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Understanding Embedded - FPGAs (Field Programmable Gate Array)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications,

Details	
Product Status	Active
Number of LABs/CLBs	600
Number of Logic Elements/Cells	2700
Total RAM Bits	40960
Number of I/O	92
Number of Gates	100000
Voltage - Supply	2.375V ~ 2.625V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 85°C (TJ)
Package / Case	144-LQFP
Supplier Device Package	144-TQFP (20x20)
Purchase URL	https://www.e-xfl.com/product-detail/xilinx/xc2s100-6tq144c

The three IOB registers function either as edge-triggered D-type flip-flops or as level-sensitive latches. Each IOB has a clock signal (CLK) shared by the three registers and independent Clock Enable (CE) signals for each register. In addition to the CLK and CE control signals, the three registers share a Set/Reset (SR). For each register, this signal can be independently configured as a synchronous Set, a synchronous Reset, an asynchronous Preset, or an asynchronous Clear.

A feature not shown in the block diagram, but controlled by the software, is polarity control. The input and output buffers and all of the IOB control signals have independent polarity controls.

Optional pull-up and pull-down resistors and an optional weak-keeper circuit are attached to each pad. Prior to configuration all outputs not involved in configuration are forced into their high-impedance state. The pull-down resistors and the weak-keeper circuits are inactive, but inputs may optionally be pulled up.

Table 3: **Standards Supported by I/O (Typical Values)**

I/O Standard	Input Reference Voltage (V_{REF})	Output Source Voltage (V_{CCO})	Board Termination Voltage (V_{TT})
LVTTL (2-24 mA)	N/A	3.3	N/A
LVCMOS2	N/A	2.5	N/A
PCI (3V/5V, 33 MHz/66 MHz)	N/A	3.3	N/A
GTL	0.8	N/A	1.2
GTL+	1.0	N/A	1.5
HSTL Class I	0.75	1.5	0.75
HSTL Class III	0.9	1.5	1.5
HSTL Class IV	0.9	1.5	1.5
SSTL3 Class I and II	1.5	3.3	1.5
SSTL2 Class I and II	1.25	2.5	1.25
CTT	1.5	3.3	1.5
AGP-2X	1.32	3.3	N/A

The activation of pull-up resistors prior to configuration is controlled on a global basis by the configuration mode pins. If the pull-up resistors are not activated, all the pins will float. Consequently, external pull-up or pull-down resistors must be provided on pins required to be at a well-defined logic level prior to configuration.

All pads are protected against damage from electrostatic discharge (ESD) and from over-voltage transients. Two forms of over-voltage protection are provided, one that permits 5V compliance, and one that does not. For 5V compliance, a zener-like structure connected to ground turns on when the output rises to approximately 6.5V. When 5V compliance is not required, a conventional clamp diode may be connected to the output supply voltage, V_{CCO}. The type of over-voltage protection can be selected independently for each pad.

All Spartan-II FPGA IOBs support IEEE 1149.1-compatible boundary scan testing.

Input Path

A buffer In the Spartan-II FPGA IOB input path routes the input signal either directly to internal logic or through an optional input flip-flop.

An optional delay element at the D-input of this flip-flop eliminates pad-to-pad hold time. The delay is matched to the internal clock-distribution delay of the FPGA, and when used, assures that the pad-to-pad hold time is zero.

Each input buffer can be configured to conform to any of the low-voltage signaling standards supported. In some of these standards the input buffer utilizes a user-supplied threshold voltage, V_{REF}. The need to supply V_{REF} imposes constraints on which standards can used in close proximity to each other. See "I/O Banking," page 9.

There are optional pull-up and pull-down resistors at each input for use after configuration.

Output Path

The output path includes a 3-state output buffer that drives the output signal onto the pad. The output signal can be routed to the buffer directly from the internal logic or through an optional IOB output flip-flop.

The 3-state control of the output can also be routed directly from the internal logic or through a flip-flip that provides synchronous enable and disable.

Each output driver can be individually programmed for a wide range of low-voltage signaling standards. Each output buffer can source up to 24 mA and sink up to 48 mA. Drive strength and slew rate controls minimize bus transients.

In most signaling standards, the output high voltage depends on an externally supplied V_{CCO} voltage. The need to supply V_{CCO} imposes constraints on which standards can be used in close proximity to each other. See "I/O Banking".

An optional weak-keeper circuit is connected to each output. When selected, the circuit monitors the voltage on the pad and weakly drives the pin High or Low to match the input signal. If the pin is connected to a multiple-source signal, the weak keeper holds the signal in its last state if all

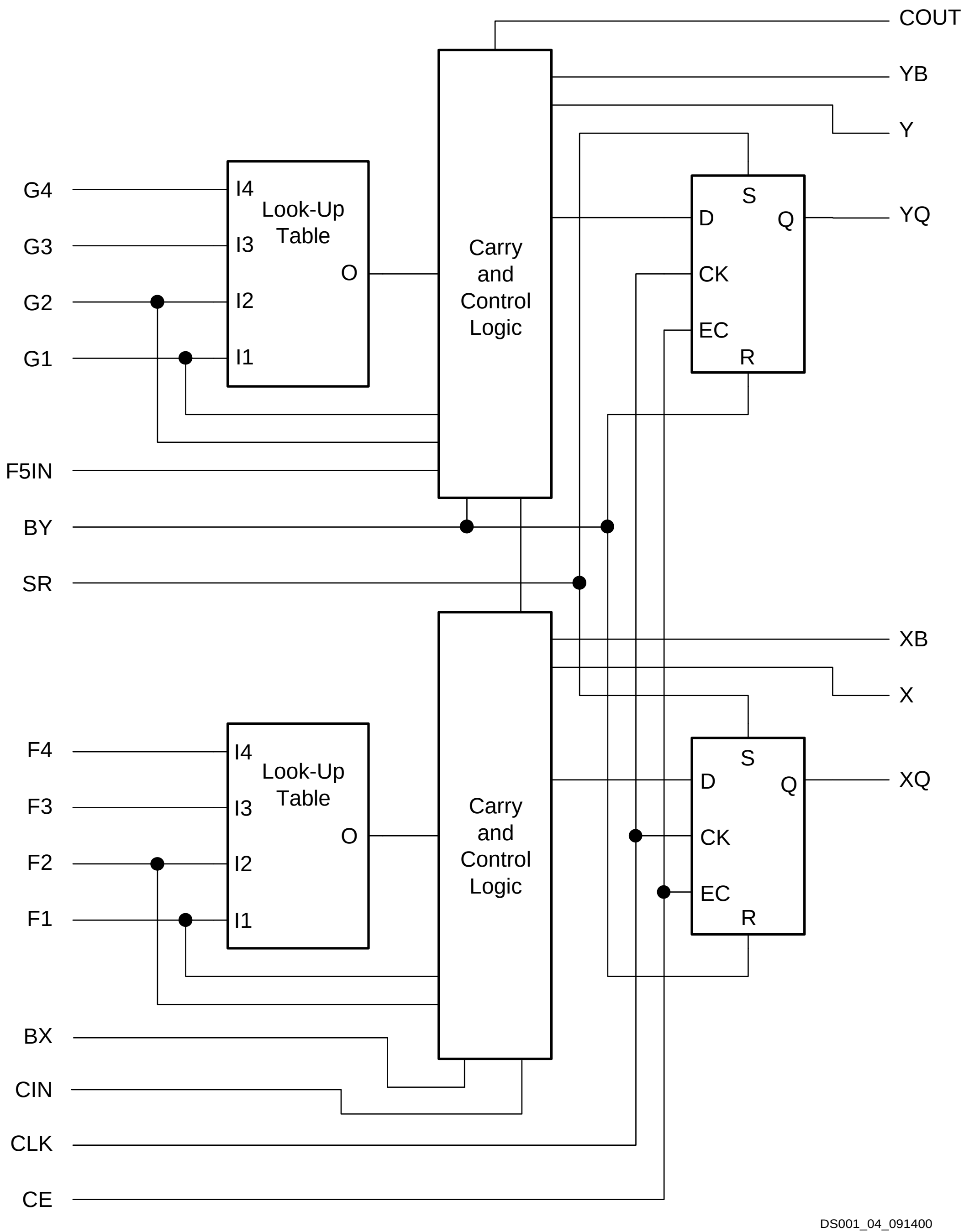

Figure 4: **Spartan-II CLB Slice** (two identical slices in each CLB)

Storage Elements

Storage elements in the Spartan-II FPGA slice can be configured either as edge-triggered D-type flip-flops or as level-sensitive latches. The D inputs can be driven either by function generators within the slice or directly from slice inputs, bypassing the function generators.

In addition to Clock and Clock Enable signals, each slice has synchronous set and reset signals (SR and BY). SR forces a storage element into the initialization state specified for it in the configuration. BY forces it into the opposite state. Alternatively, these signals may be configured to operate asynchronously.

All control signals are independently invertible, and are shared by the two flip-flops within the slice.

Additional Logic

The F5 multiplexer in each slice combines the function generator outputs. This combination provides either a function generator that can implement any 5-input function, a 4:1 multiplexer, or selected functions of up to nine inputs.

Signals

There are two kinds of pins that are used to configure Spartan-II devices: Dedicated pins perform only specific configuration-related functions; the other pins can serve as general purpose I/Os once user operation has begun.

The dedicated pins comprise the mode pins (M2, M1, M0), the configuration clock pin (CCLK), the PROGRAM pin, the DONE pin and the boundary-scan pins (TDI, TDO, TMS, TCK). Depending on the selected configuration mode, CCLK may be an output generated by the FPGA, or may be generated externally, and provided to the FPGA as an input.

Note that some configuration pins can act as outputs. For correct operation, these pins require a V_{CCO} of 3.3V to drive an LVTTL signal or 2.5V to drive an LVCMOS signal. All the relevant pins fall in banks 2 or 3. The CS and WRITE pins for Slave Parallel mode are located in bank 1.

For a more detailed description than that given below, see "Pinout Tables" in Module 4 and XAPP176, *Spartan-II FPGA Series Configuration and Readback*.

The Process

The sequence of steps necessary to configure Spartan-II devices are shown in Figure 11. The overall flow can be divided into three different phases.

- Initiating Configuration
- Configuration memory clear
- Loading data frames
- Start-up

The memory clearing and start-up phases are the same for all configuration modes; however, the steps for the loading of data frames are different. Thus, the details for data frame loading are described separately in the sections devoted to each mode.

Initiating Configuration

There are two different ways to initiate the configuration process: applying power to the device or asserting the PROGRAM input.

Configuration on power-up occurs automatically unless it is delayed by the user, as described in a separate section below. The waveform for configuration on power-up is shown in Figure 12, page 19. Before configuration can begin, V_{CCO} Bank 2 must be greater than 1.0V. Furthermore, all V_{CCINT} power pins must be connected to a 2.5V supply. For more information on delaying configuration, see "Clearing Configuration Memory," page 19.

Once in user operation, the device can be re-configured simply by pulling the PROGRAM pin Low. The device acknowledges the beginning of the configuration process by driving DONE Low, then enters the memory-clearing phase.

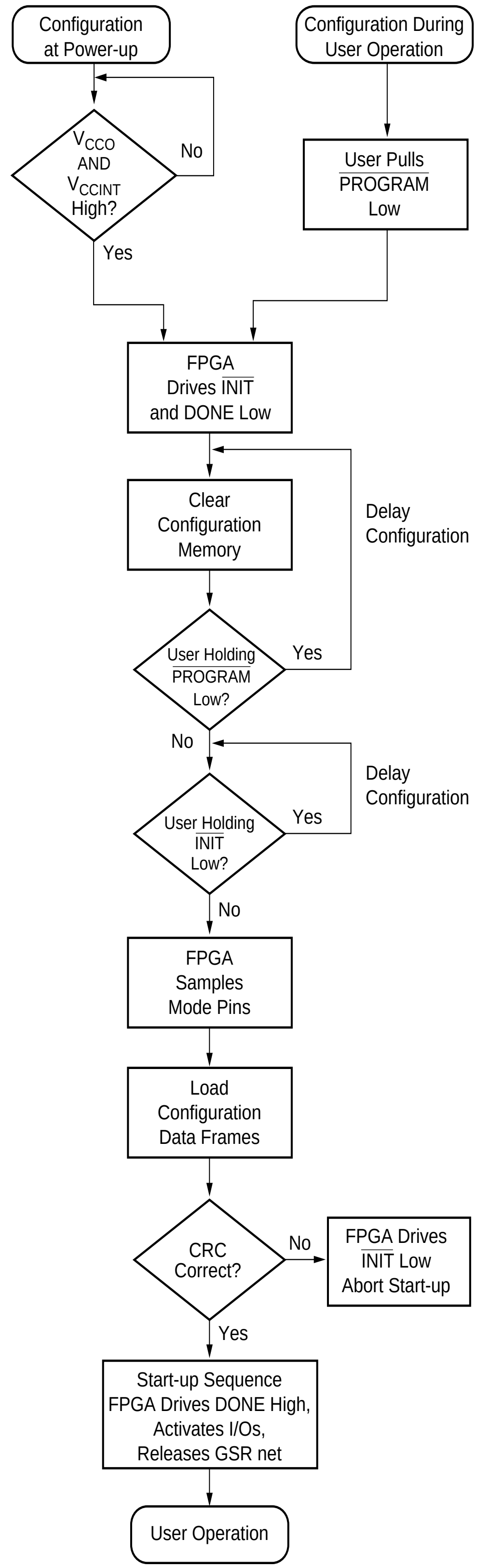

Figure 11: **Configuration Flow Diagram**

If CCLK is slower than F_{CCNH}, the FPGA will never assert BUSY. In this case, the above handshake is unnecessary, and data can simply be entered into the FPGA every CCLK cycle.

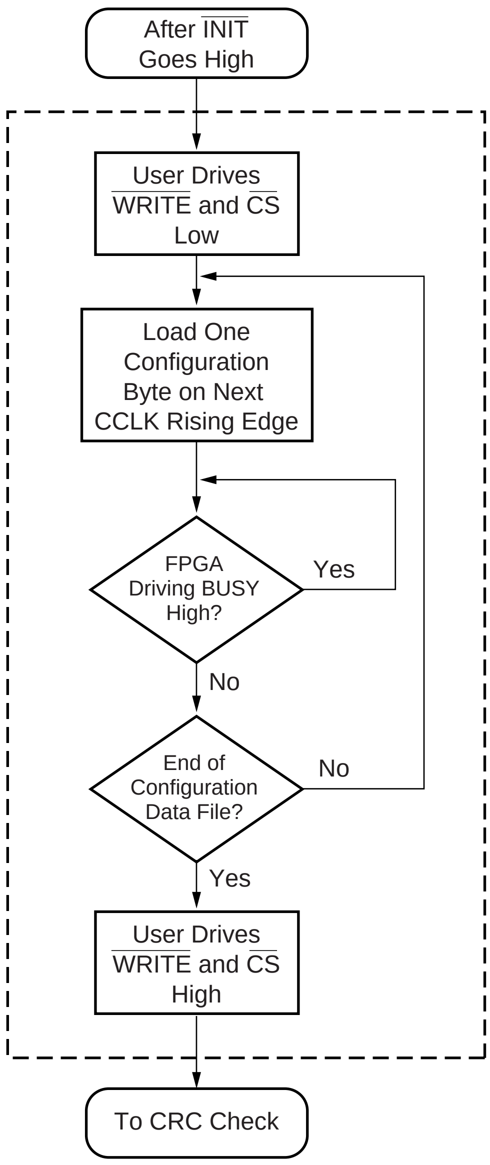

Figure 19: **Loading Configuration Data for the Slave Parallel Mode**

A configuration packet does not have to be written in one continuous stretch, rather it can be split into many write sequences. Each sequence would involve assertion of CS.

In applications where multiple clock cycles may be required to access the configuration data before each byte can be loaded into the Slave Parallel interface, a new byte of data may not be ready for each consecutive CCLK edge. In such a case the CS signal may be de-asserted until the next byte is valid on D0-D7. While CS is High, the Slave Parallel interface does not expect any data and ignores all CCLK transitions. However, to avoid aborting configuration, WRITE must continue to be asserted while CS is asserted.

Abort

To abort configuration during a write sequence, de-assert WRITE while holding CS Low. The abort operation is initiated at the rising edge of CCLK, as shown in Figure 21, page 26. The device will remain BUSY until the aborted operation is complete. After aborting configuration, data is assumed to be unaligned to word boundaries and the FPGA requires a new synchronization word prior to accepting any new packets.

Boundary-Scan Mode

In the boundary-scan mode, no nondedicated pins are required, configuration being done entirely through the IEEE 1149.1 Test Access Port.

Configuration through the TAP uses the special CFG_IN instruction. This instruction allows data input on TDI to be converted into data packets for the internal configuration bus.

The following steps are required to configure the FPGA through the boundary-scan port.

1. Load the CFG_IN instruction into the boundary-scan instruction register (IR)
2. Enter the Shift-DR (SDR) state
3. Shift a standard configuration bitstream into TDI
4. Return to Run-Test-Idle (RTI)
5. Load the JSTART instruction into IR
6. Enter the SDR state
7. Clock TCK through the sequence (the length is programmable)
8. Return to RTI

Configuration and readback via the TAP is always available. The boundary-scan mode simply locks out the other modes. The boundary-scan mode is selected by a <10x> on the mode pins (M0, M1, M2).

Readback

The configuration data stored in the Spartan-II FPGA configuration memory can be readback for verification. Along with the configuration data it is possible to readback the contents of all flip-flops/latches, LUT RAMs, and block RAMs. This capability is used for real-time debugging.

For more detailed information see XAPP176, *Spartan-II FPGA Family Configuration and Readback.*

Using Block RAM Features

The Spartan-II FPGA family provides dedicated blocks of on-chip, true dual-read/write port synchronous RAM, with 4096 memory cells. Each port of the block RAM memory can be independently configured as a read/write port, a read port, a write port, and can be configured to a specific data width. The block RAM memory offers new capabilities allowing the FPGA designer to simplify designs.

Operating Modes

Block RAM memory supports two operating modes.

- Read Through
- Write Back

Read Through (One Clock Edge)

The read address is registered on the read port clock edge and data appears on the output after the RAM access time. Some memories may place the latch/register at the outputs depending on the desire to have a faster clock-to-out versus setup time. This is generally considered to be an inferior solution since it changes the read operation to an asynchronous function with the possibility of missing an address/control line transition during the generation of the read pulse clock.

Write Back (One Clock Edge)

The write address is registered on the write port clock edge and the data input is written to the memory and mirrored on the write port input.

Block RAM Characteristics

1. All inputs are registered with the port clock and have a setup to clock timing specification.
2. All outputs have a read through or write back function depending on the state of the port WE pin. The outputs relative to the port clock are available after the clock-to-out timing specification.
3. The block RAM are true SRAM memories and do not have a combinatorial path from the address to the output. The LUT cells in the CLBs are still available with this function.
4. The ports are completely independent from each other (*i.e.,* clocking, control, address, read/write function, and data width) without arbitration.
5. A write operation requires only one clock edge.
6. A read operation requires only one clock edge.

The output ports are latched with a self timed circuit to guarantee a glitch free read. The state of the output port will not change until the port executes another read or write operation.

Library Primitives

Figure 31 and Figure 32 show the two generic library block RAM primitives. Table 11 describes all of the available primitives for synthesis and simulation.

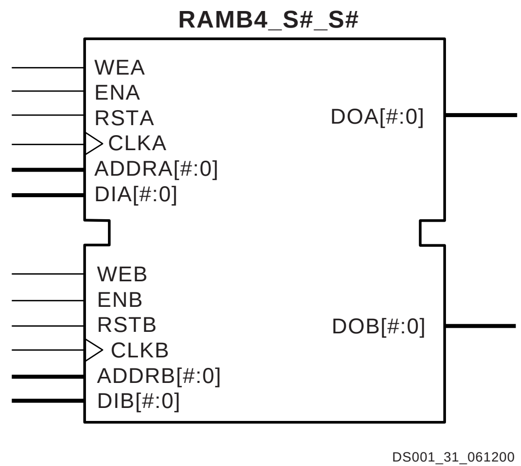

Figure 31: **Dual-Port Block RAM Memory**

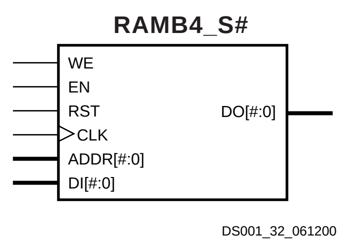

Figure 32: **Single-Port Block RAM Memory**

Table 11: **Available Library Primitives**

Primitive	Port A Width	Port B Width
RAMB4_S1	1	N/A
RAMB4_S1_S1		1
RAMB4_S1_S2		2
RAMB4_S1_S4		4
RAMB4_S1_S8		8
RAMB4_S1_S16		16
RAMB4_S2	2	N/A
RAMB4_S2_S2		2
RAMB4_S2_S4		4
RAMB4_S2_S8		8
RAMB4_S2_S16		16

At the third rising edge of CLKA, the T_{BCCS} parameter is violated with two writes to memory location 0x0F. The DOA and DOB busses reflect the contents of the DIA and DIB busses, but the stored value at 0x7E is invalid.

At the fourth rising edge of CLKA, a read operation is performed at memory location 0x0F and invalid data is present on the DOA bus. Port B also executes a read operation to memory location 0x0F and also reads invalid data.

At the fifth rising edge of CLKA a read operation is performed that does not violate the T_{BCCS} parameter to the previous write of 0x7E by Port B. THe DOA bus reflects the recently written value by Port B.

Initialization

The block RAM memory can initialize during the device configuration sequence. The 16 initialization properties of 64 hex values each (a total of 4096 bits) set the initialization of each RAM. These properties appear in Table 14. Any initialization properties not explicitly set configure as zeros. Partial initialization strings pad with zeros. Initialization strings greater than 64 hex values generate an error. The RAMs can be simulated with the initialization values using generics in VHDL simulators and parameters in Verilog simulators.

Initialization in VHDL

The block RAM structures may be initialized in VHDL for both simulation and synthesis for inclusion in the EDIF output file. The simulation of the VHDL code uses a generic to pass the initialization.

Initialization in Verilog

The block RAM structures may be initialized in Verilog for both simulation and synthesis for inclusion in the EDIF output file. The simulation of the Verilog code uses a defparam to pass the initialization.

Block Memory Generation

The CORE Generator™ software generates memory structures using the block RAM features. This program outputs VHDL or Verilog simulation code templates and an EDIF file for inclusion in a design.

Table 14: **RAM Initialization Properties**

Property	Memory Cells
INIT_00	255 to 0
INIT_01	511 to 256
INIT_02	767 to 512
INIT_03	1023 to 768
INIT_04	1279 to 1024
INIT_05	1535 to 1280
INIT_06	1791 to 1536
INIT_07	2047 to 1792
INIT_08	2303 to 2048
INIT_09	2559 to 2304
INIT_0a	2815 to 2560
INIT_0b	3071 to 2816
INIT_0c	3327 to 3072
INIT_0d	3583 to 3328
INIT_0e	3839 to 3584
INIT_0f	4095 to 3840

Table 14: **RAM Initialization Properties**

For design examples and more information on using the Block RAM, see XAPP173, *Using Block SelectRAM+ Memory in Spartan-II FPGAs*.

Using Versatile I/O

The Spartan-II FPGA family includes a highly configurable, high-performance I/O resource called Versatile I/O to provide support for a wide variety of I/O standards. The Versatile I/O resource is a robust set of features including programmable control of output drive strength, slew rate, and input delay and hold time. Taking advantage of the flexibility and Versatile I/O features and the design considerations described in this document can improve and simplify system level design.

Introduction

As FPGAs continue to grow in size and capacity, the larger and more complex systems designed for them demand an increased variety of I/O standards. Furthermore, as system clock speeds continue to increase, the need for high-performance I/O becomes more important. While chip-to-chip delays have an increasingly substantial impact on overall system speed, the task of achieving the desired system performance becomes more difficult with the proliferation of low-voltage I/O standards. Versatile I/O, the revolutionary input/output resources of Spartan-II devices, has resolved this potential problem by providing a highly configurable, high-performance alternative to the I/O resources of more conventional programmable devices. The Spartan-II FPGA Versatile I/O features combine the flexibility and time-to-market advantages of programmable logic with the high performance previously available only with ASICs and custom ICs.

Each Versatile I/O block can support up to 16 I/O standards. Supporting such a variety of I/O standards allows the support of a wide variety of applications, from general purpose standard applications to high-speed low-voltage memory busses.

Versatile I/O blocks also provide selectable output drive strengths and programmable slew rates for the LVTTL output buffers, as well as an optional, programmable weak pull-up, weak pull-down, or weak "keeper" circuit ideal for use in external bussing applications.

Each Input/Output Block (IOB) includes three registers, one each for the input, output, and 3-state signals within the IOB. These registers are optionally configurable as either a D-type flip-flop or as a level sensitive latch.

The input buffer has an optional delay element used to guarantee a zero hold time requirement for input signals registered within the IOB.

The Versatile I/O features also provide dedicated resources for input reference voltage (V_{REF}) and output source voltage (V_{CCO}), along with a convenient banking system that simplifies board design.

By taking advantage of the built-in features and wide variety of I/O standards supported by the Versatile I/O features, system-level design and board design can be greatly simplified and improved.

Fundamentals

Modern bus applications, pioneered by the largest and most influential companies in the digital electronics industry, are commonly introduced with a new I/O standard tailored specifically to the needs of that application. The bus I/O standards provide specifications to other vendors who create products designed to interface with these applications. Each standard often has its own specifications for current, voltage, I/O buffering, and termination techniques.

The ability to provide the flexibility and time-to-market advantages of programmable logic is increasingly dependent on the capability of the programmable logic device to support an ever increasing variety of I/O standards

The Versatile I/O resources feature highly configurable input and output buffers which provide support for a wide variety of I/O standards. As shown in Table 15, each buffer type can support a variety of voltage requirements.

Table 15: **Versatile I/O Supported Standards (Typical Values)**

I/O Standard	Input Reference Voltage (V_{REF})	Output Source Voltage (V_{CCO})	Board Termination Voltage (V_{TT})
LVTTL (2-24 mA)	N/A	3.3	N/A
LVCMOS2	N/A	2.5	N/A
PCI (3V/5V, 33 MHz/66 MHz)	N/A	3.3	N/A
GTL	0.8	N/A	1.2
GTL+	1.0	N/A	1.5
HSTL Class I	0.75	1.5	0.75
HSTL Class III	0.9	1.5	1.5
HSTL Class IV	0.9	1.5	1.5
SSTL3 Class I and II	1.5	3.3	1.5
SSTL2 Class I and II	1.25	2.5	1.25
CTT	1.5	3.3	1.5
AGP-2X	1.32	3.3	N/A

Overview of Supported I/O Standards

This section provides a brief overview of the I/O standards supported by all Spartan-II devices.

While most I/O standards specify a range of allowed voltages, this document records typical voltage values only. Detailed information on each specification may be found on the Electronic Industry Alliance JEDEC website at http://www.jedec.org. For more details on the I/O standards and termination application examples, see XAPP179, "*Using SelectIO Interfaces in Spartan-II and Spartan-IIE FPGAs.*"

LVTTL — Low-Voltage TTL

The Low-Voltage TTL (LVTTL) standard is a general purpose EIA/JESDSA standard for 3.3V applications that uses an LVTTL input buffer and a Push-Pull output buffer. This standard requires a 3.3V output source voltage (V_{CCO}), but does not require the use of a reference voltage (V_{REF}) or a termination voltage (V_{TT}).

LVCMOS2 — Low-Voltage CMOS for 2.5V

The Low-Voltage CMOS for 2.5V or lower (LVCMOS2) standard is an extension of the LVCMOS standard (JESD 8.5) used for general purpose 2.5V applications. This standard requires a 2.5V output source voltage (V_{CCO}), but does not require the use of a reference voltage (V_{REF}) or a board termination voltage (V_{TT}).

the LOC property is described below. Table 16 summarizes the input standards compatibility requirements.

An optional delay element is associated with each IBUF. When the IBUF drives a flip-flop within the IOB, the delay element by default activates to ensure a zero hold-time requirement. The NODELAY=TRUE property overrides this default.

When the IBUF does not drive a flip-flop within the IOB, the delay element de-activates by default to provide higher performance. To delay the input signal, activate the delay element with the DELAY=TRUE property.

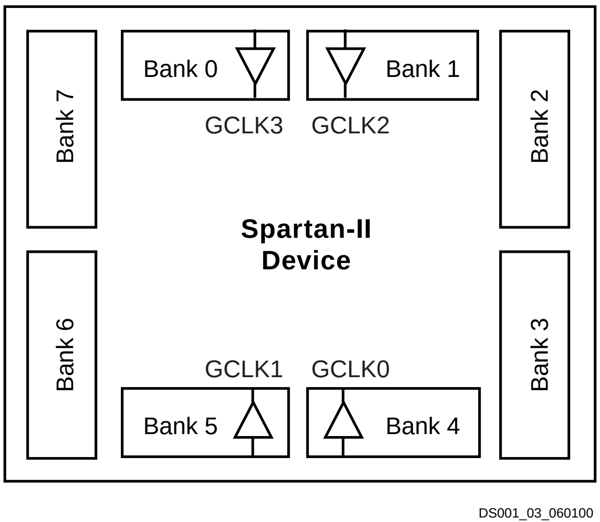

Figure 36: **I/O Banks**

Table 16: **Xilinx Input Standards Compatibility Requirements**

Rule 1	All differential amplifier input signals within a bank are required to be of the same standard.
Rule 2	There are no placement restrictions for inputs with standards that require a single-ended input buffer.

IBUFG

Signals used as high fanout clock inputs to the Spartan-II device should drive a global clock input buffer (IBUFG) via an external input port in order to take advantage of one of the four dedicated global clock distribution networks. The output of the IBUFG primitive can only drive a CLKDLL, CLKDLLHF, or a BUFG primitive. The generic IBUFG primitive appears in Figure 37.

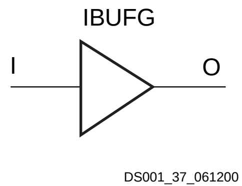

Figure 37: **Global Clock Input Buffer (IBUFG) Primitive**

With no extension or property specified for the generic IBUFG primitive, the assumed standard is LVTTL.

The voltage reference signal is "banked" within the Spartan-II device on a half-edge basis such that for all packages there are eight independent V_{REF} banks internally. See Figure 36 for a representation of the I/O banks. Within each bank approximately one of every six I/O pins is automatically configured as a V_{REF} input.

IBUFG placement restrictions require any differential amplifier input signals within a bank be of the same standard. The LOC property can specify a location for the IBUFG.

As an added convenience, the BUFGP can be used to instantiate a high fanout clock input. The BUFGP primitive represents a combination of the LVTTL IBUFG and BUFG primitives, such that the output of the BUFGP can connect directly to the clock pins throughout the design.

The Spartan-II FPGA BUFGP primitive can only be placed in a global clock pad location. The LOC property can specify a location for the BUFGP.

OBUF

An OBUF must drive outputs through an external output port. The generic output buffer (OBUF) primitive appears in Figure 38.

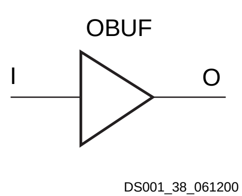

Figure 38: **Output Buffer (OBUF) Primitive**

With no extension or property specified for the generic OBUF primitive, the assumed standard is slew rate limited LVTTL with 12 mA drive strength.

The LVTTL OBUF additionally can support one of two slew rate modes to minimize bus transients. By default, the slew rate for each output buffer is reduced to minimize power bus transients when switching non-critical signals.

SSTL3 Class I

A sample circuit illustrating a valid termination technique for SSTL3_I appears in Figure 47. DC voltage specifications appear in Table 25 for the SSTL3_I standard. See "DC Specifications" in Module 3 for the actual FPGA characteristics.

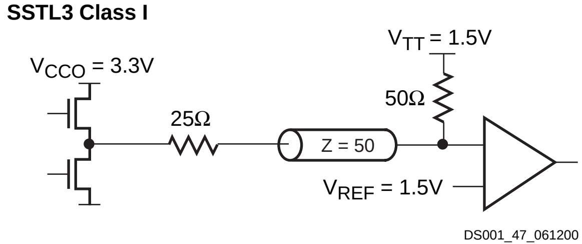

Figure 47: **Terminated SSTL3 Class I**

Table 25: **SSTL3_I Voltage Specifications**

Parameter	Min	Typ	Max
V_{CCO}	3.0	3.3	3.6
$V_{REF} = 0.45 \times V_{CCO}$	1.3	1.5	1.7
$V_{TT} = V_{REF}$	1.3	1.5	1.7
$V_{IH} \geq V_{REF} + 0.2$	1.5	1.7	3.9(1)
$V_{IL} \leq V_{REF} - 0.2$	–0.3(2)	1.3	1.5
$V_{OH} \geq V_{REF} + 0.6$	1.9	-	-
$V_{OL} \leq V_{REF} - 0.6$	-	-	1.1
I_{OH} at V_{OH} (mA)	–8	-	-
I_{OL} at V_{OL} (mA)	8	-	-

Notes:
1. V_{IH} maximum is V_{CCO} + 0.3.
2. V_{IL} minimum does not conform to the formula.

SSTL3 Class II

A sample circuit illustrating a valid termination technique for SSTL3_II appears in Figure 48. DC voltage specifications appear in Table 26 for the SSTL3_II standard. See "DC Specifications" in Module 3 for the actual FPGA characteristics.

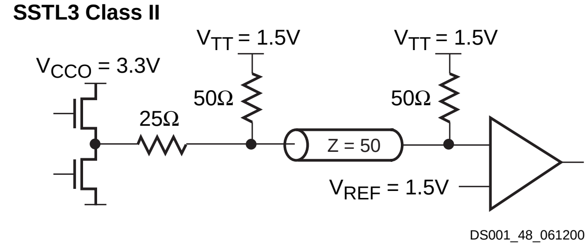

Figure 48: **Terminated SSTL3 Class II**

Table 26: **SSTL3_II Voltage Specifications**

Parameter	Min	Typ	Max
V_{CCO}	3.0	3.3	3.6
$V_{REF} = 0.45 \times V_{CCO}$	1.3	1.5	1.7
$V_{TT} = V_{REF}$	1.3	1.5	1.7
$V_{IH} \geq V_{REF} + 0.2$	1.5	1.7	3.9(1)
$V_{IL} \leq V_{REF} - 0.2$	–0.3(2)	1.3	1.5
$V_{OH} \geq V_{REF} + 0.8$	2.1	-	-
$V_{OL} \leq V_{REF} - 0.8$	-	-	0.9
I_{OH} at V_{OH} (mA)	–16	-	-
I_{OL} at V_{OL} (mA)	16	-	-

Notes:
1. V_{IH} maximum is V_{CCO} + 0.3
2. V_{IL} minimum does not conform to the formula

IOB Input Switching Characteristics(1)

Input delays associated with the pad are specified for LVTTL levels. For other standards, adjust the delays with the values shown in "IOB Input Delay Adjustments for Different Standards," page 57.

Symbol	Description	Device	Speed Grade -6 Min	Speed Grade -6 Max	Speed Grade -5 Min	Speed Grade -5 Max	Units
Propagation Delays							
T_{IOPI}	Pad to I output, no delay	All	-	0.8	-	1.0	ns
T_{IOPID}	Pad to I output, with delay	All	-	1.5	-	1.8	ns
T_{IOPLI}	Pad to output IQ via transparent latch, no delay	All	-	1.7	-	2.0	ns
T_{IOPLID}	Pad to output IQ via transparent latch, with delay	XC2S15	-	3.8	-	4.5	ns
		XC2S30	-	3.8	-	4.5	ns
		XC2S50	-	3.8	-	4.5	ns
		XC2S100	-	3.8	-	4.5	ns
		XC2S150	-	4.0	-	4.7	ns
		XC2S200	-	4.0	-	4.7	ns
Sequential Delays							
T_{IOCKIQ}	Clock CLK to output IQ	All	-	0.7	-	0.8	ns
Setup/Hold Times with Respect to Clock CLK(2)							
T_{IOPICK} / T_{IOICKP}	Pad, no delay	All	1.7 / 0	-	1.9 / 0	-	ns
$T_{IOPICKD}$ / $T_{IOICKPD}$	Pad, with delay(1)	XC2S15	3.8 / 0	-	4.4 / 0	-	ns
		XC2S30	3.8 / 0	-	4.4 / 0	-	ns
		XC2S50	3.8 / 0	-	4.4 / 0	-	ns
		XC2S100	3.8 / 0	-	4.4 / 0	-	ns
		XC2S150	3.9 / 0	-	4.6 / 0	-	ns
		XC2S200	3.9 / 0	-	4.6 / 0	-	ns
$T_{IOICECK}$ / $T_{IOCKICE}$	ICE input	All	0.9 / 0.01	-	0.9 / 0.01	-	ns
Set/Reset Delays							
$T_{IOSRCKI}$	SR input (IFF, synchronous)	All	-	1.1	-	1.2	ns
T_{IOSRIQ}	SR input to IQ (asynchronous)	All	-	1.5	-	1.7	ns
T_{GSRQ}	GSR to output IQ	All	-	9.9	-	11.7	ns

Notes:
1. Input timing for LVTTL is measured at 1.4V. For other I/O standards, see the table "Delay Measurement Methodology," page 60.
2. A zero hold time listing indicates no hold time or a negative hold time.

IOB Input Delay Adjustments for Different Standards[1]

Input delays associated with the pad are specified for LVTTL. For other standards, adjust the delays by the values shown. A delay adjusted in this way constitutes a worst-case limit.

Symbol	Description	Standard	Speed Grade		Units
			-6	-5	
Data Input Delay Adjustments					
T_{ILVTTL}	Standard-specific data input delay adjustments	LVTTL	0	0	ns
$T_{ILVCMOS2}$		LVCMOS2	–0.04	–0.05	ns
T_{IPCI33_3}		PCI, 33 MHz, 3.3V	–0.11	–0.13	ns
T_{IPCI33_5}		PCI, 33 MHz, 5.0V	0.26	0.30	ns
T_{IPCI66_3}		PCI, 66 MHz, 3.3V	–0.11	–0.13	ns
T_{IGTL}		GTL	0.20	0.24	ns
T_{IGTLP}		GTL+	0.11	0.13	ns
T_{IHSTL}		HSTL	0.03	0.04	ns
T_{ISSTL2}		SSTL2	–0.08	–0.09	ns
T_{ISSTL3}		SSTL3	–0.04	–0.05	ns
T_{ICTT}		CTT	0.02	0.02	ns
T_{IAGP}		AGP	–0.06	–0.07	ns

Notes:
1. Input timing for LVTTL is measured at 1.4V. For other I/O standards, see the table "Delay Measurement Methodology," page 60.

IOB Output Switching Characteristics

Output delays terminating at a pad are specified for LVTTL with 12 mA drive and fast slew rate. For other standards, adjust the delays with the values shown in "IOB Output Delay Adjustments for Different Standards," page 59.

Symbol	Description	Speed Grade				Units
		-6		-5		
		Min	Max	Min	Max	
Propagation Delays						
T_{IOOP}	O input to pad	-	2.9	-	3.4	ns
T_{IOOLP}	O input to pad via transparent latch	-	3.4	-	4.0	ns
3-state Delays						
T_{IOTHZ}	T input to pad high-impedance(1)	-	2.0	-	2.3	ns
T_{IOTON}	T input to valid data on pad	-	3.0	-	3.6	ns
$T_{IOTLPHZ}$	T input to pad high impedance via transparent latch(1)	-	2.5	-	2.9	ns
$T_{IOTLPON}$	T input to valid data on pad via transparent latch	-	3.5	-	4.2	ns
T_{GTS}	GTS to pad high impedance(1)	-	5.0	-	5.9	ns
Sequential Delays						
T_{IOCKP}	Clock CLK to pad	-	2.9	-	3.4	ns
T_{IOCKHZ}	Clock CLK to pad high impedance (synchronous)(1)	-	2.3	-	2.7	ns
T_{IOCKON}	Clock CLK to valid data on pad (synchronous)	-	3.3	-	4.0	ns
Setup/Hold Times with Respect to Clock CLK(2)						
T_{IOOCK} / T_{IOCKO}	O input	1.1 / 0	-	1.3 / 0	-	ns
$T_{IOOCECK}$ / $T_{IOCKOCE}$	OCE input	0.9 / 0.01	-	0.9 / 0.01	-	ns
$T_{IOSRCKO}$ / $T_{IOCKOSR}$	SR input (OFF)	1.2 / 0	-	1.3 / 0	-	ns
T_{IOTCK} / T_{IOCKT}	3-state setup times, T input	0.8 / 0	-	0.9 / 0	-	ns
$T_{IOTCECK}$ / $T_{IOCKTCE}$	3-state setup times, TCE input	1.0 / 0	-	1.0 / 0	-	ns
$T_{IOSRCKT}$ / $T_{IOCKTSR}$	3-state setup times, SR input (TFF)	1.1 / 0	-	1.2 / 0	-	ns
Set/Reset Delays						
T_{IOSRP}	SR input to pad (asynchronous)	-	3.7	-	4.4	ns
T_{IOSRHZ}	SR input to pad high impedance (asynchronous)(1)	-	3.1	-	3.7	ns
T_{IOSRON}	SR input to valid data on pad (asynchronous)	-	4.1	-	4.9	ns
T_{IOGSRQ}	GSR to pad	-	9.9	-	11.7	ns

Notes:
1. Three-state turn-off delays should not be adjusted.
2. A zero hold time listing indicates no hold time or a negative hold time.

Calculation of T_{IOOP} as a Function of Capacitance

T_{IOOP} is the propagation delay from the O Input of the IOB to the pad. The values for T_{IOOP} are based on the standard capacitive load (C_{SL}) for each I/O standard as listed in the table "Constants for Calculating TIOOP", below.

For other capacitive loads, use the formulas below to calculate an adjusted propagation delay, T_{IOOP1}.

$$T_{IOOP1} = T_{IOOP} + Adj + (C_{LOAD} – C_{SL}) * F_L$$

Where:

Adj is selected from "IOB Output Delay Adjustments for Different Standards", page 59, according to the I/O standard used

C_{LOAD} is the capacitive load for the design

F_L is the capacitance scaling factor

Delay Measurement Methodology

Standard	V_L (1)	V_H (1)	Meas. Point	V_{REF} Typ (2)
LVTTL	0	3	1.4	-
LVCMOS2	0	2.5	1.125	-
PCI33_5	Per PCI Spec			-
PCI33_3	Per PCI Spec			-
PCI66_3	Per PCI Spec			-
GTL	V_{REF} – 0.2	V_{REF} + 0.2	V_{REF}	0.80
GTL+	V_{REF} – 0.2	V_{REF} + 0.2	V_{REF}	1.0
HSTL Class I	V_{REF} – 0.5	V_{REF} + 0.5	V_{REF}	0.75
HSTL Class III	V_{REF} – 0.5	V_{REF} + 0.5	V_{REF}	0.90
HSTL Class IV	V_{REF} – 0.5	V_{REF} + 0.5	V_{REF}	0.90
SSTL3 I and II	V_{REF} – 1.0	V_{REF} + 1.0	V_{REF}	1.5
SSTL2 I and II	V_{REF} – 0.75	V_{REF} + 0.75	V_{REF}	1.25
CTT	V_{REF} – 0.2	V_{REF} + 0.2	V_{REF}	1.5
AGP	V_{REF} – (0.2xV_{CCO})	V_{REF} + (0.2xV_{CCO})	V_{REF}	Per AGP Spec

Notes:
1. Input waveform switches between V_L and V_H.
2. Measurements are made at V_{REF} Typ, Maximum, and Minimum. Worst-case values are reported.
3. I/O parameter measurements are made with the capacitance values shown in the table, "Constants for Calculating TIOOP". See Xilinx application note XAPP179 for the appropriate terminations.
4. I/O standard measurements are reflected in the IBIS model information except where the IBIS format precludes it.

Constants for Calculating T_{IOOP}

Standard	C_{SL} (1) (pF)	F_L (ns/pF)
LVTTL Fast Slew Rate, 2 mA drive	35	0.41
LVTTL Fast Slew Rate, 4 mA drive	35	0.20
LVTTL Fast Slew Rate, 6 mA drive	35	0.13
LVTTL Fast Slew Rate, 8 mA drive	35	0.079
LVTTL Fast Slew Rate, 12 mA drive	35	0.044
LVTTL Fast Slew Rate, 16 mA drive	35	0.043
LVTTL Fast Slew Rate, 24 mA drive	35	0.033
LVTTL Slow Slew Rate, 2 mA drive	35	0.41
LVTTL Slow Slew Rate, 4 mA drive	35	0.20
LVTTL Slow Slew Rate, 6 mA drive	35	0.100
LVTTL Slow Slew Rate, 8 mA drive	35	0.086
LVTTL Slow Slew Rate, 12 mA drive	35	0.058
LVTTL Slow Slew Rate, 16 mA drive	35	0.050
LVTTL Slow Slew Rate, 24 mA drive	35	0.048
LVCMOS2	35	0.041
PCI 33 MHz 5V	50	0.050
PCI 33 MHZ 3.3V	10	0.050
PCI 66 MHz 3.3V	10	0.033
GTL	0	0.014
GTL+	0	0.017
HSTL Class I	20	0.022
HSTL Class III	20	0.016
HSTL Class IV	20	0.014
SSTL2 Class I	30	0.028
SSTL2 Class II	30	0.016
SSTL3 Class I	30	0.029
SSTL3 Class II	30	0.016
CTT	20	0.035
AGP	10	0.037

Notes:
1. I/O parameter measurements are made with the capacitance values shown above. See Xilinx application note XAPP179 for the appropriate terminations.
2. I/O standard measurements are reflected in the IBIS model information except where the IBIS format precludes it.

DLL Timing Parameters

All devices are 100 percent functionally tested. Because of the difficulty in directly measuring many internal timing parameters, those parameters are derived from benchmark timing patterns. The following guidelines reflect worst-case values across the recommended operating conditions.

Symbol	Description	Speed Grade				Units
		-6		-5		
		Min	Max	Min	Max	
$F_{CLKINHF}$	Input clock frequency (CLKDLLHF)	60	200	60	180	MHz
$F_{CLKINLF}$	Input clock frequency (CLKDLL)	25	100	25	90	MHz
$T_{DLLPWHF}$	Input clock pulse width (CLKDLLHF)	2.0	-	2.4	-	ns
$T_{DLLPWLF}$	Input clock pulse width (CLKDLL)	2.5	-	3.0	-	ns

DLL Clock Tolerance, Jitter, and Phase Information

All DLL output jitter and phase specifications were determined through statistical measurement at the package pins using a clock mirror configuration and matched drivers. Figure 52, page 63, provides definitions for various parameters in the table below.

Symbol	Description	F_{CLKIN}	CLKDLLHF		CLKDLL		Units
			Min	Max	Min	Max	
T_{IPTOL}	Input clock period tolerance		-	1.0	-	1.0	ns
T_{IJITCC}	Input clock jitter tolerance (cycle-to-cycle)		-	±150	-	±300	ps
T_{LOCK}	Time required for DLL to acquire lock	> 60 MHz	-	20	-	20	µs
		50-60 MHz	-	-	-	25	µs
		40-50 MHz	-	-	-	50	µs
		30-40 MHz	-	-	-	90	µs
		25-30 MHz	-	-	-	120	µs
T_{OJITCC}	Output jitter (cycle-to-cycle) for any DLL clock output(1)		-	±60	-	±60	ps
T_{PHIO}	Phase offset between CLKIN and CLKO(2)		-	±100	-	±100	ps
T_{PHOO}	Phase offset between clock outputs on the DLL(3)		-	±140	-	±140	ps
T_{PHIOM}	Maximum phase difference between CLKIN and CLKO(4)		-	±160	-	±160	ps
T_{PHOOM}	Maximum phase difference between clock outputs on the DLL(5)		-	±200	-	±200	ps

Notes:
1. **Output Jitter** is cycle-to-cycle jitter measured on the DLL output clock, *excluding* input clock jitter.
2. **Phase Offset between CLKIN and CLKO** is the worst-case fixed time difference between rising edges of CLKIN and CLKO, *excluding* output jitter and input clock jitter.
3. **Phase Offset between Clock Outputs on the DLL** is the worst-case fixed time difference between rising edges of any two DLL outputs, *excluding* Output Jitter and input clock jitter.
4. **Maximum Phase Difference between CLKIN an CLKO** is the sum of Output Jitter and Phase Offset between CLKIN and CLKO, or the greatest difference between CLKIN and CLKO rising edges due to DLL alone (*excluding* input clock jitter).
5. **Maximum Phase Difference between Clock Outputs on the DLL** is the sum of Output JItter and Phase Offset between any DLL clock outputs, or the greatest difference between any two DLL output rising edges due to DLL alone (*excluding* input clock jitter).

Table 36: **Spartan-II Family Package Options**

Package	Leads	Type	Maximum I/O	Lead Pitch (mm)	Footprint Area (mm)	Height (mm)	Mass[(1)] (g)
VQ100 / VQG100	100	Very Thin Quad Flat Pack (VQFP)	60	0.5	16 x 16	1.20	0.6
TQ144 / TQG144	144	Thin Quad Flat Pack (TQFP)	92	0.5	22 x 22	1.60	1.4
CS144 / CSG144	144	Chip Scale Ball Grid Array (CSBGA)	92	0.8	12 x 12	1.20	0.3
PQ208 / PQG208	208	Plastic Quad Flat Pack (PQFP)	140	0.5	30.6 x 30.6	3.70	5.3
FG256 / FGG256	256	Fine-pitch Ball Grid Array (FBGA)	176	1.0	17 x 17	2.00	0.9
FG456 / FGG456	456	Fine-pitch Ball Grid Array (FBGA)	284	1.0	23 x 23	2.60	2.2

Notes:
1. Package mass is ±10%.

Note: Some early versions of Spartan-II devices, including the XC2S15 and XC2S30 ES devices and the XC2S150 with date code 0045 or earlier, included a power-down pin. For more information, see Answer Record 10500.

VCCO Banks

Some of the I/O standards require specific V_{CCO} voltages. These voltages are externally connected to device pins that serve groups of IOBs, called banks. Eight I/O banks result from separating each edge of the FPGA into two banks (see Figure 3 in Module 2). Each bank has multiple V_{CCO} pins which must be connected to the same voltage. In the smaller packages, the V_{CCO} pins are connected between banks, effectively reducing the number of independent banks available (see Table 37). These interconnected banks are shown in the Pinout Tables with V_{CCO} pads for multiple banks connected to the same pin.

Table 37: **Independent VCCO Banks Available**

Package	VQ100 PQ208	CS144 TQ144	FG256 FG456
Independent Banks	1	4	8

Package Overview

Table 36 shows the six low-cost, space-saving production package styles for the Spartan-II family.

Each package style is available in an environmentally friendly lead-free (Pb-free) option. The Pb-free packages include an extra 'G' in the package style name. For example, the standard "CS144" package becomes "CSG144" when ordered as the Pb-free option. Leaded (non-Pb-free) packages may be available for selected devices, with the same pin-out and without the "G" in the ordering code; contact Xilinx sales for more information. The mechanical dimensions of the standard and Pb-free packages are similar, as shown in the mechanical drawings provided in Table 38.

For additional package information, see UG112: *Device Package User Guide*.

Mechanical Drawings

Detailed mechanical drawings for each package type are available from the Xilinx web site at the specified location in Table 38.

Material Declaration Data Sheets (MDDS) are also available on the Xilinx web site for each package.

Table 38: **Xilinx Package Documentation**

Package	Drawing	MDDS
VQ100	Package Drawing	PK173_VQ100
VQG100		PK130_VQG100
TQ144	Package Drawing	PK169_TQ144
TQG144		PK126_TQG144
CS144	Package Drawing	PK149_CS144
CSG144		PK103_CSG144
PQ208	Package Drawing	PK166_PQ208
PQG208		PK123_PQG208
FG256	Package Drawing	PK151_FG256
FGG256		PK105_FGG256
FG456	Package Drawing	PK154_FG456
FGG456		PK109_FGG456

Package Thermal Characteristics

Table 39 provides the thermal characteristics for the various Spartan-II FPGA package offerings. This information is also available using the Thermal Query tool on xilinx.com (www.xilinx.com/cgi-bin/thermal/thermal.pl).

The junction-to-case thermal resistance (θ_{JC}) indicates the difference between the temperature measured on the package body (case) and the die junction temperature per watt of power consumption. The junction-to-board (θ_{JB}) value similarly reports the difference between the board and junction temperature. The junction-to-ambient (θ_{JA}) value reports the temperature difference between the ambient environment and the junction temperature. The θ_{JA} value is reported at different air velocities, measured in linear feet per minute (LFM). The "Still Air (0 LFM)" column shows the θ_{JA} value in a system without a fan. The thermal resistance drops with increasing air flow.

Table 39: **Spartan-II Package Thermal Characteristics**

Package	Device	Junction-to-Case (θ_{JC})	Junction-to-Board (θ_{JB})	Junction-to-Ambient (θ_{JA}) at Different Air Flows				Units
				Still Air (0 LFM)	250 LFM	500 LFM	750 LFM	
VQ100 VQG100	XC2S15	11.3	N/A	44.1	36.7	34.2	33.3	°C/Watt
	XC2S30	10.1	N/A	40.7	33.9	31.5	30.8	°C/Watt
TQ144 TQG144	XC2S15	7.3	N/A	38.6	30.0	25.7	24.1	°C/Watt
	XC2S30	6.7	N/A	34.7	27.0	23.1	21.7	°C/Watt
	XC2S50	5.8	N/A	32.2	25.1	21.4	20.1	°C/Watt
	XC2S100	5.3	N/A	31.4	24.4	20.9	19.6	°C/Watt
CS144 CSG144	XC2S30	2.8	N/A	34.0	26.0	23.9	23.2	°C/Watt
PQ208 PQG208	XC2S50	6.7	N/A	25.2	18.6	16.4	15.2	°C/Watt
	XC2S100	5.9	N/A	24.6	18.1	16.0	14.9	°C/Watt
	XC2S150	5.0	N/A	23.8	17.6	15.6	14.4	°C/Watt
	XC2S200	4.1	N/A	23.0	17.0	15.0	13.9	°C/Watt
FG256 FGG256	XC2S50	7.1	17.6	27.2	21.4	20.3	19.8	°C/Watt
	XC2S100	5.8	15.1	25.1	19.5	18.3	17.8	°C/Watt
	XC2S150	4.6	12.7	23.0	17.6	16.3	15.8	°C/Watt
	XC2S200	3.5	10.7	21.4	16.1	14.7	14.2	°C/Watt
FG456 FGG456	XC2S150	2.0	N/A	21.9	17.3	15.8	15.2	°C/Watt
	XC2S200	2.0	N/A	21.0	16.6	15.1	14.5	°C/Watt

Pinout Tables

The following device-specific pinout tables include all packages available for each Spartan®-II device. They follow the pad locations around the die, and include Boundary Scan register locations.

XC2S15 Device Pinouts

XC2S15 Pad Name					
Function	Bank	VQ100	TQ144	CS144	Bndry Scan
GND	-	P1	P143	A1	-
TMS	-	P2	P142	B1	-
I/O	7	P3	P141	C2	77
I/O	7	-	P140	C1	80
I/O, V_{REF}	7	P4	P139	D4	83
I/O	7	P5	P137	D2	86
I/O	7	P6	P136	D1	89
GND	-	-	P135	E4	-
I/O	7	P7	P134	E3	92
I/O	7	-	P133	E2	95
I/O, V_{REF}	7	P8	P132	E1	98
I/O	7	P9	P131	F4	101
I/O	7	-	P130	F3	104
I/O, IRDY(1)	7	P10	P129	F2	107
GND	-	P11	P128	F1	-
V_{CCO}	7	P12	P127	G2	-
V_{CCO}	6	P12	P127	G2	-
I/O, TRDY(1)	6	P13	P126	G1	110
V_{CCINT}	-	P14	P125	G3	-
I/O	6	-	P124	G4	113
I/O	6	P15	P123	H1	116
I/O, V_{REF}	6	P16	P122	H2	119
I/O	6	-	P121	H3	122
I/O	6	P17	P120	H4	125
GND	-	-	P119	J1	-
I/O	6	P18	P118	J2	128
I/O	6	P19	P117	J3	131
I/O, V_{REF}	6	P20	P115	K1	134
I/O	6	-	P114	K2	137
I/O	6	P21	P113	K3	140
I/O	6	P22	P112	L1	143
M1	-	P23	P111	L2	146
GND	-	P24	P110	L3	-
M0	-	P25	P109	M1	147
V_{CCO}	6	P26	P108	M2	-
V_{CCO}	5	P26	P107	N1	-

XC2S15 Device Pinouts *(Continued)*

XC2S15 Pad Name					
Function	Bank	VQ100	TQ144	CS144	Bndry Scan
M2	-	P27	P106	N2	148
I/O	5	-	P103	K4	155
I/O, V_{REF}	5	P30	P102	L4	158
I/O	5	P31	P100	N4	161
I/O	5	P32	P99	K5	164
GND	-	-	P98	L5	-
V_{CCINT}	-	P33	P97	M5	-
I/O	5	-	P96	N5	167
I/O	5	-	P95	K6	170
I/O, V_{REF}	5	P34	P94	L6	173
I/O	5	-	P93	M6	176
V_{CCINT}	-	P35	P92	N6	-
I, GCK1	5	P36	P91	M7	185
V_{CCO}	5	P37	P90	N7	-
V_{CCO}	4	P37	P90	N7	-
GND	-	P38	P89	L7	-
I, GCK0	4	P39	P88	K7	186
I/O	4	P40	P87	N8	190
I/O	4	-	P86	M8	193
I/O, V_{REF}	4	P41	P85	L8	196
I/O	4	-	P84	K8	199
I/O	4	-	P83	N9	202
V_{CCINT}	-	P42	P82	M9	-
GND	-	-	P81	L9	-
I/O	4	P43	P80	K9	205
I/O	4	P44	P79	N10	208
I/O, V_{REF}	4	P45	P77	L10	211
I/O	4	-	P76	N11	214
I/O	4	P46	P75	M11	217
I/O	4	P47	P74	L11	220
GND	-	P48	P73	N12	-
DONE	3	P49	P72	M12	223
V_{CCO}	4	P50	P71	N13	-
V_{CCO}	3	P50	P70	M13	-
$\overline{\text{PROGRAM}}$	-	P51	P69	L12	226
I/O ($\overline{\text{INIT}}$)	3	P52	P68	L13	227
I/O (D7)	3	P53	P67	K10	230
I/O	3	-	P66	K11	233
I/O, V_{REF}	3	P54	P65	K12	236
I/O	3	P55	P63	J10	239
I/O (D6)	3	P56	P62	J11	242

XC2S100 Device Pinouts *(Continued)*

XC2S100 Pad Name						
Function	Bank	TQ144	PQ208	FG256	FG456	Bndry Scan
V_{CCINT}	-	-	P38	V_{CCINT}*	V_{CCINT}*	-
V_{CCO}	6	-	P39	V_{CCO} Bank 6*	V_{CCO} Bank 6*	-
GND	-	P119	P40	GND*	GND*	-
I/O	6	P118	P41	K4	T1	314
I/O, V_{REF}	6	P117	P42	M1	R4	317
I/O	6	-	-	-	T2	320
I/O	6	P116	P43	L4	U1	323
I/O	6	-	-	M2	R5	326
I/O	6	-	P44	L3	U2	332
I/O, V_{REF}	6	P115	P45	N1	T3	335
V_{CCO}	6	-	-	V_{CCO} Bank 6*	V_{CCO} Bank 6*	-
GND	-	-	-	GND*	GND*	-
I/O	6	-	P46	P1	T4	338
I/O	6	-	-	L5	W1	341
I/O	6	-	-	-	U4	344
I/O	6	P114	P47	N2	Y1	347
I/O	6	-	-	M4	W2	350
I/O	6	P113	P48	R1	Y2	356
I/O	6	P112	P49	M3	W3	359
M1	-	P111	P50	P2	U5	362
GND	-	P110	P51	GND*	GND*	-
M0	-	P109	P52	N3	AB2	363
V_{CCO}	6	P108	P53	V_{CCO} Bank 6*	V_{CCO} Bank 6*	-
V_{CCO}	5	P107	P53	V_{CCO} Bank 5*	V_{CCO} Bank 5*	-
M2	-	P106	P54	R3	Y4	364
I/O	5	-	-	N5	V7	374
I/O	5	P103	P57	T2	Y6	377
I/O	5	-	-	-	AA4	380
I/O	5	-	-	P5	W6	383
I/O	5	-	P58	T3	Y7	386
GND	-	-	-	GND*	GND*	-
V_{CCO}	5	-	-	V_{CCO} Bank 5*	V_{CCO} Bank 5*	-
I/O, V_{REF}	5	P102	P59	T4	AA5	389
I/O	5	-	P60	M6	AB5	392
I/O	5	-	-	T5	AB6	398
I/O	5	P101	P61	N6	AA7	401
I/O	5	-	-	-	W7	404

XC2S100 Device Pinouts *(Continued)*

XC2S100 Pad Name						
Function	Bank	TQ144	PQ208	FG256	FG456	Bndry Scan
I/O, V_{REF}	5	P100	P62	R5	W8	407
I/O	5	P99	P63	P6	Y8	410
GND	-	P98	P64	GND*	GND*	-
V_{CCO}	5	-	P65	V_{CCO} Bank 5*	V_{CCO} Bank 5*	-
V_{CCINT}	-	P97	P66	V_{CCINT}*	V_{CCINT}*	-
I/O	5	P96	P67	R6	AA8	413
I/O	5	P95	P68	M7	V9	416
I/O	5	-	-	-	AB9	419
I/O	5	-	P69	N7	Y9	422
I/O	5	-	P70	T6	W10	428
I/O	5	-	P71	P7	AB10	431
GND	-	-	P72	GND*	GND*	-
I/O, V_{REF}	5	P94	P73	P8	Y10	434
I/O	5	-	P74	R7	V11	437
I/O	5	-	-	T7	W11	440
I/O	5	P93	P75	T8	AB11	443
V_{CCINT}	-	P92	P76	V_{CCINT}*	V_{CCINT}*	-
I, GCK1	5	P91	P77	R8	Y11	455
V_{CCO}	5	P90	P78	V_{CCO} Bank 5*	V_{CCO} Bank 5*	-
V_{CCO}	4	P90	P78	V_{CCO} Bank 4*	V_{CCO} Bank 4*	-
GND	-	P89	P79	GND*	GND*	-
I, GCK0	4	P88	P80	N8	W12	456
I/O	4	P87	P81	N9	U12	460
I/O	4	P86	P82	R9	Y12	466
I/O	4	-	-	N10	AA12	469
I/O	4	-	P83	T9	AB13	472
I/O, V_{REF}	4	P85	P84	P9	AA13	475
GND	-	-	P85	GND*	GND*	-
I/O	4	-	P86	M10	Y13	478
I/O	4	-	P87	R10	V13	481
I/O	4	-	P88	P10	AA14	487
I/O	4	-	-	-	V14	490
I/O	4	P84	P89	T10	AB15	493
I/O	4	P83	P90	R11	AA15	496
V_{CCINT}	-	P82	P91	V_{CCINT}*	V_{CCINT}*	-
V_{CCO}	4	-	P92	V_{CCO} Bank 4*	V_{CCO} Bank 4*	-
GND	-	P81	P93	GND*	GND*	-
I/O	4	P80	P94	M11	Y15	499

XC2S150 Device Pinouts *(Continued)*

XC2S150 Pad Name					Bndry
Function	Bank	PQ208	FG256	FG456	Scan
I/O	1	P174	B10	C14	72
I/O	1	-	-	B14	75
I/O	1	P175	D10	D13	81
I/O	1	P176	A10	C13	84
GND	-	P177	GND*	GND*	-
V_{CCO}	1	-	V_{CCO} Bank 1*	V_{CCO} Bank 1*	-
I/O, V_{REF}	1	P178	B9	B13	87
I/O	1	P179	E10	E12	90
I/O	1	-	A9	B12	93
I/O	1	P180	D9	D12	96
I/O	1	-	-	C12	99
I/O	1	P181	A8	D11	102
I, GCK2	1	P182	C9	A11	108
GND	-	P183	GND*	GND*	-
V_{CCO}	1	P184	V_{CCO} Bank 1*	V_{CCO} Bank 1*	-
V_{CCO}	0	P184	V_{CCO} Bank 0*	V_{CCO} Bank 0*	-
I, GCK3	0	P185	B8	C11	109
V_{CCINT}	-	P186	V_{CCINT}*	V_{CCINT}*	-
I/O	0	-	-	E11	116
I/O	0	P187	A7	A10	119
I/O	0	-	D8	B10	122
I/O	0	P188	A6	C10	125
I/O, V_{REF}	0	P189	B7	A9	128
V_{CCO}	0	-	V_{CCO} Bank 0*	V_{CCO} Bank 0*	-
GND	-	P190	GND*	GND*	-
I/O	0	P191	C8	B9	131
I/O	0	P192	D7	E10	134
I/O	0	-	-	D10	140
I/O	0	P193	E7	A8	143
I/O	0	-	-	D9	146
I/O	0	-	-	B8	149
I/O	0	P194	C7	E9	155
I/O	0	P195	B6	A7	158

XC2S150 Device Pinouts *(Continued)*

XC2S150 Pad Name					Bndry
Function	Bank	PQ208	FG256	FG456	Scan
V_{CCINT}	-	P196	V_{CCINT}*	V_{CCINT}*	-
V_{CCO}	0	P197	V_{CCO} Bank 0*	V_{CCO} Bank 0*	-
GND	-	P198	GND*	GND*	-
I/O	0	P199	A5	B7	161
I/O, V_{REF}	0	P200	C6	E8	164
I/O	0	-	-	D8	167
I/O	0	P201	B5	C7	170
I/O	0	-	D6	D7	173
I/O	0	-	-	B6	176
I/O	0	-	-	A5	179
I/O	0	P202	A4	D6	182
I/O, V_{REF}	0	P203	B4	C6	185
V_{CCO}	0	-	V_{CCO} Bank 0*	V_{CCO} Bank 0*	-
GND	-	-	GND*	GND*	-
I/O	0	P204	E6	B5	188
I/O	0	-	D5	E7	191
I/O	0	-	-	A4	194
I/O	0	-	-	E6	197
I/O	0	P205	A3	B4	200
GND	-	-	GND*	GND*	-
I/O	0	-	C5	A3	203
I/O	0	-	-	B3	206
I/O	0	-	-	D5	209
I/O	0	P206	B3	C5	212
TCK	-	P207	C4	C4	-
V_{CCO}	0	P208	V_{CCO} Bank 0*	V_{CCO} Bank 0*	-
V_{CCO}	7	P208	V_{CCO} Bank 7*	V_{CCO} Bank 7*	-

04/18/01

Notes:

1. IRDY and TRDY can only be accessed when using Xilinx PCI cores.
2. Pads labelled GND*, V_{CCINT}*, V_{CCO} Bank 0*, V_{CCO} Bank 1*, V_{CCO} Bank 2*, V_{CCO} Bank 3*, V_{CCO} Bank 4*, V_{CCO} Bank 5*, V_{CCO} Bank 6*, V_{CCO} Bank 7* are internally bonded to independent ground or power planes within the package.
3. See "VCCO Banks" for details on V_{CCO} banking.

XC2S200 Device Pinouts

XC2S200 Pad Name					Bndry
Function	Bank	PQ208	FG256	FG456	Scan
GND	-	P1	GND*	GND*	-
TMS	-	P2	D3	D3	-
I/O	7	P3	C2	B1	257
I/O	7	-	-	E4	263
I/O	7	-	-	C1	266
I/O	7	-	A2	F5	269
GND	-	-	GND*	GND*	-
I/O, V_{REF}	7	P4	B1	D2	272
I/O	7	-	-	E3	275
I/O	7	-	-	F4	281
GND	-	-	GND*	GND*	-
I/O	7	-	E3	G5	284
I/O	7	P5	D2	F3	287
GND	-	-	GND*	GND*	-
V_{CCO}	7	-	V_{CCO} Bank 7*	V_{CCO} Bank 7*	-
I/O, V_{REF}	7	P6	C1	E2	290
I/O	7	P7	F3	E1	293
I/O	7	-	-	G4	296
I/O	7	-	-	G3	299
I/O	7	-	E2	H5	302
GND	-	-	GND*	GND*	-
I/O	7	P8	E4	F2	305
I/O	7	-	-	F1	308
I/O, V_{REF}	7	P9	D1	H4	314
I/O	7	P10	E1	G1	317
GND	-	P11	GND*	GND*	-
V_{CCO}	7	P12	V_{CCO} Bank 7*	V_{CCO} Bank 7*	-
V_{CCINT}	-	P13	V_{CCINT}*	V_{CCINT}*	-
I/O	7	P14	F2	H3	320
I/O	7	P15	G3	H2	323
I/O	7	-	-	J4	326
I/O	7	-	-	H1	329
I/O	7	-	F1	J5	332
GND	-	-	GND*	GND*	-
I/O	7	P16	F4	J2	335
I/O	7	-	-	J3	338
I/O	7	-	-	J1	341
I/O	7	P17	F5	K5	344
I/O	7	P18	G2	K1	347
GND	-	P19	GND*	GND*	-

XC2S200 Device Pinouts *(Continued)*

XC2S200 Pad Name					Bndry
Function	Bank	PQ208	FG256	FG456	Scan
V_{CCO}	7	-	V_{CCO} Bank 7*	V_{CCO} Bank 7*	-
I/O, V_{REF}	7	P20	H3	K3	350
I/O	7	P21	G4	K4	353
I/O	7	-	-	K2	359
I/O	7	-	H2	L6	362
I/O	7	P22	G5	L1	365
I/O	7	-	-	L5	368
I/O	7	P23	H4	L4	374
I/O, IRDY(1)	7	P24	G1	L3	377
GND	-	P25	GND*	GND*	-
V_{CCO}	7	P26	V_{CCO} Bank 7*	V_{CCO} Bank 7*	-
V_{CCO}	6	P26	V_{CCO} Bank 6*	V_{CCO} Bank 6*	-
I/O, TRDY(1)	6	P27	J2	M1	380
V_{CCINT}	-	P28	V_{CCINT}*	V_{CCINT}*	-
I/O	6	-	-	M6	389
I/O	6	P29	H1	M3	392
I/O	6	-	J4	M4	395
I/O	6	-	-	N1	398
I/O	6	P30	J1	M5	404
I/O, V_{REF}	6	P31	J3	N2	407
V_{CCO}	6	-	V_{CCO} Bank 6*	V_{CCO} Bank 6*	-
GND	-	P32	GND*	GND*	-
I/O	6	P33	K5	N3	410
I/O	6	P34	K2	N4	413
I/O	6	-	-	P1	416
I/O	6	-	-	N5	419
I/O	6	P35	K1	P2	422
GND	-	-	GND*	GND*	-
I/O	6	-	K3	P4	425
I/O	6	-	-	R1	428
I/O	6	-	-	P5	431
I/O	6	P36	L1	P3	434
I/O	6	P37	L2	R2	437
V_{CCINT}	-	P38	V_{CCINT}*	V_{CCINT}*	-
V_{CCO}	6	P39	V_{CCO} Bank 6*	V_{CCO} Bank 6*	-
GND	-	P40	GND*	GND*	-
I/O	6	P41	K4	T1	440
I/O, V_{REF}	6	P42	M1	R4	443