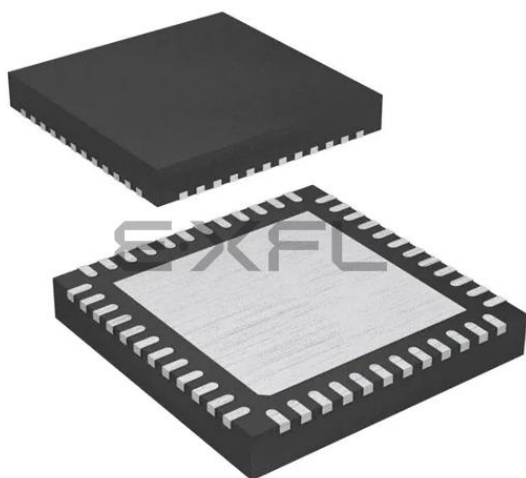




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Details

Product Status	Not For New Designs
Core Processor	S08
Core Size	8-Bit
Speed	40MHz
Connectivity	I ² C, SCI, SPI
Peripherals	LVD, POR, PWM, WDT
Number of I/O	39
Program Memory Size	60KB (60K x 8)
Program Memory Type	FLASH
EEPROM Size	-
RAM Size	4K x 8
Voltage - Supply (Vcc/Vdd)	1.8V ~ 3.6V
Data Converters	A/D 8x10b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	48-VFQFN Exposed Pad
Supplier Device Package	48-QFN-EP (7x7)
Purchase URL	https://www.e-xfl.com/product-detail/nxp-semiconductors/mc9s08gt60acfder



into stop2, the states of the I/O pins are latched. The states are held while in stop2 mode and after exiting stop2 mode until a 1 is written to PPDACK in SPMSC2.

Exit from stop2 is performed by asserting either of the wake-up pins: $\overline{\text{RESET}}$ or IRQ, or by an RTI interrupt. IRQ is always an active low input when the MCU is in stop2, regardless of how it was configured before entering stop2.

Upon wake-up from stop2 mode, the MCU will start up as from a power-on reset (POR) except pin states remain latched. The CPU will take the reset vector. The system and all peripherals will be in their default reset states and must be initialized.

After waking up from stop2, the PPDF bit in SPMSC2 is set. This flag may be used to direct user code to go to a stop2 recovery routine. PPDF remains set and the I/O pin states remain latched until a 1 is written to PPDACK in SPMSC2.

To maintain I/O state for pins that were configured as general-purpose I/O, the user must restore the contents of the I/O port registers, which have been saved in RAM, to the port registers before writing to the PPDACK bit. If the port registers are not restored from RAM before writing to PPDACK, then the register bits will assume their reset states when the I/O pin latches are opened and the I/O pins will switch to their reset states.

For pins that were configured as peripheral I/O, the user must reconfigure the peripheral module that interfaces to the pin before writing to the PPDACK bit. If the peripheral module is not enabled before writing to PPDACK, the pins will be controlled by their associated port control registers when the I/O latches are opened.

3.6.3 Stop3 Mode

Upon entering the stop3 mode, all of the clocks in the MCU, including the oscillator itself, are halted. The ICG is turned off, the ATD is disabled, and the voltage regulator is put in standby. The states of all of the internal registers and logic, as well as the RAM content, are maintained. The I/O pin states are not latched at the pin as in stop2. Instead they are maintained by virtue of the states of the internal logic driving the pins being maintained.

Exit from stop3 is performed by asserting $\overline{\text{RESET}}$, an asynchronous interrupt pin, or through the real-time interrupt. The asynchronous interrupt pins are the IRQ or KBI pins.

If stop3 is exited by means of the $\overline{\text{RESET}}$ pin, then the MCU will be reset and operation will resume after taking the reset vector. Exit by means of an asynchronous interrupt or the real-time interrupt will result in the MCU taking the appropriate interrupt vector.

A separate self-clocked source (≈ 1 kHz) for the real-time interrupt allows a wakeup from stop2 or stop3 mode with no external components. When RTIS2:RTIS1:RTIS0 = 0:0:0, the real-time interrupt function and this 1-kHz source are disabled. Power consumption is lower when the 1-kHz source is disabled, but in that case the real-time interrupt cannot wake the MCU from stop.

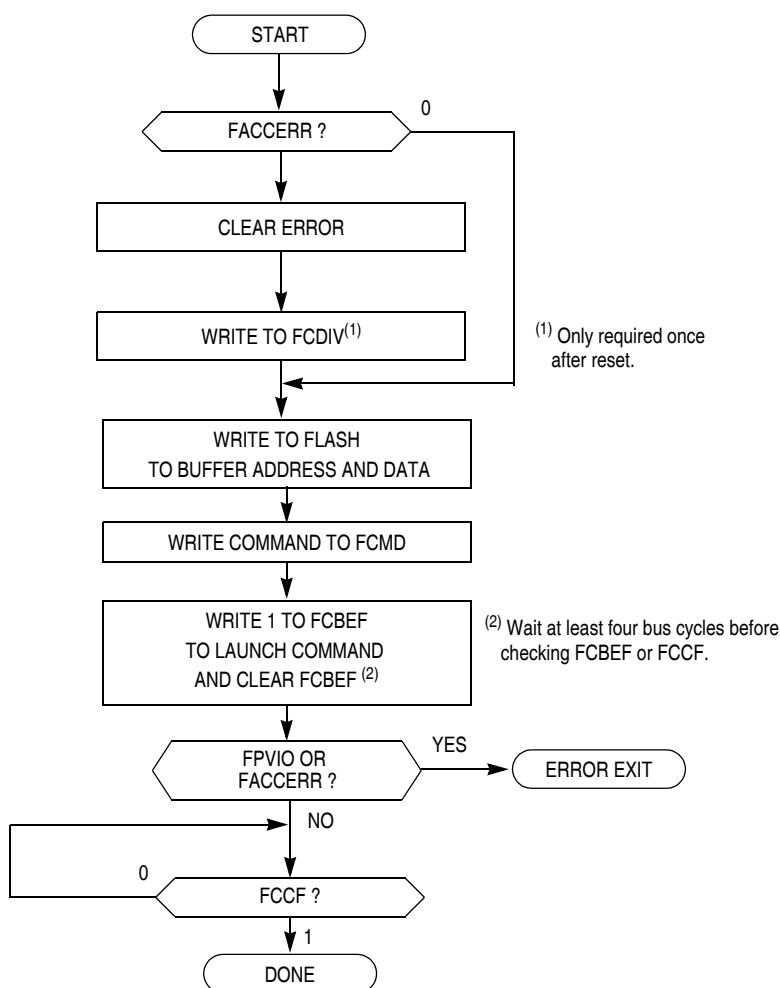


Figure 4-2. Flash Program and Erase Flowchart

4.4.4 Burst Program Execution

The burst program command is used to program sequential bytes of data in less time than would be required using the standard program command. This is possible because the high voltage to the flash array does not need to be disabled between program operations. Ordinarily, when a program or erase command is issued, an internal charge pump associated with the flash memory must be enabled to supply high voltage to the array. Upon completion of the command, the charge pump is turned off. When a burst program command is issued, the charge pump is enabled and then remains enabled after completion of the burst program operation if the following two conditions are met:

1. The next burst program command has been queued before the current program operation has completed.
2. The next sequential address selects a byte on the same physical row as the current byte being programmed. A row of flash memory consists of 64 bytes. A byte within a row is selected by addresses A5 through A0. A new row begins when addresses A5 through A0 are all zero.

The first byte of a series of sequential bytes being programmed in burst mode will take the same amount of time to program as a byte programmed in standard mode. Subsequent bytes will program in the burst

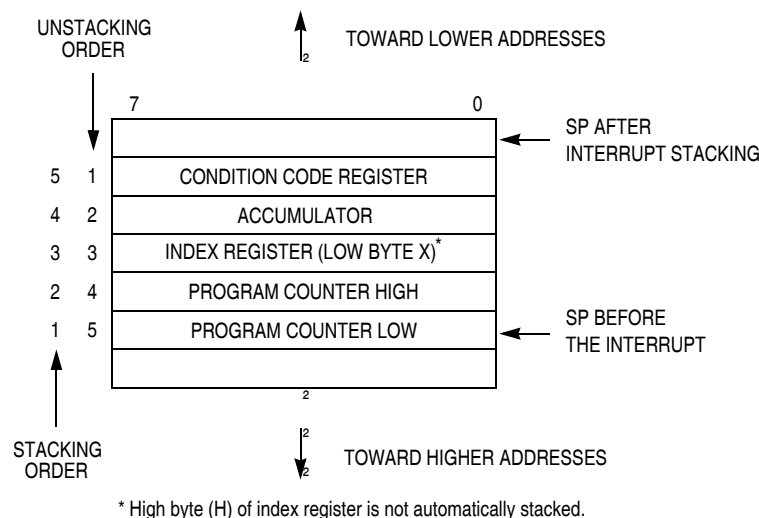


Figure 5-1. Interrupt Stack Frame

When an RTI instruction is executed, these values are recovered from the stack in reverse order. As part of the RTI sequence, the CPU fills the instruction pipeline by reading three bytes of program information, starting from the PC address just recovered from the stack.

The status flag causing the interrupt must be acknowledged (cleared) before returning from the ISR. Typically, the flag should be cleared at the beginning of the ISR so that if another interrupt is generated by this same source, it will be registered so it can be serviced after completion of the current ISR.

5.5.2 External Interrupt Request (IRQ) Pin

External interrupts are managed by the IRQSC status and control register. When the IRQ function is enabled, synchronous logic monitors the pin for edge-only or edge-and-level events. When the MCU is in stop mode and system clocks are shut down, a separate asynchronous path is used so the IRQ (if enabled) can wake the MCU.

5.5.2.1 Pin Configuration Options

The IRQ pin enable (IRQPE) control bit in the IRQSC register must be 1 for the IRQ pin to act as the interrupt request (IRQ) input. When the pin is configured as an IRQ input, the user can choose the polarity of edges or levels detected (IRQEDG), whether the pin detects edges-only or edges and levels (IRQMOD), and whether an event causes an interrupt or only sets the IRQF flag (which can be polled by software).

When the IRQ pin is configured to detect rising edges, an optional pulldown resistor is available rather than a pullup resistor. BIH and BIL instructions may be used to detect the level on the IRQ pin when the pin is configured to act as the IRQ input.

NOTE

The voltage measured on the pulled up IRQ pin may be as low as $V_{DD} - 0.7$ V. The internal gates connected to this pin are pulled all the way to V_{DD} . All other pins with enabled pullup resistors will have an unloaded measurement of V_{DD} .

7.3.3 FLL Engaged, Internal Clock (FEI) Mode

FLL engaged internal (FEI) is entered when any of the following conditions occur:

- CLKS bits are written to 01
- The DCO clock stabilizes ($DCOS = 1$) while in SCM upon exiting the off state with $CLKS = 01$

In FLL engaged internal mode, the reference clock is derived from the internal reference clock ICGIRCLK, and the FLL loop will attempt to lock the ICGDCLK frequency to the desired value, as selected by the MFD bits.

7.3.3.1 FLL Engaged Internal Unlocked

FEI unlocked is a temporary state that is entered when FEI is entered and the count error (Δn) output from the subtractor is greater than the maximum n_{unlock} or less than the minimum n_{unlock} , as required by the lock detector to detect the unlock condition.

The ICG will remain in this state while the count error (Δn) is greater than the maximum n_{lock} or less than the minimum n_{lock} , as required by the lock detector to detect the lock condition.

In this state the output clock signal ICGOUT frequency is given by $f_{ICGDCLK} / R$.

7.3.3.2 FLL Engaged Internal Locked

FLL engaged internal locked is entered from FEI unlocked when the count error (Δn), which comes from the subtractor, is less than n_{lock} (max) and greater than n_{lock} (min) for a given number of samples, as required by the lock detector to detect the lock condition. The output clock signal ICGOUT frequency is given by $f_{ICGDCLK} / R$. In FEI locked, the filter value is only updated once every four comparison cycles. The update made is an average of the error measurements taken in the four previous comparisons.

7.3.4 FLL Bypassed, External Clock (FBE) Mode

FLL bypassed external (FBE) is entered when any of the following conditions occur:

- From SCM when $CLKS = 10$ and ERCS is high
- When $CLKS = 10$, $ERCS = 1$ upon entering off mode, and off is then exited
- From FLL engaged external mode if a loss of DCO clock occurs and the external reference is still valid (both $LOCS = 1$ and $ERCS = 1$)

In this state, the DCO and IRG are off and the reference clock is derived from the external reference clock, ICGERCLK. The output clock signal ICGOUT frequency is given by $f_{ICGERCLK} / R$. If an external clock source is used ($REFS = 0$), then the input frequency on the EXTAL pin can be anywhere in the range 0 MHz to 40 MHz. If a crystal or resonator is used ($REFS = 1$), then frequency range is either low for $RANGE = 0$ or high for $RANGE = 1$.

7.3.5 FLL Engaged, External Clock (FEE) Mode

The FLL engaged external (FEE) mode is entered when any of the following conditions occur:

- $CLKS = 11$ and ERCS and DCOS are both high.
- The DCO stabilizes ($DCOS = 1$) while in SCM upon exiting the off state with $CLKS = 11$.

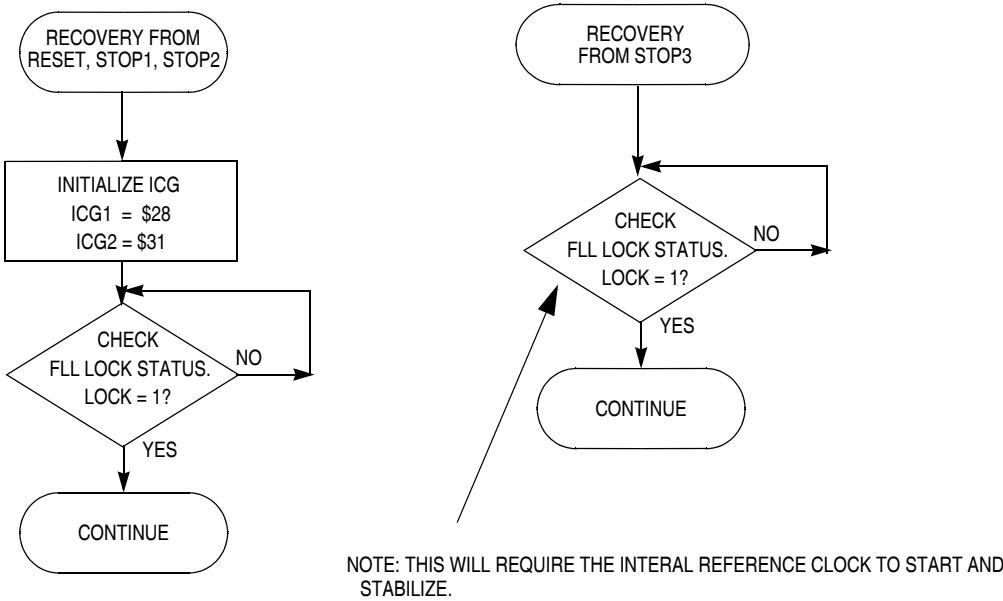


Figure 7-10. ICG Initialization and Stop Recovery for Example #3

7.4.5 Example #4: Internal Clock Generator Trim

The internally generated clock source is guaranteed to have a period $\pm 25\%$ of the nominal value. In some case this may be sufficient accuracy. For other applications that require a tight frequency tolerance, a trimming procedure is provided that will allow a very accurate source. This section outlines one example of trimming the internal oscillator. Many other possible trimming procedures are valid and can be used.

interrupt service routine, this would allow nesting of interrupts (which is not recommended because it leads to programs that are difficult to debug and maintain).

For compatibility with the earlier M68HC05 MCUs, the high-order half of the H:X index register pair (H) is not saved on the stack as part of the interrupt sequence. The user must use a PSHH instruction at the beginning of the service routine to save H and then use a PULH instruction just before the RTI that ends the interrupt service routine. It is not necessary to save H if you are certain that the interrupt service routine does not use any instructions or auto-increment addressing modes that might change the value of H.

The software interrupt (SWI) instruction is like a hardware interrupt except that it is not masked by the global I bit in the CCR and it is associated with an instruction opcode within the program so it is not asynchronous to program execution.

8.4.3 Wait Mode Operation

The WAIT instruction enables interrupts by clearing the I bit in the CCR. It then halts the clocks to the CPU to reduce overall power consumption while the CPU is waiting for the interrupt or reset event that will wake the CPU from wait mode. When an interrupt or reset event occurs, the CPU clocks will resume and the interrupt or reset event will be processed normally.

If a serial BACKGROUND command is issued to the MCU through the background debug interface while the CPU is in wait mode, CPU clocks will resume and the CPU will enter active background mode where other serial background commands can be processed. This ensures that a host development system can still gain access to a target MCU even if it is in wait mode.

8.4.4 Stop Mode Operation

Usually, all system clocks, including the crystal oscillator (when used), are halted during stop mode to minimize power consumption. In such systems, external circuitry is needed to control the time spent in stop mode and to issue a signal to wake up the target MCU when it is time to resume processing. Unlike the earlier M68HC05 and M68HC08 MCUs, the HCS08 can be configured to keep a minimum set of clocks running in stop mode. This optionally allows an internal periodic signal to wake the target MCU from stop mode.

When a host debug system is connected to the background debug pin (BKGD) and the ENBDM control bit has been set by a serial command through the background interface (or because the MCU was reset into active background mode), the oscillator is forced to remain active when the MCU enters stop mode. In this case, if a serial BACKGROUND command is issued to the MCU through the background debug interface while the CPU is in stop mode, CPU clocks will resume and the CPU will enter active background mode where other serial background commands can be processed. This ensures that a host development system can still gain access to a target MCU even if it is in stop mode.

Recovery from stop mode depends on the particular HCS08 and whether the oscillator was stopped in stop mode. Refer to the [Modes of Operation](#) chapter for more details.

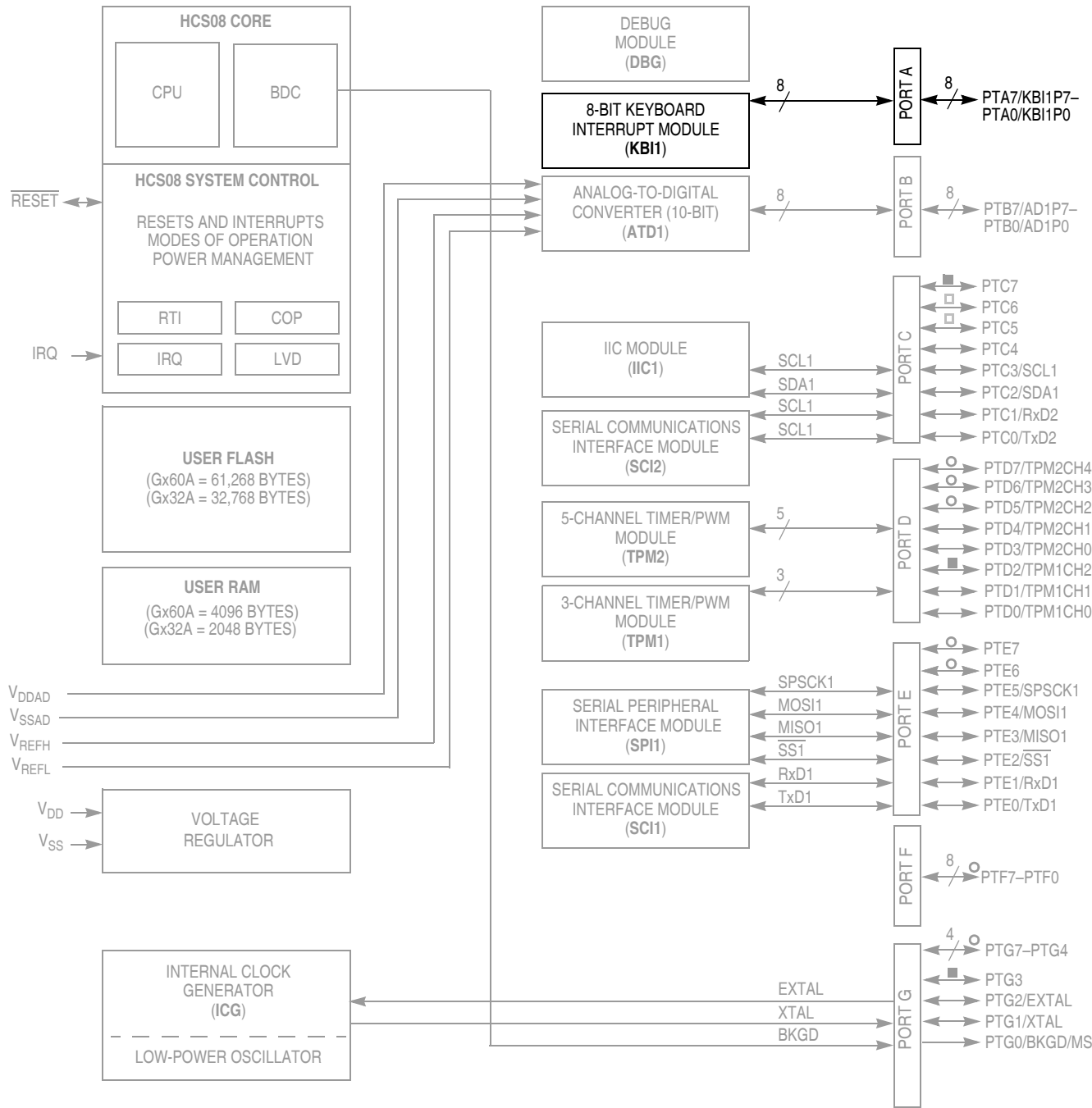
8.4.5 BGND Instruction

The BGND instruction is new to the HCS08 compared to the M68HC08. BGND would not be used in normal user programs because it forces the CPU to stop processing user instructions and enter the active background mode. The only way to resume execution of the user program is through reset or by a host debug system issuing a GO, TRACE1, or TAGGO serial command through the background debug interface.

Software-based breakpoints can be set by replacing an opcode at the desired breakpoint address with the BGND opcode. When the program reaches this breakpoint address, the CPU is forced to active background mode rather than continuing the user program.

Table 8-2. HCS08 Instruction Set Summary (Sheet 2 of 7)

Source Form	Operation	Description	Effect on CCR						Address Mode	Opcode	Operand	Bus Cycles ¹
			V	H	I	N	Z	C				
BCLR <i>n,opr8a</i>	Clear Bit n in Memory	$M_n \leftarrow 0$	–	–	–	–	–	–	DIR (b0) DIR (b1) DIR (b2) DIR (b3) DIR (b4) DIR (b5) DIR (b6) DIR (b7)	11 13 15 17 19 1B 1D 1F	dd dd dd dd dd dd dd dd	5 5 5 5 5 5 5 5
BCS <i>rel</i>	Branch if Carry Bit Set (Same as BLO)	Branch if (C) = 1	–	–	–	–	–	–	REL	25	rr	3
BEQ <i>rel</i>	Branch if Equal	Branch if (Z) = 1	–	–	–	–	–	–	REL	27	rr	3
BGE <i>rel</i>	Branch if Greater Than or Equal To (Signed Operands)	Branch if $(N \oplus V) = 0$	–	–	–	–	–	–	REL	90	rr	3
BGND	Enter Active Background if ENBDM = 1	Waits For and Processes BDM Commands Until GO, TRACE1, or TAGGO	–	–	–	–	–	–	INH	82		5+
BGT <i>rel</i>	Branch if Greater Than (Signed Operands)	Branch if $(Z) \mid (N \oplus V) = 0$	–	–	–	–	–	–	REL	92	rr	3
BHCC <i>rel</i>	Branch if Half Carry Bit Clear	Branch if (H) = 0	–	–	–	–	–	–	REL	28	rr	3
BHCS <i>rel</i>	Branch if Half Carry Bit Set	Branch if (H) = 1	–	–	–	–	–	–	REL	29	rr	3
BHI <i>rel</i>	Branch if Higher	Branch if $(C) \mid (Z) = 0$	–	–	–	–	–	–	REL	22	rr	3
BHS <i>rel</i>	Branch if Higher or Same (Same as BCC)	Branch if (C) = 0	–	–	–	–	–	–	REL	24	rr	3
BIH <i>rel</i>	Branch if IRQ Pin High	Branch if IRQ pin = 1	–	–	–	–	–	–	REL	2F	rr	3
BIL <i>rel</i>	Branch if IRQ Pin Low	Branch if IRQ pin = 0	–	–	–	–	–	–	REL	2E	rr	3
BIT # <i>opr8i</i> BIT <i>opr8a</i> BIT <i>opr16a</i> BIT <i>opr16,X</i> BIT <i>opr8,X</i> BIT <i>,X</i> BIT <i>opr16,SP</i> BIT <i>opr8,SP</i>	Bit Test	(A) & (M) (CCR Updated but Operands Not Changed)	0	–	–	↑	↑	–	IMM DIR EXT IX2 IX1 IX SP2 SP1	A5 B5 C5 D5 E5 F5 9ED5 9EE5	ii dd hh ll ee ff ff ff ee ff ff	2 3 4 4 3 3 5 4
BLE <i>rel</i>	Branch if Less Than or Equal To (Signed Operands)	Branch if $(Z) \mid (N \oplus V) = 1$	–	–	–	–	–	–	REL	93	rr	3
BLO <i>rel</i>	Branch if Lower (Same as BCS)	Branch if (C) = 1	–	–	–	–	–	–	REL	25	rr	3
BLS <i>rel</i>	Branch if Lower or Same	Branch if $(C) \mid (Z) = 1$	–	–	–	–	–	–	REL	23	rr	3
BLT <i>rel</i>	Branch if Less Than (Signed Operands)	Branch if $(N \oplus V) = 1$	–	–	–	–	–	–	REL	91	rr	3
BMC <i>rel</i>	Branch if Interrupt Mask Clear	Branch if (I) = 0	–	–	–	–	–	–	REL	2C	rr	3
BMI <i>rel</i>	Branch if Minus	Branch if (N) = 1	–	–	–	–	–	–	REL	2B	rr	3
BMS <i>rel</i>	Branch if Interrupt Mask Set	Branch if (I) = 1	–	–	–	–	–	–	REL	2D	rr	3
BNE <i>rel</i>	Branch if Not Equal	Branch if (Z) = 0	–	–	–	–	–	–	REL	26	rr	3
BPL <i>rel</i>	Branch if Plus	Branch if (N) = 0	–	–	–	–	–	–	REL	2A	rr	3
BRA <i>rel</i>	Branch Always	No Test	–	–	–	–	–	–	REL	20	rr	3



Note: Not all pins are bonded out in all packages. See Table 2-2 for complete details.

Block Diagram Symbol Key:	
○	Not connected in 48-, 44-, and 42-pin packages
■	Not connected in 44- and 42-pin packages
□	Not connected in 42-pin packages

Figure 9-2. Block Diagram Highlighting KBI Module

10.6.4 PWM End-of-Duty-Cycle Events

For channels that are configured for PWM operation, there are two possibilities:

- When the channel is configured for edge-aligned PWM, the channel flag is set when the timer counter matches the channel value register that marks the end of the active duty cycle period.
- When the channel is configured for center-aligned PWM, the timer count matches the channel value register twice during each PWM cycle. In this CPWM case, the channel flag is set at the start and at the end of the active duty cycle, which are the times when the timer counter matches the channel value register.

The flag is cleared by the 2-step sequence described in [Section 10.6.1, “Clearing Timer Interrupt Flags.”](#)

10.7 TPM Registers and Control Bits

The TPM includes:

- An 8-bit status and control register (TPM_xSC)
- A 16-bit counter (TPM_xCNTH:TPM_xCNTL)
- A 16-bit modulo register (TPM_xMODH:TPM_xMODL)

Each timer channel has:

- An 8-bit status and control register (TPM_xCnSC)
- A 16-bit channel value register (TPM_xCnVH:TPM_xCnVL)

Refer to the direct-page register summary in the [Memory](#) chapter of this data sheet for the absolute address assignments for all TPM registers. This section refers to registers and control bits only by their names. A Freescale-provided equate or header file is used to translate these names into the appropriate absolute addresses.

Some MCU systems have more than one TPM, so register names include placeholder characters to identify which TPM and which channel is being referenced. For example, TPM_xCnSC refers to timer (TPM) *x*, channel *n* and TPM1C2SC is the status and control register for timer 1, channel 2.

12.3 Modes of Operation

12.3.1 SPI in Stop Modes

The SPI is disabled in all stop modes, regardless of the settings before executing the STOP instruction. During either stop1 or stop2 mode, the SPI module will be fully powered down. Upon wake-up from stop1 or stop2 mode, the SPI module will be in the reset state. During stop3 mode, clocks to the SPI module are halted. No registers are affected. If stop3 is exited with a reset, the SPI will be put into its reset state. If stop3 is exited with an interrupt, the SPI continues from the state it was in when stop3 was entered.

12.4 Register Definition

The SPI has five 8-bit registers to select SPI options, control baud rate, report SPI status, and for transmit/receive data.

Refer to the direct-page register summary in the [Memory](#) chapter of this data sheet for the absolute address assignments for all SPI registers. This section refers to registers and control bits only by their names, and a Freescale-provided equate or header file is used to translate these names into the appropriate absolute addresses.

12.4.1 SPI Control Register 1 (SPI1C1)

This read/write register includes the SPI enable control, interrupt enables, and configuration options.

	7	6	5	4	3	2	1	0
R								
W								
Reset	0	0	0	0	0	1	0	0

Figure 12-5. SPI Control Register 1 (SPI1C1)

Table 12-1. SPI1C1 Field Descriptions

Field	Description
7 SPIE	SPI Interrupt Enable (for SPRF and MODF) — This is the interrupt enable for SPI receive buffer full (SPRF) and mode fault (MODF) events. 0 Interrupts from SPRF and MODF inhibited (use polling) 1 When SPRF or MODF is 1, request a hardware interrupt
6 SPE	SPI System Enable — Disabling the SPI halts any transfer that is in progress, clears data buffers, and initializes internal state machines. SPRF is cleared and SPTEF is set to indicate the SPI transmit data buffer is empty. 0 SPI system inactive 1 SPI system enabled
5 SPTIE	SPI Transmit Interrupt Enable — This is the interrupt enable bit for SPI transmit buffer empty (SPTEF). 0 Interrupts from SPTEF inhibited (use polling) 1 When SPTEF is 1, hardware interrupt requested

Table 12-3. SPI1C2 Register Field Descriptions

Field	Description
4 MODFEN	Master Mode-Fault Function Enable — When the SPI is configured for slave mode, this bit has no meaning or effect. (The $\overline{SS}$ pin is the slave select input.) In master mode, this bit determines how the $\overline{SS}$ pin is used (refer to Table 12-2 for more details). 0 Mode fault function disabled, master $\overline{SS}$ pin reverts to general-purpose I/O not controlled by SPI 1 Mode fault function enabled, master $\overline{SS}$ pin acts as the mode fault input or the slave select output
3 BIDIROE	Bidirectional Mode Output Enable — When bidirectional mode is enabled by SPI pin control 0 (SPC0) = 1, BIDIROE determines whether the SPI data output driver is enabled to the single bidirectional SPI I/O pin. Depending on whether the SPI is configured as a master or a slave, it uses either the MOSI (MOMI) or MISO (SISO) pin, respectively, as the single SPI data I/O pin. When SPC0 = 0, BIDIROE has no meaning or effect. 0 Output driver disabled so SPI data I/O pin acts as an input 1 SPI I/O pin enabled as an output
1 SPISWAI	SPI Stop in Wait Mode 0 SPI clocks continue to operate in wait mode 1 SPI clocks stop when the MCU enters wait mode
0 SPC0	SPI Pin Control 0 — The SPC0 bit chooses single-wire bidirectional mode. If MSTR = 0 (slave mode), the SPI uses the MISO (SISO) pin for bidirectional SPI data transfers. If MSTR = 1 (master mode), the SPI uses the MOSI (MOMI) pin for bidirectional SPI data transfers. When SPC0 = 1, BIDIROE is used to enable or disable the output driver for the single bidirectional SPI I/O pin. 0 SPI uses separate pins for data input and data output 1 SPI configured for single-wire bidirectional operation

12.4.3 SPI Baud Rate Register (SPI1BR)

This register is used to set the prescaler and bit rate divisor for an SPI master. This register may be read or written at any time.

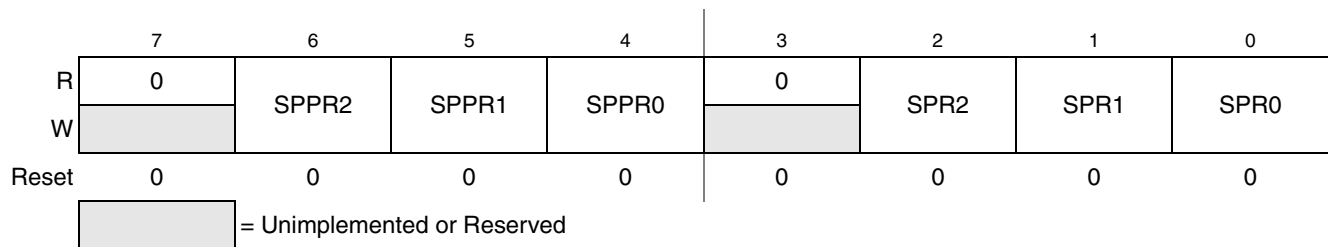


Figure 12-7. SPI Baud Rate Register (SPI1BR)

Table 12-4. SPI1BR Register Field Descriptions

Field	Description
6:4 SPPR[2:0]	SPI Baud Rate Prescale Divisor — This 3-bit field selects one of eight divisors for the SPI baud rate prescaler as shown in Table 12-5. The input to this prescaler is the bus rate clock (BUSCLK). The output of this prescaler drives the input of the SPI baud rate divider (see Figure 12-4).
2:0 SPR[2:0]	SPI Baud Rate Divisor — This 3-bit field selects one of eight divisors for the SPI baud rate divider as shown in Table 12-6. The input to this divider comes from the SPI baud rate prescaler (see Figure 12-4). The output of this divider is the SPI bit rate clock for master mode.

Chapter 14

Analog-to-Digital Converter (S08ATDV3)

The MC9S08GBxxA/GTxxA provides one 8-channel analog-to-digital (ATD) module. The eight ATD channels share port B. Each channel individually can be configured for general-purpose I/O or for ATD functionality. All features of the ATD module as described in this section are available on the MC9S08GBxxA/GTxxA. Electrical parametric information for the ATD may be found in [Appendix A](#), “Electrical Characteristics.”

Table 14-5. Clock Prescaler Values

PRS	Factor = (PRS + 1) × 2	Max Bus Clock MHz (2 MHz max ATD Clock) ¹	Max Bus Clock MHz (1 MHz max ATD Clock) ²	Min Bus Clock ³ MHz (500 kHz min ATD Clock)
0000	2	4	2	1
0001	4	8	4	2
0010	6	12	6	3
0011	8	16	8	4
0100	10	20	10	5
0101	12	20	12	6
0110	14	20	14	7
0111	16	20	16	8
1000	18	20	18	9
1001	20	20	20	10
1010	22	20	20	11
1011	24	20	20	12
1100	26	20	20	13
1101	28	20	20	14
1110	30	20	20	15
1111	32	20	20	16

¹ Maximum ATD conversion clock frequency is 2 MHz. The max bus clock frequency is computed from the max ATD conversion clock frequency times the indicated prescaler setting; i.e., for a PRS of 0, max bus clock = 2 (max ATD conversion clock frequency) × 2 (Factor) = 4 MHz.

² Use these settings if the maximum desired ATD conversion clock frequency is 1 MHz. The max bus clock frequency is computed from the max ATD conversion clock frequency times the indicated prescaler setting; i.e., for a PRS of 0, max bus clock = 1 (max ATD conversion clock frequency) × 2 (Factor) = 2 MHz.

³ Minimum ATD conversion clock frequency is 500 kHz. The min bus clock frequency is computed from the min ATD conversion clock frequency times the indicated prescaler setting; i.e., for a PRS of 1, min bus clock = 0.5 (min ATD conversion clock frequency) × 2 (Factor) = 1 MHz.

14.6.2 ATD Status and Control (ATD1SC)

Writes to the ATD status and control register clears the CCF flag, cancels any pending interrupts, and initiates a new conversion.

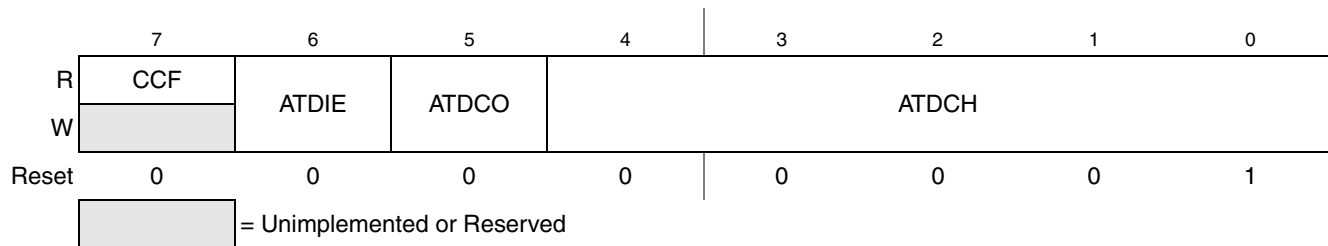


Figure 14-6. ATD Status and Control Register (ATD1SC)

15.3 On-Chip Debug System (DBG)

Because HCS08 devices do not have external address and data buses, the most important functions of an in-circuit emulator have been built onto the chip with the MCU. The debug system consists of an 8-stage FIFO that can store address or data bus information, and a flexible trigger system to decide when to capture bus information and what information to capture. The system relies on the single-wire background debug system to access debug control registers and to read results out of the eight stage FIFO.

The debug module includes control and status registers that are accessible in the user's memory map. These registers are located in the high register space to avoid using valuable direct page memory space.

Most of the debug module's functions are used during development, and user programs rarely access any of the control and status registers for the debug module. The one exception is that the debug system can provide the means to implement a form of ROM patching. This topic is discussed in greater detail in [Section 15.3.6, "Hardware Breakpoints."](#)

15.3.1 Comparators A and B

Two 16-bit comparators (A and B) can optionally be qualified with the R/W signal and an opcode tracking circuit. Separate control bits allow you to ignore R/W for each comparator. The opcode tracking circuitry optionally allows you to specify that a trigger will occur only if the opcode at the specified address is actually executed as opposed to only being read from memory into the instruction queue. The comparators are also capable of magnitude comparisons to support the inside range and outside range trigger modes. Comparators are disabled temporarily during all BDC accesses.

The A comparator is always associated with the 16-bit CPU address. The B comparator compares to the CPU address or the 8-bit CPU data bus, depending on the trigger mode selected. Because the CPU data bus is separated into a read data bus and a write data bus, the RWAEN and RWA control bits have an additional purpose, in full address plus data comparisons they are used to decide which of these buses to use in the comparator B data bus comparisons. If RWAEN = 1 (enabled) and RWA = 0 (write), the CPU's write data bus is used. Otherwise, the CPU's read data bus is used.

The currently selected trigger mode determines what the debugger logic does when a comparator detects a qualified match condition. A match can cause:

- Generation of a breakpoint to the CPU
- Storage of data bus values into the FIFO
- Starting to store change-of-flow addresses into the FIFO (begin type trace)
- Stopping the storage of change-of-flow addresses into the FIFO (end type trace)

15.3.2 Bus Capture Information and FIFO Operation

The usual way to use the FIFO is to setup the trigger mode and other control options, then arm the debugger. When the FIFO has filled or the debugger has stopped storing data into the FIFO, you would read the information out of it in the order it was stored into the FIFO. Status bits indicate the number of words of valid information that are in the FIFO as data is stored into it. If a trace run is manually halted by writing 0 to ARM before the FIFO is full (CNT = 1:0:0:0), the information is shifted by one position and

- ¹ Self-clocked mode frequency is the frequency that the DCO generates when the FLL is open-loop.
- ² Loss of reference frequency is the reference frequency detected internally, which transitions the ICG into self-clocked mode if it is not in the desired range.
- ³ Loss of DCO frequency is the DCO frequency detected internally, which transitions the ICG into FLL bypassed external mode (if an external reference exists) if it is not in the desired range.
- ⁴ This parameter is characterized before qualification rather than 100% tested.
- ⁵ Proper PC board layout procedures must be followed to achieve specifications.
- ⁶ This specification applies to the period of time required for the FLL to lock after entering FLL engaged internal or external modes. If a crystal/resonator is being used as the reference, this specification assumes it is already running.
- ⁷ Jitter is the average deviation from the programmed frequency measured over the specified interval at maximum f_{ICGOUT} . Measurements are made with the device powered by filtered supplies and clocked by a stable external clock signal. Noise injected into the FLL circuitry via V_{DDA} and V_{SSA} and variation in crystal oscillator frequency increase the C_{Jitter} percentage for a given interval.
- ⁸ See [Figure A-10](#)

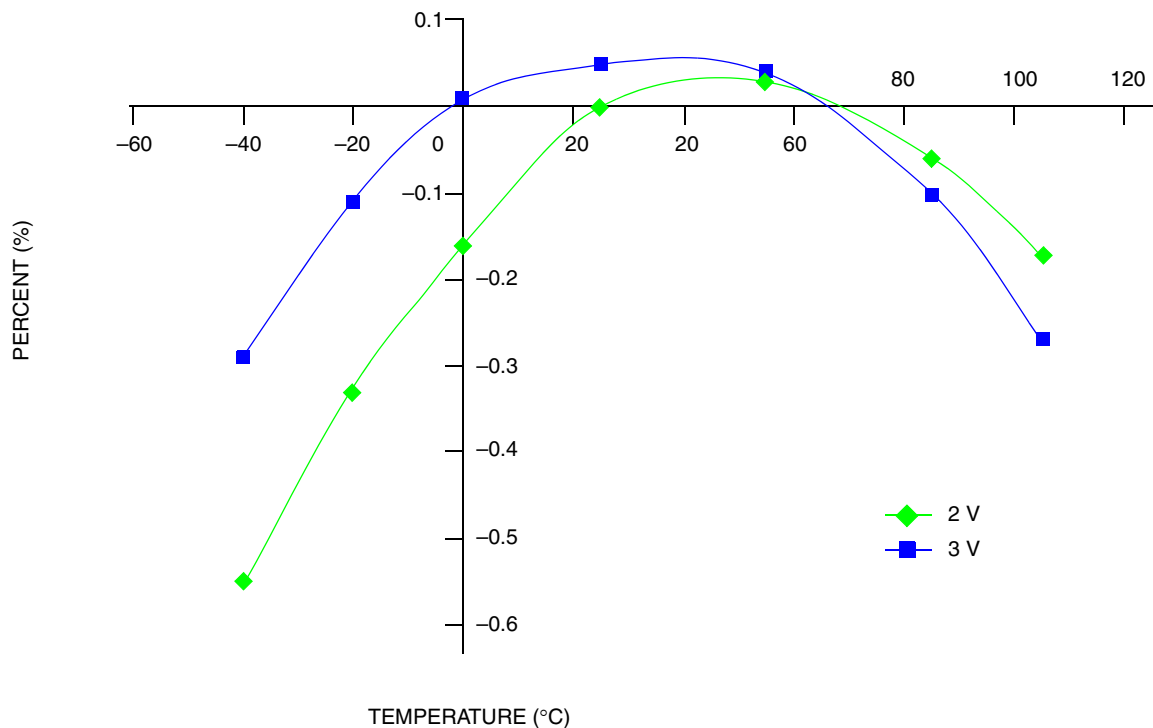


Figure A-10. Internal Oscillator Deviation from Trimmed Frequency

A.9 AC Characteristics

This section describes ac timing characteristics for each peripheral system. For detailed information about how clocks for the bus are generated, see [Chapter 7, “Internal Clock Generator \(S08ICGV2\).”](#)

improved noise immunity in the oscillator circuit. The low-power oscillator is also available for power-sensitive applications.

This new oscillator option available on the GB60A series is selected by a new control bit in the ICG control register 1 (ICGC1): the HGO bit. HGO is bit 7 of the ICGC1 register, formerly an unimplemented bit that always read '0'. The reset value is '0', which selects the low-power oscillator option—which is consistent with the GB60 series external oscillator.

Setting HGO to '1' selects the high gain external oscillator which increases the voltage swing across the external crystal or resonator, making it more immune to external noise.

The values of the feedback and series resistors for the external oscillator will be different in most cases between HGO=0 and HGO=1. Consult the ICG DC Electrical Specifications table in the MC9S08GB60A data sheet for the proper values.

B.5 Internal Clock Generator: Low-Power Oscillator Maximum Frequency

On the GB60 series, the external oscillator's maximum frequency is 10 MHz when in FEE mode and 16 MHz when in FBE mode.

On the GB60A series, when HGO=1, the same maximum frequencies apply. However, when HGO=0, the maximum frequency is 10 MHz in FEE and FBE modes.

B.6 Internal Clock Generator: Loss-of-Clock Disable Option

The ICG module has a clock monitor which will generate a loss-of-clock signal when either the reference clock or the DCO clock does not meet minimum frequency requirements. This signal is used to generate either a reset or an interrupt, depending on the settings in the ICGC2 register.

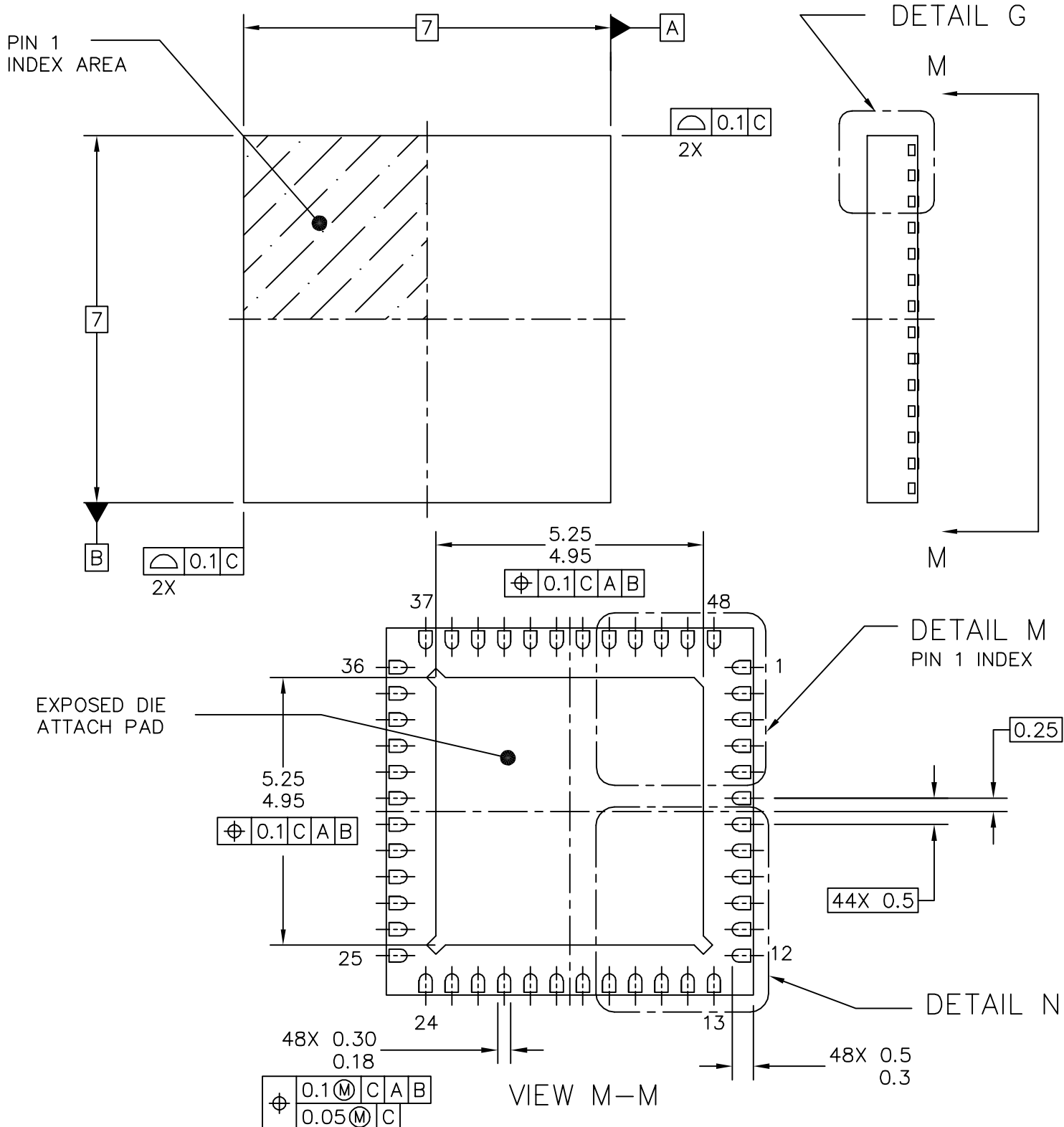
On the GB60 series, this clock monitor cannot be turned on or off by the user. The on/off status of the clock monitor is determined by the state of the ICG module.

On the GB60A series, an option has been added to allow the user to disable the clock monitor. A new control bit, LOCD, has been added to the ICGC1 register at bit position 1, formerly an unimplemented bit. The reset state is '0', which enables the clock monitor. Setting LOCD = 1 will disable the clock monitor and thereby eliminate any loss-of-clock resets or interrupts.

The advantage of disabling the clock monitor is to reduce the current draw of the ICG module. Disabling the clock monitor when running in stop3 mode with a low-range external oscillator enabled will save approximately 9 μA of current. With LOCD=0 in this configuration, the stop3 I_{DD} is about 14 μA . When LOCD=1 in this configuration, the stop3 I_{DD} is about 5 μA .

For the best combination of power conservation and system protection, Freescale Semiconductor recommends setting the LOCD=0 whenever the MCU is in active run mode and then setting LOCD=1 just before entering stop3 mode when OSCSTEN=1. If OSCSTEN=0, then the LOCD bit will not make a difference in the stop3 current.

PIN 1 —
INDEX AREA



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			CASE NUMBER: 1314-04		20 APR 2005
			STANDARD: JEDEC-MO-220 VKKD-2		