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### What is "[Embedded - Microcontrollers](#)"?

"[Embedded - Microcontrollers](#)" refer to small, integrated circuits designed to perform specific tasks within larger systems. These microcontrollers are essentially compact computers on a single chip, containing a processor core, memory, and programmable input/output peripherals. They are called "embedded" because they are embedded within electronic devices to control various functions, rather than serving as standalone computers. Microcontrollers are crucial in modern electronics, providing the intelligence and control needed for a wide range of applications.

### Applications of "[Embedded - Microcontrollers](#)"

#### Details

Product Status	Active
Core Processor	e200z7
Core Size	32-Bit Dual-Core
Speed	220MHz
Connectivity	CANbus, EBI/EMI, Ethernet, FlexCANbus, LINbus, SCI, SPI
Peripherals	DMA, LVD, POR, Zipwire
Number of I/O	293
Program Memory Size	4MB (4M x 8)
Program Memory Type	FLASH
EEPROM Size	-
RAM Size	512K x 8
Voltage - Supply (Vcc/Vdd)	3V ~ 5.5V
Data Converters	A/D 40x12b eQADCx2
Oscillator Type	External
Operating Temperature	-40°C ~ 125°C (TA)
Mounting Type	Surface Mount
Package / Case	416-BGA
Supplier Device Package	416-MAPBGA (27x27)
Purchase URL	<a href="https://www.e-xfl.com/product-detail/nxp-semiconductors/spc5775bdk3mme2">https://www.e-xfl.com/product-detail/nxp-semiconductors/spc5775bdk3mme2</a>

# 1 Introduction

## 1.1 Features summary

On-chip modules available within the family include the following features:

- Three dual issue, 32-bit CPU core complexes (e200z7), two of which run in lockstep
  - Power Architecture embedded specification compliance
  - Instruction set enhancement allowing variable length encoding (VLE), optional encoding of mixed 16-bit and 32-bit instructions, for code size footprint reduction
  - On the two computational cores: Signal processing extension (SPE1.1) instruction support for digital signal processing (DSP)
  - Single-precision floating point operations
  - On the two computational cores: 16 KB I-Cache and 16 KB D-Cache
  - Hardware cache coherency between cores
- 16 hardware semaphores
- 3-channel CRC module
- 4 MB on-chip flash memory
  - Supports read during program and erase operations, and multiple blocks allowing EEPROM emulation
- 512 KB on-chip general-purpose SRAM including 64 KB standby RAM
- Two multichannel direct memory access controllers (eDMA)
  - 64 channels per eDMA
- Dual core Interrupt Controller (INTC)
- Dual phase-locked loops (PLLs) with stable clock domain for peripherals and frequency modulation (FM) domain for computational shell
- Crossbar Switch architecture for concurrent access to peripherals, flash memory, or RAM from multiple bus masters with End-To-End ECC
- System Integration Unit (SIU)
- Error Injection Module (EIM) and Error Reporting Module (ERM)
- Four protected port output (PPO) pins
- Boot Assist Module (BAM) supports serial bootload via CAN or SCI
- Up to three second-generation Enhanced Time Processor Units (eTPUs)
  - 32 channels per eTPU
  - Total of 36 KB code RAM
  - Total of 9 KB parameter RAM

## Electrical characteristics

	1	2	3	4	5	6	7	8	9	10	11	12	13	14	15	16	17	18	19	20	21	22	23	24	25	26	
A	VSS	VDD	RSTOUT	ANA0_SDA0	ANA4	ANA8	ANA11	ANA14	VDDA_SD	REFBYP25	VRL_SD	VRH_SD	AN28	AN32	AN36	VDDA_EG	REFBYP25	VRL_EQ	VRH_EQ	ANB7_SDD7	ANB11	ANB14	ANB17	ANB21	ANB23	VSS	A
B	VDDH1	VSS	VDD	TEST	ANA1_SDA1	ANA5	ANA10	ANA14	VDDA_MIS0	VSSA_SD	REFBYP75	AN24	AN27	AN29	AN33	VDDA_EG	VSSA_EQ	REFBYP75	ANB6_SDD6	ANB8	ANB10	ANB15	ANB18	ANB22	VSS	TCRCLK	B
C	ETPUA30	ETPUA31	VSS	VDD	ANA2_SDA2	ANA6	ANA9	ANA13	ANA17_SDB1	ANA19_SDB3	ANA21_SDC1	ANA23_SDC3	AN26	AN30	AN34	AN37	AN38	ANB0_SDD0	ANB4_SDD4	ANB5_SDD5	ANB12	ANB16	ANB19	VSS	ETPUC0	ETPUC1	C
D	ETPUA27	ETPUA28	ETPUA29	VSS	VDD	ANA3_SDA3	ANA7	ANA12	ANA16_SDB0	ANA18_SDB2	ANA20_SDC0	ANA22_SDC2	AN25	AN31	AN35	AN39	ANB1_SDD1	ANB2_SDD2	ANB3_SDD3	ANB9	ANB13	ANB20	VSS	SEN72_A	ETPUC2	ETPUC3	D
E	ETPUA23	ETPUA24	ETPUA25	ETPUA26																			VDDH7	ETPUC4	ETPUC5	ETPUC6	E
F	ETPUA19	ETPUA20	ETPUA21	ETPUA22																			ETPUC7	ETPUC8	ETPUC9	ETPUC10	F
G	ETPUA15	ETPUA16	ETPUA17	ETPUA18																			ETPUC11	ETPUC12	ETPUC13	ETPUC14	G
H	ETPUA11	ETPUA12	ETPUA14	ETPUA13																			ETPUC15	ETPUC16	ETPUC17	ETPUC18	H
J	ETPUA7	ETPUA8	ETPUA9	ETPUA10																			ETPUC19	ETPUC20	ETPUC21	ETPUC22	J
K	ETPUA3	ETPUA4	ETPUA5	ETPUA6																			ETPUC23	ETPUC24	ETPUC25	ETPUC26	K
L	TCRCLKA	ETPUA0	ETPUA1	ETPUA2																			ETPUC27	ETPUC28	ETPUC29	ETPUC30	L
M	NC	TXDA	RXDA	VSTBY																			ETPUC31	ETPU815	ETPU814	VDDH7	M
N	RXDB	BOOTCFG1	WKPCFG	VDD																			ETPU86	ETPU811	ETPU812	ETPU813	N
P	TXDB	PLLCFG1	PLLCFG2	VDDH1																			ETPU87	ETPU88	ETPU89	ETPU810	P
R	JCOMP	RESET	PLLCFG0	RDY																			ETPU83	ETPU84	ETPU85	ETPU86	R
T	VDDH2	MCKO	MSE01	EVTI																			TCRCLK8	ETPU80	ETPU81	ETPU82	T
U	EVTO	MSE00	MDO0	MDO1																			ETPU819	ETPU818	ETPU817	ETPU816	U
V	MDO2	MDO3	MDO4	MDO5																			ETPU826	ETPU822	ETPU821	ETPU820	V
W	MDO6	MDO7	MDO8	VDDH2																			REGSEL	ETPU825	ETPU824	ETPU823	W
Y	MDO9	MDO10	MDO11	MDO15																			ETPU829	ETPU828	ETPU827	REGCTL	Y
AA	MDO12	MDO13	MDO14	NC																			VDDPMC	ETPU830	VDDPWR	VSS5YN	AA
AB	TDO	TCK	TMS	VDD																			VDD	ETPU831	VSSPWR	XTAL	AB
AC	VDDH2	TDI	VDD	VSS	FEC_TXCLKREFCLK	PCSA1	PCSA2	PCSA4	PCSB1	VDDH3	VDDH4	VDD	EMIOS8	EMIOS14	EMIOS18	EMIOS22	EMIOS27	EMIOS31	CNRX8	CNRXD	VDDH5	PCSC1	VSSPMC	VDD	VDDH6	XTAL	AC
AD	ENGCLK	VDD	VSS	FEC_TXD0	FEC_TXD1	PCSA5	SOUTA	SCKA	PCSB0	PCSB3	EMIOS2	EMIOS5	EMIOS9	EMIOS15	EMIOS19	EMIOS23	EMIOS26	EMIOS30	CNTX8	CNTXD	SCKC	RXDC	PCSC3	VSS	VDD	VDDFLA	AD
AE	VDD	VSS	FEC_RXDV	FEC_TXEN	PCSA4	PCSA0	PCSA3	SCKB	SINB	EMIOS0	EMIOS3	EMIOS6	EMIOS10	EMIOS13	EMIOS17	EMIOS21	EMIOS25	EMIOS29	CNRXA	CNRXC	PCSC0	SINC	PCSC2	PCSC5	VSS	VDD	AE
AF	VSS	VDDH2A	FEC_RXD0	FEC_RXD1	VDDH3A	PCSB5	SINB	SCKB	SOUTB	EMIOS1	EMIOS4	EMIOS7	EMIOS11	EMIOS12	EMIOS16	EMIOS20	EMIOS24	EMIOS28	CNTXA	CNTXC	SOUTC	VDDH4	TXDC	PCSC4	VDDH5	VSS	AF

Figure 2. MPC5775E 416-ball MAPBGA (full diagram)

## 3 Electrical characteristics

The following information includes details about power considerations, DC/AC electrical characteristics, and AC timing specifications.

### 3.1 Absolute maximum ratings

Absolute maximum specifications are stress ratings only. Functional operation at these maxima is not guaranteed.

#### CAUTION

Stress beyond listed maxima may affect device reliability or cause permanent damage to the device.

See [Operating conditions](#) for functional operation specifications.

Table 3. Device operating conditions (continued)

Symbol	Parameter	Conditions	Value			Unit
			Min	Typ	Max	
$V_{RL\_SD}$	SDADC ground reference voltage	—	$V_{SSA\_SD}$			V
$V_{DDA\_SD}$	SDADC supply voltage <sup>15</sup>	—	4.5	—	5.5	V
$V_{DDA\_EQA/B}$	eQADC supply voltage	—	4.75	—	5.25	V
$V_{RH\_SD}$	SDADC reference	—	4.5	$V_{DDA\_SD}$	5.5	V
$V_{DDA\_SD} - V_{RH\_SD}$	SDADC reference differential voltage	—	—	—	25	mV
$V_{SSA\_SD} - V_{RL\_SD}$	$V_{RL\_SD}$ differential voltage	—	–25	—	25	mV
$V_{RH\_EQ}$	eQADC reference	—	4.75	—	5.25	V
$V_{DDA\_EQA/B} - V_{RH\_EQ}$	eQADC reference differential voltage	—	—	—	25	mV
$V_{SSA\_EQ} - V_{RL\_EQ}$	$V_{RL\_EQ}$ differential voltage	—	–25	—	25	mV
$V_{SSA\_EQ} - V_{SS}$	$V_{SSA\_EQ}$ differential voltage	—	–25	—	25	mV
$V_{SSA\_SD} - V_{SS}$	$V_{SSA\_SD}$ differential voltage	—	–25	—	25	mV
$V_{RAMP}$	Slew rate on power supply pins	—	—	—	100	V/ms
<b>Current</b>						
$I_{IC}$	DC injection current (per pin) <sup>16, 17, 18</sup>	Digital pins and analog pins	–3.0	—	3.0	mA
$I_{MAXSEG}$	Maximum current per power segment <sup>19, 20</sup>	—	–80	—	80	mA

- Maximum operating frequency is applicable to the computational cores and platform for the device. See the Clocking chapter in the MPC5775E Microcontroller Reference Manual for more information on the clock limitations for the various IP blocks on the device.
- If frequency modulation (FM) is enabled, the maximum frequency still cannot exceed this value.
- MPC5775B Max value is 220 MHz.
- MPC5775B Max value is 110 MHz
- The maximum specification for operating junction temperature  $T_J$  must be respected. [Thermal characteristics](#) provides details.
- Core voltage as measured on device pin to guarantee published silicon performance
- During power ramp, voltage measured on silicon might be lower. Maximum performance is not guaranteed, but correct silicon operation is guaranteed. See power management and reset management for description.
- Maximum core voltage is not permitted for entire product life. See absolute maximum rating.
- When internal LVD/HVDs are disabled, external monitoring is required to guarantee device operation. Failure to monitor externally supply voltage may result in erroneous operation of the device.
- This LVD/HVD disabled supply voltage condition only applies after LVD/HVD are disabled by the application during the reset sequence, and the LVD/HVD are active until that point.
- This spec does not apply to  $V_{DDEH1}$ .
- When internal flash memory regulator is used:
  - Flash memory read operation is supported for a minimum  $V_{DDPMC}$  value of 3.15 V.
  - Flash memory read, program, and erase operations are supported for a minimum  $V_{DDPMC}$  value of 3.5 V.

When flash memory power is supplied externally ( $V_{DDPMC}$  shorted to  $V_{DDFLA}$ ): The  $V_{DDPMC}$  range must be within the limits specified for LVD\_FLASH and HVD\_FLASH monitoring. [Table 26](#) provides the monitored LVD\_FLASH and HVD\_FLASH limits.

- If the standby RAM regulator is not used, the  $V_{STBY}$  supply input pin must be tied to ground.
- $V_{STBY\_BO}$  is the maximum voltage that sets the standby RAM brownout flag in the device logic. The minimum voltage for RAM data retention is guaranteed always to be less than the  $V_{STBY\_BO}$  maximum value.

**Table 7. I/O pullup/pulldown DC electrical characteristics**

Symbol	Parameter	Conditions	Value			Unit
			Min	Typ	Max	
I <sub>WPU</sub>	Weak pullup current	V <sub>IN</sub> = 0.35 * V <sub>DDEX</sub> 4.5 V < V <sub>DDEX</sub> < 5.5 V	40	—	120	μA
		V <sub>IN</sub> = 0.35 * V <sub>DDEX</sub> 3.0 V < V <sub>DDEX</sub> < 3.6 V	25	—	80	
I <sub>WPD</sub>	Weak pulldown current	V <sub>IN</sub> = 0.65 * V <sub>DDEX</sub> 4.5 V < V <sub>DDEX</sub> < 5.5 V	40	—	120	μA
		V <sub>IN</sub> = 0.65 * V <sub>DDEX</sub> 3.0 V < V <sub>DDEX</sub> < 3.6 V	25	—	80	

The specifications in [Table 8](#) apply to the pins ANA0\_SDA0 to ANA7, ANA16\_SDB0 to ANA23\_SDC3, and ANB0\_SDD0 to ANB7\_SDD7.

**Table 8. I/O pullup/pulldown resistance electrical characteristics**

Symbol	Parameter	Conditions	Value			Unit
			Min	Typ	Max	
R <sub>PUPD</sub>	Analog input bias / diagnostic pullup/pulldown resistance	200 kΩ	130	200	280	kΩ
		100 kΩ	65	100	140	
		5 kΩ	1.4	5	7.5	
Δ <sub>PUPD</sub>	R <sub>PUPD</sub> pullup/pulldown resistance mismatch	—	—	—	5	%

### 3.6.2 Output pad specifications

[Figure 4](#) shows output DC electrical characteristics.

**Table 12. PLL1 electrical characteristics (continued)**

Symbol	Parameter	Conditions	Value			Unit
			Min	Typ	Max	
$\Delta_{\text{PLL1IN}}$	PLL1 input clock duty cycle <sup>1</sup>	—	35	—	65	%
$f_{\text{PLL1VCO}}$	PLL1 VCO frequency	—	600	—	1250	MHz
$f_{\text{PLL1PHI}}$	PLL1 output clock PHI	—	4.762	—	264 <sup>2</sup>	MHz
$t_{\text{PLL1LOCK}}$	PLL1 lock time	—	—	—	100	$\mu\text{s}$
$ \Delta_{\text{PLL1PHISPJ}} $	PLL1_PHI single period peak-to-peak jitter	$f_{\text{PLL1PHI}} = 200 \text{ MHz}$ , 6-sigma	—	—	500 <sup>3</sup>	ps
$f_{\text{PLL1MOD}}$	PLL1 modulation frequency	—	—	—	250	kHz
$ \delta_{\text{PLL1MOD}} $	PLL1 modulation depth (when enabled)	Center spread	0.25	—	2	%
		Down spread	0.5	—	4	%
$I_{\text{PLL1}}$	PLL1 consumption	FINE LOCK state	—	—	6	mA

1. PLL1IN clock retrieved directly from either internal PLL0 or external XOSC clock. Input characteristics are granted when using internal PLL0 or external oscillator in functional mode.
2. MPC5775B Max value is 220MHz.
3. Noise on the  $V_{\text{DD}}$  supply with frequency content below 40 kHz and above 50 MHz is filtered by the PLL. Noise on the  $V_{\text{DD}}$  supply with frequency content in the range of 40 kHz – 50 MHz must be filtered externally to the device.

### 3.7.2 Oscillator electrical specifications

#### NOTE

All oscillator specifications in Table 13 are valid for  $V_{\text{DDEH6}} = 3.0 \text{ V}$  to  $5.5 \text{ V}$ .

**Table 13. External oscillator (XOSC) electrical specifications**

Symbol	Parameter	Conditions	Value		Unit
			Min	Max	
$f_{\text{XTAL}}$	Crystal frequency range	—	8	40	MHz
$t_{\text{cst}}$	Crystal start-up time <sup>1, 2</sup>	$T_{\text{J}} = 150 \text{ }^{\circ}\text{C}$	—	5	ms
$t_{\text{rec}}$	Crystal recovery time <sup>3</sup>	—	—	0.5	ms
$V_{\text{IHEXT}}$	EXTAL input high voltage (external reference)	$V_{\text{REF}} = 0.28 * V_{\text{DDEH6}}$	$V_{\text{REF}} + 0.6$	—	V
$V_{\text{ILEXT}}$	EXTAL input low voltage (external reference)	$V_{\text{REF}} = 0.28 * V_{\text{DDEH6}}$	—	$V_{\text{REF}} - 0.6$	V
$C_{\text{S\_EXTAL}}$	Total on-chip stray capacitance on EXTAL pin <sup>4</sup>	416-ball MAPBGA	2.3	3.0	pF
$C_{\text{S\_XTAL}}$	Total on-chip stray capacitance on XTAL pin <sup>4</sup>	416-ball MAPBGA	2.3	3.0	pF
$g_{\text{m}}$	Oscillator transconductance <sup>5</sup>	Low	3	10	mA/V
		Medium	10	27	
		High	12	35	
$V_{\text{EXTAL}}$	Oscillation amplitude on the EXTAL pin after startup <sup>6</sup>	—	0.5	1.6	V
$V_{\text{HYS}}$	Comparator hysteresis	—	0.1	1.0	V
$I_{\text{XTAL}}$	XTAL current <sup>6, 7</sup>	—	—	14	mA

11. All channels have same  $10\text{ k}\Omega < R_s < 100\text{ k}\Omega$ . Channel under test has  $R_s = 10\text{ k}\Omega$ ,  $I_{INJ} = I_{INJMAX}, I_{INJMIN}$ .
12. The TUE specification is always less than the sum of the INL, DNL, offset, and gain errors due to cancelling errors.
13. TUE does not apply to differential conversions.
14. Variable gain is controlled by setting the PRE\_GAIN bits in the ADC\_ACR1-8 registers to select a gain factor of  $\times 1$ ,  $\times 2$ , or  $\times 4$ . Settings are for differential input only. Tested at  $\times 1$  gain. Values for other settings are guaranteed as indicated.
15. Guaranteed 10-bit monotonicity.
16. At  $V_{RH\_EQ} - V_{RL\_EQ} = 5.12\text{ V}$ , one LSB =  $1.25\text{ mV}$ .

### 3.8.2 Sigma-Delta ADC (SDADC)

The SDADC is a 16-bit Sigma-Delta analog-to-digital converter with a 333 Ksps maximum output conversion rate.

#### NOTE

The voltage range is 4.5 V to 5.5 V for SDADC specifications, except where noted otherwise.

**Table 17. SDADC electrical specifications**

Symbol	Parameter	Conditions	Value			Unit
			Min	Typ	Max	
V <sub>IN</sub>	ADC input signal	—	0	—	V <sub>DDA_SD</sub>	V
V <sub>IN_PK2PK</sub> <sup>1</sup>	Input range peak to peak  V <sub>IN_PK2PK</sub> = V <sub>INP</sub> <sup>2</sup> – V <sub>INM</sub> <sup>3</sup>	Single ended V <sub>INM</sub> = V <sub>RL_SD</sub>	V <sub>RH_SD</sub> /GAIN			V
		Single ended V <sub>INM</sub> = 0.5*V <sub>RH_SD</sub> GAIN = 1	±0.5*V <sub>RH_SD</sub>			
		Single ended V <sub>INM</sub> = 0.5*V <sub>RH_SD</sub> GAIN = 2,4,8,16	±V <sub>RH_SD</sub> /GAIN			
		Differential 0 < V <sub>IN</sub> < V <sub>DDEx</sub>	±V <sub>RH_SD</sub> /GAIN			
f <sub>ADCD_M</sub>	SD clock frequency <sup>4</sup>	—	4	14.4	16	MHz
f <sub>ADCD_S</sub>	Conversion rate	—	—	—	333	Ksps
—	Oversampling ratio	Internal modulator	24	—	256	—
RESOLUTION	SD register resolution <sup>5</sup>	2's complement notation	16			bit
GAIN	ADC gain	Defined through SDADC_MCR[PGAN]. Only integer powers of 2 are valid gain values.	1	—	16	—

Table continues on the next page...

Table 17. SDADC electrical specifications (continued)

Symbol	Parameter	Conditions	Value			Unit
			Min	Typ	Max	
$ \delta_{\text{GAIN}} $	Absolute value of the ADC gain error <sup>6, 7</sup>	Before calibration (applies to gain setting = 1)	—	—	1.5	%
		After calibration $\Delta V_{\text{RH\_SD}} < 5\%$ , $\Delta V_{\text{DDA\_SD}} < 10\%$ $\Delta T_{\text{J}} < 50\text{ }^{\circ}\text{C}$	—	—	5	mV
		After calibration $\Delta V_{\text{RH\_SD}} < 5\%$ , $\Delta V_{\text{DDA\_SD}} < 10\%$ $\Delta T_{\text{J}} < 100\text{ }^{\circ}\text{C}$	—	—	7.5	
		After calibration $\Delta V_{\text{RH\_SD}} < 5\%$ , $\Delta V_{\text{DDA\_SD}} < 10\%$ $\Delta T_{\text{J}} < 150\text{ }^{\circ}\text{C}$	—	—	10	
$V_{\text{OFFSET}}$	Conversion offset <sup>6, 7</sup>	Before calibration (applies to all gain settings: 1, 2, 4, 8, 16)	—	$10 \cdot (1 + 1/\text{gain})$	20	mV
		After calibration $\Delta V_{\text{DDA\_SD}} < 10\%$ $\Delta T_{\text{J}} < 50\text{ }^{\circ}\text{C}$	—	—	5	
		After calibration $\Delta V_{\text{DDA\_SD}} < 10\%$ $\Delta T_{\text{J}} < 100\text{ }^{\circ}\text{C}$	—	—	7.5	
		After calibration $\Delta V_{\text{DDA\_SD}} < 10\%$ $\Delta T_{\text{J}} < 150\text{ }^{\circ}\text{C}$	—	—	10	
$\text{SNR}_{\text{DIFF150}}$	Signal to noise ratio in differential mode, 150 Ksps output rate	$4.5\text{ V} < V_{\text{DDA\_SD}} < 5.5\text{ V}$ <sup>8, 9</sup> $V_{\text{RH\_SD}} = V_{\text{DDA\_SD}}$ GAIN = 1	80	—	—	dB
		$4.5\text{ V} < V_{\text{DDA\_SD}} < 5.5\text{ V}$ <sup>8, 9</sup> $V_{\text{RH\_SD}} = V_{\text{DDA\_SD}}$ GAIN = 2	77	—	—	
		$4.5\text{ V} < V_{\text{DDA\_SD}} < 5.5\text{ V}$ <sup>8, 9</sup> $V_{\text{RH\_SD}} = V_{\text{DDA\_SD}}$ GAIN = 4	74	—	—	
		$4.5\text{ V} < V_{\text{DDA\_SD}} < 5.5\text{ V}$ <sup>8, 9</sup> $V_{\text{RH\_SD}} = V_{\text{DDA\_SD}}$ GAIN = 8	71	—	—	
		$4.5\text{ V} < V_{\text{DDA\_SD}} < 5.5\text{ V}$ <sup>8, 9</sup> $V_{\text{RH\_SD}} = V_{\text{DDA\_SD}}$ GAIN = 16	68	—	—	

Table continues on the next page...



**Table 18. Temperature Sensor electrical characteristics**

Symbol	Parameter	Conditions	Value			Unit
			Min	Typ	Max	
—	Temperature monitoring range	—	−40	—	150	°C
T <sub>SENS</sub>	Sensitivity	—	—	5.18	—	mV/°C
T <sub>ACC</sub>	Accuracy	−40°C < T <sub>J</sub> < 150°C	−5	—	5	°C
I <sub>TEMP_SENS</sub>	V <sub>DDA_EQA</sub> power supply current, per Temp Sensor	—	—	—	700	μA

### 3.10 LVDS pad electrical characteristics

The LVDS pad is used for the Microsecond Channel (MSC) and DSPI LVDS interfaces, with different characteristics given in the following tables.

### 3.10.1 MSC/DSPi LVDS interface timing diagrams

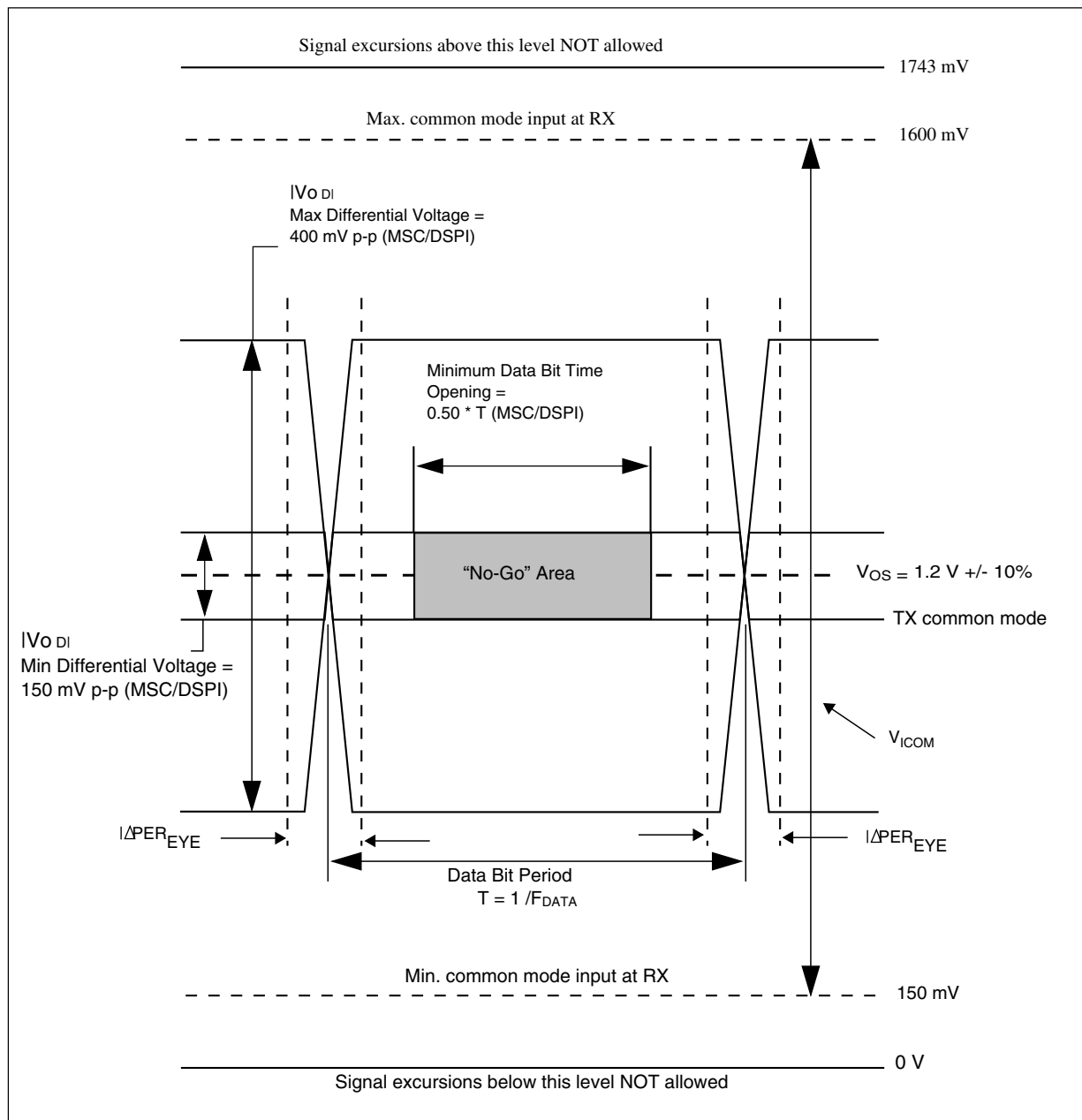


Figure 7. MSC/DSPi LVDS timing definition

**Table 26. Voltage monitor electrical characteristics<sup>1, 2</sup> (continued)**

Symbol	Parameter	Conditions	Configuration			Value			Unit
			Trim bits	Mask Opt.	Pow. Up	Min	Typ	Max	
POR_HV	HV $V_{DDPMC}$ supply power on reset threshold	Rising voltage (powerup)	N/A	No	Enab.	2444	2600	2756	mV
		Falling voltage (power down)				2424	2580	2736	
LVD_HV	HV internal $V_{DDPMC}$ supply low voltage monitoring	Rising voltage (untrimmed)	4bit	No	Enab.	2935	3023	3112	mV
		Falling voltage (untrimmed)				2922	3010	3099	
		Rising voltage (trimmed)				2946	3010	3066	
		Falling voltage (trimmed)				2934	2998	3044	
HVD_HV	HV internal $V_{DDPMC}$ supply high voltage monitoring	Rising voltage	4bit	Yes	Disab.	5696	5860	5968	mV
		Falling voltage				5666	5830	5938	
LVD_FLASH	FLASH supply low voltage monitoring <sup>6</sup>	Rising voltage (untrimmed)	4bit	No	Enab.	2935	3023	3112	mV
		Falling voltage (untrimmed)				2922	3010	3099	
		Rising voltage (trimmed)				2956	3010	3053	
		Falling voltage (trimmed)				2944	2998	3041	
HVD_FLASH	FLASH supply high voltage monitoring <sup>6</sup>	Rising voltage	4bit	Yes	Disab.	3456	3530	3584	mV
		Falling voltage				3426	3500	3554	
LVD_IO	Main I/O $V_{DDEH1}$ supply low voltage monitoring	Rising voltage (untrimmed)	4bit	No	Enab.	3250	3350	3488	mV
		Falling voltage (untrimmed)				3220	3320	3458	
		Rising voltage (trimmed)				3347	3420	3468	
		Falling voltage (trimmed)				3317	3390	3438	
$t_{VDASSERT}$	Voltage detector threshold crossing assertion	—	—	—	—	0.1	—	2.0	$\mu$ s
$t_{VDRELEASE}$	Voltage detector threshold crossing de-assertion	—	—	—	—	5	—	20	$\mu$ s

1. LVD is released after  $t_{VDRELEASE}$  temporization when upper threshold is crossed; LVD is asserted  $t_{VDASSERT}$  after detection when lower threshold is crossed.
2. HVD is released after  $t_{VDRELEASE}$  temporization when lower threshold is crossed; HVD is asserted  $t_{VDASSERT}$  after detection when upper threshold is crossed.
3. POR098\_c threshold is an untrimmed value, before the completion of the power-up sequence. All other LVD/HVD thresholds are provided after trimming.
4. LV internal supply levels are measured on device internal supply grid after internal voltage drop.
5. LV external supply levels are measured on the die side of the package bond wire after package voltage drop.
6.  $V_{DDFLA}$  range is guaranteed when internal flash memory regulator is used.

### 3.11.4 Power sequencing requirements

Requirements for power sequencing include the following.

**Table 30. Flash memory AC timing specifications (continued)**

Symbol	Characteristic	Min	Typical	Max	Units
$t_{done}$	Time from 0 to 1 transition on the MCR-EHV bit initiating a program/erase until the MCR-DONE bit is cleared.	—	—	5	ns
$t_{dones}$	Time from 1 to 0 transition on the MCR-EHV bit aborting a program/erase until the MCR-DONE bit is set to a 1.	—	16 plus four system clock periods	20.8 plus four system clock periods	$\mu$ s
$t_{drcv}$	Time to recover once exiting low power mode.	16 plus seven system clock periods.	—	45 plus seven system clock periods	$\mu$ s
$t_{aistart}$	Time from 0 to 1 transition of UT0-AIE initiating a Margin Read or Array Integrity until the UT0-AID bit is cleared. This time also applies to the resuming from a suspend or breakpoint by clearing AISUS or clearing NAIBP	—	—	5	ns
$t_{aistop}$	Time from 1 to 0 transition of UT0-AIE initiating an Array Integrity abort until the UT0-AID bit is set. This time also applies to the UT0-AISUS to UT0-AID setting in the event of a Array Integrity suspend request.	—	—	80 plus fifteen system clock periods	ns
$t_{mrstop}$	Time from 1 to 0 transition of UT0-AIE initiating a Margin Read abort until the UT0-AID bit is set. This time also applies to the UT0-AISUS to UT0-AID setting in the event of a Margin Read suspend request.	10.36 plus four system clock periods	—	20.42 plus four system clock periods	$\mu$ s

### 3.12.6 Flash memory read wait-state and address-pipeline control settings

The following table describes the recommended settings of the Flash Memory Controller's PFCR1[RWSC] and PFCR1[APC] fields at various flash memory operating frequencies, based on specified intrinsic flash memory access times of the C55FMC array at 150°C.

**Table 31. Flash memory read wait-state and address-pipeline control combinations**

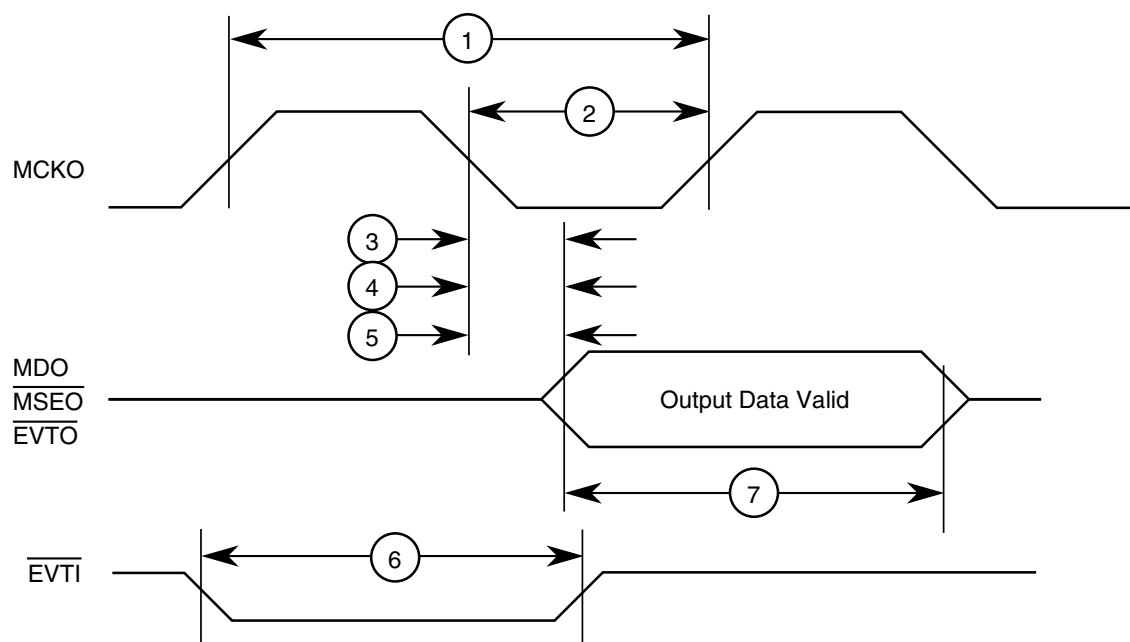
Flash memory frequency	RWSC	APC	Flash memory read latency on mini-cache miss (# of $f_{PLATF}$ clock periods)	Flash memory read latency on mini-cache hit (# of $f_{PLATF}$ clock periods)
0 MHz < $f_{PLATF}$ ≤ 33 MHz	0	0	3	1
33 MHz < $f_{PLATF}$ ≤ 100 MHz	2	1	5	1

Table continues on the next page...

**Table 34. Nexus debug port timing<sup>1</sup> (continued)**

Spec	Characteristic	Symbol	Min	Max	Unit
8	Absolute minimum TCK cycle time <sup>4</sup> (TDO sampled on posedge of TCK)	$t_{TCYC}$	40 <sup>5</sup>	—	ns
	Absolute minimum TCK cycle time <sup>4</sup> (TDO sampled on negedge of TCK)		20 <sup>5</sup>	—	
9	TCK Duty Cycle	$t_{TDC}$	40	60	%
10	TDI, TMS Data Setup Time <sup>6</sup>	$t_{NTDIS}, t_{NTMSS}$	8	—	ns
11	TDI, TMS Data Hold Time <sup>6</sup>	$T_{NTDIH}, t_{NTMSH}$	5	—	ns
12	TCK Low to TDO Data Valid <sup>6</sup>	$t_{NTDOV}$	0	18	ns
13	$\overline{RDY}$ Valid to MCKO <sup>7</sup>	—	—	—	—
14	TDO hold time after TCLK low <sup>6</sup>	$t_{NTDOH}$	1	—	ns

1. All Nexus timing relative to MCKO is measured from 50% of MCKO and 50% of the respective signal. Nexus timing specified at  $V_{DD} = 1.08\text{ V to }1.32\text{ V}$ ,  $V_{DDE} = 3.0\text{ V to }3.6\text{ V}$ ,  $V_{DD33}$  and  $V_{DDSYN} = 3.0\text{ V to }3.6\text{ V}$ ,  $T_A = T_L$  to  $T_H$ , and  $C_L = 30\text{ pF}$  with  $DSC = 0b10$ .
2. MDO, MSEO, and EVTO data is held valid until next MCKO low cycle.
3. This is a functionally allowable feature. However, it may be limited by the maximum frequency specified by the absolute minimum TCK period specification.
4. This value is TDO propagation time plus 2 ns setup time to sampling edge.
5. This may require a maximum clock speed that is less than the maximum functional capability of the design depending on the actual system frequency being used.
6. Applies to TMS pin timing for the bit frame when using the 1149.7 advanced protocol.
7. The  $\overline{RDY}$  pin timing is asynchronous to MCKO. The timing is guaranteed by design to function correctly.

**Figure 22. Nexus timings**

**Table 38. DSPI channel frequency support (continued)**

DSPI use mode		Max usable frequency (MHz) <sup>1, 2</sup>
LVDS (Master mode)	Full duplex – Modified timing (Table 41)	30
	Output only mode TSB mode (SCK/SOUT/PCS) (Table 42)	40

1. Maximum usable frequency can be achieved if used with fastest configuration of the highest drive pads.

2. Maximum usable frequency does not take into account external device propagation delay.

### 3.13.8.1 DSPI master mode full duplex timing with CMOS and LVDS pads

#### 3.13.8.1.1 DSPI CMOS Master Mode — Classic Timing

**Table 39. DSPI CMOS master classic timing (full duplex and output only) –**  
MTFE = 0, CPHA = 0 or 1<sup>1</sup>

#	Symbol	Characteristic	Condition <sup>2</sup>		Value <sup>3</sup>		Unit
			Pad drive <sup>4</sup>	Load (C <sub>L</sub> )	Min	Max	
1	t <sub>SCK</sub>	SCK cycle time	PCR[SRC]=11b	25 pF	33.0	—	ns
			PCR[SRC]=10b	50 pF	80.0	—	
			PCR[SRC]=01b	50 pF	200.0	—	
2	t <sub>CSC</sub>	PCS to SCK delay	PCR[SRC]=11b	25 pF	(N <sup>5</sup> × t <sub>SYS</sub> <sup>6</sup> ) – 16	—	ns
			PCR[SRC]=10b	50 pF	(N <sup>5</sup> × t <sub>SYS</sub> <sup>6</sup> ) – 16	—	
			PCR[SRC]=01b	50 pF	(N <sup>5</sup> × t <sub>SYS</sub> <sup>6</sup> ) – 18	—	
			PCS: PCR[SRC]=01b SCK: PCR[SRC]=10b	50 pF	(N <sup>5</sup> × t <sub>SYS</sub> <sup>6</sup> ) – 45	—	
3	t <sub>ASC</sub>	After SCK delay	PCR[SRC]=11b	PCS: 0 pF SCK: 50 pF	(M <sup>7</sup> × t <sub>SYS</sub> <sup>6</sup> ) – 35	—	ns
			PCR[SRC]=10b	PCS: 0 pF SCK: 50 pF	(M <sup>7</sup> × t <sub>SYS</sub> <sup>6</sup> ) – 35	—	
			PCR[SRC]=01b	PCS: 0 pF SCK: 50 pF	(M <sup>7</sup> × t <sub>SYS</sub> <sup>6</sup> ) – 35	—	
			PCS: PCR[SRC]=01b SCK: PCR[SRC]=10b	PCS: 0 pF SCK: 50 pF	(M <sup>7</sup> × t <sub>SYS</sub> <sup>6</sup> ) – 35	—	
4	t <sub>SDC</sub>	SCK duty cycle <sup>8</sup>	PCR[SRC]=11b	0 pF	1/2t <sub>SCK</sub> – 2	1/2t <sub>SCK</sub> + 2	ns
			PCR[SRC]=10b	0 pF	1/2t <sub>SCK</sub> – 2	1/2t <sub>SCK</sub> + 2	
			PCR[SRC]=01b	0 pF	1/2t <sub>SCK</sub> – 5	1/2t <sub>SCK</sub> + 5	
PCS strobe timing							
5	t <sub>PCSC</sub>	PCSx to PCSS time <sup>9</sup>	PCR[SRC]=10b	25 pF	13.0	—	ns
6	t <sub>PASC</sub>	PCSS to PCSx time <sup>9</sup>	PCR[SRC]=10b	25 pF	13.0	—	ns
SIN setup time							

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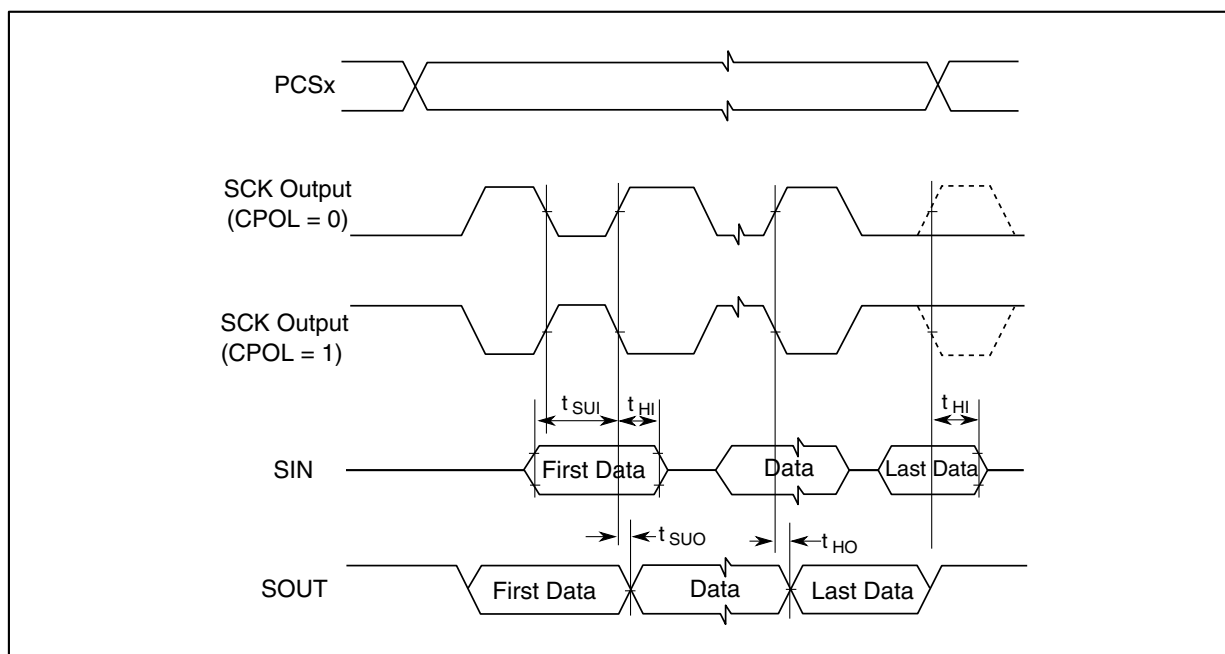


Figure 31. DSPI CMOS master mode – modified timing, CPHA = 1

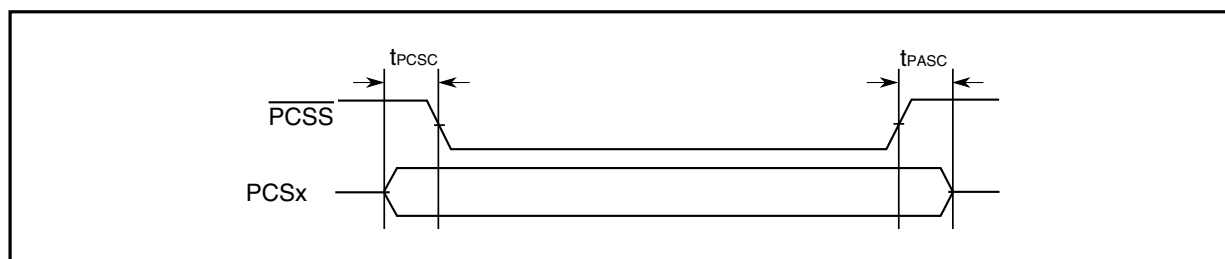


Figure 32. DSPI PCS strobe ( $\overline{\text{PCSS}}$ ) timing (master mode)

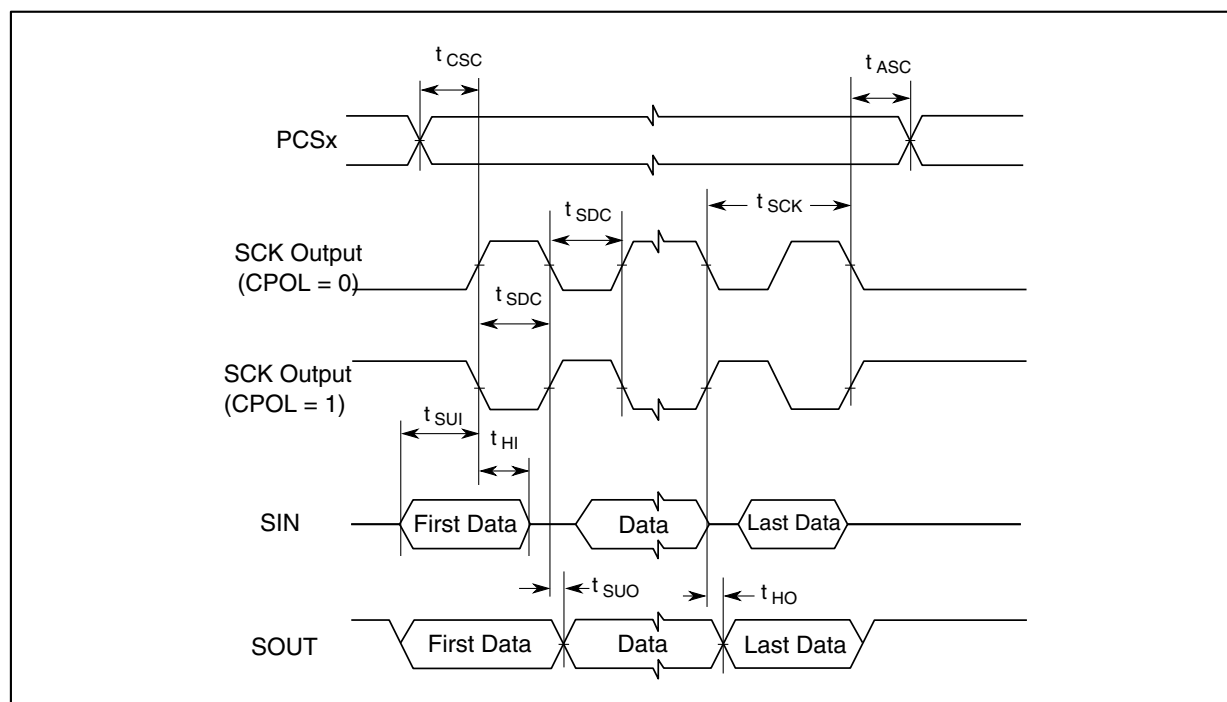
### 3.13.8.1.3 DSPI LVDS Master Mode – Modified Timing

Table 41. DSPI LVDS master timing – full duplex – modified transfer format (MTFE = 1), CPHA = 0 or 1

#	Symbol	Characteristic	Condition <sup>1</sup>		Value <sup>2</sup>		Unit
			Pad drive <sup>3</sup>	Load ( $C_L$ )	Min	Max	
1	$t_{\text{SCK}}$	SCK cycle time	LVDS	15 pF to 25 pF differential	33.3	—	ns
2	$t_{\text{CSC}}$	PCS to SCK delay (LVDS SCK)	PCS: PCR[SRC]=11b	25 pF	$(N^4 \times t_{\text{SYS}}^5) - 10$	—	ns
			PCS: PCR[SRC]=10b	50 pF	$(N^4 \times t_{\text{SYS}}^5) - 10$	—	ns
			PCS: PCR[SRC]=01b	50 pF	$(N^4 \times t_{\text{SYS}}^5) - 32$	—	ns

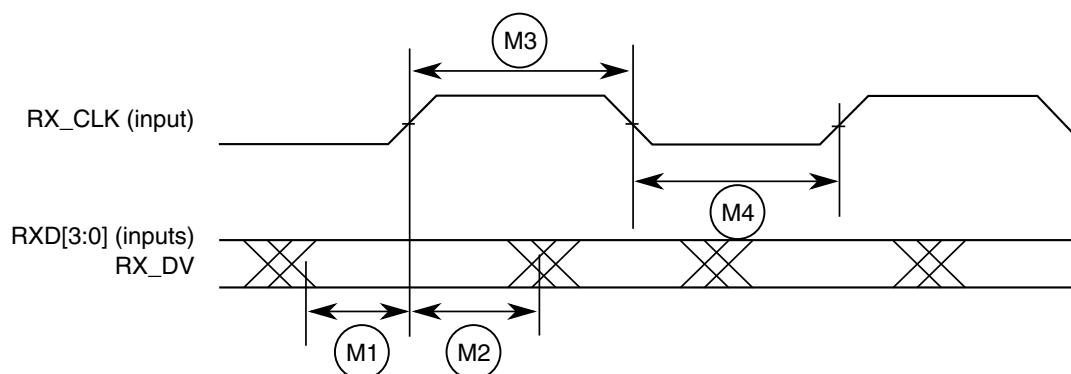
Table continues on the next page...

4. N is the number of clock cycles added to time between PCS assertion and SCK assertion and is software programmable using DSPI\_CTARx[PSSCK] and DSPI\_CTARx[CSSCK]. The minimum value is 2 cycles unless TSB mode or Continuous SCK clock mode is selected, in which case, N is automatically set to 0 clock cycles (PCS and SCK are driven by the same edge of DSPI\_CLKn).
5.  $t_{SYS}$  is the period of DSPI\_CLKn clock, the input clock to the DSPI module. Maximum frequency is 100 MHz (min  $t_{SYS}$  = 10 ns).
6. M is the number of clock cycles added to time between SCK negation and PCS negation and is software programmable using DSPI\_CTARx[PASC] and DSPI\_CTARx[ASC]. The minimum value is 2 cycles unless TSB mode or Continuous SCK clock mode is selected, in which case, M is automatically set to 0 clock cycles (PCS and SCK are driven by the same edge of DSPI\_CLKn).
7.  $t_{SDC}$  is only valid for even divide ratios. For odd divide ratios the fundamental duty cycle is not 50:50. For these odd divide ratios cases, the absolute spec number is applied as jitter/uncertainty to the nominal high time and low time.
8. Input timing assumes an input slew rate of 1 ns (10% – 90%) and LVDS differential voltage =  $\pm 100$  mV.
9. P is the number of clock cycles added to delay the DSPI input sample point and is software programmable using DSPI\_MCR[SMPL\_PT]. The value must be 0, 1 or 2. If the baud rate divide ratio is /2 or /3, this value is automatically set to 1.
10. SOUT Data Valid and Data hold are independent of load capacitance if SCK and SOUT load capacitances are the same value.



**Figure 33. DSPI LVDS master mode – modified timing, CPHA = 0**





**Figure 36. MII receive signal timing diagram**

### 3.13.9.2 MII transmit signal timing (TXD[3:0], TX\_EN, and TX\_CLK)

The transmitter functions correctly up to a TX\_CLK maximum frequency of 25 MHz +1%. There is no minimum frequency requirement. The system clock frequency must be at least equal to or greater than the TX\_CLK frequency.

The transmit outputs (TXD[3:0], TX\_EN) can be programmed to transition from either the rising or falling edge of TX\_CLK, and the timing is the same in either case. This options allows the use of noncompliant MII PHYs.

Refer to the *MPC5775E Microcontroller Reference Manual's* Fast Ethernet Controller (FEC) chapter for details of this option and how to enable it.

**Table 45. MII transmit signal timing<sup>1</sup>**

Symbol	Characteristic	Value <sup>2</sup>		Unit
		Min	Max	
M5	TX_CLK to TXD[3:0], TX_EN invalid	4.5	—	ns
M6	TX_CLK to TXD[3:0], TX_EN valid	—	25	ns
M7	TX_CLK pulse width high	35%	65%	TX_CLK period
M8	TX_CLK pulse width low	35%	65%	TX_CLK period

1. All timing specifications valid to the pad input levels defined in [I/O pad specifications](#).

2. Output parameters are valid for  $C_L = 25$  pF, where  $C_L$  is the external load to the device. The internal package capacitance is accounted for, and does not need to be subtracted from the 25 pF value.

The thermal performance of any component depends on the power dissipation of the surrounding components. In addition, the ambient temperature varies widely within the application. For many natural convection and especially closed box applications, the board temperature at the perimeter (edge) of the package is approximately the same as the local air temperature near the device. Specifying the local ambient conditions explicitly as the board temperature provides a more precise description of the local ambient conditions that determine the temperature of the device.

At a known board temperature, the junction temperature is estimated using the following equation:

$$T_J = T_B + (R_{\theta JB} * P_D)$$

where:

$T_B$  = board temperature for the package perimeter (°C)

$R_{\theta JB}$  = junction-to-board thermal resistance (°C/W) per JESD51-8

$P_D$  = power dissipation in the package (W)

When the heat loss from the package case to the air does not factor into the calculation, the junction temperature is predictable if the application board is similar to the thermal test condition, with the component soldered to a board with internal planes.

The thermal resistance is expressed as the sum of a junction-to-case thermal resistance plus a case-to-ambient thermal resistance:

$$R_{\theta JA} = R_{\theta JC} + R_{\theta CA}$$

where:

$R_{\theta JA}$  = junction-to-ambient thermal resistance (°C/W)

$R_{\theta JC}$  = junction-to-case thermal resistance (°C/W)

$R_{\theta CA}$  = case to ambient thermal resistance (°C/W)

$R_{\theta JC}$  is device related and is not affected by other factors. The thermal environment can be controlled to change the case-to-ambient thermal resistance,  $R_{\theta CA}$ . For example, change the air flow around the device, add a heat sink, change the mounting arrangement on the printed circuit board, or change the thermal dissipation on the printed circuit board surrounding the device. This description is most useful for packages with heat sinks where 90% of the heat flow is through the case to heat sink to ambient. For most packages, a better model is required.

A more accurate two-resistor thermal model can be constructed from the junction-to-board thermal resistance and the junction-to-case thermal resistance. The junction-to-case thermal resistance describes when using a heat sink or where a substantial amount of heat is dissipated from the top of the package. The junction-to-board thermal resistance describes the thermal performance when most of the heat is conducted to the printed circuit board. This model can be used to generate simple estimations and for computational fluid dynamics (CFD) thermal models. More accurate compact Flotherm models can be generated upon request.

To determine the junction temperature of the device in the application on a prototype board, use the thermal characterization parameter ( $\Psi_{JT}$ ) to determine the junction temperature by measuring the temperature at the top center of the package case using the following equation:

$$T_J = T_T + (\Psi_{JT} \times P_D)$$

where:

$T_T$  = thermocouple temperature on top of the package (°C)

$\Psi_{JT}$  = thermal characterization parameter (°C/W)

$P_D$  = power dissipation in the package (W)

The thermal characterization parameter is measured in compliance with the JESD51-2 specification using a 40-gauge type T thermocouple epoxied to the top center of the package case. Position the thermocouple so that the thermocouple junction rests on the package. Place a small amount of epoxy on the thermocouple junction and approximately 1 mm of wire extending from the junction. Place the thermocouple wire flat against the package case to avoid measurement errors caused by the cooling effects of the thermocouple wire.

When board temperature is perfectly defined below the device, it is possible to use the thermal characterization parameter ( $\Psi_{JPB}$ ) to determine the junction temperature by measuring the temperature at the bottom center of the package case (exposed pad) using the following equation:

$$T_J = T_B + (\Psi_{JPB} \times P_D)$$

where:

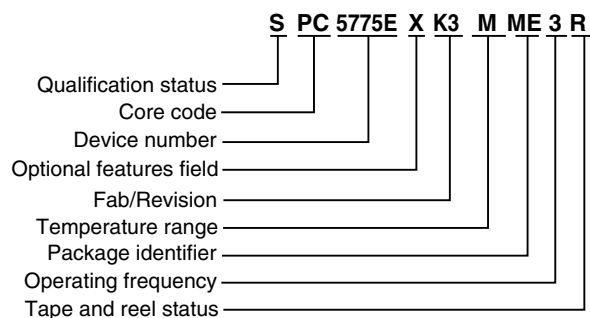
$T_T$  = thermocouple temperature on bottom of the package (°C)

$\Psi_{JT}$  = thermal characterization parameter (°C/W)

$P_D$  = power dissipation in the package (W)

## 5 Ordering information

Figure 42 and Table 51 describe orderable part numbers for the MPC5775B and MPC5775E.



**Device Number**  
MPC5775B or MPC5775E

**Temperature range**  
M = -40 °C to 125 °C

**Package identifier**  
ME = 416 MAPBGA Pb-Free

**Tape and reel status**  
R = Tape and reel  
(blank) = Trays

**Qualification status**  
S = Fully spec. qualified, automotive flow

**Operating frequency**  
3 = 264 MHz  
2 = 220 MHz

**Optional features field**  
D = ISO-compliant CAN FD available, trimmed for SMPS or external regulator, and include SHE compliant security firmware

**Figure 42. MPC5775B and MPC5775E Orderable part number description**

**Table 51. Example orderable part numbers**

Part number	Package description	Speed (MHz)	Operating temperature <sup>1</sup>	
			Min (T <sub>L</sub> )	Max (T <sub>H</sub> )
SPC5775BDK3MME2	SPC5775B 416 package Lead-free (Pb-free)	220	-40 °C	125 °C
SPC5775EDK3MME3	SPC5775E 416 package Lead-free (Pb-free)	264	-40 °C	125 °C

1. The lowest ambient operating temperature is referenced by T<sub>L</sub>; the highest ambient operating temperature is referenced by T<sub>H</sub>.

## 6 Document revision history

The following table summarizes revisions to this document since the previous release.

**Table 52. Revision history**

Revision	Date	Description of changes
1	05/2018	Initial release

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