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Understanding Embedded - Microprocessors

Embedded microprocessors are specialized computing chips designed to perform specific tasks within an embedded system. Unlike general-purpose microprocessors found in personal computers, embedded microprocessors are tailored for dedicated functions within larger systems, offering optimized performance, efficiency, and reliability. These microprocessors are integral to the operation of countless electronic devices, providing the computational power necessary for controlling processes, handling data, and managing communications.

Applications of Embedded - Microprocessors

Embedded microprocessors are utilized across a broad spectrum of applications, making them indispensable in

Details	
Product Status	Obsolete
Core Processor	PowerPC e300
Number of Cores/Bus Width	1 Core, 32-Bit
Speed	266MHz
Co-Processors/DSP	Communications; QUICC Engine
RAM Controllers	DDR, DDR2
Graphics Acceleration	No
Display & Interface Controllers	-
Ethernet	10/100/1000Mbps (1)
SATA	-
USB	USB 1.x (1)
Voltage - I/O	1.8V, 2.5V, 3.3V
Operating Temperature	-40°C ~ 105°C (TA)
Security Features	-
Package / Case	740-LBGA
Supplier Device Package	740-TBGA (37.5x37.5)
Purchase URL	https://www.e-xfl.com/product-detail/nxp-semiconductors/mpc8358czuaddea

This figure shows the MPC8358E block diagram.

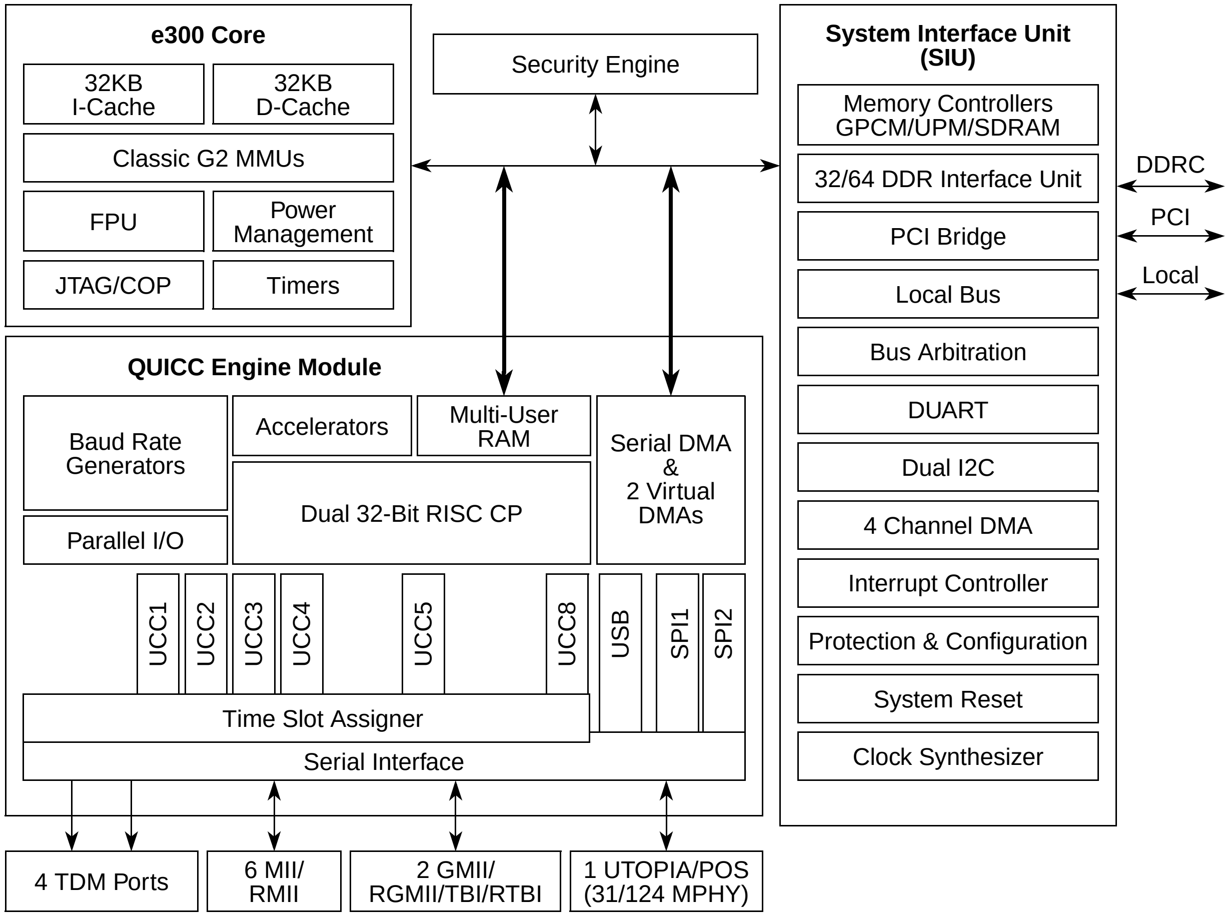

Figure 2. MPC8358E Block Diagram

Major features of the MPC8360E/58E are as follows:

- e300 PowerPC processor core (enhanced version of the MPC603e core)
  - Operates at up to 667 MHz (for the MPC8360E) and 400 MHz (for the MPC8358E)
  - High-performance, superscalar processor core
  - Floating-point, integer, load/store, system register, and branch processing units
  - 32-Kbyte instruction cache, 32-Kbyte data cache
  - Lockable portion of L1 cache
  - Dynamic power management
  - Software-compatible with the Freescale processor families implementing the Power Architecture™ technology
- QUICC Engine unit
  - Two 32-bit RISC controllers for flexible support of the communications peripherals, each operating up to 500 MHz (for the MPC8360E) and 400 MHz (for the MPC8358E)
  - Serial DMA channel for receive and transmit on all serial channels
  - QUICC Engine module peripheral request interface (for SEC, PCI, IEEE Std. 1588™)
  - Eight universal communication controllers (UCCs) on the MPC8360E and six UCCs on the MPC8358E supporting the following protocols and interfaces (not all of them simultaneously):
    - IEEE 1588 protocol supported

- DRAM chip configurations from 64 Mbits to 1 Gigabit with ×8/×16 data ports
- Full ECC support (when the MPC8360E is configured as 2×32-bit DDR memory controllers, both support ECC)
- Page mode support (up to 16 simultaneous open pages for DDR1, up to 32 simultaneous open pages for DDR2)
- Contiguous or discontiguous memory mapping
- Read-modify-write support
- Sleep mode support for self refresh SDRAM
- Supports auto refreshing
- Supports source clock mode
- On-the-fly power management using CKE
- Registered DIMM support
- 2.5-V SSTL2 compatible I/O for DDR1, 1.8-V SSTL2 compatible I/O for DDR2
- External driver impedance calibration
- On-die termination (ODT)

- PCI interface
  - PCI Specification Revision 2.3 compatible
  - Data bus widths:
    - Single 32-bit data PCI interface that operates at up to 66 MHz
  - PCI 3.3-V compatible (not 5-V compatible)
  - PCI host bridge capabilities on both interfaces
  - PCI agent mode supported on PCI interface
  - Support for PCI-to-memory and memory-to-PCI streaming
  - Memory prefetching of PCI read accesses and support for delayed read transactions
  - Support for posting of processor-to-PCI and PCI-to-memory writes
  - On-chip arbitration, supporting five masters on PCI
  - Support for accesses to all PCI address spaces
  - Parity support
  - Selectable hardware-enforced coherency
  - Address translation units for address mapping between host and peripheral
  - Dual address cycle supported when the device is the target
  - Internal configuration registers accessible from PCI
- Local bus controller (LBC)
  - Multiplexed 32-bit address and data operating at up to 133 MHz
  - Eight chip selects support eight external slaves
  - Up to eight-beat burst transfers
  - 32-, 16-, and 8-bit port sizes are controlled by an on-chip memory controller
  - Three protocol engines available on a per chip select basis:
    - General-purpose chip select machine (GPCM)
    - Three user programmable machines (UPMs)
    - Dedicated single data rate SDRAM controller
  - Parity support
  - Default boot ROM chip select with configurable bus width (8-, 16-, or 32-bit)
- Programmable interrupt controller (PIC)
  - Functional and programming compatibility with the MPC8260 interrupt controller
  - Support for 8 external and 35 internal discrete interrupt sources
  - Support for one external (optional) and seven internal machine checkstop interrupt sources

- Programmable highest priority request
- Four groups of interrupts with programmable priority
- External and internal interrupts directed to communication processor
- Redirects interrupts to external $\overline{\text{INTA}}$ pin when in core disable mode
- Unique vector number for each interrupt source

- Dual industry-standard I^2C interfaces
  - Two-wire interface
  - Multiple master support
  - Master or slave I^2C mode support
  - On-chip digital filtering rejects spikes on the bus
  - System initialization data is optionally loaded from I^2C-1 EPROM by boot sequencer embedded hardware
- DMA controller
  - Four independent virtual channels
  - Concurrent execution across multiple channels with programmable bandwidth control
  - All channels accessible by local core and remote PCI masters
  - Misaligned transfer capability
  - Data chaining and direct mode
  - Interrupt on completed segment and chain
  - DMA external handshake signals: $\overline{\text{DMA\_DREQ}}$[0:3]/$\overline{\text{DMA\_DACK}}$[0:3]/$\overline{\text{DMA\_DONE}}$[0:3]. There is one set for each DMA channel. The pins are multiplexed to the parallel IO pins with other QE functions.
- DUART
  - Two 4-wire interfaces (RxD, TxD, RTS, CTS)
  - Programming model compatible with the original 16450 UART and the PC16550D
- System timers
  - Periodic interrupt timer
  - Real-time clock
  - Software watchdog timer
  - Eight general-purpose timers
- IEEE Std. 1149.1™-compliant, JTAG boundary scan
- Integrated PCI bus and SDRAM clock generation

2 Electrical Characteristics

This section provides the AC and DC electrical specifications and thermal characteristics for the MPC8360E/58E. The device is currently targeted to these specifications. Some of these specifications are independent of the I/O cell, but are included for a more complete reference. These are not purely I/O buffer design specifications.

4.1 DC Electrical Characteristics

This table provides the clock input (CLKIN/PCI_SYNC_IN) DC timing specifications for the device.

Table 7. CLKIN DC Electrical Characteristics

Parameter	Condition	Symbol	Min	Max	Unit
Input high voltage	—	V_{IH}	2.7	OV_{DD} + 0.3	V
Input low voltage	—	V_{IL}	–0.3	0.4	V
CLKIN input current	0 V ≤ V_{IN} ≤ OV_{DD}	I_{IN}	—	±10	μA
PCI_SYNC_IN input current	0 V ≤ V_{IN} ≤ 0.5V or OV_{DD} – 0.5V ≤ V_{IN} ≤ OV_{DD}	I_{IN}	—	±10	μA
PCI_SYNC_IN input current	0.5 V ≤ V_{IN} ≤ OV_{DD} – 0.5 V	I_{IN}	—	±100	μA

4.2 AC Electrical Characteristics

The primary clock source for the device can be one of two inputs, CLKIN or PCI_CLK, depending on whether the device is configured in PCI host or PCI agent mode. This table provides the clock input (CLKIN/PCI_CLK) AC timing specifications for the device.

Table 8. CLKIN AC Timing Specifications

Parameter/Condition	Symbol	Min	Typical	Max	Unit	Notes
CLKIN/PCI_CLK frequency	f_{CLKIN}	—	—	66.67	MHz	1
CLKIN/PCI_CLK cycle time	t_{CLKIN}	15	—	—	ns	—
CLKIN/PCI_CLK rise and fall time	t_{KH}, t_{KL}	0.6	1.0	2.3	ns	2
CLKIN/PCI_CLK duty cycle	t_{KHK}/t_{CLKIN}	40	—	60	%	3
CLKIN/PCI_CLK jitter	—	—	—	±150	ps	4, 5

Notes:
1. **Caution:** The system, core, USB, security, and 10/100/1000 Ethernet must not exceed their respective maximum or minimum operating frequencies.
2. Rise and fall times for CLKIN/PCI_CLK are measured at 0.4 V and 2.7 V.
3. Timing is guaranteed by design and characterization.
4. This represents the total input jitter—short term and long term—and is guaranteed by design.
5. The CLKIN/PCI_CLK driver's closed loop jitter bandwidth should be <500 kHz at –20 dB. The bandwidth must be set low to allow cascade-connected PLL-based devices to track CLKIN drivers with the specified jitter.

4.3 Gigabit Reference Clock Input Timing

This table provides the Gigabit reference clocks (GTX_CLK125) AC timing specifications.

Table 9. GTX_CLK125 AC Timing Specifications

At recommended operating conditions with LV_{DD} = 2.5 ± 0.125 mV/ 3.3 V ± 165 mV

Parameter/Condition	Symbol	Min	Typical	Max	Unit	Notes
GTX_CLK125 frequency	t_{G125}	—	125	—	MHz	—
GTX_CLK125 cycle time	t_{G125}	—	8	—	ns	—

This figure provides the AC test load for the DDR bus.

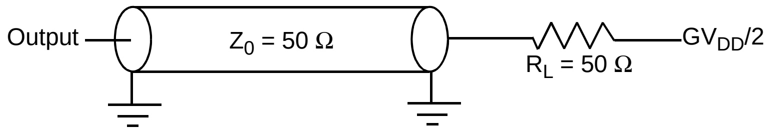

Figure 8. DDR AC Test Load

Table 22. DDR and DDR2 SDRAM Measurement Conditions

Symbol	DDR	DDR2	Unit	Notes
V_{TH}	$MV_{REF} \pm 0.31$ V	$MV_{REF} \pm 0.25$ V	V	1
V_{OUT}	$0.5 \times GV_{DD}$	$0.5 \times GV_{DD}$	V	2

Notes:
1. Data input threshold measurement point.
2. Data output measurement point.

This figure shows the DDR SDRAM output timing diagram for source synchronous mode.

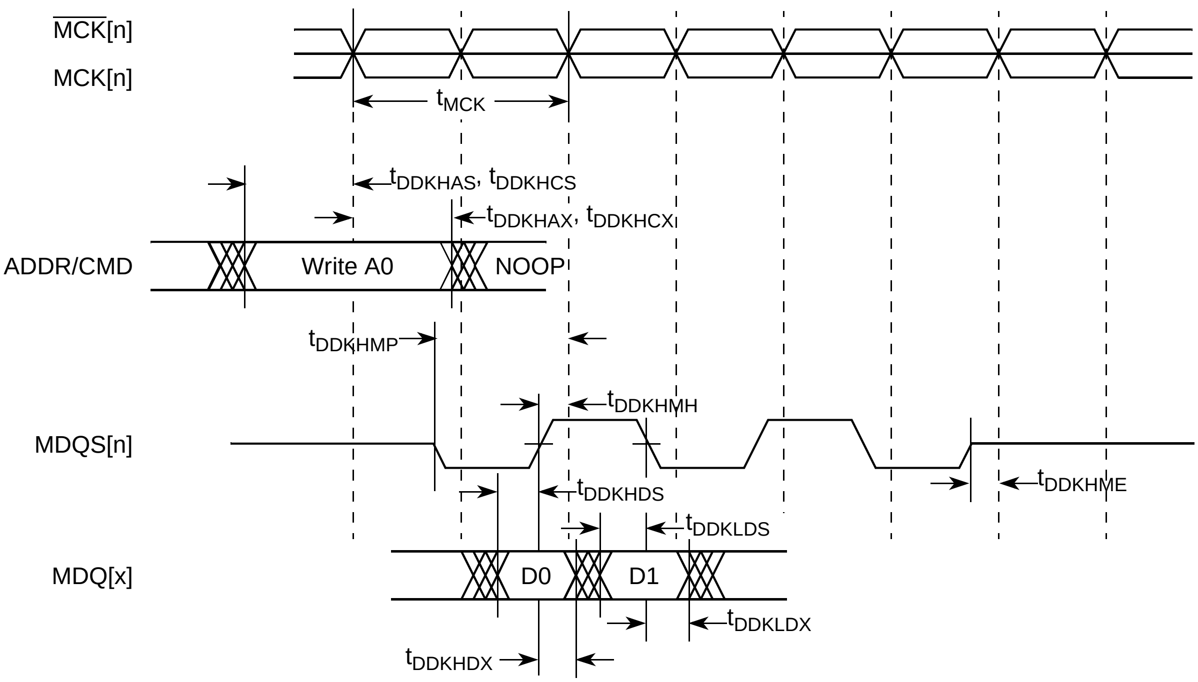

Figure 9. DDR SDRAM Output Timing Diagram for Source Synchronous Mode

8.2.2 MII AC Timing Specifications

This section describes the MII transmit and receive AC timing specifications.

8.2.2.1 MII Transmit AC Timing Specifications

This table provides the MII transmit AC timing specifications.

Table 29. MII Transmit AC Timing Specifications

At recommended operating conditions with LV_{DD}/OV_{DD} of 3.3 V ± 10%.

Parameter/Condition	Symbol[1]	Min	Typ	Max	Unit
TX_CLK clock period 10 Mbps	t_{MTX}	—	400	—	ns
TX_CLK clock period 100 Mbps	t_{MTX}	—	40	—	ns
TX_CLK duty cycle	t_{MTXH}/t_{MTX}	35	—	65	%
TX_CLK to MII data TXD[3:0], TX_ER, TX_EN delay	t_{MTKHDX} t_{MTKHDV}	1 —	5	— 15	ns
TX_CLK data clock rise time, (20% to 80%)	t_{MTXR}	1.0	—	4.0	ns
TX_CLK data clock fall time, (80% to 20%)	t_{MTXF}	1.0	—	4.0	ns

Note:

1. The symbols used for timing specifications follow the pattern of $t_{\text{(first two letters of functional block)(signal)(state)(reference)(state)}}$ for inputs and $t_{\text{(first two letters of functional block)(reference)(state)(signal)(state)}}$ for outputs. For example, t_{MTKHDX} symbolizes MII transmit timing (MT) for the time t_{MTX} clock reference (K) going high (H) until data outputs (D) are invalid (X). Note that, in general, the clock reference symbol representation is based on two to three letters representing the clock of a particular functional. For example, the subscript of t_{MTX} represents the MII(M) transmit (TX) clock. For rise and fall times, the latter convention is used with the appropriate letter: R (rise) or F (fall).

This figure shows the MII transmit AC timing diagram.

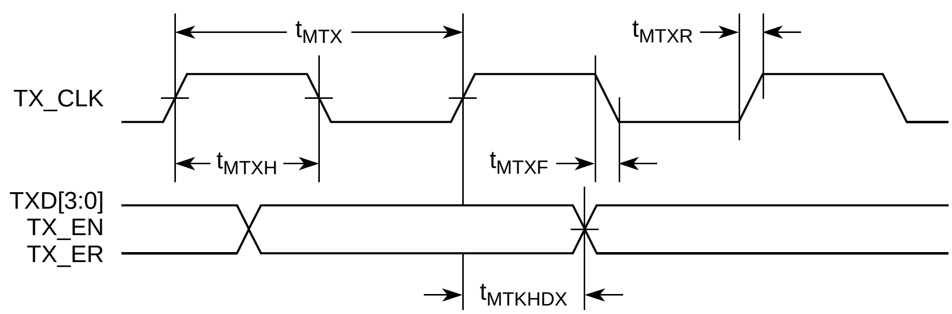

Figure 12. MII Transmit AC Timing Diagram

Table 32. RMII Receive AC Timing Specifications (continued)

At recommended operating conditions with LV_{DD}/OV_{DD} of 3.3 V ± 10%.

Parameter/Condition	Symbol[1]	Min	Typ	Max	Unit
RXD[1:0], CRS_DV, RX_ER setup time to REF_CLK	$t_{RMRDVKH}$	4.0	—	—	ns
RXD[1:0], CRS_DV, RX_ER hold time to REF_CLK	$t_{RMRDXKH}$	2.0	—	—	ns
REF_CLK clock rise time	t_{RMXR}	1.0	—	4.0	ns
REF_CLK clock fall time	t_{RMXF}	1.0	—	4.0	ns

Note:

1. The symbols used for timing specifications follow the pattern of $t_{\text{(first three letters of functional block)(signal)(state)(reference)(state)}}$ for inputs and $t_{\text{(first two letters of functional block)(reference)(state)(signal)(state)}}$ for outputs. For example, $t_{RMRDVKH}$ symbolizes RMII receive timing (RMR) with respect to the time data input signals (D) reach the valid state (V) relative to the t_{RMX} clock reference (K) going to the high (H) state or setup time. Also, $t_{RMRDXKL}$ symbolizes RMII receive timing (RMR) with respect to the time data input signals (D) went invalid (X) relative to the t_{RMX} clock reference (K) going to the low (L) state or hold time. Note that, in general, the clock reference symbol representation is based on three letters representing the clock of a particular functional. For example, the subscript of t_{RMX} represents the RMII (RM) reference (X) clock. For rise and fall times, the latter convention is used with the appropriate letter: R (rise) or F (fall).

This figure provides the AC test load.

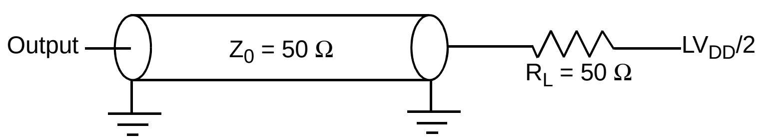

Figure 16. AC Test Load

This figure shows the RMII receive AC timing diagram.

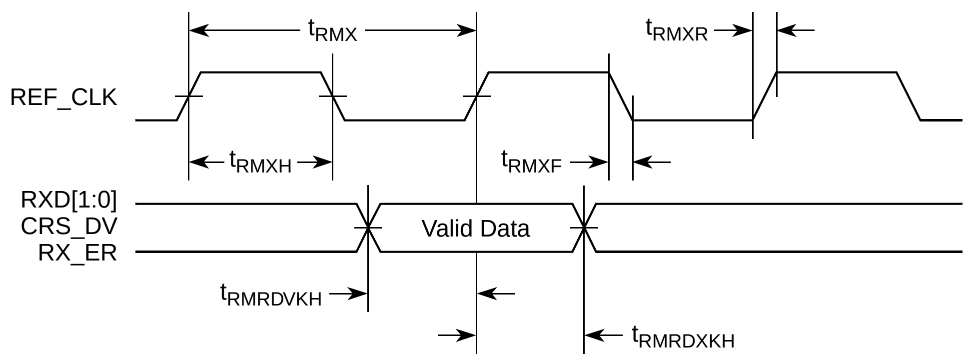

Figure 17. RMII Receive AC Timing Diagram

8.2.4 TBI AC Timing Specifications

This section describes the TBI transmit and receive AC timing specifications.

8.3.2 MII Management AC Electrical Specifications

This table provides the MII management AC timing specifications.

Table 37. MII Management AC Timing Specifications

At recommended operating conditions with LV_{DD} is 3.3 V ± 10%.

Parameter/Condition	Symbol[1]	Min	Typ	Max	Unit	Notes
MDC frequency	f_{MDC}	—	2.5	—	MHz	2
MDC period	t_{MDC}	—	400	—	ns	—
MDC clock pulse width high	t_{MDCH}	32	—	—	ns	—
MDC to MDIO delay	$t_{MDTKHDX}$ $t_{MDTKHDV}$	10 —	—	— 110	ns	3
MDIO to MDC setup time	$t_{MDRDVKH}$	10	—	—	ns	—
MDIO to MDC hold time	$t_{MDRDXKH}$	0	—	—	ns	—
MDC rise time	t_{MDCR}	—	—	10	ns	—
MDC fall time	t_{MDHF}	—	—	10	ns	—

Notes:

1. The symbols used for timing specifications follow the pattern of $t_{\text{(first two letters of functional block)(signal)(state)(reference)(state)}}$ for inputs and $t_{\text{(first two letters of functional block)(reference)(state)(signal)(state)}}$ for outputs. For example, t_{MDKHDX} symbolizes management data timing (MD) for the time t_{MDC} from clock reference (K) high (H) until data outputs (D) are invalid (X) or data hold time. Also, $t_{MDRDVKH}$ symbolizes management data timing (MD) with respect to the time data input signals (D) reach the valid state (V) relative to the t_{MDC} clock reference (K) going to the high (H) state or setup time. For rise and fall times, the latter convention is used with the appropriate letter: R (rise) or F (fall).
2. This parameter is dependent on the csb_clk speed (that is, for a csb_clk of 267 MHz, the maximum frequency is 8.3 MHz and the minimum frequency is 1.2 MHz; for a csb_clk of 375 MHz, the maximum frequency is 11.7 MHz and the minimum frequency is 1.7 MHz).
3. This parameter is dependent on the ce_clk speed (that is, for a ce_clk of 200 MHz, the delay is 90 ns and for a ce_clk of 300 MHz, the delay is 63 ns).

This figure shows the MII management AC timing diagram.

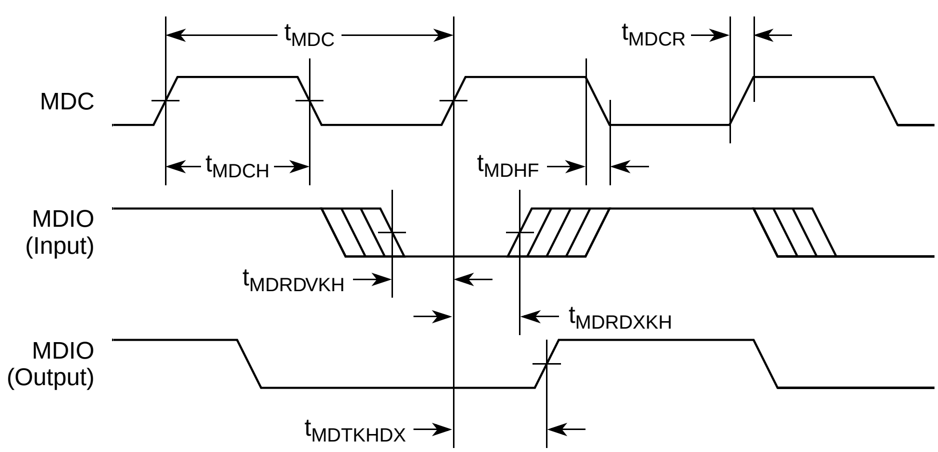

Figure 21. MII Management Interface Timing Diagram

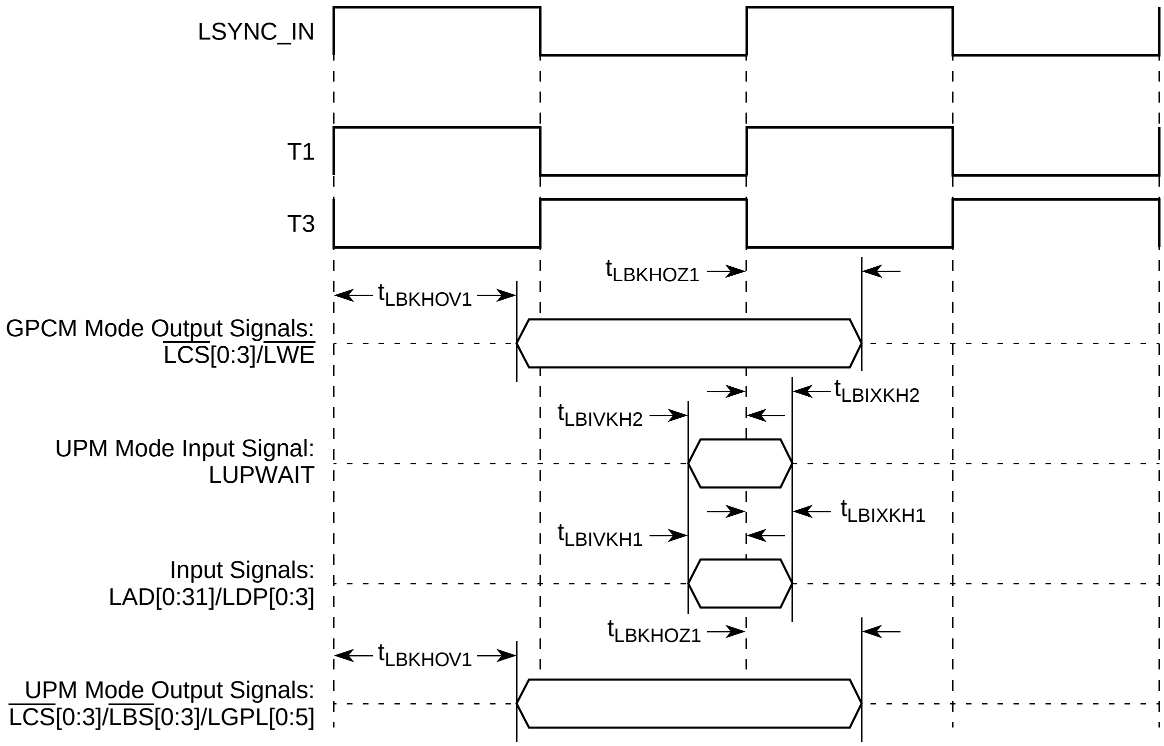

Figure 25. Local Bus Signals, GPCM/UPM Signals for LCRR[CLKDIV] = 2 (DLL Enabled)

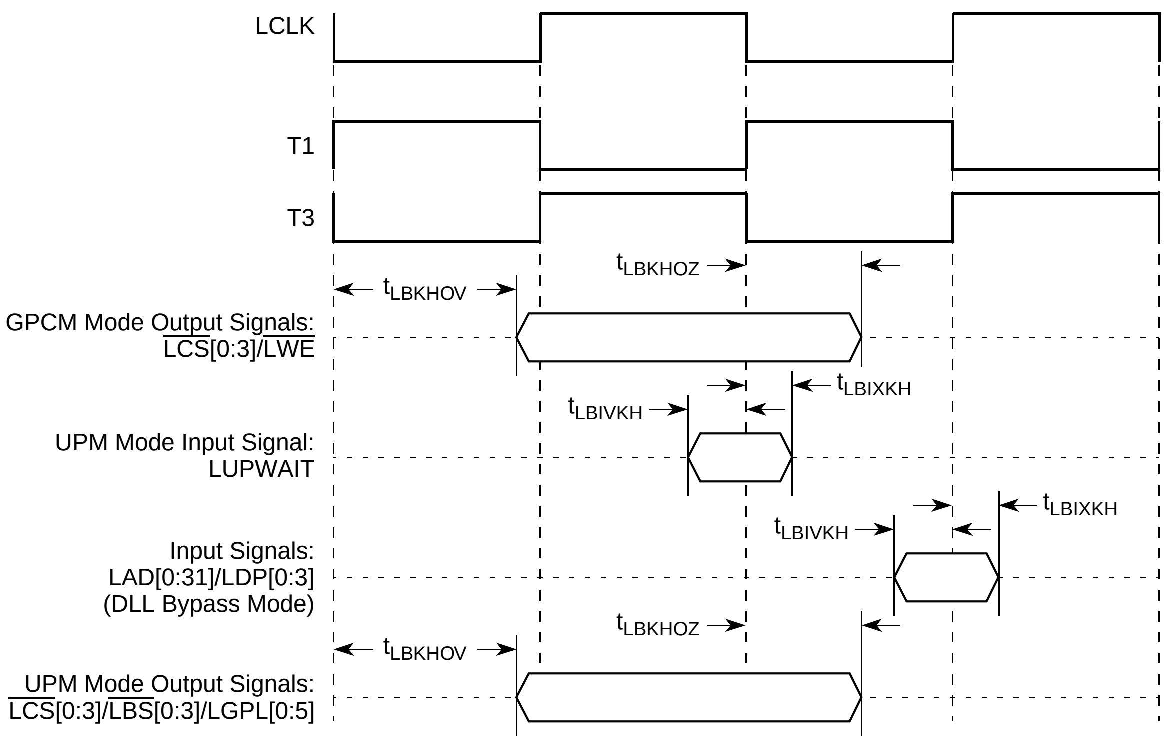

Figure 26. Local Bus Signals, GPCM/UPM Signals for LCRR[CLKDIV] = 2 (DLL Bypass Mode)

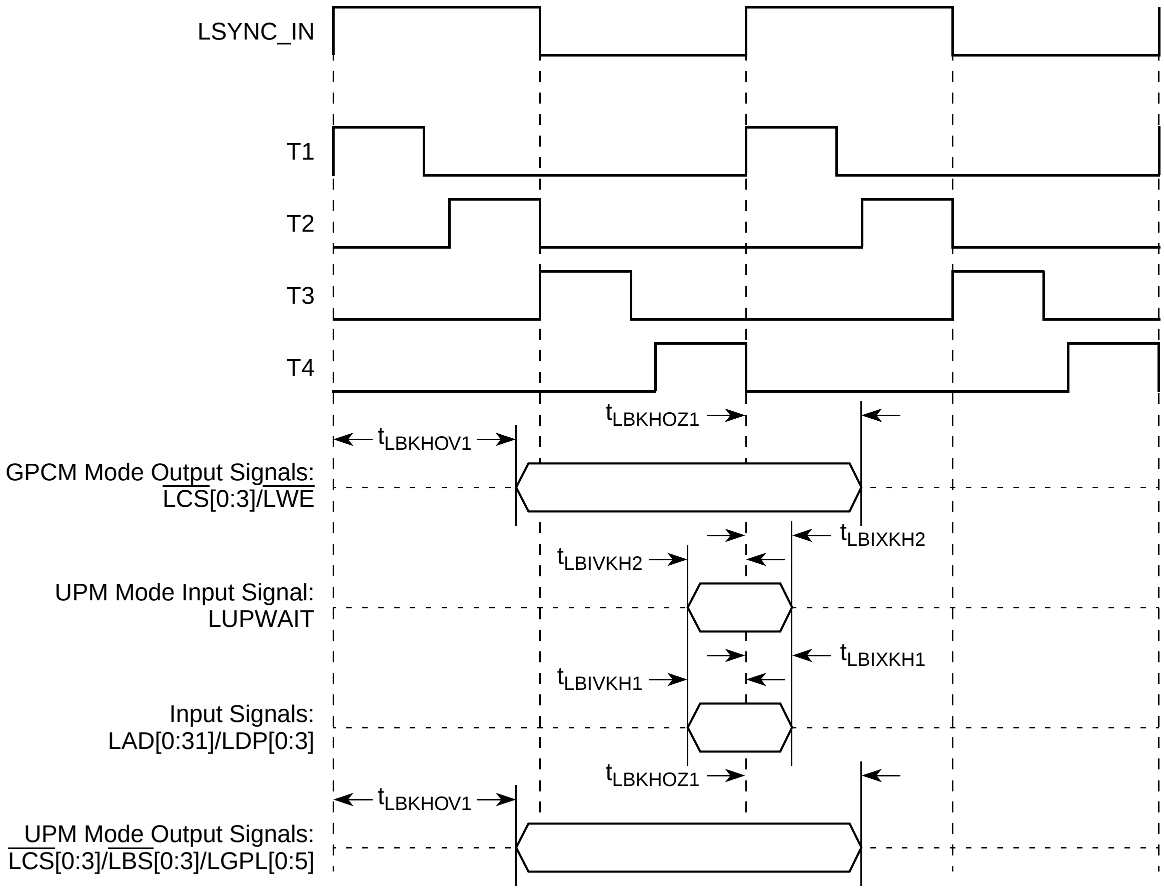

Figure 28. Local Bus Signals, GPCM/UPM Signals for LCRR[CLKDIV] = 4 (DLL Enabled)

10 JTAG

This section describes the DC and AC electrical specifications for the IEEE 1149.1 (JTAG) interface of the MPC8360E/58E.

10.1 JTAG DC Electrical Characteristics

This table provides the DC electrical characteristics for the IEEE 1149.1 (JTAG) interface of the device.

Table 42. JTAG interface DC Electrical Characteristics

Characteristic	Symbol	Condition	Min	Max	Unit
Output high voltage	V_{OH}	I_{OH} = –6.0 mA	2.4	—	V
Output low voltage	V_{OL}	I_{OL} = 6.0 mA	—	0.5	V
Output low voltage	V_{OL}	I_{OL} = 3.2 mA	—	0.4	V
Input high voltage	V_{IH}	—	2.5	OV_{DD} + 0.3	V
Input low voltage	V_{IL}	—	–0.3	0.8	V
Input current	I_{IN}	0 V ≤ V_{IN} ≤ OV_{DD}	—	±10	μA

This figure provides the AC test load for TDO and the boundary-scan outputs of the device.

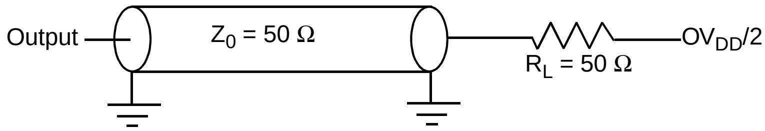

Figure 29. AC Test Load for the JTAG Interface

This figure provides the JTAG clock input timing diagram.

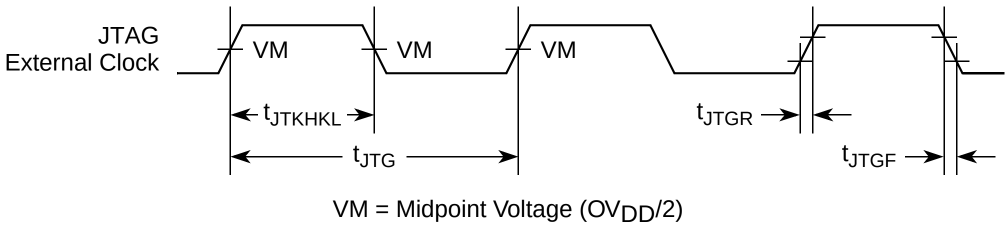

Figure 30. JTAG Clock Input Timing Diagram

This figure provides the $\overline{\text{TRST}}$ timing diagram.

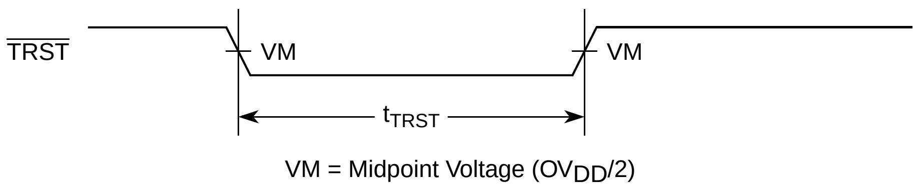

Figure 31. $\overline{\text{TRST}}$ Timing Diagram

This figure provides the boundary-scan timing diagram.

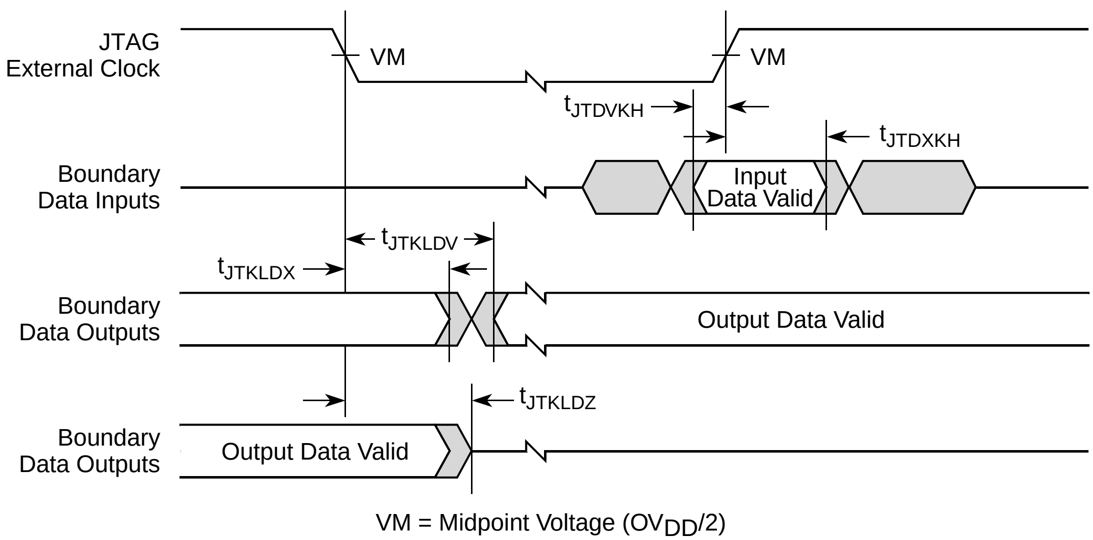

Figure 32. Boundary-Scan Timing Diagram

This figure provides the test access port timing diagram.

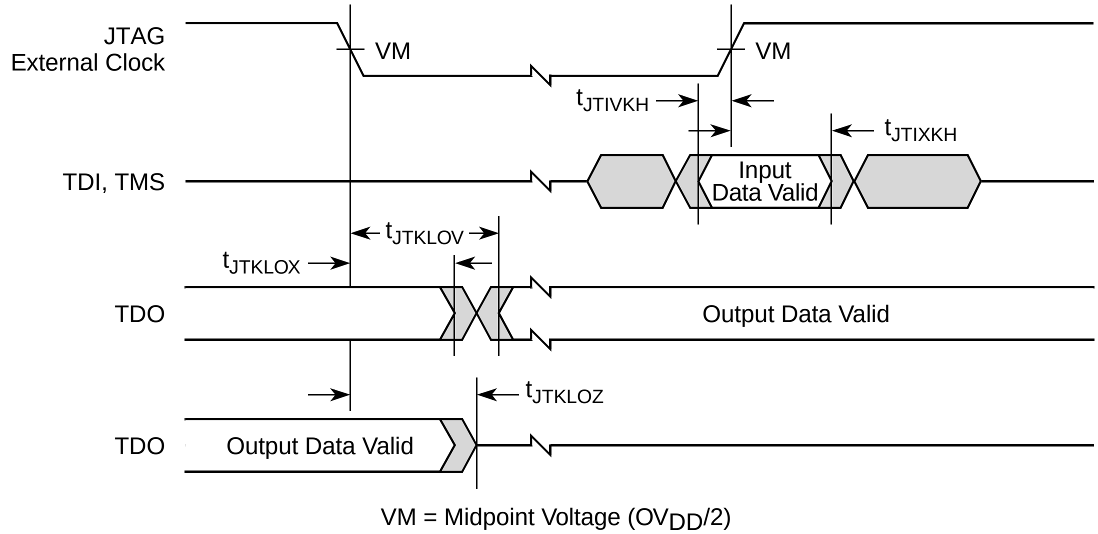

Figure 33. Test Access Port Timing Diagram

11 I²C

This section describes the DC and AC electrical characteristics for the I²C interface of the MPC8360E/58E.

11.1 I²C DC Electrical Characteristics

This table provides the DC electrical characteristics for the I²C interface of the device.

Table 44. I²C DC Electrical Characteristics

At recommended operating conditions with OV_{DD} of 3.3 V ± 10%.

Parameter	Symbol	Min	Max	Unit	Notes
Input high voltage level	V_{IH}	$0.7 \times OV_{DD}$	$OV_{DD} + 0.3$	V	—
Input low voltage level	V_{IL}	–0.3	$0.3 \times OV_{DD}$	V	—
Low level output voltage	V_{OL}	0	0.4	V	1
Output fall time from V_{IH}(min) to V_{IL}(max) with a bus capacitance from 10 to 400 pF	t_{I2KLKV}	$20 + 0.1 \times C_B$	250	ns	2
Pulse width of spikes which must be suppressed by the input filter	t_{I2KHKL}	0	50	ns	3
Capacitance for each I/O pin	C_I	—	10	pF	—
Input current (0 V $\leq V_{IN} \leq OV_{DD}$)	I_{IN}	—	±10	μA	4

Notes:
1. Output voltage (open drain or open collector) condition = 3 mA sink current.
2. C_B = capacitance of one bus line in pF.
3. Refer to the *MPC8360E Integrated Communications Processor Reference Manual* for information on the digital filter used.
4. I/O pins obstruct the SDA and SCL lines if OV_{DD} is switched off.

11.2 I^2C AC Electrical Specifications

This table provides the AC timing parameters for the I^2C interface of the device.

Table 45. I^2C AC Electrical Specifications

All values refer to V_{IH} (min) and V_{IL} (max) levels (see Table 44).

Parameter	Symbol[1]	Min	Max	Unit	Note
SCL clock frequency	f_{I2C}	0	400	kHz	2
Low period of the SCL clock	t_{I2CL}	1.3	—	µs	—
High period of the SCL clock	t_{I2CH}	0.6	—	µs	—
Setup time for a repeated START condition	t_{I2SVKH}	0.6	—	µs	—
Hold time (repeated) START condition (after this period, the first clock pulse is generated)	t_{I2SXKL}	0.6	—	µs	—
Data setup time	t_{I2DVKH}	100	—	ns	3
Data hold time: CBUS compatible masters I^2C bus devices	t_{I2DXKL}	 — 0^2	 — 0.9^3	µs	—
Rise time of both SDA and SCL signals	t_{I2CR}	$20 + 0.1\ C_b{}^4$	300	ns	—
Fall time of both SDA and SCL signals	t_{I2CF}	$20 + 0.1\ C_b{}^4$	300	ns	—
Set-up time for STOP condition	t_{I2PVKH}	0.6	—	µs	—
Bus free time between a STOP and START condition	t_{I2KHDX}	1.3	—	µs	—
Noise margin at the LOW level for each connected device (including hysteresis)	V_{NL}	$0.1 \times OV_{DD}$	—	V	—
Noise margin at the HIGH level for each connected device (including hysteresis)	V_{NH}	$0.2 \times OV_{DD}$	—	V	—

Notes:

1. The symbols used for timing specifications follow the pattern of $t_{\text{(first two letters of functional block)(signal)(state)(reference)(state)}}$ for inputs and $t_{\text{(first two letters of functional block)(reference)(state)(signal)(state)}}$ for outputs. For example, t_{I2DVKH} symbolizes I^2C timing (I2) with respect to the time data input signals (D) reach the valid state (V) relative to the t_{I2C} clock reference (K) going to the high (H) state or setup time. Also, t_{I2SXKL} symbolizes I^2C timing (I2) for the time that the data with respect to the start condition (S) went invalid (X) relative to the t_{I2C} clock reference (K) going to the low (L) state or hold time. Also, t_{I2PVKH} symbolizes I^2C timing (I2) for the time that the data with respect to the stop condition (P) reaching the valid state (V) relative to the t_{I2C} clock reference (K) going to the high (H) state or setup time. For rise and fall times, the latter convention is used with the appropriate letter: R (rise) or F (fall).
2. The device provides a hold time of at least 300 ns for the SDA signal (referred to the V_{IH} min of the SCL signal) to bridge the undefined region of the falling edge of SCL.
3. The maximum t_{I2DVKH} has only to be met if the device does not stretch the LOW period (t_{I2CL}) of the SCL signal.
4. C_B = capacitance of one bus line in pF.

This figure provides the AC test load for the I^{2}C.

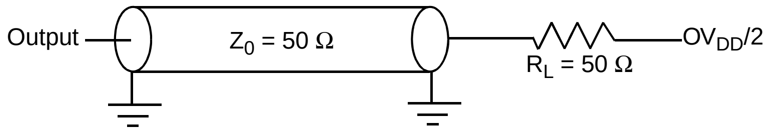

Figure 34. I^{2}C AC Test Load

This figure shows the AC timing diagram for the I^{2}C bus.

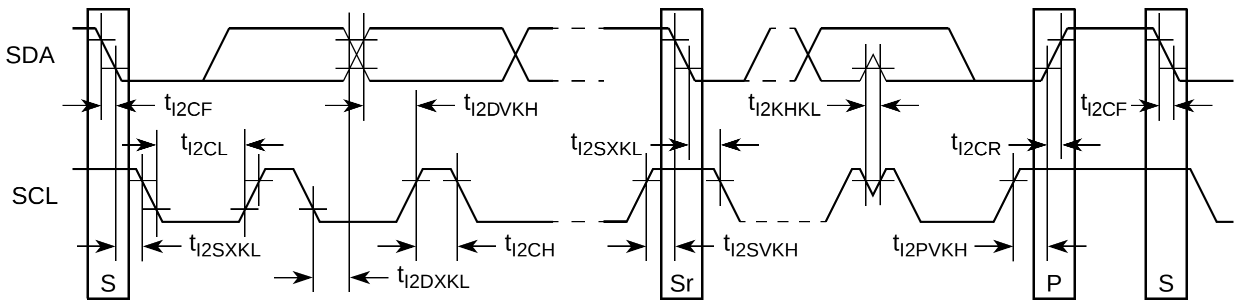

Figure 35. I^{2}C Bus AC Timing Diagram

12 PCI

This section describes the DC and AC electrical specifications for the PCI bus of the MPC8360E/58E.

12.1 PCI DC Electrical Characteristics

This table provides the DC electrical characteristics for the PCI interface of the device.

Table 46. PCI DC Electrical Characteristics

Parameter	Symbol	Test Condition	Min	Max	Unit
High-level input voltage	V_{IH}	$V_{OUT} \ge V_{OH}$ (min) or $V_{OUT} \le V_{OL}$ (max)	0.5 × OV_{DD}	OV_{DD} + 0.5	V
Low-level input voltage	V_{IL}		-0.5	0.3 × OV_{DD}	V
High-level output voltage	V_{OH}	I_{OH} = –500 µA	0.9 × OV_{DD}	—	V
Low-level output voltage	V_{OL}	I_{OL} = 1500 µA	—	0.1 × OV_{DD}	V
Input current	I_{IN}	0 V ≤ V_{IN}[1] ≤ OV_{DD}	—	±10	µA

12.2 PCI AC Electrical Specifications

This section describes the general AC timing parameters of the PCI bus of the device. Note that the PCI_CLK or PCI_SYNC_IN signal is used as the PCI input clock depending on whether the device is configured as a host or agent device. This table provides the PCI AC timing specifications at 66 MHz.

Table 47. PCI AC Timing Specifications at 66 MHz

Parameter	Symbol[1]	Min	Max	Unit	Notes
Clock to output valid	t_{PCKHOV}	—	6.0	ns	2, 5
Output hold from clock	t_{PCKHOX}	1	—	ns	2

18.3 AC Test Load

These figures represent the AC timing from Table 62 and Table 63. Note that although the specifications generally reference the rising edge of the clock, these AC timing diagrams also apply when the falling edge is the active edge.

This figure shows the timing with external clock.

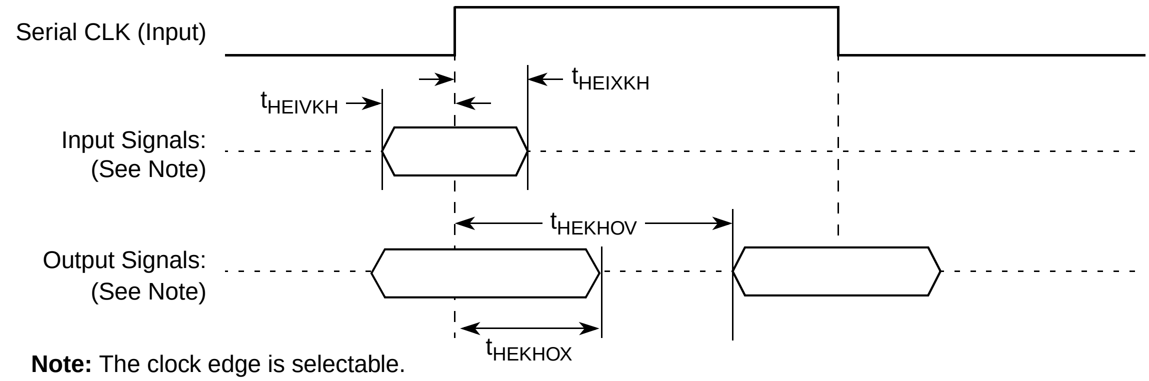

Figure 50. AC Timing (External Clock) Diagram

This figure shows the timing with internal clock.

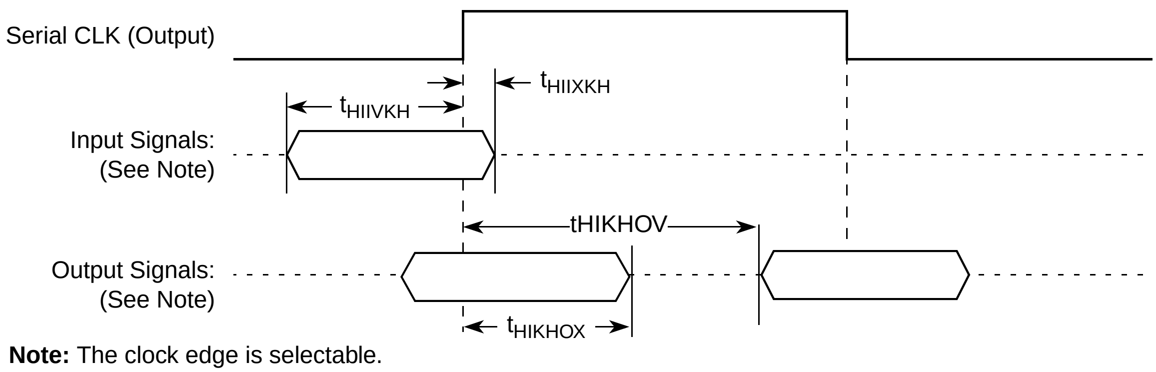

Figure 51. AC Timing (Internal Clock) Diagram

Table 66. MPC8360E TBGA Pinout Listing (continued)

Signal	Package Pin Number	Pin Type	Power Supply	Notes
PCI_DEVSEL/CE_PF[16]	E26	I/O	OV_{DD}	5
PCI_IDSEL/CE_PF[17]	F22	I/O	OV_{DD}	—
PCI_SERR/CE_PF[18]	B29	I/O	OV_{DD}	5
PCI_PERR/CE_PF[19]	A29	I/O	OV_{DD}	5
PCI_REQ[0]/CE_PF[20]	F19	I/O	$LV_{DD}2$	—
PCI_REQ[1]/CPCI_HS_ES/ CE_PF[21]	A21	I/O	$LV_{DD}2$	—
PCI_REQ[2]/CE_PF[22]	C21	I/O	$LV_{DD}2$	—
PCI_GNT[0]/CE_PF[23]	E20	I/O	$LV_{DD}2$	—
PCI_GNT[1]/CPCI1_HS_LED/ CE_PF[24]	B20	I/O	$LV_{DD}2$	—
PCI_GNT[2]/CPCI1_HS_ENUM/ CE_PF[25]	C20	I/O	$LV_{DD}2$	—
PCI_MODE	D36	I	OV_{DD}	—
M66EN/CE_PF[4]	B37	I/O	OV_{DD}	—
Local Bus Controller Interface				
LAD[0:31]	N32, N33, N35, N36, P37, P32, P34, R36, R35, R34, R33, T37, T35, T34, T33, U37, T32, U36, U34, V36, V35, W37, W35, V33, V32, W34, Y36, W32, AA37, Y33, AA35, AA34	I/O	OV_{DD}	—
LDP[0]/CKSTOP_OUT	AB37	I/O	OV_{DD}	—
LDP[1]/CKSTOP_IN	AB36	I/O	OV_{DD}	—
LDP[2]/LCS[6]	AB35	I/O	OV_{DD}	—
LDP[3]/LCS[7]	AA33	I/O	OV_{DD}	—
LA[27:31]	AC37, AA32, AC36, AC34, AD36	O	OV_{DD}	—
LCS[0:5]	AD33, AG37, AF34, AE33, AD32, AH37	O	OV_{DD}	—
LWE[0:3]/LSDDQM[0:3]/LBS[0:3]	AG35, AG34, AH36, AE32	O	OV_{DD}	—
LBCTL	AD35	O	OV_{DD}	—
LALE	M37	O	OV_{DD}	—
LGPL0/LSDA10/cfg_reset_source0	AB32	I/O	OV_{DD}	—
LGPL1/LSDWE/cfg_reset_source1	AE37	I/O	OV_{DD}	—
LGPL2/LSDRAS/LOE	AC33	O	OV_{DD}	—
LGPL3/LSDCAS/cfg_reset_source2	AD34	I/O	OV_{DD}	—
LGPL4/LGTA/LUPWAIT/LPBSE	AE35	I/O	OV_{DD}	—
LGPL5/cfg_clkin_div	AF36	I/O	OV_{DD}	—
LCKE	G36	O	OV_{DD}	—
LCLK[0]	J33	O	OV_{DD}	—
LCLK[1]/LCS[6]	J34	O	OV_{DD}	—

Table 66. MPC8360E TBGA Pinout Listing (continued)

Signal	Package Pin Number	Pin Type	Power Supply	Notes
PMC				
$\overline{\text{QUIESCE}}$	B36	O	OV_{DD}	—
System Control				
$\overline{\text{PORESET}}$	L37	I	OV_{DD}	—
$\overline{\text{HRESET}}$	L36	I/O	OV_{DD}	1
$\overline{\text{SRESET}}$	M33	I/O	OV_{DD}	2
Thermal Management				
THERM0	AP19	I	GV_{DD}	—
THERM1	AT31	I	GV_{DD}	—
Power and Ground Signals				
$AV_{DD}1$	K35	Power for LBIU DLL (1.2 V)	$AV_{DD}1$	—
$AV_{DD}2$	K36	Power for CE PLL (1.2 V)	$AV_{DD}2$	—
$AV_{DD}5$	AM29	Power for e300 PLL (1.2 V)	$AV_{DD}5$	—
$AV_{DD}6$	K37	Power for system PLL (1.2 V)	$AV_{DD}6$	—
GND	A2, A8, A13, A19, A22, A25, A31, A33, A36, B7, B12, B24, B27, B30, C4, C6, C9, C15, C26, C32, D3, D8, D11, D14, D17, D19, D23, D27, E7, E13, E25, E30, E36, F4, F37, G34, H1, H5, H32, H33, J4, J32, J37, K1, L3, L5, L33, L34, M1, M34, M35, N37, P2, P5, P35, P36, R4, T3, U1, U5, U35, V37, W1, W4, W33, W36, Y34, AA3, AA5, AC3, AC32, AC35, AD1, AD37, AE4, AE34, AE36, AF33, AG4, AG6, AG32, AH35, AJ1, AJ4, AJ32, AJ35, AJ37, AK36, AL3, AL34, AM4, AN6, AN23, AN30, AP8, AP12, AP14, AP16, AP17, AP20, AP25, AR6, AR8, AR9, AR19, AR24, AR31, AR35, AR37, AT4, AT10, AT19, AT20, AT25, AU14, AU22, AU28, AU35	—	—	—
GV_{DD}	AD4, AE3, AF1, AF5, AF35, AF37, AG2, AG36, AH33, AH34, AK5, AM1, AM35, AM37, AN2, AN10, AN11, AN12, AN14, AN32, AN36, AP5, AP23, AP28, AR1, AR7, AR10, AR12, AR21, AR25, AR27, AR33, AT15, AT22, AT28, AT33, AU2, AU5, AU16, AU31, AU36	Power for DDR DRAM I/O voltage (2.5 or 1.8 V)	GV_{DD}	—

Table 67. MPC8358E TBGA Pinout Listing (continued)

Signal	Package Pin Number	Pin Type	Power Supply	Notes
CE_PB[0:27]	AE2, AE1, AD5, AD3, AD2, AC6, AC5, AC4, AC2, AC1, AB5, AB4, AB3, AB1, AA6, AA4, AA2, Y6, Y4, Y3, Y2, Y1, W6, W5, W2, V5, V3, V2	I/O	OV_{DD}	—
CE_PC[0:1]	V1, U6	I/O	OV_{DD}	
CE_PC[2:3]	C16, A15	I/O	$LV_{DD}1$	—
CE_PC[4:6]	U4, U3, T6	I/O	OV_{DD}	—
CE_PC[7]	C19	I/O	$LV_{DD}2$	—
CE_PC[8:9]	A4, C5	I/O	$LV_{DD}0$	—
CE_PC[10:30]	T5, T4, T2, T1, R5, R3, R1, C11, D12, F13, B10, C10, E12, A9, B8, D10, A14, E15, B14, D15, AH2	I/O	OV_{DD}	—
CE_PD[0:27]	E11, D9, C8, F11, A7, E9, C7, A6, F10, B6, D7, E8, B5, A5, C2, E4, F5, B1, D2, G5, D1, E2, H6, F3, E1, F2, G3, H4	I/O	OV_{DD}	—
CE_PE[0:31]	K3, J2, F1, G2, J5, H3, G1, H2, K6, J3, K5, K4, L6, P6, P4, P3, P1, N4, N5, N2, N1, M2, M3, M5, M6, L1, L2, L4, E14, C13, C14, B13	I/O	OV_{DD}	—
CE_PF[0:3]	F14, D13, A12, A11	I/O	OV_{DD}	—
Clocks				
PCI_CLK_OUT[0]/CE_PF[26]	B22	I/O	$LV_{DD}2$	—
PCI_CLK_OUT[1:2]/CE_PF[27:28]	D22, A23	I/O	OV_{DD}	—
CLKIN	E37	I	OV_{DD}	—
PCI_CLOCK/PCI_SYNC_IN	M36	I	OV_{DD}	—
PCI_SYNC_OUT/CE_PF[29]	D37	I/O	OV_{DD}	3
JTAG				
TCK	K33	I	OV_{DD}	—
TDI	K34	I	OV_{DD}	4
TDO	H37	O	OV_{DD}	3
TMS	J36	I	OV_{DD}	4
$\overline{\text{TRST}}$	L32	I	OV_{DD}	4
Test				
TEST	L35	I	OV_{DD}	7
$\overline{\text{TEST\_SEL}}$	AU34	I	GV_{DD}	10
PMC				
$\overline{\text{QUIESCE}}$	B36	O	OV_{DD}	—
System Control				

21.2 Core PLL Configuration

RCWL[COREPLL] selects the ratio between the internal coherent system bus clock (*csb_clk*) and the e300 core clock (*core_clk*). This table shows the encodings for RCWL[COREPLL]. COREPLL values not listed in this table should be considered reserved.

Table 73. e300 Core PLL Configuration

RCWL[COREPLL]			*core_clk*:*csb_clk* Ratio	VCO divider
0–1	2–5	6		
nn	0000	n	PLL bypassed (PLL off, *csb_clk* clocks core directly)	PLL bypassed (PLL off, *csb_clk* clocks core directly)
00	0001	0	1:1	÷2
01	0001	0	1:1	÷4
10	0001	0	1:1	÷8
11	0001	0	1:1	÷8
00	0001	1	1.5:1	÷2
01	0001	1	1.5:1	÷4
10	0001	1	1.5:1	÷8
11	0001	1	1.5:1	÷8
00	0010	0	2:1	÷2
01	0010	0	2:1	÷4
10	0010	0	2:1	÷8
11	0010	0	2:1	÷8
00	0010	1	2.5:1	÷2
01	0010	1	2.5:1	÷4
10	0010	1	2.5:1	÷8
11	0010	1	2.5:1	÷8
00	0011	0	3:1	÷2
01	0011	0	3:1	÷4
10	0011	0	3:1	÷8
11	0011	0	3:1	÷8

NOTE

Core VCO frequency = Core frequency × VCO divider. The VCO divider (RCWL[COREPLL[0:1]]) must be set properly so that the core VCO frequency is in the range of 800–1800 MHz. Having a core frequency below the CSB frequency is not a possible option because the core frequency must be equal to or greater than the CSB frequency.

22.3.1 Experimental Determination of the Junction Temperature with a Heat Sink

When heat sink is used, the junction temperature is determined from a thermocouple inserted at the interface between the case of the package and the interface material. A clearance slot or hole is normally required in the heat sink. Minimizing the size of the clearance is important to minimize the change in thermal performance caused by removing part of the thermal interface to the heat sink. Because of the experimental difficulties with this technique, many engineers measure the heat sink temperature and then back calculate the case temperature using a separate measurement of the thermal resistance of the interface. From this case temperature, the junction temperature is determined from the junction-to-case thermal resistance.

$$T_J = T_C + (R_{\theta JC} \times P_D)$$

where:

T_J = junction temperature (° C)

T_C = case temperature of the package (° C)

$R_{\theta JC}$ = junction to case thermal resistance (° C/W)

P_D = power dissipation (W)

23 System Design Information

This section provides electrical and thermal design recommendations for successful application of the MPC8360E/58E. Additional information can be found in *MPC8360E/MPC8358E PowerQUICC Design Checklist* (AN3097).

23.1 System Clocking

The device includes two PLLs, as follows.

- The platform PLL ($AV_{DD}1$) generates the platform clock from the externally supplied CLKIN input. The frequency ratio between the platform and CLKIN is selected using the platform PLL ratio configuration bits as described in Section 21.1, “System PLL Configuration.”
- The e300 core PLL ($AV_{DD}2$) generates the core clock as a slave to the platform clock. The frequency ratio between the e300 core clock and the platform clock is selected using the e300 PLL ratio configuration bits as described in Section 21.2, “Core PLL Configuration.”

23.2 PLL Power Supply Filtering

Each of the PLLs listed above is provided with power through independent power supply pins ($AV_{DD}1$, $AV_{DD}2$, respectively). The AV_{DD} level should always be equivalent to V_{DD}, and preferably these voltages are derived directly from V_{DD} through a low frequency filter scheme such as the following.

There are a number of ways to reliably provide power to the PLLs, but the recommended solution is to provide five independent filter circuits as illustrated in Figure 56, one to each of the five AV_{DD} pins. By providing independent filters to each PLL, the opportunity to cause noise injection from one PLL to the other is reduced.

This circuit is intended to filter noise in the PLLs resonant frequency range from a 500 kHz to 10 MHz range. It should be built with surface mount capacitors with minimum Effective Series Inductance (ESL). Consistent with the recommendations of Dr. Howard Johnson in *High Speed Digital Design: A Handbook of Black Magic* (Prentice Hall, 1993), multiple small capacitors of equal value are recommended over a single large value capacitor.

Each circuit should be placed as close as possible to the specific AV_{DD} pin being supplied to minimize noise coupled from nearby circuits. It should be possible to route directly from the capacitors to the AV_{DD} pin, which is on the periphery of package, without the inductance of vias.