR2 Match Interrupt	155
Effects of a Reset	175	TMR2-to-PR2 Match Interrupt	150
Enabling SPI I/O	171	Timer3	141
Master Mode	172	16-Bit Read/Write Mode	143
Master/Slave Connection	171	Associated Registers	143
Operation	170	Operation	142
Operation in Power-Managed Modes	175	Oscillator	141, 143
Serial Clock	167	Overflow Interrupt	141, 143
Serial Data In	167	Special Event Trigger (CCP)	143
Serial Data Out	167	TMR3H Register	141
Slave Mode	173	TMR3L Register	141
Slave Select	167	Timing Diagrams	
Slave Select Synchronization	173	A/D Conversion	371
SPI Clock	172	Acknowledge Sequence	204
Typical Connection	171	Asynchronous Reception	225
SS	167	Asynchronous Transmission	222
SSPOV	201	Asynchronous Transmission (Back to Back)	222
SSPOV Status Flag	201	Automatic Baud Rate Calculation	220
SSPSTAT Register		Auto-Wake-up Bit (WUE) During	
R/W Bit	181, 183	Normal Operation	226
Stack Full/Underflow Resets	62	Auto-Wake-up Bit (WUE) During Sleep	226
SUBFSR	325	Baud Rate Generator with Clock Arbitration	198
SUBFWB	314	BRG Overflow Sequence	220
SUBLW	315	BRG Reset Due to SDA Arbitration During	
SUBULNK	325	Start Condition	207
SUBWF	315	Brown-out Reset (BOR)	357
SUBWFB	316	Bus Collision During a Repeated Start	
SWAPF	316	Condition (Case 1)	208
T		Bus Collision During a Repeated Start	
Table Reads/Table Writes	62	Condition (Case 2)	208
TBLRD	317	Bus Collision During a Start Condition	
TBLWT	318	(SCL = 0)	207
Time-out in Various Situations (table)	51	Bus Collision During a Stop Condition (Case 1)	209
Timer0	129	Bus Collision During a Stop Condition (Case 2)	209
Associated Registers	131	Bus Collision During Start Condition	
Operation	130	(SDA Only)	206
Overflow Interrupt	131	Bus Collision for Transmit and Acknowledge	205
Prescaler	131	Capture/Compare/PWM (All CCP Modules)	359
Prescaler Assignment (PSA Bit)	131	CLKO and I/O	356
Prescaler Select (T0PS2:T0PS0 Bits)	131	Clock Synchronization	191
Prescaler. <i>See</i> Prescaler, Timer0.		Clock/Instruction Cycle	63
Reads and Writes in 16-Bit Mode	130	EUSART Synchronous Receive (Master/Slave)	369
Source Edge Select (T0SE Bit)	130	EUSART Synchronous Transmission	
Source Select (T0CS Bit)	130	(Master/Slave)	369
Switching Prescaler Assignment	131	Example SPI Master Mode (CKE = 0)	361
Timer1	133	Example SPI Master Mode (CKE = 1)	362
16-Bit Read/Write Mode	135	Example SPI Slave Mode (CKE = 0)	363
Associated Registers	137	Example SPI Slave Mode (CKE = 1)	364
Interrupt	136	External Clock (All Modes Except PLL)	354
Operation	134	Fail-Safe Clock Monitor	273
Oscillator	133, 135	First Start Bit Timing	199
Layout Considerations	136	Full-Bridge PWM Output	159
Low-Power Option	135	Half-Bridge PWM Output	158
Overflow Interrupt	133	High/Low-Voltage Detect Characteristics	351
Resetting, Using the CCP Special Event Trigger	136	High-Voltage Detect Operation (VDIRMAG = 1)	256
Special Event Trigger (ECCP)	154	I ² C Bus Data	365
TMR1H Register	133	I ² C Bus Start/Stop Bits	365