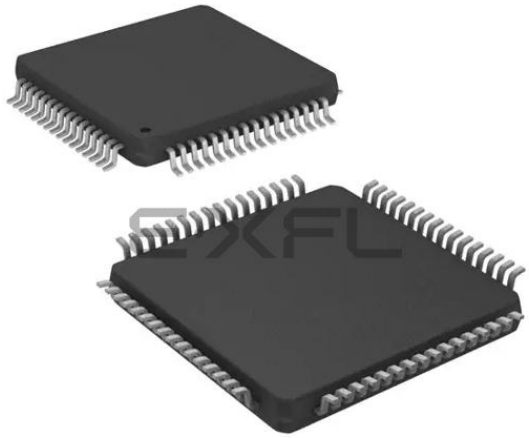

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Core Processor	PIC
Core Size	8-Bit
Speed	4MHz
Connectivity	I²C, SPI
Peripherals	LCD, POR, PWM, WDT
Number of I/O	25
Program Memory Size	7KB (4K x 14)
Program Memory Type	OTP
EEPROM Size	-
RAM Size	176 x 8
Voltage - Supply (Vcc/Vdd)	2.5V ~ 6V
Data Converters	-
Oscillator Type	External
Operating Temperature	0°C ~ 70°C (TA)
Mounting Type	Surface Mount
Package / Case	64-TQFP
Supplier Device Package	64-TQFP (10x10)
Purchase URL	https://www.e-xfl.com/product-detail/microchip-technology/pic16lc923t-04-pt

Table of Contents

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TABLE 3-1: PIC16C9XX PINOUT DESCRIPTION (Cont.'d)

Pin Name	DIP Pin#	PLCC Pin#	TQFP Pin#	Pin Type	Buffer Type	Description
$V_{LCD}3$	19	20	11	P	—	LCD Voltage.
V_{DD}	20, 60	22, 64	12, 52	P	—	Digital power.
V_{SS}	6, 21	7, 23	13, 62	P	—	Ground reference.
NC	—	1	—	—	—	These pins are not internally connected. These pins should be left unconnected.

Legend: I = input O = output P = power L = LCD Driver
— = Not used TTL = TTL input ST = Schmitt Trigger input

3.1 Clocking Scheme/Instruction Cycle

The clock input (from OSC1) is internally divided by four to generate four non-overlapping quadrature clocks namely Q1, Q2, Q3 and Q4. Internally, the program counter (PC) is incremented every Q1, the instruction is fetched from the program memory and latched into the instruction register in Q4. The instruction is decoded and executed during the following Q1 through Q4. The clocks and instruction execution flow is shown in Figure 3-3.

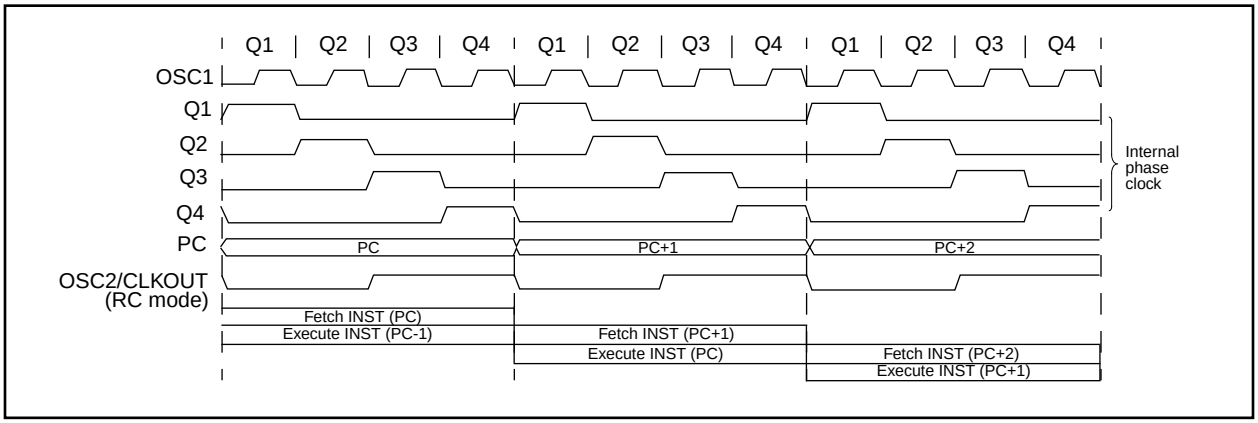
3.2 Instruction Flow/Pipelining

An “Instruction Cycle” consists of four Q cycles (Q1, Q2, Q3 and Q4). The instruction fetch and execute are pipelined such that fetch takes one instruction cycle while decode and execute takes another instruction cycle. However, due to the pipelining, each instruction effectively executes in one cycle. If an instruction causes the program counter to change (e.g. GOTO) then two cycles are required to complete the instruction (Example 3-1).

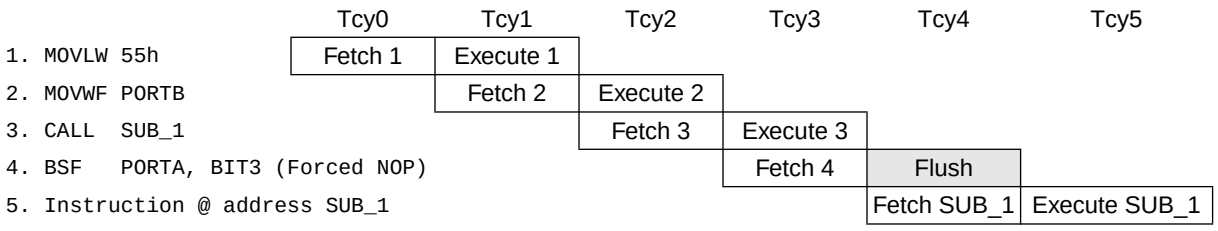
A fetch cycle begins with the program counter (PC) incrementing in Q1.

In the execution cycle, the fetched instruction is latched into the “Instruction Register" in cycle Q1. This instruction is then decoded and executed during the Q2, Q3, and Q4 cycles. Data memory is read during Q2 (operand read) and written during Q4 (destination write).

FIGURE 3-3: CLOCK/INSTRUCTION CYCLE

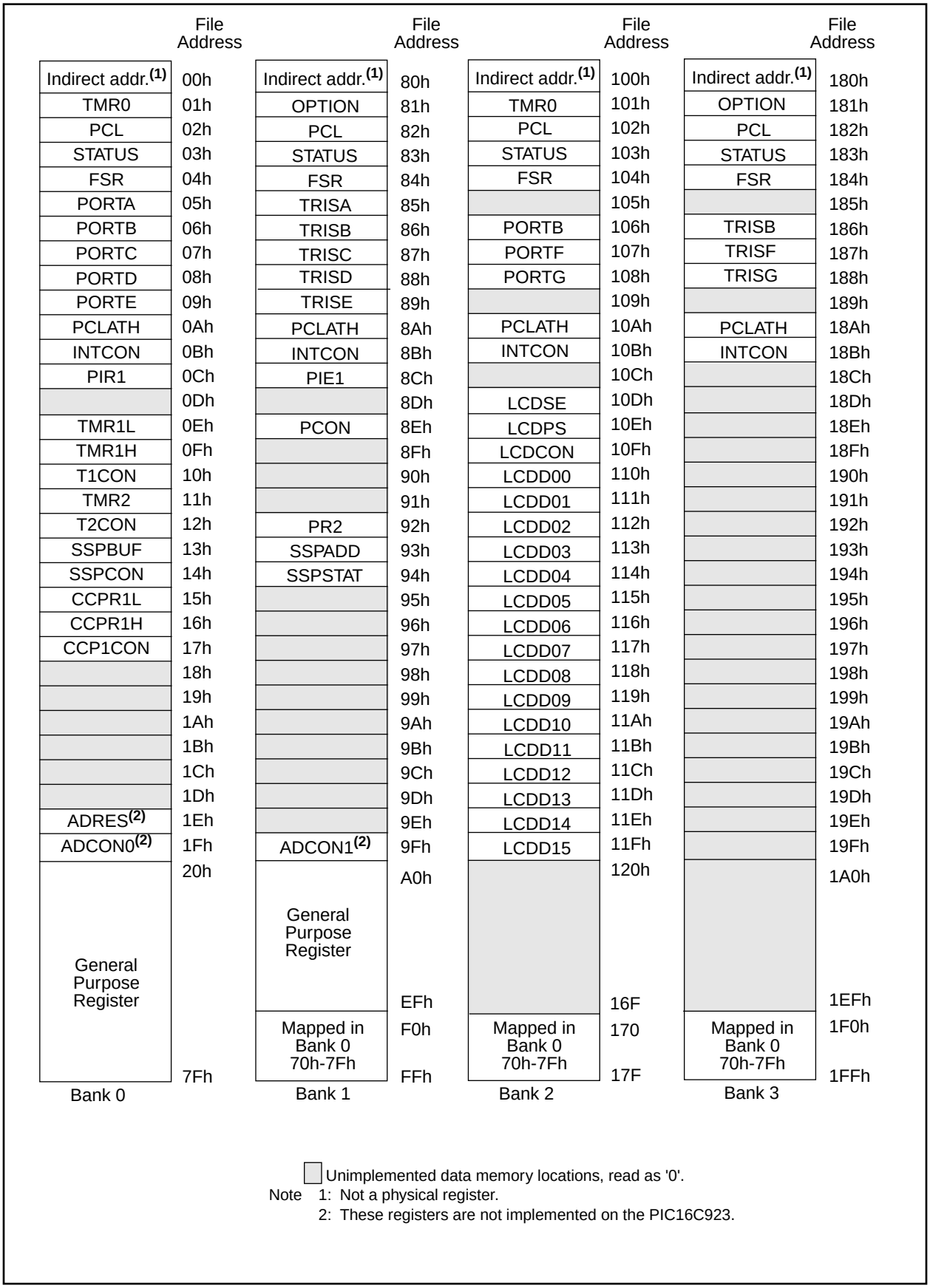


EXAMPLE 3-1: INSTRUCTION PIPELINE FLOW



All instructions are single cycle, except for any program branches. These take two cycles since the fetch instruction is “flushed” from the pipeline while the new instruction is being fetched and then executed.

FIGURE 4-2: REGISTER FILE MAP



Unimplemented data memory locations, read as '0'.

Note 1: Not a physical register.

2: These registers are not implemented on the PIC16C923.

TABLE 4-1: SPECIAL FUNCTION REGISTER SUMMARY (Cont.'d)

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on Power-on Reset	Value on all other resets
Bank 1											
80h	INDF	Addressing this location uses contents of FSR to address data memory (not a physical register)								0000 0000	0000 0000
81h	OPTION	$\overline{RBPU}$	INTEDG	T0CS	T0SE	PSA	PS2	PS1	PS0	1111 1111	1111 1111
82h	PCL	Program Counter's (PC) Least Significant Byte								0000 0000	0000 0000
83h	STATUS	IRP	RP1	RP0	$\overline{TO}$	$\overline{PD}$	Z	DC	C	0001 1xxx	000q quuu
84h	FSR	Indirect data memory address pointer								xxxx xxxx	uuuu uuuu
85h	TRISA	—	—	PORTA Data Direction Register						--11 1111	--11 1111
86h	TRISB	PORTB Data Direction Register								1111 1111	1111 1111
87h	TRISC	—	—	PORTC Data Direction Register						--11 1111	--11 1111
88h	TRISD	PORTD Data Direction Register								1111 1111	1111 1111
89h	TRISE	PORTE Data Direction Register								1111 1111	1111 1111
8Ah	PCLATH	—	—	—	Write Buffer for the upper 5 bits of the PC					---0 0000	---0 0000
8Bh	INTCON	GIE	PEIE	T0IE	INTE	RBIE	T0IF	INTF	RBIF	0000 000x	0000 000u
8Ch	PIE1	LCDIE	ADIE(2)	—	—	SSPIE	CCP1IE	TMR2IE	TMR1IE	00-- 0000	00-- 0000
8Dh	—	Unimplemented								—	—
8Eh	PCON	—	—	—	—	—	—	$\overline{POR}$	—	---- --0-	---- --u-
8Fh	—	Unimplemented								—	—
90h	—	Unimplemented								—	—
91h	—	Unimplemented								—	—
92h	PR2	Timer2 Period Register								1111 1111	1111 1111
93h	SSPADD	Synchronous Serial Port (I^2C mode) Address Register								0000 0000	0000 0000
94h	SSPSTAT	SMP	CKE	D/$\overline{A}$	P	S	R/$\overline{W}$	UA	BF	0000 0000	0000 0000
95h	—	Unimplemented								—	—
96h	—	Unimplemented								—	—
97h	—	Unimplemented								—	—
98h	—	Unimplemented								—	—
99h	—	Unimplemented								—	—
9Ah	—	Unimplemented								—	—
9Bh	—	Unimplemented								—	—
9Ch	—	Unimplemented								—	—
9Dh	—	Unimplemented								—	—
9Eh	—	Unimplemented								—	—
9Fh(1)	ADCON1	—	—	—	—	—	PCFG2	PCFG1	PCFG0	---- -000	---- -000

Legend: x = unknown, u = unchanged, q = value depends on condition, - = unimplemented read as '0', shaded locations are unimplemented, read as '0'.

Note 1: Registers ADRES, ADCON0, and ADCON1 are not implemented in the PIC16C923, read as '0'.
2: These bits are reserved on the PIC16C923, always maintain these bits clear.
3: These pixels do not display, but can be used as general purpose RAM.
4: PIC16C923 reset values for PORTA: `--xx xxxx` for a POR, and `--uu uuuu` for all other resets, PIC16C924 reset values for PORTA: `--0x 0000` when read.
5: Bit1 of ADCON0 is reserved on the PIC16C924, always maintain this bit clear.

5.3 PORTC and TRISC Register

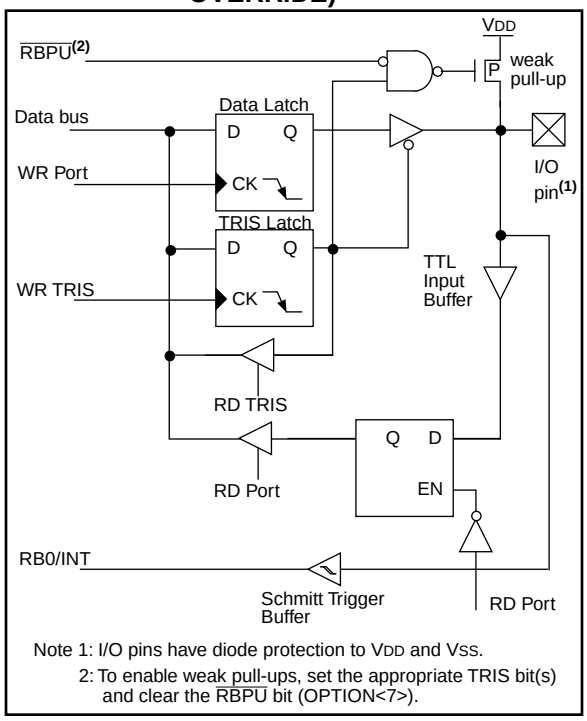
PORTC is an 6-bit bi-directional port. Each pin is individually configurable as an input or output through the TRISC register. PORTC is multiplexed with several peripheral functions (Table 5-5). PORTC pins have Schmitt Trigger input buffers.

When enabling peripheral functions, care should be taken in defining TRIS bits for each PORTC pin. Some peripherals override the TRIS bit to make a pin an output, while other peripherals override the TRIS bit to make a pin an input. Since the TRIS bit override is in effect while the peripheral is enabled, read-modify-write instructions (BSF, BCF, XORWF) with TRISC as destination should be avoided. The user should refer to the corresponding peripheral section for the correct TRIS bit settings.

EXAMPLE 5-3: INITIALIZING PORTC

```
BCF     STATUS,RP0 ; Select Bank0
BCF     STATUS,RP1
CLRF    PORTC      ; Initialize PORTC
BSF     STATUS,RP0 ;
MOVLW   0xCF       ; Value used to
                   ; initialize data
                   ; direction
MOVWF   TRISC      ; Set RC<3:0> as inputs
                   ; RC<5:4> as outputs
                   ; RC<7:6> always read 0
```

FIGURE 5-5: PORTC BLOCK DIAGRAM (PERIPHERAL OUTPUT OVERRIDE)



Note 1: I/O pins have diode protection to VDD and VSS.
2: To enable weak pull-ups, set the appropriate TRIS bit(s) and clear the RBPU bit (OPTION<7>).

TABLE 5-5: PORTC FUNCTIONS

Name	Bit#	Buffer Type	Function
RC0/T1OSO/T1CKI	bit0	ST	Input/output port pin or Timer1 oscillator output or Timer1 clock input
RC1/T1OSI	bit1	ST	Input/output port pin or Timer1 oscillator input
RC2/CCP1	bit2	ST	Input/output port pin or Capture input/Compare output/PWM output
RC3/SCK/SCL	bit3	ST	Input/output port pin or the synchronous serial clock for both SPI and I^2C modes.
RC4/SDI/SDA	bit4	ST	Input/output port pin or the SPI Data In (SPI mode) or data I/O (I^2C mode).
RC5/SDO	bit5	ST	Input/output port pin or Synchronous Serial Port data out

Legend: ST = Schmitt Trigger input

TABLE 5-6: SUMMARY OF REGISTERS ASSOCIATED WITH PORTC

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on Power-on Reset	Value on all other resets
07h	PORTC	—	—	RC5	RC4	RC3	RC2	RC1	RC0	--xx xxxx	--uu uuuu
87h	TRISC	—	—	PORTC Data Direction Control Register						--11 1111	--11 1111

Legend: x = unknown, u = unchanged, - = unimplemented, read as '0'. Shaded cells are not used by PORTC.

NOTES:

8.3 Timer1 Operation in Asynchronous Counter Mode

If control bit $\overline{\text{T1SYNC}}$ (T1CON<2>) is set, the external clock input is not synchronized. The timer continues to increment asynchronous to the internal phase clocks. The timer will continue to run during SLEEP and can generate an interrupt on overflow which will wake-up the processor. However, special precautions in software are needed to read-from or write-to the Timer1 register pair (TMR1H:TMR1L) (Section 8.3.2).

In asynchronous counter mode, Timer1 cannot be used as a time-base for capture or compare operations.

8.3.1 EXTERNAL CLOCK INPUT TIMING WITH UNSYNCHRONIZED CLOCK

If control bit $\overline{\text{T1SYNC}}$ is set, the timer will increment completely asynchronously. The input clock must meet certain minimum high time and low time requirements, as specified in timing parameters 45, 46, and 47.

8.3.2 READING AND WRITING TMR1 IN ASYNCHRONOUS COUNTER MODE

Reading TMR1H or TMR1L while the timer is running, from an external asynchronous clock, will ensure a valid read (taken care of in hardware). However, the user should keep in mind that reading the 16-bit timer in two 8-bit values itself poses certain problems since the timer may overflow between the reads.

For writes, it is recommended that the user simply stop the timer and write the desired values. A write contention may occur by writing to the timer registers while the register is incrementing. This may produce an unpredictable value in the timer register.

Reading the 16-bit value requires some care. Example 8-1 is an example routine to read the 16-bit timer value. This is useful if the timer cannot be stopped.

EXAMPLE 8-1: READING A 16-BIT FREE-RUNNING TIMER

```
; All interrupts are disabled
   MOVF   TMR1H, W  ;Read high byte
   MOVWF  TMPH      ;
   MOVF   TMR1L, W  ;Read low byte
   MOVWF  TMPL      ;
   MOVF   TMR1H, W  ;Read high byte
   SUBWF  TMPH,  W  ;Sub 1st read
                    ; with 2nd read
   BTFSC  STATUS,Z  ;Is result = 0
   GOTO   CONTINUE  ;Good 16-bit read
;
; TMR1L may have rolled over between the read
; of the high and low bytes. Reading the high
; and low bytes now will read a good value.
;
   MOVF   TMR1H, W  ;Read high byte
   MOVWF  TMPH      ;
   MOVF   TMR1L, W  ;Read low byte
   MOVWF  TMPL      ;
; Re-enable the Interrupt (if required)
CONTINUE            ;Continue with your code
```

8.4 Timer1 Oscillator

A crystal oscillator circuit is built in between pins T1OSI (input) and T1OSO (amplifier output). It is enabled by setting control bit T1OSCEN (T1CON<3>). The oscillator is a low power oscillator rated up to 200 kHz. It will continue to run during SLEEP. It is primarily intended for a 32 kHz crystal. Table 8-1 shows the capacitor selection for the Timer1 oscillator.

The Timer1 oscillator is identical to the LP oscillator. The user must provide a software time delay to ensure proper oscillator start-up.

TABLE 8-1: CAPACITOR SELECTION FOR THE TIMER1 OSCILLATOR

Osc Type	Freq	C1	C2
LP	32 kHz	33 pF	33 pF
	100 kHz	15 pF	15 pF
	200 kHz	15 pF	15 pF
These values are for design guidance only.			
Crystals Tested:			
32.768 kHz	Epson C-001R32.768K-A		± 20 PPM
100 kHz	Epson C-2 100.00 KC-P		± 20 PPM
200 kHz	STD XTL 200.000 kHz		± 20 PPM

Note 1: Higher capacitance increases the stability of oscillator but also increases the start-up time.

2: Since each resonator/crystal has its own characteristics, the user should consult the resonator/crystal manufacturer for appropriate values of external components.

Note:	The Timer2 postscaler (Section 9.0) is not used in the determination of the PWM frequency. The postscaler could be used to have a servo update rate at a different frequency than the PWM output.

10.3.2 PWM DUTY CYCLE

The PWM duty cycle is specified by writing to the CCPR1L register and to the CCP1CON<5:4> bits. Up to 10-bit resolution is available: the CCPR1L contains the eight MSbs and CCP1CON<5:4> contains the two LSbs. This 10-bit value is represented by CCPR1L:CCP1CON<5:4>. The following equation is used to calculate the PWM duty cycle in time:

PWM duty cycle = (CCPR1L:CCP1CON<5:4>) • Tosc • (TMR2 prescale value)

CCPR1L and CCP1CON<5:4> can be written to at any time, but the duty cycle value is not latched into CCPR1H until after a match between PR2 and TMR2 occurs (i.e., the period is complete). In PWM mode, CCPR1H is a read-only register.

The CCPR1H register and a 2-bit internal latch are used to double buffer the PWM duty cycle. This double buffering is essential for glitchless PWM operation.

When the CCPR1H and 2-bit latch match TMR2 concatenated with an internal 2-bit Q clock or 2 bits of the TMR2 prescaler, the CCP1 pin is cleared.

Maximum PWM resolution (bits) for a given PWM frequency:

$$= \frac{\log\left(\frac{FOSC}{FPWM}\right)}{\log(2)} \text{ bits}$$

Note:	If the PWM duty cycle value is longer than the PWM period the CCP1 pin will not be cleared.

EXAMPLE 10-2: PWM PERIOD AND DUTY CYCLE CALCULATION

Desired PWM frequency is 31.25 kHz,
Fosc = 8 MHz
TMR2 prescale = 1

$1/31.25\ kHz = [\ (PR2) + 1\ ] \cdot 4 \cdot 1/8\ MHz \cdot 1$

$32\ \mu s = [\ (PR2) + 1\ ] \cdot 4 \cdot 125\ ns \cdot 1$

$PR2 = 63$

Find the maximum resolution of the duty cycle that can be used with a 31.25 kHz frequency and 8 MHz oscillator:

$1/31.25\ kHz = 2^{PWM\ RESOLUTION} \cdot 1/8\ MHz \cdot 1$

$32\ \mu s = 2^{PWM\ RESOLUTION} \cdot 125\ ns \cdot 1$

$256 = 2^{PWM\ RESOLUTION}$

$\log(256) = (PWM\ Resolution) \cdot \log(2)$

$8.0 = PWM\ Resolution$

At most, an 8-bit resolution duty cycle can be obtained from a 31.25 kHz frequency and a 8 MHz oscillator, i.e., $0 \le CCPR1L:CCP1CON<5:4> \le 255$. Any value greater than 255 will result in a 100% duty cycle.

In order to achieve higher resolution, the PWM frequency must be decreased. In order to achieve higher PWM frequency, the resolution must be decreased.

Table 10-2 lists example PWM frequencies and resolutions for Fosc = 8 MHz. TMR2 prescaler and PR2 values are also shown.

10.3.3 SET-UP FOR PWM OPERATION

The following steps should be taken when configuring the CCP module for PWM operation:

1. Set the PWM period by writing to the PR2 register.
2. Set the PWM duty cycle by writing to the CCPR1L register and CCP1CON<5:4> bits.
3. Make the CCP1 pin an output by clearing the TRISC<2> bit.
4. Set the TMR2 prescale value and enable Timer2 by writing to T2CON.
5. Configure the CCP module for PWM operation.

TABLE 10-2: EXAMPLE PWM FREQUENCIES AND RESOLUTIONS AT 8 MHz

PWM Frequency	488 Hz	1.95 kHz	7.81 kHz	31.25 kHz	62.5 kHz	250 kHz
Timer Prescaler (1, 4, 16)	16	4	1	1	1	1
PR2 Value	0xFF	0xFF	0xFF	0x3F	0x1F	0x07
Maximum Resolution (bits)	10	10	10	8	7	5

The $\overline{SS}$ pin allows a synchronous slave mode. The SPI must be in slave mode (SSPCON<3:0> = 04h) and the TRISA<5> bit must be set for the synchronous slave mode to be enabled. When the $\overline{SS}$ pin is low, transmission and reception are enabled and the SDO pin is driven. When the $\overline{SS}$ pin goes high, the SDO pin is no longer driven, even if in the middle of a transmitted byte, and becomes a floating output. External pull-up/ pull-down resistors may be desirable, depending on the application.

Note:	When the SPI is in Slave Mode with $\overline{SS}$ pin control enabled, (SSPCON<3:0> = 0100) the SPI module will reset if the $\overline{SS}$ pin is set to V_{DD}.
Note:	If the SPI is used in Slave Mode with CKE = '1', then the $\overline{SS}$ pin control must be enabled.

To emulate two-wire communication, the SDO pin can be connected to the SDI pin. When the SPI needs to operate as a receiver the SDO pin can be configured as an input. This disables transmissions from the SDO. The SDI can always be left as an input (SDI function) since it cannot create a bus conflict.

FIGURE 11-5: SPI MODE TIMING, MASTER MODE

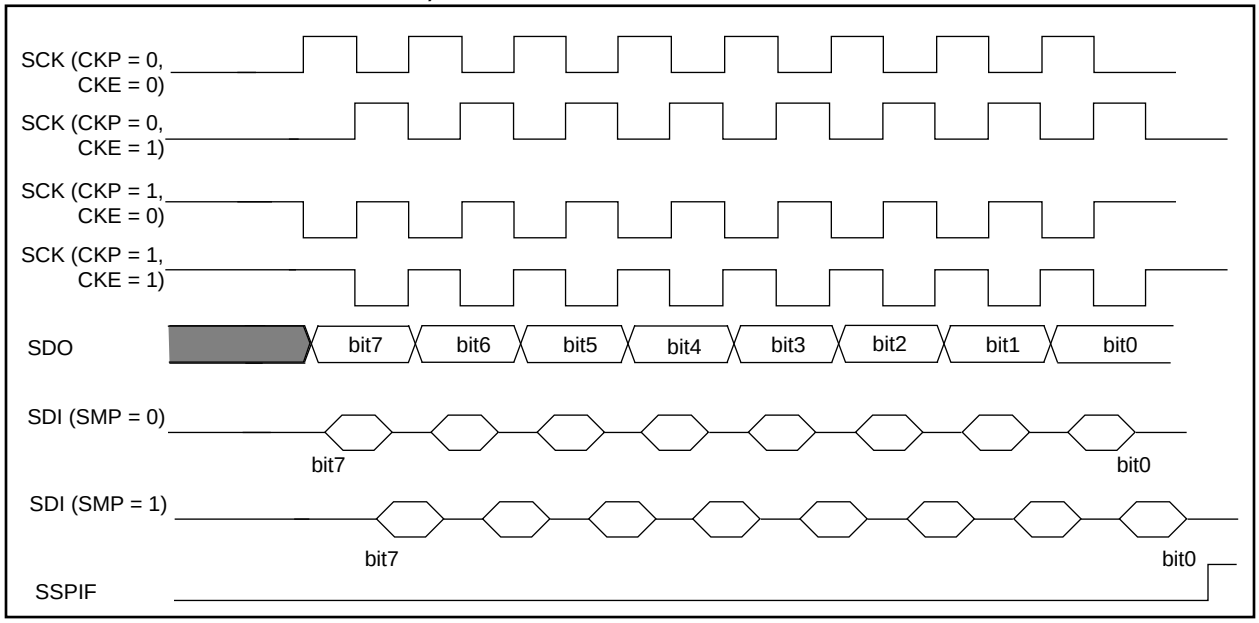
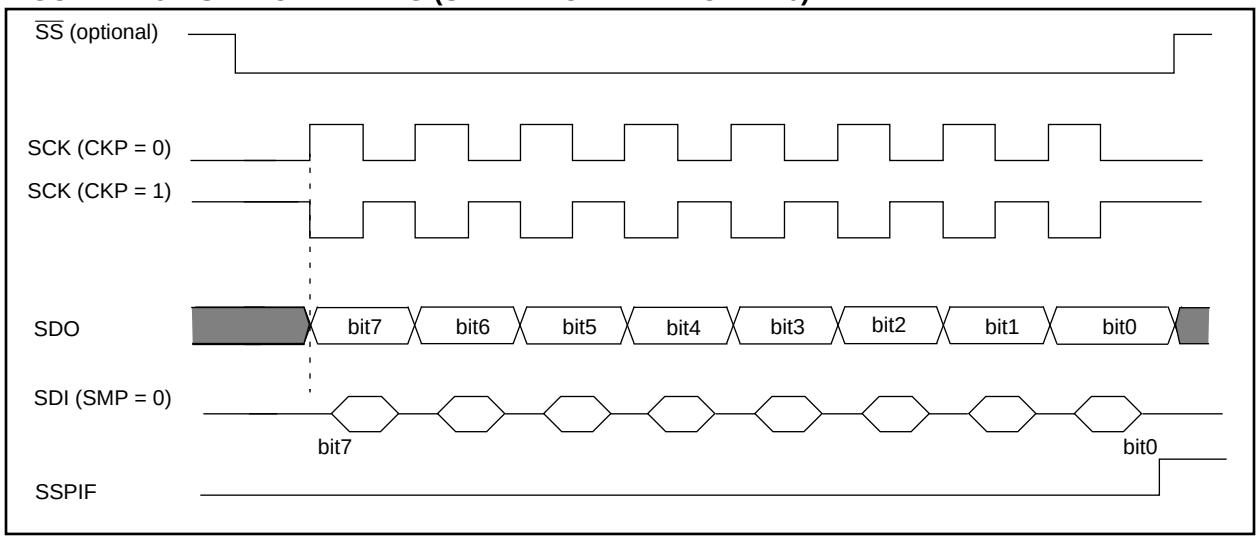


FIGURE 11-6: SPI MODE TIMING (SLAVE MODE WITH CKE = 0)



11.2 I²C™ Overview

This section provides an overview of the Inter-Integrated Circuit (I^2C) bus, with Section 11.3 discussing the operation of the SSP module in I^2C mode.

The I^2C bus is a two-wire serial interface developed by the Philips Corporation. The original specification, or standard mode, was for data transfers of up to 100 Kbps. An enhanced specification, or fast mode is not supported. This device will communicate with fast mode devices if attached to the same bus.

The I^2C interface employs a comprehensive protocol to ensure reliable transmission and reception of data. When transmitting data, one device is the "master" which initiates transfer on the bus and generates the clock signals to permit that transfer, while the other device(s) acts as the "slave." All portions of the slave protocol are implemented in the SSP module's hardware, except general call support, while portions of the master protocol need to be addressed in the PIC16CXXX software. Table 11-2 defines some of the I^2C bus terminology. For additional information on the I^2C interface specification, refer to the Philips document "*The I^2C bus and how to use it.*" #939839340011, which can be obtained from the Philips Corporation.

In the I^2C interface protocol each device has an address. When a master wishes to initiate a data transfer, it first transmits the address of the device that it wishes to "talk" to. All devices "listen" to see if this is their address. Within this address, a bit specifies if the master wishes to read-from/write-to the slave device. The master and slave are always in opposite modes (transmitter/receiver) of operation during a data transfer. That is they can be thought of as operating in either of these two relations:

- Master-transmitter and Slave-receiver
- Slave-transmitter and Master-receiver

In both cases the master generates the clock signal.

The output stages of the clock (SCL) and data (SDA) lines must have an open-drain or open-collector in order to perform the wired-AND function of the bus. External pull-up resistors are used to ensure a high level when no device is pulling the line down. The number of devices that may be attached to the I^2C bus is limited only by the maximum bus loading specification of 400 pF.

11.2.1 INITIATING AND TERMINATING DATA TRANSFER

During times of no data transfer (idle time), both the clock line (SCL) and the data line (SDA) are pulled high through the external pull-up resistors. The START and STOP conditions determine the start and stop of data transmission. The START condition is defined as a high to low transition of the SDA when the SCL is high. The STOP condition is defined as a low to high transition of the SDA when the SCL is high. Figure 11-8 shows the START and STOP conditions. The master generates these conditions for starting and terminating data transfer. Due to the definition of the START and STOP conditions, when data is being transmitted, the SDA line can only change state when the SCL line is low.

FIGURE 11-8: START AND STOP CONDITIONS

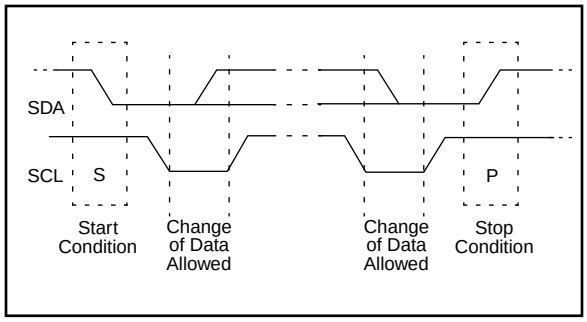


TABLE 11-2: I^2C BUS TERMINOLOGY

Term	Description
Transmitter	The device that sends the data to the bus.
Receiver	The device that receives the data from the bus.
Master	The device which initiates the transfer, generates the clock and terminates the transfer.
Slave	The device addressed by a master.
Multi-master	More than one master device in a system. These masters can attempt to control the bus at the same time without corrupting the message.
Arbitration	Procedure that ensures that only one of the master devices will control the bus. This ensure that the transfer data does not get corrupted.
Synchronization	Procedure where the clock signals of two or more devices are synchronized.

12.2 Selecting the A/D Conversion Clock

The A/D conversion time per bit is defined as TAD. The A/D conversion requires 9.5 TAD per 8-bit conversion. The source of the A/D conversion clock is software selected. The four possible options for TAD are:

- 2TOSC
- 8TOSC
- 32TOSC
- Internal RC oscillator

For correct A/D conversions, the A/D conversion clock (TAD) must be selected to ensure a minimum TAD time of 1.6 µs.

Table 12-1 shows the resultant TAD times derived from the device operating frequencies and the A/D clock source selected.

12.3 Configuring Analog Port Pins

The ADCON1 and TRISA registers control the operation of the A/D port pins. The port pins that are desired as analog inputs must have their corresponding TRIS bits set (input). If the TRIS bit is cleared (output), the digital output level (VOH or VOL) will be converted.

The A/D operation is independent of the state of the CHS2:CHS0 bits and the TRIS bits.

Note 1: When reading the port register, all pins configured as analog inputs will read as cleared (a low level). Pins configured as digital inputs, will convert an analog input. Analog levels on a digitally configured input will not affect the conversion accuracy.

Note 2: Analog levels on any pin that is defined as a digital input (including the AN4:AN0 pins), may cause the input buffer to consume current that is out of the devices specification.

TABLE 12-1: TAD vs. DEVICE OPERATING FREQUENCIES

A/D Clock Source (TAD)		Device Frequency			
Operation	**ADCS1:ADCS0**	**8 MHz**	**5 MHz**	**1.25 MHz**	**333.33 kHz**
2TOSC	00	250 ns(2)	400 ns(2)	1.6 µs	6 µs
8TOSC	01	1 µs	1.6 µs	6.4 µs	24 µs(3)
32TOSC	10	4 µs	6.4 µs	25.6 µs(3)	96 µs(3)
RC	11	2 - 6 µs(1,4)	2 - 6 µs(1,4)	2 - 6 µs(1,4)	2 - 6 µs(1)

Legend: Shaded cells are outside of recommended range.

Note 1: The RC source has a typical TAD time of 4 µs.

2: These values violate the minimum required TAD time.

3: For faster conversion times, the selection of another clock source is recommended.

4: When derived frequency is greater than 1 MHz, the RC A/D conversion clock source is recommended for sleep mode only

5: For extended voltage devices (LC), please refer to the electrical specifications section.

13.3 Pixel Control

13.3.1 LCDD (PIXEL DATA) REGISTERS

The pixel registers contain bits which define the state of each pixel. Each bit defines one unique pixel.

Table 13-4 shows the correlation of each bit in the LCDD registers to the respective common and segment signals.

Any LCD pixel location not being used for display can be used as general purpose RAM.

FIGURE 13-10:GENERIC LCDD REGISTER LAYOUT

R/W-x	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x
SEGs COMc	SEGs COMc	SEGs COMc	SEGs COMc	SEGs COMc	SEGs COMc	SEGs COMc	SEGs COMc
bit7							bit0

R =Readable bit
W =Writable bit
U =Unimplemented bit, Read as '0'
-n =Value at POR reset

bit 7-0: **SEGsCOMc**: Pixel Data Bit for segment s and common c
1 = Pixel on (dark)
0 = Pixel off (clear)

14.4 Power-on Reset (POR), Power-up Timer (PWRT) and Oscillator Start-up Timer (OST)

14.4.1 POWER-ON RESET (POR)

A Power-on Reset pulse is generated on-chip when VDD rise is detected (in the range of 1.5V - 2.1V). To take advantage of the POR, just tie the $\overline{MCLR}$ pin directly (or through a resistor) to VDD. This will eliminate external RC components usually needed to create a Power-on Reset. A maximum rise time for VDD is specified. See Electrical Specifications for details.

When the device starts normal operation (exits the reset condition), device operating parameters (voltage, frequency, temperature, ...) must be met to ensure operation. If these conditions are not met, the device must be held in reset until the operating conditions are met.

For additional information, refer to Application Note AN607, "*Power-up Trouble Shooting*."

14.4.2 POWER-UP TIMER (PWRT)

The Power-up Timer provides a fixed 72 ms nominal time-out on power-up only, from the POR. The Power-up Timer operates on an internal RC oscillator. The chip is kept in reset as long as the PWRT is active. The PWRT's time delay allows VDD to rise to an acceptable level. A configuration bit is provided to enable/disable the PWRT.

The power-up time delay will vary from chip to chip due to VDD, temperature, and process variation. See DC parameters for details.

14.4.3 OSCILLATOR START-UP TIMER (OST)

The Oscillator Start-up Timer (OST) provides 1024 oscillator cycle (from OSC1 input) delay after the PWRT delay is over. This ensures that the crystal oscillator or resonator has started and stabilized.

The OST time-out is invoked only for XT, LP and HS modes and only on Power-on Reset or wake-up from SLEEP.

14.4.4 TIME-OUT SEQUENCE

On power-up the time-out sequence is as follows: First PWRT time-out is invoked after the POR time delay has expired. Then OST is activated. The total time-out will vary based on oscillator configuration and the status of the PWRT. For example, in RC mode with the PWRT disabled, there will be no time-out at all. Figure 14-8, Figure 14-9, and Figure 14-10 depict time-out sequences on power-up.

Since the time-outs occur from the POR pulse, if $\overline{MCLR}$ is kept low long enough, the time-outs will expire. Then bringing $\overline{MCLR}$ high will begin execution immediately (Figure 14-9). This is useful for testing purposes or to synchronize more than one PIC16CXXX device operating in parallel.

Table 14-5 shows the reset conditions for some special function registers, while Table 14-6 shows the reset conditions for all the registers.

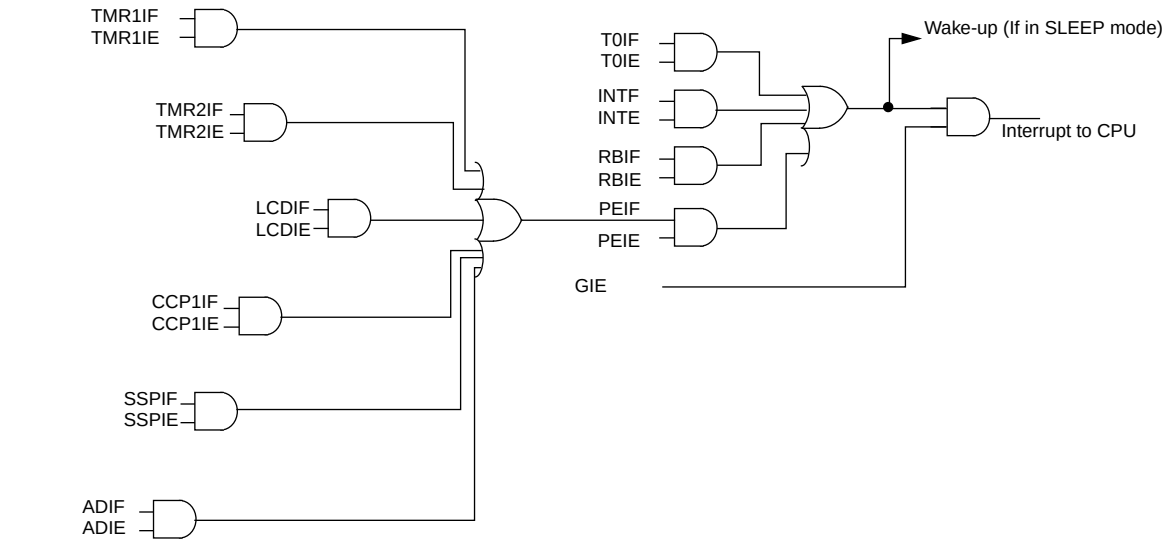
14.4.5 POWER CONTROL/STATUS REGISTER (PCON)

Bit1 is Power-on Reset Status bit $\overline{POR}$. It is cleared on a Power-on Reset and unaffected otherwise. The user must set this bit following a Power-on Reset.

TABLE 14-3: TIME-OUT IN VARIOUS SITUATIONS

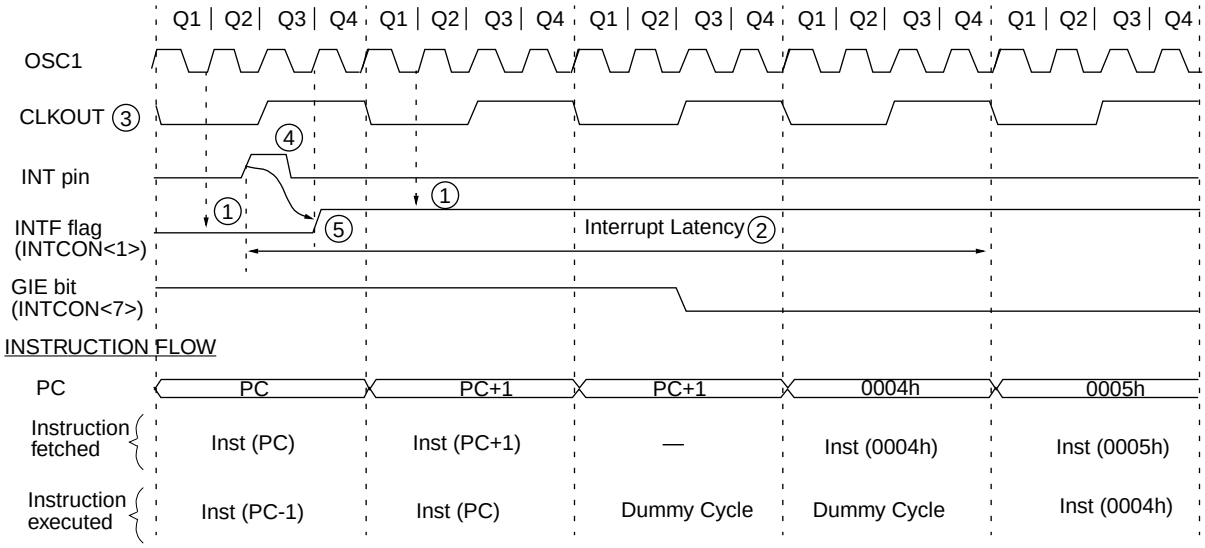
Oscillator Configuration	Power-up		Wake-up from SLEEP
	$\overline{PWRTE}$ = 1	$\overline{PWRTE}$ = 0	
XT, HS, LP	1024TOSC	72 ms + 1024TOSC	1024 TOSC
RC	—	72 ms	—

FIGURE 14-14:INTERRUPT LOGIC



The A/D module interrupt is implemented on the PIC16C924 only.

FIGURE 14-15:INT PIN INTERRUPT TIMING



Note 1: INTF flag is sampled here (every Q1).
2: Interrupt latency = 3-4 T_{CY} where T_{CY} = instruction cycle time.
Latency is the same whether Inst (PC) is a single cycle or a 2-cycle instruction.
3: CLKOUT is available only in RC oscillator mode.
4: For minimum width of INT pulse, refer to AC specs.
5: INTF can be set anytime during the Q4-Q1 cycles.

SUBWF **Subtract W from f**

Syntax: [*label*] SUBWF f,d

Operands: $0 \le f \le 127$
$d \in [0,1]$

Operation: (f) - (W) → (destination)

Status Affected: C, DC, Z

Encoding:

00	0010	dfff	ffff

Description: Subtract (2's complement method) W register from register 'f'. If 'd' is 0 the result is stored in the W register. If 'd' is 1 the result is stored back in register 'f'.

Words: 1

Cycles: 1

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	Read register 'f'	Process data	Write to destination

Example 1: SUBWF REG1,1

Before Instruction
REG1 = 3
W = 2
C = ?
Z = ?

After Instruction
REG1 = 1
W = 2
C = 1; result is positive
Z = 0

Example 2: Before Instruction
REG1 = 2
W = 2
C = ?
Z = ?

After Instruction
REG1 = 0
W = 2
C = 1; result is zero
Z = 1

Example 3: Before Instruction
REG1 = 1
W = 2
C = ?
Z = ?

After Instruction
REG1 = 0xFF
W = 2
C = 0; result is negative
Z = 0

SWAPF **Swap Nibbles in f**

Syntax: [*label*] SWAPF f,d

Operands: $0 \le f \le 127$
$d \in [0,1]$

Operation: (f<3:0>) → (destination<7:4>),
(f<7:4>) → (destination<3:0>)

Status Affected: None

Encoding:

00	1110	dfff	ffff

Description: The upper and lower nibbles of register 'f' are exchanged. If 'd' is 0 the result is placed in W register. If 'd' is 1 the result is placed in register 'f'.

Words: 1

Cycles: 1

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	Read register 'f'	Process data	Write to destination

Example SWAPF REG, 0

Before Instruction
REG1 = 0xA5

After Instruction
REG1 = 0xA5
W = 0x5A

TRIS **Load TRIS Register**

Syntax: [*label*] TRIS f

Operands: $5 \le f \le 7$

Operation: (W) → TRIS register f;

Status Affected: None

Encoding:

00	0000	0110	0fff

Description: The instruction is supported for code compatibility with the PIC16C5X products. Since TRIS registers are readable and writable, the user can directly address them.

Words: 1

Cycles: 1

Example

To maintain upward compatibility with future PIC16CXX products, do not use this instruction.

FIGURE 17-2: LOAD CONDITIONS

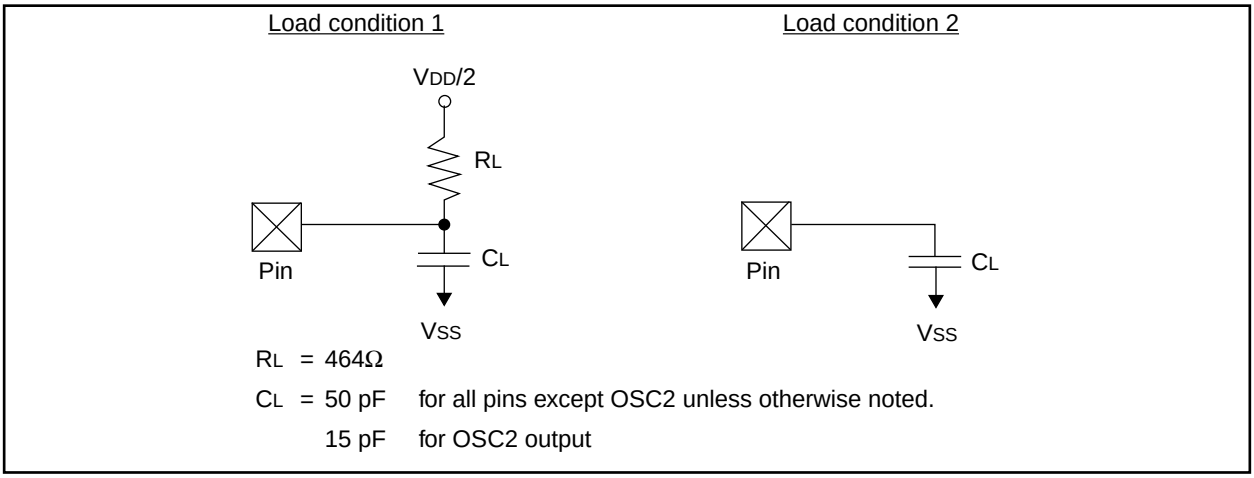
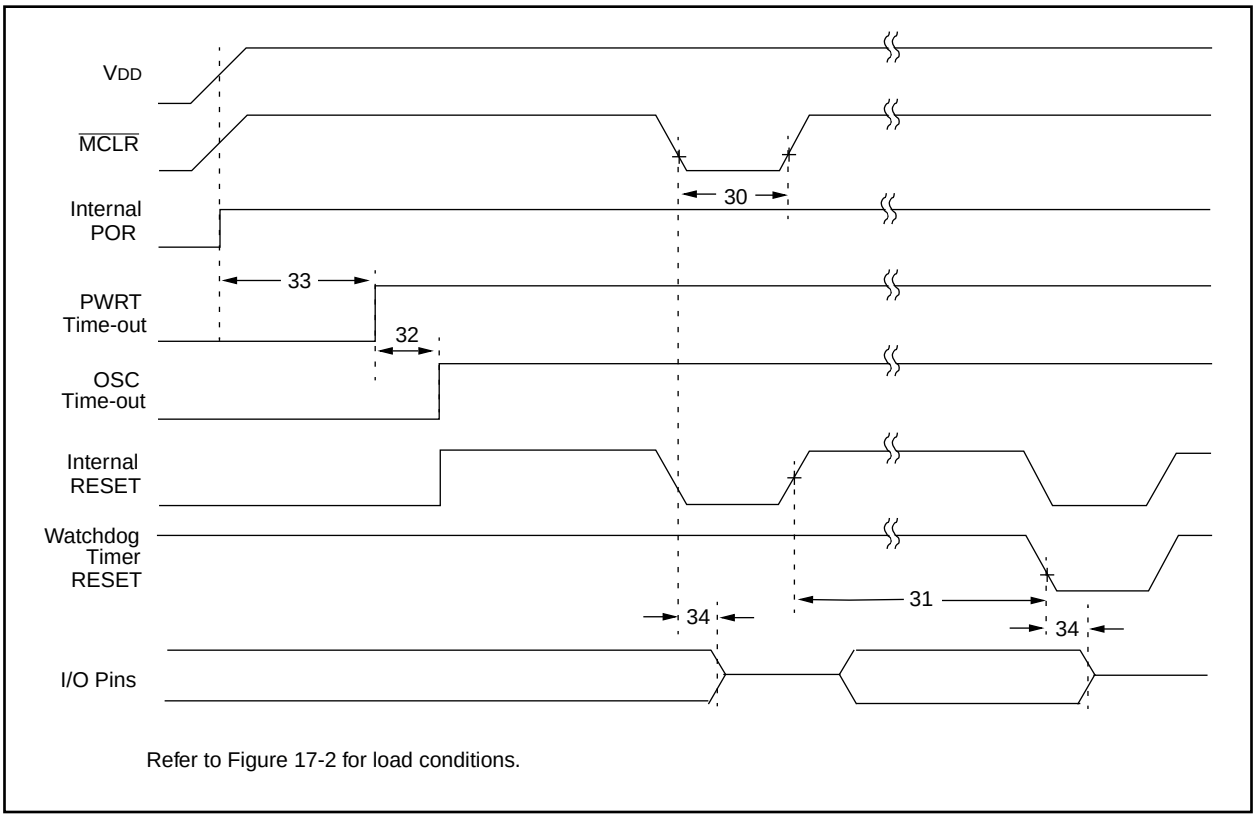


FIGURE 17-5: RESET, WATCHDOG TIMER, OSCILLATOR START-UP TIMER AND POWER-UP TIMER TIMING



Refer to Figure 17-2 for load conditions.

TABLE 17-6: RESET, WATCHDOG TIMER, OSCILLATOR START-UP TIMER AND POWER-UP TIMER REQUIREMENTS

Parameter No.	Sym	Characteristic	Min	Typ†	Max	Units	Conditions
30	TmcL	$\overline{MCLR}$ Pulse Width (low)	2	—	—	µs	
31*	Twdt	Watchdog Timer Time-out Period (No Prescaler)	7	18	33	ms	VDD = 5V, -40˚C to +85˚C
32	Tost	Oscillation Start-up Timer Period	—	1024TOSC	—	—	TOSC = OSC1 period
33*	Tpwrt	Power-up Timer Period	28	72	132	ms	VDD = 5V, -40˚C to +85˚C
34	TIOZ	I/O Hi-impedance from $\overline{MCLR}$ Low or Watchdog Timer Reset	—	—	2.1	µs	

* These parameters are characterized but not tested.

† Data in "Typ" column is at 5V, 25˚C unless otherwise stated. These parameters are for design guidance only and are not tested.

TABLE 17-9: SPI MODE REQUIREMENTS

Param No.	Sym	Characteristic		Min	Typ†	Max	Units	Conditions
70*	TssL2scH, TssL2scL	$\overline{SS}\downarrow$ to SCK$\downarrow$ or SCK$\uparrow$ input		TCY	—	—	ns	
71*	TscH	SCK input high time (slave mode)	Continuous	1.25TCY + 30	—	—	ns	
71A*			Single Byte	40	—	—	ns	
72*	TscL	SCK input low time (slave mode)	Continuous	1.25TCY + 30	—	—	ns	
72A*			Single Byte	40				
73*	TdiV2scH, TdiV2scL	Setup time of SDI data input to SCK edge		50	—	—	ns	
74*	TscH2diL, TscL2diL	Hold time of SDI data input to SCK edge		50	—	—	ns	
75*	TdoR	SDO data output rise time		—	10	25	ns	
76*	TdoF	SDO data output fall time		—	10	25	ns	
77*	TssH2doZ	$\overline{SS}\uparrow$ to SDO output hi-impedance		10	—	50	ns	
78*	TscR	SCK output rise time (master mode)		—	10	25	ns	
79*	TscF	SCK output fall time (master mode)		—	10	25	ns	
80*	TscH2doV, TscL2doV	SDO data output valid after SCK edge		—	—	50	ns	
81*	TdoV2scH, TdoV2scL	SDO data output setup to SCK edge		TCY	—	—	ns	
82*	TssL2doV	SDO data output valid after $\overline{SS}\downarrow$ edge		—	—	50	ns	
83*	TscH2ssH, TscL2ssH	$\overline{SS}\uparrow$ after SCK edge		1.5TCY + 40	—	—	ns	
84*	Tb2b	Delay between consecutive bytes		1.5TCY + 40	—	—	ns	

* Characterized but not tested.

† Data in "Typ" column is at 5V, 25˚C unless otherwise stated. These parameters are for design guidance only and are not tested.

18.0 DC AND AC CHARACTERISTICS GRAPHS AND TABLES

The graphs and tables provided in this section are for design guidance and are not tested or guaranteed.

In some graphs or tables the data presented are outside specified operating range (i.e., outside specified VDD range). This is for information only and devices are guaranteed to operate properly only within the specified range.

Note:	The data presented in this section is a statistical summary of data collected on units from different lots over a period of time and process characterization samples. 'Typical' represents the mean of the distribution at, 25°C, while 'max' or 'min' represents (mean +3σ) and (mean -3σ) respectively where σ is standard deviation.

FIGURE 18-1: TYPICAL IPD vs. VDD (WDT DISABLED, RC MODE @ 25°C)

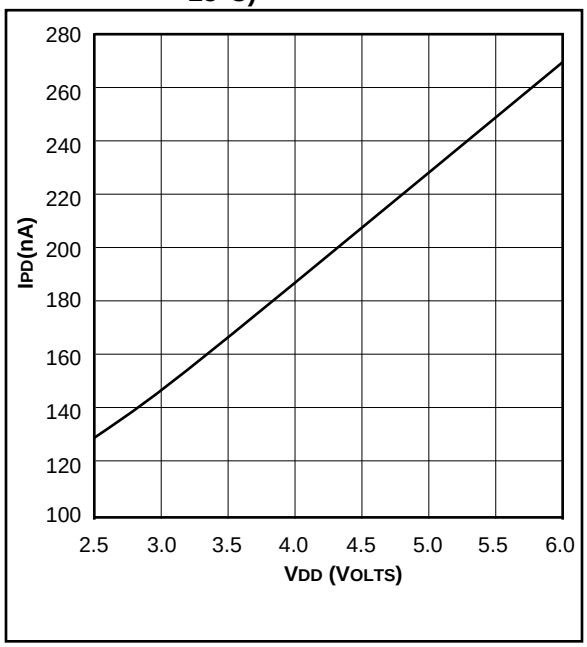


FIGURE 18-2: MAXIMUM IPD vs. VDD (WDT DISABLED, RC MODE -40°C TO +85°C)

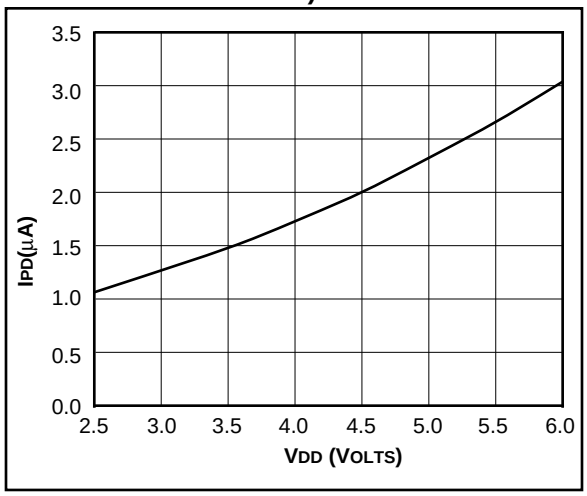


FIGURE 18-3: TYPICAL IPD vs. VDD (WDT ENABLED, RC MODE @ 25°C)

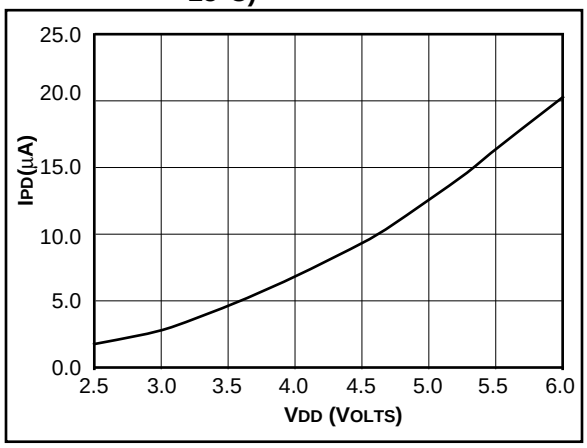


FIGURE 18-4: MAXIMUM IPD vs. VDD (WDT ENABLED, RC MODE -40°C TO +85°C)