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Details

Product Status	Active
Core Processor	PIC
Core Size	8-Bit
Speed	20MHz
Connectivity	I ² C, SPI
Peripherals	Brown-out Detect/Reset, POR, PWM, WDT
Number of I/O	11
Program Memory Size	3.5KB (2K x 14)
Program Memory Type	FLASH
EEPROM Size	-
RAM Size	128 x 8
Voltage - Supply (Vcc/Vdd)	1.8V ~ 3.6V
Data Converters	A/D 8x10b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 125°C (TA)
Mounting Type	Surface Mount
Package / Case	14-SOIC (0.154", 3.90mm Width)
Supplier Device Package	14-SOIC
Purchase URL	https://www.e-xfl.com/product-detail/microchip-technology/pic16lf1503-e-sl

3.6.1 TRADITIONAL DATA MEMORY

The traditional data memory is a region from FSR address 0x000 to FSR address 0xFFF. The addresses correspond to the absolute addresses of all SFR, GPR and common registers.

FIGURE 3-9: TRADITIONAL DATA MEMORY MAP

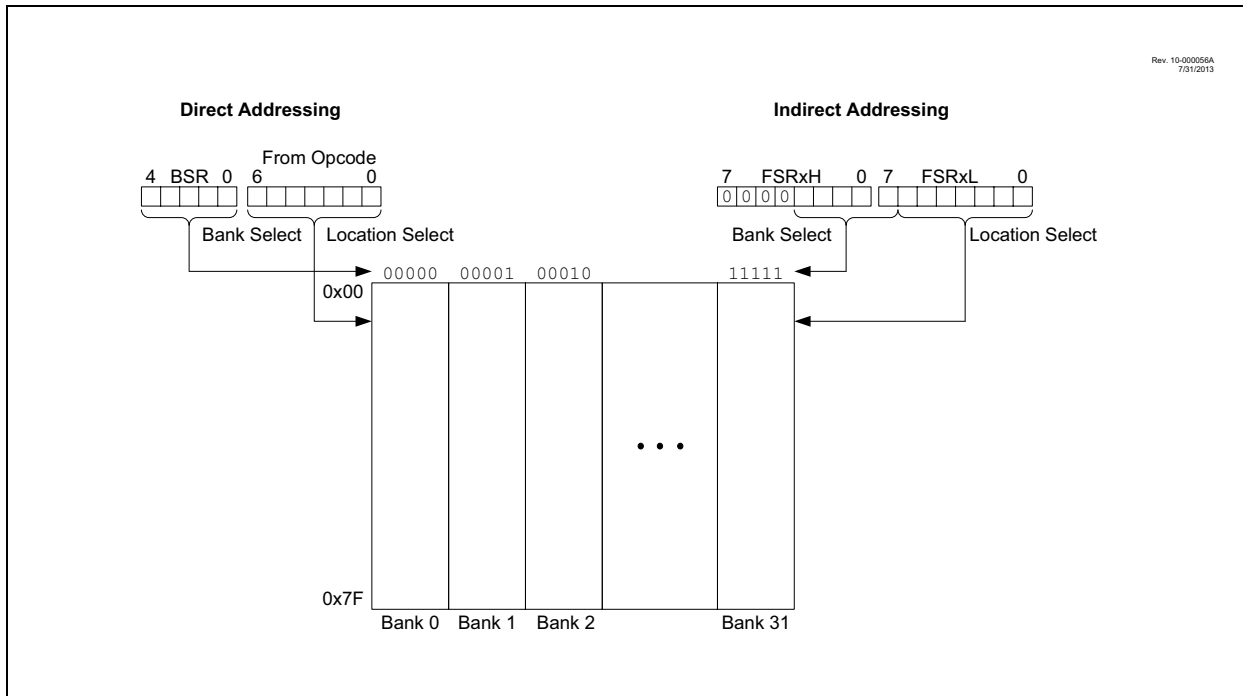
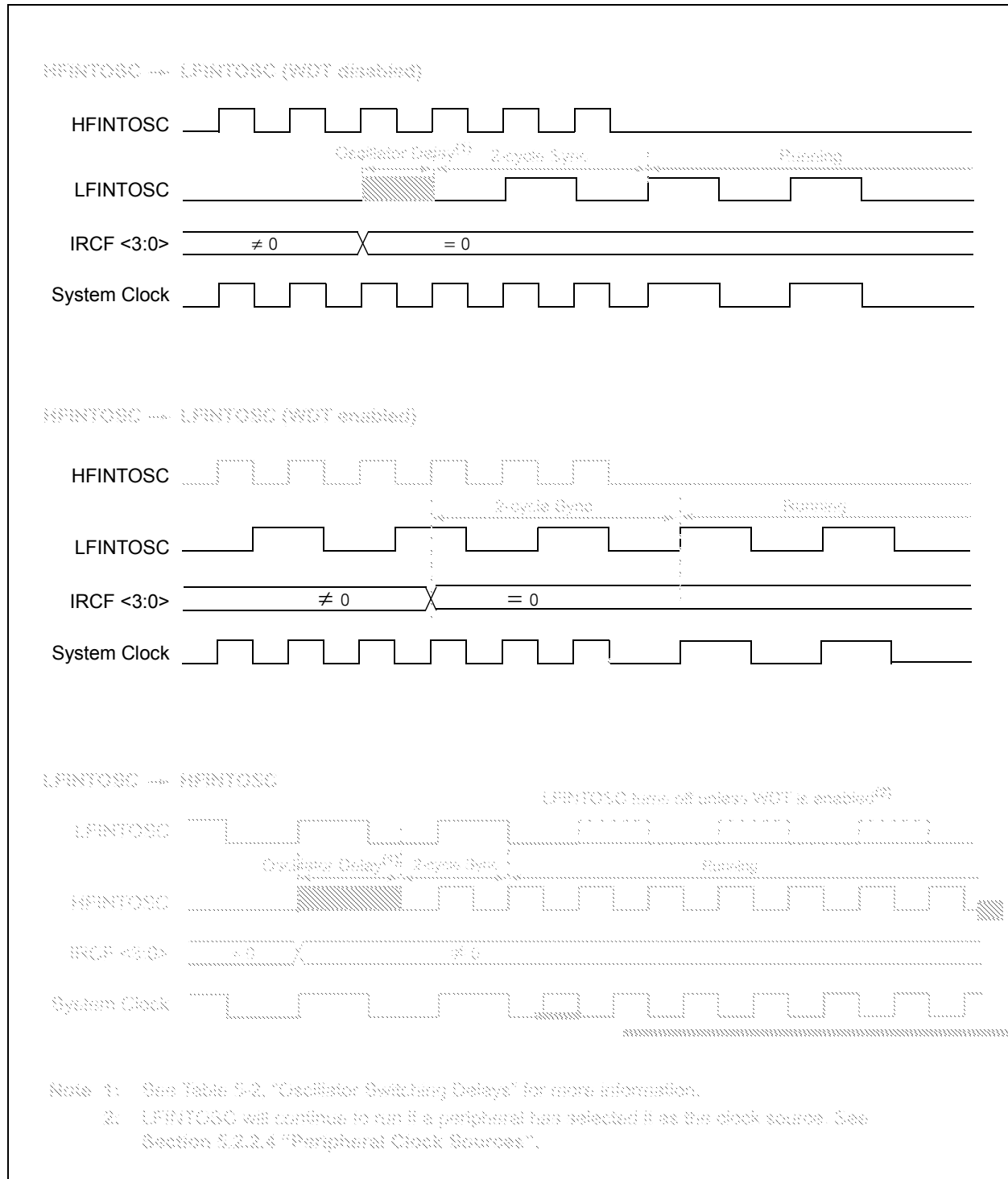


FIGURE 5-3: INTERNAL OSCILLATOR SWITCH TIMING



9.6 Register Definitions: Watchdog Timer Control

REGISTER 9-1: WDTCON: WATCHDOG TIMER CONTROL REGISTER

U-0	U-0	R/W-0/0	R/W-1/1	R/W-0/0	R/W-1/1	R/W-1/1	R/W-0/0
—	—	WDTPS<4:0>					SWDTEN
bit 7							bit 0

Legend:

R = Readable bit W = Writable bit U = Unimplemented bit, read as '0'
 u = Bit is unchanged x = Bit is unknown -n/n = Value at POR and BOR/Value at all other Resets
 '1' = Bit is set '0' = Bit is cleared

bit 7-6 **Unimplemented:** Read as '0'

bit 5-1 **WDTPS<4:0>:** Watchdog Timer Period Select bits⁽¹⁾

Bit Value = Prescale Rate

11111 = Reserved. Results in minimum interval (1:32)

•
•
•

10011 = Reserved. Results in minimum interval (1:32)

10010 = 1:8388608 (2^{23}) (Interval 256s nominal)

10001 = 1:4194304 (2^{22}) (Interval 128s nominal)

10000 = 1:2097152 (2^{21}) (Interval 64s nominal)

01111 = 1:1048576 (2^{20}) (Interval 32s nominal)

01110 = 1:524288 (2^{19}) (Interval 16s nominal)

01101 = 1:262144 (2^{18}) (Interval 8s nominal)

01100 = 1:131072 (2^{17}) (Interval 4s nominal)

01011 = 1:65536 (Interval 2s nominal) (Reset value)

01010 = 1:32768 (Interval 1s nominal)

01001 = 1:16384 (Interval 512 ms nominal)

01000 = 1:8192 (Interval 256 ms nominal)

00111 = 1:4096 (Interval 128 ms nominal)

00110 = 1:2048 (Interval 64 ms nominal)

00101 = 1:1024 (Interval 32 ms nominal)

00100 = 1:512 (Interval 16 ms nominal)

00011 = 1:256 (Interval 8 ms nominal)

00010 = 1:128 (Interval 4 ms nominal)

00001 = 1:64 (Interval 2 ms nominal)

00000 = 1:32 (Interval 1 ms nominal)

bit 0 **SWDTEN:** Software Enable/Disable for Watchdog Timer bit

If WDTE<1:0> = 1x:

This bit is ignored.

If WDTE<1:0> = 01:

1 = WDT is turned on

0 = WDT is turned off

If WDTE<1:0> = 00:

This bit is ignored.

Note 1: Times are approximate. WDT time is based on 31 kHz LFINTOSC.

10.2.1 READING THE FLASH PROGRAM MEMORY

To read a program memory location, the user must:

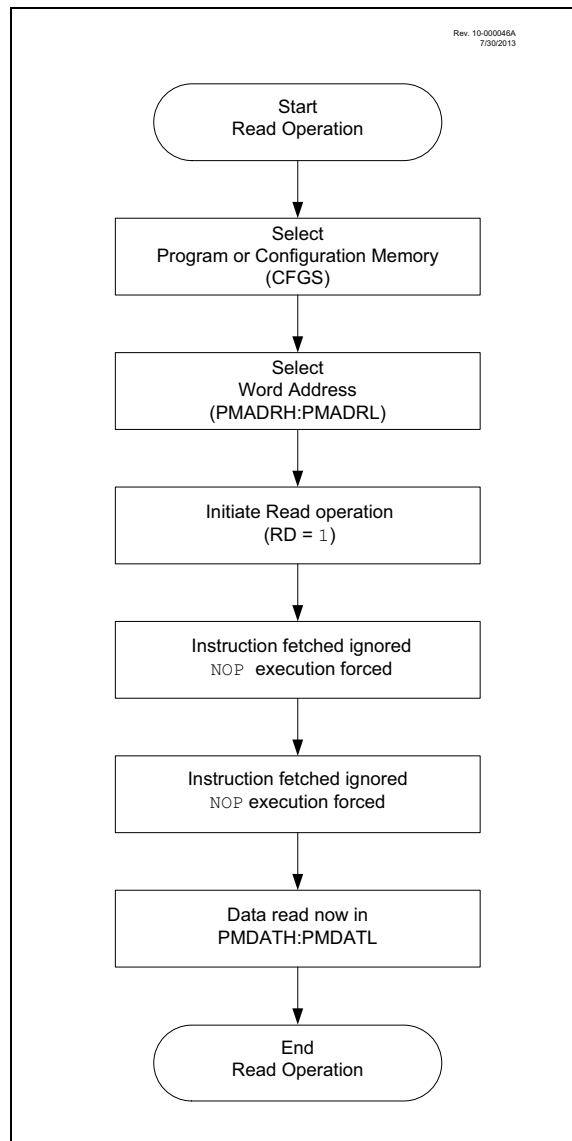
1. Write the desired address to the PMADRH:PMADRL register pair.
2. Clear the CFGS bit of the PMCON1 register.
3. Then, set control bit RD of the PMCON1 register.

Once the read control bit is set, the program memory Flash controller will use the second instruction cycle to read the data. This causes the second instruction immediately following the “BSF PMCON1, RD” instruction to be ignored. The data is available in the very next cycle, in the PMDATH:PMDATL register pair; therefore, it can be read as two bytes in the following instructions.

PMDATH:PMDATL register pair will hold this value until another read or until it is written to by the user.

Note: The two instructions following a program memory read are required to be NOPs. This prevents the user from executing a 2-cycle instruction on the next instruction after the RD bit is set.

FIGURE 10-1: FLASH PROGRAM MEMORY READ FLOWCHART



10.2.4 WRITING TO FLASH PROGRAM MEMORY

Program memory is programmed using the following steps:

1. Load the address in PMADRH:PMADRL of the row to be programmed.
2. Load each write latch with data.
3. Initiate a programming operation.
4. Repeat steps 1 through 3 until all data is written.

Before writing to program memory, the word(s) to be written must be erased or previously unwritten. Program memory can only be erased one row at a time. No automatic erase occurs upon the initiation of the write.

Program memory can be written one or more words at a time. The maximum number of words written at one time is equal to the number of write latches. See Figure 10-5 (row writes to program memory with 16 write latches) for more details.

The write latches are aligned to the Flash row address boundary defined by the upper 10-bits of PMADRH:PMADRL, (PMADRH<6:0>:PMADRL<7:5>) with the lower five bits of PMADRL, (PMADRL<4:0>) determining the write latch being loaded. Write operations do not cross these boundaries. At the completion of a program memory write operation, the data in the write latches is reset to contain 0x3FFF.

The following steps should be completed to load the write latches and program a row of program memory. These steps are divided into two parts. First, each write latch is loaded with data from the PMDATH:PMDATL using the unlock sequence with LWLO = 1. When the last word to be loaded into the write latch is ready, the LWLO bit is cleared and the unlock sequence executed. This initiates the programming operation, writing all the latches into Flash program memory.

Note: The special unlock sequence is required to load a write latch with data or initiate a Flash programming operation. If the unlock sequence is interrupted, writing to the latches or program memory will not be initiated.

1. Set the WREN bit of the PMCON1 register.
2. Clear the CFGS bit of the PMCON1 register.
3. Set the LWLO bit of the PMCON1 register. When the LWLO bit of the PMCON1 register is '1', the write sequence will only load the write latches and will not initiate the write to Flash program memory.
4. Load the PMADRH:PMADRL register pair with the address of the location to be written.
5. Load the PMDATH:PMDATL register pair with the program memory data to be written.
6. Execute the unlock sequence (**Section 10.2.2 "Flash Memory Unlock Sequence"**). The write latch is now loaded.
7. Increment the PMADRH:PMADRL register pair to point to the next location.
8. Repeat steps 5 through 7 until all but the last write latch has been loaded.
9. Clear the LWLO bit of the PMCON1 register. When the LWLO bit of the PMCON1 register is '0', the write sequence will initiate the write to Flash program memory.
10. Load the PMDATH:PMDATL register pair with the program memory data to be written.
11. Execute the unlock sequence (**Section 10.2.2 "Flash Memory Unlock Sequence"**). The entire program memory latch content is now written to Flash program memory.

Note: The program memory write latches are reset to the blank state (0x3FFF) at the completion of every write or erase operation. As a result, it is not necessary to load all the program memory write latches. Unloaded latches will remain in the blank state.

An example of the complete write sequence is shown in Example 10-3. The initial address is loaded into the PMADRH:PMADRL register pair; the data is loaded using indirect addressing.

15.2 ADC Operation

15.2.1 STARTING A CONVERSION

To enable the ADC module, the ADON bit of the ADCON0 register must be set to a '1'. Setting the GO/DONE bit of the ADCON0 register to a '1' will start the Analog-to-Digital conversion.

Note: The GO/DONE bit should not be set in the same instruction that turns on the ADC. Refer to **Section 15.2.6 “ADC Conversion Procedure”**.

15.2.2 COMPLETION OF A CONVERSION

When the conversion is complete, the ADC module will:

- Clear the GO/DONE bit
- Set the ADIF Interrupt Flag bit
- Update the ADRESH and ADRESL registers with new conversion result

15.2.3 TERMINATING A CONVERSION

If a conversion must be terminated before completion, the GO/DONE bit can be cleared in software. The ADRESH and ADRESL registers will be updated with the partially complete Analog-to-Digital conversion sample. Incomplete bits will match the last bit converted.

Note: A device Reset forces all registers to their Reset state. Thus, the ADC module is turned off and any pending conversion is terminated.

15.2.4 ADC OPERATION DURING SLEEP

The ADC module can operate during Sleep. This requires the ADC clock source to be set to the FRC option. Performing the ADC conversion during Sleep can reduce system noise. If the ADC interrupt is enabled, the device will wake-up from Sleep when the conversion completes. If the ADC interrupt is disabled, the ADC module is turned off after the conversion completes, although the ADON bit remains set.

When the ADC clock source is something other than FRC, a SLEEP instruction causes the present conversion to be aborted and the ADC module is turned off, although the ADON bit remains set.

15.2.5 AUTO-CONVERSION TRIGGER

The auto-conversion trigger allows periodic ADC measurements without software intervention. When a rising edge of the selected source occurs, the GO/DONE bit is set by hardware.

The auto-conversion trigger source is selected with the TRIGSEL<3:0> bits of the ADCON2 register.

Using the auto-conversion trigger does not assure proper ADC timing. It is the user's responsibility to ensure that the ADC timing requirements are met.

See Table 15-2 for auto-conversion sources.

TABLE 15-2: AUTO-CONVERSION SOURCES

Source Peripheral	Signal Name
Timer0	T0_overflow
Timer1	T1_overflow
Timer2	T2_match
Comparator C1	C1OUT_sync
Comparator C2	C2OUT_sync
CLC1	LC1_out
CLC2	LC2_out

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17.0 COMPARATOR MODULE

Comparators are used to interface analog circuits to a digital circuit by comparing two analog voltages and providing a digital indication of their relative magnitudes. Comparators are very useful mixed signal building blocks because they provide analog functionality independent of program execution. The analog comparator module includes the following features:

- Independent comparator control
- Programmable input selection
- Comparator output is available internally/externally
- Programmable output polarity
- Interrupt-on-change
- Wake-up from Sleep
- Programmable Speed/Power optimization
- PWM shutdown
- Programmable and fixed voltage reference

17.1 Comparator Overview

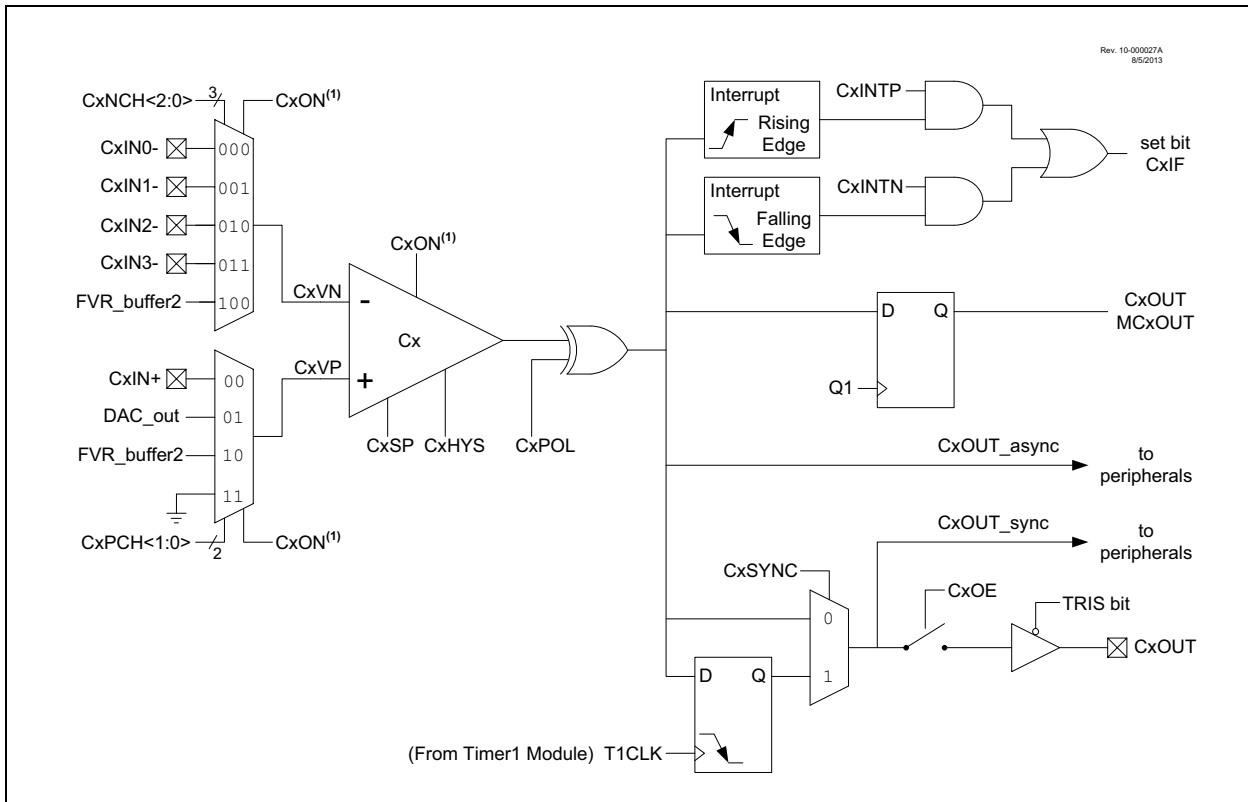
A single comparator is shown in Figure 17-2 along with the relationship between the analog input levels and the digital output. When the analog voltage at V_{IN+} is less than the analog voltage at V_{IN-} , the output of the comparator is a digital low level. When the analog voltage at V_{IN+} is greater than the analog voltage at V_{IN-} , the output of the comparator is a digital high level.

The comparators available for this device are listed in Table 17-1.

TABLE 17-1: AVAILABLE COMPARATORS

Device	C1	C2
PIC16(L)F1503	•	•

FIGURE 17-1: COMPARATOR MODULE SIMPLIFIED BLOCK DIAGRAM



18.2 Register Definitions: Option Register

REGISTER 18-1: OPTION_REG: OPTION REGISTER

R/W-1/1	R/W-1/1	R/W-1/1	R/W-1/1	R/W-1/1	R/W-1/1	R/W-1/1	R/W-1/1
WPUEN	INTEDG	TMR0CS	TMR0SE	PSA	PS<2:0>		
bit 7							bit 0

Legend:

R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7	WPUEN: Weak Pull-Up Enable bit 1 = All weak pull-ups are disabled (except MCLR, if it is enabled) 0 = Weak pull-ups are enabled by individual WPUx latch values
bit 6	INTEDG: Interrupt Edge Select bit 1 = Interrupt on rising edge of INT pin 0 = Interrupt on falling edge of INT pin
bit 5	TMR0CS: Timer0 Clock Source Select bit 1 = Transition on T0CKI pin 0 = Internal instruction cycle clock (Fosc/4)
bit 4	TMR0SE: Timer0 Source Edge Select bit 1 = Increment on high-to-low transition on T0CKI pin 0 = Increment on low-to-high transition on T0CKI pin
bit 3	PSA: Prescaler Assignment bit 1 = Prescaler is not assigned to the Timer0 module 0 = Prescaler is assigned to the Timer0 module
bit 2-0	PS<2:0>: Prescaler Rate Select bits

Bit Value	Timer0 Rate
000	1 : 2
001	1 : 4
010	1 : 8
011	1 : 16
100	1 : 32
101	1 : 64
110	1 : 128
111	1 : 256

TABLE 18-1: SUMMARY OF REGISTERS ASSOCIATED WITH TIMER0

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Register on Page
ADCON2	TRIGSEL<3:0>				—	—	—	—	121
INTCON	GIE	PEIE	TMR0IE	INTE	IOCIE	TMR0IF	INTF	IOCIF	64
OPTION_REG	WPUEN	INTEDG	TMR0CS	TMR0SE	PSA	PS<2:0>			139
TMR0	Holding Register for the 8-bit Timer0 Count								137*
TRISA	—	—	TRISA5	TRISA4	— ⁽¹⁾	TRISA2	TRISA1	TRISA0	98

Legend: — = Unimplemented location, read as '0'. Shaded cells are not used by the Timer0 module.

* Page provides register information.

Note 1: Unimplemented, read as '1'.

19.5.2.1 T1G Pin Gate Operation

The T1G pin is one source for Timer1 gate control. It can be used to supply an external source to the Timer1 gate circuitry.

19.5.2.2 Timer0 Overflow Gate Operation

When Timer0 increments from FFh to 00h, a low-to-high pulse will automatically be generated and internally supplied to the Timer1 gate circuitry.

19.5.3 TIMER1 GATE TOGGLE MODE

When Timer1 Gate Toggle mode is enabled, it is possible to measure the full-cycle length of a Timer1 gate signal, as opposed to the duration of a single level pulse.

The Timer1 gate source is routed through a flip-flop that changes state on every incrementing edge of the signal. See Figure 19-4 for timing details.

Timer1 Gate Toggle mode is enabled by setting the T1GTM bit of the T1GCON register. When the T1GTM bit is cleared, the flip-flop is cleared and held clear. This is necessary in order to control which edge is measured.

Note: Enabling Toggle mode at the same time as changing the gate polarity may result in indeterminate operation.

19.5.4 TIMER1 GATE SINGLE-PULSE MODE

When Timer1 Gate Single-Pulse mode is enabled, it is possible to capture a single pulse gate event. Timer1 Gate Single-Pulse mode is first enabled by setting the T1GSPM bit in the T1GCON register. Next, the T1GGO/DONE bit in the T1GCON register must be set. The Timer1 will be fully enabled on the next incrementing edge. On the next trailing edge of the pulse, the T1GGO/DONE bit will automatically be cleared. No other gate events will be allowed to increment Timer1 until the T1GGO/DONE bit is once again set in software. See Figure 19-5 for timing details.

If the Single Pulse Gate mode is disabled by clearing the T1GSPM bit in the T1GCON register, the T1GGO/DONE bit should also be cleared.

Enabling the Toggle mode and the Single-Pulse mode simultaneously will permit both sections to work together. This allows the cycle times on the Timer1 gate source to be measured. See Figure 19-6 for timing details.

19.5.5 TIMER1 GATE VALUE STATUS

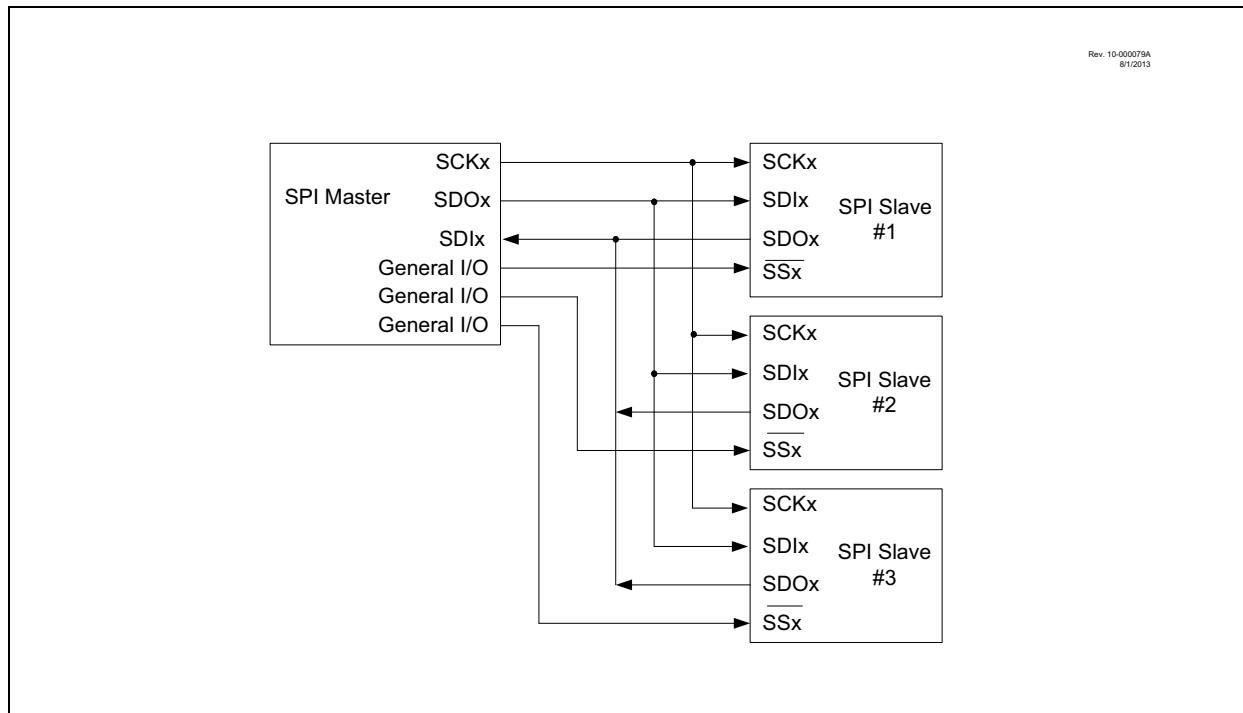
When Timer1 Gate Value Status is utilized, it is possible to read the most current level of the gate control value. The value is stored in the T1GVAL bit in the T1GCON register. The T1GVAL bit is valid even when the Timer1 gate is not enabled (TMR1GE bit is cleared).

19.5.6 TIMER1 GATE EVENT INTERRUPT

When Timer1 Gate Event Interrupt is enabled, it is possible to generate an interrupt upon the completion of a gate event. When the falling edge of T1GVAL occurs, the TMR1GIF flag bit in the PIR1 register will be set. If the TMR1GIE bit in the PIE1 register is set, then an interrupt will be recognized.

The TMR1GIF flag bit operates even when the Timer1 gate is not enabled (TMR1GE bit is cleared).

FIGURE 21-4: SPI MASTER AND MULTIPLE SLAVE CONNECTION



21.2.1 SPI MODE REGISTERS

The MSSP module has five registers for SPI mode operation. These are:

- MSSP STATUS register (SSPxSTAT)
- MSSP Control Register 1 (SSPxCON1)
- MSSP Control Register 3 (SSPxCON3)
- MSSP Data Buffer register (SSPxBUF)
- MSSP Address register (SSPxADD)
- MSSP Shift register (SSPxSR)
(Not directly accessible)

SSPxCON1 and SSPxSTAT are the control and STATUS registers in SPI mode operation. The SSPxCON1 register is readable and writable. The lower six bits of the SSPxSTAT are read-only. The upper two bits of the SSPxSTAT are read/write.

In SPI master mode, SSPxADD can be loaded with a value used in the Baud Rate Generator. More information on the Baud Rate Generator is available in **Section 21.7 “Baud Rate Generator”**.

SSPxSR is the shift register used for shifting data in and out. SSPxBUF provides indirect access to the SSPxSR register. SSPxBUF is the buffer register to which data bytes are written, and from which data bytes are read.

In receive operations, SSPxSR and SSPxBUF together create a buffered receiver. When SSPxSR receives a complete byte, it is transferred to SSPxBUF and the SSPxIF interrupt is set.

During transmission, the SSPxBUF is not buffered. A write to SSPxBUF will write to both SSPxBUF and SSPxSR.

PIC16(L)F1503

21.5.3.3 7-bit Transmission with Address Hold Enabled

Setting the AHEN bit of the SSPxCON3 register enables additional clock stretching and interrupt generation after the eighth falling edge of a received matching address. Once a matching address has been clocked in, CKP is cleared and the SSPxIF interrupt is set.

Figure 21-19 displays a standard waveform of a 7-bit Address Slave Transmission with AHEN enabled.

1. Bus starts idle.
2. Master sends Start condition; the S bit of SSPxSTAT is set; SSPxIF is set if interrupt on Start detect is enabled.
3. Master sends matching address with $\overline{R/\overline{W}}$ bit set. After the eighth falling edge of the SCLx line the CKP bit is cleared and SSPxIF interrupt is generated.
4. Slave software clears SSPxIF.
5. Slave software reads ACKTIM bit of SSPxCON3 register, and $\overline{R/\overline{W}}$ and $\overline{D/\overline{A}}$ of the SSPxSTAT register to determine the source of the interrupt.
6. Slave reads the address value from the SSPxBUF register clearing the BF bit.
7. Slave software decides from this information if it wishes to \overline{ACK} or not \overline{ACK} and sets the ACKDT bit of the SSPxCON2 register accordingly.
8. Slave sets the CKP bit releasing SCLx.
9. Master clocks in the \overline{ACK} value from the slave.
10. Slave hardware automatically clears the CKP bit and sets SSPxIF after the \overline{ACK} if the $\overline{R/\overline{W}}$ bit is set.
11. Slave software clears SSPxIF.
12. Slave loads value to transmit to the master into SSPxBUF setting the BF bit.

Note: SSPxBUF cannot be loaded until after the \overline{ACK} .

13. Slave sets the CKP bit, releasing the clock.
14. Master clocks out the data from the slave and sends an \overline{ACK} value on the ninth SCLx pulse.
15. Slave hardware copies the \overline{ACK} value into the ACKSTAT bit of the SSPxCON2 register.
16. Steps 10-15 are repeated for each byte transmitted to the master from the slave.
17. If the master sends a not \overline{ACK} the slave releases the bus allowing the master to send a Stop and end the communication.

Note: Master must send a not \overline{ACK} on the last byte to ensure that the slave releases the SCLx line to receive a Stop.

REGISTER 23-6: CLCxGLS1: GATE 2 LOGIC SELECT REGISTER

R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u
LCxG2D4T	LCxG2D4N	LCxG2D3T	LCxG2D3N	LCxG2D2T	LCxG2D2N	LCxG2D1T	LCxG2D1N
bit 7							bit 0

Legend:

R = Readable bit

W = Writable bit

U = Unimplemented bit, read as '0'

u = Bit is unchanged

x = Bit is unknown

-n/n = Value at POR and BOR/Value at all other Resets

'1' = Bit is set

'0' = Bit is cleared

- bit 7 **LCxG2D4T:** Gate 2 Data 4 True (non-inverted) bit
 1 = lcx4T is gated into lcxg2
 0 = lcx4T is not gated into lcxg2
- bit 6 **LCxG2D4N:** Gate 2 Data 4 Negated (inverted) bit
 1 = lcx4N is gated into lcxg2
 0 = lcx4N is not gated into lcxg2
- bit 5 **LCxG2D3T:** Gate 2 Data 3 True (non-inverted) bit
 1 = lcx3T is gated into lcxg2
 0 = lcx3T is not gated into lcxg2
- bit 4 **LCxG2D3N:** Gate 2 Data 3 Negated (inverted) bit
 1 = lcx3N is gated into lcxg2
 0 = lcx3N is not gated into lcxg2
- bit 3 **LCxG2D2T:** Gate 2 Data 2 True (non-inverted) bit
 1 = lcx2T is gated into lcxg2
 0 = lcx2T is not gated into lcxg2
- bit 2 **LCxG2D2N:** Gate 2 Data 2 Negated (inverted) bit
 1 = lcx2N is gated into lcxg2
 0 = lcx2N is not gated into lcxg2
- bit 1 **LCxG2D1T:** Gate 2 Data 1 True (non-inverted) bit
 1 = lcx1T is gated into lcxg2
 0 = lcx1T is not gated into lcxg2
- bit 0 **LCxG2D1N:** Gate 2 Data 1 Negated (inverted) bit
 1 = lcx1N is gated into lcxg2
 0 = lcx1N is not gated into lcxg2

REGISTER 23-8: CLCxGLS3: GATE 4 LOGIC SELECT REGISTER

R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u	R/W-x/u
LCxG4D4T	LCxG4D4N	LCxG4D3T	LCxG4D3N	LCxG4D2T	LCxG4D2N	LCxG4D1T	LCxG4D1N
bit 7							bit 0

Legend:

R = Readable bit

W = Writable bit

U = Unimplemented bit, read as '0'

u = Bit is unchanged

x = Bit is unknown

-n/n = Value at POR and BOR/Value at all other Resets

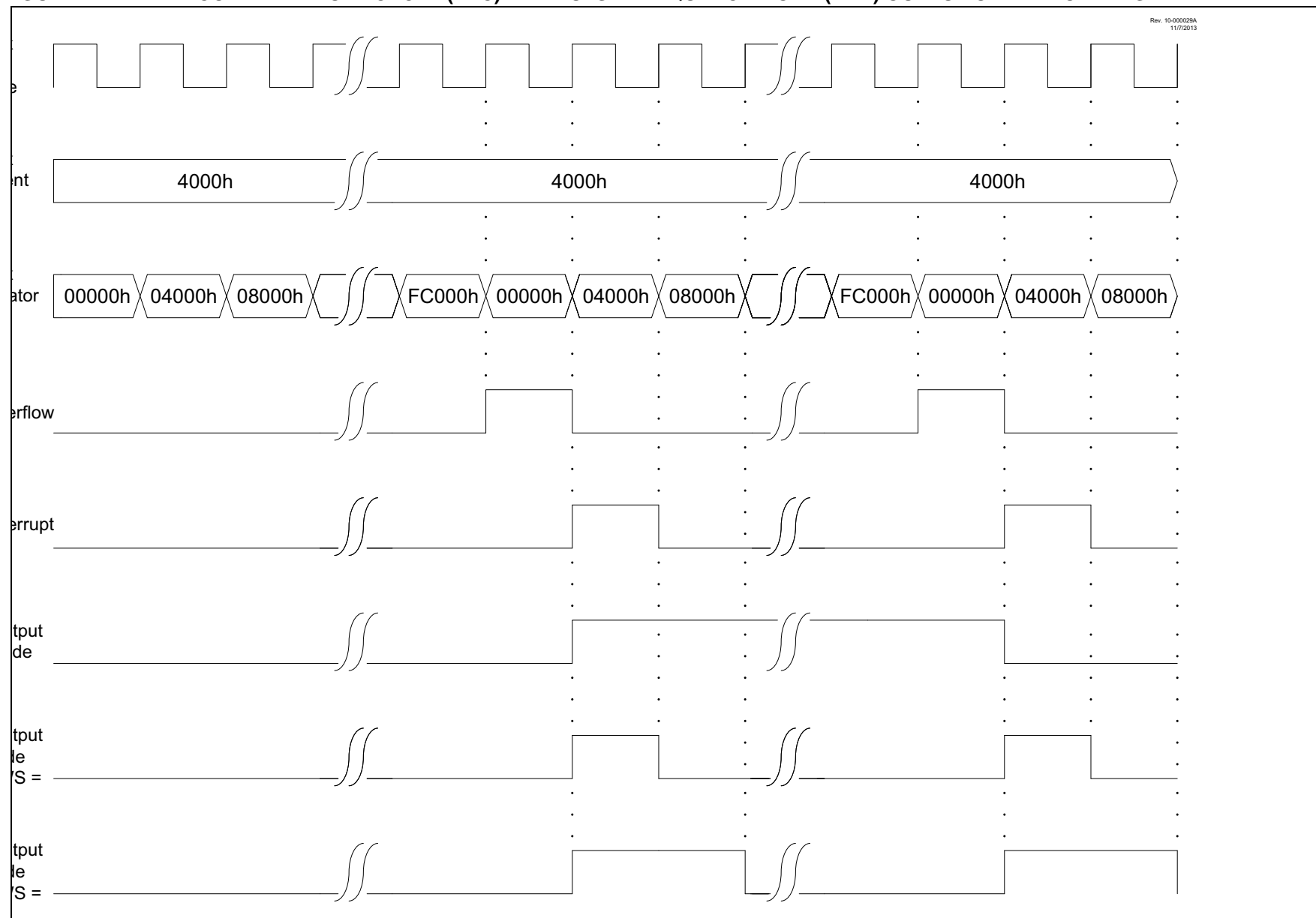
'1' = Bit is set

'0' = Bit is cleared

- bit 7 **LCxG4D4T:** Gate 4 Data 4 True (non-inverted) bit
 1 = lcx4T is gated into lcxg4
 0 = lcx4T is not gated into lcxg4
- bit 6 **LCxG4D4N:** Gate 4 Data 4 Negated (inverted) bit
 1 = lcx4N is gated into lcxg4
 0 = lcx4N is not gated into lcxg4
- bit 5 **LCxG4D3T:** Gate 4 Data 3 True (non-inverted) bit
 1 = lcx3T is gated into lcxg4
 0 = lcx3T is not gated into lcxg4
- bit 4 **LCxG4D3N:** Gate 4 Data 3 Negated (inverted) bit
 1 = lcx3N is gated into lcxg4
 0 = lcx3N is not gated into lcxg4
- bit 3 **LCxG4D2T:** Gate 4 Data 2 True (non-inverted) bit
 1 = lcx2T is gated into lcxg4
 0 = lcx2T is not gated into lcxg4
- bit 2 **LCxG4D2N:** Gate 4 Data 2 Negated (inverted) bit
 1 = lcx2N is gated into lcxg4
 0 = lcx2N is not gated into lcxg4
- bit 1 **LCxG4D1T:** Gate 4 Data 1 True (non-inverted) bit
 1 = lcx1T is gated into lcxg4
 0 = lcx1T is not gated into lcxg4
- bit 0 **LCxG4D1N:** Gate 4 Data 1 Negated (inverted) bit
 1 = lcx1N is gated into lcxg4
 0 = lcx1N is not gated into lcxg4

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11/7/2013

FIGURE 24-2: NCO – FIXED DUTY CYCLE (FDC) AND PULSE FREQUENCY MODE (PFM) OUTPUT OPERATION DIAGRAM



PIC16(L)F1503

25.12 Register Definitions: CWG Control

REGISTER 25-1: CWGxCON0: CWG CONTROL REGISTER 0

R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	U-0	U-0	R/W-0/0
GxEN	GxOEB	GxOEA	GxPOLB	GxPOLA	—	—	GxCS0
bit 7							bit 0

Legend:

R = Readable bit

W = Writable bit

U = Unimplemented bit, read as '0'

u = Bit is unchanged

x = Bit is unknown

-n/n = Value at POR and BOR/Value at all other Resets

'1' = Bit is set

'0' = Bit is cleared

q = Value depends on condition

- bit 7 **GxEN:** CWGx Enable bit
1 = Module is enabled
0 = Module is disabled
- bit 6 **GxOEB:** CWGxB Output Enable bit
1 = CWGxB is available on appropriate I/O pin
0 = CWGxB is not available on appropriate I/O pin
- bit 5 **GxOEA:** CWGxA Output Enable bit
1 = CWGxA is available on appropriate I/O pin
0 = CWGxA is not available on appropriate I/O pin
- bit 4 **GxPOLB:** CWGxB Output Polarity bit
1 = Output is inverted polarity
0 = Output is normal polarity
- bit 3 **GxPOLA:** CWGxA Output Polarity bit
1 = Output is inverted polarity
0 = Output is normal polarity
- bit 2-1 **Unimplemented:** Read as '0'
- bit 0 **GxCS0:** CWGx Clock Source Select bit
1 = HFINTOSC
0 = Fosc

RRF Rotate Right f through Carry

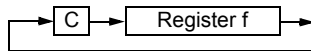
Syntax: [*label*] RRF *f*,*d*

Operands: $0 \leq f \leq 127$
 $d \in [0,1]$

Operation: See description below

Status Affected: C

Description: The contents of register 'f' are rotated one bit to the right through the Carry flag. If 'd' is '0', the result is placed in the W register. If 'd' is '1', the result is placed back in register 'f'.



SUBLW Subtract W from literal

Syntax: [*label*] SUBLW *k*

Operands: $0 \leq k \leq 255$

Operation: $k - (W) \rightarrow (W)$

Status Affected: C, DC, Z

Description: The W register is subtracted (2's complement method) from the 8-bit literal 'k'. The result is placed in the W register.

C = 0	$W > k$
C = 1	$W \leq k$
DC = 0	$W<3:0> > k<3:0>$
DC = 1	$W<3:0> \leq k<3:0>$

SLEEP Enter Sleep mode

Syntax: [*label*] SLEEP

Operands: None

Operation: 00h → WDT,
 0 → WDT prescaler,
 1 → \overline{TO} ,
 0 → \overline{PD}

Status Affected: \overline{TO} , \overline{PD}

Description: The power-down Status bit, \overline{PD} is cleared. Time-out Status bit, \overline{TO} is set. Watchdog Timer and its prescaler are cleared. The processor is put into Sleep mode with the oscillator stopped.

SUBWF Subtract W from f

Syntax: [*label*] SUBWF *f*,*d*

Operands: $0 \leq f \leq 127$
 $d \in [0,1]$

Operation: $(f) - (W) \rightarrow (\text{destination})$

Status Affected: C, DC, Z

Description: Subtract (2's complement method) W register from register 'f'. If 'd' is '0', the result is stored in the W register. If 'd' is '1', the result is stored back in register 'f'.

C = 0	$W > f$
C = 1	$W \leq f$
DC = 0	$W<3:0> > f<3:0>$
DC = 1	$W<3:0> \leq f<3:0>$

SUBWFB Subtract W from f with Borrow

Syntax: SUBWFB *f* {,*d*}

Operands: $0 \leq f \leq 127$
 $d \in [0,1]$

Operation: $(f) - (W) - (\overline{B}) \rightarrow \text{dest}$

Status Affected: C, DC, Z

Description: Subtract W and the BORROW flag (CARRY) from register 'f' (2's complement method). If 'd' is '0', the result is stored in W. If 'd' is '1', the result is stored back in register 'f'.

FIGURE 28-1: VOLTAGE FREQUENCY GRAPH, $-40^{\circ}\text{C} \leq T_A \leq +125^{\circ}\text{C}$, PIC16F1503 ONLY

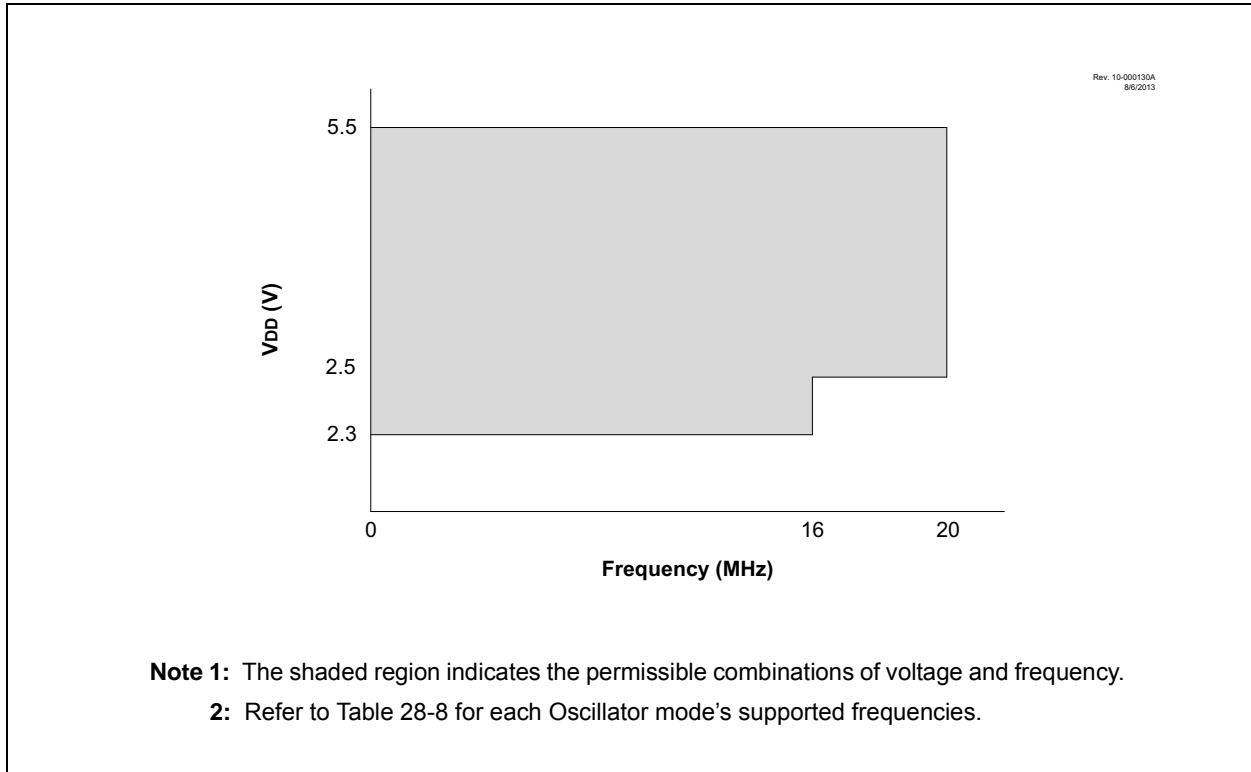
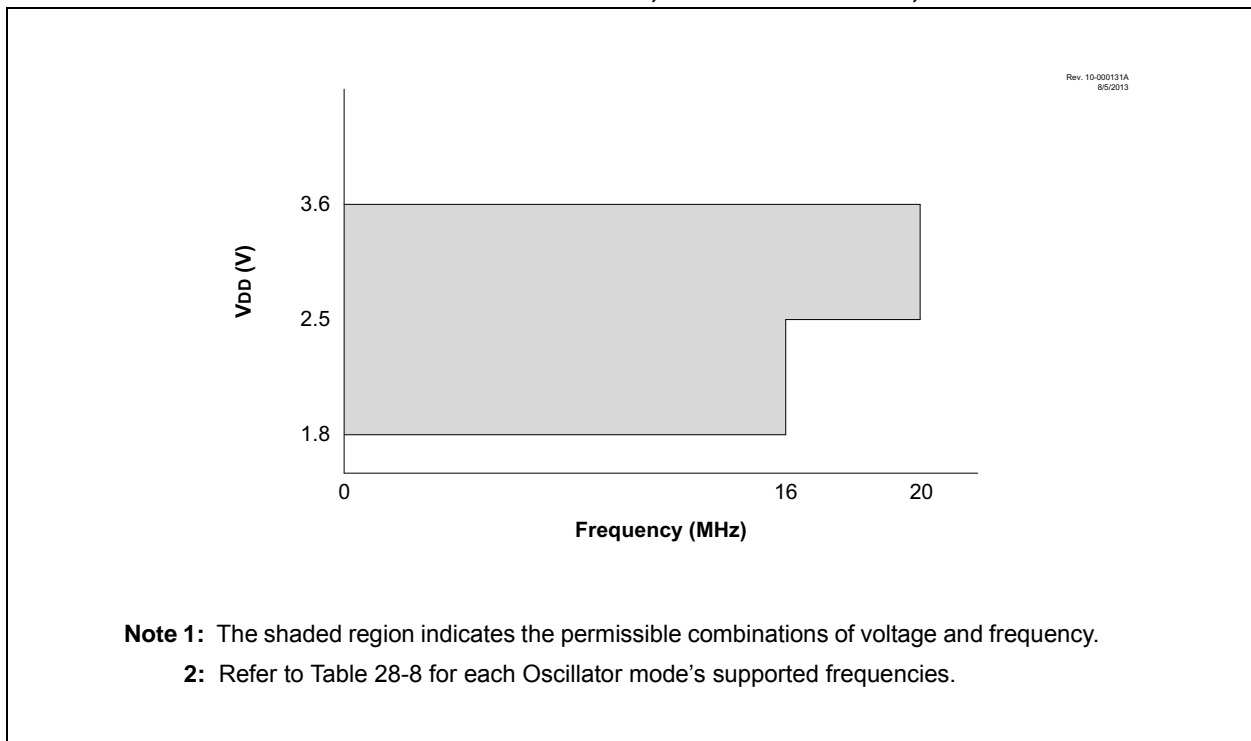


FIGURE 28-2: VOLTAGE FREQUENCY GRAPH, $-40^{\circ}\text{C} \leq T_A \leq +125^{\circ}\text{C}$, PIC16LF1503 ONLY



PIC16(L)F1503

FIGURE 29-29: I_{PD}, BROWN-OUT RESET (BOR), BORV = 0, PIC16F1503 ONLY

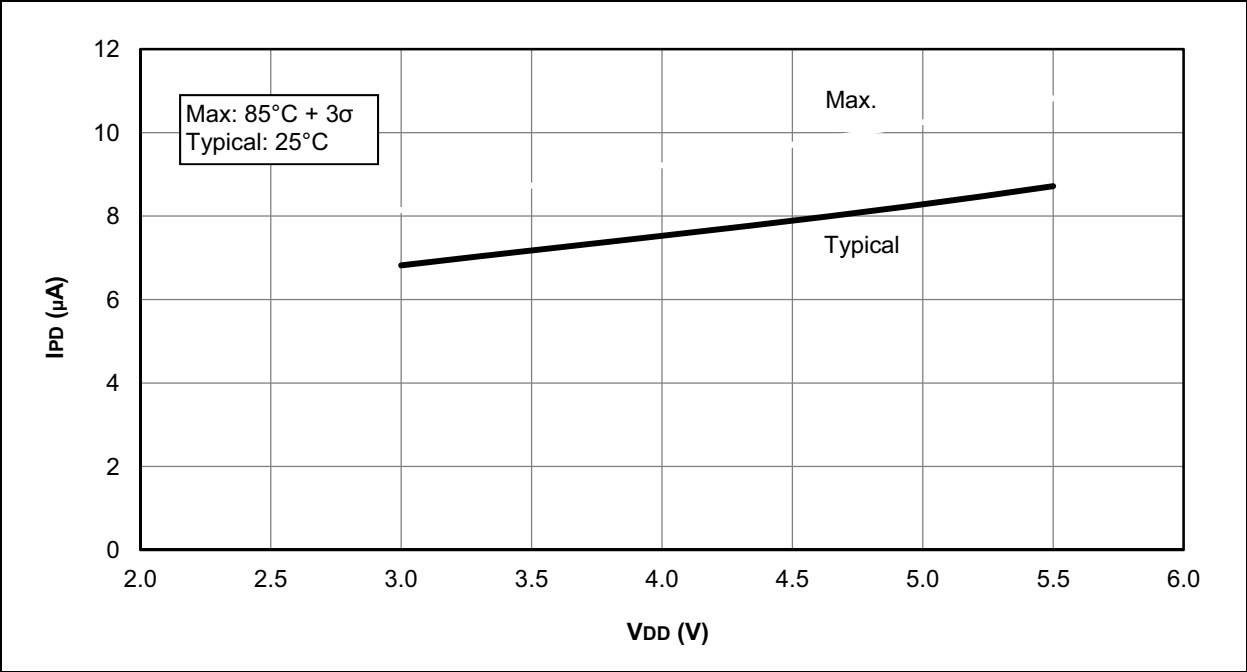
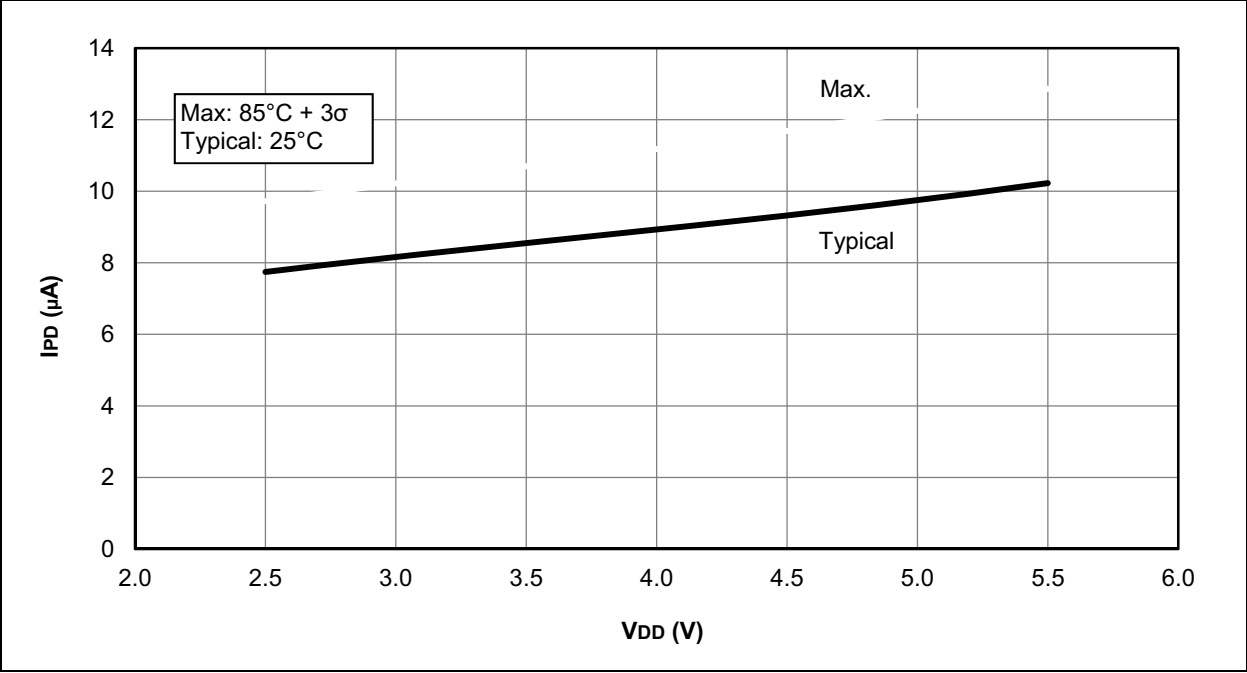


FIGURE 29-30: I_{PD}, BROWN-OUT RESET (BOR), BORV = 1, PIC16F1503 ONLY



PIC16(L)F1503

FIGURE 29-41: POR RELEASE VOLTAGE

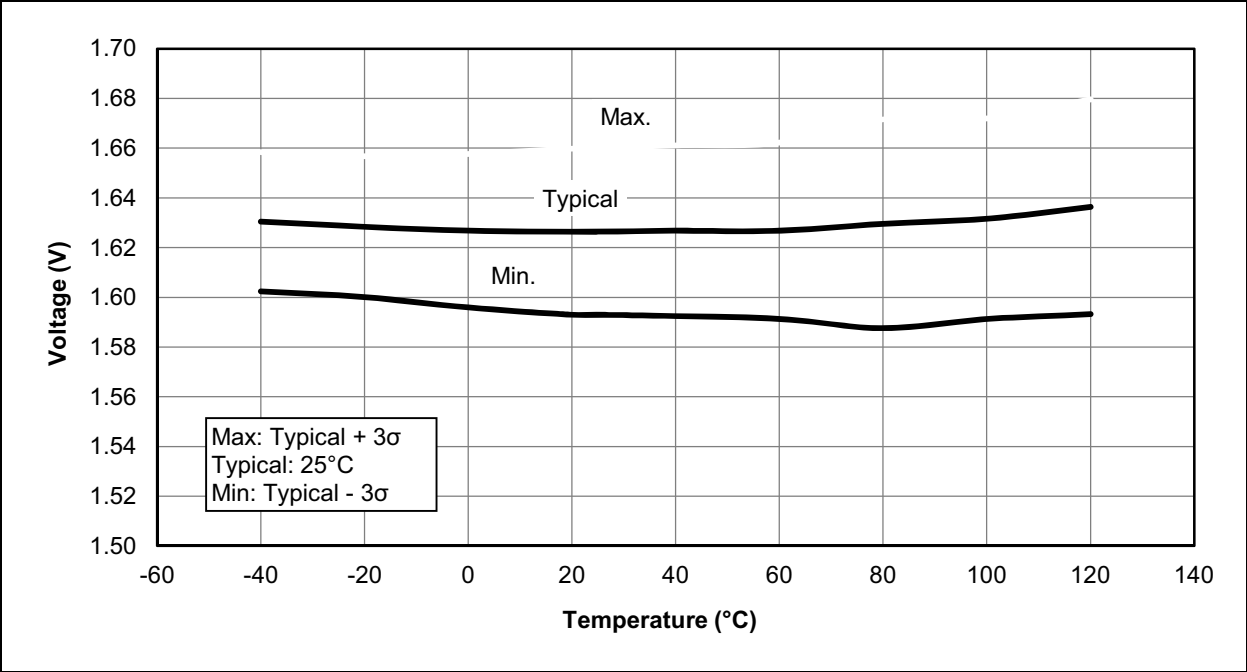
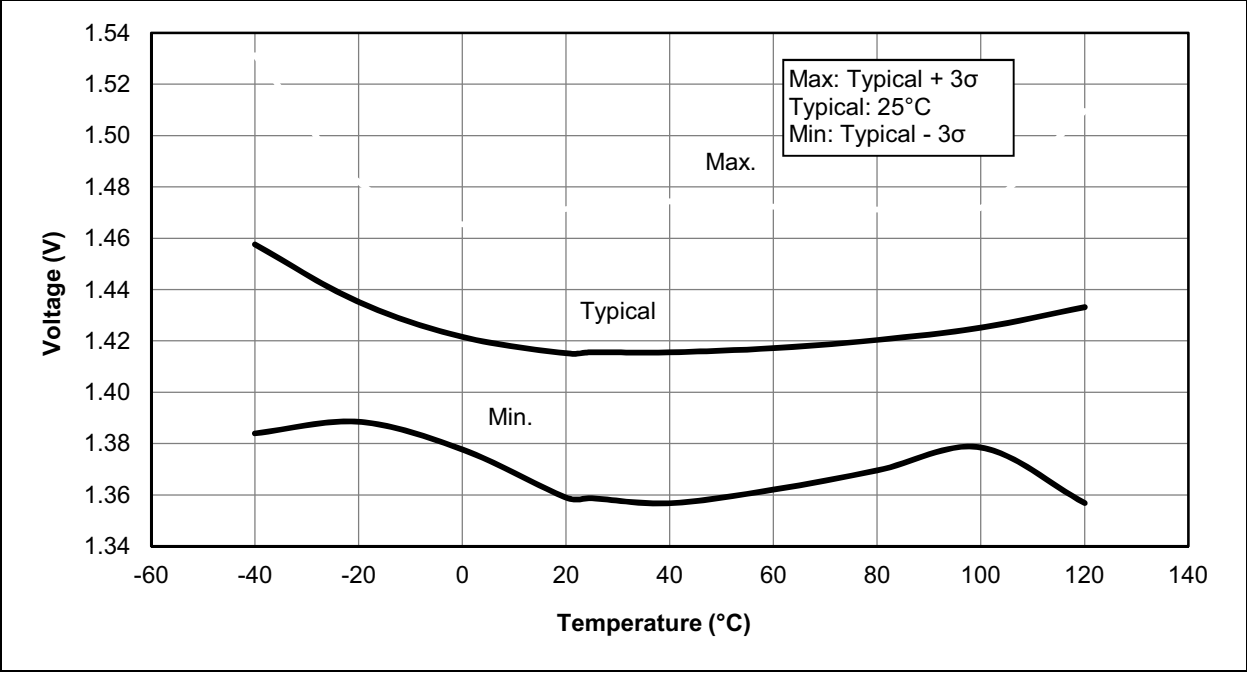


FIGURE 29-42: POR REARM VOLTAGE, PIC16F1503 ONLY



PIC16(L)F1503

FIGURE 29-45: BROWN-OUT RESET VOLTAGE, BORV = 1, PIC16F1503 ONLY

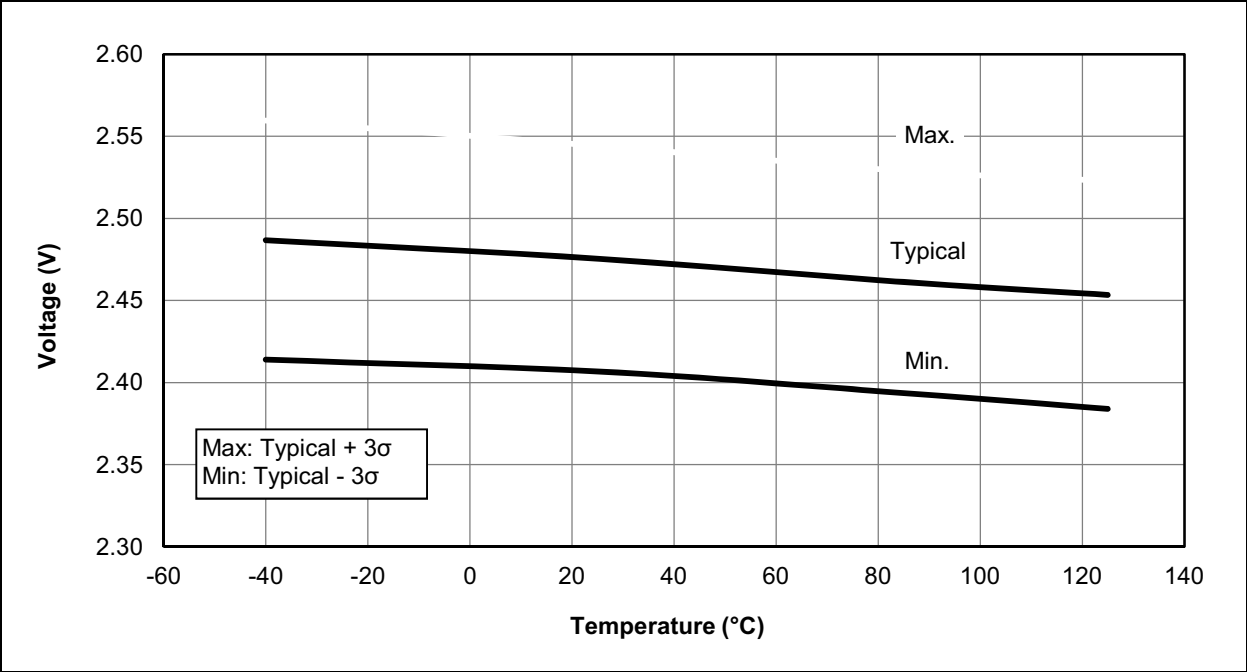


FIGURE 29-46: BROWN-OUT RESET HYSTERESIS, BORV = 1, PIC16F1503 ONLY

