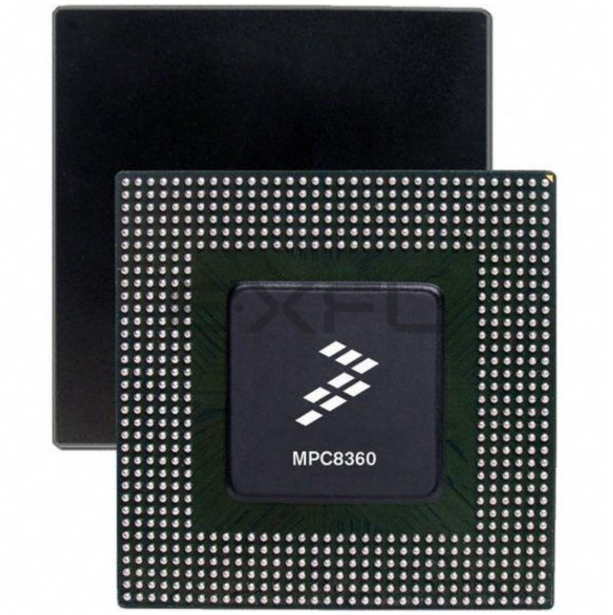
NXP USA Inc. - KMPC8358CZUAGDG Datasheet



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Understanding Embedded - Microprocessors

Embedded microprocessors are specialized computing chips designed to perform specific tasks within an embedded system. Unlike general-purpose microprocessors found in personal computers, embedded microprocessors are tailored for dedicated functions within larger systems, offering optimized performance, efficiency, and reliability. These microprocessors are integral to the operation of countless electronic devices, providing the computational power necessary for controlling processes, handling data, and managing communications.

Applications of Embedded - Microprocessors

Embedded microprocessors are utilized across a broad spectrum of applications, making them indispensable in

Details	
Product Status	Obsolete
Core Processor	PowerPC e300
Number of Cores/Bus Width	1 Core, 32-Bit
Speed	400MHz
Co-Processors/DSP	Communications; QUICC Engine
RAM Controllers	DDR, DDR2
Graphics Acceleration	No
Display & Interface Controllers	-
Ethernet	10/100/1000Mbps (1)
SATA	-
USB	USB 1.x (1)
Voltage - I/O	1.8V, 2.5V, 3.3V
Operating Temperature	-40°C ~ 105°C (TA)
Security Features	-
Package / Case	740-LBGA
Supplier Device Package	740-TBGA (37.5x37.5)
Purchase URL	https://www.e-xfl.com/product-detail/nxp-semiconductors/kmpc8358czuagdg

  - Eight TDM interfaces on the MPC8360E and four TDM interfaces on the MPC8358E with 1-bit mode for E3/T3 rates in clear channel
  - Sixteen independent baud rate generators and 30 input clock pins for supplying clocks to UCC and MCC serial channels (MCC is only available on the MPC8360E)
  - Four independent 16-bit timers that can be interconnected as four 32-bit timers
  - Interworking functionality:
    - Layer 2 10/100-Base T Ethernet switch
    - ATM-to-ATM switching (AAL0, 2, 5)
    - Ethernet-to-ATM switching with L3/L4 support
    - PPP interworking
- Security engine is optimized to handle all the algorithms associated with IPSec, SSL/TLS, SRTP, 802.11i®, iSCSI, and IKE processing. The security engine contains four crypto-channels, a controller, and a set of crypto execution units (EUs).
  - Public key execution unit (PKEU) supporting the following:
    - RSA and Diffie-Hellman
    - Programmable field size up to 2048 bits
    - Elliptic curve cryptography
    - F2m and F(p) modes
    - Programmable field size up to 511 bits
  - Data encryption standard execution unit (DEU)
    - DES, 3DES
    - Two key (K1, K2) or three key (K1, K2, K3)
    - ECB and CBC modes for both DES and 3DES
  - Advanced encryption standard unit (AESU)
  - Implements the Rinjdael symmetric key cipher
  - Key lengths of 128, 192, and 256 bits, two key
    - ECB, CBC, CCM, and counter modes
  - ARC four execution unit (AFEU)
    - Implements a stream cipher compatible with the RC4 algorithm
    - 40- to 128-bit programmable key
  - Message digest execution unit (MDEU)
    - SHA with 160-, 224-, or 256-bit message digest
    - MD5 with 128-bit message digest
    - HMAC with either SHA or MD5 algorithm
  - Random number generator (RNG)
  - Four crypto-channels, each supporting multi-command descriptor chains
    - Static and/or dynamic assignment of crypto-execution units via an integrated controller
    - Buffer size of 256 bytes for each execution unit, with flow control for large data sizes
  - Storage/NAS XOR parity generation accelerator for RAID applications
- Dual DDR SDRAM memory controllers on the MPC8360E and a single DDR SDRAM memory controller on the MPC8358E
  - Programmable timing supporting both DDR1 and DDR2 SDRAM
  - On the MPC8360E, the DDR buses can be configured as two 32-bit buses or one 64-bit bus; on the MPC8358E, the DDR bus can be configured as a 32- or 64-bit bus
  - 32- or 64-bit data interface, up to 333 MHz (for the MPC8360E) and 266 MHz (for the MPC8358E) data rate
  - Four banks of memory, each up to 1 Gbyte

  - Programmable highest priority request
  - Four groups of interrupts with programmable priority
  - External and internal interrupts directed to communication processor
  - Redirects interrupts to external $\overline{\text{INTA}}$ pin when in core disable mode
  - Unique vector number for each interrupt source
- Dual industry-standard I^{2}C interfaces
  - Two-wire interface
  - Multiple master support
  - Master or slave I^{2}C mode support
  - On-chip digital filtering rejects spikes on the bus
  - System initialization data is optionally loaded from I^{2}C-1 EPROM by boot sequencer embedded hardware
- DMA controller
  - Four independent virtual channels
  - Concurrent execution across multiple channels with programmable bandwidth control
  - All channels accessible by local core and remote PCI masters
  - Misaligned transfer capability
  - Data chaining and direct mode
  - Interrupt on completed segment and chain
  - DMA external handshake signals: $\overline{\text{DMA\_DREQ}}$[0:3]/$\overline{\text{DMA\_DACK}}$[0:3]/$\overline{\text{DMA\_DONE}}$[0:3]. There is one set for each DMA channel. The pins are multiplexed to the parallel IO pins with other QE functions.
- DUART
  - Two 4-wire interfaces (RxD, TxD, RTS, CTS)
  - Programming model compatible with the original 16450 UART and the PC16550D
- System timers
  - Periodic interrupt timer
  - Real-time clock
  - Software watchdog timer
  - Eight general-purpose timers
- IEEE Std. 1149.1™-compliant, JTAG boundary scan
- Integrated PCI bus and SDRAM clock generation

2 Electrical Characteristics

This section provides the AC and DC electrical specifications and thermal characteristics for the MPC8360E/58E. The device is currently targeted to these specifications. Some of these specifications are independent of the I/O cell, but are included for a more complete reference. These are not purely I/O buffer design specifications.

Table 4. MPC8360E TBGA Core Power Dissipation[1] (continued)

Core Frequency (MHz)	CSB Frequency (MHz)	QUICC Engine Frequency (MHz)	Typical	Maximum	Unit	Notes
667	333	500	6.1	6.8	W	2, 3, 5, 9

Notes:

1. The values do not include I/O supply power (OV_{DD}, LV_{DD}, GV_{DD}) or AV_{DD}. For I/O power values, see Table 6.
2. Typical power is based on a voltage of V_{DD} = 1.2 V or 1.3 V, a junction temperature of T_J = 105° C, and a Dhrystone benchmark application.
3. Thermal solutions need to design to a value higher than typical power on the end application, T_A target, and I/O power.
4. Maximum power is based on a voltage of V_{DD} = 1.2 V, WC process, a junction T_J = 105° C, and an artificial smoke test.
5. Maximum power is based on a voltage of V_{DD} = 1.3 V for applications that use 667 MHz (CPU)/500 (QE) with WC process, a junction T_J = 105° C, and an artificial smoke test.
6. Typical power is based on a voltage of V_{DD} = 1.3 V, a junction temperature of T_J = 70° C, and a Dhrystone benchmark application.
7. Maximum power is based on a voltage of V_{DD} = 1.3 V for applications that use 667 MHz (CPU) or 500 (QE) with WC process, a junction T_J = 70° C, and an artificial smoke test.
8. This frequency combination is only available for rev. 2.0 silicon.
9. This frequency combination is not available for rev. 2.0 silicon.

Table 5. MPC8358E TBGA Core Power Dissipation[1]

Core Frequency (MHz)	CSB Frequency (MHz)	QUICC Engine Frequency (MHz)	Typical	Maximum	Unit	Notes
266	266	300	4.1	4.5	W	2, 3, 4
400	266	400	4.5	5.0	W	2, 3, 4

Notes:

1. The values do not include I/O supply power (OV_{DD}, LV_{DD}, GV_{DD}) or AV_{DD}. For I/O power values, see Table 6.
2. Typical power is based on a voltage of V_{DD} = 1.2 V, a junction temperature of T_J = 105° C, and a Dhrystone benchmark application.
3. Thermal solutions need to design to a value higher than typical power on the end application, T_A target, and I/O power.
4. Maximum power is based on a voltage of V_{DD} = 1.2 V, WC process, a junction T_J = 105° C, and an artificial smoke test.

4.1 DC Electrical Characteristics

This table provides the clock input (CLKIN/PCI_SYNC_IN) DC timing specifications for the device.

Table 7. CLKIN DC Electrical Characteristics

Parameter	Condition	Symbol	Min	Max	Unit
Input high voltage	—	V_{IH}	2.7	OV_{DD} + 0.3	V
Input low voltage	—	V_{IL}	–0.3	0.4	V
CLKIN input current	$0 V \le V_{IN} \le OV_{DD}$	I_{IN}	—	±10	µA
PCI_SYNC_IN input current	$0 V \le V_{IN} \le 0.5V$ or $OV_{DD} - 0.5V \le V_{IN} \le OV_{DD}$	I_{IN}	—	±10	µA
PCI_SYNC_IN input current	$0.5 V \le V_{IN} \le OV_{DD} - 0.5 V$	I_{IN}	—	±100	µA

4.2 AC Electrical Characteristics

The primary clock source for the device can be one of two inputs, CLKIN or PCI_CLK, depending on whether the device is configured in PCI host or PCI agent mode. This table provides the clock input (CLKIN/PCI_CLK) AC timing specifications for the device.

Table 8. CLKIN AC Timing Specifications

Parameter/Condition	Symbol	Min	Typical	Max	Unit	Notes
CLKIN/PCI_CLK frequency	f_{CLKIN}	—	—	66.67	MHz	1
CLKIN/PCI_CLK cycle time	t_{CLKIN}	15	—	—	ns	—
CLKIN/PCI_CLK rise and fall time	t_{KH}, t_{KL}	0.6	1.0	2.3	ns	2
CLKIN/PCI_CLK duty cycle	t_{KHK}/t_{CLKIN}	40	—	60	%	3
CLKIN/PCI_CLK jitter	—	—	—	±150	ps	4, 5

Notes:

1. **Caution:** The system, core, USB, security, and 10/100/1000 Ethernet must not exceed their respective maximum or minimum operating frequencies.
2. Rise and fall times for CLKIN/PCI_CLK are measured at 0.4 V and 2.7 V.
3. Timing is guaranteed by design and characterization.
4. This represents the total input jitter—short term and long term—and is guaranteed by design.
5. The CLKIN/PCI_CLK driver's closed loop jitter bandwidth should be <500 kHz at –20 dB. The bandwidth must be set low to allow cascade-connected PLL-based devices to track CLKIN drivers with the specified jitter.

4.3 Gigabit Reference Clock Input Timing

This table provides the Gigabit reference clocks (GTX_CLK125) AC timing specifications.

Table 9. GTX_CLK125 AC Timing Specifications

At recommended operating conditions with LV_{DD} = 2.5 ± 0.125 mV/ 3.3 V ± 165 mV

Parameter/Condition	Symbol	Min	Typical	Max	Unit	Notes
GTX_CLK125 frequency	t_{G125}	—	125	—	MHz	—
GTX_CLK125 cycle time	t_{G125}	—	8	—	ns	—

Table 9. GTX_CLK125 AC Timing Specifications

At recommended operating conditions with LV_{DD} = 2.5 ± 0.125 mV/ 3.3 V ± 165 mV (continued)

Parameter/Condition	Symbol	Min	Typical	Max	Unit	Notes
GTX_CLK rise and fall time LV_{DD} = 2.5 V LV_{DD} = 3.3 V	t_{G125R}/t_{G125F}	—	—	 0.75 1.0	ns	1
GTX_CLK125 duty cycle GMII & TBI 1000Base-T for RGMII & RTBI	t_{G125H}/t_{G125}	 45 47	—	 55 53	%	2
GTX_CLK125 jitter	—	—	—	±150	ps	2

Notes:

1. Rise and fall times for GTX_CLK125 are measured from 0.5 and 2.0 V for LV_{DD} = 2.5 V and from 0.6 and 2.7 V for LV_{DD} = 3.3 V.
2. GTX_CLK125 is used to generate the GTX clock for the UCC Ethernet transmitter with 2% degradation. The GTX_CLK125 duty cycle can be loosened from 47%/53% as long as the PHY device can tolerate the duty cycle generated by GTX_CLK. See Section 8.2.2, “MII AC Timing Specifications,” Section 8.2.3, “RMII AC Timing Specifications,” and Section 8.2.5, “RGMII and RTBI AC Timing Specifications” for the duty cycle for 10Base-T and 100Base-T reference clock.

5 RESET Initialization

This section describes the DC and AC electrical specifications for the reset initialization timing and electrical requirements of the MPC8360E/58E.

5.1 RESET DC Electrical Characteristics

This table provides the DC electrical characteristics for the RESET pins of the device.

Table 10. RESET Pins DC Electrical Characteristics [1]

Characteristic	Symbol	Condition	Min	Max	Unit
Input high voltage	V_{IH}	—	2.0	OV_{DD} + 0.3	V
Input low voltage	V_{IL}	—	–0.3	0.8	V
Input current	I_{IN}	—	—	±10	µA
Output high voltage	V_{OH}[2]	I_{OH} = –8.0 mA	2.4	—	V
Output low voltage	V_{OL}	I_{OL} = 8.0 mA	—	0.5	V
Output low voltage	V_{OL}	I_{OL} = 3.2 mA	—	0.4	V

Notes:

1. This table applies for pins $\overline{\text{PORESET}}$, $\overline{\text{HRESET}}$, $\overline{\text{SRESET}}$, and $\overline{\text{QUIESCE}}$.
2. $\overline{\text{HRESET}}$ and $\overline{\text{SRESET}}$ are open drain pins, thus V_{OH} is not relevant for those pins.

8.2.1.2 GMII Receive AC Timing Specifications

This table provides the GMII receive AC timing specifications.

Table 28. GMII Receive AC Timing Specifications

At recommended operating conditions with LV_{DD}/OV_{DD} of 3.3 V ± 10%.

Parameter/Condition	Symbol[1]	Min	Typ	Max	Unit	Notes
RX_CLK clock period	t_{GRX}	—	8.0	—	ns	—
RX_CLK duty cycle	t_{GRXH}/t_{GRX}	40	—	60	%	—
RXD[7:0], RX_DV, RX_ER setup time to RX_CLK	t_{GRDVKH}	2.0	—	—	ns	—
RXD[7:0], RX_DV, RX_ER hold time to RX_CLK	t_{GRDXKH}	0.2	—	—	ns	2
RX_CLK clock rise time, (20% to 80%)	t_{GRXR}	—	—	1.0	ns	—
RX_CLK clock fall time, (80% to 20%)	t_{GRXF}	—	—	1.0	ns	—

Notes:

1. The symbols used for timing specifications follow the pattern of $t_{\text{(first two letters of functional block)(signal)(state)(reference)(state)}}$ for inputs and $t_{\text{(first two letters of functional block)(reference)(state)(signal)(state)}}$ for outputs. For example, t_{GRDVKH} symbolizes GMII receive timing (GR) with respect to the time data input signals (D) reaching the valid state (V) relative to the t_{RX} clock reference (K) going to the high state (H) or setup time. Also, t_{GRDXKL} symbolizes GMII receive timing (GR) with respect to the time data input signals (D) went invalid (X) relative to the t_{GRX} clock reference (K) going to the low (L) state or hold time. Note that, in general, the clock reference symbol representation is based on three letters representing the clock of a particular functional. For example, the subscript of t_{GRX} represents the GMII (G) receive (RX) clock. For rise and fall times, the latter convention is used with the appropriate letter: R (rise) or F (fall).
2. In rev. 2.0 silicon, due to errata, t_{GRDXKH} minimum is 0.5 which is not compliant with the standard. Refer to Errata *QE_ENET18* in *Chip Errata for the MPC8360E, Rev. 1*.

This figure shows the GMII receive AC timing diagram.

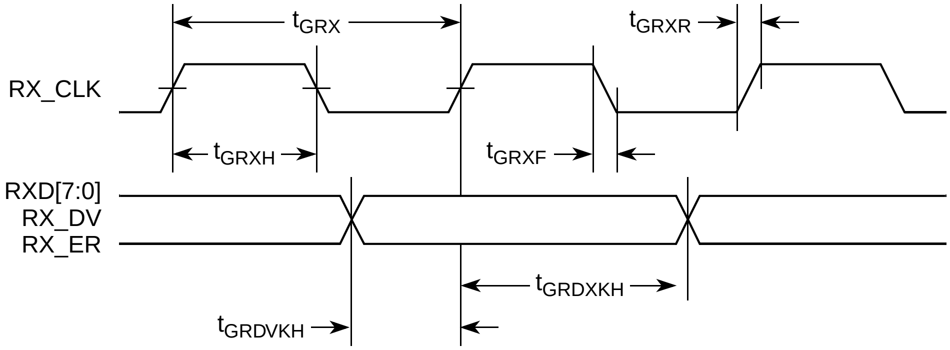

Figure 11. GMII Receive AC Timing Diagram

8.2.4.1 TBI Transmit AC Timing Specifications

This table provides the TBI transmit AC timing specifications.

Table 33. TBI Transmit AC Timing Specifications

At recommended operating conditions with LV_{DD}/OV_{DD} of 3.3 V ± 10%.

Parameter/Condition	Symbol[1]	Min	Typ	Max	Unit	Notes
GTX_CLK clock period	t_{TTX}	—	8.0	—	ns	—
GTX_CLK duty cycle	t_{TTXH}/t_{TTX}	40	—	60	%	—
GTX_CLK to TBI data TCG[9:0] delay	t_{TTKHDX} t_{TTKHDV}	1.0 —	—	— 5.0	ns	3
GTX_CLK clock rise time, (20% to 80%)	t_{TTXR}	—	—	1.0	ns	—
GTX_CLK clock fall time, (80% to 20%)	t_{TTXF}	—	—	1.0	ns	—
GTX_CLK125 reference clock period	t_{G125}	—	8.0	—	ns	2
GTX_CLK125 reference clock duty cycle	t_{G125H}/t_{G125}	45	—	55	ns	—

Notes:

1. The symbols used for timing specifications follow the pattern of $t_{\text{(first two letters of functional block)(signal)(state)(reference)(state)}}$ for inputs and $t_{\text{(first two letters of functional block)(reference)(state)(signal)(state)}}$ for outputs. For example, t_{TTKHDV} symbolizes the TBI transmit timing (TT) with respect to the time from t_{TTX} (K) going high (H) until the referenced data signals (D) reach the valid state (V) or setup time. Also, t_{TTKHDX} symbolizes the TBI transmit timing (TT) with respect to the time from t_{TTX} (K) going high (H) until the referenced data signals (D) reach the invalid state (X) or hold time. Note that, in general, the clock reference symbol representation is based on three letters representing the clock of a particular functional. For example, the subscript of t_{TTX} represents the TBI (T) transmit (TX) clock. For rise and fall times, the latter convention is used with the appropriate letter: R (rise) or F (fall).
2. This symbol is used to represent the external GTX_CLK125 and does not follow the original symbol naming convention.
3. In rev. 2.0 silicon, due to errata, t_{TTKHDX} minimum is 0.7 ns for UCC1. Refer to Errata *QE_ENET19* in *Chip Errata for the MPC8360E, Rev. 1*.

This figure shows the TBI transmit AC timing diagram.

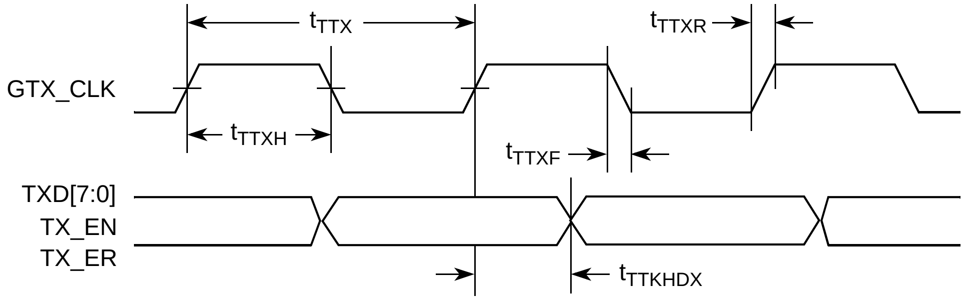

Figure 18. TBI Transmit AC Timing Diagram

Table 41. Local Bus General Timing Parameters—DLL Bypass Mode[9] (continued)

Parameter	Symbol[1]	Min	Max	Unit	Notes
Local bus clock to output valid	t_{LBKHOV}	—	3	ns	3
Local bus clock to output high impedance for LAD/LDP	t_{LBKHOZ}	—	4	ns	8

Notes:

1. The symbols used for timing specifications follow the pattern of $t_{\text{(first two letters of functional block)(signal)(state)(reference)(state)}}$ for inputs and $t_{\text{(first two letters of functional block)(reference)(state)(signal)(state)}}$ for outputs. For example, $t_{LBIXKH1}$ symbolizes local bus timing (LB) for the input (I) to go invalid (X) with respect to the time the t_{LBK} clock reference (K) goes high (H), in this case for clock one (1). Also, t_{LBKHOX} symbolizes local bus timing (LB) for the t_{LBK} clock reference (K) to go high (H), with respect to the output (O) going invalid (X) or output hold time.
2. All timings are in reference to falling edge of LCLK0 (for all outputs and for $\overline{\text{LGTA}}$ and LUPWAIT inputs) or rising edge of LCLK0 (for all other inputs).
3. All signals are measured from $OV_{DD}/2$ of the rising/falling edge of LCLK0 to $0.4 \times OV_{DD}$ of the signal in question for 3.3-V signaling levels.
4. Input timings are measured at the pin.
5. $t_{LBOTOT1}$ should be used when RCWH[LALE] is not set and when the load on LALE output pin is at least 10 pF less than the load on LAD output pins.
6. $t_{LBOTOT2}$ should be used when RCWH[LALE] is set and when the load on LALE output pin is at least 10 pF less than the load on LAD output pins.
7. $t_{LBOTOT3}$ should be used when RCWH[LALE] is set and when the load on LALE output pin equals to the load on LAD output pins.
8. For purposes of active/float timing measurements, the Hi-Z or off-state is defined to be when the total current delivered through the component pin is less than or equal to the leakage current specification.
9. DLL bypass mode is not recommended for use at frequencies above 66 MHz.

This figure provides the AC test load for the local bus.

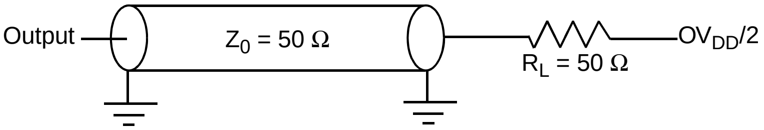

Figure 22. Local Bus C Test Load

This figure provides the AC test load for the I^2C.

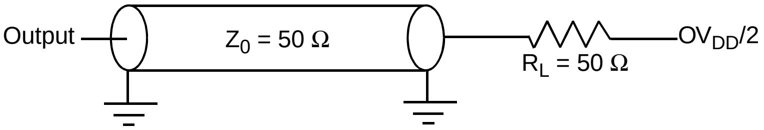

Figure 34. I^2C AC Test Load

This figure shows the AC timing diagram for the I^2C bus.

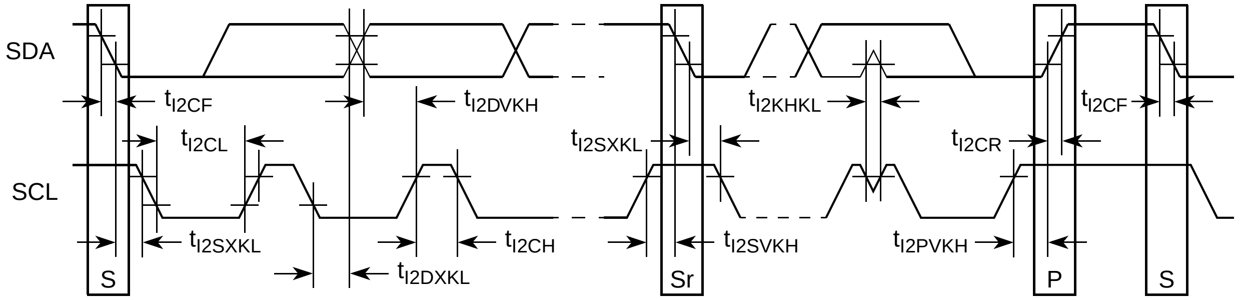

Figure 35. I^2C Bus AC Timing Diagram

12 PCI

This section describes the DC and AC electrical specifications for the PCI bus of the MPC8360E/58E.

12.1 PCI DC Electrical Characteristics

This table provides the DC electrical characteristics for the PCI interface of the device.

Table 46. PCI DC Electrical Characteristics

Parameter	Symbol	Test Condition	Min	Max	Unit
High-level input voltage	V_{IH}	$V_{OUT} \ge V_{OH}$ (min) or $V_{OUT} \le V_{OL}$ (max)	0.5 × OV_{DD}	OV_{DD} + 0.5	V
Low-level input voltage	V_{IL}		-0.5	0.3 × OV_{DD}	V
High-level output voltage	V_{OH}	I_{OH} = –500 μA	0.9 × OV_{DD}	—	V
Low-level output voltage	V_{OL}	I_{OL} = 1500 μA	—	0.1 × OV_{DD}	V
Input current	I_{IN}	0 V $\le V_{IN}^1 \le OV_{DD}$	—	±10	μA

12.2 PCI AC Electrical Specifications

This section describes the general AC timing parameters of the PCI bus of the device. Note that the PCI_CLK or PCI_SYNC_IN signal is used as the PCI input clock depending on whether the device is configured as a host or agent device. This table provides the PCI AC timing specifications at 66 MHz.

Table 47. PCI AC Timing Specifications at 66 MHz

Parameter	Symbol[1]	Min	Max	Unit	Notes
Clock to output valid	t_{PCKHOV}	—	6.0	ns	2, 5
Output hold from clock	t_{PCKHOX}	1	—	ns	2

20 Package and Pin Listings

This section details package parameters, pin assignments, and dimensions. The MPC8360E/58E is available in a tape ball grid array (TBGA), see Section 20.1, “Package Parameters for the TBGA Package,” and Section 20.2, “Mechanical Dimensions of the TBGA Package,” for information on the package.

20.1 Package Parameters for the TBGA Package

The package parameters for rev. 2.0 silicon are as provided in the following list. The package type is 37.5 mm × 37.5 mm, 740 tape ball grid array (TBGA).

Package outline	37.5 mm × 37.5 mm
Interconnects	740
Pitch	1.00 mm
Module height (typical)	1.46 mm
Solder Balls	62 Sn/36 Pb/2 Ag (ZU package)
	95.5 Sn/0.5 Cu/4Ag (VV package)
Ball diameter (typical)	0.64 mm

Table 66. MPC8360E TBGA Pinout Listing (continued)

Signal	Package Pin Number	Pin Type	Power Supply	Notes
PCI_DEVSEL/CE_PF[16]	E26	I/O	OV_{DD}	5
PCI_IDSEL/CE_PF[17]	F22	I/O	OV_{DD}	—
PCI_SERR/CE_PF[18]	B29	I/O	OV_{DD}	5
PCI_PERR/CE_PF[19]	A29	I/O	OV_{DD}	5
PCI_REQ[0]/CE_PF[20]	F19	I/O	$LV_{DD}2$	—
PCI_REQ[1]/CPCI_HS_ES/ CE_PF[21]	A21	I/O	$LV_{DD}2$	—
PCI_REQ[2]/CE_PF[22]	C21	I/O	$LV_{DD}2$	—
PCI_GNT[0]/CE_PF[23]	E20	I/O	$LV_{DD}2$	—
PCI_GNT[1]/CPCI1_HS_LED/ CE_PF[24]	B20	I/O	$LV_{DD}2$	—
PCI_GNT[2]/CPCI1_HS_ENUM/ CE_PF[25]	C20	I/O	$LV_{DD}2$	—
PCI_MODE	D36	I	OV_{DD}	—
M66EN/CE_PF[4]	B37	I/O	OV_{DD}	—
Local Bus Controller Interface				
LAD[0:31]	N32, N33, N35, N36, P37, P32, P34, R36, R35, R34, R33, T37, T35, T34, T33, U37, T32, U36, U34, V36, V35, W37, W35, V33, V32, W34, Y36, W32, AA37, Y33, AA35, AA34	I/O	OV_{DD}	—
LDP[0]/CKSTOP_OUT	AB37	I/O	OV_{DD}	—
LDP[1]/CKSTOP_IN	AB36	I/O	OV_{DD}	—
LDP[2]/LCS[6]	AB35	I/O	OV_{DD}	—
LDP[3]/LCS[7]	AA33	I/O	OV_{DD}	—
LA[27:31]	AC37, AA32, AC36, AC34, AD36	O	OV_{DD}	—
LCS[0:5]	AD33, AG37, AF34, AE33, AD32, AH37	O	OV_{DD}	—
LWE[0:3]/LSDDQM[0:3]/LBS[0:3]	AG35, AG34, AH36, AE32	O	OV_{DD}	—
LBCTL	AD35	O	OV_{DD}	—
LALE	M37	O	OV_{DD}	—
LGPL0/LSDA10/cfg_reset_source0	AB32	I/O	OV_{DD}	—
LGPL1/LSDWE/cfg_reset_source1	AE37	I/O	OV_{DD}	—
LGPL2/LSDRAS/LOE	AC33	O	OV_{DD}	—
LGPL3/LSDCAS/cfg_reset_source2	AD34	I/O	OV_{DD}	—
LGPL4/LGTA/LUPWAIT/LPBSE	AE35	I/O	OV_{DD}	—
LGPL5/cfg_clkin_div	AF36	I/O	OV_{DD}	—
LCKE	G36	O	OV_{DD}	—
LCLK[0]	J33	O	OV_{DD}	—
LCLK[1]/LCS[6]	J34	O	OV_{DD}	—

Table 66. MPC8360E TBGA Pinout Listing (continued)

Signal	Package Pin Number	Pin Type	Power Supply	Notes
LCLK[2]/LCS[7]	G37	O	OV_{DD}	—
LSYNC_OUT	F34	O	OV_{DD}	—
LSYNC_IN	G35	I	OV_{DD}	—
Programmable Interrupt Controller				
$\overline{\text{MCP\_OUT}}$	E34	O	OV_{DD}	2
$\overline{\text{IRQ0}}/\overline{\text{MCP\_IN}}$	C37	I	OV_{DD}	—
$\overline{\text{IRQ}}$[1]/M1SRCID[4]/M2SRCID[4]/ LSRCID[4]	F35	I/O	OV_{DD}	—
$\overline{\text{IRQ}}$[2]/M1DVAL/M2DVAL/LDVAL	F36	I/O	OV_{DD}	—
$\overline{\text{IRQ}}$[3]/CORE_SRESET	H34	I/O	OV_{DD}	—
$\overline{\text{IRQ}}$[4:5]	G33, G32	I/O	OV_{DD}	—
$\overline{\text{IRQ}}$[6]/$\overline{\text{LCS}}$[6]/$\overline{\text{CKSTOP\_OUT}}$	E35	I/O	OV_{DD}	—
$\overline{\text{IRQ}}$[7]/$\overline{\text{LCS}}$[7]/$\overline{\text{CKSTOP\_IN}}$	H36	I/O	OV_{DD}	—
DUART				
UART1_SOUT/M1SRCID[0]/ M2SRCID[0]/LSRCID[0]	E32	O	OV_{DD}	—
UART1_SIN/M1SRCID[1]/ M2SRCID[1]/LSRCID[1]	B34	I/O	OV_{DD}	—
$\overline{\text{UART1\_CTS}}$/M1SRCID[2]/ M2SRCID[2]/LSRCID[2]	C34	I/O	OV_{DD}	—
$\overline{\text{UART1\_RTS}}$/M1SRCID[3]/ M2SRCID[3]/LSRCID[3]	A35	O	OV_{DD}	—
I^2C Interface				
IIC1_SDA	D34	I/O	OV_{DD}	2
IIC1_SCL	B35	I/O	OV_{DD}	2
IIC2_SDA	E33	I/O	OV_{DD}	2
IIC2_SCL	C35	I/O	OV_{DD}	2
QUICC Engine Block				
CE_PA[0]	F8	I/O	LV_{DD0}	—
CE_PA[1:2]	AH1, AG5	I/O	OV_{DD}	—
CE_PA[3:7]	F6, D4, C3, E5, A3	I/O	$LV_{DD}0$	—
CE_PA[8]	AG3	I/O	OV_{DD}	—
CE_PA[9:12]	F7, B3, E6, B4	I/O	$LV_{DD}0$	—
CE_PA[13:14]	AG1, AF6	I/O	OV_{DD}	—
CE_PA[15]	B2	I/O	$LV_{DD}0$	—
CE_PA[16]	AF4	I/O	OV_{DD}	—
CE_PA[17:21]	B16, A16, E17, A17, B17	I/O	$LV_{DD}1$	—

Table 67. MPC8358E TBGA Pinout Listing (continued)

Signal	Package Pin Number	Pin Type	Power Supply	Notes
MEMC_MWE	AT26	O	GV_{DD}	—
MEMC_MRAS	AT29	O	GV_{DD}	—
MEMC_MCAS	AT24	O	GV_{DD}	—
MEMC_MCS[0:3]	AU27, AT27, AU8, AU7	O	GV_{DD}	—
MEMC_MCKE[0:1]	AL32, AU33	O	GV_{DD}	3
MEMC_MCK[0:5]	AK37, AT37, AN1, AR2, AN25, AK1	O	GV_{DD}	—
MEMC_MCK[0:5]	AL37, AT36, AP2, AT2, AN24, AL1	O	GV_{DD}	—
MDIC[0:1]	AH6, AP30	I/O	GV_{DD}	11
PCI				
PCI_INTA/IRQ_OUT/CE_PF[5]	A20	I/O	$LV_{DD}2$	2
PCI_RESET_OUT/CE_PF[6]	E19	I/O	$LV_{DD}2$	—
PCI_AD[31:30]/CE_PG[31:30]	D20, D21	I/O	$LV_{DD}2$	—
PCI_AD[29:25]/CE_PG[29:25]	A24, B23, C23, E23, A26	I/O	OV_{DD}	—
PCI_AD[24]/CE_PG[24]	B21	I/O	$LV_{DD}2$	—
PCI_AD[23:0]/CE_PG[23:0]	C24, C25, D25, B25, E24, F24, A27, A28, F27, A30, C30, D30, E29, B31, C31, D31, D32, A32, C33, B33, F30, E31, A34, D33	I/O	OV_{DD}	—
PCI_C/BE[3:0]/CE_PF[10:7]	E22, B26, E28, F28	I/O	OV_{DD}	—
PCI_PAR/CE_PF[11]	D28	I/O	OV_{DD}	—
PCI_FRAME/CE_PF[12]	D26	I/O	OV_{DD}	5
PCI_TRDY/CE_PF[13]	C27	I/O	OV_{DD}	5
PCI_IRDY/CE_PF[14]	C28	I/O	OV_{DD}	5
PCI_STOP/CE_PF[15]	B28	I/O	OV_{DD}	5
PCI_DEVSEL/CE_PF[16]	E26	I/O	OV_{DD}	5
PCI_IDSEL/CE_PF[17]	F22	I/O	OV_{DD}	—
PCI_SERR/CE_PF[18]	B29	I/O	OV_{DD}	5
PCI_PERR/CE_PF[19]	A29	I/O	OV_{DD}	5
PCI_REQ[0]/CE_PF[20]	F19	I/O	$LV_{DD}2$	—
PCI_REQ[1]/CPCI_HS_ES/CE_PF[21]	A21	I/O	$LV_{DD}2$	—
PCI_REQ[2]/CE_PF[22]	C21	I/O	$LV_{DD}2$	—
PCI_GNT[0]/CE_PF[23]	E20	I/O	$LV_{DD}2$	—
PCI_GNT[1]/CPCI1_HS_LED/CE_PF[24]	B20	I/O	$LV_{DD}2$	—
PCI_GNT[2]/CPCI1_HS_ENUM/CE_PF[25]	C20	I/O	$LV_{DD}2$	—

Table 67. MPC8358E TBGA Pinout Listing (continued)

Signal	Package Pin Number	Pin Type	Power Supply	Notes
$\overline{\text{PCI\_MODE}}$	D36	I	OV_{DD}	—
M66EN/CE_PF[4]	B37	I/O	OV_{DD}	—
Local Bus Controller Interface				
LAD[0:31]	N32, N33, N35, N36, P37, P32, P34, R36, R35, R34, R33, T37, T35, T34, T33, U37, T32, U36, U34, V36, V35, W37, W35, V33, V32, W34, Y36, W32, AA37, Y33, AA35, AA34	I/O	OV_{DD}	—
LDP[0]/$\overline{\text{CKSTOP\_OUT}}$	AB37	I/O	OV_{DD}	—
LDP[1]/$\overline{\text{CKSTOP\_IN}}$	AB36	I/O	OV_{DD}	—
LDP[2]/$\overline{\text{LCS}}$[6]	AB35	I/O	OV_{DD}	—
LDP[3]/$\overline{\text{LCS}}$[7]	AA33	I/O	OV_{DD}	—
LA[27:31]	AC37, AA32, AC36, AC34, AD36	O	OV_{DD}	—
$\overline{\text{LCS}}$[0:5]	AD33, AG37, AF34, AE33, AD32, AH37	O	OV_{DD}	—
$\overline{\text{LWE}}$[0:3]/LSDDQM[0:3]/$\overline{\text{LBS}}$[0:3]	AG35, AG34, AH36, AE32	O	OV_{DD}	—
LBCTL	AD35	O	OV_{DD}	—
LALE	M37	O	OV_{DD}	—
LGPL0/LSDA10/cfg_reset_source0	AB32	I/O	OV_{DD}	—
LGPL1/$\overline{\text{LSDWE}}$/cfg_reset_source1	AE37	I/O	OV_{DD}	—
LGPL2/$\overline{\text{LSDRAS}}$/$\overline{\text{LOE}}$	AC33	O	OV_{DD}	—
LGPL3/$\overline{\text{LSDCAS}}$/cfg_reset_source2	AD34	I/O	OV_{DD}	—
LGPL4/$\overline{\text{LGTA}}$/LUPWAIT/LPBSE	AE35	I/O	OV_{DD}	—
LGPL5/cfg_clkin_div	AF36	I/O	OV_{DD}	—
LCKE	G36	O	OV_{DD}	—
LCLK[0]	J33	O	OV_{DD}	—
LCLK[1]/LCS[6]	J34	O	OV_{DD}	—
LCLK[2]/LCS[7]	G37	O	OV_{DD}	—
LSYNC_OUT	F34	O	OV_{DD}	—
LSYNC_IN	G35	I	OV_{DD}	—
Programmable Interrupt Controller				
$\overline{\text{MCP\_OUT}}$	E34	O	OV_{DD}	2
$\overline{\text{IRQ0}}$/$\overline{\text{MCP\_IN}}$	C37	I	OV_{DD}	—
$\overline{\text{IRQ}}$[1]/M1SRCID[4]/M2SRCID[4]/LSRCID[4]	F35	I/O	OV_{DD}	—
$\overline{\text{IRQ}}$[2]/M1DVAL/M2DVAL/LDVAL	F36	I/O	OV_{DD}	—
$\overline{\text{IRQ}}$[3]/CORE_SRESET	H34	I/O	OV_{DD}	—

clock. When the device is configured as a PCI agent device the CLKIN and the CFG_CLKIN_DIV signals should be tied to GND.

When the device is configured as a PCI host device (RCWH[PCIHOST] = 1) and PCI clock output is disabled (RCWH[PCICKDRV] = 0), clock distribution and balancing done externally on the board. Therefore, PCI_SYNC_IN is the primary input clock.

As shown in Figure 54 and Figure 55, the primary clock input (frequency) is multiplied by the QUICC Engine block phase-locked loop (PLL), the system PLL, and the clock unit to create the QUICC Engine clock (*ce_clk*), the coherent system bus clock (*csb_clk*), the internal DDRC1 controller clock (*ddr1_clk*), and the internal clock for the local bus interface unit and DDR2 memory controller (*lb_clk*).

The *csb_clk* frequency is derived from a complex set of factors that can be simplified into the following equation:

$$csb\_clk = \{\text{PCI\_SYNC\_IN} \times (1 + \text{CFG\_CLKIN\_DIV})\} \times \text{SPMF}$$

In PCI host mode, PCI_SYNC_IN × (1 + CFG_CLKIN_DIV) is the CLKIN frequency; in PCI agent mode, CFG_CLKIN_DIV must be pulled down (low), so PCI_SYNC_IN × (1 + CFG_CLKIN_DIV) is the PCI_CLK frequency.

The *csb_clk* serves as the clock input to the e300 core. A second PLL inside the e300 core multiplies up the *csb_clk* frequency to create the internal clock for the e300 core (*core_clk*). The system and core PLL multipliers are selected by the SPMF and COREPLL fields in the reset configuration word low (RCWL) which is loaded at power-on reset or by one of the hard-coded reset options. See Chapter 4, “Reset, Clocking, and Initialization,” in the *MPC8360E PowerQUICC II Pro Integrated Communications Processor Reference Manual* for more information on the clock subsystem.

The *ce_clk* frequency is determined by the QUICC Engine PLL multiplication factor (RCWL[CEPMF) and the QUICC Engine PLL division factor (RCWL[CEPDF]) according to the following equation:

$$ce\_clk = (\text{primary clock input} \times \text{CEPMF}) \div (1 + \text{CEPDF})$$

The internal *ddr1_clk* frequency is determined by the following equation:

$$ddr1\_clk = csb\_clk \times (1 + \text{RCWL[DDR1CM]})$$

Note that the lb_clk clock frequency (for DDRC2) is determined by RCWL[LBCM]. The *internal ddr1_clk* frequency is not the external memory bus frequency; *ddr1_clk* passes through the DDRC1 clock divider (÷2) to create the differential DDRC1 memory bus clock outputs (MEMC1_MCK and $\overline{\text{MEMC1\_MCK}}$). However, the data rate is the same frequency as *ddr1_clk*.

The internal *lb_clk* frequency is determined by the following equation:

$$lb\_clk = csb\_clk \times (1 + \text{RCWL[LBCM]})$$

Note that *lb_clk* is not the external local bus or DDRC2 frequency; *lb_clk* passes through the a LB clock divider to create the external local bus clock outputs (LSYNC_OUT and LCLK[0:2]). The LB clock divider ratio is controlled by LCRR[CLKDIV].

Additionally, some of the internal units may be required to be shut off or operate at lower frequency than the *csb_clk* frequency. Those units have a default clock ratio that can be configured by a memory mapped register after the device comes out of reset. This table specifies which units have a configurable clock frequency.

Table 68. Configurable Clock Units

Unit	Default Frequency	Options
Security core	*csb_clk*/3	Off, *csb_clk*[1], *csb_clk*/2, *csb_clk*/3
PCI and DMA complex	csb_clk	Off, *csb_clk*

[1] With limitation, only for slow csb_clk rates, up to 166 MHz.

This table provides the operating frequencies for the TBGA package under recommended operating conditions (see Table 2). All frequency combinations shown in the table below may not be available. Maximum operating frequencies depend on the part

Table 72. CSB Frequency Options (continued)

CFG_CLKIN_DIV at Reset[1]	**SPMF**	***csb_clk*: Input Clock Ratio**[2]	**Input Clock Frequency (MHz)**[2]			
			16.67	**25**	**33.33**	**66.67**
			***csb_clk* Frequency (MHz)**			
Low	0110	6:1	100	150	200	
Low	0111	7:1	116	175	233	
Low	1000	8:1	133	200	266	
Low	1001	9:1	150	225	300	
Low	1010	10:1	166	250	333	
Low	1011	11:1	183	275		
Low	1100	12:1	200	300		
Low	1101	13:1	216	325		
Low	1110	14:1	233			
Low	1111	15:1	250			
Low	0000	16:1	266			
High	0010	2:1				133
High	0011	3:1			100	200
High	0100	4:1			133	266
High	0101	5:1			166	333
High	0110	6:1			200	
High	0111	7:1			233	
High	1000	8:1				
High	1001	9:1				
High	1010	10:1				
High	1011	11:1				
High	1100	12:1				
High	1101	13:1				
High	1110	14:1				
High	1111	15:1				
High	0000	16:1				

[1] CFG_CLKIN_DIV is only used for host mode; CLKIN must be tied low and CFG_CLKIN_DIV must be pulled down (low) in agent mode.

[2] CLKIN is the input clock in host mode; PCI_CLK is the input clock in agent mode.

21.2 Core PLL Configuration

RCWL[COREPLL] selects the ratio between the internal coherent system bus clock (*csb_clk*) and the e300 core clock (*core_clk*). This table shows the encodings for RCWL[COREPLL]. COREPLL values not listed in this table should be considered reserved.

Table 73. e300 Core PLL Configuration

RCWL[COREPLL]			***core_clk*:*csb_clk* Ratio**	**VCO divider**
0–1	**2–5**	**6**		
nn	**0000**	n	PLL bypassed (PLL off, *csb_clk* clocks core directly)	PLL bypassed (PLL off, *csb_clk* clocks core directly)
00	**0001**	0	1:1	÷2
01	**0001**	0	1:1	÷4
10	**0001**	0	1:1	÷8
11	**0001**	0	1:1	÷8
00	**0001**	1	1.5:1	÷2
01	**0001**	1	1.5:1	÷4
10	**0001**	1	1.5:1	÷8
11	**0001**	1	1.5:1	÷8
00	**0010**	0	2:1	÷2
01	**0010**	0	2:1	÷4
10	**0010**	0	2:1	÷8
11	**0010**	0	2:1	÷8
00	**0010**	1	2.5:1	÷2
01	**0010**	1	2.5:1	÷4
10	**0010**	1	2.5:1	÷8
11	**0010**	1	2.5:1	÷8
00	**0011**	0	3:1	÷2
01	**0011**	0	3:1	÷4
10	**0011**	0	3:1	÷8
11	**0011**	0	3:1	÷8

NOTE

Core VCO frequency = Core frequency × VCO divider. The VCO divider (RCWL[COREPLL[0:1]]) must be set properly so that the core VCO frequency is in the range of 800–1800 MHz. Having a core frequency below the CSB frequency is not a possible option because the core frequency must be equal to or greater than the CSB frequency.

Table 77. Package Thermal Characteristics for the TBGA Package (continued)

Characteristic	Symbol	Value	Unit	Notes
Junction-to-package natural convection on top	ψ_{JT}	1	° C/W	6

Notes

1. Junction temperature is a function of die size, on-chip power dissipation, package thermal resistance, mounting site (board) temperature, ambient temperature, airflow, power dissipation of other components on the board, and board thermal resistance.
2. Per JEDEC JESD51-2 and SEMI G38-87 with the single layer board horizontal.
3. Per JEDEC JESD51-6 with the board horizontal. 1 m/sec is approximately equal to 200 linear feet per minute (LFM).
4. Thermal resistance between the die and the printed-circuit board per JEDEC JESD51-8. Board temperature is measured on the top surface of the board near the package.
5. Thermal resistance between the die and the case top surface as measured by the cold plate method (MIL SPEC-883 Method 1012.1).
6. Thermal characterization parameter indicating the temperature difference between package top and the junction temperature per JEDEC JESD51-2. When Greek letters are not available, the thermal characterization parameter is written as Psi-JT.

22.2 Thermal Management Information

For the following sections, $P_D = (V_{DD} \times I_{DD}) + P_{I/O}$ where $P_{I/O}$ is the power dissipation of the I/O drivers. See Table 6 for typical power dissipations values.

22.2.1 Estimation of Junction Temperature with Junction-to-Ambient Thermal Resistance

An estimation of the chip junction temperature, T_J, can be obtained from the equation:

$$T_J = T_A + (R_{\theta JA} \times P_D)$$

where:

T_J = junction temperature (° C)
T_A = ambient temperature for the package (° C)
$R_{\theta JA}$ = junction-to-ambient thermal resistance (° C/W)
P_D = power dissipation in the package (W)

The junction-to-ambient thermal resistance is an industry standard value that provides a quick and easy estimation of thermal performance. As a general statement, the value obtained on a single-layer board is appropriate for a tightly packed printed-circuit board. The value obtained on the board with the internal planes is usually appropriate if the board has low power dissipation and the components are well separated. Test cases have demonstrated that errors of a factor of two (in the quantity $T_J – T_A$) are possible.

22.2.2 Estimation of Junction Temperature with Junction-to-Board Thermal Resistance

The thermal performance of a device cannot be adequately predicted from the junction-to-ambient thermal resistance. The thermal performance of any component is strongly dependent on the power dissipation of surrounding components. Additionally, the ambient temperature varies widely within the application. For many natural convection and especially closed box applications, the board temperature at the perimeter (edge) of the package is approximately the same as the local air temperature near the device. Specifying the local ambient conditions explicitly as the board temperature provides a more precise description of the local ambient conditions that determine the temperature of the device. At a known board temperature, the junction temperature is estimated using the following equation:

This figure shows the PLL power supply filter circuit.

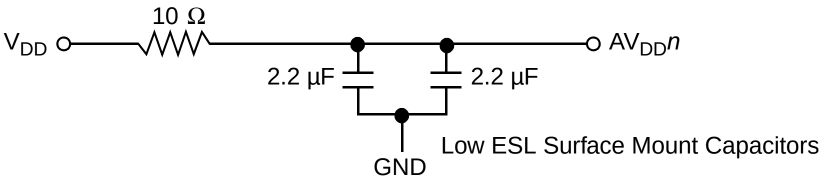

Figure 56. PLL Power Supply Filter Circuit

23.3 Decoupling Recommendations

Due to large address and data buses as well as high operating frequencies, the device can generate transient power surges and high frequency noise in its power supply, especially while driving large capacitive loads. This noise must be prevented from reaching other components in the device system, and the device itself requires a clean, tightly regulated source of power. Therefore, it is recommended that the system designer place at least one decoupling capacitor at each V_{DD}, OV_{DD}, GV_{DD}, and LV_{DD} pins of the device. These decoupling capacitors should receive their power from separate V_{DD}, OV_{DD}, GV_{DD}, LV_{DD}, and GND power planes in the PCB, utilizing short traces to minimize inductance. Capacitors may be placed directly under the device using a standard escape pattern. Others may surround the part.

These capacitors should have a value of 0.01 or 0.1 µF. Only ceramic SMT (surface mount technology) capacitors should be used to minimize lead inductance, preferably 0402 or 0603 sizes.

Additionally, it is recommended that there be several bulk storage capacitors distributed around the PCB, feeding the V_{DD}, OV_{DD}, GV_{DD}, and LV_{DD} planes, to enable quick recharging of the smaller chip capacitors. These bulk capacitors should have a low ESR (equivalent series resistance) rating to ensure the quick response time necessary. They should also be connected to the power and ground planes through two vias to minimize inductance. Suggested bulk capacitors—100–330 µF (AVX TPS tantalum or Sanyo OSCON).

23.4 Connection Recommendations

To ensure reliable operation, it is highly recommended to connect unused inputs to an appropriate signal level. Unused active low inputs should be tied to OV_{DD}, GV_{DD}, or LV_{DD} as required. Unused active high inputs should be connected to GND. All NC (no-connect) signals must remain unconnected.

Power and ground connections must be made to all external V_{DD}, GV_{DD}, LV_{DD}, OV_{DD}, and GND pins of the device.

23.5 Output Buffer DC Impedance

The device drivers are characterized over process, voltage, and temperature. For all buses, the driver is a push-pull single-ended driver type (open drain for I^2C).

To measure Z_0 for the single-ended drivers, an external resistor is connected from the chip pad to OV_{DD} or GND. Then, the value of each resistor is varied until the pad voltage is $OV_{DD}/2$ (see Figure 57). The output impedance is the average of two components, the resistances of the pull-up and pull-down devices. When data is held high, SW1 is closed (SW2 is open) and R_P is trimmed until the voltage at the pad equals $OV_{DD}/2$. R_P then becomes the resistance of the pull-up devices. R_P and R_N are designed to be close to each other in value. Then, $Z_0 = (R_P + R_N)/2$.

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Freescale Semiconductor Japan Ltd.
Headquarters
ARCO Tower 15F
1-8-1, Shimo-Meguro, Meguro-ku
Tokyo 153-0064
Japan
0120 191014 or
+81 3 5437 9125
support.japan@freescale.com

Asia/Pacific:
Freescale Semiconductor China Ltd.
Exchange Building 23F
No. 118 Jianguo Road
Chaoyang District
Beijing 100022
China
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