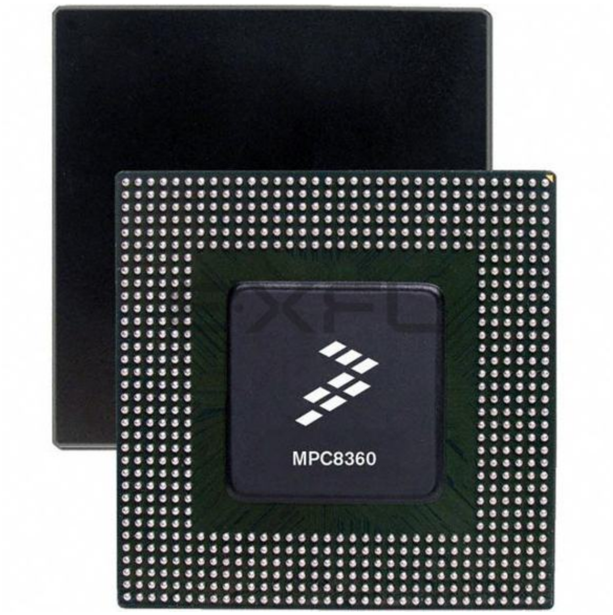

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Understanding Embedded - Microprocessors

Embedded microprocessors are specialized computing chips designed to perform specific tasks within an embedded system. Unlike general-purpose microprocessors found in personal computers, embedded microprocessors are tailored for dedicated functions within larger systems, offering optimized performance, efficiency, and reliability. These microprocessors are integral to the operation of countless electronic devices, providing the computational power necessary for controlling processes, handling data, and managing communications.

Applications of Embedded - Microprocessors

Embedded microprocessors are utilized across a broad spectrum of applications, making them indispensable in

Details	
Product Status	Obsolete
Core Processor	PowerPC e300
Number of Cores/Bus Width	1 Core, 32-Bit
Speed	500MHz
Co-Processors/DSP	Communications; QUICC Engine
RAM Controllers	DDR, DDR2
Graphics Acceleration	No
Display & Interface Controllers	-
Ethernet	10/100/1000Mbps (1)
SATA	-
USB	USB 1.x (1)
Voltage - I/O	1.8V, 2.5V, 3.3V
Operating Temperature	0°C ~ 105°C (TA)
Security Features	-
Package / Case	740-LBGA
Supplier Device Package	740-TBGA (37.5x37.5)
Purchase URL	https://www.e-xfl.com/product-detail/nxp-semiconductors/mpc8360vvahfh

wide range of protocols including ATM, Ethernet, HDLC, and POS. The QUICC Engine module’s enhanced interworking eases the transition and reduces investment costs from ATM to IP based systems. The other major features include a dual DDR SDRAM memory controller for the MPC8360E, which allows equipment providers to partition system parameters and data in an extremely efficient way, such as using one 32-bit DDR memory controller for control plane processing and the other for data plane processing. The MPC8358E has a single DDR SDRAM memory controller. The MPC8360E/58E also offers a 32-bit PCI controller, a flexible local bus, and a dedicated security engine.

This figure shows the MPC8360Eblock diagram.

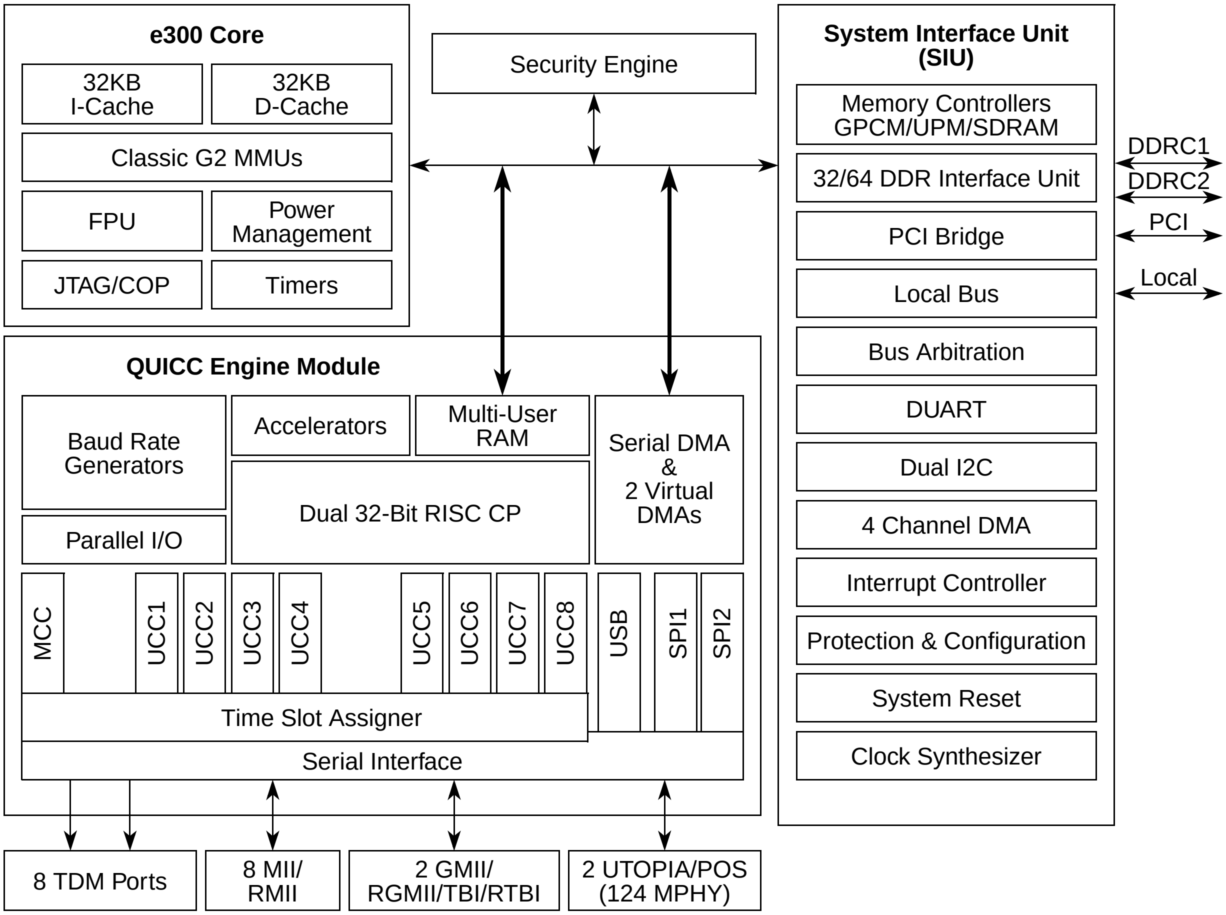


Figure 1. MPC8360E Block Diagram

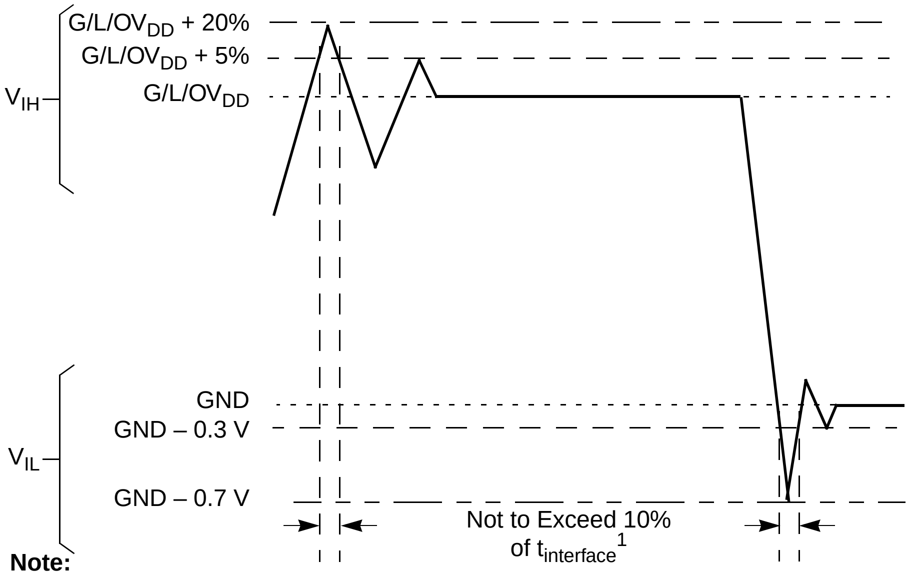
Table 2. Recommended Operating Conditions (continued)

Characteristic	Symbol	Recommended Value	Unit	Notes
PCI, local bus, DUART, system control and power management, I^2C, SPI, and JTAG I/O voltage	OV_{DD}	3.3 V ± 330 mV	V	—
Junction temperature	T_J	0 to 105 –40 to 105	°C	2

Notes:

1. GV_{DD}, LV_{DD}, OV_{DD}, AV_{DD}, and V_{DD} must track each other and must vary in the same direction—either in the positive or negative direction.
2. The operating conditions for junction temperature, T_J, on the 600/333/400 MHz and 500/333/500 MHz on rev. 2.0 silicon is 0° to 70 °C. Refer to Errata General9 in *Chip Errata for the MPC8360E, Rev. 1*.
3. For more information on Part Numbering, refer to Table 80.

This figure shows the undershoot and overshoot voltages at the interfaces of the device.



Note:
1. Note that $t_{interface}$ refers to the clock period associated with the bus clock interface.

Figure 3. Overshoot/Undershoot Voltage for $GV_{DD}/OV_{DD}/LV_{DD}$

Table 9. GTX_CLK125 AC Timing Specifications

At recommended operating conditions with LV_{DD} = 2.5 ± 0.125 mV/ 3.3 V ± 165 mV (continued)

Parameter/Condition	Symbol	Min	Typical	Max	Unit	Notes
GTX_CLK rise and fall time LV_{DD} = 2.5 V LV_{DD} = 3.3 V	t_{G125R}/t_{G125F}	—	—	 0.75 1.0	ns	1
GTX_CLK125 duty cycle GMII & TBI 1000Base-T for RGMII & RTBI	t_{G125H}/t_{G125}	 45 47	—	 55 53	%	2
GTX_CLK125 jitter	—	—	—	±150	ps	2

Notes:

1. Rise and fall times for GTX_CLK125 are measured from 0.5 and 2.0 V for LV_{DD} = 2.5 V and from 0.6 and 2.7 V for LV_{DD} = 3.3 V.
2. GTX_CLK125 is used to generate the GTX clock for the UCC Ethernet transmitter with 2% degradation. The GTX_CLK125 duty cycle can be loosened from 47%/53% as long as the PHY device can tolerate the duty cycle generated by GTX_CLK. See Section 8.2.2, “MII AC Timing Specifications,” Section 8.2.3, “RMII AC Timing Specifications,” and Section 8.2.5, “RGMII and RTBI AC Timing Specifications” for the duty cycle for 10Base-T and 100Base-T reference clock.

5 RESET Initialization

This section describes the DC and AC electrical specifications for the reset initialization timing and electrical requirements of the MPC8360E/58E.

5.1 RESET DC Electrical Characteristics

This table provides the DC electrical characteristics for the RESET pins of the device.

Table 10. RESET Pins DC Electrical Characteristics [1]

Characteristic	Symbol	Condition	Min	Max	Unit
Input high voltage	V_{IH}	—	2.0	OV_{DD} + 0.3	V
Input low voltage	V_{IL}	—	–0.3	0.8	V
Input current	I_{IN}	—	—	±10	μA
Output high voltage	V_{OH}[2]	I_{OH} = –8.0 mA	2.4	—	V
Output low voltage	V_{OL}	I_{OL} = 8.0 mA	—	0.5	V
Output low voltage	V_{OL}	I_{OL} = 3.2 mA	—	0.4	V

Notes:

1. This table applies for pins $\overline{\text{PORESET}}$, $\overline{\text{HRESET}}$, $\overline{\text{SRESET}}$, and $\overline{\text{QUIESCE}}$.
2. $\overline{\text{HRESET}}$ and $\overline{\text{SRESET}}$ are open drain pins, thus V_{OH} is not relevant for those pins.

5.2 RESET AC Electrical Characteristics

This section describes the AC electrical specifications for the reset initialization timing requirements of the device. This table provides the reset initialization AC timing specifications for the DDR SDRAM component(s).

Table 11. RESET Initialization Timing Specifications

Parameter/Condition	Min	Max	Unit	Notes
Required assertion time of $\overline{\text{HRESET}}$ or $\overline{\text{SRESET}}$ (input) to activate reset flow	32	—	$t_{\text{PCI\_SYNC\_IN}}$	1
Required assertion time of $\overline{\text{PORESET}}$ with stable clock applied to CLKIN when the device is in PCI host mode	32	—	t_{CLKIN}	2
Required assertion time of $\overline{\text{PORESET}}$ with stable clock applied to PCI_SYNC_IN when the device is in PCI agent mode	32	—	$t_{\text{PCI\_SYNC\_IN}}$	1
$\overline{\text{HRESET}}$/$\overline{\text{SRESET}}$ assertion (output)	512	—	$t_{\text{PCI\_SYNC\_IN}}$	1
$\overline{\text{HRESET}}$ negation to $\overline{\text{SRESET}}$ negation (output)	16	—	$t_{\text{PCI\_SYNC\_IN}}$	1
Input setup time for POR config signals (CFG_RESET_SOURCE[0:2] and CFG_CLKIN_DIV) with respect to negation of $\overline{\text{PORESET}}$ when the device is in PCI host mode	4	—	t_{CLKIN}	2
Input setup time for POR config signals (CFG_RESET_SOURCE[0:2] and CFG_CLKIN_DIV) with respect to negation of $\overline{\text{PORESET}}$ when the device is in PCI agent mode	4	—	$t_{\text{PCI\_SYNC\_IN}}$	1
Input hold time for POR config signals with respect to negation of $\overline{\text{HRESET}}$	0	—	ns	—
Time for the device to turn off POR config signals with respect to the assertion of $\overline{\text{HRESET}}$	—	4	ns	3
Time for the device to turn on POR config signals with respect to the negation of $\overline{\text{HRESET}}$	1	—	$t_{\text{PCI\_SYNC\_IN}}$	1, 3

Notes:

1. $t_{\text{PCI\_SYNC\_IN}}$ is the clock period of the input clock applied to PCI_SYNC_IN. When the device is In PCI host mode the primary clock is applied to the CLKIN input, and PCI_SYNC_IN period depends on the value of CFG_CLKIN_DIV. Refer *MPC8360E PowerQUICC II Pro Integrated Communications Processor Reference Manual* for more details.
2. t_{CLKIN} is the clock period of the input clock applied to CLKIN. It is only valid when the device is in PCI host mode. Refer *MPC8360E PowerQUICC II Pro Integrated Communications Processor Reference Manual* for more details.
3. POR config signals consists of CFG_RESET_SOURCE[0:2] and CFG_CLKIN_DIV.

This table provides the PLL and DLL lock times.

Table 12. PLL and DLL Lock Times

Parameter/Condition	Min	Max	Unit	Notes
PLL lock times	—	100	µs	—
DLL lock times	7680	122,880	csb_clk cycles	1, 2

Notes:

1. DLL lock times are a function of the ratio between the output clock and the coherency system bus clock (csb_clk). A 2:1 ratio results in the minimum and an 8:1 ratio results in the maximum.
2. The csb_clk is determined by the CLKIN and system PLL ratio. See Section 21, “Clocking,” for more information.

8.2.2.2 MII Receive AC Timing Specifications

This table provides the MII receive AC timing specifications.

Table 30. MII Receive AC Timing Specifications

At recommended operating conditions with LV_{DD}/OV_{DD} of 3.3 V ± 10%.

Parameter/Condition	Symbol[1]	Min	Typ	Max	Unit
RX_CLK clock period 10 Mbps	t_{MRX}	—	400	—	ns
RX_CLK clock period 100 Mbps	t_{MRX}	—	40	—	ns
RX_CLK duty cycle	t_{MRXH}/t_{MRX}	35	—	65	%
RXD[3:0], RX_DV, RX_ER setup time to RX_CLK	t_{MRDVKH}	10.0	—	—	ns
RXD[3:0], RX_DV, RX_ER hold time to RX_CLK	t_{MRDXKH}	10.0	—	—	ns
RX_CLK clock rise time, (20% to 80%)	t_{MRXR}	1.0	—	4.0	ns
RX_CLK clock fall time, (80% to 20%)	t_{MRXF}	1.0	—	4.0	ns

Note:

1. The symbols used for timing specifications follow the pattern of $t_{(\text{first two letters of functional block})(\text{signal})(\text{state})(\text{reference})(\text{state})}$ for inputs and $t_{(\text{first two letters of functional block})(\text{reference})(\text{state})(\text{signal})(\text{state})}$ for outputs. For example, t_{MRDVKH} symbolizes MII receive timing (MR) with respect to the time data input signals (D) reach the valid state (V) relative to the t_{MRX} clock reference (K) going to the high (H) state or setup time. Also, t_{MRDXKL} symbolizes MII receive timing (GR) with respect to the time data input signals (D) went invalid (X) relative to the t_{MRX} clock reference (K) going to the low (L) state or hold time. Note that, in general, the clock reference symbol representation is based on three letters representing the clock of a particular functional. For example, the subscript of t_{MRX} represents the MII (M) receive (RX) clock. For rise and fall times, the latter convention is used with the appropriate letter: R (rise) or F (fall).

This figure provides the AC test load.

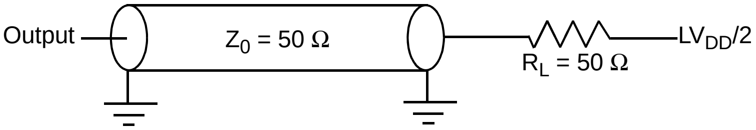


Figure 13. AC Test Load

This figure shows the MII receive AC timing diagram.

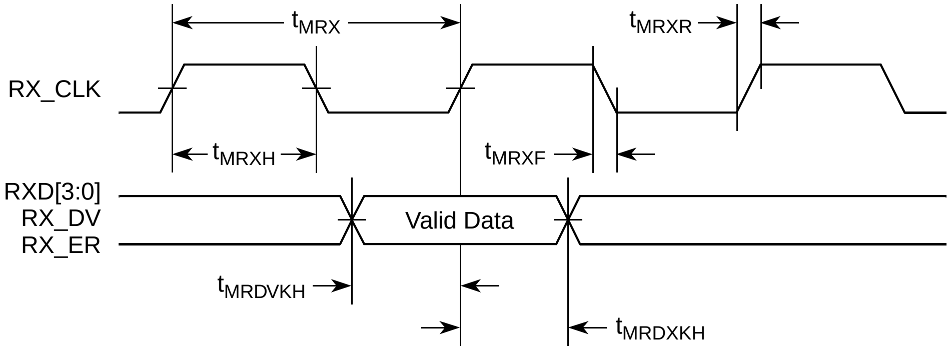


Figure 14. MII Receive AC Timing Diagram

This figure shows the RGMII and RTBI AC timing and multiplexing diagrams.

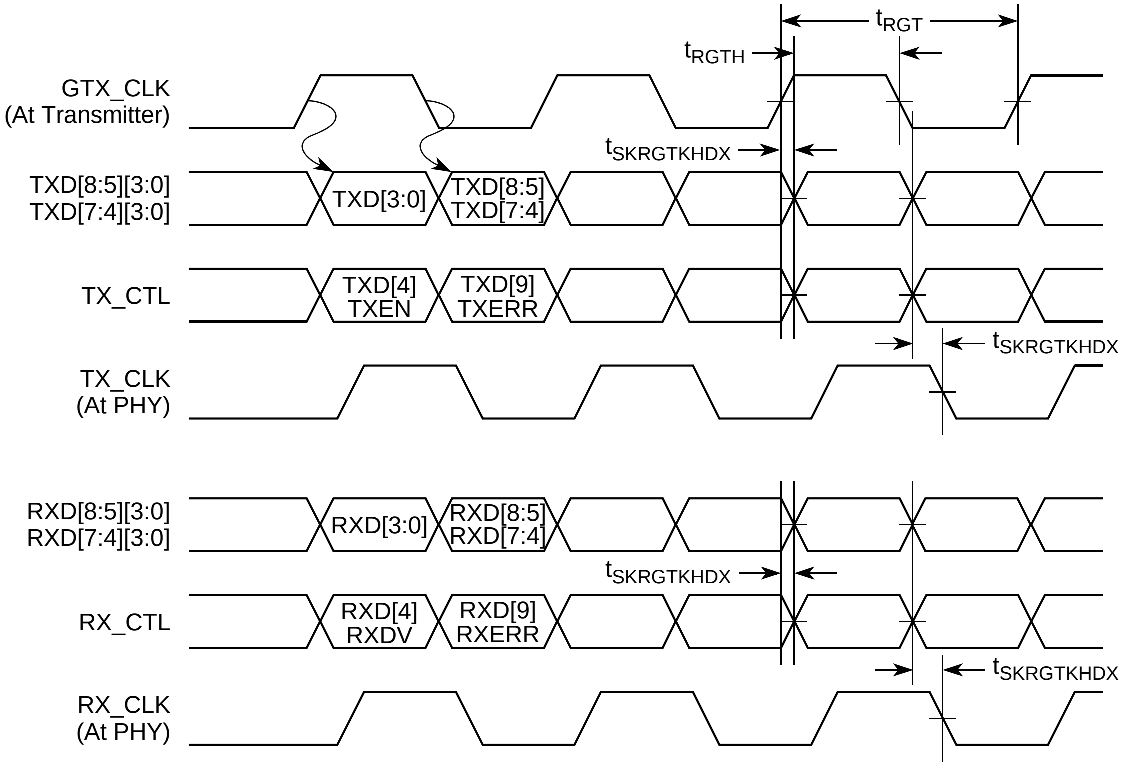


Figure 20. RGMII and RTBI AC Timing and Multiplexing Diagrams

8.3 Ethernet Management Interface Electrical Characteristics

The electrical characteristics specified here apply to MII management interface signals MDIO (management data input/output) and MDC (management data clock). The electrical characteristics for GMII, RGMII, TBI, and RTBI are specified in Section 8.1, “Three-Speed Ethernet Controller (10/100/1000 Mbps)— GMII/MII/RMII/TBI/RGMII/RTBI Electrical Characteristics.”

8.3.1 MII Management DC Electrical Characteristics

The MDC and MDIO are defined to operate at a supply voltage of 3.3 V. The DC electrical characteristics for MDIO and MDC are provided in this table.

Table 36. MII Management DC Electrical Characteristics When Powered at 3.3 V

Parameter	Symbol	Conditions		Min	Max	Unit
Supply voltage (3.3 V)	OV_{DD}	—		2.97	3.63	V
Output high voltage	V_{OH}	I_{OH} = –1.0 mA	OV_{DD} = Min	2.10	OV_{DD} + 0.3	V
Output low voltage	V_{OL}	I_{OL} = 1.0 mA	OV_{DD} = Min	GND	0.50	V
Input high voltage	V_{IH}	—		2.00	—	V
Input low voltage	V_{IL}	—		—	0.80	V
Input current	I_{IN}	$0\ V \leq V_{IN} \leq OV_{DD}$		—	±10	μA

Table 41. Local Bus General Timing Parameters—DLL Bypass Mode[9] (continued)

Parameter	Symbol[1]	Min	Max	Unit	Notes
Local bus clock to output valid	t_{LBKHOV}	—	3	ns	3
Local bus clock to output high impedance for LAD/LDP	t_{LBKHOZ}	—	4	ns	8

Notes:

1. The symbols used for timing specifications follow the pattern of $t_{\text{(first two letters of functional block)(signal)(state)(reference)(state)}}$ for inputs and $t_{\text{(first two letters of functional block)(reference)(state)(signal)(state)}}$ for outputs. For example, $t_{LBIXKH1}$ symbolizes local bus timing (LB) for the input (I) to go invalid (X) with respect to the time the t_{LBK} clock reference (K) goes high (H), in this case for clock one (1). Also, t_{LBKHOX} symbolizes local bus timing (LB) for the t_{LBK} clock reference (K) to go high (H), with respect to the output (O) going invalid (X) or output hold time.
2. All timings are in reference to falling edge of LCLK0 (for all outputs and for $\overline{\text{LGTA}}$ and LUPWAIT inputs) or rising edge of LCLK0 (for all other inputs).
3. All signals are measured from $OV_{DD}/2$ of the rising/falling edge of LCLK0 to $0.4 \times OV_{DD}$ of the signal in question for 3.3-V signaling levels.
4. Input timings are measured at the pin.
5. $t_{LBOTOT1}$ should be used when RCWH[LALE] is not set and when the load on LALE output pin is at least 10 pF less than the load on LAD output pins.
6. $t_{LBOTOT2}$ should be used when RCWH[LALE] is set and when the load on LALE output pin is at least 10 pF less than the load on LAD output pins.
7. $t_{LBOTOT3}$ should be used when RCWH[LALE] is set and when the load on LALE output pin equals to the load on LAD output pins.
8. For purposes of active/float timing measurements, the Hi-Z or off-state is defined to be when the total current delivered through the component pin is less than or equal to the leakage current specification.
9. DLL bypass mode is not recommended for use at frequencies above 66 MHz.

This figure provides the AC test load for the local bus.

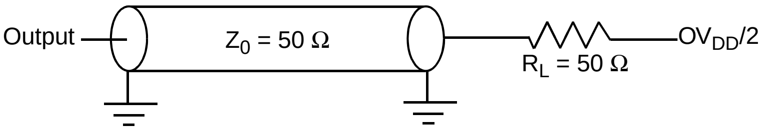


Figure 22. Local Bus C Test Load

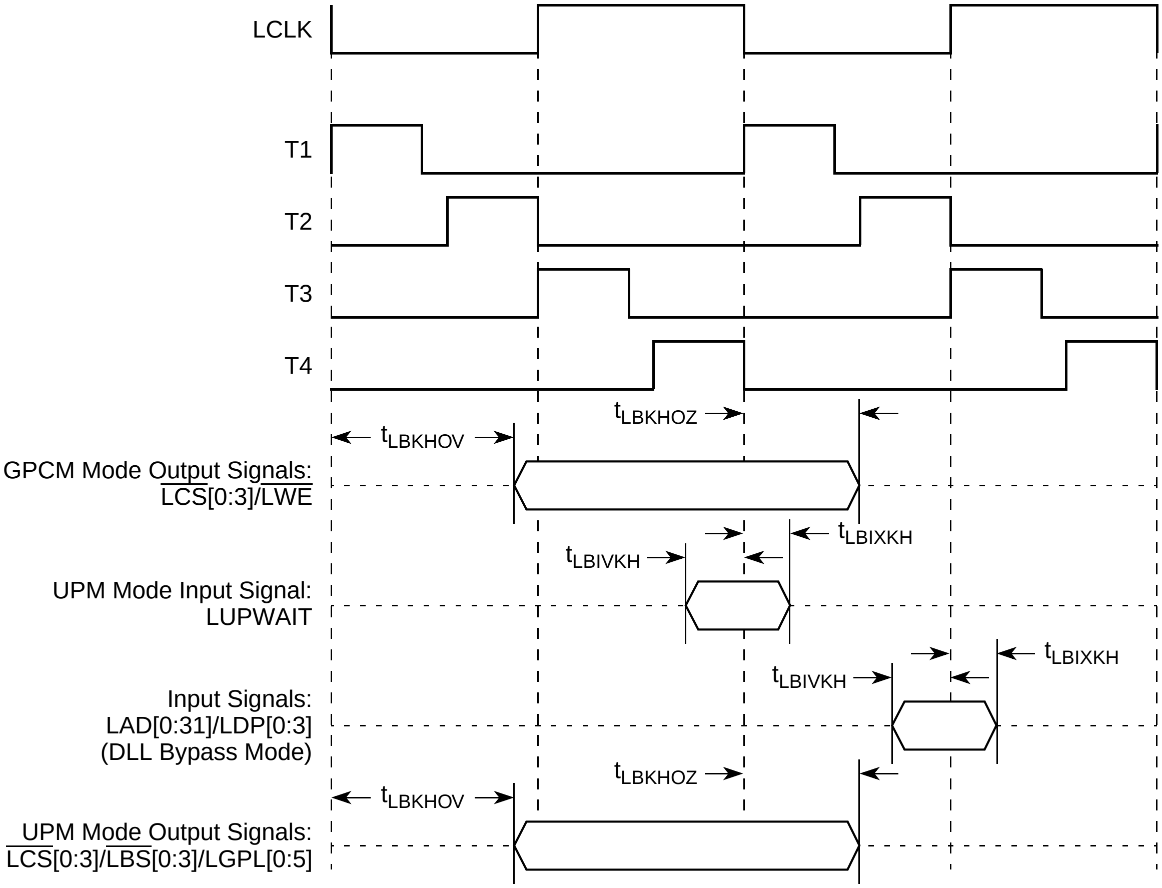


Figure 27. Local Bus Signals, GPCM/UPM Signals for LCRR[CLKDIV] = 4 (DLL Bypass Mode)

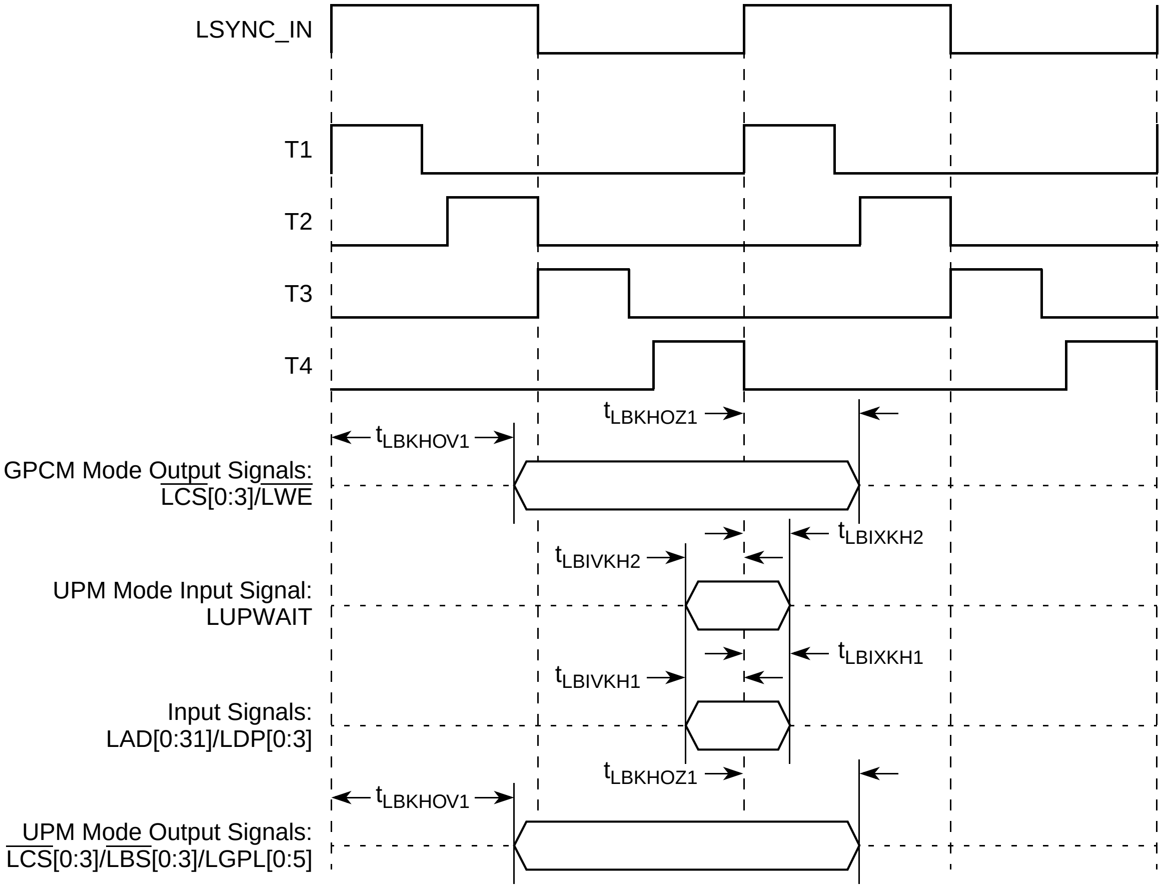


Figure 28. Local Bus Signals, GPCM/UPM Signals for LCRR[CLKDIV] = 4 (DLL Enabled)

10 JTAG

This section describes the DC and AC electrical specifications for the IEEE 1149.1 (JTAG) interface of the MPC8360E/58E.

10.1 JTAG DC Electrical Characteristics

This table provides the DC electrical characteristics for the IEEE 1149.1 (JTAG) interface of the device.

Table 42. JTAG interface DC Electrical Characteristics

Characteristic	Symbol	Condition	Min	Max	Unit
Output high voltage	V_{OH}	I_{OH} = –6.0 mA	2.4	—	V
Output low voltage	V_{OL}	I_{OL} = 6.0 mA	—	0.5	V
Output low voltage	V_{OL}	I_{OL} = 3.2 mA	—	0.4	V
Input high voltage	V_{IH}	—	2.5	OV_{DD} + 0.3	V
Input low voltage	V_{IL}	—	–0.3	0.8	V
Input current	I_{IN}	0 V $\leq V_{IN} \leq OV_{DD}$	—	±10	μA

10.2 JTAG AC Electrical Characteristics

This section describes the AC electrical specifications for the IEEE 1149.1 (JTAG) interface of the device.

This table provides the JTAG AC timing specifications as defined in Figure 30 through Figure 33.

Table 43. JTAG AC Timing Specifications (Independent of CLKIN)[1]

At recommended operating conditions (see Table 2).

Parameter	Symbol [2]	Min	Max	Unit	Notes
JTAG external clock frequency of operation	f_{JTG}	0	33.3	MHz	—
JTAG external clock cycle time	t_{JTG}	30	—	ns	—
JTAG external clock duty cycle	t_{JTKHKL}/t_{JTG}	45	55	%	—
JTAG external clock rise and fall times	t_{JTGR} & t_{JTGF}	0	2	ns	—
$\overline{\text{TRST}}$ assert time	t_{TRST}	25	—	ns	3
Input setup times: Boundary-scan data TMS, TDI	 t_{JTDVKH} t_{JTIVKH}	 4 4	 — —	ns	4
Input hold times: Boundary-scan data TMS, TDI	 t_{JTDXKH} t_{JTIXKH}	 10 10	 — —	ns	4
Valid times: Boundary-scan data TDO	 t_{JTKLDV} t_{JTKLOV}	 2 2	 11 11	ns	5
Output hold times: Boundary-scan data TDO	 t_{JTKLDX} t_{JTKLOX}	 2 2	 — —	ns	5
JTAG external clock to output high impedance: Boundary-scan data TDO	 t_{JTKLDZ} t_{JTKLOZ}	 2 2	 19 9	ns	5, 6

Notes:

1. All outputs are measured from the midpoint voltage of the falling/rising edge of t_{TCLK} to the midpoint of the signal in question. The output timings are measured at the pins. All output timings assume a purely resistive 50-Ω load (see Figure 22). Time-of-flight delays must be added for trace lengths, vias, and connectors in the system.
2. The symbols used for timing specifications herein follow the pattern of $t_{\text{(first two letters of functional block)(signal)(state)(reference)(state)}}$ for inputs and $t_{\text{(first two letters of functional block)(reference)(state)(signal)(state)}}$ for outputs. For example, t_{JTDVKH} symbolizes JTAG device timing (JT) with respect to the time data input signals (D) reaching the valid state (V) relative to the t_{JTG} clock reference (K) going to the high (H) state or setup time. Also, t_{JTDXKH} symbolizes JTAG timing (JT) with respect to the time data input signals (D) went invalid (X) relative to the t_{JTG} clock reference (K) going to the high (H) state. Note that, in general, the clock reference symbol representation is based on three letters representing the clock of a particular functional. For rise and fall times, the latter convention is used with the appropriate letter: R (rise) or F (fall).
3. $\overline{\text{TRST}}$ is an asynchronous level sensitive signal. The setup time is for test purposes only.
4. Non-JTAG signal input timing with respect to t_{TCLK}.
5. Non-JTAG signal output timing with respect to t_{TCLK}.
6. Guaranteed by design and characterization.

This figure shows the PCI input AC timing conditions.

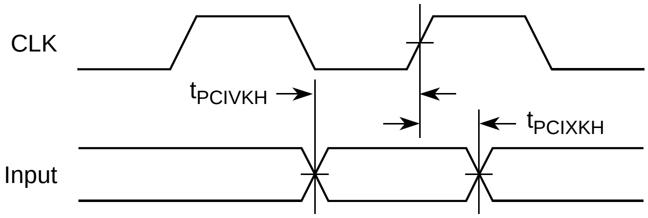


Figure 37. PCI Input AC Timing Measurement Conditions

This figure shows the PCI output AC timing conditions.

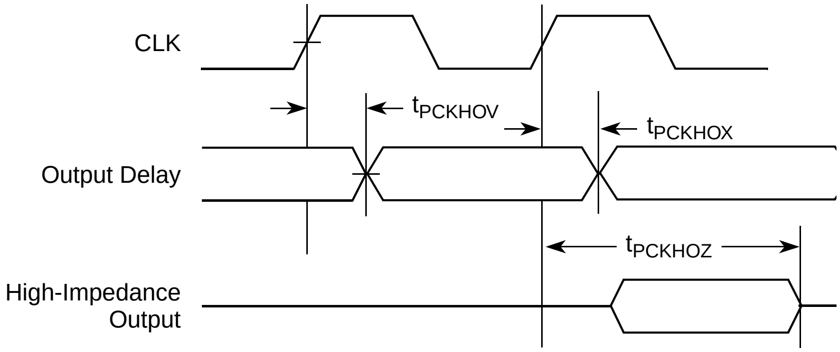


Figure 38. PCI Output AC Timing Measurement Condition

13 Timers

This section describes the DC and AC electrical specifications for the timers of the MPC8360E/58E.

13.1 Timers DC Electrical Characteristics

This table provides the DC electrical characteristics for the device timer pins, including TIN, $\overline{\text{TOUT}}$, $\overline{\text{TGATE}}$, and RTC_CLK.

Table 49. Timers DC Electrical Characteristics

Characteristic	Symbol	Condition	Min	Max	Unit
Output high voltage	V_{OH}	I_{OH} = –6.0 mA	2.4	—	V
Output low voltage	V_{OL}	I_{OL} = 6.0 mA	—	0.5	V
Output low voltage	V_{OL}	I_{OL} = 3.2 mA	—	0.4	V
Input high voltage	V_{IH}	—	2.0	OV_{DD} + 0.3	V
Input low voltage	V_{IL}	—	–0.3	0.8	V
Input current	I_{IN}	$0 V \le V_{IN} \le OV_{DD}$	—	±10	μA

13.2 Timers AC Timing Specifications

This table provides the timer input and output AC timing specifications.

Table 50. Timers Input AC Timing Specifications[1]

Characteristic	Symbol[2]	Typ	Unit
Timers inputs—minimum pulse width	t_{TIWID}	20	ns

Notes:

1. Input specifications are measured from the 50% level of the signal to the 50% level of the rising edge of CLKIN. Timings are measured at the pin.
2. Timers inputs and outputs are asynchronous to any visible clock. Timers outputs should be synchronized before use by any external synchronous logic. Timers inputs are required to be valid for at least t_{TIWID} ns to ensure proper operation.

This figure provides the AC test load for the timers.

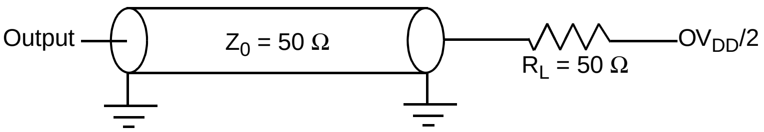


Figure 39. Timers AC Test Load

14 GPIO

This section describes the DC and AC electrical specifications for the GPIO of the MPC8360E/58E.

14.1 GPIO DC Electrical Characteristics

This table provides the DC electrical characteristics for the device GPIO.

Table 51. GPIO DC Electrical Characteristics

Characteristic	Symbol	Condition	Min	Max	Unit	Notes
Output high voltage	V_{OH}	$I_{OH} = -6.0$ mA	2.4	—	V	1
Output low voltage	V_{OL}	$I_{OL} = 6.0$ mA	—	0.5	V	1
Output low voltage	V_{OL}	$I_{OL} = 3.2$ mA	—	0.4	V	1
Input high voltage	V_{IH}	—	2.0	$OV_{DD} + 0.3$	V	1
Input low voltage	V_{IL}	—	–0.3	0.8	V	—
Input current	I_{IN}	$0\ V \le V_{IN} \le OV_{DD}$	—	±10	μA	—

Note:

1. This specification applies when operating from 3.3-V supply.

14.2 GPIO AC Timing Specifications

This table provides the GPIO input and output AC timing specifications.

Table 52. GPIO Input AC Timing Specifications[1]

Characteristic	Symbol[2]	Typ	Unit
GPIO inputs—minimum pulse width	t_{PIWID}	20	ns

Notes:

1. Input specifications are measured from the 50% level of the signal to the 50% level of the rising edge of CLKIN. Timings are measured at the pin.
2. GPIO inputs and outputs are asynchronous to any visible clock. GPIO outputs should be synchronized before use by any external synchronous logic. GPIO inputs are required to be valid for at least t_{PIWID} ns to ensure proper operation.

This figure provides the AC test load for the GPIO.

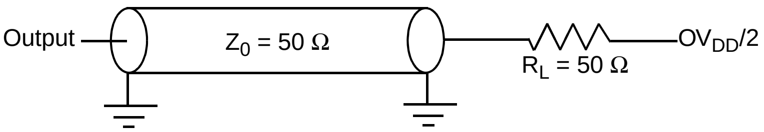


Figure 40. GPIO AC Test Load

15 IPIC

This section describes the DC and AC electrical specifications for the external interrupt pins of the MPC8360E/58E.

15.1 IPIC DC Electrical Characteristics

This table provides the DC electrical characteristics for the external interrupt pins of the IPIC.

Table 53. IPIC DC Electrical Characteristics

Characteristic	Symbol	Condition	Min	Max	Unit
Input high voltage	V_{IH}	—	2.0	OV_{DD} + 0.3	V
Input low voltage	V_{IL}	—	–0.3	0.8	V
Input current	I_{IN}	—	—	±10	µA
Output low voltage	V_{OL}	I_{OL} = 6.0 mA	—	0.5	V
Output low voltage	V_{OL}	I_{OL} = 3.2 mA	—	0.4	V

Notes:

1. This table applies for pins $\overline{\text{IRQ}}$[0:7], $\overline{\text{IRQ\_OUT}}$, $\overline{\text{MCP\_OUT}}$, and CE ports Interrupts.
2. $\overline{\text{IRQ\_OUT}}$ and $\overline{\text{MCP\_OUT}}$ are open drain pins, thus V_{OH} is not relevant for those pins.

Table 66. MPC8360E TBGA Pinout Listing (continued)

Signal	Package Pin Number	Pin Type	Power Supply	Notes
PCI_DEVSEL/CE_PF[16]	E26	I/O	OV_{DD}	5
PCI_IDSEL/CE_PF[17]	F22	I/O	OV_{DD}	—
PCI_SERR/CE_PF[18]	B29	I/O	OV_{DD}	5
PCI_PERR/CE_PF[19]	A29	I/O	OV_{DD}	5
PCI_REQ[0]/CE_PF[20]	F19	I/O	$LV_{DD}2$	—
PCI_REQ[1]/CPCI_HS_ES/CE_PF[21]	A21	I/O	$LV_{DD}2$	—
PCI_REQ[2]/CE_PF[22]	C21	I/O	$LV_{DD}2$	—
PCI_GNT[0]/CE_PF[23]	E20	I/O	$LV_{DD}2$	—
PCI_GNT[1]/CPCI1_HS_LED/CE_PF[24]	B20	I/O	$LV_{DD}2$	—
PCI_GNT[2]/CPCI1_HS_ENUM/CE_PF[25]	C20	I/O	$LV_{DD}2$	—
PCI_MODE	D36	I	OV_{DD}	—
M66EN/CE_PF[4]	B37	I/O	OV_{DD}	—
Local Bus Controller Interface				
LAD[0:31]	N32, N33, N35, N36, P37, P32, P34, R36, R35, R34, R33, T37, T35, T34, T33, U37, T32, U36, U34, V36, V35, W37, W35, V33, V32, W34, Y36, W32, AA37, Y33, AA35, AA34	I/O	OV_{DD}	—
LDP[0]/CKSTOP_OUT	AB37	I/O	OV_{DD}	—
LDP[1]/CKSTOP_IN	AB36	I/O	OV_{DD}	—
LDP[2]/LCS[6]	AB35	I/O	OV_{DD}	—
LDP[3]/LCS[7]	AA33	I/O	OV_{DD}	—
LA[27:31]	AC37, AA32, AC36, AC34, AD36	O	OV_{DD}	—
LCS[0:5]	AD33, AG37, AF34, AE33, AD32, AH37	O	OV_{DD}	—
LWE[0:3]/LSDDQM[0:3]/LBS[0:3]	AG35, AG34, AH36, AE32	O	OV_{DD}	—
LBCTL	AD35	O	OV_{DD}	—
LALE	M37	O	OV_{DD}	—
LGPL0/LSDA10/cfg_reset_source0	AB32	I/O	OV_{DD}	—
LGPL1/LSDWE/cfg_reset_source1	AE37	I/O	OV_{DD}	—
LGPL2/LSDRAS/LOE	AC33	O	OV_{DD}	—
LGPL3/LSDCAS/cfg_reset_source2	AD34	I/O	OV_{DD}	—
LGPL4/LGTA/LUPWAIT/LPBSE	AE35	I/O	OV_{DD}	—
LGPL5/cfg_clkin_div	AF36	I/O	OV_{DD}	—
LCKE	G36	O	OV_{DD}	—
LCLK[0]	J33	O	OV_{DD}	—
LCLK[1]/LCS[6]	J34	O	OV_{DD}	—

Table 67. MPC8358E TBGA Pinout Listing (continued)

Signal	Package Pin Number	Pin Type	Power Supply	Notes
$LV_{DD}1$	C17, D16	Power for UCC2 Ethernet interface option 1 (2.5 V, 3.3 V)	$LV_{DD}1$	9
$LV_{DD}2$	B18, E21	Power for UCC2 Ethernet interface option 2 (2.5 V, 3.3 V)	$LV_{DD}2$	9
V_{DD}	C36, D29, D35, E16, F9, F12, F15, F17, F18, F20, F21, F23, F25, F26, F29, F31, F32, F33, G6, J6, K32, M32, N6, P33, R6, R32, U32, V6, Y5, Y32, AB6, AB33, AD6, AF32, AK6, AL6, AM7, AM9, AM10, AM11, AM12, AM13, AM14, AM15, AM18, AM21, AM25, AM28, AM32, AN15, AN21, AN26, AU9, AU17	Power for core (1.2 V)	V_{DD}	—
OV_{DD}	A10, B9, B15, B32, C1, C12, C22, C29, D24, E3, E10, E27, G4, H35, J1, J35, K2, M4, N3, N34, R2, R37, T36, U2, U33, V4, V34, W3, Y35, Y37, AA1, AA36, AB2, AB34	PCI, 10/100 Ethernet, and other standard (3.3 V)	OV_{DD}	—
MVREF1	AN20	I	DDR reference voltage	—
MVREF2	AU32	I	DDR reference voltage	—
SPARE1	B11	I/O	OV_{DD}	8
SPARE3	AH32	—	GV_{DD}	8
SPARE4	AU18	—	GV_{DD}	7
SPARE5	AP1	—	GV_{DD}	8

Table 67. MPC8358E TBGA Pinout Listing (continued)

Signal	Package Pin Number	Pin Type	Power Supply	Notes
No Connect				
NC	AM16, AM17, AM20, AN13, AN16, AN17, AP10, AP11, AP13, AP15, AP18, AR11, AR13, AR14, AR15, AR16, AR17, AR20, AT11, AT12, AT13, AT14, AT16, AT17, AT18, AU10, AU11, AU12, AU13, AU15, AU19	—	—	—

Notes:

1. This pin is an open drain signal. A weak pull-up resistor (1 kΩ) should be placed on this pin to OV_{DD}.
2. This pin is an open drain signal. A weak pull-up resistor (2–10 kΩ) should be placed on this pin to OV_{DD}.
3. This output is actively driven during reset rather than being three-stated during reset.
4. These JTAG pins have weak internal pull-up P-FETs that are always enabled.
5. This pin should have a weak pull up if the chip is in PCI host mode. Follow PCI specifications recommendation.
6. These are On Die Termination pins, used to control DDR2 memories internal termination resistance.
7. This pin must always be tied to GND.
8. This pin must always be left not connected.
9. Refer to *MPC8360E PowerQUICC II Pro Integrated Communications Processor Reference Manual* section on “RGMII Pins,” for information about the two UCC2 Ethernet interface options.
10. This pin must always be tied to GV_{DD}.
11. It is recommended that MDIC0 be tied to GND using an 18.2 Ω resistor and MDIC1 be tied to DDR power using an 18.2 Ω resistor for DDR2.

This figure shows the internal distribution of clocks within the MPC8358E.

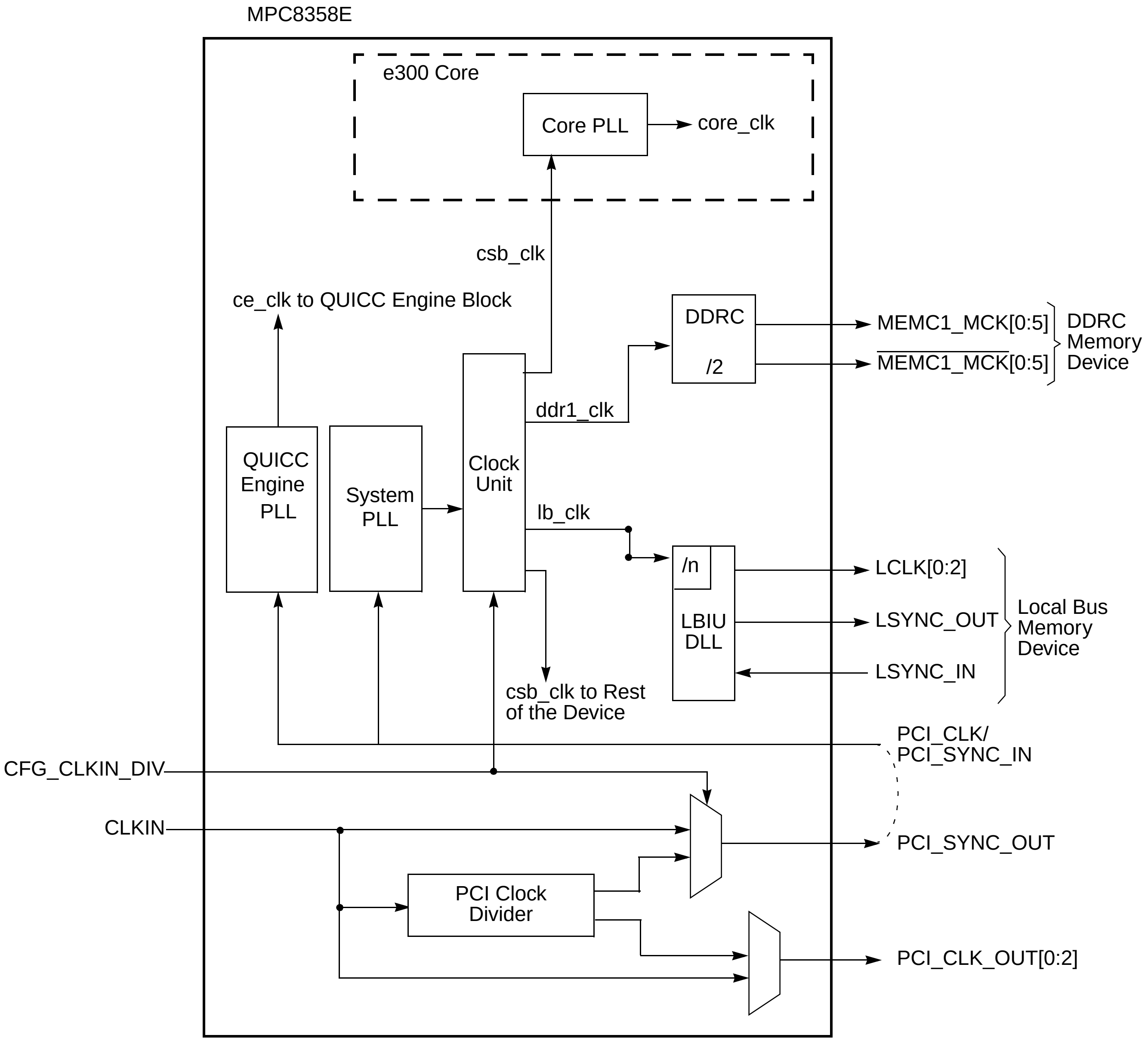


Figure 55. MPC8358E Clock Subsystem

The primary clock source for the device can be one of two inputs, CLKIN or PCI_CLK, depending on whether the device is configured in PCI host or PCI agent mode. Note that in PCI host mode, the primary clock input also depends on whether PCI clock outputs are selected with RCWH[PCICKDRV]. When the device is configured as a PCI host device (RCWH[PCIHOST] = 1) and PCI clock output is selected (RCWH[PCICKDRV] = 1), CLKIN is its primary input clock. CLKIN feeds the PCI clock divider (÷2) and the multiplexors for PCI_SYNC_OUT and PCI_CLK_OUT. The CFG_CLKIN_DIV configuration input selects whether CLKIN or CLKIN/2 is driven out on the PCI_SYNC_OUT signal. The OCCR[PCIOEN*n*] parameters enable the PCI_CLK_OUT*n*, respectively.

PCI_SYNC_OUT is connected externally to PCI_SYNC_IN to allow the internal clock subystem to synchronize to the system PCI clocks. PCI_SYNC_OUT must be connected properly to PCI_SYNC_IN, with equal delay to all PCI agent devices in the system, to allow the device to function. When the device is configured as a PCI agent device, PCI_CLK is the primary input clock. When the device is configured as a PCI agent device the CLKIN and the CFG_CLKIN_DIV signals should be tied to GND.

When the device is configured as a PCI host device (RCWH[PCIHOST] = 1) and PCI clock output is disabled (RCWH[PCICKDRV] = 0), clock distribution and balancing done externally on the board. Therefore, PCI_SYNC_IN is the primary input clock.

As shown in Figure 54 and Figure 55, the primary clock input (frequency) is multiplied by the QUICC Engine block phase-locked loop (PLL), the system PLL, and the clock unit to create the QUICC Engine clock (*ce_clk*), the coherent system bus clock (*csb_clk*), the internal DDRC1 controller clock (*ddr1_clk*), and the internal clock for the local bus interface unit and DDR2 memory controller (*lb_clk*).

The *csb_clk* frequency is derived from a complex set of factors that can be simplified into the following equation:

$$csb\_clk = \{\text{PCI\_SYNC\_IN} \times (1 + \text{CFG\_CLKIN\_DIV})\} \times \text{SPMF}$$

In PCI host mode, PCI_SYNC_IN × (1 + CFG_CLKIN_DIV) is the CLKIN frequency; in PCI agent mode, CFG_CLKIN_DIV must be pulled down (low), so PCI_SYNC_IN × (1 + CFG_CLKIN_DIV) is the PCI_CLK frequency.

The *csb_clk* serves as the clock input to the e300 core. A second PLL inside the e300 core multiplies up the *csb_clk* frequency to create the internal clock for the e300 core (*core_clk*). The system and core PLL multipliers are selected by the SPMF and COREPLL fields in the reset configuration word low (RCWL) which is loaded at power-on reset or by one of the hard-coded reset options. See Chapter 4, “Reset, Clocking, and Initialization,” in the *MPC8360E PowerQUICC II Pro Integrated Communications Processor Reference Manual* for more information on the clock subsystem.

The *ce_clk* frequency is determined by the QUICC Engine PLL multiplication factor (RCWL[CEPMF) and the QUICC Engine PLL division factor (RCWL[CEPDF]) according to the following equation:

$$ce\_clk = (\text{primary clock input} \times \text{CEPMF}) \div (1 + \text{CEPDF})$$

The internal *ddr1_clk* frequency is determined by the following equation:

$$ddr1\_clk = csb\_clk \times (1 + \text{RCWL[DDR1CM]})$$

Note that the lb_clk clock frequency (for DDRC2) is determined by RCWL[LBCM]. The *internal ddr1_clk* frequency is not the external memory bus frequency; *ddr1_clk* passes through the DDRC1 clock divider (÷2) to create the differential DDRC1 memory bus clock outputs (MEMC1_MCK and $\overline{\text{MEMC1\_MCK}}$). However, the data rate is the same frequency as *ddr1_clk*.

The internal *lb_clk* frequency is determined by the following equation:

$$lb\_clk = csb\_clk \times (1 + \text{RCWL[LBCM]})$$

Note that *lb_clk* is not the external local bus or DDRC2 frequency; *lb_clk* passes through the a LB clock divider to create the external local bus clock outputs (LSYNC_OUT and LCLK[0:2]). The LB clock divider ratio is controlled by LCRR[CLKDIV].

Additionally, some of the internal units may be required to be shut off or operate at lower frequency than the *csb_clk* frequency. Those units have a default clock ratio that can be configured by a memory mapped register after the device comes out of reset. This table specifies which units have a configurable clock frequency.

Table 68. Configurable Clock Units

Unit	Default Frequency	Options
Security core	*csb_clk*/3	Off, *csb_clk*[1], *csb_clk*/2, *csb_clk*/3
PCI and DMA complex	csb_clk	Off, *csb_clk*

[1] With limitation, only for slow csb_clk rates, up to 166 MHz.

This table provides the operating frequencies for the TBGA package under recommended operating conditions (see Table 2). All frequency combinations shown in the table below may not be available. Maximum operating frequencies depend on the part

22.3.1 Experimental Determination of the Junction Temperature with a Heat Sink

When heat sink is used, the junction temperature is determined from a thermocouple inserted at the interface between the case of the package and the interface material. A clearance slot or hole is normally required in the heat sink. Minimizing the size of the clearance is important to minimize the change in thermal performance caused by removing part of the thermal interface to the heat sink. Because of the experimental difficulties with this technique, many engineers measure the heat sink temperature and then back calculate the case temperature using a separate measurement of the thermal resistance of the interface. From this case temperature, the junction temperature is determined from the junction-to-case thermal resistance.

$$T_J = T_C + (R_{\theta JC} \times P_D)$$

where:

T_J = junction temperature (° C)
T_C = case temperature of the package (° C)
$R_{\theta JC}$ = junction to case thermal resistance (° C/W)
P_D = power dissipation (W)

23 System Design Information

This section provides electrical and thermal design recommendations for successful application of the MPC8360E/58E. Additional information can be found in *MPC8360E/MPC8358E PowerQUICC Design Checklist* (AN3097).

23.1 System Clocking

The device includes two PLLs, as follows.

- The platform PLL ($AV_{DD}1$) generates the platform clock from the externally supplied CLKIN input. The frequency ratio between the platform and CLKIN is selected using the platform PLL ratio configuration bits as described in Section 21.1, “System PLL Configuration.”
- The e300 core PLL ($AV_{DD}2$) generates the core clock as a slave to the platform clock. The frequency ratio between the e300 core clock and the platform clock is selected using the e300 PLL ratio configuration bits as described in Section 21.2, “Core PLL Configuration.”

23.2 PLL Power Supply Filtering

Each of the PLLs listed above is provided with power through independent power supply pins ($AV_{DD}1$, $AV_{DD}2$, respectively). The AV_{DD} level should always be equivalent to V_{DD}, and preferably these voltages are derived directly from V_{DD} through a low frequency filter scheme such as the following.

There are a number of ways to reliably provide power to the PLLs, but the recommended solution is to provide five independent filter circuits as illustrated in Figure 56, one to each of the five AV_{DD} pins. By providing independent filters to each PLL, the opportunity to cause noise injection from one PLL to the other is reduced.

This circuit is intended to filter noise in the PLLs resonant frequency range from a 500 kHz to 10 MHz range. It should be built with surface mount capacitors with minimum Effective Series Inductance (ESL). Consistent with the recommendations of Dr. Howard Johnson in *High Speed Digital Design: A Handbook of Black Magic* (Prentice Hall, 1993), multiple small capacitors of equal value are recommended over a single large value capacitor.

Each circuit should be placed as close as possible to the specific AV_{DD} pin being supplied to minimize noise coupled from nearby circuits. It should be possible to route directly from the capacitors to the AV_{DD} pin, which is on the periphery of package, without the inductance of vias.

This figure shows the PLL power supply filter circuit.

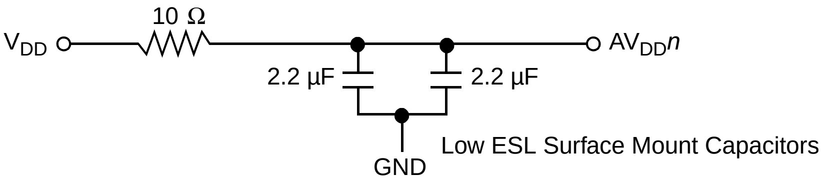


Figure 56. PLL Power Supply Filter Circuit

23.3 Decoupling Recommendations

Due to large address and data buses as well as high operating frequencies, the device can generate transient power surges and high frequency noise in its power supply, especially while driving large capacitive loads. This noise must be prevented from reaching other components in the device system, and the device itself requires a clean, tightly regulated source of power. Therefore, it is recommended that the system designer place at least one decoupling capacitor at each V_{DD}, OV_{DD}, GV_{DD}, and LV_{DD} pins of the device. These decoupling capacitors should receive their power from separate V_{DD}, OV_{DD}, GV_{DD}, LV_{DD}, and GND power planes in the PCB, utilizing short traces to minimize inductance. Capacitors may be placed directly under the device using a standard escape pattern. Others may surround the part.

These capacitors should have a value of 0.01 or 0.1 µF. Only ceramic SMT (surface mount technology) capacitors should be used to minimize lead inductance, preferably 0402 or 0603 sizes.

Additionally, it is recommended that there be several bulk storage capacitors distributed around the PCB, feeding the V_{DD}, OV_{DD}, GV_{DD}, and LV_{DD} planes, to enable quick recharging of the smaller chip capacitors. These bulk capacitors should have a low ESR (equivalent series resistance) rating to ensure the quick response time necessary. They should also be connected to the power and ground planes through two vias to minimize inductance. Suggested bulk capacitors—100–330 µF (AVX TPS tantalum or Sanyo OSCON).

23.4 Connection Recommendations

To ensure reliable operation, it is highly recommended to connect unused inputs to an appropriate signal level. Unused active low inputs should be tied to OV_{DD}, GV_{DD}, or LV_{DD} as required. Unused active high inputs should be connected to GND. All NC (no-connect) signals must remain unconnected.

Power and ground connections must be made to all external V_{DD}, GV_{DD}, LV_{DD}, OV_{DD}, and GND pins of the device.

23.5 Output Buffer DC Impedance

The device drivers are characterized over process, voltage, and temperature. For all buses, the driver is a push-pull single-ended driver type (open drain for I^2C).

To measure Z_0 for the single-ended drivers, an external resistor is connected from the chip pad to OV_{DD} or GND. Then, the value of each resistor is varied until the pad voltage is $OV_{DD}/2$ (see Figure 57). The output impedance is the average of two components, the resistances of the pull-up and pull-down devices. When data is held high, SW1 is closed (SW2 is open) and R_P is trimmed until the voltage at the pad equals $OV_{DD}/2$. R_P then becomes the resistance of the pull-up devices. R_P and R_N are designed to be close to each other in value. Then, $Z_0 = (R_P + R_N)/2$.