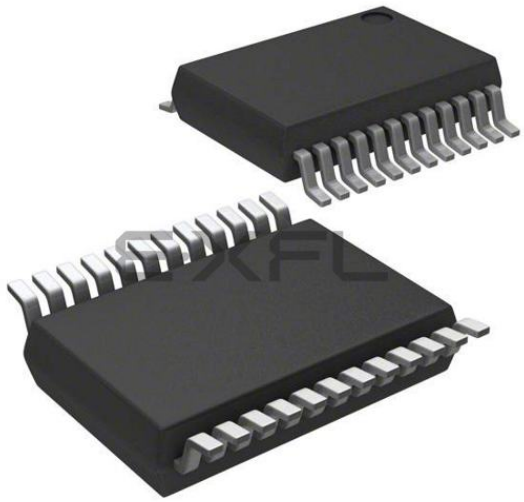

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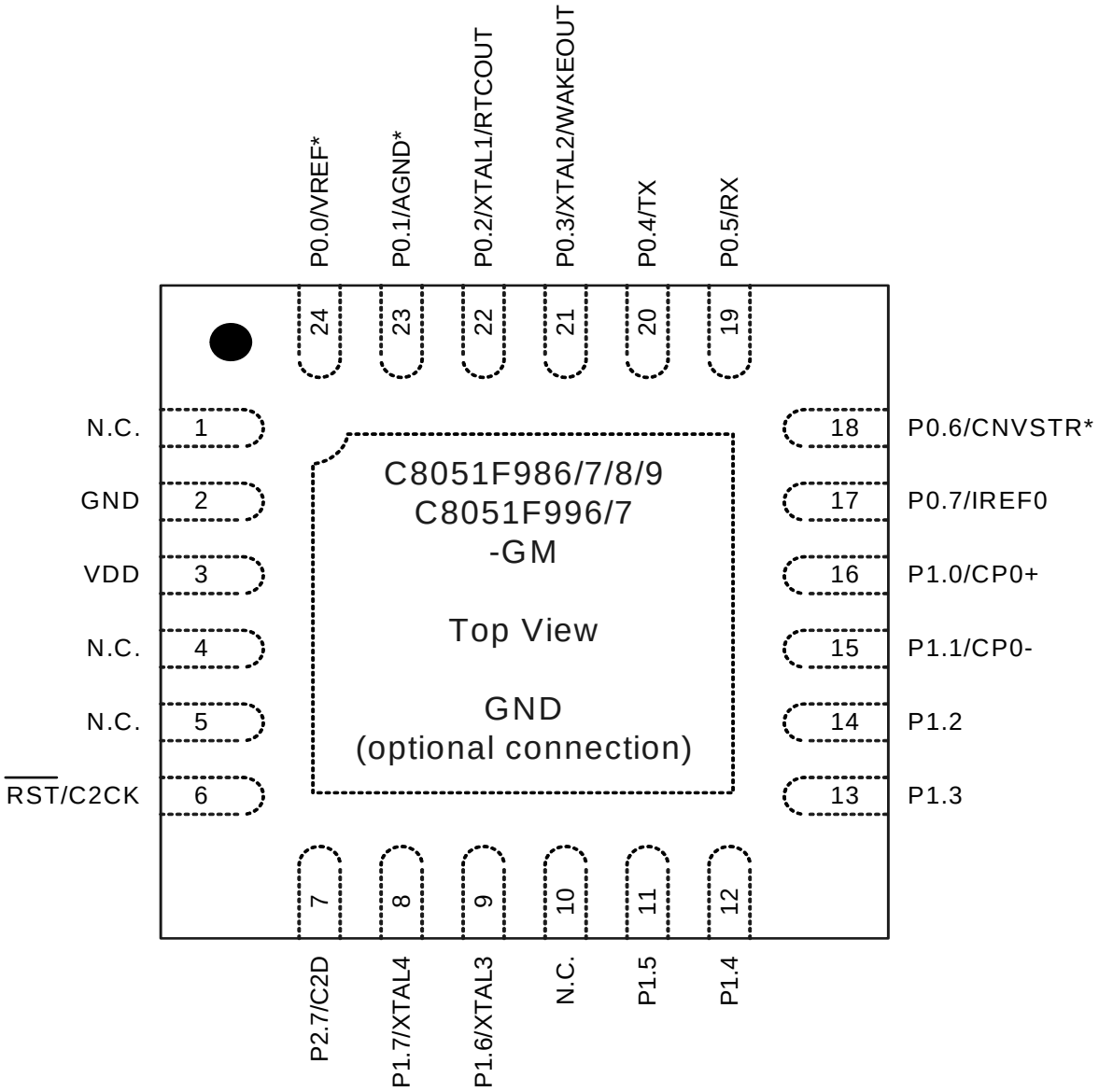
What is "Embedded - Microcontrollers"?

"Embedded - Microcontrollers" refer to small, integrated circuits designed to perform specific tasks within larger systems. These microcontrollers are essentially compact computers on a single chip, containing a processor core, memory, and programmable input/output peripherals. They are called "embedded" because they are embedded within electronic devices to control various functions, rather than serving as standalone computers. Microcontrollers are crucial in modern electronics, providing the intelligence and control needed for a wide range of applications.

Applications of "Embedded - Microcontrollers"

Details	
Product Status	Active
Core Processor	CIP-51 8051
Core Size	8-Bit
Speed	25MHz
Connectivity	SMBus (2-Wire/I²C), SPI, UART/USART
Peripherals	Brown-out Detect/Reset, POR, PWM, WDT
Number of I/O	17
Program Memory Size	8KB (8K x 8)
Program Memory Type	FLASH
EEPROM Size	-
RAM Size	512 x 8
Voltage - Supply (Vcc/Vdd)	1.8V ~ 3.6V
Data Converters	-
Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	24-SSOP (0.154", 3.90mm Width)
Supplier Device Package	24-QSOP
Purchase URL	https://www.e-xfl.com/product-detail/silicon-labs/c8051f987-c-gu



***Note:** Signal only available on ‘F986, ‘F988, and ‘F996 devices.

Figure 3.2. QFN-24 Pinout Diagram (Top View)

Table 3.3. PCB Land Pattern

Dimension	Min	Max
D	2.71 REF	
D2	1.60	1.80
e	0.50 BSC	
E	2.71 REF	
E2	1.60	1.80
f	2.53 REF	
GD	2.10	—
GE	2.10	—
W	—	0.34
X	—	0.28
Y	0.61 REF	
ZE	—	3.31
ZD	—	3.31

Notes:

General

1. All dimensions shown are in millimeters (mm) unless otherwise noted.
2. Dimensioning and Tolerancing is per the ANSI Y14.5M-1994 specification.
3. This Land Pattern Design is based on IPC-SM-782 guidelines.
4. All dimensions shown are at Maximum Material Condition (MMC). Least Material Condition (LMC) is calculated based on a Fabrication Allowance of 0.05 mm.

Solder Mask Design

1. All metal pads are to be non-solder mask defined (NSMD). Clearance between the solder mask and the metal pad is to be 60 µm minimum, all the way around the pad.

Stencil Design

1. A stainless steel, laser-cut and electro-polished stencil with trapezoidal walls should be used to assure good solder paste release.
2. The stencil thickness should be 0.125 mm (5 mils).
3. The ratio of stencil aperture to land pad size should be 1:1 for all perimeter pads.
4. A 1.45 x 1.45 mm square aperture should be used for the center pad. This provides approximately 70% solder paste coverage on the pad, which is optimum to assure correct component stand-off.

Card Assembly

1. A No-Clean, Type-3 solder paste is recommended.
2. The recommended card reflow profile is per the JEDEC/IPC J-STD-020 specification for Small Body Components.

SFR Definition 5.3. ADC0AC: ADC0 Accumulator Configuration

Bit	7	6	5	4	3	2	1	0
Name	AD012BE	AD0AE	AD0SJST[2:0]			AD0RPT[2:0]		
Type	R/W	W	R/W			R/W		
Reset	0	0	0	0	0	0	0	0

SFR Page = 0x0; SFR Address = 0xBA

Bit	Name	Function
7	AD012BE	**ADC0 12-Bit Mode Enable.** Enables 12-bit Mode on C8051F980/6 and C8051F990/6 devices. 0: 12-bit Mode Disabled. 1: 12-bit Mode Enabled.
6	AD0AE	**ADC0 Accumulate Enable.** Enables multiple conversions to be accumulated when burst mode is disabled. 0: ADC0H:ADC0L contain the result of the latest conversion when Burst Mode is disabled. 1: ADC0H:ADC0L contain the accumulated conversion results when Burst Mode is disabled. Software must write 0x0000 to ADC0H:ADC0L to clear the accumulated result. This bit is write-only. Always reads 0b.
5:3	AD0SJST[2:0]	**ADC0 Accumulator Shift and Justify.** Specifies the format of data read from ADC0H:ADC0L. 000: Right justified. No shifting applied. 001: Right justified. Shifted right by 1 bit. 010: Right justified. Shifted right by 2 bits. 011: Right justified. Shifted right by 3 bits. 100: Left justified. No shifting applied. All remaining bit combinations are reserved.
2:0	AD0RPT[2:0]	**ADC0 Repeat Count.** Selects the number of conversions to perform and accumulate in Burst Mode. This bit field must be set to 000 if Burst Mode is disabled. 000: Perform and Accumulate 1 conversion. 001: Perform and Accumulate 4 conversions. 010: Perform and Accumulate 8 conversions. 011: Perform and Accumulate 16 conversions. 100: Perform and Accumulate 32 conversions. 101: Perform and Accumulate 64 conversions. All remaining bit combinations are reserved.

SFR Definition 5.4. ADC0PWR: ADC0 Burst Mode Power-Up Time

Bit	7	6	5	4	3	2	1	0
Name	AD0LPM				AD0PWR[3:0]			
Type	R/W	R	R	R	R/W			
Reset	0	0	0	0	1	1	1	1

SFR Page = All; SFR Address = 0xBB

Bit	Name	Function
7	AD0LPM	**ADC0 Low Power Mode Enable.** Enables Low Power Mode Operation. 0: Low Power Mode disabled. 1: Low Power Mode enabled.
6:4	Unused	Read = 0000b; Write = Don't Care.
3:0	AD0PWR[3:0]	**ADC0 Burst Mode Power-Up Time.** Sets the time delay required for ADC0 to power up from a low power state. For BURSTEN = 0: ADC0 power state controlled by AD0EN. For BURSTEN = 1 and AD0EN = 1: ADC0 remains enabled and does not enter a low power state after all conversions are complete. Conversions can begin immediately following the start-of-conversion signal. For BURSTEN = 1 and AD0EN = 0: ADC0 enters a low power state after all conversions are complete. Conversions can begin a programmed delay after the start-of-conversion signal. The ADC0 Burst Mode Power-Up time is programmed according to the following equation: $AD0PWR = \frac{Tstartup}{400ns} - 1$ or $Tstartup = (AD0PWR + 1)400ns$ **Note:** Setting AD0PWR to 0x04 provides a typical tracking time of 2 us for the first sample taken after the start of conversion.

5.6.1. Window Detector In Single-Ended Mode

Figure 5.5 shows two example window comparisons for right-justified data, with ADC0LTH:ADC0LTL = 0x0080 (128d) and ADC0GTH:ADC0GTL = 0x0040 (64d). The input voltage can range from 0 to VREF x (1023/1024) with respect to GND, and is represented by a 10-bit unsigned integer value. In the left example, an AD0WINT interrupt will be generated if the ADC0 conversion word (ADC0H:ADC0L) is within the range defined by ADC0GTH:ADC0GTL and ADC0LTH:ADC0LTL (if 0x0040 < ADC0H:ADC0L < 0x0080). In the right example, and AD0WINT interrupt will be generated if the ADC0 conversion word is outside of the range defined by the ADC0GT and ADC0LT registers (if ADC0H:ADC0L < 0x0040 or ADC0H:ADC0L > 0x0080). Figure 5.6 shows an example using left-justified data with the same comparison values.

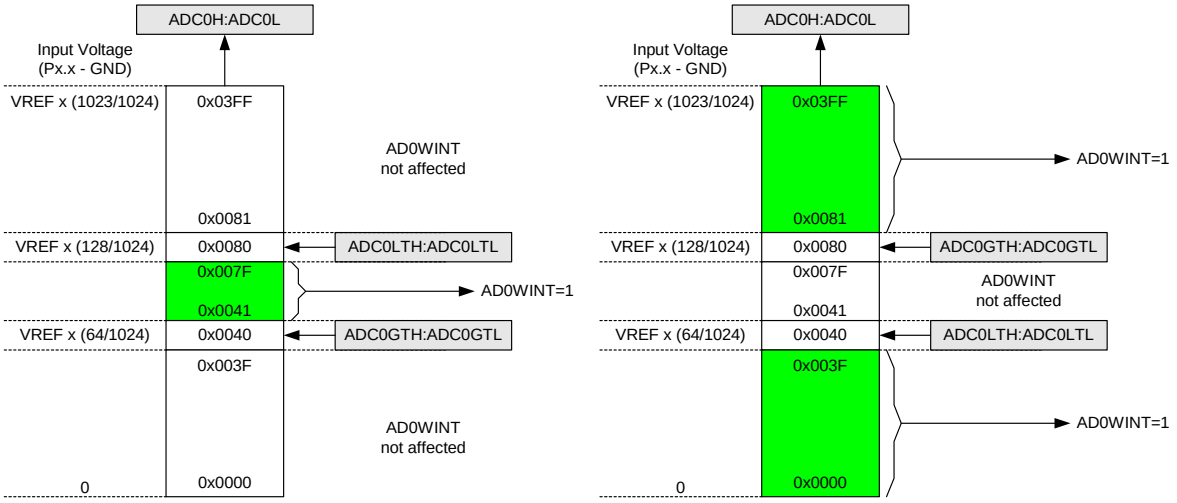


Figure 5.5. ADC Window Compare Example: Right-Justified Single-Ended Data

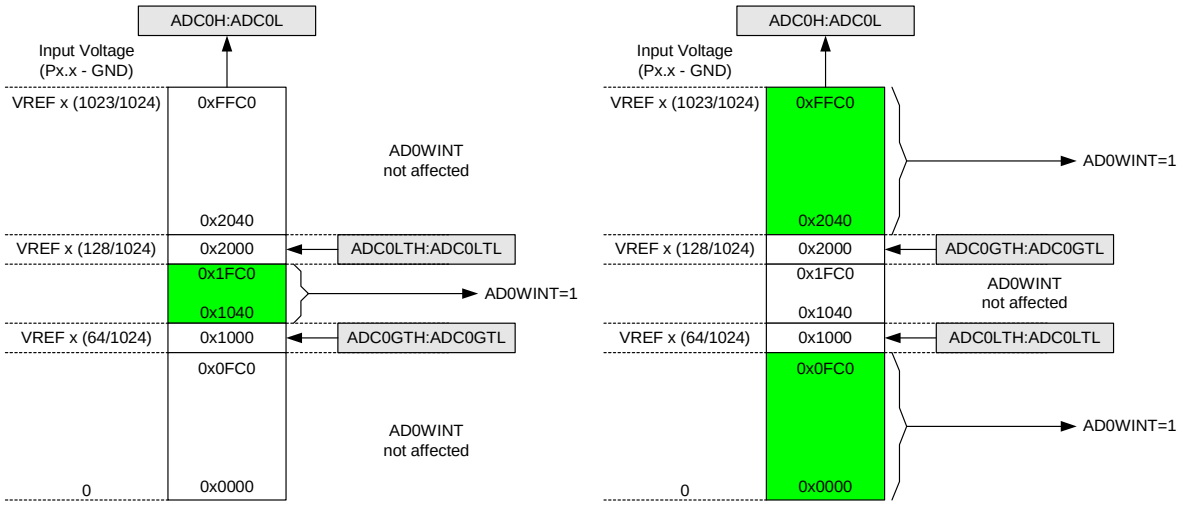


Figure 5.6. ADC Window Compare Example: Left-Justified Single-Ended Data

5.6.2. ADC0 Specifications

See “4. Electrical Characteristics” on page 48 for a detailed listing of ADC0 specifications.

7. Comparator

C8051F99x-C8051F98x devices include an on-chip programmable voltage comparator: Comparator 0 (CPT0) shown in Figure 7.1.

The Comparator offers programmable response time and hysteresis, an analog input multiplexer, and two outputs that are optionally available at the Port pins: a digital synchronous “latched” output (CP0), or a digital asynchronous “raw” output (CP0A). The asynchronous CP0A signal is available even when the system clock is not active. This allows the Comparator to operate and generate an output when the device is in some low power modes.

7.1. Comparator Inputs

Each Comparator performs an analog comparison of the voltage levels at its positive (CP0+) and negative (CP0-) input. The analog input multiplexers are completely under software control and configured using SFR registers. See Section “7.6. Comparator0 Analog Multiplexer” on page 98 for details on how to select and configure Comparator inputs.

Important Note About Comparator Inputs: The Port pins selected as Comparator inputs should be configured as analog inputs and skipped by the Crossbar. See the Port I/O chapter for more details on how to configure Port I/O pins as Analog Inputs. The Comparator may also be used to compare the logic level of digital signals, however, Port I/O pins configured as digital inputs must be driven to a valid logic state (HIGH or LOW) to avoid increased power consumption.

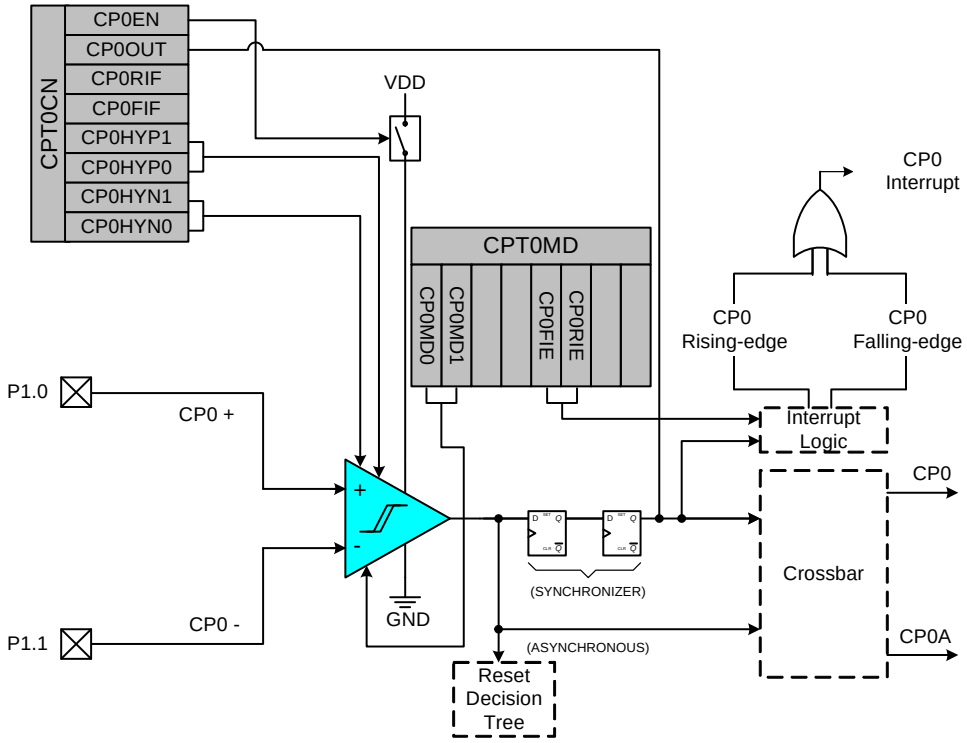


Figure 7.1. Comparator 0 Functional Block Diagram

7.6. Comparator0 Analog Multiplexer

Comparator0 on C8051F99x-C8051F98x devices has an analog input multiplexer to connect Port I/O pins and internal signals the comparator inputs; CP0+/CP0- are the positive and negative input multiplexers for Comparator0.

The comparator input multiplexers directly support capacitive touch switches. When the Capacitive Touch Sense Compare input is selected on the positive or negative multiplexer, any Port I/O pin connected to the other multiplexer can be directly connected to a capacitive touch switch with no additional external components. The Capacitive Touch Sense Compare provides the appropriate reference level for detecting when the capacitive touch switches have charged or discharged through the on-chip Rsense resistor. The Comparator outputs can be routed to Timer2 or Timer3 for capturing sense capacitor's charge and discharge time. See Section "25. Timers" on page 278 for details.

Any of the following may be selected as comparator inputs: Port I/O pins, Capacitive Touch Sense Compare, VDD Supply Voltage, Regulated Digital Supply Voltage (Output of VREG0) or ground. The Comparator's supply voltage divided by 2 is also available as an input; the resistors used to divide the voltage only draw current when this setting is selected. The Comparator input multiplexers are configured using the CPT0MX register described in SFR Definition 7.3.

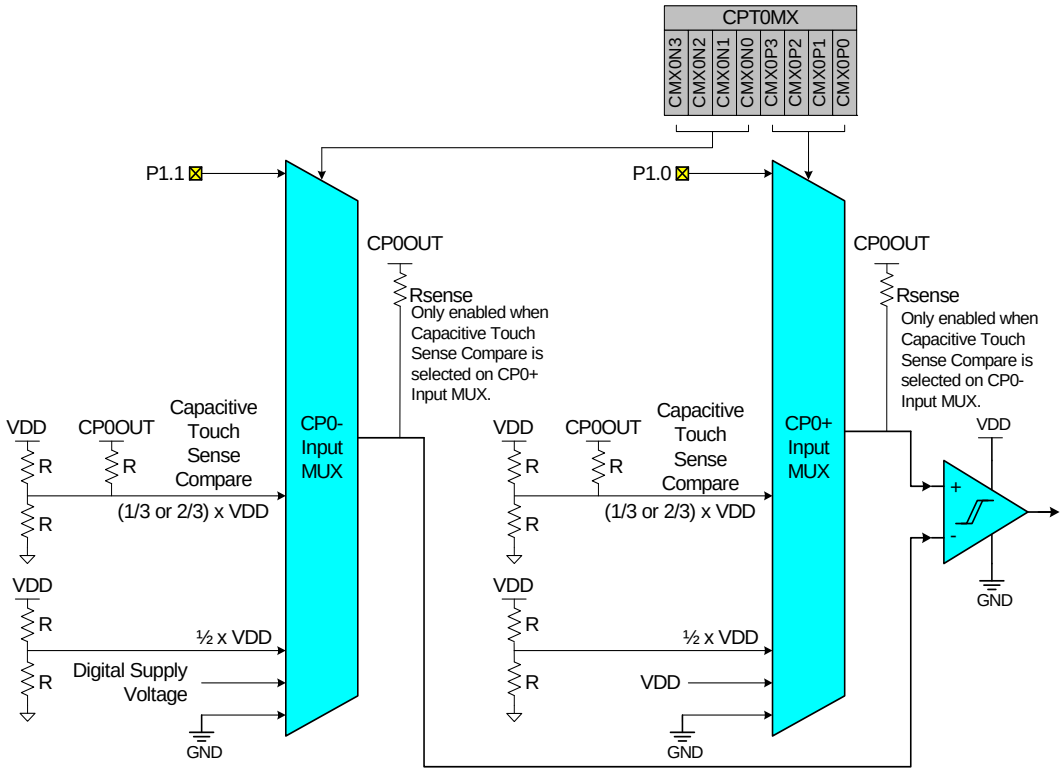


Figure 7.3. CP0 Multiplexer Block Diagram

Important Note About Comparator Input Configuration: Port pins selected as comparator inputs should be configured as analog inputs, and should be skipped by the Digital Crossbar. To configure a Port pin for analog input, set to 0 the corresponding bit in register PnMDIN and disable the digital driver (PnMDOUT = 0 and Port Latch = 1). To force the Crossbar to skip a Port pin, set to 1 the corresponding bit in register PnSKIP. See Section "21. Port Input/Output" on page 215 for more Port I/O configuration details.

SFR Definition 8.9. CS0THH: Capacitive Sense Comparator Threshold High Byte

Bit	7	6	5	4	3	2	1	0
Name	CS0THH[7:0]							
Type	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
Reset	0	0	0	0	0	0	0	0

SFR Page = 0x0; SFR Address = 0xFE

Bit	Name	Description
7:0	CS0THH[7:0]	**CS0 Comparator Threshold High Byte.** High byte of the 16-bit value compared to the Capacitive Sense conversion result.

SFR Definition 8.10. CS0THL: Capacitive Sense Comparator Threshold Low Byte

Bit	7	6	5	4	3	2	1	0
Name	CS0THL[7:0]							
Type	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
Reset	0	0	0	0	0	0	0	0

SFR Page = 0x0; SFR Address = 0xFD

Bit	Name	Description
7:0	CS0THL[7:0]	**CS0 Comparator Threshold Low Byte.** Low byte of the 16-bit value compared to the Capacitive Sense conversion result.

SFR Definition 8.15. CS0MX: Capacitive Sense Mux Channel Select

Bit	7	6	5	4	3	2	1	0
Name	Reserved	Reserved	Reserved	Reserved	CS0MX[3:0]			
Type	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
Reset	0	0	0	0	0	0	0	0

SFR Page = 0x0; SFR Address = 0xAB

Bit	Name	Description	
7:5	Reserved	Read = 0000b; Write = 0000b.	
4:0	CS0MX[4:0]	**CS0 Mux Channel Select.** Selects one of the 14 input channels for Capacitive Sense conversion.	
		Value	**Channel**
		0000	P0.0
		0001	P0.1
		0010	P0.2
		0011	P0.3
		0100	P0.4
		0101	P0.5
		0110	P0.6
		0111	P0.7
		1000	P1.0
		1001	P1.1
		1010	P1.2
		1011	P1.3
		1100	P1.4 (24-pin packages only)
		1101	P1.5
		1110	Reserved
		1111	Reserved

14. Flash Memory

On-chip, re-programmable Flash memory is included for program code and non-volatile data storage. The Flash memory can be programmed in-system through the C2 interface or by software using the MOVX write instruction. Once cleared to logic 0, a Flash bit must be erased to set it back to logic 1. Flash bytes would typically be erased (set to 0xFF) before being reprogrammed. The write and erase operations are automatically timed by hardware for proper execution; data polling to determine the end of the write/erase operations is not required. Code execution is stalled during Flash write/erase operations. Refer to Table 4.6 for complete Flash memory electrical characteristics.

14.1. Programming the Flash Memory

The simplest means of programming the Flash memory is through the C2 interface using programming tools provided by Silicon Laboratories or a third party vendor. This is the only means for programming a non-initialized device. For details on the C2 commands to program Flash memory, see Section “27. C2 Interface” on page 319.

The Flash memory can be programmed by software using the MOVX write instruction with the address and data byte to be programmed provided as normal operands. Before programming Flash memory using MOVX, Flash programming operations must be enabled by: (1) setting the PSWE Program Store Write Enable bit (PSCTL.0) to logic 1 (this directs the MOVX writes to target Flash memory); and (2) Writing the Flash key codes in sequence to the Flash Lock register (FLKEY). The PSWE bit remains set until cleared by software. For detailed guidelines on programming Flash from firmware, please see Section “14.5. Flash Write and Erase Guidelines” on page 156.

To ensure the integrity of the Flash contents, the on-chip VDD Monitor must be enabled and enabled as a reset source in any system that includes code that writes and/or erases Flash memory from software. Furthermore, there should be no delay between enabling the V_{DD} Monitor and enabling the V_{DD} Monitor as a reset source. Any attempt to write or erase Flash memory while the V_{DD} Monitor is disabled, or not enabled as a reset source, will cause a Flash Error device reset.

14.1.1. Flash Lock and Key Functions

Flash writes and erases by user software are protected with a lock and key function. The Flash Lock and Key Register (FLKEY) must be written with the correct key codes, in sequence, before Flash operations may be performed. The key codes are: 0xA5, 0xF1. The timing does not matter, but the codes must be written in order. If the key codes are written out of order, or the wrong codes are written, Flash writes and erases will be disabled until the next system reset. Flash writes and erases will also be disabled if a Flash write or erase is attempted before the key codes have been written properly. The Flash lock resets after each write or erase; the key codes must be written again before a following Flash operation can be performed. The FLKEY register is detailed in SFR Definition 14.4.

14.5.2. PSWE Maintenance

1. Reduce the number of places in code where the PSWE bit (b0 in PSCTL) is set to a 1. There should be exactly one routine in code that sets PSWE to a 1 to write Flash bytes and one routine in code that sets both PSWE and PSEE both to a 1 to erase Flash pages.
2. Minimize the number of variable accesses while PSWE is set to a 1. Handle pointer address updates and loop maintenance outside the "PSWE = 1;... PSWE = 0;" area. Code examples showing this can be found in :AN201: "Writing to Flash from Firmware", available from the Silicon Laboratories website.
3. Disable interrupts prior to setting PSWE to a 1 and leave them disabled until after PSWE has been reset to 0. Any interrupts posted during the Flash write or erase operation will be serviced in priority order after the Flash operation has been completed and interrupts have been re-enabled by software.
4. Make certain that the Flash write and erase pointer variables are not located in XRAM. See your compiler documentation for instructions regarding how to explicitly locate variables in different memory areas.
5. Add address bounds checking to the routines that write or erase Flash memory to ensure that a routine called with an illegal address does not result in modification of the Flash.

14.5.3. System Clock

1. If operating from an external crystal, be advised that crystal performance is susceptible to electrical interference and is sensitive to layout and to changes in temperature. If the system is operating in an electrically noisy environment, use the internal oscillator or use an external CMOS clock.
2. If operating from the external oscillator, switch to the internal oscillator during Flash write or erase operations. The external oscillator can continue to run, and the CPU can switch back to the external oscillator after the Flash operation has completed.

Additional Flash recommendations and example code can be found in “AN201: Writing to Flash from Firmware", available from the Silicon Laboratories web site.

20.1.2. Using RTC0ADR and RTC0DAT to Access SmaRTClock Internal Registers

The SmaRTClock internal registers can be read and written using RTC0ADR and RTC0DAT. The RTC0ADR register selects the SmaRTClock internal register that will be targeted by subsequent reads or writes. Recommended instruction timing is provided in this section. If the recommended instruction timing is not followed, then BUSY (RTC0ADR.7) should be checked prior to each read or write operation to make sure the SmaRTClock Interface is not busy performing the previous read or write operation. A SmaRTClock Write operation is initiated by writing to the RTC0DAT register. Below is an example of writing to a SmaRTClock internal register.

1. Poll BUSY (RTC0ADR.7) until it returns 0 or follow recommended instruction timing.
2. Write 0x05 to RTC0ADR. This selects the internal RTC0CN register at SmaRTClock Address 0x05.
3. Write 0x00 to RTC0DAT. This operation writes 0x00 to the internal RTC0CN register.

A SmaRTClock Read operation is initiated by setting the SmaRTClock Interface Busy bit. This transfers the contents of the internal register selected by RTC0ADR to RTC0DAT. The transferred data will remain in RTC0DAT until the next read or write operation. Below is an example of reading a SmaRTClock internal register.

1. Poll BUSY (RTC0ADR.7) until it returns 0 or follow recommended instruction timing.
2. Write 0x05 to RTC0ADR. This selects the internal RTC0CN register at SmaRTClock Address 0x05.
3. Write 1 to BUSY. This initiates the transfer of data from RTC0CN to RTC0DAT.
4. Poll BUSY (RTC0ADR.7) until it returns 0 or follow recommend instruction timing.
5. Read data from RTC0DAT. This data is a copy of the RTC0CN register.

Note: The RTC0ADR and RTC0DAT registers will retain their state upon a device reset.

20.1.3. RTC0ADR Short Strobe Feature

Reads and writes to indirect SmaRTClock registers normally take 7 system clock cycles. To minimize the indirect register access time, the Short Strobe feature decreases the read and write access time to 6 system clocks. The Short Strobe feature is automatically enabled on reset and can be manually enabled/disabled using the SHORT (RTC0ADR.4) control bit.

Recommended Instruction Timing for a single register read with short strobe enabled:

```
mov RTC0ADR, #095h
nop
nop
nop
mov A, RTC0DAT
```

Recommended Instruction Timing for a single register write with short strobe enabled:

```
mov RTC0ADR, #095h
mov RTC0DAT, #000h
nop
```

20.1.4. SmaRTClock Interface Autoread Feature

When Autoread is enabled, each read from RTC0DAT initiates the next indirect read operation on the SmaRTClock internal register selected by RTC0ADR. Software should set the BUSY bit once at the beginning of each series of consecutive reads. Software should follow recommended instruction timing or check if the SmaRTClock Interface is busy prior to reading RTC0DAT. Autoread is enabled by setting AUTORD (RTC0ADR.6) to logic 1.

20.3.3. Software Considerations for using the SmaRTClock Timer and Alarm

The SmaRTClock timer and alarm have two operating modes to suit varying applications. The two modes are described below:

Mode 1:
The first mode uses the SmaRTClock timer as a perpetual timebase which is never reset to zero. Every 36 hours, the timer is allowed to overflow without being stopped or disrupted. The alarm interval is software managed and is added to the ALRMn registers by software after each alarm. This allows the alarm match value to always stay ahead of the timer by one software managed interval. If software uses 32-bit unsigned addition to increment the alarm match value, then it does not need to handle overflows since both the timer and the alarm match value will overflow in the same manner.

This mode is ideal for applications which have a long alarm interval (e.g., 24 or 36 hours) and/or have a need for a perpetual timebase. An example of an application that needs a perpetual timebase is one whose wake-up interval is constantly changing. For these applications, software can keep track of the number of timer overflows in a 16-bit variable, extending the 32-bit (36 hour) timer to a 48-bit (272 year) perpetual timebase.

Mode 2:
The second mode uses the SmaRTClock timer as a general purpose up counter which is auto reset to zero by hardware after each alarm. The alarm interval is managed by hardware and stored in the ALRMn registers. Software only needs to set the alarm interval once during device initialization. After each alarm, software should keep a count of the number of alarms that have occurred in order to keep track of time.

This mode is ideal for applications that require minimal software intervention and/or have a fixed alarm interval. This mode is the most power efficient since it requires less CPU time per alarm.

22.4. Using the SMBus

The SMBus can operate in both Master and Slave modes. The interface provides timing and shifting control for serial transfers; higher level protocol is determined by user software. The SMBus interface provides the following application-independent features:

- Byte-wise serial data transfers
- Clock signal generation on SCL (Master Mode only) and SDA data synchronization
- Timeout/bus error recognition, as defined by the SMB0CF configuration register
- START/STOP timing, detection, and generation
- Bus arbitration
- Interrupt generation
- Status information
- Optional hardware recognition of slave address and automatic acknowledgement of address/data

SMBus interrupts are generated for each data byte or slave address that is transferred. When hardware acknowledgement is disabled, the point at which the interrupt is generated depends on whether the hardware is acting as a data transmitter or receiver. When a transmitter (i.e. sending address/data, receiving an ACK), this interrupt is generated after the ACK cycle so that software may read the received ACK value; when receiving data (i.e., receiving address/data, sending an ACK), this interrupt is generated before the ACK cycle so that software may define the outgoing ACK value. If hardware acknowledgement is enabled, these interrupts are always generated after the ACK cycle. See Section 22.5 for more details on transmission sequences.

Interrupts are also generated to indicate the beginning of a transfer when a master (START generated), or the end of a transfer when a slave (STOP detected). Software should read the SMB0CN (SMBus Control register) to find the cause of the SMBus interrupt. The SMB0CN register is described in Section 22.4.2; Table 22.5 provides a quick SMB0CN decoding reference.

Table 22.5. SMBus Status Decoding With Hardware ACK Generation Disabled (EHACK = 0)

Mode	Values Read				Current SMbus State	Typical Response Options	Values to Write			Next Status Vector Expected
	Status Vector	ACKRQ	ARBLOST	ACK			STA	STO	ACK	
Master Transmitter	1110	0	0	X	A master START was generated.	Load slave address + R/W into SMB0DAT.	0	0	X	1100
	1100	0	0	0	A master data or address byte was transmitted; NACK received.	Set STA to restart transfer.	1	0	X	1110
						Abort transfer.	0	1	X	—
		0	0	1	A master data or address byte was transmitted; ACK received.	Load next data byte into SMB0DAT.	0	0	X	1100
						End transfer with STOP.	0	1	X	—
						End transfer with STOP and start another transfer.	1	1	X	—
						Send repeated START.	1	0	X	1110
						Switch to Master Receiver Mode (clear SI without writing new data to SMB0DAT).	0	0	X	1000
Master Receiver	1000	1	0	X	A master data byte was received; ACK requested.	Acknowledge received byte; Read SMB0DAT.	0	0	1	1000
						Send NACK to indicate last byte, and send STOP.	0	1	0	—
						Send NACK to indicate last byte, and send STOP followed by START.	1	1	0	1110
						Send ACK followed by repeated START.	1	0	1	1110
						Send NACK to indicate last byte, and send repeated START.	1	0	0	1110
						Send ACK and switch to Master Transmitter Mode (write to SMB0DAT before clearing SI).	0	0	1	1100
						Send NACK and switch to Master Transmitter Mode (write to SMB0DAT before clearing SI).	0	0	0	1100

23.2.2. 9-Bit UART

9-bit UART mode uses a total of eleven bits per data byte: a start bit, 8 data bits (LSB first), a programmable ninth data bit, and a stop bit. The state of the ninth transmit data bit is determined by the value in TB80 (SCON0.3), which is assigned by user software. It can be assigned the value of the parity flag (bit P in register PSW) for error detection, or used in multiprocessor communications. On receive, the ninth data bit goes into RB80 (SCON0.2) and the stop bit is ignored.

Data transmission begins when an instruction writes a data byte to the SBUF0 register. The TI0 Transmit Interrupt Flag (SCON0.1) is set at the end of the transmission (the beginning of the stop-bit time). Data reception can begin any time after the REN0 Receive Enable bit (SCON0.4) is set to 1. After the stop bit is received, the data byte will be loaded into the SBUF0 receive register if the following conditions are met: (1) RI0 must be logic 0, and (2) if MCE0 is logic 1, the 9th bit must be logic 1 (when MCE0 is logic 0, the state of the ninth data bit is unimportant). If these conditions are met, the eight bits of data are stored in SBUF0, the ninth bit is stored in RB80, and the RI0 flag is set to 1. If the above conditions are not met, SBUF0 and RB80 will not be loaded and the RI0 flag will not be set to 1. A UART0 interrupt will occur if enabled when either TI0 or RI0 is set to 1.

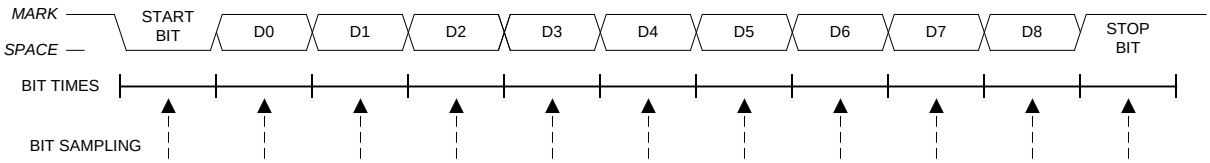


Figure 23.5. 9-Bit UART Timing Diagram

23.3. Multiprocessor Communications

9-Bit UART mode supports multiprocessor communication between a master processor and one or more slave processors by special use of the ninth data bit. When a master processor wants to transmit to one or more slaves, it first sends an address byte to select the target(s). An address byte differs from a data byte in that its ninth bit is logic 1; in a data byte, the ninth bit is always set to logic 0.

Setting the MCE0 bit (SCON0.5) of a slave processor configures its UART such that when a stop bit is received, the UART will generate an interrupt only if the ninth bit is logic 1 (RB80 = 1) signifying an address byte has been received. In the UART interrupt handler, software will compare the received address with the slave's own assigned 8-bit address. If the addresses match, the slave will clear its MCE0 bit to enable interrupts on the reception of the following data byte(s). Slaves that weren't addressed leave their MCE0 bits set and do not generate interrupts on the reception of the following data bytes, thereby ignoring the data. Once the entire message is received, the addressed slave resets its MCE0 bit to ignore all transmissions until it receives the next address byte.

Multiple addresses can be assigned to a single slave and/or a single address can be assigned to multiple slaves, thereby enabling "broadcast" transmissions to more than one slave simultaneously. The master processor can be configured to receive all transmissions or a protocol can be implemented such that the master/slave role is temporarily reversed to enable half-duplex transmission between the original master and slave(s).

24.1. Signal Descriptions

The four signals used by SPI0 (MOSI, MISO, SCK, NSS) are described below.

24.1.1. Master Out, Slave In (MOSI)

The master-out, slave-in (MOSI) signal is an output from a master device and an input to slave devices. It is used to serially transfer data from the master to the slave. This signal is an output when SPI0 is operating as a master and an input when SPI0 is operating as a slave. Data is transferred most-significant bit first. When configured as a master, MOSI is driven by the MSB of the shift register in both 3- and 4-wire mode.

24.1.2. Master In, Slave Out (MISO)

The master-in, slave-out (MISO) signal is an output from a slave device and an input to the master device. It is used to serially transfer data from the slave to the master. This signal is an input when SPI0 is operating as a master and an output when SPI0 is operating as a slave. Data is transferred most-significant bit first. The MISO pin is placed in a high-impedance state when the SPI module is disabled and when the SPI operates in 4-wire mode as a slave that is not selected. When acting as a slave in 3-wire mode, MISO is always driven by the MSB of the shift register.

24.1.3. Serial Clock (SCK)

The serial clock (SCK) signal is an output from the master device and an input to slave devices. It is used to synchronize the transfer of data between the master and slave on the MOSI and MISO lines. SPI0 generates this signal when operating as a master. The SCK signal is ignored by a SPI slave when the slave is not selected (NSS = 1) in 4-wire slave mode.

24.1.4. Slave Select (NSS)

The function of the slave-select (NSS) signal is dependent on the setting of the NSSMD1 and NSSMD0 bits in the SPI0CN register. There are three possible modes that can be selected with these bits:

1. NSSMD[1:0] = 00: 3-Wire Master or 3-Wire Slave Mode: SPI0 operates in 3-wire mode, and NSS is disabled. When operating as a slave device, SPI0 is always selected in 3-wire mode. Since no select signal is present, SPI0 must be the only slave on the bus in 3-wire mode. This is intended for point-to-point communication between a master and one slave.
2. NSSMD[1:0] = 01: 4-Wire Slave or Multi-Master Mode: SPI0 operates in 4-wire mode, and NSS is enabled as an input. When operating as a slave, NSS selects the SPI0 device. When operating as a master, a 1-to-0 transition of the NSS signal disables the master function of SPI0 so that multiple master devices can be used on the same SPI bus.
3. NSSMD[1:0] = 1x: 4-Wire Master Mode: SPI0 operates in 4-wire mode, and NSS is enabled as an output. The setting of NSSMD0 determines what logic level the NSS pin will output. This configuration should only be used when operating SPI0 as a master device.

See Figure 24.2, Figure 24.3, and Figure 24.4 for typical connection diagrams of the various operational modes. **Note that the setting of NSSMD bits affects the pinout of the device.** When in 3-wire master or 3-wire slave mode, the NSS pin will not be mapped by the crossbar. In all other modes, the NSS signal will be mapped to a pin on the device. See Section “21. Port Input/Output” on page 215 for general purpose port I/O and crossbar information.

24.2. SPI0 Master Mode Operation

A SPI master device initiates all data transfers on a SPI bus. SPI0 is placed in master mode by setting the Master Enable flag (MSTEN, SPI0CN.6). Writing a byte of data to the SPI0 data register (SPI0DAT) when in master mode writes to the transmit buffer. If the SPI shift register is empty, the byte in the transmit buffer is moved to the shift register, and a data transfer begins. The SPI0 master immediately shifts out the data serially on the MOSI line while providing the serial clock on SCK. The SPIF (SPI0CN.7) flag is set to logic

SFR Definition 25.4. TL0: Timer 0 Low Byte

Bit	7	6	5	4	3	2	1	0
Name	TL0[7:0]							
Type	R/W							
Reset	0	0	0	0	0	0	0	0

SFR Page = 0x0; SFR Address = 0x8A

Bit	Name	Function
7:0	TL0[7:0]	**Timer 0 Low Byte.** The TL0 register is the low byte of the 16-bit Timer 0.

SFR Definition 25.5. TL1: Timer 1 Low Byte

Bit	7	6	5	4	3	2	1	0
Name	TL1[7:0]							
Type	R/W							
Reset	0	0	0	0	0	0	0	0

SFR Page = 0x0; SFR Address = 0x8B

Bit	Name	Function
7:0	TL1[7:0]	**Timer 1 Low Byte.** The TL1 register is the low byte of the 16-bit Timer 1.

SFR Definition 25.6. TH0: Timer 0 High Byte

Bit	7	6	5	4	3	2	1	0
Name	TH0[7:0]							
Type	R/W							
Reset	0	0	0	0	0	0	0	0

SFR Page = 0x0; SFR Address = 0x8C

Bit	Name	Function
7:0	TH0[7:0]	**Timer 0 High Byte.** The TH0 register is the high byte of the 16-bit Timer 0.

SFR Definition 25.7. TH1: Timer 1 High Byte

Bit	7	6	5	4	3	2	1	0
Name	TH1[7:0]							
Type	R/W							
Reset	0	0	0	0	0	0	0	0

SFR Page = 0x0; SFR Address = 0x8D

Bit	Name	Function
7:0	TH1[7:0]	**Timer 1 High Byte.** The TH1 register is the high byte of the 16-bit Timer 1.

Document Change List

Revision 0.3 to Revision 0.4

- QFN-20 package and landing diagram updated.

Revision 0.4 to Revision 1.0

- IREF0CF register description updated.
- Updated ADC0 Chapter Text.
- Corrected an error in the Product Selector Guide.
- Updated SmaRTClock chapter to indicate how the Alarm value should be set when using Auto Reset and the LFO.
- Updated electrical specifications to fill TBDs and updated power specifications based on Rev B characterization data.
- Added a note to the OSCICL register description.
- Added a note to the CRC0CN register description.
- Updated equation in the CRC0CNT register description.
- Updated Power On Reset description.

Revision 1.0 to Revision 1.1

- Removed references to AN338.

Revision 1.1 to Revision 1.2

- Removed QuickSense references.
- Updated part numbers to Revision C in “Ordering Information” on page 31 and added Figure 3.4, Figure 3.5, and Figure 3.6 to identify the silicon revision.
- Updated REVID register (SFR Definition 14.2) and REVID C2 register (C2 Register Definition 27.3) with the 0x02 value for Revision C.
- Updated Figure “7.3 CP0 Multiplexer Block Diagram” on page 98 to remove the bar over the CPnOUT signals.
- Updated the “Reset Sources” on page 181 chapter to reflect the correct state of the $\overline{\text{RST}}$ pin during power-on reset.
- Updated Figure “1.14 Port I/O Functional Block Diagram” on page 26 and Figure “21.1 Port I/O Functional Block Diagram” on page 215 to mention P1.4 is not available on 20-pin devices.
- Removed references to the EMI0CN register, which does not exist.
- Updated Figure “8.2 Auto-Scan Example” on page 103 to refer to the correct pins.
- Updated POR Monitor Threshold (V_{POR}) Brownout Condition (VDD Falling) specification minimum, typical, and maximum values.
- Updated the reset value of the CLKSEL register (SFR Definition 19.1).
- Updated description of WEAKPUD in SFR Definition 21.3.
- Corrected SFR addresses for P0DRV (SFR Definition 21.12), P1DRV (SFR Definition 21.17), P2DRV (SFR Definition 21.20), PMU0MD (SFR Definition 15.3), FLSCL (SFR Definition 14.5), REF0CN (SFR Definition 5.15), CS0SCAN0 (SFR Definition 8.5), and CS0SCAN1 (SFR Definition 8.6).
- Replaced all instances of V_{BAT} with V_{DD}.
- Added a note to “11.1. Accessing XRAM” , “15.5. Sleep Mode” , and “18. Reset Sources” regarding an issue with the first address of XRAM.
- Added a note to “15.5. Sleep Mode” and “19. Clocking Sources” regarding using the internal low power