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Embedded - System On Chip (SoC): The Heart of Modern Embedded Systems

Embedded - System On Chip (SoC) refers to an integrated circuit that consolidates all the essential components of a computer system into a single chip. This includes a microprocessor, memory, and other peripherals, all packed into one compact and efficient package. SoCs are designed to provide a complete computing solution, optimizing both space and power consumption, making them ideal for a wide range of embedded applications.

What are Embedded - System On Chip (SoC)?

System On Chip (SoC) integrates multiple functions of a computer or electronic system onto a single chip. Unlike traditional multi-chip solutions, SoCs combine a central

Details	
Product Status	Obsolete
Architecture	MCU, FPGA
Core Processor	ARM® Cortex®-M3
Flash Size	128KB
RAM Size	16KB
Peripherals	DMA, POR, WDT
Connectivity	EBI/EMI, I²C, SPI, UART/USART
Speed	100MHz
Primary Attributes	ProASIC®3 FPGA, 60K Gates, 1536D-Flip-Flops
Operating Temperature	0°C ~ 85°C (TJ)
Package / Case	288-TFBGA, CSPBGA
Supplier Device Package	288-CSP (11x11)
Purchase URL	https://www.e-xfl.com/product-detail/microsemi/a2f060m3e-1csg288

1 – SmartFusion Family Overview

Introduction

The SmartFusion® family of cSoCs builds on the technology first introduced with the Fusion mixed signal FPGAs. SmartFusion cSoCs are made possible by integrating FPGA technology with programmable high-performance analog and hardened ARM Cortex-M3 microcontroller blocks on a flash semiconductor process. The SmartFusion cSoC takes its name from the fact that these three discrete technologies are integrated on a single chip, enabling the lowest cost of ownership and smallest footprint solution to you.

General Description

Microcontroller Subsystem (MSS)

The MSS is composed of a 100 MHz Cortex-M3 processor and integrated peripherals, which are interconnected via a multi-layer AHB bus matrix (ABM). This matrix allows the Cortex-M3 processor, FPGA fabric master, Ethernet media access controller (MAC), when available, and peripheral DMA (PDMA) controller to act as masters to the integrated peripherals, FPGA fabric, embedded nonvolatile memory (eNVM), embedded synchronous RAM (eSRAM), external memory controller (EMC), and analog compute engine (ACE) blocks.

SmartFusion cSoCs of different densities offer various sets of integrated peripherals. Available peripherals include SPI, I^2C, and UART serial ports, embedded FlashROM (EFROM), 10/100 Ethernet MAC, timers, phase-locked loops (PLLs), oscillators, real-time counters (RTC), and peripheral DMA controller (PDMA).

Programmable Analog

Analog Front-End (AFE)

SmartFusion cSoCs offer an enhanced analog front-end compared to Fusion devices. The successive approximation register analog-to-digital converters (SAR ADC) are similar to those found on Fusion devices. SmartFusion cSoC also adds first order sigma-delta digital-to-analog converters (SDD DAC).

SmartFusion cSoCs can handle multiple analog signals simultaneously with its signal conditioning blocks (SCBs). SCBs are made of a combination of active bipolar prescalers (ABPS), comparators, current monitors and temperature monitors. ABPS modules allow larger bipolar voltages to be fed to the ADC. Current monitors take the voltage across an external sense resistor and convert it to a voltage suitable for the ADC input range. Similarly, the temperature monitor reads the current through an external PN-junction (diode or transistor) and converts it internally for the ADC. The SCB also includes comparators to monitor fast signal thresholds without using the ADC. The output of the comparators can be fed to the analog compute engine or the ADC.

Analog Compute Engine (ACE)

The mixed signal blocks found in SmartFusion cSoCs are controlled and connected to the rest of the system via a dedicated processor called the analog compute engine (ACE). The role of the ACE is to offload control of the analog blocks from the Cortex-M3, thus offering faster throughput or better power consumption compared to a system where the main processor is in charge of monitoring the analog resources. The ACE is built to handle sampling, sequencing, and post-processing of the ADCs, DACs, and SCBs.

This enables reduction or complete removal of expensive voltage monitor and brownout detection devices from the PCB design. Flash-based SmartFusion cSoCs simplify total system design and reduce cost and design risk, while increasing system reliability.

Immunity to Firm Errors

Firm errors occur most commonly when high-energy neutrons, generated in the atmosphere, strike a configuration cell of an SRAM FPGA. The energy of the collision can change the state of the configuration cell and thus change the logic, routing, or I/O configuration behavior in an unpredictable way.

Another source of radiation-induced firm errors is alpha particles. For alpha radiation to cause a soft or firm error, its source must be in very close proximity to the affected circuit. The alpha source must be in the package molding compound or in the die itself. While low-alpha molding compounds are being used increasingly, this helps reduce but does not entirely eliminate alpha-induced firm errors.

Firm errors are impossible to prevent in SRAM FPGAs. The consequence of this type of error can be a complete system failure. Firm errors do not occur in SmartFusion cSoCs. Once it is programmed, the flash cell configuration element of SmartFusion cSoCs cannot be altered by high energy neutrons and is therefore immune to errors from them. Recoverable (or soft) errors occur in the user data SRAMs of all FPGA devices. These can easily be mitigated by using error detection and correction (EDAC) circuitry built into the FPGA fabric.

Specifying I/O States During Programming

You can modify the I/O states during programming in FlashPro. In FlashPro, this feature is supported for PDB files generated from Designer v8.5 or greater. See the *FlashPro User's Guide* for more information.

Note: PDB files generated from Designer v8.1 to Designer v8.4 (including all service packs) have limited display of Pin Numbers only.

The I/Os are controlled by the JTAG Boundary Scan register during programming, except for the analog pins (AC, AT and AV). The Boundary Scan register of the AG pin can be used to enable/disable the gate driver in software v9.0.

1. Load a PDB from the FlashPro GUI. You must have a PDB loaded to modify the I/O states during programming.
2. From the FlashPro GUI, click PDB Configuration. A FlashPoint – Programming File Generator window appears.
3. Click the Specify I/O States During Programming button to display the Specify I/O States During Programming dialog box.
4. Sort the pins as desired by clicking any of the column headers to sort the entries by that header. Select the I/Os you wish to modify (Figure 1-1 on page 1-4).
5. Set the I/O Output State. You can set Basic I/O settings if you want to use the default I/O settings for your pins, or use Custom I/O settings to customize the settings for each pin. Basic I/O state settings:

   1 – I/O is set to drive out logic High

   0 – I/O is set to drive out logic Low

   Last Known State – I/O is set to the last value that was driven out prior to entering the programming mode, and then held at that value during programming

   Z -Tri-State: I/O is tristated

Theta-JA

Junction-to-ambient thermal resistance (θ_{JA}) is determined under standard conditions specified by JEDEC (JESD-51), but it has little relevance in actual performance of the product. It should be used with caution but is useful for comparing the thermal performance of one package to another.

A sample calculation showing the maximum power dissipation allowed for the A2F200-FG484 package under forced convection of 1.0 m/s and 75°C ambient temperature is as follows:

$$\text{Maximum Power Allowed} = \frac{T_{J(MAX)} - T_{A(MAX)}}{\theta_{JA}}$$

EQ 4

where

θ_{JA} = 19.00°C/W (taken from Table 2-6 on page 2-7).

T_A = 75.00°C

$$\text{Maximum Power Allowed} = \frac{100.00°C - 75.00°C}{19.00°C/W} = 1.3\ W$$

EQ 5

The power consumption of a device can be calculated using the Microsemi SoC Products Group power calculator. The device's power consumption must be lower than the calculated maximum power dissipation by the package. If the power consumption is higher than the device's maximum allowable power dissipation, a heat sink can be attached on top of the case, or the airflow inside the system must be increased.

Theta-JB

Junction-to-board thermal resistance (θ_{JB}) measures the ability of the package to dissipate heat from the surface of the chip to the PCB. As defined by the JEDEC (JESD-51) standard, the thermal resistance from junction to board uses an isothermal ring cold plate zone concept. The ring cold plate is simply a means to generate an isothermal boundary condition at the perimeter. The cold plate is mounted on a JEDEC standard board with a minimum distance of 5.0 mm away from the package edge.

Theta-JC

Junction-to-case thermal resistance (θ_{JC}) measures the ability of a device to dissipate heat from the surface of the chip to the top or bottom surface of the package. It is applicable for packages used with external heat sinks. Constant temperature is applied to the surface in consideration and acts as a boundary condition. This only applies to situations where all or nearly all of the heat is dissipated through the surface in consideration.

Calculation for Heat Sink

For example, in a design implemented in an A2F200-FG484 package with 2.5 m/s airflow, the power consumption value using the power calculator is 3.00 W. The user-dependent T_a and T_j are given as follows:

T_J = 100.00°C

T_A = 70.00°C

From the datasheet:

θ_{JA} = 17.00°C/W

θ_{JC} = 8.28°C/W

$$P = \frac{T_J - T_A}{T_{JA}} = \frac{100°C - 70°C}{17.00\ W} = 1.76\ W$$

EQ 6

The 1.76 W power is less than the required 3.00 W. The design therefore requires a heat sink, or the airflow where the device is mounted should be increased. The design's total junction-to-air thermal resistance requirement can be estimated by EQ 7:

$$T_{JA(total)} = \frac{T_J - T_A}{P} = \frac{100°C - 70°C}{3.00\ W} = 10.00°C/W$$

EQ 7

Determining the heat sink's thermal performance proceeds as follows:

$$T_{JA(TOTAL)} = T_{JC} + T_{CS} + T_{SA}$$

EQ 8

where

T_{JA} = 0.37°C/W

= Thermal resistance of the interface material between the case and the heat sink, usually provided by the thermal interface manufacturer

T_{SA} = Thermal resistance of the heat sink in °C/W

$$T_{SA} = T_{JA(TOTAL)} - T_{JC} - T_{CS}$$

EQ 9

$$T_{SA} = 13.33°C/W - 8.28°C/W - 0.37°C/W = 5.01°C/W$$

A heat sink with a thermal resistance of 5.01°C/W or better should be used. Thermal resistance of heat sinks is a function of airflow. The heat sink performance can be significantly improved with increased airflow.

Carefully estimating thermal resistance is important in the long-term reliability of an FPGA. Design engineers should always correlate the power consumption of the device with the maximum allowable power dissipation of the package selected for that device.

Note: The junction-to-air and junction-to-board thermal resistances are based on JEDEC standard (JESD-51) and assumptions made in building the model. It may not be realized in actual application and therefore should be used with a degree of caution. Junction-to-case thermal resistance assumes that all power is dissipated through the case.

Temperature and Voltage Derating Factors

Table 2-7 • **Temperature and Voltage Derating Factors for Timing Delays (normalized to T_J = 85°C, worst-case VCC = 1.425 V)**

Array Voltage VCC (V)	Junction Temperature (°C)					
	–40°C	**0°C**	**25°C**	**70°C**	**85°C**	**100°C**
1.425	0.86	0.91	0.93	0.98	1.00	1.02
1.500	0.81	0.86	0.88	0.93	0.95	0.96
1.575	0.78	0.83	0.85	0.90	0.91	0.93

Table 2-14 • **Different Components Contributing to Dynamic Power Consumption in SmartFusion cSoCs**

Parameter	Definition	Power Supply		Device			Units
		Name	Domain	A2F060	A2F200	A2F500	
PAC24	Current Monitor Power Contribution	See Table 2-93 on page 2-78	–	1.03			mW
PAC25	ABPS Power Contribution	See Table 2-96 on page 2-82	–	0.70			mW
PAC26	Sigma-Delta DAC Power Contribution[2]	See Table 2-98 on page 2-85	–	0.58			mW
PAC27	Comparator Power Contribution	See Table 2-97 on page 2-84	–	1.02			mW
PAC28	Voltage Regulator Power Contribution[3]	See Table 2-99 on page 2-87	–	36.30			mW

Notes:

1. *For a different use of MSS peripherals and resources, refer to SmartPower.*
2. *Assumes Input = Half Scale Operation mode.*
3. *Assumes 100 mA load on 1.5 V domain.*

Table 2-15 • **Different Components Contributing to the Static Power Consumption in SmartFusion cSoCs**

Parameter	Definition	Power Supply		Device			Units
		Name	Domain	A2F060	A2F200	A2F200	
PDC1	Core static power contribution in SoC mode	VCC	1.5 V	11.10	23.70	37.95	mW
PDC2	Device static power contribution in Standby Mode	See Table 2-8 on page 2-10	–	11.10	23.70	37.95	mW
PDC3	Device static power contribution in Time Keeping mode	See Table 2-8 on page 2-10	3.3 V	33.00	33.00	33.00	µW
PDC7	Static contribution per input pin (standard dependent contribution)	VCCxxxxIOBx/VCC	See Table 2-10 and Table 2-11 on page 2-11.				
PDC8	Static contribution per output pin (standard dependent contribution)	VCCxxxxIOBx/VCC	See Table 2-12 and Table 2-13 on page 2-11.				
PDC9	Static contribution per PLL	VCC	1.5 V	2.55	2.55	2.55	mW

Table 2-16 • **eNVM Dynamic Power Consumption**

Parameter	Description	Condition	Min.	Typ.	Max.	Units
eNVM System	eNVM array operating power	Idle		795		µA
		Read operation	See Table 2-14 on page 2-12.			
		Erase		900		µA
		Write		900		µA
PNVMCTRL	eNVM controller operating power			20		µW/MHz

P_{PLL} = 0 W

Embedded Nonvolatile Memory Dynamic Contribution—P_{eNVM}

SoC Mode

The eNVM dynamic power consumption is a piecewise linear function of frequency.

$P_{eNVM} = N_{eNVM\text{-}BLOCKS} * E_4 * P_{AC15} * F_{READ\text{-}eNVM}$ when $F_{READ\text{-}eNVM}$ d33 MHz,

$P_{eNVM} = N_{eNVM\text{-}BLOCKS} * E_4 * (P_{AC16} + P_{AC17} * F_{READ\text{-}eNVM})$ when $F_{READ\text{-}eNVM}$ > 33 MHz

Where:

$N_{eNVM\text{-}BLOCKS}$ is the number of eNVM blocks used in the design.

E_4 is the eNVM enable rate for read operations. Default is 0 (eNVM mainly in idle state).

$F_{READ\text{-}eNVM}$ is the eNVM read clock frequency.

Standby Mode and Time Keeping Mode

P_{eNVM} = 0 W

Main Crystal Oscillator Dynamic Contribution—$P_{XTL\text{-}OSC}$

SoC Mode

$P_{XTL\text{-}OSC} = P_{AC18}$

Standby Mode

$P_{XTL\text{-}OSC}$ = 0 W

Time Keeping Mode

$P_{XTL\text{-}OSC}$ = 0 W

Low Power Oscillator Crystal Dynamic Contribution—$P_{LPXTAL\text{-}OSC}$

Operating, Standby, and Time Keeping Mode

$P_{LPXTAL\text{-}OSC} = P_{AC21}$

RC Oscillator Dynamic Contribution—$P_{RC\text{-}OSC}$

SoC Mode

$P_{RC\text{-}OSC} = P_{AC19A} + P_{AC19B}$

Standby Mode and Time Keeping Mode

$P_{RC\text{-}OSC}$ = 0 W

Analog System Dynamic Contribution—P_{AB}

SoC Mode

$P_{AB} = P_{AC23} * N_{TM} + P_{AC24} * N_{CM} + P_{AC25} * N_{ABPS} + P_{AC26} * N_{SDD} + P_{AC27} * N_{COMP} + P_{ADC} * N_{ADC} + P_{VR}$

Where:

N_{CM} is the number of current monitor blocks

N_{TM} is the number of temperature monitor blocks

N_{SDD} is the number of sigma-delta DAC blocks

N_{ABPS} is the number of ABPS blocks

N_{ADC} is the number of ADC blocks

N_{COMP} is the number of comparator blocks

$P_{VR} = P_{AC28}$

$P_{ADC} = P_{AC20A} + P_{AC20B}$

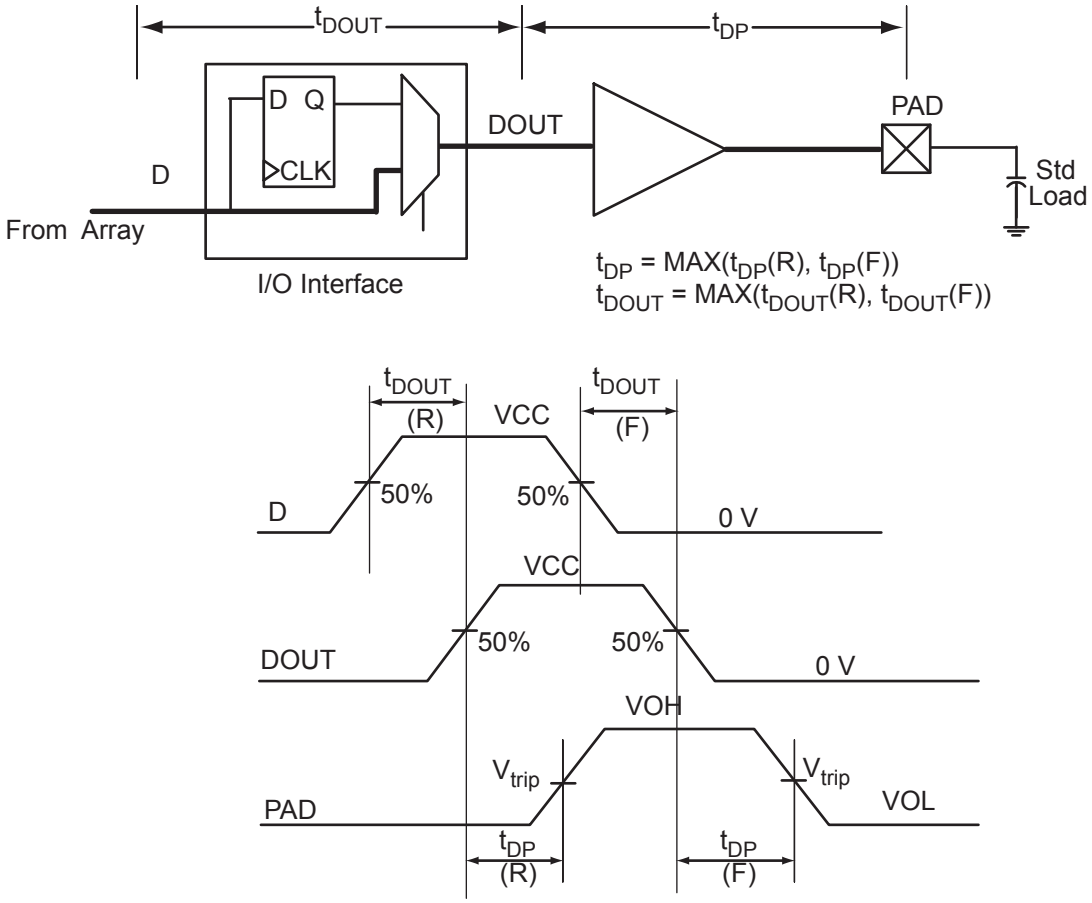

Figure 2-4 • **Output Buffer Model and Delays (example)**

1.8 V LVCMOS

Low-voltage CMOS for 1.8 V is an extension of the LVCMOS standard (JESD8-5) used for general-purpose 1.8 V applications. It uses a 1.8 V input buffer and a push-pull output buffer.

Table 2-47 • **Minimum and Maximum DC Input and Output Levels Applicable to FPGA I/O Banks**

1.8 V LVCMOS	VIL		VIH		VOL	VOH	I_{OL}	I_{OH}	I_{OSL}	I_{OSH}	I_{IL}	I_{IH}
Drive Strength	Min. V	Max. V	Min. V	Max. V	Max. V	Min. V	mA	mA	Max. mA[1]	Max. mA[1]	µA[2]	µA[2]
2 mA	–0.3	0.35 * VCCxxxxIOBx	0.65 * VCCxxxxIOBx	1.9	0.45	VCCxxxxIOBx – 0.45	2	2	11	9	15	15
4 mA	–0.3	0.35 * VCCxxxxIOBx	0.65 * VCCxxxxIOBx	1.9	0.45	VCCxxxxIOBx – 0.45	4	4	22	17	15	15
6 mA	–0.3	0.35 * VCCxxxxIOBx	0.65 * VCCxxxxIOBx	1.9	0.45	VCCxxxxIOBx – 0.45	6	6	44	35	15	15
8 mA	–0.3	0.35 * VCCxxxxIOBx	0.65 * VCCxxxxIOBx	1.9	0.45	VCCxxxxIOBx – 0.45	8	8	51	45	15	15
12 mA	–0.3	0.35 * VCCxxxxIOBx	0.65 * VCCxxxxIOBx	1.9	0.45	VCCxxxxIOBx – 0.45	12	12	74	91	15	15
16 mA	–0.3	0.35 * VCCxxxxIOBx	0.65 * VCCxxxxIOBx	1.9	0.45	VCCxxxxIOBx – 0.45	16	16	74	91	15	15

Notes:

1. *Currents are measured at high temperature (100°C junction temperature) and maximum voltage.*
2. *Currents are measured at 85°C junction temperature.*
3. *Software default selection highlighted in gray.*

Table 2-48 • **Minimum and Maximum DC Input and Output Levels Applicable to MSS I/O Banks**

1.8 V LVCMOS	VIL		VIH		VOL	VOH	I_{OL}	I_{OH}	I_{OSL}	I_{OSH}	I_{IL}	I_{IH}
Drive Strength	Min. V	Max. V	Min. V	Max. V	Max. V	Min. V	mA	mA	Max. mA[1]	Max. mA[1]	µA[2]	µA[2]
4 mA	–0.3	0.35 * VCCxxxxIOBx	0.65 * VCCxxxxIOBx	3.6	0.45	VCCxxxxIOBx – 0.45	4	4	22	17	15	15

Notes:

1. *Currents are measured at high temperature (100°C junction temperature) and maximum voltage.*
2. *Currents are measured at 85°C junction temperature.*
3. *Software default selection highlighted in gray.*

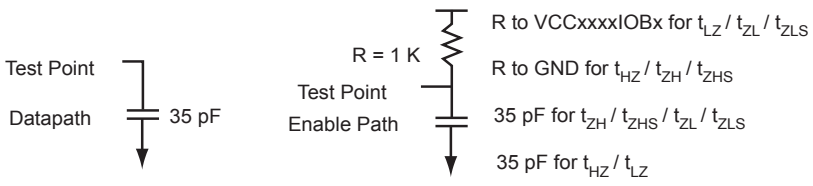

Figure 2-8 • **AC Loading**

Table 2-49 • **AC Waveforms, Measuring Points, and Capacitive Loads**

Input Low (V)	Input High (V)	Measuring Point* (V)	V_{REF} (typ.) (V)	C_{LOAD} (pF)
0	1.8	0.9	–	35

* *Measuring point = V_{trip}. See Table 2-22 on page 2-24 for a complete table of trip points.*

Input Register

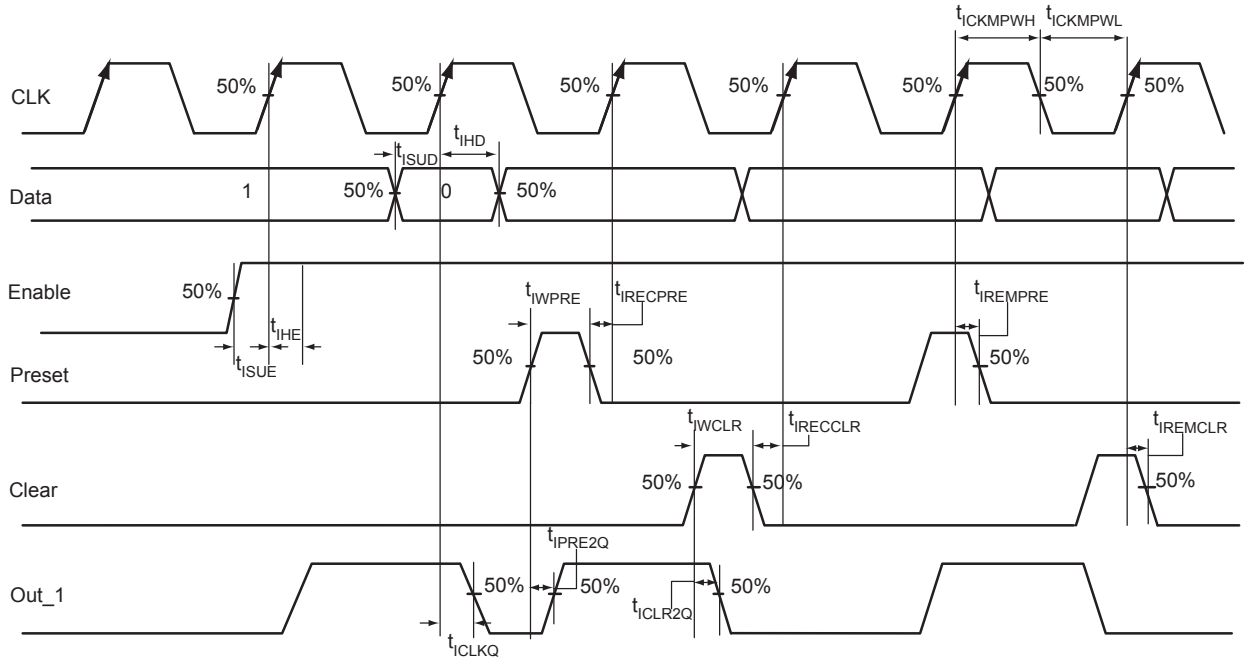

Figure 2-16 • **Input Register Timing Diagram**

Timing Characteristics

Table 2-71 • **Input Data Register Propagation Delays**
Worst Commercial-Case Conditions: T_J = 85°C, Worst-Case VCC = 1.425 V

Parameter	Description	–1	Std.	Units
t_{ICLKQ}	Clock-to-Q of the Input Data Register	0.24	0.29	ns
t_{ISUD}	Data Setup Time for the Input Data Register	0.27	0.32	ns
t_{IHD}	Data Hold Time for the Input Data Register	0.00	0.00	ns
t_{ISUE}	Enable Setup Time for the Input Data Register	0.38	0.45	ns
t_{IHE}	Enable Hold Time for the Input Data Register	0.00	0.00	ns
t_{ICLR2Q}	Asynchronous Clear-to-Q of the Input Data Register	0.46	0.55	ns
t_{IPRE2Q}	Asynchronous Preset-to-Q of the Input Data Register	0.46	0.55	ns
$t_{IREMCLR}$	Asynchronous Clear Removal Time for the Input Data Register	0.00	0.00	ns
$t_{IRECCLR}$	Asynchronous Clear Recovery Time for the Input Data Register	0.23	0.27	ns
$t_{IREMPRE}$	Asynchronous Preset Removal Time for the Input Data Register	0.00	0.00	ns
$t_{IRECPRE}$	Asynchronous Preset Recovery Time for the Input Data Register	0.23	0.27	ns
t_{IWCLR}	Asynchronous Clear Minimum Pulse Width for the Input Data Register	0.22	0.22	ns
t_{IWPRE}	Asynchronous Preset Minimum Pulse Width for the Input Data Register	0.22	0.22	ns
$t_{ICKMPWH}$	Clock Minimum Pulse Width High for the Input Data Register	0.36	0.36	ns
$t_{ICKMPWL}$	Clock Minimum Pulse Width Low for the Input Data Register	0.32	0.32	ns

Note: For the derating values at specific junction temperature and voltage supply levels, refer to Table 2-7 on page 2-9 for derating values.

DDR Module Specifications

Input DDR Module

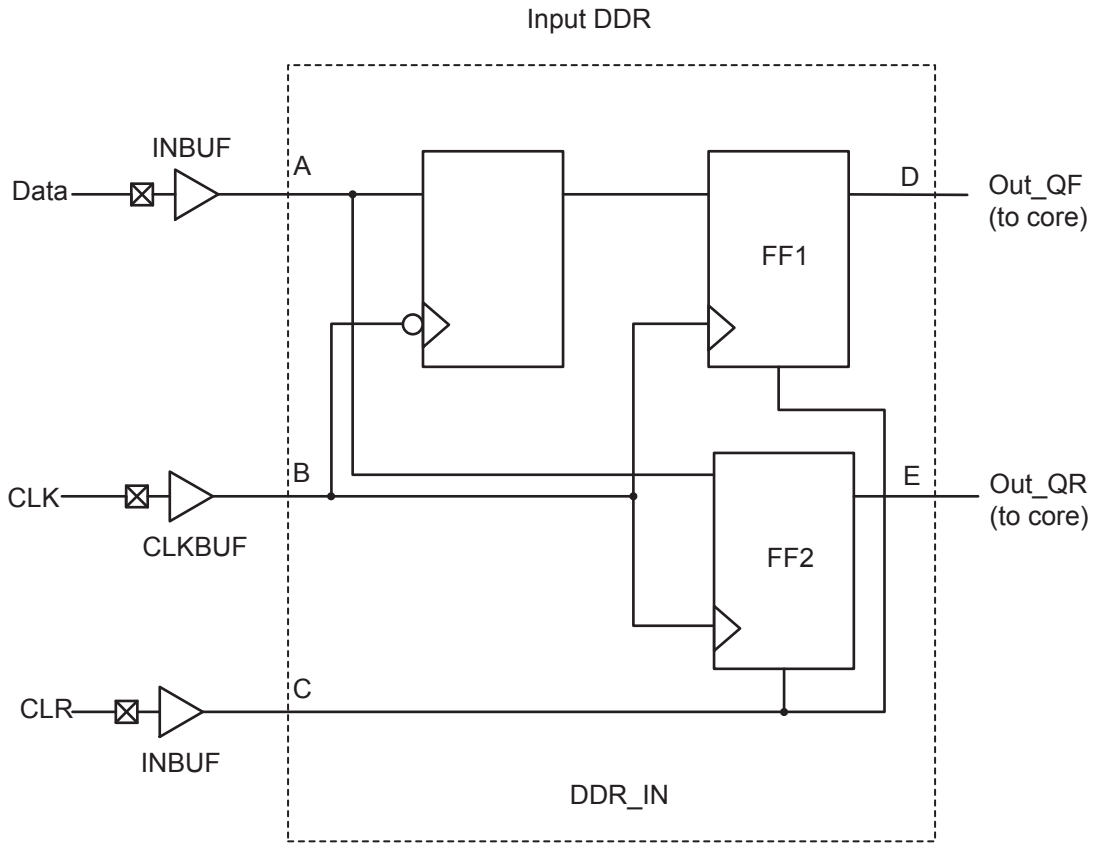

Figure 2-19 • **Input DDR Timing Model**

Table 2-74 • **Parameter Definitions**

Parameter Name	Parameter Definition	Measuring Nodes (from, to)
$t_{DDRICLKQ1}$	Clock-to-Out Out_QR	B, D
$t_{DDRICLKQ2}$	Clock-to-Out Out_QF	B, E
$t_{DDRISUD}$	Data Setup Time of DDR input	A, B
t_{DDRIHD}	Data Hold Time of DDR input	A, B
$t_{DDRICLR2Q1}$	Clear-to-Out Out_QR	C, D
$t_{DDRICLR2Q2}$	Clear-to-Out Out_QF	C, E
$t_{DDRIREMCLR}$	Clear Removal	C, B
$t_{DDRIRECCLR}$	Clear Recovery	C, B

Global Resource Characteristics

A2F200 Clock Tree Topology

Clock delays are device-specific. Figure 2-27 is an example of a global tree used for clock routing. The global tree presented in Figure 2-27 is driven by a CCC located on the west side of the A2F200 device. It is used to drive all D-flip-flops in the device.

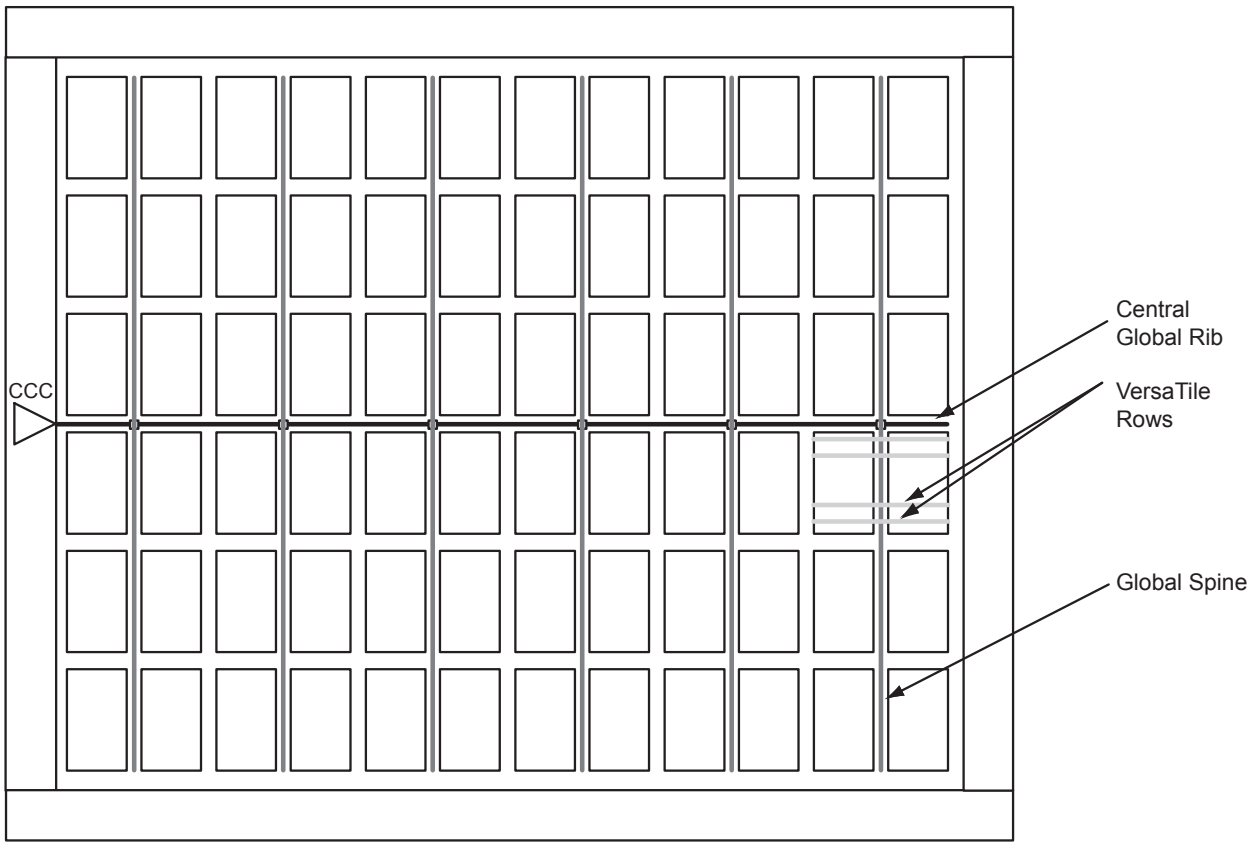

Figure 2-27 • **Example of Global Tree Use in an A2F200 Device for Clock Routing**

Global Tree Timing Characteristics

Global clock delays include the central rib delay, the spine delay, and the row delay. Delays do not include I/O input buffer clock delays, as these are I/O standard–dependent, and the clock may be driven and conditioned internally by the CCC module. For more details on clock conditioning capabilities, refer to the "Clock Conditioning Circuits" section on page 2-63. Table 2-80 through Table 2-82 on page 2-61 present minimum and maximum global clock delays for the SmartFusion cSoCs. Minimum and maximum delays are measured with minimum and maximum loading.

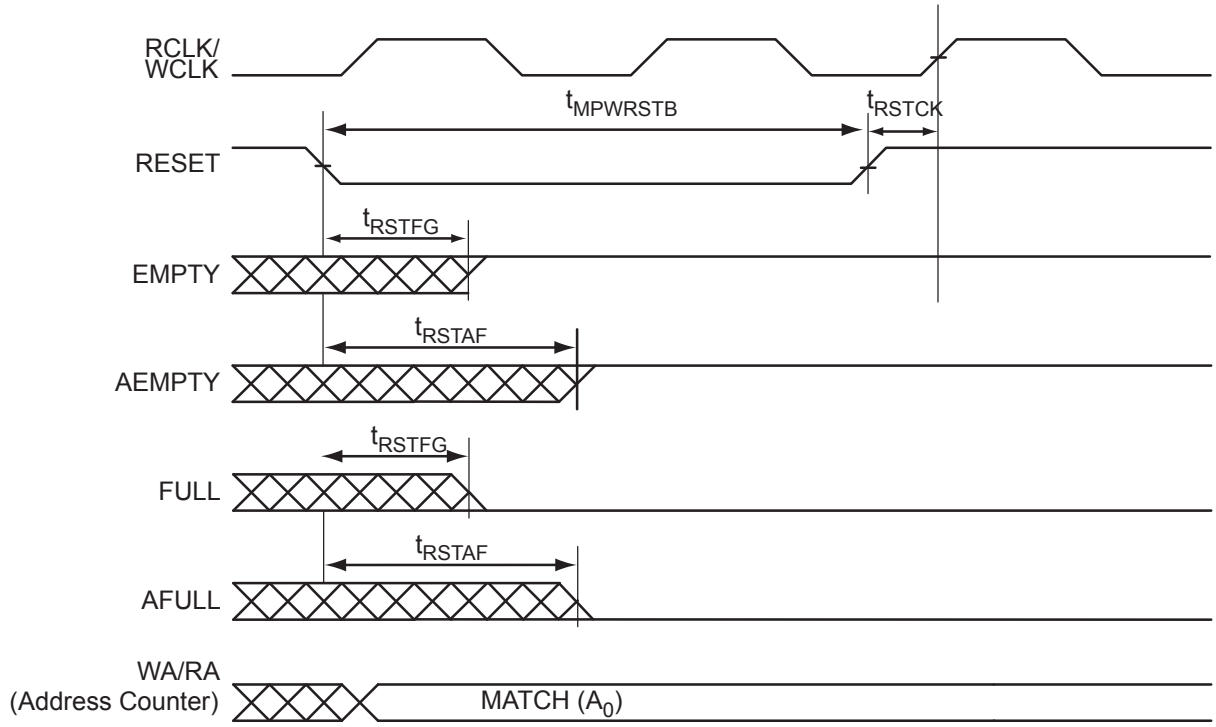

Figure 2-38 • **FIFO Reset**

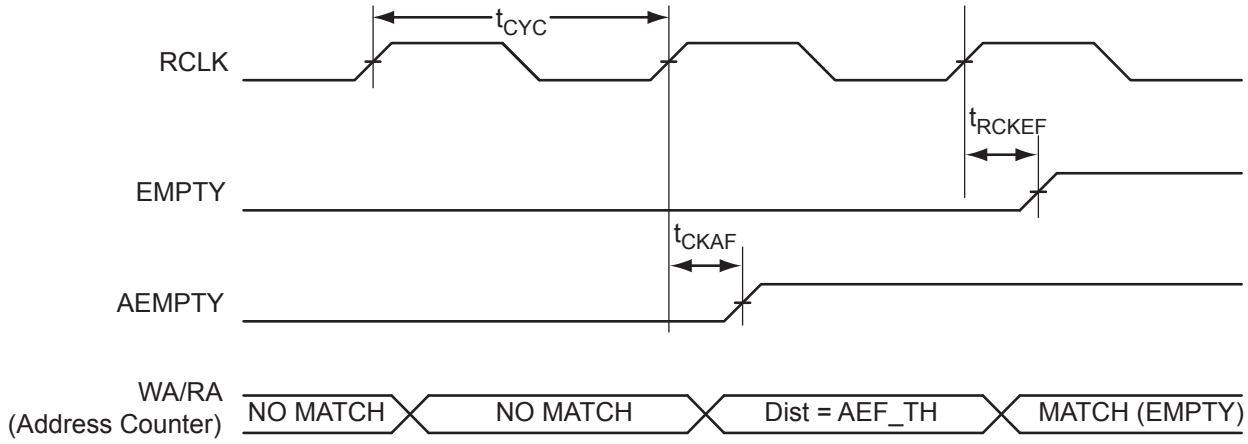

Figure 2-39 • **FIFO EMPTY Flag and AEMPTY Flag Assertion**

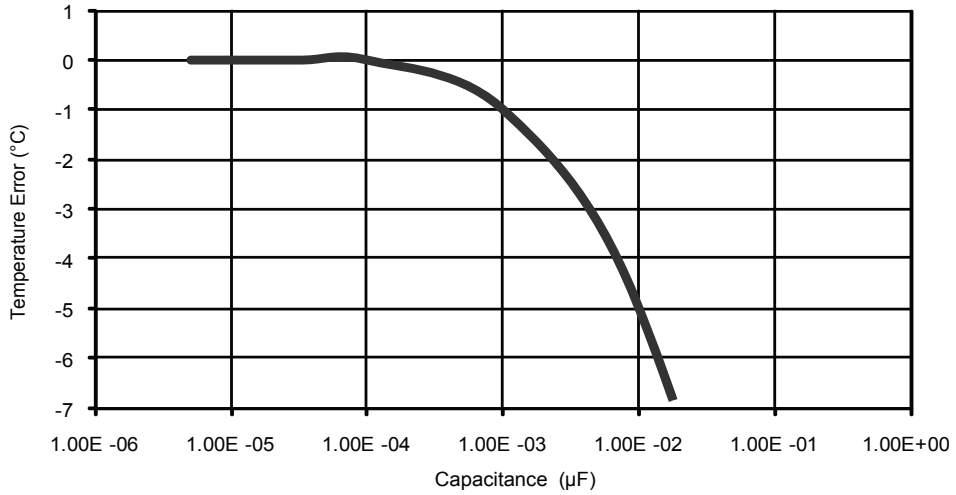

Figure 2-43 • **Temperature Error Versus External Capacitance**

Table 2-100 • **SPI Characteristics**
Commercial Case Conditions: T_J = 85ºC, VDD = 1.425 V, –1 Speed Grade (continued)

Symbol	Description and Condition	A2F060	A2F200	A2F500	Unit
sp6	Data from master (SPI_x_DO) setup time [2]	1	1	1	pclk cycles
sp7	Data from master (SPI_x_DO) hold time [2]	1	1	1	pclk cycles
sp8	SPI_x_DI setup time [2]	1	1	1	pclk cycles
sp9	SPI_x_DI hold time [2]	1	1	1	pclk cycles

Notes:

1. *These values are provided for a load of 35 pF. For board design considerations and detailed output buffer resistances, use the corresponding IBIS models located on the Microsemi SoC Products Group website: http://www.microsemi.com/index.php?option=com_microsemi&Itemid=489&lang=en&view=salescontact.*
2. *For allowable pclk configurations, refer to the Serial Peripheral Interface Controller section in the SmartFusion Microcontroller Subsystem User's Guide.*

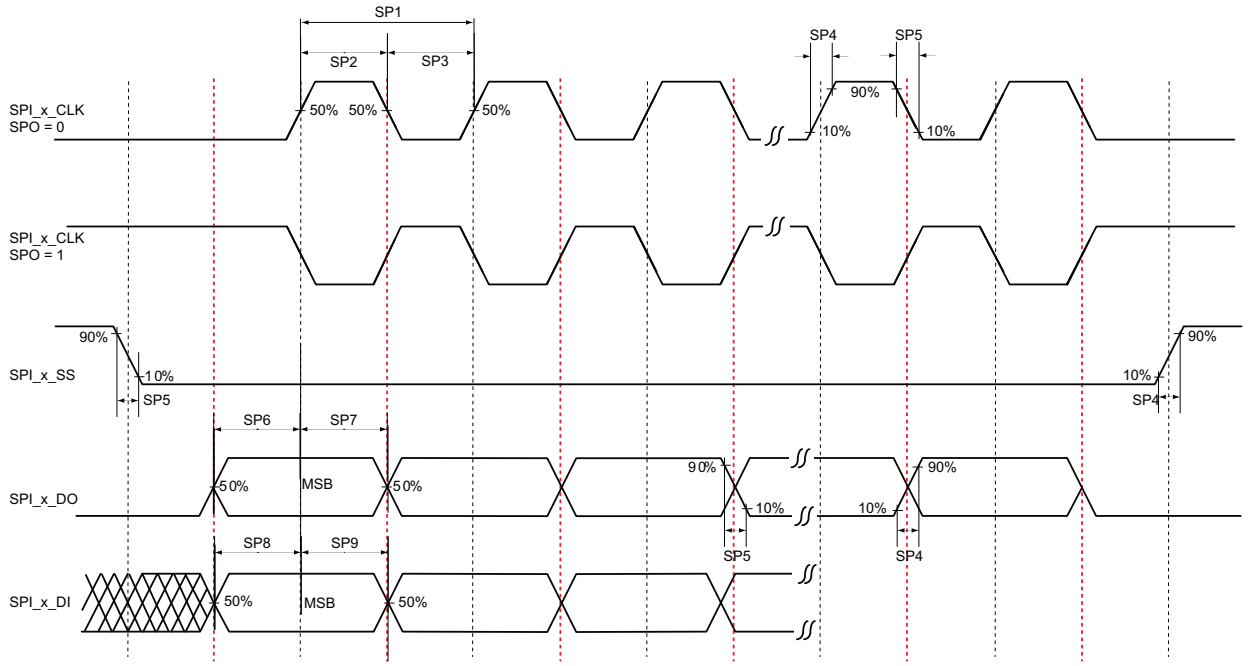

Figure 2-47 • **SPI Timing for a Single Frame Transfer in Motorola Mode (SPH = 1)**

Compile and Debug

Microsemi's SoftConsole is a free Eclipse-based IDE that enables the rapid production of C and C++ executables for Microsemi FPGA and cSoCs using Cortex-M3, Cortex-M1 and Core8051s. For SmartFusion support, SoftConsole includes the GNU C/C++ compiler and GDB debugger. Additional examples can be found on the SoftConsole page:

- Using UART with SmartFusion: SoftConsole Standalone Flow Tutorial
  - Design Files
- Displaying POT Level with LEDs: Libero SoC and SoftConsole Flow Tutorial for SmartFusion
  - Design Files

IAR Embedded Workbench® for ARM/Cortex is an integrated development environment for building and debugging embedded ARM applications using assembler, C and C++. It includes a project manager, editor, build and debugger tools with support for RTOS-aware debugging on hardware or in a simulator.

- Designing SmartFusion cSoC with IAR Systems
- IAR Embedded Workbench IDE User Guide for ARM
- Download Evaluation or Kickstart version of IAR Embedded Workbench for ARM

Keil's Microcontroller Development Kit comes in two editions: MDK-ARM and MDK Basic. Both editions feature µVision®, the ARM Compiler, MicroLib, and RTX, but the MDK Basic edition is limited to 256K so that small applications are more affordable.

- Designing SmartFusion cSoC with Keil
- Using Keil µVision and Microsemi SmartFusion cSoC
  - Programming file for use with this tutorial
- Keil Microcontroller Development Kit for ARM Product Manuals
- Download Evaluation version of Keil MDK-ARM

CMSIS COMPLIANT ARM® Cortex™ Microcontroller Software Interface Standard	Microsemi	KEIL™ An ARM® Company	IAR SYSTEMS
Software IDE	**SoftConsole**	**Vision IDE**	**Embedded Workbench**
Website	www.microsemi.com/soc	www.keil.com	www.iar.com
Free versions from SoC Products Group	Free with Libero SoC	32 K code limited	32 K code limited
Available from Vendor	N/A	Full version	Full version
Compiler	GNU GCC	RealView C/C++	IAR ARM Compiler
Debugger	GDB debug	Vision Debugger	C-SPY Debugger
Instruction Set Simulator	No	Vision Simulator	Yes
Debug Hardware	FlashPro4	ULINK2 or ULINK-ME	J-LINK or J-LINK Lite

Operating Systems

FreeRTOS™ is a portable, open source, royalty free, mini real-time kernel (a free-to-download and free-to-deploy RTOS that can be used in commercial applications without any requirement to expose your proprietary source code). FreeRTOS is scalable and designed specifically for small embedded systems. This FreeRTOS version ported by Microsemi is 6.0.1. For more information, visit the FreeRTOS website: www.freertos.org

- SmartFusion Webserver Demo Using uIP and FreeRTOS
- SmartFusion cSoC: Running Webserver, TFTP on lwIP TCP/IP Stack Application Note

Name	Type	Polarity/Bus Size	Description
NCAP		1	Negative capacitor connection. This is the negative terminal of the charge pump. A capacitor, with a 2.2 µF recommended value, is required to connect between PCAP and NCAP. Analog charge pump capacitors are not needed if none of the analog SCB features are used and none of the SDDs are used. In that case it should be left unconnected.
PCAP		1	Positive Capacitor connection. This is the positive terminal of the charge pump. A capacitor, with a 2.2 µF recommended value, is required to connect between PCAP and NCAP. If this pin is not used, it must be left unconnected/floating. In this case, no capacitor is needed. Analog charge pump capacitors are not needed if none of the analog SCB features are used, and none of the SDDs are used.
PTBASE		1	Pass transistor base connection This is the control signal of the voltage regulator. This pin should be connected to the base of an external pass transistor used with the 1.5 V internal voltage regulator and can be floating if not used.
PTEM		1	Pass transistor emitter connection. This is the feedback input of the voltage regulator. This pin should be connected to the emitter of an external pass transistor used with the 1.5 V internal voltage regulator and can be floating if not used.
MSS_RESET_N		Low	Low Reset signal which can be used as an external reset and can also be used as a system level reset under control of the Cortex-M3 processor. MSS_RESET_N is an output asserted low after power-on reset. The direction of MSS_RESET_N changes during the execution of the Microsemi System Boot when chip-level reset is enabled. The Microsemi System Boot reconfigures MSS_RESET_N to become a reset input signal when chip-level reset is enabled. It has an internal pull-up so it can be left floating. In the current software, the MSS_RESET_N is modeled as an external input signal only.
PU_N	In	Low	Push-button is the connection for the external momentary switch used to turn on the 1.5 V voltage regulator and can be floating if not used.

Pin No.	CS288		
	A2F060 Function	A2F200 Function	A2F500 Function
H19	VCCFPGAIOB1	VCCFPGAIOB1	VCCFPGAIOB1
H21	NC	GDB2/IO33PDB1V0	GDB2/IO42PDB1V0
J1	EMC_DB[4]/IO38NPB5V0	EMC_DB[4]/GEA0/IO61NPB5V0	EMC_DB[4]/GEA0/IO78NPB5V0
J3	EMC_DB[8]/IO40NPB5V0	EMC_DB[8]/GEC0/IO63NPB5V0	EMC_DB[8]/GEC0/IO80NPB5V0
J5	EMC_DB[1]/IO36PDB5V0	EMC_DB[1]/GEB2/IO59PDB5V0	EMC_DB[1]/GEB2/IO76PDB5V0
J6	EMC_DB[6]/IO39NDB5V0	EMC_DB[6]/GEB0/IO62NDB5V0	EMC_DB[6]/GEB0/IO79NDB5V0
J7	VCCFPGAIOB5	VCCFPGAIOB5	VCCFPGAIOB5
J8	VCC	VCC	VCC
J9	GND	GND	GND
J10	VCC	VCC	VCC
J11	GND	GND	GND
J12	VCC	VCC	VCC
J13	GND	GND	GND
J14	VCC	VCC	VCC
J15	VPP	VPP	VPP
J16	NC	IO32NPB1V0	IO41NPB1V0
J17	NC	GNDQ	GNDQ
J19	VCCMAINXTAL	VCCMAINXTAL	VCCMAINXTAL
J21	NC	GDA2/IO33NDB1V0	GDA2/IO42NDB1V0
K1	GND	GND	GND
K3	EMC_DB[5]/IO38PPB5V0	EMC_DB[5]/GEA1/IO61PPB5V0	EMC_DB[5]/GEA1/IO78PPB5V0
K5	EMC_DB[0]/IO36NDB5V0	EMC_DB[0]/GEA2/IO59NDB5V0	EMC_DB[0]/GEA2/IO76NDB5V0
K6	EMC_DB[3]/IO37PPB5V0	EMC_DB[3]/GEC2/IO60PPB5V0	EMC_DB[3]/GEC2/IO77PPB5V0
K8	GND	GND	GND
K9	VCC	VCC	VCC
K10	GND	GND	GND
K11	VCC	VCC	VCC
K12	GND	GND	GND
K13	VCC	VCC	VCC
K14	GND	GND	GND
K16	LPXOUT	LPXOUT	LPXOUT

Notes:

1. *Shading denotes pins that do not have completely identical functions from density to density. For example, the bank assignment can be different for an I/O, or the function might be available only on a larger density device.*
2. **: Indicates that the signal assigned to the pins as a CLKBUF/CLKBUF_LVPECL/CLKBUF_LVDS goes through a glitchless mux. In order for the glitchless mux to operate correctly, the signal must be a free-running clock signal. Refer to the 'Glitchless MUX' section in the SmartFusion Microcontroller Subsystem User's Guide for more details.*

Pin Number	PQ208	
	A2F200	A2F500
1	VCCPLL	VCCPLL0
2	VCOMPLA	VCOMPLA0
3	GNDQ	GNDQ
4	EMC_DB[15]/GAA2/IO71PDB5V0	GAA2/IO88PDB5V0
5	EMC_DB[14]/GAB2/IO71NDB5V0	GAB2/IO88NDB5V0
6	EMC_DB[13]/GAC2/IO70PDB5V0	GAC2/IO87PDB5V0
7	EMC_DB[12]/IO70NDB5V0	IO87NDB5V0
8	VCC	VCC
9	GND	GND
10	VCCFPGAIOB5	VCCFPGAIOB5
11	EMC_DB[11]/IO69PDB5V0	IO86PDB5V0
12	EMC_DB[10]/IO69NDB5V0	IO86NDB5V0
13	GFA2/IO68PSB5V0	GFA2/IO85PSB5V0
14	GFA1/IO64PDB5V0	GFA1/IO81PDB5V0
15	GFA0/IO64NDB5V0	GFA0/IO81NDB5V0
16	EMC_DB[9]/GEC1/IO63PDB5V0	GEC1/IO80PDB5V0
17	EMC_DB[8]/GEC0/IO63NDB5V0	GEC0/IO80NDB5V0
18	EMC_DB[7]/GEB1/IO62PDB5V0	GEB1/IO79PDB5V0
19	EMC_DB[6]/GEB0/IO62NDB5V0	GEB0/IO79NDB5V0
20	EMC_DB[5]/GEA1/IO61PDB5V0	GEA1/IO78PDB5V0
21	EMC_DB[4]/GEA0/IO61NDB5V0	GEA0/IO78NDB5V0
22	VCC	VCC
23	GND	GND
24	VCCFPGAIOB5	VCCFPGAIOB5
25	EMC_DB[3]/GEC2/IO60PDB5V0	GEC2/IO77PDB5V0
26	EMC_DB[2]/IO60NDB5V0	IO77NDB5V0
27	EMC_DB[1]/GEB2/IO59PDB5V0	GEB2/IO76PDB5V0
28	EMC_DB[0]/GEA2/IO59NDB5V0	GEA2/IO76NDB5V0
29	VCC	VCC
30	GND	GND
31	GNDRCOSC	GNDRCOSC

Notes:

1. *Shading denotes pins that do not have completely identical functions from density to density. For example, the bank assignment can be different for an I/O, or the function might be available only on a larger density device.*
2. **: Indicates that the signal assigned to the pins as a CLKBUF/CLKBUF_LVPECL/CLKBUF_LVDS goes through a glitchless mux. In order for the glitchless mux to operate correctly, the signal must be a free-running clock signal. Refer to the 'Glitchless MUX' section in the SmartFusion Microcontroller Subsystem User's Guide for more details.*

Pin Number	PQ208	
	A2F200	A2F500
156	GNDQ	GNDQ
157	GNDQ	GNDQ
158	VCCFPGAIOB0	VCCFPGAIOB0
159	GBA1/IO19PDB0V0	GBA1/IO23PDB0V0
160	GBA0/IO19NDB0V0	GBA0/IO23NDB0V0
161	VCCFPGAIOB0	VCCFPGAIOB0
162	GND	GND
163	VCC	VCC
164	EMC_AB[25]/IO16PDB0V0	IO21PDB0V0
165	EMC_AB[24]/IO16NDB0V0	IO21NDB0V0
166	EMC_AB[23]/IO15PDB0V0	IO20PDB0V0
167	EMC_AB[22]/IO15NDB0V0	IO20NDB0V0
168	EMC_AB[21]/IO14PDB0V0	IO19PDB0V0
169	EMC_AB[20]/IO14NDB0V0	IO19NDB0V0
170	EMC_AB[19]/IO13PDB0V0	IO18PDB0V0
171	EMC_AB[18]/IO13NDB0V0	IO18NDB0V0
172	EMC_AB[17]/IO12PDB0V0	IO17PDB0V0
173	EMC_AB[16]/IO12NDB0V0	IO17NDB0V0
174	VCCFPGAIOB0	VCCFPGAIOB0
175	GND	GND
176	VCC	VCC
177	EMC_AB[15]/IO11PDB0V0	IO14PDB0V0
178	EMC_AB[14]/IO11NDB0V0	IO14NDB0V0
179	EMC_AB[13]/IO10PDB0V0	IO13PDB0V0
180	EMC_AB[12]/IO10NDB0V0	IO13NDB0V0
181	EMC_AB[11]/IO09PDB0V0	IO12PDB0V0
182	EMC_AB[10]/IO09NDB0V0	IO12NDB0V0
183	EMC_AB[9]/IO08PDB0V0	IO11PDB0V0
184	EMC_AB[8]/IO08NDB0V0	IO11NDB0V0
185	EMC_AB[7]/IO07PDB0V0	IO10PDB0V0
186	EMC_AB[6]/IO07NDB0V0	IO10NDB0V0

Notes:

1. *Shading denotes pins that do not have completely identical functions from density to density. For example, the bank assignment can be different for an I/O, or the function might be available only on a larger density device.*
2. **: Indicates that the signal assigned to the pins as a CLKBUF/CLKBUF_LVPECL/CLKBUF_LVDS goes through a glitchless mux. In order for the glitchless mux to operate correctly, the signal must be a free-running clock signal. Refer to the 'Glitchless MUX' section in the SmartFusion Microcontroller Subsystem User's Guide for more details.*