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Understanding Embedded - Microprocessors

Embedded microprocessors are specialized computing chips designed to perform specific tasks within an embedded system. Unlike general-purpose microprocessors found in personal computers, embedded microprocessors are tailored for dedicated functions within larger systems, offering optimized performance, efficiency, and reliability. These microprocessors are integral to the operation of countless electronic devices, providing the computational power necessary for controlling processes, handling data, and managing communications.

Applications of Embedded - Microprocessors

Embedded microprocessors are utilized across a broad spectrum of applications, making them indispensable in

Details	
Product Status	Obsolete
Core Processor	PowerPC e300
Number of Cores/Bus Width	1 Core, 32-Bit
Speed	500MHz
Co-Processors/DSP	Communications; QUICC Engine, Security; SEC
RAM Controllers	DDR, DDR2
Graphics Acceleration	No
Display & Interface Controllers	-
Ethernet	10/100/1000Mbps (1)
SATA	-
USB	USB 1.x (1)
Voltage - I/O	1.8V, 2.5V, 3.3V
Operating Temperature	0°C ~ 105°C (TA)
Security Features	Cryptography, Random Number Generator
Package / Case	740-LBGA
Supplier Device Package	740-TBGA (37.5x37.5)
Purchase URL	https://www.e-xfl.com/product-detail/nxp-semiconductors/kmpc8360ezuahfh

- Eight TDM interfaces on the MPC8360E and four TDM interfaces on the MPC8358E with 1-bit mode for E3/T3 rates in clear channel
  - Sixteen independent baud rate generators and 30 input clock pins for supplying clocks to UCC and MCC serial channels (MCC is only available on the MPC8360E)
  - Four independent 16-bit timers that can be interconnected as four 32-bit timers
  - Interworking functionality:
    - Layer 2 10/100-Base T Ethernet switch
    - ATM-to-ATM switching (AAL0, 2, 5)
    - Ethernet-to-ATM switching with L3/L4 support
    - PPP interworking
- Security engine is optimized to handle all the algorithms associated with IPSec, SSL/TLS, SRTP, 802.11i®, iSCSI, and IKE processing. The security engine contains four crypto-channels, a controller, and a set of crypto execution units (EUs).
  - Public key execution unit (PKEU) supporting the following:
    - RSA and Diffie-Hellman
    - Programmable field size up to 2048 bits
    - Elliptic curve cryptography
    - F2m and F(p) modes
    - Programmable field size up to 511 bits
  - Data encryption standard execution unit (DEU)
    - DES, 3DES
    - Two key (K1, K2) or three key (K1, K2, K3)
    - ECB and CBC modes for both DES and 3DES
  - Advanced encryption standard unit (AESU)
  - Implements the Rinjdael symmetric key cipher
  - Key lengths of 128, 192, and 256 bits, two key
    - ECB, CBC, CCM, and counter modes
  - ARC four execution unit (AFEU)
    - Implements a stream cipher compatible with the RC4 algorithm
    - 40- to 128-bit programmable key
  - Message digest execution unit (MDEU)
    - SHA with 160-, 224-, or 256-bit message digest
    - MD5 with 128-bit message digest
    - HMAC with either SHA or MD5 algorithm
  - Random number generator (RNG)
  - Four crypto-channels, each supporting multi-command descriptor chains
    - Static and/or dynamic assignment of crypto-execution units via an integrated controller
    - Buffer size of 256 bytes for each execution unit, with flow control for large data sizes
  - Storage/NAS XOR parity generation accelerator for RAID applications
- Dual DDR SDRAM memory controllers on the MPC8360E and a single DDR SDRAM memory controller on the MPC8358E
  - Programmable timing supporting both DDR1 and DDR2 SDRAM
  - On the MPC8360E, the DDR buses can be configured as two 32-bit buses or one 64-bit bus; on the MPC8358E, the DDR bus can be configured as a 32- or 64-bit bus
  - 32- or 64-bit data interface, up to 333 MHz (for the MPC8360E) and 266 MHz (for the MPC8358E) data rate
  - Four banks of memory, each up to 1 Gbyte

Table 1. Absolute Maximum Ratings[1] (continued)

Characteristic	Symbol	Max Value	Unit	Notes
Storage temperature range	T_{STG}	–55 to 150	°C	—

Notes:
1. Functional and tested operating conditions are given in Table 2. Absolute maximum ratings are stress ratings only, and functional operation at the maximums is not guaranteed. Stresses beyond those listed may affect device reliability or cause permanent damage to the device.
2. **Caution:** MV_{IN} must not exceed GV_{DD} by more than 0.3 V. This limit may be exceeded for a maximum of 100 ms during power-on reset and power-down sequences.
3. **Caution:** OV_{IN} must not exceed OV_{DD} by more than 0.3 V. This limit may be exceeded for a maximum of 100 ms during power-on reset and power-down sequences.
4. **Caution:** LV_{IN} must not exceed LV_{DD} by more than 0.3 V. This limit may be exceeded for a maximum of 100 ms during power-on reset and power-down sequences.
5. (M,L,O)V_{IN} and MV_{REF} may overshoot/undershoot to a voltage and for a maximum duration as shown in Figure 3.
6. OV_{IN} on the PCI interface may overshoot/undershoot according to the PCI Electrical Specification for 3.3-V operation, as shown in Figure 4.

2.1.2 Power Supply Voltage Specification

This table provides the recommended operating conditions for the device. Note that the values in this table are the recommended and tested operating conditions. Proper device operation outside of these conditions is not guaranteed.

Table 2. Recommended Operating Conditions

Characteristic	Symbol	Recommended Value	Unit	Notes
Core and PLL supply voltage for MPC8358 Device Part Number with Processor Frequency label of AD=266MHz and AG=400MHz & QUICC Engine Frequency label of E=300MHz & G=400MHz MPC8360 Device Part Number with Processor Frequency label of AG=400MHz and AJ=533MHz & QUICC Engine Frequency label of G=400MHz	V_{DD} & AV_{DD}	1.2 V ± 60 mV	V	1, 3
Core and PLL supply voltage for MPC8360 Device Part Number with Processor Frequency label of AL=667MHz and QUICC Engine Frequency label of H=500MHz	V_{DD} & AV_{DD}	1.3 V ± 50 mV	V	1, 3
DDR and DDR2 DRAM I/O supply voltage DDR DDR2	GV_{DD}	 2.5 V ± 125 mV 1.8 V ± 90 mV	V	—
Three-speed Ethernet I/O supply voltage	$LV_{DD}0$	3.3 V ± 330 mV 2.5 V ± 125 mV	V	—
Three-speed Ethernet I/O supply voltage	$LV_{DD}1$	3.3 V ± 330 mV 2.5 V ± 125 mV	V	—
Three-speed Ethernet I/O supply voltage	$LV_{DD}2$	3.3 V ± 330 mV 2.5 V ± 125 mV	V	—

This table shows the estimated typical I/O power dissipation for the device.

Table 6. Estimated Typical I/O Power Dissipation

Interface	Parameter	GV_{DD} (1.8 V)	GV_{DD} (2.5 V)	OV_{DD} (3.3 V)	LV_{DD} (3.3 V)	LV_{DD} (2.5 V)	Unit	Comments
DDR I/O 65% utilization R_s = 20 Ω R_t = 50 Ω 2 pairs of clocks	200 MHz, 1 × 32 bits	0.3	0.46	—	—	—	W	—
	200 MHz, 1 × 64 bits	0.4	0.58	—	—	—	W	—
	200 MHz, 2 × 32 bits	0.6	0.92	—	—	—	W	—
	266 MHz, 1 × 32 bits	0.35	0.56	—	—	—	W	—
	266 MHz, 1 × 64 bits	0.46	0.7	—	—	—	W	—
	266 MHz, 2 × 32 bits	0.7	1.11	—	—	—	W	—
	333 MHz, 1 × 32 bits	0.4	0.65	—	—	—	W	—
	333 MHz, 1 × 64 bits	0.53	0.82	—	—	—	W	—
	333 MHz, 2 × 32 bits	0.81	1.3	—	—	—	W	—
Local Bus I/O Load = 25 pf 3 pairs of clocks	133 MHz, 32 bits	—	—	0.22	—	—	W	—
	83 MHz, 32 bits	—	—	0.14	—	—	W	—
	66 MHz, 32 bits	—	—	0.12	—	—	W	—
	50 MHz, 32 bits	—	—	0.09	—	—	W	—
PCI I/O Load = 30 pF	33 MHz, 32 bits	—	—	0.05	—	—	W	—
	66 MHz, 32 bits	—	—	0.07	—	—	W	—
10/100/1000 Ethernet I/O Load = 20 pF	MII or RMII	—	—	—	0.01	—	W	Multiply by number of interfaces used.
	GMII or TBI	—	—	—	0.04	—	W	
	RGMII or RTBI	—	—	—	—	0.04	W	
Other I/O	—	—	—	0.1	—	—	W	—

4 Clock Input Timing

This section provides the clock input DC and AC electrical characteristics for the MPC8360E/58E.

NOTE

The rise/fall time on QUICC Engine block input pins should not exceed 5 ns. This should be enforced especially on clock signals. Rise time refers to signal transitions from 10% to 90% of V_{DD}; fall time refers to transitions from 90% to 10% of V_{DD}.

8.2.5 RGMII and RTBI AC Timing Specifications

This table presents the RGMII and RTBI AC timing specifications.

Table 35. RGMII and RTBI AC Timing Specifications

At recommended operating conditions with LV_{DD} of 2.5 V ± 5%.

Parameter/Condition	Symbol[1]	Min	Typ	Max	Unit	Notes
Data to clock output skew (at transmitter)	$t_{SKRGTKHDX}$ $t_{SKRGTKHDV}$	–0.5 —	—	— 0.5	ns	7
Data to clock input skew (at receiver)	$t_{SKRGDXKH}$ $t_{SKRGDVKH}$	1.0 —	—	— 2.6	ns	2
Clock cycle duration	t_{RGT}	7.2	8.0	8.8	ns	3
Duty cycle for 1000Base-T	t_{RGTH}/t_{RGT}	45	50	55	%	4, 5
Duty cycle for 10BASE-T and 100BASE-TX	t_{RGTH}/t_{RGT}	40	50	60	%	3, 5
Rise time (20–80%)	t_{RGTR}	—	—	0.75	ns	—
Fall time (20–80%)	t_{RGTF}	—	—	0.75	ns	—
GTX_CLK125 reference clock period	t_{G125}	—	8.0	—	ns	6
GTX_CLK125 reference clock duty cycle	t_{G125H}/t_{G125}	47	—	53	%	—

Notes:

1. Note that, in general, the clock reference symbol representation for this section is based on the symbols RGT to represent RGMII and RTBI timing. For example, the subscript of t_{RGT} represents the TBI (T) receive (Rx) clock. Note also that the notation for rise (R) and fall (F) times follows the clock symbol that is being represented. For symbols representing skews, the subscript is skew (SK) followed by the clock that is being skewed (RGT).
2. This implies that PC board design requires clocks to be routed such that an additional trace delay of greater than 1.5 ns can be added to the associated clock signal.
3. For 10 and 100 Mbps, t_{RGT} scales to 400 ns ± 40 ns and 40 ns ± 4 ns, respectively.
4. Duty cycle may be stretched/shrunk during speed changes or while transitioning to a received packet's clock domains as long as the minimum duty cycle is not violated and stretching occurs for no more than three t_{RGT} of the lowest speed transitioned between.
5. Duty cycle reference is $LV_{DD}/2$.
6. This symbol is used to represent the external GTX_CLK125 and does not follow the original symbol naming convention.
7. In rev. 2.0 silicon, due to errata, $t_{SKRGTKHDX}$ minimum is –2.3 ns and $t_{SKRGTKHDV}$ maximum is 1 ns for UCC1, 1.2 ns for UCC2 option 1, and 1.8 ns for UCC2 option 2. In rev. 2.1 silicon, due to errata, $t_{SKRGTKHDX}$ minimum is –0.65 ns for UCC2 option 1 and –0.9 for UCC2 option 2, and $t_{SKRGTKHDV}$ maximum is 0.75 ns for UCC1 and UCC2 option 1 and 0.85 for UCC2 option 2. Refer to Errata QE_ENET10 in *Chip Errata for the MPC8360E, Rev. 1*. UCC1 does meet $t_{SKRGTKHDX}$ minimum for rev. 2.1 silicon.

Table 41. Local Bus General Timing Parameters—DLL Bypass Mode[9] (continued)

Parameter	Symbol[1]	Min	Max	Unit	Notes
Local bus clock to output valid	t_{LBKHOV}	—	3	ns	3
Local bus clock to output high impedance for LAD/LDP	t_{LBKHOZ}	—	4	ns	8

Notes:
1. The symbols used for timing specifications follow the pattern of $t_{\text{(first two letters of functional block)(signal)(state)(reference)(state)}}$ for inputs and $t_{\text{(first two letters of functional block)(reference)(state)(signal)(state)}}$ for outputs. For example, $t_{LBIXKH1}$ symbolizes local bus timing (LB) for the input (I) to go invalid (X) with respect to the time the t_{LBK} clock reference (K) goes high (H), in this case for clock one (1). Also, t_{LBKHOX} symbolizes local bus timing (LB) for the t_{LBK} clock reference (K) to go high (H), with respect to the output (O) going invalid (X) or output hold time.
2. All timings are in reference to falling edge of LCLK0 (for all outputs and for $\overline{\text{LGTA}}$ and LUPWAIT inputs) or rising edge of LCLK0 (for all other inputs).
3. All signals are measured from $OV_{DD}/2$ of the rising/falling edge of LCLK0 to $0.4 \times OV_{DD}$ of the signal in question for 3.3-V signaling levels.
4. Input timings are measured at the pin.
5. $t_{LBOTOT1}$ should be used when RCWH[LALE] is not set and when the load on LALE output pin is at least 10 pF less than the load on LAD output pins.
6. $t_{LBOTOT2}$ should be used when RCWH[LALE] is set and when the load on LALE output pin is at least 10 pF less than the load on LAD output pins.
7. $t_{LBOTOT3}$ should be used when RCWH[LALE] is set and when the load on LALE output pin equals to the load on LAD output pins.
8. For purposes of active/float timing measurements, the Hi-Z or off-state is defined to be when the total current delivered through the component pin is less than or equal to the leakage current specification.
9. DLL bypass mode is not recommended for use at frequencies above 66 MHz.

This figure provides the AC test load for the local bus.

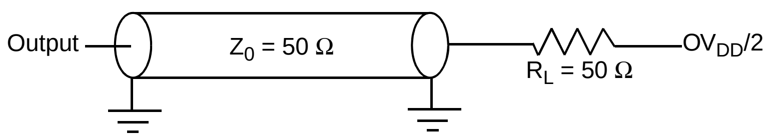

Figure 22. Local Bus C Test Load

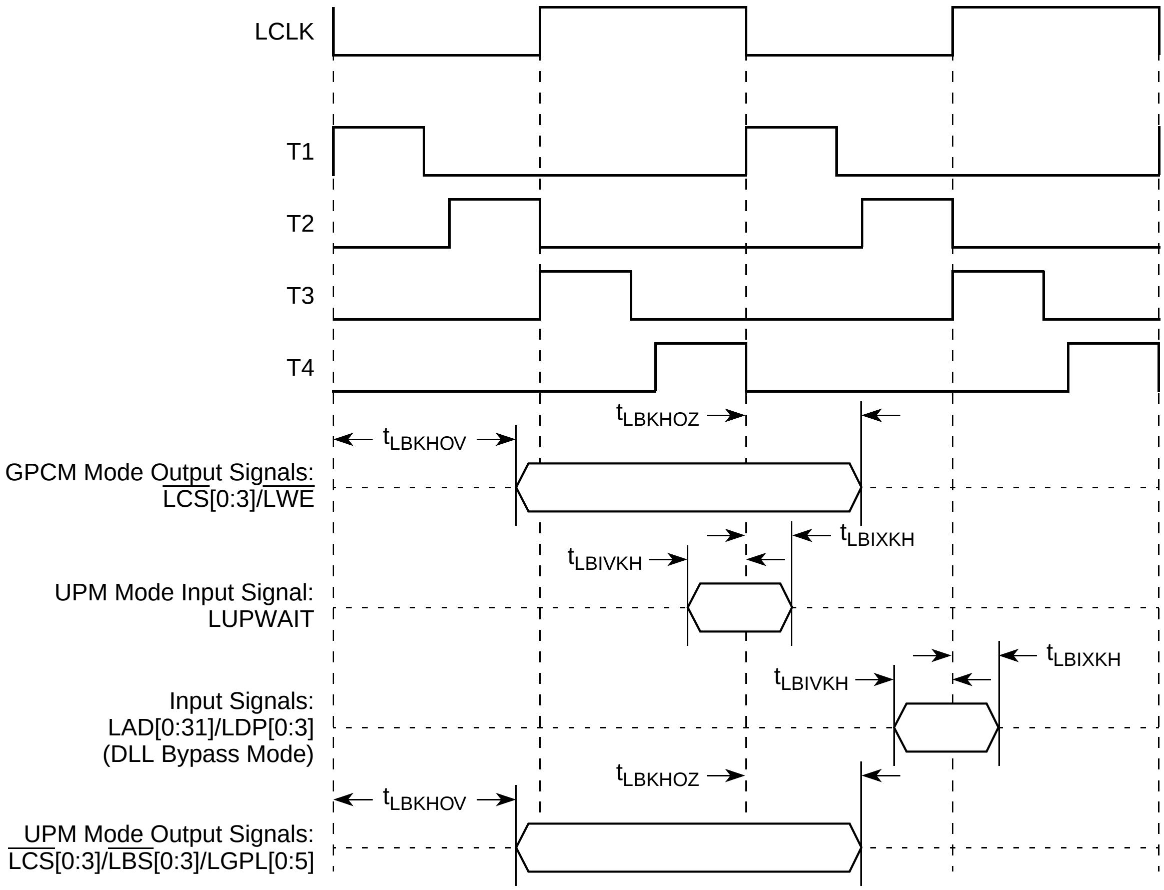

Figure 27. Local Bus Signals, GPCM/UPM Signals for LCRR[CLKDIV] = 4 (DLL Bypass Mode)

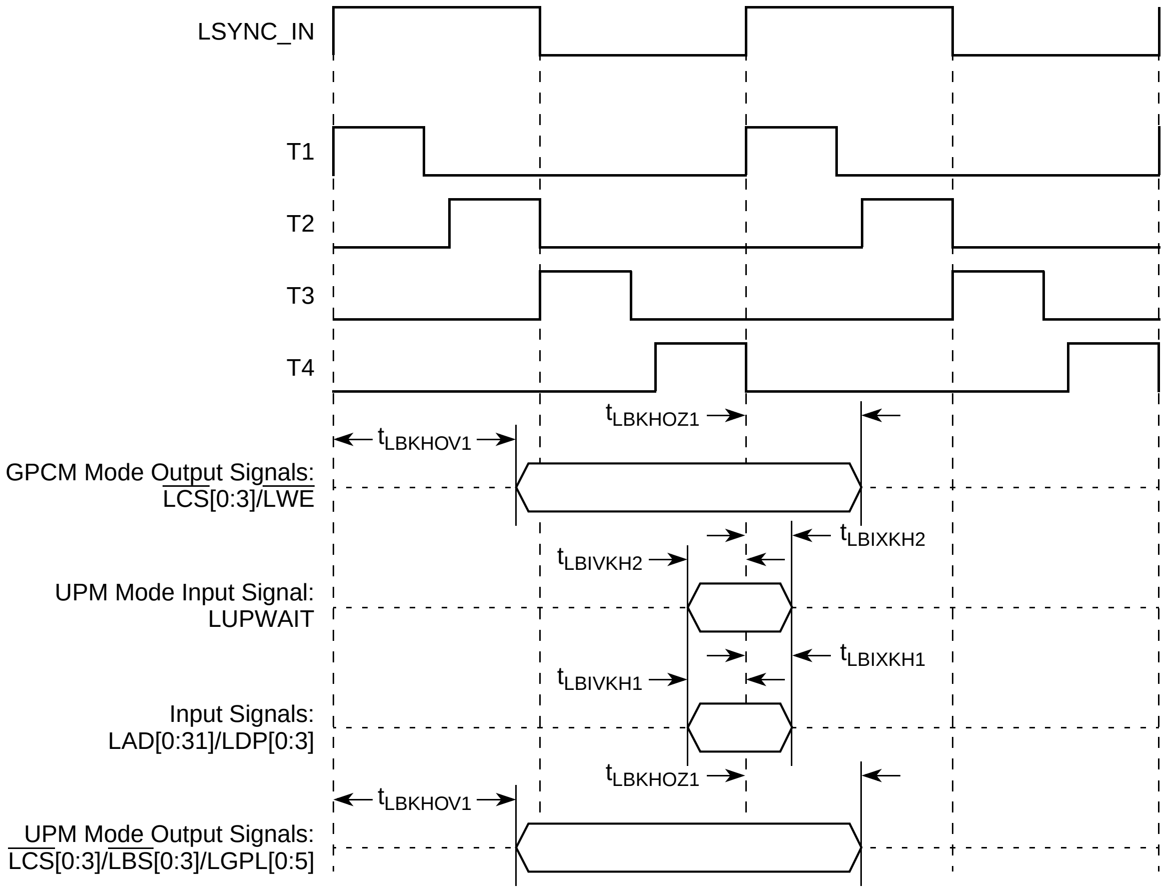

Figure 28. Local Bus Signals, GPCM/UPM Signals for LCRR[CLKDIV] = 4 (DLL Enabled)

10 JTAG

This section describes the DC and AC electrical specifications for the IEEE 1149.1 (JTAG) interface of the MPC8360E/58E.

10.1 JTAG DC Electrical Characteristics

This table provides the DC electrical characteristics for the IEEE 1149.1 (JTAG) interface of the device.

Table 42. JTAG interface DC Electrical Characteristics

Characteristic	Symbol	Condition	Min	Max	Unit
Output high voltage	V_{OH}	I_{OH} = –6.0 mA	2.4	—	V
Output low voltage	V_{OL}	I_{OL} = 6.0 mA	—	0.5	V
Output low voltage	V_{OL}	I_{OL} = 3.2 mA	—	0.4	V
Input high voltage	V_{IH}	—	2.5	OV_{DD} + 0.3	V
Input low voltage	V_{IL}	—	–0.3	0.8	V
Input current	I_{IN}	0 V ≤ V_{IN} ≤ OV_{DD}	—	±10	μA

This figure provides the AC test load for TDO and the boundary-scan outputs of the device.

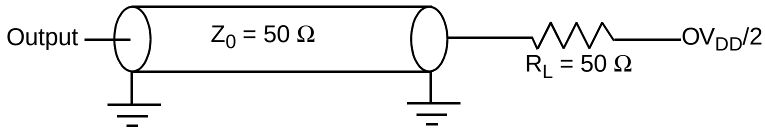

Figure 29. AC Test Load for the JTAG Interface

This figure provides the JTAG clock input timing diagram.

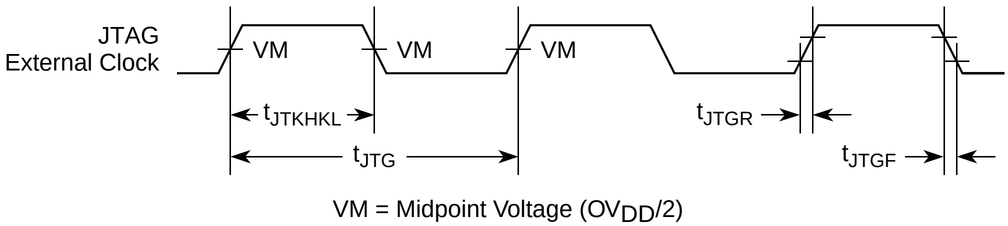

Figure 30. JTAG Clock Input Timing Diagram

This figure provides the $\overline{TRST}$ timing diagram.

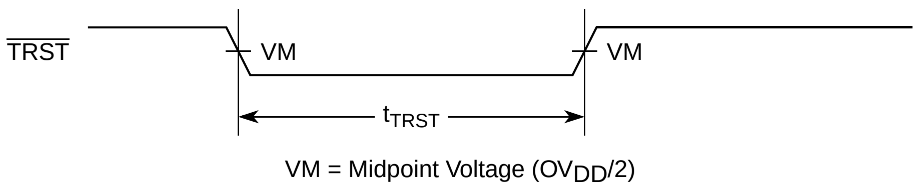

Figure 31. $\overline{TRST}$ Timing Diagram

This figure provides the boundary-scan timing diagram.

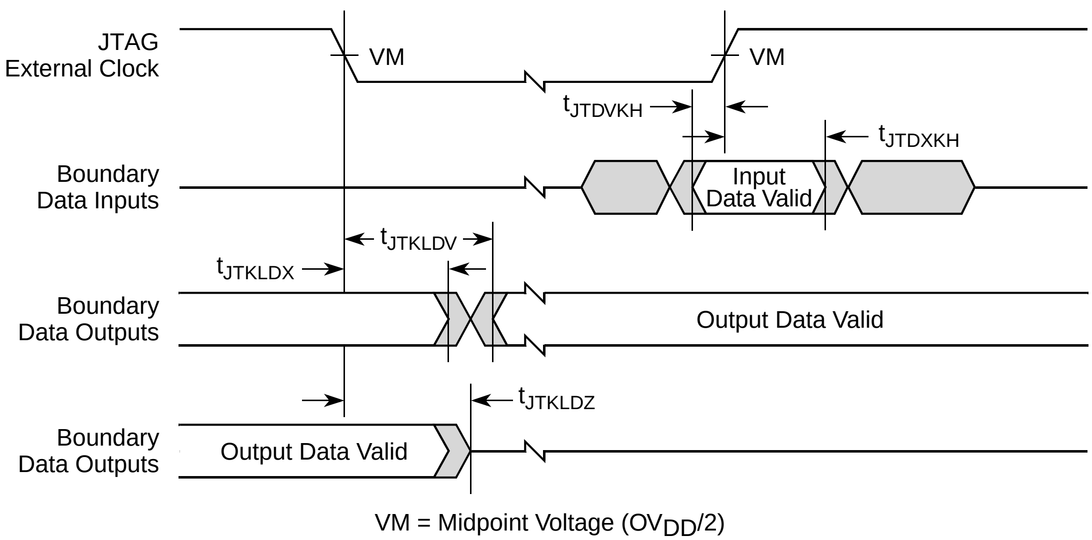

Figure 32. Boundary-Scan Timing Diagram

15.2 IPIC AC Timing Specifications

This table provides the IPIC input and output AC timing specifications.

Table 54. IPIC Input AC Timing Specifications[1]

Characteristic	Symbol[2]	Min	Unit
IPIC inputs—minimum pulse width	t_{PIWID}	20	ns

Notes:

1. Input specifications are measured from the 50% level of the signal to the 50% level of the rising edge of CLKIN. Timings are measured at the pin.
2. IPIC inputs and outputs are asynchronous to any visible clock. IPIC outputs should be synchronized before use by any external synchronous logic. IPIC inputs are required to be valid for at least t_{PIWID} ns to ensure proper operation when working in edge triggered mode.

16 SPI

This section describes the DC and AC electrical specifications for the SPI of the MPC8360E/58E.

16.1 SPI DC Electrical Characteristics

This table provides the DC electrical characteristics for the device SPI.

Table 55. SPI DC Electrical Characteristics

Characteristic	Symbol	Condition	Min	Max	Unit
Output high voltage	V_{OH}	I_{OH} = –6.0 mA	2.4	—	V
Output low voltage	V_{OL}	I_{OL} = 6.0 mA	—	0.5	V
Output low voltage	V_{OL}	I_{OL} = 3.2 mA	—	0.4	V
Input high voltage	V_{IH}	—	2.0	OV_{DD} + 0.3	V
Input low voltage	V_{IL}	—	–0.3	0.8	V
Input current	I_{IN}	0 V $\le V_{IN} \le OV_{DD}$	—	±10	µA

16.2 SPI AC Timing Specifications

This table and provide the SPI input and output AC timing specifications.

Table 56. SPI AC Timing Specifications[1]

Characteristic	Symbol[2]	Min	Max	Unit
SPI outputs—Master mode (internal clock) delay	t_{NIKHOX} t_{NIKHOV}	0.3 —	— 8	ns
SPI outputs—Slave mode (external clock) delay	t_{NEKHOX} t_{NEKHOV}	2 —	— 8	ns
SPI inputs—Master mode (internal clock) input setup time	t_{NIIVKH}	8	—	ns
SPI inputs—Master mode (internal clock) input hold time	t_{NIIXKH}	0	—	ns
SPI inputs—Slave mode (external clock) input setup time	t_{NEIVKH}	4	—	ns

Table 60. UTOPIA AC Timing Specifications[1] (continued)

Characteristic	Symbol[2]	Min	Max	Unit	Notes
UTOPIA inputs—Internal clock input hold time	t_{UIIXKH}	2.4	—	ns	—
UTOPIA inputs—External clock input hold time	t_{UEIXKH}	1	—	ns	3

Notes:
1. Output specifications are measured from the 50% level of the rising edge of CLKIN to the 50% level of the signal. Timings are measured at the pin.
2. The symbols used for timing specifications follow the pattern of $t_{\text{(first two letters of functional block)(signal)(state)(reference)(state)}}$ for inputs and $t_{\text{(first two letters of functional block)(reference)(state)(signal)(state)}}$ for outputs. For example, t_{UIKHOX} symbolizes the UTOPIA outputs internal timing (UI) for the time t_{UTOPIA} memory clock reference (K) goes from the high state (H) until outputs (O) are invalid (X).
3. In rev. 2.0 silicon, due to errata, t_{UEIVKH} minimum is 4.3 ns and t_{UEIXKH} minimum is 1.4 ns under specific conditions. Refer to Errata QE_UPC3 in *Chip Errata for the MPC8360E, Rev. 1*.

This figure provides the AC test load for the UTOPIA.

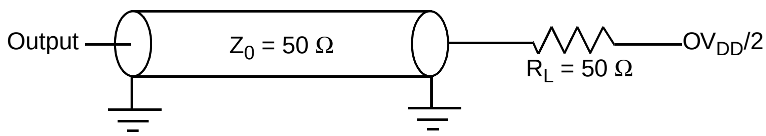

Figure 46. UTOPIA AC Test Load

These figures represent the AC timing from Table 56. Note that although the specifications generally reference the rising edge of the clock, these AC timing diagrams also apply when the falling edge is the active edge.

This figure shows the UTOPIA timing with external clock.

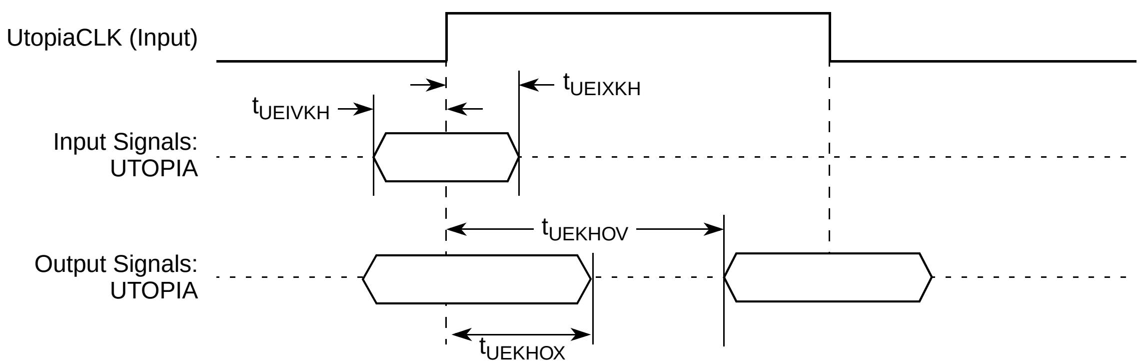

Figure 47. UTOPIA AC Timing (External Clock) Diagram

This figure shows the UTOPIA timing with internal clock.

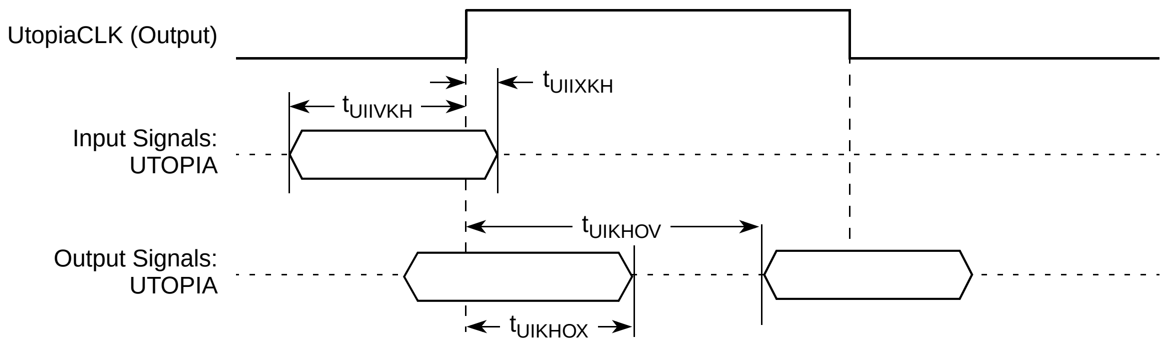

Figure 48. UTOPIA AC Timing (Internal Clock) Diagram

18 HDLC, BISYNC, Transparent, and Synchronous UART

This section describes the DC and AC electrical specifications for the high level data link control (HDLC), BISYNC, transparent, and synchronous UART protocols of the MPC8360E/58E.

18.1 HDLC, BISYNC, Transparent, and Synchronous UART DC Electrical Characteristics

This table provides the DC electrical characteristics for the device HDLC, BISYNC, transparent, and synchronous UART protocols.

Table 61. HDLC, BISYNC, Transparent, and Synchronous UART DC Electrical Characteristics

Characteristic	Symbol	Condition	Min	Max	Unit
Output high voltage	V_{OH}	I_{OH} = –2.0 mA	2.4	—	V
Output low voltage	V_{OL}	I_{OL} = 3.2 mA	—	0.5	V
Input high voltage	V_{IH}	—	2.0	OV_{DD} + 0.3	V
Input low voltage	V_{IL}	—	–0.3	0.8	V
Input current	I_{IN}	0 V $\le V_{IN} \le OV_{DD}$	—	±10	μA

18.2 HDLC, BISYNC, Transparent, and Synchronous UART AC Timing Specifications

These tables provide the input and output AC timing specifications for HDLC, BISYNC, transparent, and synchronous UART protocols.

Table 62. HDLC, BISYNC, and Transparent AC Timing Specifications[1]

Characteristic	Symbol[2]	Min	Max	Unit
Outputs—Internal clock delay	t_{HIKHOV}	0	11.2	ns
Outputs—External clock delay	t_{HEKHOV}	1	10.8	ns

Table 66. MPC8360E TBGA Pinout Listing (continued)

Signal	Package Pin Number	Pin Type	Power Supply	Notes
PCI_DEVSEL/CE_PF[16]	E26	I/O	OV_{DD}	5
PCI_IDSEL/CE_PF[17]	F22	I/O	OV_{DD}	—
PCI_SERR/CE_PF[18]	B29	I/O	OV_{DD}	5
PCI_PERR/CE_PF[19]	A29	I/O	OV_{DD}	5
PCI_REQ[0]/CE_PF[20]	F19	I/O	$LV_{DD}2$	—
PCI_REQ[1]/CPCI_HS_ES/ CE_PF[21]	A21	I/O	$LV_{DD}2$	—
PCI_REQ[2]/CE_PF[22]	C21	I/O	$LV_{DD}2$	—
PCI_GNT[0]/CE_PF[23]	E20	I/O	$LV_{DD}2$	—
PCI_GNT[1]/CPCI1_HS_LED/ CE_PF[24]	B20	I/O	$LV_{DD}2$	—
PCI_GNT[2]/CPCI1_HS_ENUM/ CE_PF[25]	C20	I/O	$LV_{DD}2$	—
PCI_MODE	D36	I	OV_{DD}	—
M66EN/CE_PF[4]	B37	I/O	OV_{DD}	—
Local Bus Controller Interface				
LAD[0:31]	N32, N33, N35, N36, P37, P32, P34, R36, R35, R34, R33, T37, T35, T34, T33, U37, T32, U36, U34, V36, V35, W37, W35, V33, V32, W34, Y36, W32, AA37, Y33, AA35, AA34	I/O	OV_{DD}	—
LDP[0]/CKSTOP_OUT	AB37	I/O	OV_{DD}	—
LDP[1]/CKSTOP_IN	AB36	I/O	OV_{DD}	—
LDP[2]/LCS[6]	AB35	I/O	OV_{DD}	—
LDP[3]/LCS[7]	AA33	I/O	OV_{DD}	—
LA[27:31]	AC37, AA32, AC36, AC34, AD36	O	OV_{DD}	—
LCS[0:5]	AD33, AG37, AF34, AE33, AD32, AH37	O	OV_{DD}	—
LWE[0:3]/LSDDQM[0:3]/LBS[0:3]	AG35, AG34, AH36, AE32	O	OV_{DD}	—
LBCTL	AD35	O	OV_{DD}	—
LALE	M37	O	OV_{DD}	—
LGPL0/LSDA10/cfg_reset_source0	AB32	I/O	OV_{DD}	—
LGPL1/LSDWE/cfg_reset_source1	AE37	I/O	OV_{DD}	—
LGPL2/LSDRAS/LOE	AC33	O	OV_{DD}	—
LGPL3/LSDCAS/cfg_reset_source2	AD34	I/O	OV_{DD}	—
LGPL4/LGTA/LUPWAIT/LPBSE	AE35	I/O	OV_{DD}	—
LGPL5/cfg_clkin_div	AF36	I/O	OV_{DD}	—
LCKE	G36	O	OV_{DD}	—
LCLK[0]	J33	O	OV_{DD}	—
LCLK[1]/LCS[6]	J34	O	OV_{DD}	—

Table 66. MPC8360E TBGA Pinout Listing (continued)

Signal	Package Pin Number	Pin Type	Power Supply	Notes
$LV_{DD}0$	D5, D6	Power for UCC1 Ethernet interface (2.5 V, 3.3 V)	$LV_{DD}0$	—
$LV_{DD}1$	C17, D16	Power for UCC2 Ethernet interface option 1 (2.5 V, 3.3 V)	$LV_{DD}1$	9
$LV_{DD}2$	B18, E21	Power for UCC2 Ethernet interface option 2 (2.5 V, 3.3 V)	$LV_{DD}2$	9
V_{DD}	C36, D29, D35, E16, F9, F12, F15, F17, F18, F20, F21, F23, F25, F26, F29, F31, F32, F33, G6, J6, K32, M32, N6, P33, R6, R32, U32, V6, Y5, Y32, AB6, AB33, AD6, AF32, AK6, AL6, AM7, AM9, AM10, AM11, AM12, AM13, AM14, AM15, AM18, AM21, AM25, AM28, AM32, AN15, AN21, AN26, AU9, AU17	Power for core (1.2 V)	V_{DD}	—
OV_{DD}	A10, B9, B15, B32, C1, C12, C22, C29, D24, E3, E10, E27, G4, H35, J1, J35, K2, M4, N3, N34, R2, R37, T36, U2, U33, V4, V34, W3, Y35, Y37, AA1, AA36, AB2, AB34	PCI, 10/100 Ethernet, and other standard (3.3 V)	OV_{DD}	—
MVREF1	AN20	I	DDR reference voltage	—
MVREF2	AU32	I	DDR reference voltage	—
SPARE1	B11	I/O	OV_{DD}	8
SPARE3	AH32	—	GV_{DD}	8
SPARE4	AU18	—	GV_{DD}	7
SPARE5	AP1	—	GV_{DD}	8

Table 67. MPC8358E TBGA Pinout Listing (continued)

Signal	Package Pin Number	Pin Type	Power Supply	Notes
$\overline{\text{IRQ}}$[4:5]	G33, G32	I/O	OV_{DD}	—
$\overline{\text{IRQ}}$[6]/$\overline{\text{LCS}}$[6]/$\overline{\text{CKSTOP\_OUT}}$	E35	I/O	OV_{DD}	—
$\overline{\text{IRQ}}$[7]/$\overline{\text{LCS}}$[7]/$\overline{\text{CKSTOP\_IN}}$	H36	I/O	OV_{DD}	—
DUART				
UART1_SOUT/M1SRCID[0]/ M2SRCID[0]/LSRCID[0]	E32	O	OV_{DD}	—
UART1_SIN/M1SRCID[1]/ M2SRCID[1]/LSRCID[1]	B34	I/O	OV_{DD}	—
$\overline{\text{UART1\_CTS}}$/M1SRCID[2]/ M2SRCID[2]/LSRCID[2]	C34	I/O	OV_{DD}	—
$\overline{\text{UART1\_RTS}}$/M1SRCID[3]/ M2SRCID[3]/LSRCID[3]	A35	O	OV_{DD}	—
I^2C Interface				
IIC1_SDA	D34	I/O	OV_{DD}	2
IIC1_SCL	B35	I/O	OV_{DD}	2
IIC2_SDA	E33	I/O	OV_{DD}	2
IIC2_SCL	C35	I/O	OV_{DD}	2
QUICC Engine				
CE_PA[0]	F8	I/O	LV_{DD0}	—
CE_PA[1:2]	AH1, AG5	I/O	OV_{DD}	—
CE_PA[3:7]	F6, D4, C3, E5, A3	I/O	$LV_{DD}0$	—
CE_PA[8]	AG3	I/O	OV_{DD}	—
CE_PA[9:12]	F7, B3, E6, B4	I/O	$LV_{DD}0$	—
CE_PA[13:14]	AG1, AF6	I/O	OV_{DD}	—
CE_PA[15]	B2	I/O	$LV_{DD}0$	—
CE_PA[16]	AF4	I/O	OV_{DD}	—
CE_PA[17:21]	B16, A16, E17, A17, B17	I/O	$LV_{DD}1$	—
CE_PA[22]	AF3	I/O	OV_{DD}	—
CE_PA[23:26]	C18, D18, E18, A18	I/O	$LV_{DD}1$	—
CE_PA[27:28]	AF2, AE6	I/O	OV_{DD}	—
CE_PA[29]	B19	I/O	$LV_{DD}1$	—
CE_PA[30]	AE5	I/O	OV_{DD}	—
CE_PA[31]	F16	I/O	$LV_{DD}1$	—

Table 67. MPC8358E TBGA Pinout Listing (continued)

Signal	Package Pin Number	Pin Type	Power Supply	Notes
$LV_{DD}1$	C17, D16	Power for UCC2 Ethernet interface option 1 (2.5 V, 3.3 V)	$LV_{DD}1$	9
$LV_{DD}2$	B18, E21	Power for UCC2 Ethernet interface option 2 (2.5 V, 3.3 V)	$LV_{DD}2$	9
V_{DD}	C36, D29, D35, E16, F9, F12, F15, F17, F18, F20, F21, F23, F25, F26, F29, F31, F32, F33, G6, J6, K32, M32, N6, P33, R6, R32, U32, V6, Y5, Y32, AB6, AB33, AD6, AF32, AK6, AL6, AM7, AM9, AM10, AM11, AM12, AM13, AM14, AM15, AM18, AM21, AM25, AM28, AM32, AN15, AN21, AN26, AU9, AU17	Power for core (1.2 V)	V_{DD}	—
OV_{DD}	A10, B9, B15, B32, C1, C12, C22, C29, D24, E3, E10, E27, G4, H35, J1, J35, K2, M4, N3, N34, R2, R37, T36, U2, U33, V4, V34, W3, Y35, Y37, AA1, AA36, AB2, AB34	PCI, 10/100 Ethernet, and other standard (3.3 V)	OV_{DD}	—
MVREF1	AN20	I	DDR reference voltage	—
MVREF2	AU32	I	DDR reference voltage	—
SPARE1	B11	I/O	OV_{DD}	8
SPARE3	AH32	—	GV_{DD}	8
SPARE4	AU18	—	GV_{DD}	7
SPARE5	AP1	—	GV_{DD}	8

Table 67. MPC8358E TBGA Pinout Listing (continued)

Signal	Package Pin Number	Pin Type	Power Supply	Notes
No Connect				
NC	AM16, AM17, AM20, AN13, AN16, AN17, AP10, AP11, AP13, AP15, AP18, AR11, AR13, AR14, AR15, AR16, AR17, AR20, AT11, AT12, AT13, AT14, AT16, AT17, AT18, AU10, AU11, AU12, AU13, AU15, AU19	—	—	—

Notes:

1. This pin is an open drain signal. A weak pull-up resistor (1 kΩ) should be placed on this pin to OV_{DD}.
2. This pin is an open drain signal. A weak pull-up resistor (2–10 kΩ) should be placed on this pin to OV_{DD}.
3. This output is actively driven during reset rather than being three-stated during reset.
4. These JTAG pins have weak internal pull-up P-FETs that are always enabled.
5. This pin should have a weak pull up if the chip is in PCI host mode. Follow PCI specifications recommendation.
6. These are On Die Termination pins, used to control DDR2 memories internal termination resistance.
7. This pin must always be tied to GND.
8. This pin must always be left not connected.
9. Refer to *MPC8360E PowerQUICC II Pro Integrated Communications Processor Reference Manual* section on “RGMII Pins,” for information about the two UCC2 Ethernet interface options.
10. This pin must always be tied to GV_{DD}.
11. It is recommended that MDIC0 be tied to GND using an 18.2 Ω resistor and MDIC1 be tied to DDR power using an 18.2 Ω resistor for DDR2.

The QUICC Engine block VCO frequency is derived from the following equations:

ce_clk = (primary clock input × CEPMF) ÷ (1 + CEPDF)

QE VCO Frequency = ce_clk × VCO divider × (1 + CEPDF)

21.4 Suggested PLL Configurations

To simplify the PLL configurations, the device might be separated into two clock domains. The first domain contains the CSB PLL and the core PLL. The core PLL is connected serially to the CSB PLL, and has the csb_clk as its input clock. The second clock domain has the QUICC Engine block PLL. The clock domains are independent, and each of their PLLs are configured separately. Both of the domains has one common input clock. This table shows suggested PLL configurations for 33 and 66 MHz input clocks and illustrates each of the clock domains separately. Any combination of clock domains setting with same input clock are valid. Refer to Section 21, “Clocking,” for the appropriate operating frequencies for your device.

Table 76. Suggested PLL Configurations

Conf No.[1]	SPMF	CORE PLL	CEPMF	CEPDF	Input Clock Freq (MHz)	CSB Freq (MHz)	Core Freq (MHz)	QUICC Engine Freq (MHz)	400 (MHz)	533 (MHz)	667 (MHz)
33 MHz CLKIN/PCI_SYNC_IN Options											
s1	0100	0000100	æ	æ	33	133	266	—	∞	∞	∞
s2	0100	0000101	æ	æ	33	133	333	—	∞	∞	∞
s3	0101	0000100	æ	æ	33	166	333	—	∞	∞	∞
s4	0101	0000101	æ	æ	33	166	416	—	—	∞	∞
s5	0110	0000100	æ	æ	33	200	400	—	∞	∞	∞
s6	0110	0000110	æ	æ	33	200	600	—	—	—	∞
s7	0111	0000011	æ	æ	33	233	350	—	∞	∞	∞
s8	0111	0000100	æ	æ	33	233	466	—	—	∞	∞
s9	0111	0000101	æ	æ	33	233	583	—	—	—	∞
s10	1000	0000011	æ	æ	33	266	400	—	∞	∞	∞
s11	1000	0000100	æ	æ	33	266	533	—	—	∞	∞
s12	1000	0000101	æ	æ	33	266	667	—	—	—	∞
s13	1001	0000010	æ	æ	33	300	300	—	∞	∞	∞
s14	1001	0000011	æ	æ	33	300	450	—	—	∞	∞
s15	1001	0000100	æ	æ	33	300	600	—	—	—	∞
s16	1010	0000010	æ	æ	33	333	333	—	∞	∞	∞
s17	1010	0000011	æ	æ	33	333	500	—	—	∞	∞
s18	1010	0000100	æ	æ	33	333	667	—	—	—	∞
c1	æ	æ	01001	0	33	—	—	300	∞	∞	∞
c2	æ	æ	01100	0	33	—	—	400	∞	∞	∞
c3	æ	æ	01110	0	33	—	—	466	—	∞	∞
c4	æ	æ	01111	0	33	—	—	500	—	∞	∞

$$T_J = T_B + (R_{\theta JB} \times P_D)$$

where:

T_J = junction temperature (° C)
T_B = board temperature at the package perimeter (° C)
$R_{\theta JA}$ = junction to board thermal resistance (° C/W) per JESD51-8
P_D = power dissipation in the package (W)

When the heat loss from the package case to the air can be ignored, acceptable predictions of junction temperature can be made. The application board should be similar to the thermal test condition: the component is soldered to a board with internal planes.

22.2.3 Experimental Determination of Junction Temperature

To determine the junction temperature of the device in the application after prototypes are available, the Thermal Characterization Parameter (Ψ_{JT}) can be used to determine the junction temperature with a measurement of the temperature at the top center of the package case using the following equation:

$$T_J = T_T + (\Psi_{JT} \times P_D)$$

where:

T_J = junction temperature (° C)
T_T = thermocouple temperature on top of package (° C)
Ψ_{JT} = junction-to-ambient thermal resistance (° C/W)
P_D = power dissipation in the package (W)

The thermal characterization parameter is measured per JESD51-2 specification using a 40 gauge type T thermocouple epoxied to the top center of the package case. The thermocouple should be positioned so that the thermocouple junction rests on the package. A small amount of epoxy is placed over the thermocouple junction and over about 1 mm of wire extending from the junction. The thermocouple wire is placed flat against the package case to avoid measurement errors caused by cooling effects of the thermocouple wire.

22.2.4 Heat Sinks and Junction-to-Ambient Thermal Resistance

In some application environments, a heat sink is required to provide the necessary thermal management of the device. When a heat sink is used, the thermal resistance is expressed as the sum of a junction to case thermal resistance and a case to ambient thermal resistance:

$$R_{\theta JA} = R_{\theta JC} + R_{\theta CA}$$

where:

$R_{\theta JA}$ = junction-to-ambient thermal resistance (° C/W)
$R_{\theta JC}$ = junction-to-case thermal resistance (° C/W)
$R_{\theta CA}$ = case-to-ambient thermal resistance (° C/W)

$R_{\theta JC}$ is device related and cannot be influenced by the user. The user controls the thermal environment to change the case-to-ambient thermal resistance, $R_{\theta CA}$. For instance, the user can change the size of the heat sink, the airflow around the device, the interface material, the mounting arrangement on printed-circuit board, or change the thermal dissipation on the printed-circuit board surrounding the device.

To illustrate the thermal performance of the devices with heat sinks, the thermal performance has been simulated with a few commercially available heat sinks. The heat sink choice is determined by the application environment (temperature, airflow, adjacent component power dissipation) and the physical space available. Because there is not a standard application environment, a standard heat sink is not required.

This table shows heat sinks and junction-to-ambient thermal resistance for TBGA package.

Table 78. Heat Sinks and Junction-to-Ambient Thermal Resistance of TBGA Package

Heat Sink Assuming Thermal Grease	Airflow	35 × 35 mm TBGA Junction-to-Ambient Thermal Resistance
AAVID 30 × 30 × 9.4 mm pin fin	Natural convection	10.7
AAVID 30 × 30 × 9.4 mm pin fin	1 m/s	6.2
AAVID 30 × 30 × 9.4 mm pin fin	2 m/s	5.3
AAVID 31 × 35 × 23 mm pin fin	Natural convection	8.1
AAVID 31 × 35 × 23 mm pin fin	1 m/s	4.4
AAVID 31 × 35 × 23 mm pin fin	2 m/s	3.7
Wakefield, 53 × 53 × 25 mm pin fin	Natural convection	5.4
Wakefield, 53 × 53 × 25 mm pin fin	1 m/s	3.2
Wakefield, 53 × 53 × 25 mm pin fin	2 m/s	2.4
MEI, 75 × 85 × 12 no adjacent board, extrusion	Natural convection	6.4
MEI, 75 × 85 × 12 no adjacent board, extrusion	1 m/s	3.8
MEI, 75 × 85 × 12 no adjacent board, extrusion	2 m/s	2.5
MEI, 75 × 85 × 12 mm, adjacent board, 40 mm side bypass	1 m/s	2.8

Accurate thermal design requires thermal modeling of the application environment using computational fluid dynamics software which can model both the conduction cooling and the convection cooling of the air moving through the application. Simplified thermal models of the packages can be assembled using the junction-to-case and junction-to-board thermal resistances listed in the thermal resistance table. More detailed thermal models can be made available on request.

Heat sink vendors include the following:

Aavid Thermalloy 603-224-9988
80 Commercial St.
Concord, NH 03301
Internet: www.aavidthermalloy.com

Alpha Novatech 408-749-7601
473 Sapena Ct. #15
Santa Clara, CA 95054
Internet: www.alphanovatech.com

International Electronic Research Corporation (IERC) 818-842-7277
413 North Moss St.
Burbank, CA 91502
Internet: www.ctscorp.com

This figure shows the PLL power supply filter circuit.

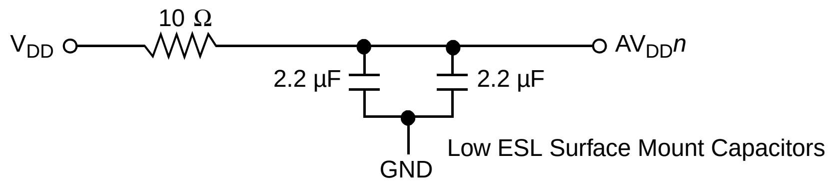

Figure 56. PLL Power Supply Filter Circuit

23.3 Decoupling Recommendations

Due to large address and data buses as well as high operating frequencies, the device can generate transient power surges and high frequency noise in its power supply, especially while driving large capacitive loads. This noise must be prevented from reaching other components in the device system, and the device itself requires a clean, tightly regulated source of power. Therefore, it is recommended that the system designer place at least one decoupling capacitor at each V_{DD}, OV_{DD}, GV_{DD}, and LV_{DD} pins of the device. These decoupling capacitors should receive their power from separate V_{DD}, OV_{DD}, GV_{DD}, LV_{DD}, and GND power planes in the PCB, utilizing short traces to minimize inductance. Capacitors may be placed directly under the device using a standard escape pattern. Others may surround the part.

These capacitors should have a value of 0.01 or 0.1 µF. Only ceramic SMT (surface mount technology) capacitors should be used to minimize lead inductance, preferably 0402 or 0603 sizes.

Additionally, it is recommended that there be several bulk storage capacitors distributed around the PCB, feeding the V_{DD}, OV_{DD}, GV_{DD}, and LV_{DD} planes, to enable quick recharging of the smaller chip capacitors. These bulk capacitors should have a low ESR (equivalent series resistance) rating to ensure the quick response time necessary. They should also be connected to the power and ground planes through two vias to minimize inductance. Suggested bulk capacitors—100–330 µF (AVX TPS tantalum or Sanyo OSCON).

23.4 Connection Recommendations

To ensure reliable operation, it is highly recommended to connect unused inputs to an appropriate signal level. Unused active low inputs should be tied to OV_{DD}, GV_{DD}, or LV_{DD} as required. Unused active high inputs should be connected to GND. All NC (no-connect) signals must remain unconnected.

Power and ground connections must be made to all external V_{DD}, GV_{DD}, LV_{DD}, OV_{DD}, and GND pins of the device.

23.5 Output Buffer DC Impedance

The device drivers are characterized over process, voltage, and temperature. For all buses, the driver is a push-pull single-ended driver type (open drain for I^2C).

To measure Z_0 for the single-ended drivers, an external resistor is connected from the chip pad to OV_{DD} or GND. Then, the value of each resistor is varied until the pad voltage is $OV_{DD}/2$ (see Figure 57). The output impedance is the average of two components, the resistances of the pull-up and pull-down devices. When data is held high, SW1 is closed (SW2 is open) and R_P is trimmed until the voltage at the pad equals $OV_{DD}/2$. R_P then becomes the resistance of the pull-up devices. R_P and R_N are designed to be close to each other in value. Then, $Z_0 = (R_P + R_N)/2$.