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Understanding Embedded - Microprocessors

Embedded microprocessors are specialized computing chips designed to perform specific tasks within an embedded system. Unlike general-purpose microprocessors found in personal computers, embedded microprocessors are tailored for dedicated functions within larger systems, offering optimized performance, efficiency, and reliability. These microprocessors are integral to the operation of countless electronic devices, providing the computational power necessary for controlling processes, handling data, and managing communications.

Applications of Embedded - Microprocessors

Embedded microprocessors are utilized across a broad spectrum of applications, making them indispensable in

Details	
Product Status	Obsolete
Core Processor	PowerPC e300
Number of Cores/Bus Width	1 Core, 32-Bit
Speed	400MHz
Co-Processors/DSP	Communications; QUICC Engine
RAM Controllers	DDR, DDR2
Graphics Acceleration	No
Display & Interface Controllers	-
Ethernet	10/100/1000Mbps (1)
SATA	-
USB	USB 1.x (1)
Voltage - I/O	1.8V, 2.5V, 3.3V
Operating Temperature	0°C ~ 105°C (TA)
Security Features	-
Package / Case	740-LBGA
Supplier Device Package	740-TBGA (37.5x37.5)
Purchase URL	https://www.e-xfl.com/product-detail/nxp-semiconductors/kmpc8358zuagdga

2.2.1 Power-Up Sequencing

MPC8360E/58E does not require the core supply voltage (V_{DD} and AV_{DD}) and I/O supply voltages (GV_{DD}, LV_{DD}, and OV_{DD}) to be applied in any particular order. During the power ramp up, before the power supplies are stable and if the I/O voltages are supplied before the core voltage, there may be a period of time that all input and output pins are actively be driven and cause contention and excessive current from 3A to 5A. In order to avoid actively driving the I/O pins and to eliminate excessive current draw, apply the core voltage (V_{DD}) before the I/O voltage (GV_{DD}, LV_{DD}, and OV_{DD}) and assert $\overline{\text{PORESET}}$ before the power supplies fully ramp up. In the case where the core voltage is applied first, the core voltage supply must rise to 90% of its nominal value before the I/O supplies reach 0.7 V, see this figure.

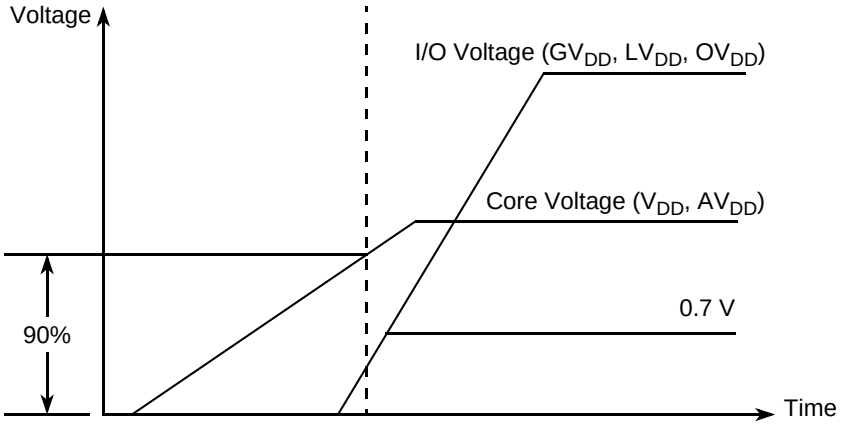


Figure 5. Power Sequencing Example

I/O voltage supplies (GV_{DD}, LV_{DD}, and OV_{DD}) do not have any ordering requirements with respect to one another.

2.2.2 Power-Down Sequencing

The MPC8360E/58E does not require the core supply voltage and I/O supply voltages to be powered down in any particular order.

3 Power Characteristics

The estimated typical power dissipation values are shown in these tables.

Table 4. MPC8360E TBGA Core Power Dissipation[1]

Core Frequency (MHz)	CSB Frequency (MHz)	QUICC Engine Frequency (MHz)	Typical	Maximum	Unit	Notes
266	266	500	5.0	5.6	W	2, 3, 5
400	266	400	4.5	5.0	W	2, 3, 4
533	266	400	4.8	5.3	W	2, 3, 4
667	333	400	5.8	6.3	W	3, 6, 7, 8
500	333	500	5.9	6.4	W	3, 6, 7, 8

5.3 QUICC Engine Block Operating Frequency Limitations

This section specify the limits of the AC electrical characteristics for the operation of the QUICC Engine block's communication interfaces.

NOTE

The settings listed below are required for correct hardware interface operation. Each protocol by itself requires a minimal QUICC Engine block operating frequency setting for meeting the performance target. Because the performance is a complex function of all the QUICC Engine block settings, the user should make use of the QUICC Engine block performance utility tool provided by Freescale to validate their system.

This table lists the maximal QUICC Engine block I/O frequencies and the minimal QUICC Engine block core frequency for each interface.

Table 13. QUICC Engine Block Operating Frequency Limitations

Interface	Interface Operating Frequency (MHz)	Max Interface Bit Rate (Mbps)	Min QUICC Engine Operating Frequency[1] (MHz)	Notes
Ethernet Management: MDC/MDIO	10 (max)	10	20	—
MII	25 (typ)	100	50	—
RMII	50 (typ)	100	50	—
GMII/RGMII/TBI/RTBI	125 (typ)	1000	250	—
SPI (master/slave)	10 (max)	10	20	—
UCC through TDM	50 (max)	70	8 × F	2
MCC	25 (max)	16.67	16 × F	2, 4
UTOPIA L2	50 (max)	800	2 × F	2
POS-PHY L2	50 (max)	800	2 × F	2
HDLC bus	10 (max)	10	20	—
HDLC/transparent	50 (max)	50	8/3 × F	2, 3
UART/async HDLC	3.68 (max internal ref clock)	115 (Kbps)	20	—
BISYNC	2 (max)	2	20	—
USB	48 (ref clock)	12	96	—

Notes:

1. The QUICC Engine module needs to run at a frequency higher than or equal to what is listed in this table.
2. 'F' is the actual interface operating frequency.\
3. The bit rate limit is independent of the data bus width (that is, the same for serial, nibble, or octal interfaces).
4. TDM in high-speed mode for serial data interface.

6 DDR and DDR2 SDRAM

This section describes the DC and AC electrical specifications for the DDR and DDR2 SDRAM interface of the MPC8360E/58E.

8.2.1.1 GMII Transmit AC Timing Specifications

This table provides the GMII transmit AC timing specifications.

Table 27. GMII Transmit AC Timing Specifications

At recommended operating conditions with LV_{DD}/OV_{DD} of 3.3 V ± 10%.

Parameter/Condition	Symbol[1]	Min	Typ	Max	Unit	Notes
GTX_CLK clock period	t_{GTX}	—	8.0	—	ns	—
GTX_CLK duty cycle	$t_{GTXH/tGTX}$	40	—	60	%	—
GTX_CLK to GMII data TXD[7:0], TX_ER, TX_EN delay	t_{GTKHDX} t_{GTKHDV}	0.5 —	—	— 5.0	ns	3
GTX_CLK clock rise time, (20% to 80%)	t_{GTXR}	—	—	1.0	ns	—
GTX_CLK clock fall time, (80% to 20%)	t_{GTXF}	—	—	1.0	ns	—
GTX_CLK125 clock period	t_{G125}	—	8.0	—	ns	2
GTX_CLK125 reference clock duty cycle measured at $LV_{DD/2}$	t_{G125H}/t_{G125}	45	—	55	%	2

Notes:

1. The symbols used for timing specifications follow the pattern $t_{\text{(first two letters of functional block)(signal)(state)(reference)(state)}}$ for inputs and $t_{\text{(first two letters of functional block)(reference)(state)(signal)(state)}}$ for outputs. For example, t_{GTKHDV} symbolizes GMII transmit timing (GT) with respect to the t_{GTX} clock reference (K) going to the high state (H) relative to the time date input signals (D) reaching the valid state (V) to state or setup time. Also, t_{GTKHDX} symbolizes GMII transmit timing (GT) with respect to the t_{GTX} clock reference (K) going to the high state (H) relative to the time date input signals (D) going invalid (X) or hold time. Note that, in general, the clock reference symbol representation is based on three letters representing the clock of a particular functional. For example, the subscript of t_{GTX} represents the GMII(G) transmit (TX) clock. For rise and fall times, the latter convention is used with the appropriate letter: R (rise) or F (fall).
2. This symbol is used to represent the external GTX_CLK125 signal and does not follow the original symbol naming convention.
3. In rev. 2.0 silicon, due to errata, t_{GTKHDX} minimum and t_{GTKHDV} maximum are not supported when the GTX_CLK is selected. Refer to Errata *QE_ENET18* in *Chip Errata for the MPC8360E, Rev. 1*.

This figure shows the GMII transmit AC timing diagram.

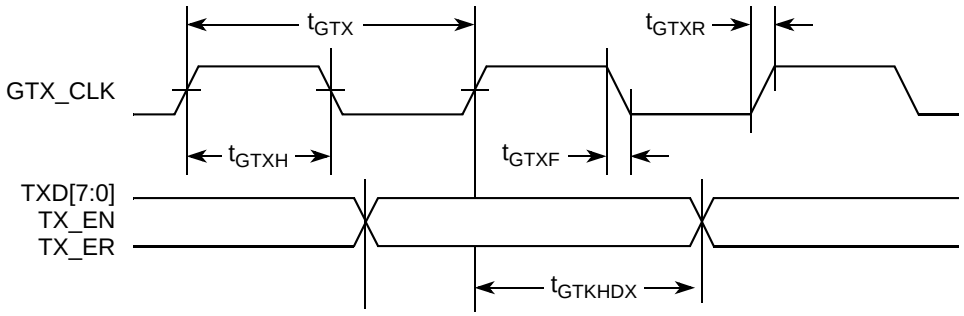


Figure 10. GMII Transmit AC Timing Diagram

8.2.1.2 GMII Receive AC Timing Specifications

This table provides the GMII receive AC timing specifications.

Table 28. GMII Receive AC Timing Specifications

At recommended operating conditions with LV_{DD}/OV_{DD} of 3.3 V ± 10%.

Parameter/Condition	Symbol[1]	Min	Typ	Max	Unit	Notes
RX_CLK clock period	t_{GRX}	—	8.0	—	ns	—
RX_CLK duty cycle	t_{GRXH}/t_{GRX}	40	—	60	%	—
RXD[7:0], RX_DV, RX_ER setup time to RX_CLK	t_{GRDVKH}	2.0	—	—	ns	—
RXD[7:0], RX_DV, RX_ER hold time to RX_CLK	t_{GRDXKH}	0.2	—	—	ns	2
RX_CLK clock rise time, (20% to 80%)	t_{GRXR}	—	—	1.0	ns	—
RX_CLK clock fall time, (80% to 20%)	t_{GRXF}	—	—	1.0	ns	—

Notes:

1. The symbols used for timing specifications follow the pattern of $t_{\text{(first two letters of functional block)(signal)(state)(reference)(state)}}$ for inputs and $t_{\text{(first two letters of functional block)(reference)(state)(signal)(state)}}$ for outputs. For example, t_{GRDVKH} symbolizes GMII receive timing (GR) with respect to the time data input signals (D) reaching the valid state (V) relative to the t_{RX} clock reference (K) going to the high state (H) or setup time. Also, t_{GRDXKL} symbolizes GMII receive timing (GR) with respect to the time data input signals (D) went invalid (X) relative to the t_{GRX} clock reference (K) going to the low (L) state or hold time. Note that, in general, the clock reference symbol representation is based on three letters representing the clock of a particular functional. For example, the subscript of t_{GRX} represents the GMII (G) receive (RX) clock. For rise and fall times, the latter convention is used with the appropriate letter: R (rise) or F (fall).
2. In rev. 2.0 silicon, due to errata, t_{GRDXKH} minimum is 0.5 which is not compliant with the standard. Refer to Errata *QE_ENET18* in *Chip Errata for the MPC8360E, Rev. 1*.

This figure shows the GMII receive AC timing diagram.

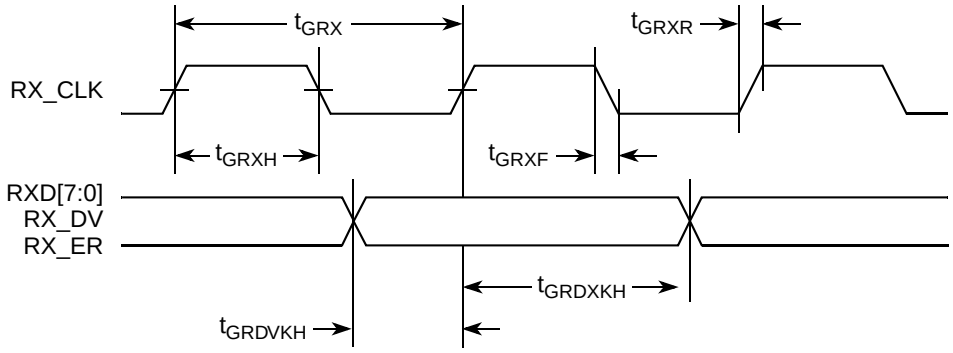


Figure 11. GMII Receive AC Timing Diagram

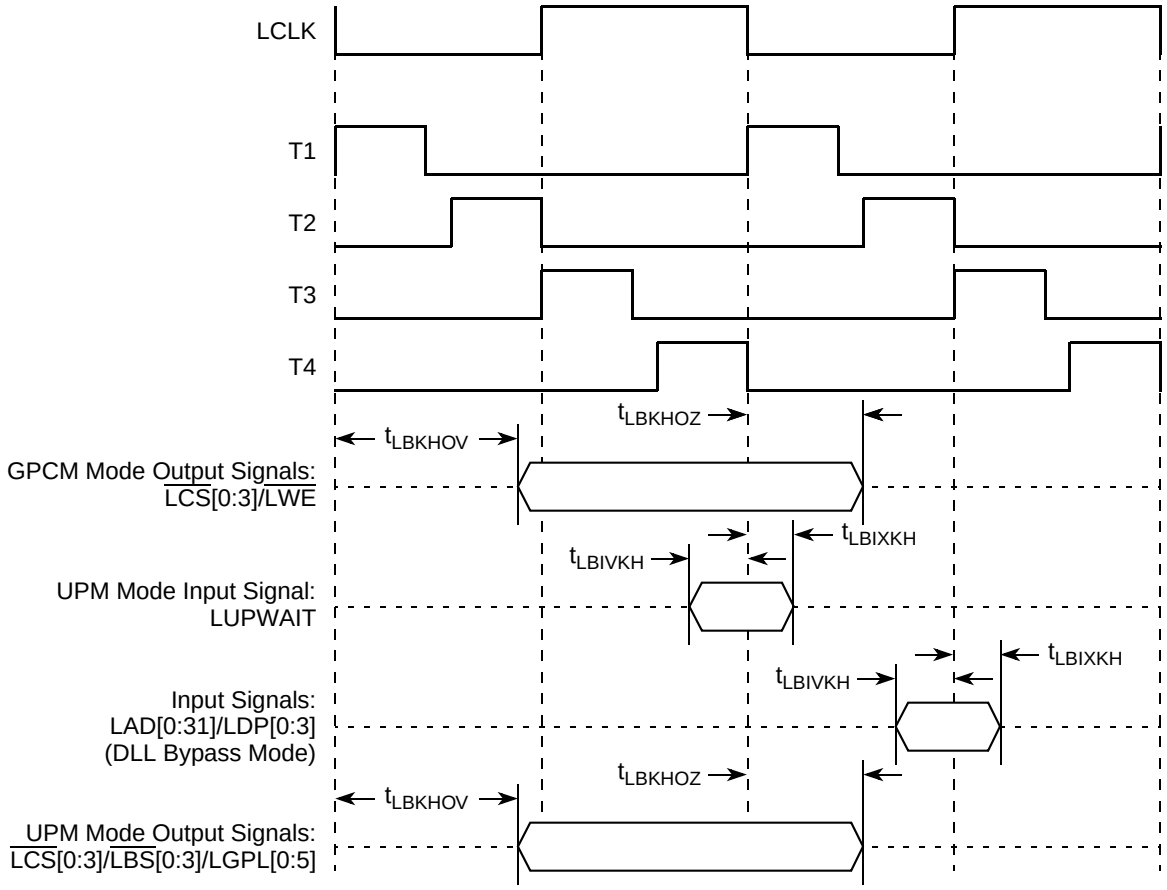


Figure 27. Local Bus Signals, GPCM/UPM Signals for LCRR[CLKDIV] = 4 (DLL Bypass Mode)

10.2 JTAG AC Electrical Characteristics

This section describes the AC electrical specifications for the IEEE 1149.1 (JTAG) interface of the device.

This table provides the JTAG AC timing specifications as defined in Figure 30 through Figure 33.

Table 43. JTAG AC Timing Specifications (Independent of CLKIN)[1]

At recommended operating conditions (see Table 2).

Parameter	Symbol [2]	Min	Max	Unit	Notes
JTAG external clock frequency of operation	f_{JTG}	0	33.3	MHz	—
JTAG external clock cycle time	t_{JTG}	30	—	ns	—
JTAG external clock duty cycle	t_{JTKHKL}/t_{JTG}	45	55	%	—
JTAG external clock rise and fall times	t_{JTGR} & t_{JTGF}	0	2	ns	—
$\overline{\text{TRST}}$ assert time	t_{TRST}	25	—	ns	3
Input setup times: Boundary-scan data TMS, TDI	 t_{JTDVKH} t_{JTIVKH}	 4 4	 — —	ns	4
Input hold times: Boundary-scan data TMS, TDI	 t_{JTDXKH} t_{JTIXKH}	 10 10	 — —	ns	4
Valid times: Boundary-scan data TDO	 t_{JTKLDV} t_{JTKLOV}	 2 2	 11 11	ns	5
Output hold times: Boundary-scan data TDO	 t_{JTKLDX} t_{JTKLOX}	 2 2	 — —	ns	5
JTAG external clock to output high impedance: Boundary-scan data TDO	 t_{JTKLDZ} t_{JTKLOZ}	 2 2	 19 9	ns	5, 6

Notes:

1. All outputs are measured from the midpoint voltage of the falling/rising edge of t_{TCLK} to the midpoint of the signal in question. The output timings are measured at the pins. All output timings assume a purely resistive 50-Ω load (see Figure 22). Time-of-flight delays must be added for trace lengths, vias, and connectors in the system.
2. The symbols used for timing specifications herein follow the pattern of $t_{\text{(first two letters of functional block)(signal)(state)(reference)(state)}}$ for inputs and $t_{\text{(first two letters of functional block)(reference)(state)(signal)(state)}}$ for outputs. For example, t_{JTDVKH} symbolizes JTAG device timing (JT) with respect to the time data input signals (D) reaching the valid state (V) relative to the t_{JTG} clock reference (K) going to the high (H) state or setup time. Also, t_{JTDXKH} symbolizes JTAG timing (JT) with respect to the time data input signals (D) went invalid (X) relative to the t_{JTG} clock reference (K) going to the high (H) state. Note that, in general, the clock reference symbol representation is based on three letters representing the clock of a particular functional. For rise and fall times, the latter convention is used with the appropriate letter: R (rise) or F (fall).
3. $\overline{\text{TRST}}$ is an asynchronous level sensitive signal. The setup time is for test purposes only.
4. Non-JTAG signal input timing with respect to t_{TCLK}.
5. Non-JTAG signal output timing with respect to t_{TCLK}.
6. Guaranteed by design and characterization.

11.2 I²C AC Electrical Specifications

This table provides the AC timing parameters for the I²C interface of the device.

Table 45. I²C AC Electrical Specifications

All values refer to V_{IH} (min) and V_{IL} (max) levels (see Table 44).

Parameter	Symbol[1]	Min	Max	Unit	Note
SCL clock frequency	f_{I2C}	0	400	kHz	2
Low period of the SCL clock	t_{I2CL}	1.3	—	μs	—
High period of the SCL clock	t_{I2CH}	0.6	—	μs	—
Setup time for a repeated START condition	t_{I2SVKH}	0.6	—	μs	—
Hold time (repeated) START condition (after this period, the first clock pulse is generated)	t_{I2SXKL}	0.6	—	μs	—
Data setup time	t_{I2DVKH}	100	—	ns	3
Data hold time: CBUS compatible masters I²C bus devices	t_{I2DXKL}	 — 0[2]	 — 0.9[3]	μs	—
Rise time of both SDA and SCL signals	t_{I2CR}	20 + 0.1 C_b[4]	300	ns	—
Fall time of both SDA and SCL signals	t_{I2CF}	20 + 0.1 C_b[4]	300	ns	—
Set-up time for STOP condition	t_{I2PVKH}	0.6	—	μs	—
Bus free time between a STOP and START condition	t_{I2KHDX}	1.3	—	μs	—
Noise margin at the LOW level for each connected device (including hysteresis)	V_{NL}	0.1 × OV_{DD}	—	V	—
Noise margin at the HIGH level for each connected device (including hysteresis)	V_{NH}	0.2 × OV_{DD}	—	V	—

Notes:

1. The symbols used for timing specifications follow the pattern of $t_{\text{(first two letters of functional block)(signal)(state)(reference)(state)}}$ for inputs and $t_{\text{(first two letters of functional block)(reference)(state)(signal)(state)}}$ for outputs. For example, t_{I2DVKH} symbolizes I²C timing (I2) with respect to the time data input signals (D) reach the valid state (V) relative to the t_{I2C} clock reference (K) going to the high (H) state or setup time. Also, t_{I2SXKL} symbolizes I²C timing (I2) for the time that the data with respect to the start condition (S) went invalid (X) relative to the t_{I2C} clock reference (K) going to the low (L) state or hold time. Also, t_{I2PVKH} symbolizes I²C timing (I2) for the time that the data with respect to the stop condition (P) reaching the valid state (V) relative to the t_{I2C} clock reference (K) going to the high (H) state or setup time. For rise and fall times, the latter convention is used with the appropriate letter: R (rise) or F (fall).
2. The device provides a hold time of at least 300 ns for the SDA signal (referred to the V_{IH} min of the SCL signal) to bridge the undefined region of the falling edge of SCL.
3. The maximum t_{I2DVKH} has only to be met if the device does not stretch the LOW period (t_{I2CL}) of the SCL signal.
4. C_B = capacitance of one bus line in pF.

13.2 Timers AC Timing Specifications

This table provides the timer input and output AC timing specifications.

Table 50. Timers Input AC Timing Specifications[1]

Characteristic	Symbol[2]	Typ	Unit
Timers inputs—minimum pulse width	t_{TIWID}	20	ns

Notes:
1. Input specifications are measured from the 50% level of the signal to the 50% level of the rising edge of CLKIN. Timings are measured at the pin.
2. Timers inputs and outputs are asynchronous to any visible clock. Timers outputs should be synchronized before use by any external synchronous logic. Timers inputs are required to be valid for at least t_{TIWID} ns to ensure proper operation.

This figure provides the AC test load for the timers.

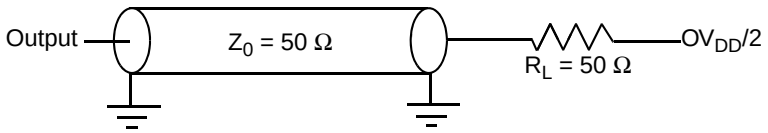


Figure 39. Timers AC Test Load

14 GPIO

This section describes the DC and AC electrical specifications for the GPIO of the MPC8360E/58E.

14.1 GPIO DC Electrical Characteristics

This table provides the DC electrical characteristics for the device GPIO.

Table 51. GPIO DC Electrical Characteristics

Characteristic	Symbol	Condition	Min	Max	Unit	Notes
Output high voltage	V_{OH}	I_{OH} = –6.0 mA	2.4	—	V	1
Output low voltage	V_{OL}	I_{OL} = 6.0 mA	—	0.5	V	1
Output low voltage	V_{OL}	I_{OL} = 3.2 mA	—	0.4	V	1
Input high voltage	V_{IH}	—	2.0	OV_{DD} + 0.3	V	1
Input low voltage	V_{IL}	—	–0.3	0.8	V	—
Input current	I_{IN}	0 V ≤V_{IN} ≤OV_{DD}	—	±10	μA	—

Note:
1. This specification applies when operating from 3.3-V supply.

15.2 IPIC AC Timing Specifications

This table provides the IPIC input and output AC timing specifications.

Table 54. IPIC Input AC Timing Specifications[1]

Characteristic	Symbol[2]	Min	Unit
IPIC inputs—minimum pulse width	t_{PIWID}	20	ns

Notes:
1. Input specifications are measured from the 50% level of the signal to the 50% level of the rising edge of CLKIN. Timings are measured at the pin.
2. IPIC inputs and outputs are asynchronous to any visible clock. IPIC outputs should be synchronized before use by any external synchronous logic. IPIC inputs are required to be valid for at least t_{PIWID} ns to ensure proper operation when working in edge triggered mode.

16 SPI

This section describes the DC and AC electrical specifications for the SPI of the MPC8360E/58E.

16.1 SPI DC Electrical Characteristics

This table provides the DC electrical characteristics for the device SPI.

Table 55. SPI DC Electrical Characteristics

Characteristic	Symbol	Condition	Min	Max	Unit
Output high voltage	V_{OH}	I_{OH} = –6.0 mA	2.4	—	V
Output low voltage	V_{OL}	I_{OL} = 6.0 mA	—	0.5	V
Output low voltage	V_{OL}	I_{OL} = 3.2 mA	—	0.4	V
Input high voltage	V_{IH}	—	2.0	OV_{DD} + 0.3	V
Input low voltage	V_{IL}	—	–0.3	0.8	V
Input current	I_{IN}	0 V ≤V_{IN} ≤OV_{DD}	—	±10	μA

16.2 SPI AC Timing Specifications

This table and provide the SPI input and output AC timing specifications.

Table 56. SPI AC Timing Specifications[1]

Characteristic	Symbol[2]	Min	Max	Unit
SPI outputs—Master mode (internal clock) delay	t_{NIKHOX} t_{NIKHOV}	0.3 —	— 8	ns
SPI outputs—Slave mode (external clock) delay	t_{NEKHOX} t_{NEKHOV}	2 —	— 8	ns
SPI inputs—Master mode (internal clock) input setup time	t_{NIIVKH}	8	—	ns
SPI inputs—Master mode (internal clock) input hold time	t_{NIIXKH}	0	—	ns
SPI inputs—Slave mode (external clock) input setup time	t_{NEIVKH}	4	—	ns

Table 56. SPI AC Timing Specifications[1]

Characteristic	Symbol[2]	Min	Max	Unit
SPI inputs—Slave mode (external clock) input hold time	t_{NEIXKH}	2	—	ns

Notes:
1. Output specifications are measured from the 50% level of the rising edge of CLKIN to the 50% level of the signal. Timings are measured at the pin.
2. The symbols used for timing specifications follow the pattern of $t_{\text{(first two letters of functional block)(signal)(state)(reference)(state)}}$ for inputs and $t_{\text{(first two letters of functional block)(reference)(state)(signal)(state)}}$ for outputs. For example, t_{NIKHOV} symbolizes the NMSI outputs internal timing (NI) for the time t_{SPI} memory clock reference (K) goes from the high state (H) until outputs (O) are valid (V).

This figure provides the AC test load for the SPI.

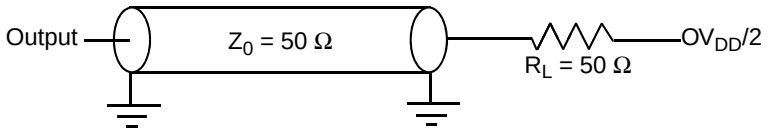
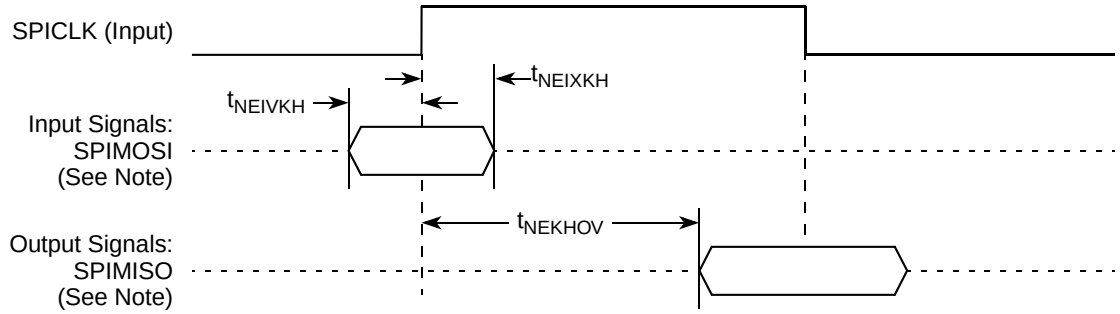


Figure 41. SPI AC Test Load

These figures represent the AC timing from Table 56. Note that although the specifications generally reference the rising edge of the clock, these AC timing diagrams also apply when the falling edge is the active edge.

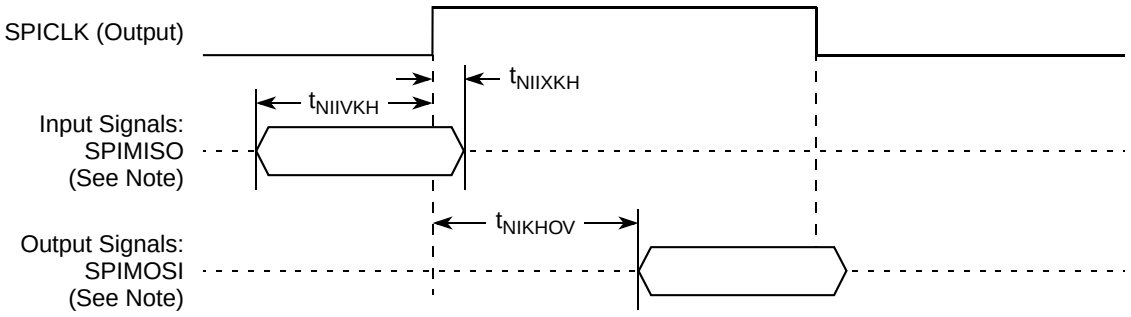
This figure shows the SPI timing in slave mode (external clock).



Note: The clock edge is selectable on SPI.

Figure 42. SPI AC Timing in Slave Mode (External Clock) Diagram

This figure shows the SPI timing in Master mode (internal clock).



Note: The clock edge is selectable on SPI.

Figure 43. SPI AC Timing in Master Mode (Internal Clock) Diagram

Table 60. UTOPIA AC Timing Specifications[1] (continued)

Characteristic	Symbol[2]	Min	Max	Unit	Notes
UTOPIA inputs—Internal clock input hold time	t_{UIIXKH}	2.4	—	ns	—
UTOPIA inputs—External clock input hold time	t_{UEIXKH}	1	—	ns	3

Notes:

1. Output specifications are measured from the 50% level of the rising edge of CLKIN to the 50% level of the signal. Timings are measured at the pin.
2. The symbols used for timing specifications follow the pattern of $t_{\text{(first two letters of functional block)(signal)(state)(reference)(state)}}$ for inputs and $t_{\text{(first two letters of functional block)(reference)(state)(signal)(state)}}$ for outputs. For example, t_{UIKHOX} symbolizes the UTOPIA outputs internal timing (UI) for the time t_{UTOPIA} memory clock reference (K) goes from the high state (H) until outputs (O) are invalid (X).
3. In rev. 2.0 silicon, due to errata, t_{UEIVKH} minimum is 4.3 ns and t_{UEIXKH} minimum is 1.4 ns under specific conditions. Refer to Errata QE_UPC3 in *Chip Errata for the MPC8360E, Rev. 1*.

This figure provides the AC test load for the UTOPIA.

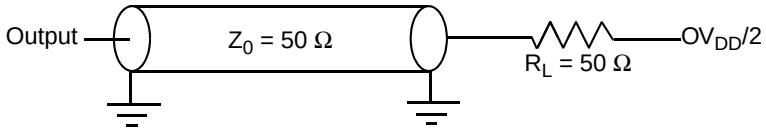


Figure 46. UTOPIA AC Test Load

These figures represent the AC timing from Table 56. Note that although the specifications generally reference the rising edge of the clock, these AC timing diagrams also apply when the falling edge is the active edge.

This figure shows the UTOPIA timing with external clock.

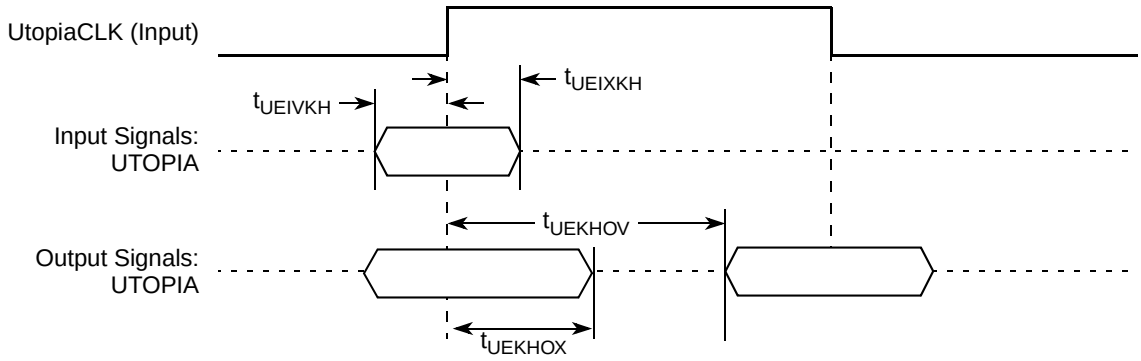


Figure 47. UTOPIA AC Timing (External Clock) Diagram

19 USB

This section provides the AC and DC electrical specifications for the USB interface of the MPC8360E/58E.

19.1 USB DC Electrical Characteristics

This table provides the DC electrical characteristics for the USB interface.

Table 64. USB DC Electrical Characteristics

Parameter	Symbol	Min	Max	Unit
High-level input voltage	V_{IH}	2	OV_{DD} + 0.3	V
Low-level input voltage	V_{IL}	–0.3	0.8	V
High-level output voltage, I_{OH} = –100 µA	V_{OH}	OV_{DD} – 0.4	—	V
Low-level output voltage, I_{OL} = 100 µA	V_{OL}	—	0.2	V
Input current	I_{IN}	—	±10	µA

19.2 USB AC Electrical Specifications

This table describes the general timing parameters of the USB interface of the device.

Table 65. USB General Timing Parameters

Parameter	Symbol[1]	Min	Max	Unit	Notes	Note
USB clock cycle time	t_{USCK}	20.83	—	ns	Full speed 48 MHz	—
USB clock cycle time	t_{USCK}	166.67	—	ns	Low speed 6 MHz	—
Skew between TXP and TXN	t_{USTSPN}	—	5	ns	—	2
Skew among RXP, RXN, and RXD	$t_{USRSPND}$	—	10	ns	Full speed transitions	2
Skew among RXP, RXN, and RXD	t_{USRPND}	—	100	ns	Low speed transitions	2

Notes:

1. The symbols used for timing specifications follow the pattern of $t_{\text{(first two letters of functional block)(state)(signal)}}$ for receive signals and $t_{\text{(first two letters of functional block)(state)(signal)}}$ for transmit signals. For example, $t_{USRSPND}$ symbolizes USB timing (US) for the USB receive signals skew (RS) among RXP, RXN, and RXD (PND). Also, t_{USTSPN} symbolizes USB timing (US) for the USB transmit signals skew (TS) between TXP and TXN (PN).
2. Skew measurements are done at $OV_{DD}/2$ of the rising or falling edge of the signals.

This figure provide the AC test load for the USB.

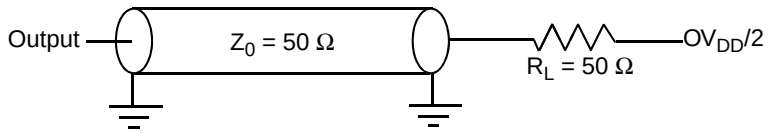


Figure 52. USB AC Test Load

20 Package and Pin Listings

This section details package parameters, pin assignments, and dimensions. The MPC8360E/58E is available in a tape ball grid array (TBGA), see Section 20.1, “Package Parameters for the TBGA Package,” and Section 20.2, “Mechanical Dimensions of the TBGA Package,” for information on the package.

20.1 Package Parameters for the TBGA Package

The package parameters for rev. 2.0 silicon are as provided in the following list. The package type is 37.5 mm × 37.5 mm, 740 tape ball grid array (TBGA).

Package outline	37.5 mm × 37.5 mm
Interconnects	740
Pitch	1.00 mm
Module height (typical)	1.46 mm
Solder Balls	62 Sn/36 Pb/2 Ag (ZU package) 95.5 Sn/0.5 Cu/4Ag (VV package)
Ball diameter (typical)	0.64 mm

20.3 Pinout Listings

Refer to AN3097, “MPC8360/MPC8358E PowerQUICC Design Checklist,” for proper pin termination and usage.

This table shows the pin list of the MPC8360E TBGA package.

Table 66. MPC8360E TBGA Pinout Listing

Signal	Package Pin Number	Pin Type	Power Supply	Notes
Primary DDR SDRAM Memory Controller Interface				
MEMC1_MDQ[0:31]	AJ34, AK33, AL33, AL35, AJ33, AK34, AK32, AM36, AN37, AN35, AR34, AT34, AP37, AP36, AR36, AT35, AP34, AR32, AP32, AM31, AN33, AM34, AM33, AM30, AP31, AM27, AR30, AT32, AN29, AP29, AN27, AR29	I/O	GV_{DD}	—
MEMC1_MDQ[32:63]/ MEMC2_MDQ[0:31]	AN8, AN7, AM8, AM6, AP9, AN9, AT7, AP7, AU6, AP6, AR4, AR3, AT6, AT5, AR5, AT3, AP4, AM5, AP3, AN3, AN5, AL5, AN4, AM2, AL2, AH5, AK3, AJ2, AJ3, AH4, AK4, AH3	I/O	GV_{DD}	—
MEMC1_MECC[0:4]/ MSRCID[0:4]	AP24, AN22, AM19, AN19, AM24	I/O	GV_{DD}	—
MEMC1_MECC[5]/ MDVAL	AM23	I/O	GV_{DD}	—
MEMC1_MECC[6:7]	AM22, AN18	I/O	GV_{DD}	—
MEMC1_MDM[0:3]	AL36, AN34, AP33, AN28	O	GV_{DD}	—
MEMC1_MDM[4:7]/ MEMC2_MDM[0:3]	AT9, AU4, AM3, AJ6	O	GV_{DD}	—
MEMC1_MDM[8]	AP27	O	GV_{DD}	—
MEMC1_MDQS[0:3]	AK35, AP35, AN31, AM26	I/O	GV_{DD}	—
MEMC1_MDQS[4:7]/ MEMC2_MDQS[0:3]	AT8, AU3, AL4, AJ5	I/O	GV_{DD}	—
MEMC1_MDQS[8]	AP26	I/O	GV_{DD}	—
MEMC1_MBA[0:1]	AU29, AU30	O	GV_{DD}	—
MEMC1_MBA[2]	AT30	O	GV_{DD}	—
MEMC1_MA[0:14]	AU21, AP22, AP21, AT21, AU25, AU26, AT23, AR26, AU24, AR23, AR28, AU23, AR22, AU20, AR18	O	GV_{DD}	—
MEMC1_MODT[0:1]	AG33, AJ36	O	GV_{DD}	6
MEMC1_MODT[2:3]/ MEMC2_MODT[0:1]	AT1, AK2	O	GV_{DD}	6
$\overline{\text{MEMC1\_MWE}}$	AT26	O	GV_{DD}	—
$\overline{\text{MEMC1\_MRAS}}$	AT29	O	GV_{DD}	—
$\overline{\text{MEMC1\_MCAS}}$	AT24	O	GV_{DD}	—
$\overline{\text{MEMC1\_MCS}}$[0:1]	AU27, AT27	O	GV_{DD}	—
$\overline{\text{MEMC1\_MCS}}$[2:3]/ $\overline{\text{MEMC2\_MCS}}$[0:1]	AU8, AU7	O	GV_{DD}	—

Table 67. MPC8358E TBGA Pinout Listing (continued)

Signal	Package Pin Number	Pin Type	Power Supply	Notes
CE_PB[0:27]	AE2, AE1, AD5, AD3, AD2, AC6, AC5, AC4, AC2, AC1, AB5, AB4, AB3, AB1, AA6, AA4, AA2, Y6, Y4, Y3, Y2, Y1, W6, W5, W2, V5, V3, V2	I/O	OV_{DD}	—
CE_PC[0:1]	V1, U6	I/O	OV_{DD}	
CE_PC[2:3]	C16, A15	I/O	$LV_{DD}1$	—
CE_PC[4:6]	U4, U3, T6	I/O	OV_{DD}	—
CE_PC[7]	C19	I/O	$LV_{DD}2$	—
CE_PC[8:9]	A4, C5	I/O	$LV_{DD}0$	—
CE_PC[10:30]	T5, T4, T2, T1, R5, R3, R1, C11, D12, F13, B10, C10, E12, A9, B8, D10, A14, E15, B14, D15, AH2	I/O	OV_{DD}	—
CE_PD[0:27]	E11, D9, C8, F11, A7, E9, C7, A6, F10, B6, D7, E8, B5, A5, C2, E4, F5, B1, D2, G5, D1, E2, H6, F3, E1, F2, G3, H4	I/O	OV_{DD}	—
CE_PE[0:31]	K3, J2, F1, G2, J5, H3, G1, H2, K6, J3, K5, K4, L6, P6, P4, P3, P1, N4, N5, N2, N1, M2, M3, M5, M6, L1, L2, L4, E14, C13, C14, B13	I/O	OV_{DD}	—
CE_PF[0:3]	F14, D13, A12, A11	I/O	OV_{DD}	—
Clocks				
PCI_CLK_OUT[0]/CE_PF[26]	B22	I/O	$LV_{DD}2$	—
PCI_CLK_OUT[1:2]/CE_PF[27:28]	D22, A23	I/O	OV_{DD}	—
CLKIN	E37	I	OV_{DD}	—
PCI_CLOCK/PCI_SYNC_IN	M36	I	OV_{DD}	—
PCI_SYNC_OUT/CE_PF[29]	D37	I/O	OV_{DD}	3
JTAG				
TCK	K33	I	OV_{DD}	—
TDI	K34	I	OV_{DD}	4
TDO	H37	O	OV_{DD}	3
TMS	J36	I	OV_{DD}	4
$\overline{\text{TRST}}$	L32	I	OV_{DD}	4
Test				
TEST	L35	I	OV_{DD}	7
$\overline{\text{TEST\_SEL}}$	AU34	I	GV_{DD}	10
PMC				
$\overline{\text{QUIESCE}}$	B36	O	OV_{DD}	—
System Control				

Table 67. MPC8358E TBGA Pinout Listing (continued)

Signal	Package Pin Number	Pin Type	Power Supply	Notes
$LV_{DD}1$	C17, D16	Power for UCC2 Ethernet interface option 1 (2.5 V, 3.3 V)	$LV_{DD}1$	9
$LV_{DD}2$	B18, E21	Power for UCC2 Ethernet interface option 2 (2.5 V, 3.3 V)	$LV_{DD}2$	9
V_{DD}	C36, D29, D35, E16, F9, F12, F15, F17, F18, F20, F21, F23, F25, F26, F29, F31, F32, F33, G6, J6, K32, M32, N6, P33, R6, R32, U32, V6, Y5, Y32, AB6, AB33, AD6, AF32, AK6, AL6, AM7, AM9, AM10, AM11, AM12, AM13, AM14, AM15, AM18, AM21, AM25, AM28, AM32, AN15, AN21, AN26, AU9, AU17	Power for core (1.2 V)	V_{DD}	—
OV_{DD}	A10, B9, B15, B32, C1, C12, C22, C29, D24, E3, E10, E27, G4, H35, J1, J35, K2, M4, N3, N34, R2, R37, T36, U2, U33, V4, V34, W3, Y35, Y37, AA1, AA36, AB2, AB34	PCI, 10/100 Ethernet, and other standard (3.3 V)	OV_{DD}	—
MVREF1	AN20	I	DDR reference voltage	—
MVREF2	AU32	I	DDR reference voltage	—
SPARE1	B11	I/O	OV_{DD}	8
SPARE3	AH32	—	GV_{DD}	8
SPARE4	AU18	—	GV_{DD}	7
SPARE5	AP1	—	GV_{DD}	8

This figure shows the internal distribution of clocks within the MPC8358E.

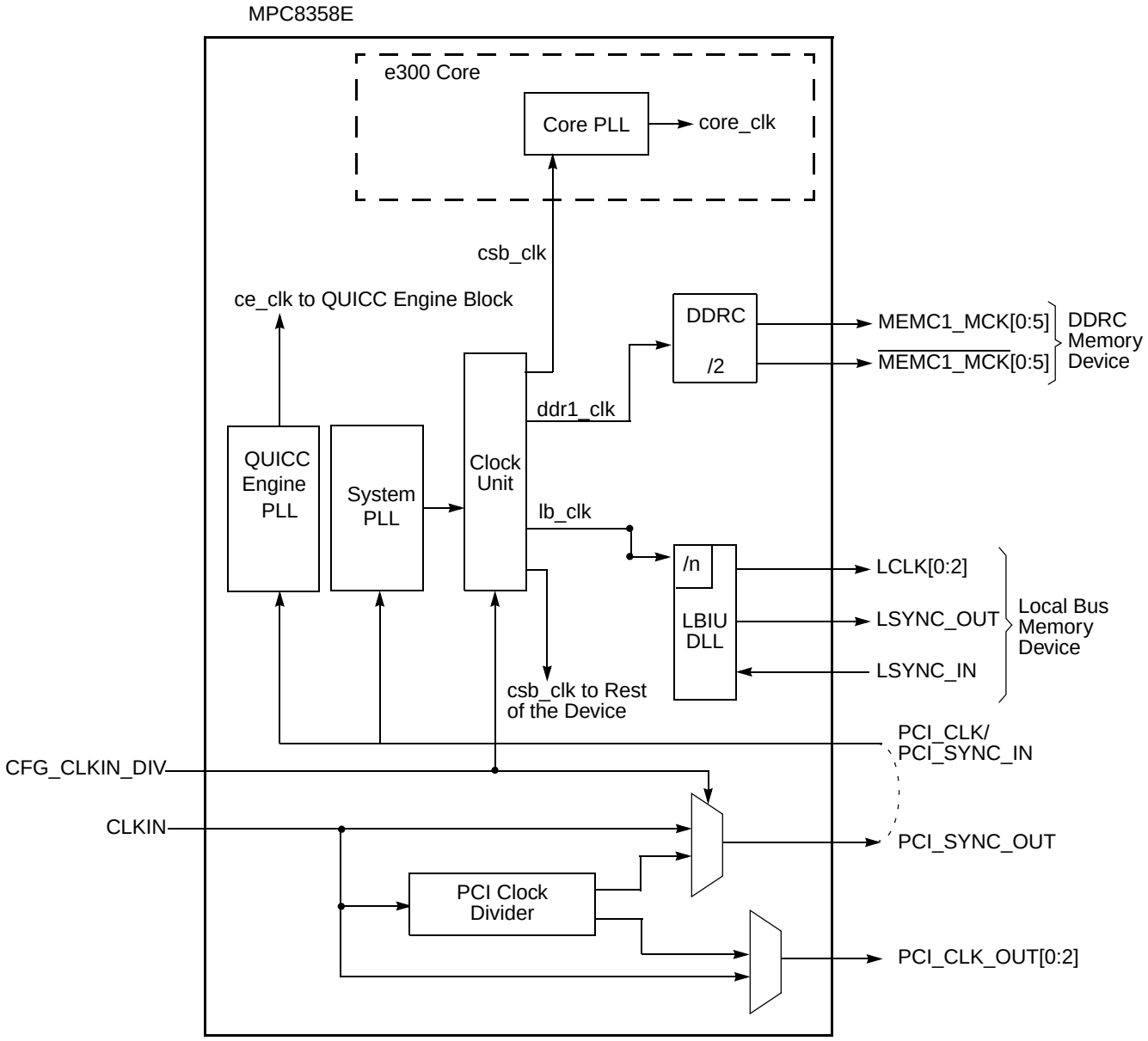


Figure 55. MPC8358E Clock Subsystem

The primary clock source for the device can be one of two inputs, CLKIN or PCI_CLK, depending on whether the device is configured in PCI host or PCI agent mode. Note that in PCI host mode, the primary clock input also depends on whether PCI clock outputs are selected with RCWH[PCICKDRV]. When the device is configured as a PCI host device (RCWH[PCIHOST] = 1) and PCI clock output is selected (RCWH[PCICKDRV] = 1), CLKIN is its primary input clock. CLKIN feeds the PCI clock divider (÷2) and the multiplexors for PCI_SYNC_OUT and PCI_CLK_OUT. The CFG_CLKIN_DIV configuration input selects whether CLKIN or CLKIN/2 is driven out on the PCI_SYNC_OUT signal. The OCCR[PCIOEN*n*] parameters enable the PCI_CLK_OUT*n*, respectively.

PCI_SYNC_OUT is connected externally to PCI_SYNC_IN to allow the internal clock subsystem to synchronize to the system PCI clocks. PCI_SYNC_OUT must be connected properly to PCI_SYNC_IN, with equal delay to all PCI agent devices in the system, to allow the device to function. When the device is configured as a PCI agent device, PCI_CLK is the primary input

The QUICC Engine block VCO frequency is derived from the following equations:

ce_clk = (primary clock input × CEPMF) ÷ (1 + CEPDF)

QE VCO Frequency = *ce_clk* × VCO divider × (1 + CEPDF)

21.4 Suggested PLL Configurations

To simplify the PLL configurations, the device might be separated into two clock domains. The first domain contains the CSB PLL and the core PLL. The core PLL is connected serially to the CSB PLL, and has the csb_clk as its input clock. The second clock domain has the QUICC Engine block PLL. The clock domains are independent, and each of their PLLs are configured separately. Both of the domains has one common input clock. This table shows suggested PLL configurations for 33 and 66 MHz input clocks and illustrates each of the clock domains separately. Any combination of clock domains setting with same input clock are valid. Refer to Section 21, "Clocking," for the appropriate operating frequencies for your device.

Table 76. Suggested PLL Configurations

Conf No.[1]	SPMF	CORE PLL	CEPMF	CEPDF	Input Clock Freq (MHz)	CSB Freq (MHz)	Core Freq (MHz)	QUICC Engine Freq (MHz)	400 (MHz)	533 (MHz)	667 (MHz)
33 MHz CLKIN/PCI_SYNC_IN Options											
s1	**0100**	**0000100**	æ	æ	33	133	266	—	∞	∞	∞
s2	**0100**	**0000101**	æ	æ	33	133	333	—	∞	∞	∞
s3	**0101**	**0000100**	æ	æ	33	166	333	—	∞	∞	∞
s4	**0101**	**0000101**	æ	æ	33	166	416	—	—	∞	∞
s5	**0110**	**0000100**	æ	æ	33	200	400	—	∞	∞	∞
s6	**0110**	**0000110**	æ	æ	33	200	600	—	—	—	∞
s7	**0111**	**0000011**	æ	æ	33	233	350	—	∞	∞	∞
s8	**0111**	**0000100**	æ	æ	33	233	466	—	—	∞	∞
s9	**0111**	**0000101**	æ	æ	33	233	583	—	—	—	∞
s10	**1000**	**0000011**	æ	æ	33	266	400	—	∞	∞	∞
s11	**1000**	**0000100**	æ	æ	33	266	533	—	—	∞	∞
s12	**1000**	**0000101**	æ	æ	33	266	667	—	—	—	∞
s13	**1001**	**0000010**	æ	æ	33	300	300	—	∞	∞	∞
s14	**1001**	**0000011**	æ	æ	33	300	450	—	—	∞	∞
s15	**1001**	**0000100**	æ	æ	33	300	600	—	—	—	∞
s16	**1010**	**0000010**	æ	æ	33	333	333	—	∞	∞	∞
s17	**1010**	**0000011**	æ	æ	33	333	500	—	—	∞	∞
s18	**1010**	**0000100**	æ	æ	33	333	667	—	—	—	∞
c1	æ	æ	**01001**	**0**	33	—	—	300	∞	∞	∞
c2	æ	æ	**01100**	**0**	33	—	—	400	∞	∞	∞
c3	æ	æ	**01110**	**0**	33	—	—	466	—	∞	∞
c4	æ	æ	**01111**	**0**	33	—	—	500	—	∞	∞

Table 76. Suggested PLL Configurations (continued)

Conf No.[1]	SPMF	CORE PLL	CEPMF	CEPDF	Input Clock Freq (MHz)	CSB Freq (MHz)	Core Freq (MHz)	QUICC Engine Freq (MHz)	400 (MHz)	533 (MHz)	667 (MHz)
c5	æ	æ	**10000**	**0**	33	—	—	533	—	∞	∞
c6	æ	æ	**10001**	**0**	33	—	—	566	—	—	∞
66 MHz CLKIN/PCI_SYNC_IN Options											
s1h	**0011**	**0000110**	æ	æ	66	200	400	—	∞	∞	∞
s2h	**0011**	**0000101**	æ	æ	66	200	500	—	—	∞	∞
s3h	**0011**	**0000110**	æ	æ	66	200	600	—	—	—	∞
s4h	**0100**	**0000011**	æ	æ	66	266	400	—	∞	∞	∞
s5h	**0100**	**0000100**	æ	æ	66	266	533	—	—	∞	∞
s6h	**0100**	**0000101**	æ	æ	66	266	667	—	—	—	∞
s7h	**0101**	**0000010**	æ	æ	66	333	333	—	∞	∞	∞
s8h	**0101**	**0000011**	æ	æ	66	333	500	—	—	∞	∞
s9h	**0101**	**0000100**	æ	æ	66	333	667	—	—	—	∞
c1h	æ	æ	**00101**	**0**	66	—	—	333	∞	∞	∞
c2h	æ	æ	**00110**	**0**	66	—	—	400	∞	∞	∞
c3h	æ	æ	**00111**	**0**	66	—	—	466	—	∞	∞
c4h	æ	æ	**01000**	**0**	66	—	—	533	—	∞	∞
c5h	æ	æ	**01001**	**0**	66	—	—	600	—	—	∞

Note:

1. The Conf No. consist of prefix, an index and a postfix. The prefix “s” and “c” stands for “syset” and “ce” respectively. The postfix “h” stands for “high input clock.”"The index is a serial number.

The following steps describe how to use above table. See Example 1.

2. Choose the up or down sections in the table according to input clock rate 33 MHz or 66 MHz.
3. Select a suitable CSB and core clock rates from Table 76. Copy the SPMF and CORE PLL configuration bits.
4. Select a suitable QUICC Engine block clock rate from Table 76. Copy the CEPMF and CEPDF configuration bits.
5. Insert the chosen SPMF, COREPLL, CEPMF and CEPDF to the RCWL fields, respectively.

22.3.1 Experimental Determination of the Junction Temperature with a Heat Sink

When heat sink is used, the junction temperature is determined from a thermocouple inserted at the interface between the case of the package and the interface material. A clearance slot or hole is normally required in the heat sink. Minimizing the size of the clearance is important to minimize the change in thermal performance caused by removing part of the thermal interface to the heat sink. Because of the experimental difficulties with this technique, many engineers measure the heat sink temperature and then back calculate the case temperature using a separate measurement of the thermal resistance of the interface. From this case temperature, the junction temperature is determined from the junction-to-case thermal resistance.

$$T_J = T_C + (R_{\theta JC} \times P_D)$$

where:

T_J = junction temperature (° C)
T_C = case temperature of the package (° C)
$R_{\theta JC}$ = junction to case thermal resistance (° C/W)
P_D = power dissipation (W)

23 System Design Information

This section provides electrical and thermal design recommendations for successful application of the MPC8360E/58E. Additional information can be found in *MPC8360E/MPC8358E PowerQUICC Design Checklist* (AN3097).

23.1 System Clocking

The device includes two PLLs, as follows.

- The platform PLL ($AV_{DD}1$) generates the platform clock from the externally supplied CLKIN input. The frequency ratio between the platform and CLKIN is selected using the platform PLL ratio configuration bits as described in Section 21.1, “System PLL Configuration.”
- The e300 core PLL ($AV_{DD}2$) generates the core clock as a slave to the platform clock. The frequency ratio between the e300 core clock and the platform clock is selected using the e300 PLL ratio configuration bits as described in Section 21.2, “Core PLL Configuration.”

23.2 PLL Power Supply Filtering

Each of the PLLs listed above is provided with power through independent power supply pins ($AV_{DD}1$, $AV_{DD}2$, respectively). The AV_{DD} level should always be equivalent to V_{DD}, and preferably these voltages are derived directly from V_{DD} through a low frequency filter scheme such as the following.

There are a number of ways to reliably provide power to the PLLs, but the recommended solution is to provide five independent filter circuits as illustrated in Figure 56, one to each of the five AV_{DD} pins. By providing independent filters to each PLL, the opportunity to cause noise injection from one PLL to the other is reduced.

This circuit is intended to filter noise in the PLLs resonant frequency range from a 500 kHz to 10 MHz range. It should be built with surface mount capacitors with minimum Effective Series Inductance (ESL). Consistent with the recommendations of Dr. Howard Johnson in *High Speed Digital Design: A Handbook of Black Magic* (Prentice Hall, 1993), multiple small capacitors of equal value are recommended over a single large value capacitor.

Each circuit should be placed as close as possible to the specific AV_{DD} pin being supplied to minimize noise coupled from nearby circuits. It should be possible to route directly from the capacitors to the AV_{DD} pin, which is on the periphery of package, without the inductance of vias.