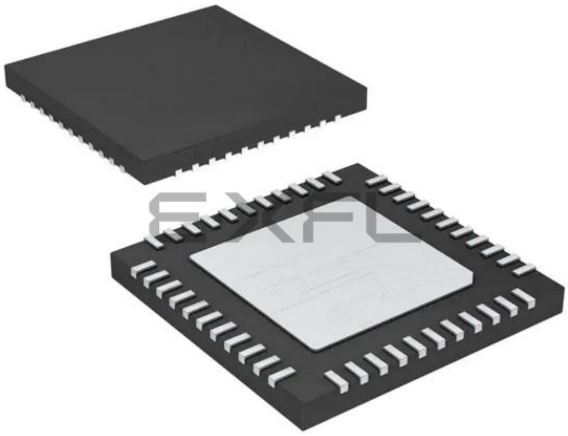


Welcome to E-XFL.COM

What is "Embedded - Microcontrollers"?

"Embedded - Microcontrollers" refer to small, integrated circuits designed to perform specific tasks within larger systems. These microcontrollers are essentially compact computers on a single chip, containing a processor core, memory, and programmable input/output peripherals. They are called "embedded" because they are embedded within electronic devices to control various functions, rather than serving as standalone computers. Microcontrollers are crucial in modern electronics, providing the intelligence and control needed for a wide range of applications.

Applications of "Embedded - Microcontrollers"

Details	
Product Status	Active
Core Processor	PIC
Core Size	8-Bit
Speed	32MHz
Connectivity	I²C, LINbus, SPI, UART/USART
Peripherals	Brown-out Detect/Reset, POR, PWM, WDT
Number of I/O	36
Program Memory Size	28KB (16K x 14)
Program Memory Type	FLASH
EEPROM Size	128 x 8
RAM Size	2K x 8
Voltage - Supply (Vcc/Vdd)	1.8V ~ 3.6V
Data Converters	A/D 28x10b; D/A 4x5b, 4x10b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	44-VQFN Exposed Pad
Supplier Device Package	44-QFN (8x8)
Purchase URL	https://www.e-xfl.com/product-detail/microchip-technology/pic16lf1779t-i-ml

1.1 Register and Bit naming conventions

1.1.1 REGISTER NAMES

When there are multiple instances of the same peripheral in a device, the peripheral control registers will be depicted as the concatenation of a peripheral identifier, peripheral instance, and control identifier. The control registers section will show just one instance of all the register names with an 'x' in the place of the peripheral instance number. This naming convention may also be applied to peripherals when there is only one instance of that peripheral in the device to maintain compatibility with other devices in the family that contain more than one.

1.1.2 BIT NAMES

There are two variants for bit names:

- Short name: Bit function abbreviation
- Long name: Peripheral abbreviation + short name

1.1.2.1 Short Bit Names

Short bit names are an abbreviation for the bit function. For example, some peripherals are enabled with the EN bit. The bit names shown in the registers are the short name variant.

Short bit names are useful when accessing bits in C programs. The general format for accessing bits by the short name is *RegisterName*bits.*ShortName*. For example, the enable bit, EN, in the COG1CON0 register can be set in C programs with the instruction `COG1CON0bits.EN = 1`.

Short names are generally not useful in assembly programs because the same name may be used by different peripherals in different bit positions. When this occurs, during the include file generation, all instances of that short bit name are appended with an underscore plus the name of the register in which the bit resides to avoid naming contentions.

1.1.2.2 Long Bit Names

Long bit names are constructed by adding a peripheral abbreviation prefix to the short name. The prefix is unique to the peripheral, thereby making every long bit name unique. The long bit name for the COG1 enable bit is the COG1 prefix, G1, appended with the enable bit short name, EN, resulting in the unique bit name G1EN.

Long bit names are useful in both C and assembly programs. For example, in C the COG1CON0 enable bit can be set with the `G1EN = 1` instruction. In assembly, this bit can be set with the `BSF COG1CON0,G1EN` instruction.

1.1.2.3 Bit Fields

Bit fields are two or more adjacent bits in the same register. Bit fields adhere only to the short bit naming convention. For example, the three Least Significant bits of the COG1CON0 register contain the mode control bits. The short name for this field is MD. There is no long bit name variant. Bit field access is only possible in C programs. The following example demonstrates a C program instruction for setting the COG1 to the Push-Pull mode:

```
COG1CON0bits.MD = 0x5;
```

Individual bits in a bit field can also be accessed with long and short bit names. Each bit is the field name appended with the number of the bit position within the field. For example, the Most Significant mode bit has the short bit name MD2 and the long bit name is G1MD2. The following two examples demonstrate assembly program sequences for setting the COG1 to the Push-Pull mode:

EXAMPLE 1-1:

```
MOVLW  ~(1<<G1MD1)
ANDWF  COG1CON0,F
MOVLW  1<<G1MD2 | 1<<G1MD0
IORWF  COG1CON0,F
```

EXAMPLE 1-2:

```
BSF    COG1CON0,G1MD2
BCF    COG1CON0,G1MD1
BSF    COG1CON0,G1MD0
```

1.1.3 REGISTER AND BIT NAMING EXCEPTIONS

1.1.3.1 Status, Interrupt, and Mirror Bits

Status, interrupt enables, interrupt flags, and mirror bits are contained in registers that span more than one peripheral. In these cases, the bit name shown is unique so there is no prefix or short name variant.

1.1.3.2 Legacy Peripherals

There are some peripherals that do not strictly adhere to these naming conventions. Peripherals that have existed for many years and are present in almost every device are the exceptions. These exceptions were necessary to limit the adverse impact of the new conventions on legacy code. Peripherals that do adhere to the new convention will include a table in the registers section indicating the long name prefix for each peripheral instance. Peripherals that fall into the exception category will not have this table. These peripherals include, but are not limited to, the following:

- EUSART
- MSSP

TABLE 1-2: PIC16(L)F1778 PINOUT DESCRIPTION (CONTINUED)

Name	Function	Input Type	Output Type	Description
RB1/AN10/PRG1IN1/PRG2IN0/HIB1/C1IN3-/C2IN3-/C3IN3-/C4IN3-/OPA2OUT/OPA1IN1+/OPA1IN1-/COG2IN	RB1	TTL/ST	CMOS	General purpose I/O.
	AN10	AN	—	ADC Channel 10 input.
	PRG1IN1	AN	—	Ramp generator 1 reference voltage input.
	PRG2IN0	AN	—	Ramp generator 2 reference voltage input.
	HIB1	HP	HP	High-Power output.
	C1IN3-	AN	—	Comparator 1 negative input.
	C2IN3-	AN	—	Comparator 2 negative input.
	C3IN3-	AN	—	Comparator 3 negative input.
	C4IN3-	AN	—	Comparator 4 negative input.
	OPA2OUT	—	AN	Operational amplifier 2 output.
	OPA1IN1+	AN	—	Operational amplifier 1 non-inverting input.
	OPA1IN1-	AN	—	Operational amplifier 1 inverting input.
	COG2IN(1)	TTL/ST	—	Complementary output generator 2 input.
RB2/AN8/OPA2IN0-/DAC3OUT1/COG3IN	RB2	TTL/ST	CMOS	General purpose I/O.
	AN8	AN	—	ADC Channel 8 input.
	OPA2IN0-	AN	—	Operational amplifier 2 inverting input.
	DAC3OUT1	—	AN	DAC3 voltage output.
	COG3IN(1)	TTL/ST	—	Complementary output generator 3 input.
RB3/AN9/C1IN2-/C2IN2-/C3IN2-/OPA2IN0+/MD3CL	RB3	TTL/ST	CMOS	General purpose I/O.
	AN9	AN	—	ADC Channel 9 input.
	C1IN2-	AN	—	Comparator 1 negative input.
	C2IN2-	AN	—	Comparator 2 negative input.
	C3IN2-	AN	—	Comparator 3 negative input.
	OPA2IN0+	AN		Operational amplifier 2 non-inverting input.
	MD3CL(1)	TTL/ST	—	Data signal modulator 3 low carrier input.
RB4/AN11/C3IN1+/T5G/MD3CH	RB4	TTL/ST	CMOS	General purpose I/O.
	AN11	AN	—	ADC Channel 11 input.
	C3IN1+	AN	—	Comparator 3 positive input.
	T5G(1)	TTL/ST	—	Timer5 gate input.
	MD3CH(1)	TTL/ST	—	Data signal modulator 3 high carrier input.
RB5/AN13/DAC5REF1-/DAC7REF1-/C4IN2-/T1G/CCP7/MD3MOD	RB5	TTL/ST	CMOS	General purpose I/O.
	AN13	AN	—	ADC Channel 11 input.
	DAC5REF1-	AN	—	DAC5 negative reference.
	DAC7REF1-	AN	—	DAC7 negative reference.
	C4IN2-	AN	—	Comparator 4 negative input.
	T1G(1)	TTL/ST	—	Timer1 gate input.
	CCP7(1)	TTL/ST	—	CCP7 capture input.
	MD3MOD(1)	TTL/ST	—	Data signal modulator modulation input.

Legend: AN = Analog input or output CMOS = CMOS compatible input or output OD = Open-Drain
TTL = TTL compatible input ST = Schmitt Trigger input with CMOS levels I^2C = Schmitt Trigger input with I^2C
HP = High Power XTAL = Crystal levels

Note 1: Default peripheral input. Alternate pins can be selected as the peripheral input with the PPS input selection registers.
2: All pin digital outputs default to PORT latch data. Alternate outputs can be selected as the peripheral digital output with the PPS output selection registers.
3: These peripheral functions are bidirectional. The output pin selections must be the same as the input pin selections.

5.3 Clock Switching

The system clock source can be switched between external and internal clock sources via software using the System Clock Select (SCS) bits of the OSCCON register. The following clock sources can be selected using the SCS bits:

- Default system oscillator determined by FOSC bits in Configuration Words
- Timer1 32 kHz crystal oscillator
- Internal Oscillator Block (INTOSC)

5.3.1 SYSTEM CLOCK SELECT (SCS) BITS

The System Clock Select (SCS) bits of the OSCCON register select the system clock source that is used for the CPU and peripherals.

- When the SCS bits of the OSCCON register = `00`, the system clock source is determined by the value of the FOSC<2:0> bits in the Configuration Words.
- When the SCS bits of the OSCCON register = `01`, the system clock source is the secondary oscillator.
- When the SCS bits of the OSCCON register = `1x`, the system clock source is chosen by the internal oscillator frequency selected by the IRCF<3:0> bits of the OSCCON register. After a Reset, the SCS bits of the OSCCON register are always cleared.

Note:	Any automatic clock switch, which may occur from Two-Speed Start-up or Fail-Safe Clock Monitor, does not update the SCS bits of the OSCCON register. The user can monitor the OSTS bit of the OSCSTAT register to determine the current system clock source.

When switching between clock sources, a delay is required to allow the new clock to stabilize. These oscillator delays are shown in Table 5-1.

5.3.2 OSCILLATOR START-UP TIMER STATUS (OSTS) BIT

The Oscillator Start-up Timer Status (OSTS) bit of the OSCSTAT register indicates whether the system clock is running from the external clock source, as defined by the FOSC<2:0> bits in the Configuration Words, or from the internal clock source. In particular, OSTS indicates that the Oscillator Start-up Timer (OST) has timed out for LP, XT or HS modes. The OST does not reflect the status of the secondary oscillator.

5.3.3 SECONDARY OSCILLATOR

The secondary oscillator is a separate crystal oscillator associated with the Timer1 peripheral. It is optimized for timekeeping operations with a 32.768 kHz crystal connected between the SOSCO and SOSCI device pins.

The secondary oscillator is enabled using the OSCEN control bit in the T1CON register. See **Section 22.0 "Timer1/3/5 Module with Gate Control"** for more information about the Timer1 peripheral.

5.3.4 SECONDARY OSCILLATOR READY (SOSCR) BIT

The user must ensure that the secondary oscillator is ready to be used before it is selected as a system clock source. The Secondary Oscillator Ready (SOSCR) bit of the OSCSTAT register indicates whether the secondary oscillator is ready to be used. After the SOSCR bit is set, the SCS bits can be configured to select the secondary oscillator.

5.3.5 CLOCK SWITCH BEFORE SLEEP

When a clock switch from an old clock to a new clock is requested just prior to entering Sleep mode, it is necessary to confirm that the switch is complete before the sleep instruction is executed. Failure to do so may result in an incomplete switch and consequential loss of the system clock altogether. Clock switching is confirmed by monitoring the clock status bits in the OSCSTAT register. Switch confirmation can be accomplished by sensing that the ready bit for the new clock is set or the ready bit for the old clock is cleared. For example, when switching between the internal oscillator with the PLL and the internal oscillator without the PLL, monitor the PLLR bit. When PLLR is set, the switch to 32 MHz operation is complete. Conversely, when PLLR is cleared the switch from 32 MHz operation to the selected internal clock is complete.

7.3 Interrupts During Sleep

Some interrupts can be used to wake from Sleep. To wake from Sleep, the peripheral must be able to operate without the system clock. The interrupt source must have the appropriate Interrupt Enable bit(s) set prior to entering Sleep.

On waking from Sleep, if the GIE bit is also set, the processor will branch to the interrupt vector. Otherwise, the processor will continue executing instructions after the SLEEP instruction. The instruction directly after the SLEEP instruction will always be executed before branching to the ISR. Refer to **Section 8.0 "Power-Down Mode (Sleep)"** for more details.

7.4 INT Pin

The INT pin can be used to generate an asynchronous edge-triggered interrupt. This interrupt is enabled by setting the INTE bit of the INTCON register. The INTEDG bit of the OPTION_REG register determines on which edge the interrupt will occur. When the INTEDG bit is set, the rising edge will cause the interrupt. When the INTEDG bit is clear, the falling edge will cause the interrupt. The INTF bit of the INTCON register will be set when a valid edge appears on the INT pin. If the GIE and INTE bits are also set, the processor will redirect program execution to the interrupt vector.

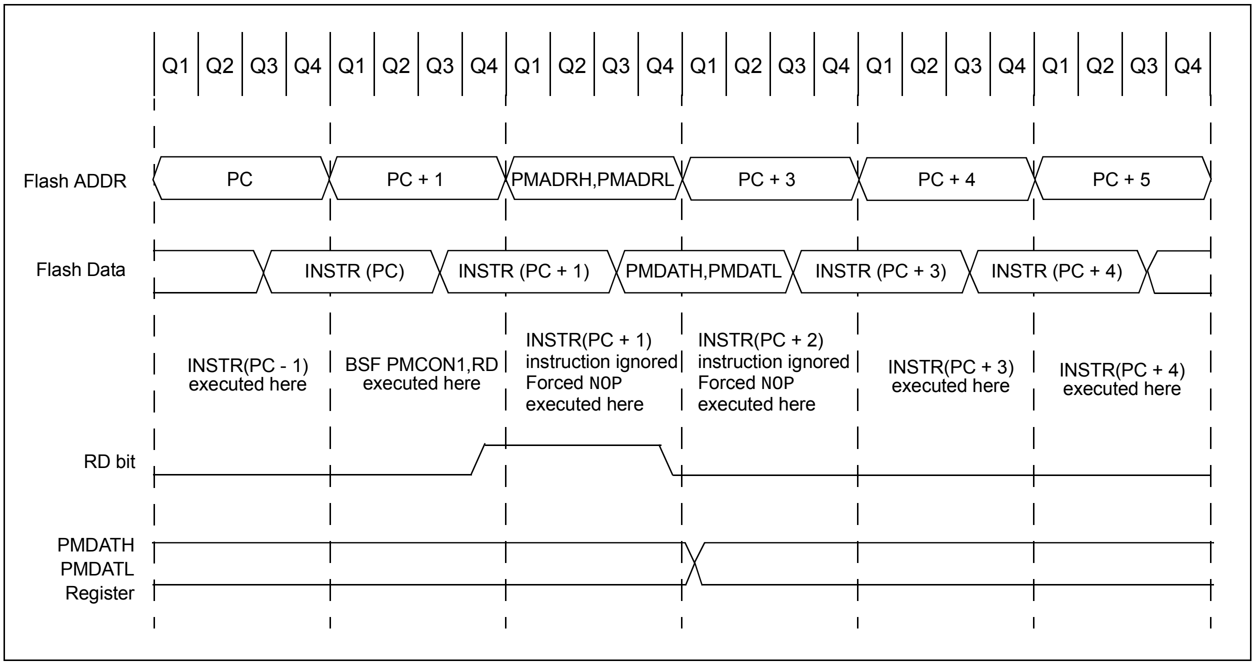
7.5 Automatic Context Saving

Upon entering an interrupt, the return PC address is saved on the stack. Additionally, the following registers are automatically saved in the shadow registers:

- W register
- STATUS register (except for $\overline{TO}$ and $\overline{PD}$)
- BSR register
- FSR registers
- PCLATH register

Upon exiting the Interrupt Service Routine, these registers are automatically restored. Any modifications to these registers during the ISR will be lost. If modifications to any of these registers are desired, the corresponding shadow register should be modified and the value will be restored when exiting the ISR. The shadow registers are available in Bank 31 and are readable and writable. Depending on the user's application, other registers may also need to be saved.

FIGURE 10-2: FLASH PROGRAM MEMORY READ CYCLE EXECUTION



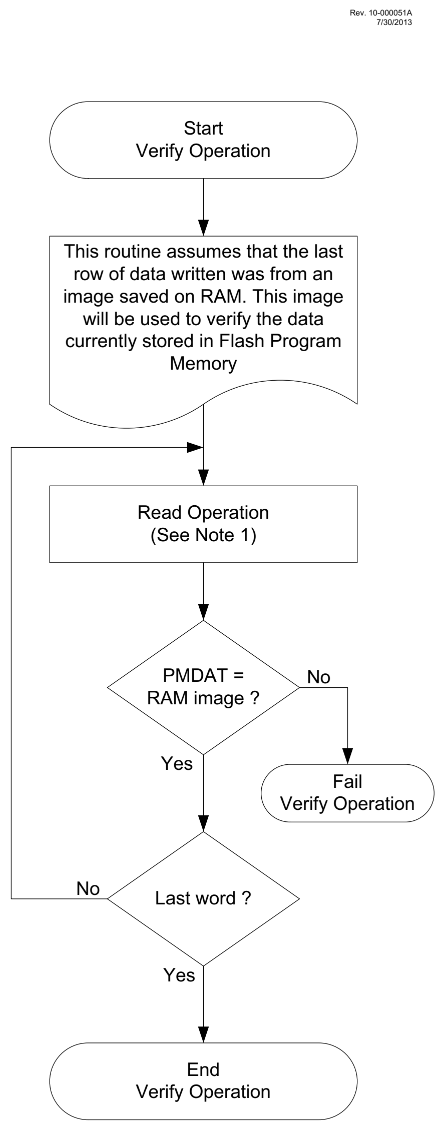
EXAMPLE 10-1: FLASH PROGRAM MEMORY READ

```
* This code block will read 1 word of program
* memory at the memory address:
    PROG_ADDR_HI : PROG_ADDR_LO
*   data will be returned in the variables;
*   PROG_DATA_HI, PROG_DATA_LO

    BANKSEL  PMADRL           ; Select Bank for PMCON registers
    MOVLW    PROG_ADDR_LO     ;
    MOVWF    PMADRL           ; Store LSB of address
    MOVLW    PROG_ADDR_HI     ;
    MOVWF    PMADRH           ; Store MSB of address

    BCF      PMCON1,CFGS      ; Do not select Configuration Space
    BSF      PMCON1,RD        ; Initiate read
    NOP                       ; Ignored (Figure 10-1)
    NOP                       ; Ignored (Figure 10-1)

    MOVF     PMDATL,W         ; Get LSB of word
    MOVWF    PROG_DATA_LO     ; Store in user location
    MOVF     PMDATH,W         ; Get MSB of word
    MOVWF    PROG_DATA_HI     ; Store in user location
```

10.5 Write/Verify

It is considered good programming practice to verify that program memory writes agree with the intended value. Since program memory is stored as a full page then the stored program memory contents are compared with the intended data stored in RAM after the last write is complete.

FIGURE 10-8: FLASH PROGRAM MEMORY VERIFY FLOWCHART



Note 1: See Figure 10-1.

12.8 Register Definitions: PPS Input Selection

REGISTER 12-1: xxxPPS: PERIPHERAL xxx INPUT SELECTION

U-0	U-0	R/W-q/u	R/W-q/u	R/W-q/u	R/W-q/u	R/W-q/u	R/W-q/u
—	—	xxxPPS<5:0>					
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as ‘0’
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
‘1’ = Bit is set	‘0’ = Bit is cleared	q = value depends on peripheral

bit 7-6 **Unimplemented:** Read as ‘0’

bit 5-3 **xxxPPS<5:3>:** Peripheral xxx Input PORT Selection bits
100 = Peripheral input is PORTE
011 = Peripheral input is PORTD(1)
010 = Peripheral input is PORTC
001 = Peripheral input is PORTB
000 = Peripheral input is PORTA

bit 2-0 **xxxPPS<2:0>:** Peripheral xxx Input Bit Selection bits(1)
111 = Peripheral input is from PORTx Bit 7 (Rx7)
110 = Peripheral input is from PORTx Bit 6 (Rx6)
101 = Peripheral input is from PORTx Bit 5 (Rx5)
100 = Peripheral input is from PORTx Bit 4 (Rx4)
011 = Peripheral input is from PORTx Bit 3 (Rx3)
010 = Peripheral input is from PORTx Bit 2 (Rx2)
001 = Peripheral input is from PORTx Bit 1 (Rx1)
000 = Peripheral input is from PORTx Bit 0 (Rx0)

Note 1: See Table 12-1 for xxxPPS register list and Reset values.
2: PIC16(L)F1777/9 only.

REGISTER 12-2: RxyPPS: PIN Rxy OUTPUT SOURCE SELECTION REGISTER

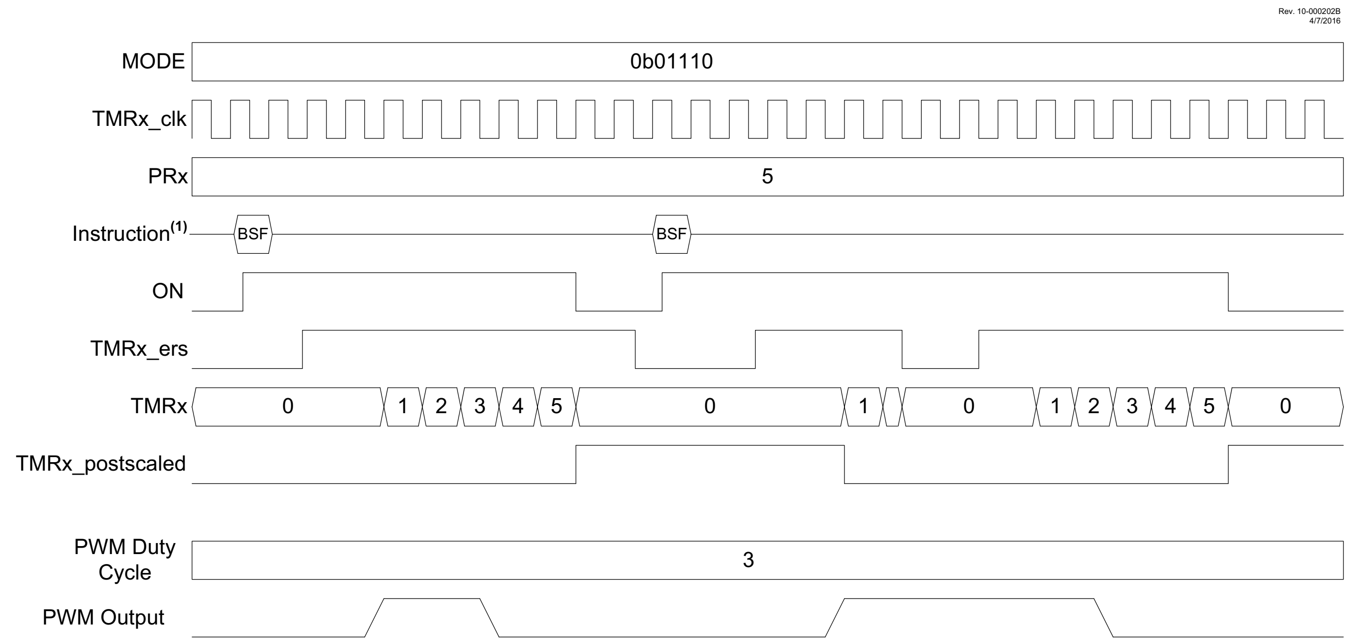
U-0	U-0	R/W-0/u	R/W-0/u	R/W-0/u	R/W-0/u	R/W-0/u	R/W-0/u
—	—	RxyPPS<5:0>					
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as ‘0’
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
‘1’ = Bit is set	‘0’ = Bit is cleared	

bit 7-6 **Unimplemented:** Read as ‘0’

bit 5-0 **RxyPPS<5:0>:** Pin Rxy Output Source Selection bits
Selection code determines the output signal on the port pin.
See Table 12-2 for the selection codes

FIGURE 23-11: LOW LEVEL RESET, EDGE-TRIGGERED HARDWARE LIMIT ONE-SHOT MODE TIMING DIAGRAM (MODE = `01110`)



Note 1: BSF and BCF represent Bit-Set File and Bit-Clear File instructions executed by the CPU to set or clear the ON bit of TxCON. CPU execution is asynchronous to the timer clock input.

REGISTER 23-2: TxCON: TIMERx CONTROL REGISTER

R/W/HC-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0
ON(1)	CKPS<2:0>			OUTPS<3:0>			
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as ‘0’
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
‘1’ = Bit is set	‘0’ = Bit is cleared	HC = Bit is cleared by hardware

bit 7 **ON:** Timerx On bit
`1` = Timerx is on
`0` = Timerx is off: all counters and state machines are reset

bit 6-4 **CKPS<2:0>:** Timer2-type Clock Prescale Select bits
`111` = 1:128 Prescaler
`110` = 1:64 Prescaler
`101` = 1:32 Prescaler
`100` = 1:16 Prescaler
`011` = 1:8 Prescaler
`010` = 1:4 Prescaler
`001` = 1:2 Prescaler
`000` = 1:1 Prescaler

bit 3-0 **OUTPS<3:0>:** Timerx Output Postscaler Select bits
`1111` = 1:16 Postscaler
`1110` = 1:15 Postscaler
`1101` = 1:14 Postscaler
`1100` = 1:13 Postscaler
`1011` = 1:12 Postscaler
`1010` = 1:11 Postscaler
`1001` = 1:10 Postscaler
`1000` = 1:9 Postscaler
`0111` = 1:8 Postscaler
`0110` = 1:7 Postscaler
`0101` = 1:6 Postscaler
`0100` = 1:5 Postscaler
`0011` = 1:4 Postscaler
`0010` = 1:3 Postscaler
`0001` = 1:2 Postscaler
`0000` = 1:1 Postscaler

Note 1: In certain modes, the ON bit will be auto-cleared by hardware. See **Section 23.6 “Operation Examples”**.

24.3 PWM Overview

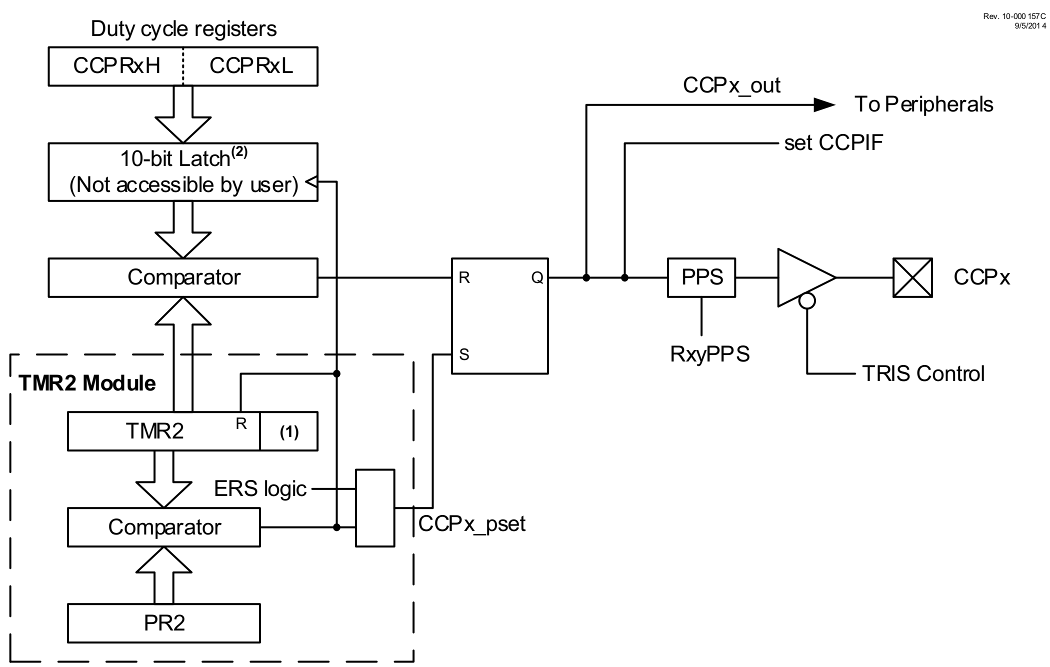
Pulse-Width Modulation (PWM) is a scheme that provides power to a load by switching quickly between fully on and fully off states. The PWM signal resembles a square wave where the high portion of the signal is considered the on state and the low portion of the signal is considered the off state. The high portion, also known as the pulse width, can vary in time and is defined in steps. A larger number of steps applied, which lengthens the pulse width, also supplies more power to the load. Lowering the number of steps applied, which shortens the pulse width, supplies less power. The PWM period is defined as the duration of one complete cycle or the total amount of on and off time combined.

PWM resolution defines the maximum number of steps that can be present in a single PWM period. A higher resolution allows for more precise control of the pulse width time and in turn the power that is applied to the load.

The term duty cycle describes the proportion of the on time to the off time and is expressed in percentages, where 0% is fully off and 100% is fully on. A lower duty cycle corresponds to less power applied and a higher duty cycle corresponds to more power applied.

Figure 24-3 shows a typical waveform of the PWM signal.

FIGURE 24-3: SIMPLIFIED PWM BLOCK DIAGRAM



Notes:
1. 8-bit timer is concatenated with two bits generated by Fosc or two bits of the internal prescaler to create 10-bit time-base.
2. The alignment of the 10 bits from the CCPR register is determined by the FMT bit.

30.0 PROGRAMMABLE RAMP GENERATOR (PRG) MODULE

The Programmable Ramp Generator (PRG) module is designed to provide rising and falling linear ramps. Typical applications include slope compensation for fixed frequency, continuous current, and Current mode switched power supplies. Slope compensation is a necessary feature of these power supplies because it prevents frequency instabilities at duty cycles greater than 50%.

The PRG has the following features:

- Linear positive and negative voltage ramp outputs
- Programmable current source/sink
- Internal and external reference voltage selection
- Internal and external timing source selection

A simplified block diagram of the PRG is shown in Figure 30-1.

TABLE 30-1: AVAILABLE PRG MODULES

Device	PRG1	PRG2	PRG3	PRG4
PIC16(L)F1778	●	●	●	
PIC16(L)F1777/9	●	●	●	●

30.1 Fundamental Operation

The PRG can be operated in three voltage ramp generator modes:

- Falling Voltage (slope compensation)
- Rising Voltage
- Alternating Rising and Falling Voltage

In the Rising or Falling mode an internal capacitor is discharged when the set_falling timing input is true and charged by an internally generated constant current when the set_rising timing input is true. The resulting linear ramp starts at the selected voltage input level and resets back to that level when the ramp is terminated by the set_falling timing input. The set_falling input dominates when both timing inputs are true.

To control the operation with a single-ended source, select the same source for both the set_rising and set_falling inputs and invert the polarity of one of them with the corresponding polarity control bit.

In the Alternating mode the capacitor is not discharged but alternates between being charged in one direction then the other.

Input selections are identical for all modes. The input voltage is supplied by any of the following:

- The PRGxIN0 or PRGxIN1 pins
- The buffered output of the internal Fixed Voltage Reference (FVR),
- Any of the internal DACs.

The timing sources are selected from the following:

- The synchronized output of any comparator
- Any PWM output
- Any I/O pin

The ramp output is available as an input to any of the comparators or op amps.

30.1.1 SLOPE COMPENSATION

Slope compensation works by quickly discharging an internal capacitor at the beginning of each PWM period. One side of the internal capacitor is connected to the voltage input source and the other side is connected to the internal current sink. The internal current sink charges this capacitor at a programmable rate. As the capacitor charges, the capacitor voltage is subtracted from the voltage source, producing a linear voltage decay at the required rate (see Figure 30-2). The ramp terminates and the capacitor is discharged when the set_falling timing input goes true. The next ramp starts when the set_rising timing input goes true.

Enabling the optional one-shot by setting the OS bit of the PRGxCON0 register ensures that the capacitor is fully discharged by overriding the set_rising timing input and holding the shorting switch closed for at least the one-shot period, typically 50 ns. Edge sensitive timing inputs that occur during the one-shot period will be ignored. Level sensitive timing inputs that occur during, and extend beyond, the one-shot period will be suspended until the end of the one-shot time.

30.1.2 RAMP GENERATION

Ramp generation is similar to slope compensation except that the slope is either both rising and falling or just rising.

30.1.2.1 Alternating Rising/Falling Ramps

The alternating rising/falling ramp generation function works by employing the built-in current source and sink and relying on the synchronous control of the internal analog switches and timing sources to ramp the module's output voltage up and then subsequently down.

Once initialized, the output voltage is ramped up linearly by the current source at a programmable rate until the set_falling timing source goes true, at which point the current source is disengaged. At the same time, the current sink is engaged to linearly ramp down the output voltage, also at a programmable rate, until the set_rising timing input goes true, thereby reversing the ramp slope. The process then repeats to create a saw tooth like waveform as shown in Figure 30-3 and Figure 30-4.

The set_rising and set_falling timing inputs can be either edge or level sensitive which is selected with the respective REDG and FEDG bits of the PRGxCON0 register. Edge sensitive operation is recommended for periodic signals such as clocks, and level sensitive operation is recommended for analog limit triggers such as comparator outputs.

When the one-shot is enabled (OS bit is set) then both the falling and rising ramps will persist for a minimum of the one-shot period. Edge sensitive timing inputs that occur during the one-shot period will be ignored. Level

32.0 MASTER SYNCHRONOUS SERIAL PORT (MSSP) MODULE

32.1 MSSP Module Overview

The Master Synchronous Serial Port (MSSP) module is a serial interface useful for communicating with other peripheral or microcontroller devices. These peripheral devices may be serial EEPROMs, shift registers, display drivers, A/D converters, etc. The MSSP module can operate in one of two modes:

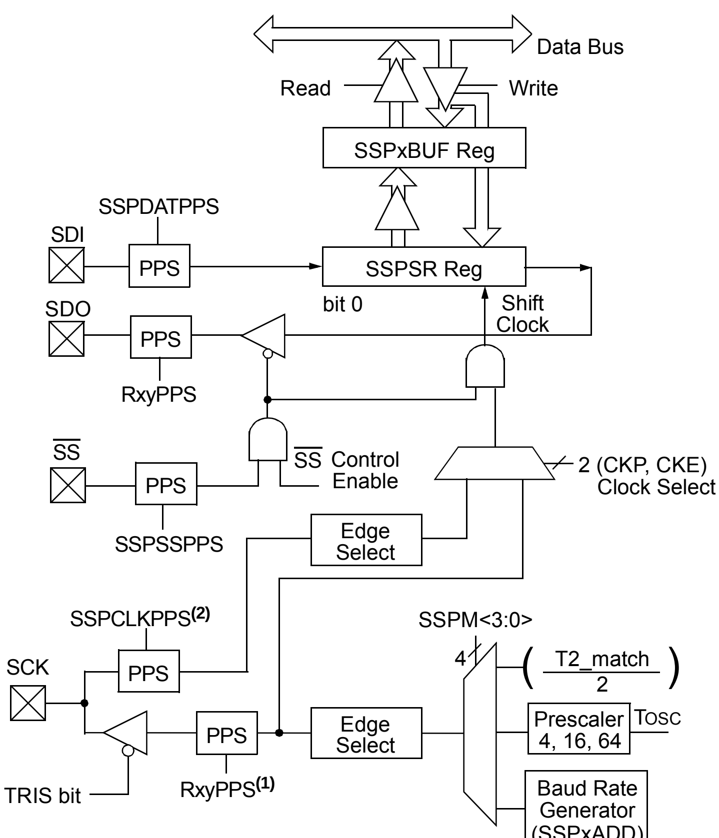
- Serial Peripheral Interface (SPI)
- Inter-Integrated Circuit (I^2C)

The SPI interface supports the following modes and features:

- Master mode
- Slave mode
- Clock Parity
- Slave Select Synchronization (Slave mode only)
- Daisy-chain connection of slave devices

Figure 32-1 is a block diagram of the SPI interface module.

FIGURE 32-1: MSSP BLOCK DIAGRAM (SPI MODE)



Note 1: Output selection for master mode

2: Input selection for slave mode

FIGURE 32-8: SLAVE SELECT SYNCHRONOUS WAVEFORM

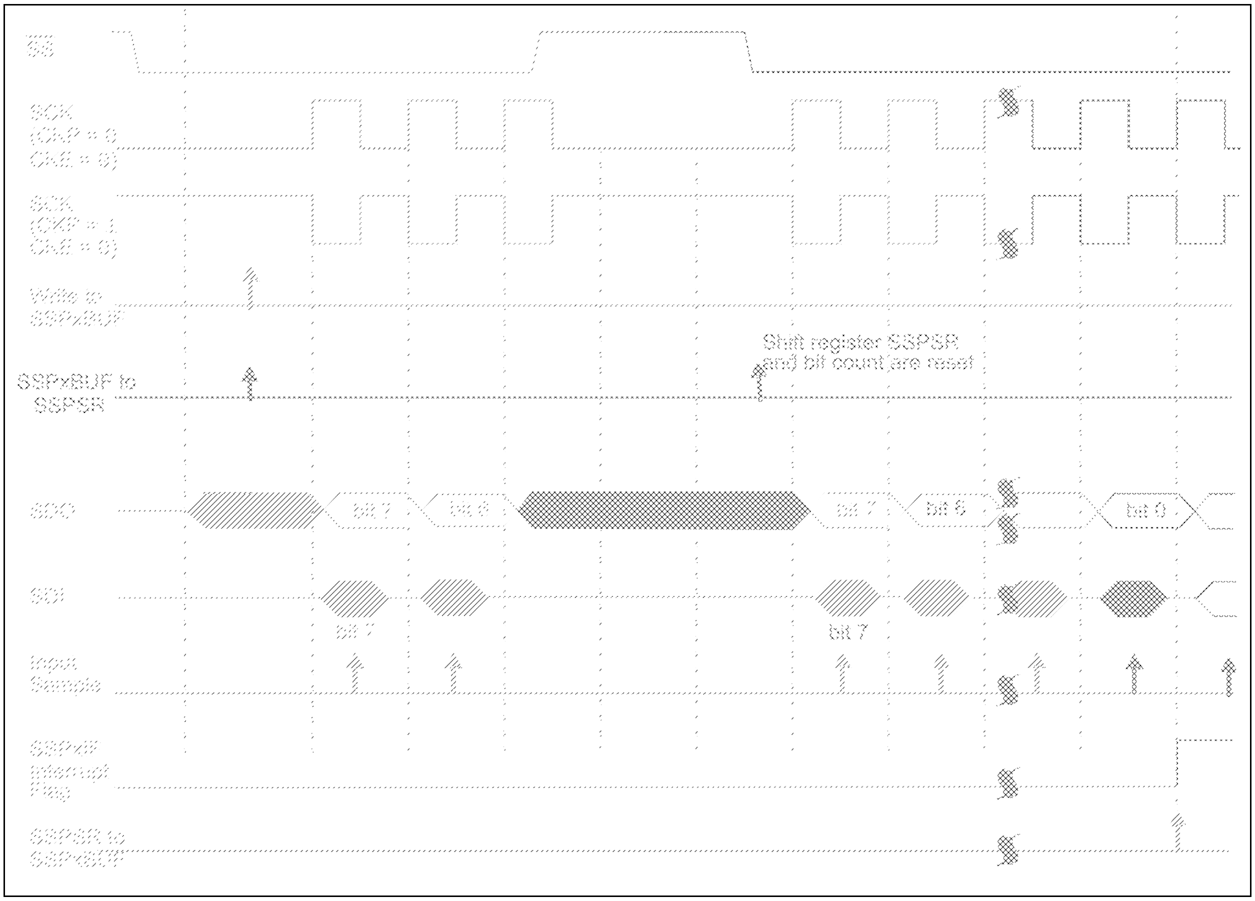


FIGURE 32-14: I²C SLAVE, 7-BIT ADDRESS, RECEPTION (SEN = 0, AHEN = 0, DHEN = 0)

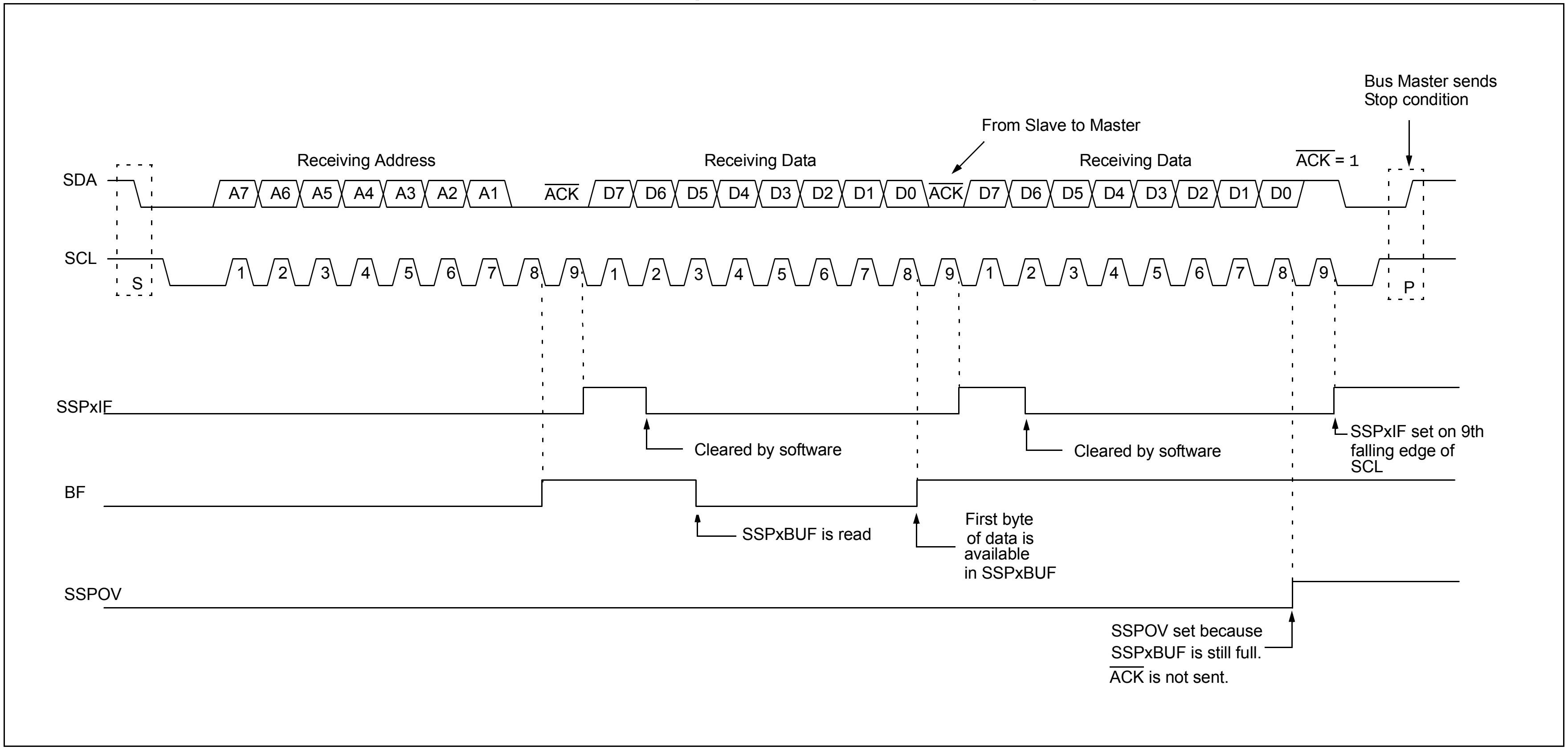
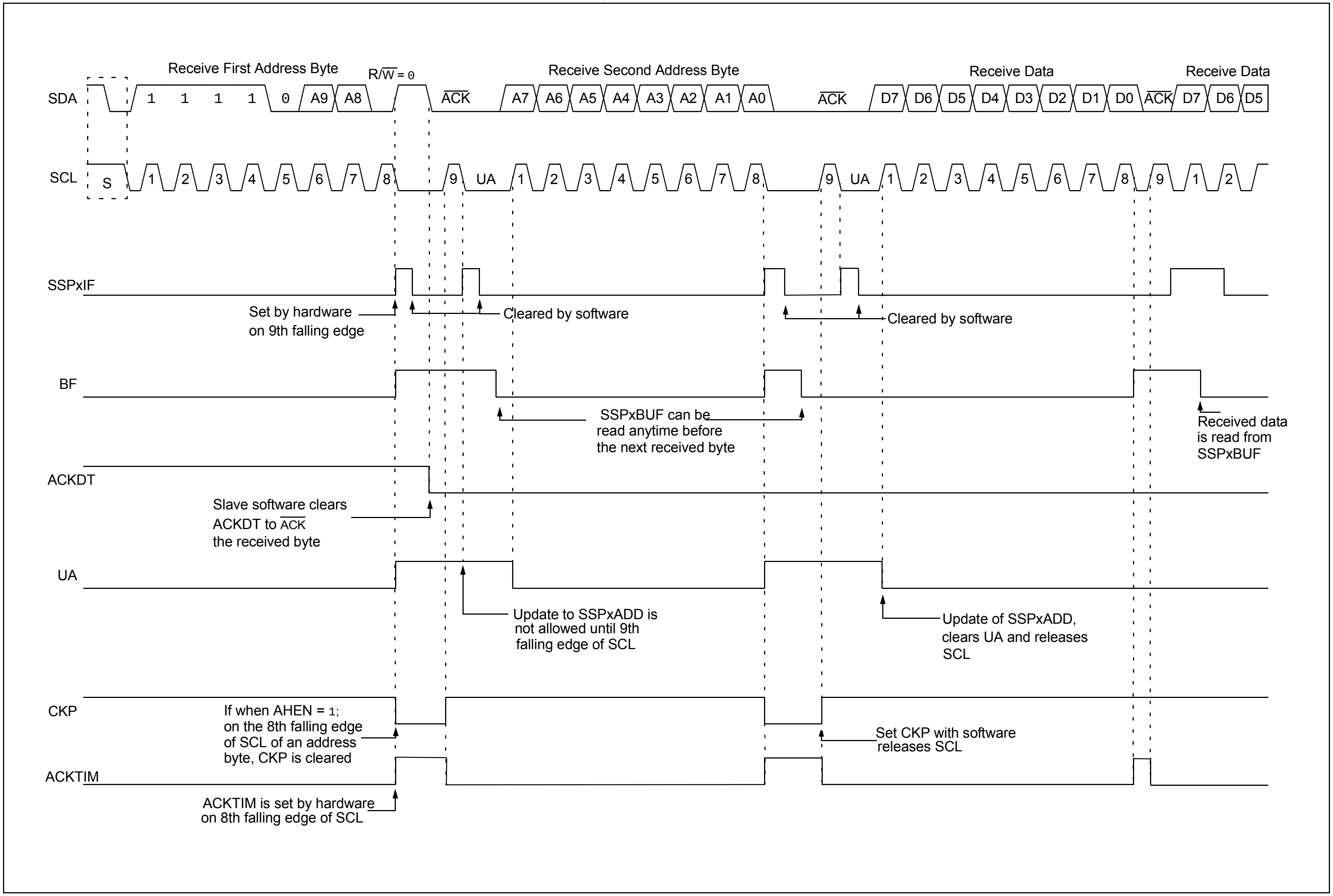


FIGURE 32-21: I²C SLAVE, 10-BIT ADDRESS, RECEPTION (SEN = 0, AHEN = 1, DHEN = 0)



33.1 EUSART Asynchronous Mode

The EUSART transmits and receives data using the standard non-return-to-zero (NRZ) format. NRZ is implemented with two levels: a VOH Mark state which represents a '1' data bit, and a VOL Space state which represents a '0' data bit. NRZ refers to the fact that consecutively transmitted data bits of the same value stay at the output level of that bit without returning to a neutral level between each bit transmission. An NRZ transmission port idles in the Mark state. Each character transmission consists of one Start bit followed by eight or nine data bits and is always terminated by one or more Stop bits. The Start bit is always a space and the Stop bits are always marks. The most common data format is eight bits. Each transmitted bit persists for a period of 1/(baud rate). An on-chip dedicated 8-bit/16-bit Baud Rate Generator is used to derive standard baud rate frequencies from the system oscillator. See Table 33-5 for examples of baud rate configurations.

The EUSART transmits and receives the LSb first. The EUSART's transmitter and receiver are functionally independent, but share the same data format and baud rate. Parity is not supported by the hardware, but can be implemented in software and stored as the ninth data bit.

33.1.1 EUSART ASYNCHRONOUS TRANSMITTER

The EUSART transmitter block diagram is shown in Figure 33-1. The heart of the transmitter is the serial Transmit Shift Register (TSR), which is not directly accessible by software. The TSR obtains its data from the transmit buffer, which is the TXxREG register.

33.1.1.1 Enabling the Transmitter

The EUSART transmitter is enabled for asynchronous operations by configuring the following three control bits:

- TXEN = 1
- SYNC = 0
- SPEN = 1

All other EUSART control bits are assumed to be in their default state.

Setting the TXEN bit of the TXxSTA register enables the transmitter circuitry of the EUSART. Clearing the SYNC bit of the TXxSTA register configures the EUSART for asynchronous operation. Setting the SPEN bit of the RCxSTA register enables the EUSART and automatically configures the TX/CK I/O pin as an output. If the TX/CK pin is shared with an analog peripheral, the analog I/O function must be disabled by clearing the corresponding ANSEL bit.

Note:	The TXIF Transmitter Interrupt flag is set when the TXEN enable bit is set.

33.1.1.2 Transmitting Data

A transmission is initiated by writing a character to the TXxREG register. If this is the first character, or the previous character has been completely flushed from the TSR, the data in the TXxREG is immediately transferred to the TSR register. If the TSR still contains all or part of a previous character, the new character data is held in the TXxREG until the Stop bit of the previous character has been transmitted. The pending character in the TXxREG is then transferred to the TSR in one TCY immediately following the Stop bit transmission. The transmission of the Start bit, data bits and Stop bit sequence commences immediately following the transfer of the data to the TSR from the TXxREG.

33.1.1.3 Transmit Data Polarity

The polarity of the transmit data can be controlled with the SCKP bit of the BAUDxCON register. The default state of this bit is '0', which selects high true transmit idle and data bits. Setting the SCKP bit to '1' will invert the transmit data resulting in low true idle and data bits. The SCKP bit controls transmit data polarity in Asynchronous mode only. In Synchronous mode, the SCKP bit has a different function. See **Section 33.5.1.2 "Clock Polarity"**.

33.1.1.4 Transmit Interrupt Flag

The TXIF interrupt flag bit of the PIR1 register is set whenever the EUSART transmitter is enabled and no character is being held for transmission in the TXxREG. In other words, the TXIF bit is only clear when the TSR is busy with a character and a new character has been queued for transmission in the TXxREG. The TXIF flag bit is not cleared immediately upon writing TXxREG. TXIF becomes valid in the second instruction cycle following the write execution. Polling TXIF immediately following the TXxREG write will return invalid results. The TXIF bit is read-only, it cannot be set or cleared by software.

The TXIF interrupt can be enabled by setting the TXIE interrupt enable bit of the PIE1 register. However, the TXIF flag bit will be set whenever the TXxREG is empty, regardless of the state of the TXIE enable bit.

To use interrupts when transmitting data, set the TXIE bit only when there is more data to send. Clear the TXIE interrupt enable bit upon writing the last character of the transmission to the TXxREG.

35.2 Instruction Descriptions

ADDFSR	Add Literal to FSRn
Syntax:	[*label*] ADDFSR FSRn, k
Operands:	$-32 \le k \le 31$ $n \in [0, 1]$
Operation:	$FSR(n) + k \rightarrow FSR(n)$
Status Affected:	None
Description:	The signed 6-bit literal 'k' is added to the contents of the FSRnH:FSRnL register pair. FSRn is limited to the range 0000h-FFFFh. Moving beyond these bounds will cause the FSR to wrap-around.

ADDLW	Add literal and W
Syntax:	[*label*] ADDLW k
Operands:	$0 \le k \le 255$
Operation:	$(W) + k \rightarrow (W)$
Status Affected:	C, DC, Z
Description:	The contents of the W register are added to the 8-bit literal 'k' and the result is placed in the W register.

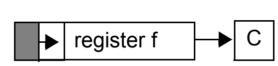
ADDWF	Add W and f
Syntax:	[*label*] ADDWF f,d
Operands:	$0 \le f \le 127$ $d \in [0,1]$
Operation:	(W) + (f) $\rightarrow$ (destination)
Status Affected:	C, DC, Z
Description:	Add the contents of the W register with register 'f'. If 'd' is '0', the result is stored in the W register. If 'd' is '1', the result is stored back in register 'f'.

ADDWFC	ADD W and CARRY bit to f
Syntax:	[*label*] ADDWFC f {,d}
Operands:	$0 \le f \le 127$ $d \in [0,1]$
Operation:	(W) + (f) + (C) $\rightarrow$ dest
Status Affected:	C, DC, Z
Description:	Add W, the Carry flag and data memory location 'f'. If 'd' is '0', the result is placed in W. If 'd' is '1', the result is placed in data memory location 'f'.

ANDLW	AND literal with W
Syntax:	[*label*] ANDLW k
Operands:	$0 \le k \le 255$
Operation:	(W) .AND. (k) $\rightarrow$ (W)
Status Affected:	Z
Description:	The contents of W register are AND'ed with the 8-bit literal 'k'. The result is placed in the W register.

ANDWF	AND W with f
Syntax:	[*label*] ANDWF f,d
Operands:	$0 \le f \le 127$ $d \in [0,1]$
Operation:	(W) .AND. (f) $\rightarrow$ (destination)
Status Affected:	Z
Description:	AND the W register with register 'f'. If 'd' is '0', the result is stored in the W register. If 'd' is '1', the result is stored back in register 'f'.

ASRF	Arithmetic Right Shift
Syntax:	[*label*] ASRF f {,d}
Operands:	$0 \le f \le 127$ $d \in [0,1]$
Operation:	(f<7>) $\rightarrow$ dest<7> (f<7:1>) $\rightarrow$ dest<6:0>, (f<0>) $\rightarrow$ C,
Status Affected:	C, Z
Description:	The contents of register 'f' are shifted one bit to the right through the Carry flag. The MSb remains unchanged. If 'd' is '0', the result is placed in W. If 'd' is '1', the result is stored back in register 'f'.



RETFIE — Return from Interrupt

Syntax:	[*label*] RETFIE k
Operands:	None
Operation:	TOS → PC, 1 → GIE
Status Affected:	None
Description:	Return from Interrupt. Stack is POPed and Top-of-Stack (TOS) is loaded in the PC. Interrupts are enabled by setting Global Interrupt Enable bit, GIE (INTCON<7>). This is a 2-cycle instruction.
Words:	1
Cycles:	2

<u>Example:</u>

```
RETFIE
```

After Interrupt
PC = TOS
GIE = 1

RETLW — Return with literal in W

Syntax:	[*label*] RETLW k
Operands:	$0 \le k \le 255$
Operation:	k → (W); TOS → PC
Status Affected:	None
Description:	The W register is loaded with the 8-bit literal ‘k’. The program counter is loaded from the top of the stack (the return address). This is a 2-cycle instruction.
Words:	1
Cycles:	2

<u>Example:</u>

```
      CALL TABLE;W contains table
            ;offset value
      •     ;W now has table value
TABLE •
      •
      ADDWF PC  ;W = offset
      RETLW k1  ;Begin table
      RETLW k2  ;
      •
      •
      •
      RETLW kn  ; End of table
```

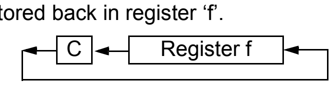
Before Instruction
W = 0x07
After Instruction
W = value of k8

RETURN — Return from Subroutine

Syntax:	[*label*] RETURN
Operands:	None
Operation:	TOS → PC
Status Affected:	None
Description:	Return from subroutine. The stack is POPed and the top of the stack (TOS) is loaded into the program counter. This is a 2-cycle instruction.

RLF — Rotate Left f through Carry

Syntax:	[*label*] RLF f,d
Operands:	$0 \le f \le 127$ $d \in [0,1]$
Operation:	See description below
Status Affected:	C
Description:	The contents of register ‘f’ are rotated one bit to the left through the Carry flag. If ‘d’ is ‘0’, the result is placed in the W register. If ‘d’ is ‘1’, the result is stored back in register ‘f’.



Words:	1
Cycles:	1

<u>Example:</u>

```
RLF    REG1,0
```

Before Instruction
REG1 = 1110 0110
C = 0
After Instruction
REG1 = 1110 0110
W = 1100 1100
C = 1

FIGURE 36-19: SPI SLAVE MODE TIMING (CKE = 0)

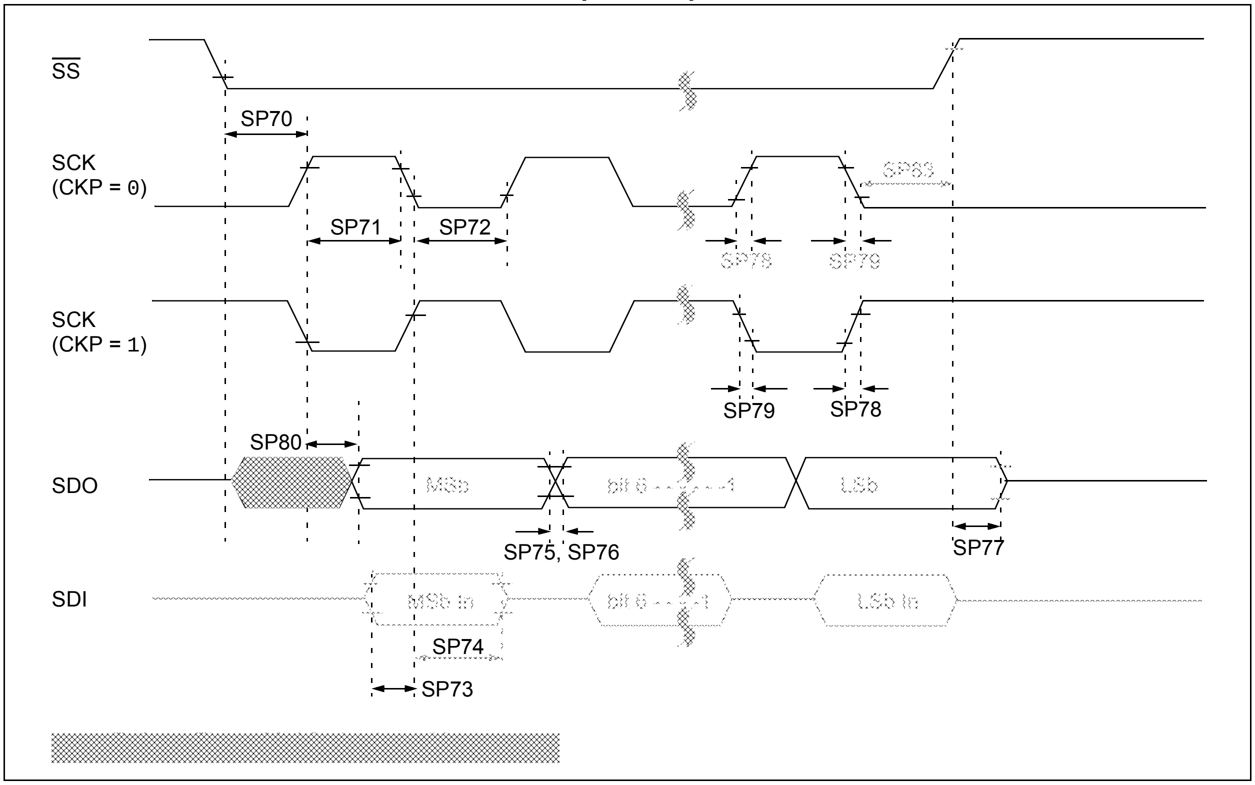
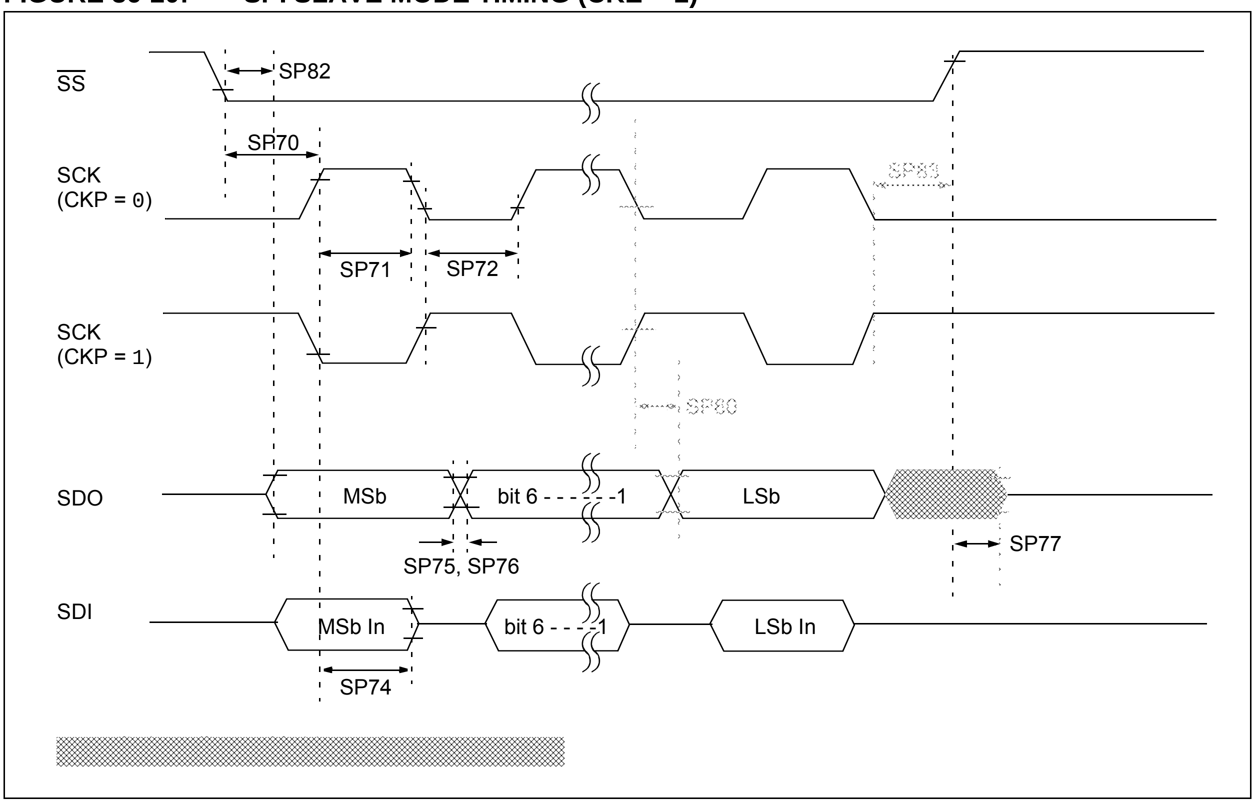


FIGURE 36-20: SPI SLAVE MODE TIMING (CKE = 1)



Note the following details of the code protection feature on Microchip devices:

- Microchip products meet the specification contained in their particular Microchip Data Sheet.
- Microchip believes that its family of products is one of the most secure families of its kind on the market today, when used in the intended manner and under normal conditions.
- There are dishonest and possibly illegal methods used to breach the code protection feature. All of these methods, to our knowledge, require using the Microchip products in a manner outside the operating specifications contained in Microchip's Data Sheets. Most likely, the person doing so is engaged in theft of intellectual property.
- Microchip is willing to work with the customer who is concerned about the integrity of their code.
- Neither Microchip nor any other semiconductor manufacturer can guarantee the security of their code. Code protection does not mean that we are guaranteeing the product as "unbreakable."

Code protection is constantly evolving. We at Microchip are committed to continuously improving the code protection features of our products. Attempts to break Microchip's code protection feature may be a violation of the Digital Millennium Copyright Act. If such acts allow unauthorized access to your software or other copyrighted work, you may have a right to sue for relief under that Act.

Microchip received ISO/TS-16949:2009 certification for its worldwide headquarters, design and wafer fabrication facilities in Chandler and Tempe, Arizona; Gresham, Oregon and design centers in California and India. The Company's quality system processes and procedures are for its PIC® MCUs and dsPIC® DSCs, KEELOQ® code hopping devices, Serial EEPROMs, microperipherals, nonvolatile memory and analog products. In addition, Microchip's quality system for the design and manufacture of development systems is ISO 9001:2000 certified.

QUALITY MANAGEMENT SYSTEM CERTIFIED BY DNV

== ISO/TS 16949 ==

Trademarks

ISBN: 978-1-5224-1016-4