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Details	
Product Status	Active
Number of LABs/CLBs	1176
Number of Logic Elements/Cells	5292
Total RAM Bits	57344
Number of I/O	176
Number of Gates	200000
Voltage - Supply	2.375V ~ 2.625V
Mounting Type	Surface Mount
Operating Temperature	-40°C ~ 100°C (TJ)
Package / Case	256-BGA
Supplier Device Package	256-FBGA (17x17)
Purchase URL	https://www.e-xfl.com/product-detail/xilinx/xc2s200-5fg256i

Spartan-II FPGA Family: Introduction and Ordering Information

DS001-1 (v2.8) June 13, 2008 **Product Specification**

Introduction

The Spartan®-II Field-Programmable Gate Array family gives users high performance, abundant logic resources, and a rich feature set, all at an exceptionally low price. The six-member family offers densities ranging from 15,000 to 200,000 system gates, as shown in Table 1. System performance is supported up to 200 MHz. Features include block RAM (to 56K bits), distributed RAM (to 75,264 bits), 16 selectable I/O standards, and four DLLs. Fast, predictable interconnect means that successive design iterations continue to meet timing requirements.

The Spartan-II family is a superior alternative to mask-programmed ASICs. The FPGA avoids the initial cost, lengthy development cycles, and inherent risk of conventional ASICs. Also, FPGA programmability permits design upgrades in the field with no hardware replacement necessary (impossible with ASICs).

Features

- Second generation ASIC replacement technology
  - Densities as high as 5,292 logic cells with up to 200,000 system gates
  - Streamlined features based on Virtex® FPGA architecture
  - Unlimited reprogrammability
  - Very low cost
  - Cost-effective 0.18 micron process
- System level features
  - SelectRAM™ hierarchical memory:
    - 16 bits/LUT distributed RAM
    - Configurable 4K bit block RAM
    - Fast interfaces to external RAM
  - Fully PCI compliant
  - Low-power segmented routing architecture
  - Full readback ability for verification/observability
  - Dedicated carry logic for high-speed arithmetic
  - Efficient multiplier support
  - Cascade chain for wide-input functions
  - Abundant registers/latches with enable, set, reset
  - Four dedicated DLLs for advanced clock control
  - Four primary low-skew global clock distribution nets
  - IEEE 1149.1 compatible boundary scan logic
- Versatile I/O and packaging
  - Pb-free package options
  - Low-cost packages available in all densities
  - Family footprint compatibility in common packages
  - 16 high-performance interface standards
  - Hot swap Compact PCI friendly
  - Zero hold time simplifies system timing
- Core logic powered at 2.5V and I/Os powered at 1.5V, 2.5V, or 3.3V
- Fully supported by powerful Xilinx® ISE® development system
  - Fully automatic mapping, placement, and routing

Table 1: **Spartan-II FPGA Family Members**

Device	Logic Cells	System Gates (Logic and RAM)	CLB Array (R x C)	Total CLBs	Maximum Available User I/O[(1)]	Total Distributed RAM Bits	Total Block RAM Bits
XC2S15	432	15,000	8 x 12	96	86	6,144	16K
XC2S30	972	30,000	12 x 18	216	92	13,824	24K
XC2S50	1,728	50,000	16 x 24	384	176	24,576	32K
XC2S100	2,700	100,000	20 x 30	600	176	38,400	40K
XC2S150	3,888	150,000	24 x 36	864	260	55,296	48K
XC2S200	5,292	200,000	28 x 42	1,176	284	75,264	56K

Notes:
1. All user I/O counts do not include the four global clock/user input pins. See details in Table 2, page 4.

Revision History

Date	Version No.	Description
09/18/00	2.0	Sectioned the Spartan-II Family data sheet into four modules. Added industrial temperature range information.
10/31/00	2.1	Removed Power down feature.
03/05/01	2.2	Added statement on PROMs.
11/01/01	2.3	Updated Product Availability chart. Minor text edits.
09/03/03	2.4	Added device part marking.
08/02/04	2.5	Added information on Pb-free packaging options and removed discontinued options.
06/13/08	2.8	Updated description and links. Updated all modules for continuous page, figure, and table numbering. Synchronized all modules to v2.8.

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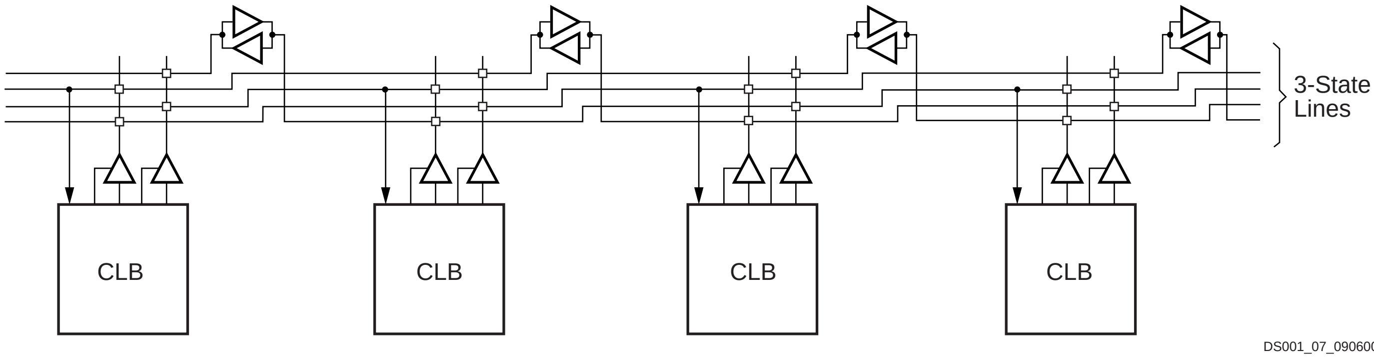

Figure 7: **BUFT Connections to Dedicated Horizontal Bus Lines**

Clock Distribution

The Spartan-II family provides high-speed, low-skew clock distribution through the primary global routing resources described above. A typical clock distribution net is shown in Figure 8.

Four global buffers are provided, two at the top center of the device and two at the bottom center. These drive the four primary global nets that in turn drive any clock pin.

Four dedicated clock pads are provided, one adjacent to each of the global buffers. The input to the global buffer is selected either from these pads or from signals in the general purpose routing. Global clock pins do not have the option for internal, weak pull-up resistors.

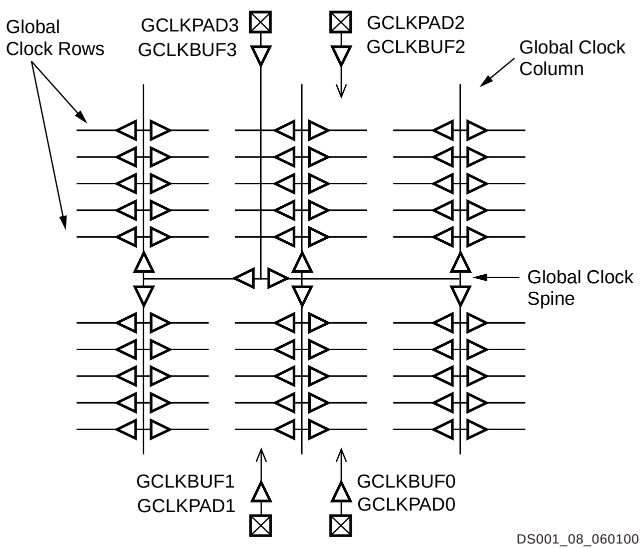

Figure 8: **Global Clock Distribution Network**

Delay-Locked Loop (DLL)

Associated with each global clock input buffer is a fully digital Delay-Locked Loop (DLL) that can eliminate skew between the clock input pad and internal clock-input pins throughout the device. Each DLL can drive two global clock networks. The DLL monitors the input clock and the distributed clock, and automatically adjusts a clock delay element. Additional delay is introduced such that clock edges reach internal flip-flops exactly one clock period after they arrive at the input. This closed-loop system effectively eliminates clock-distribution delay by ensuring that clock edges arrive at internal flip-flops in synchronism with clock edges arriving at the input.

In addition to eliminating clock-distribution delay, the DLL provides advanced control of multiple clock domains. The DLL provides four quadrature phases of the source clock, can double the clock, or divide the clock by 1.5, 2, 2.5, 3, 4, 5, 8, or 16. It has six outputs.

The DLL also operates as a clock mirror. By driving the output from a DLL off-chip and then back on again, the DLL can be used to deskew a board level clock among multiple Spartan-II devices.

In order to guarantee that the system clock is operating correctly prior to the FPGA starting up after configuration, the DLL can delay the completion of the configuration process until after it has achieved lock.

Boundary Scan

Spartan-II devices support all the mandatory boundary-scan instructions specified in the IEEE standard 1149.1. A Test Access Port (TAP) and registers are provided that implement the EXTEST, SAMPLE/PRELOAD, and BYPASS instructions. The TAP also supports two USERCODE instructions and internal scan chains.

The TAP uses dedicated package pins that always operate using LVTTL. For TDO to operate using LVTTL, the V_{CCO} for Bank 2 must be 3.3V. Otherwise, TDO switches rail-to-rail between ground and V_{CCO}. TDI, TMS, and TCK have a default internal weak pull-up resistor, and TDO has no default resistor. Bitstream options allow setting any of the four TAP pins to have an internal pull-up, pull-down, or neither.

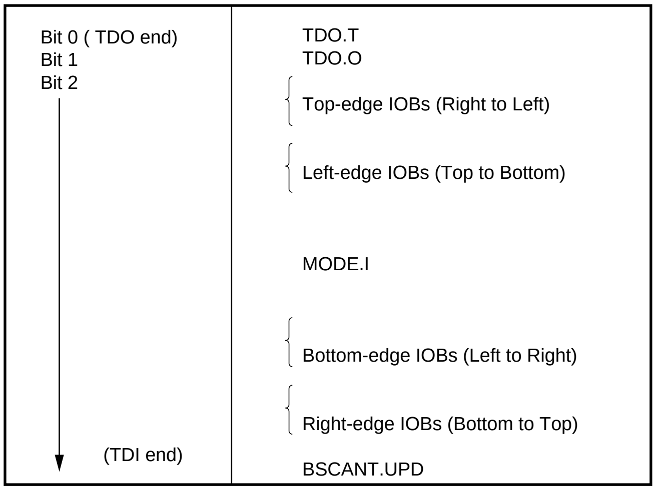

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Figure 10: **Boundary Scan Bit Sequence**

Development System

Spartan-II FPGAs are supported by the Xilinx ISE® development tools. The basic methodology for Spartan-II FPGA design consists of three interrelated steps: design entry, implementation, and verification. Industry-standard tools are used for design entry and simulation, while Xilinx provides proprietary architecture-specific tools for implementation.

The Xilinx development system is integrated under a single graphical interface, providing designers with a common user interface regardless of their choice of entry and verification tools. The software simplifies the selection of implementation options with pull-down menus and on-line help.

For HDL design entry, the Xilinx FPGA development system provides interfaces to several synthesis design environments.

A standard interface-file specification, Electronic Design Interchange Format (EDIF), simplifies file transfers into and out of the development system.

Spartan-II FPGAs supported by a unified library of standard functions. This library contains over 400 primitives and macros, ranging from 2-input AND gates to 16-bit accumulators, and includes arithmetic functions, comparators, counters, data registers, decoders, encoders, I/O functions, latches, Boolean functions, multiplexers, shift registers, and barrel shifters.

The design environment supports hierarchical design entry. These hierarchical design elements are automatically combined by the implementation tools. Different design entry tools can be combined within a hierarchical design, thus allowing the most convenient entry method to be used for each portion of the design.

Design Implementation

The place-and-route tools (PAR) automatically provide the implementation flow described in this section. The partitioner takes the EDIF netlist for the design and maps the logic into the architectural resources of the FPGA (CLBs and IOBs, for example). The placer then determines the best locations for these blocks based on their interconnections and the desired performance. Finally, the router interconnects the blocks.

The PAR algorithms support fully automatic implementation of most designs. For demanding applications, however, the user can exercise various degrees of control over the process. User partitioning, placement, and routing information is optionally specified during the design-entry process. The implementation of highly structured designs can benefit greatly from basic floorplanning.

The implementation software incorporates timing-driven placement and routing. Designers specify timing requirements along entire paths during design entry. The timing path analysis routines in PAR then recognize these user-specified requirements and accommodate them.

Timing requirements are entered in a form directly relating to the system requirements, such as the targeted clock frequency, or the maximum allowable delay between two registers. In this way, the overall performance of the system along entire signal paths is automatically tailored to user-generated specifications. Specific timing information for individual nets is unnecessary.

Design Verification

In addition to conventional software simulation, FPGA users can use in-circuit debugging techniques. Because Xilinx devices are infinitely reprogrammable, designs can be verified in real time without the need for extensive sets of software simulation vectors.

The development system supports both software simulation and in-circuit debugging techniques. For simulation, the system extracts the post-layout timing information from the design database, and back-annotates this information into the netlist for use by the simulator. Alternatively, the user can verify timing-critical portions of the design using the static timing analyzer.

For in-circuit debugging, the development system includes a download cable, which connects the FPGA in the target system to a PC or workstation. After downloading the design into the FPGA, the designer can read back the contents of the flip-flops, and so observe the internal logic state. Simple modifications can be downloaded into the system in a matter of minutes.

By default, these operations are synchronized to CCLK. The entire start-up sequence lasts eight cycles, called C0-C7, after which the loaded design is fully functional. The default timing for start-up is shown in the top half of Figure 13. The four operations can be selected to switch on any CCLK cycle C1-C6 through settings in the Xilinx software. Heavy lines show default settings.

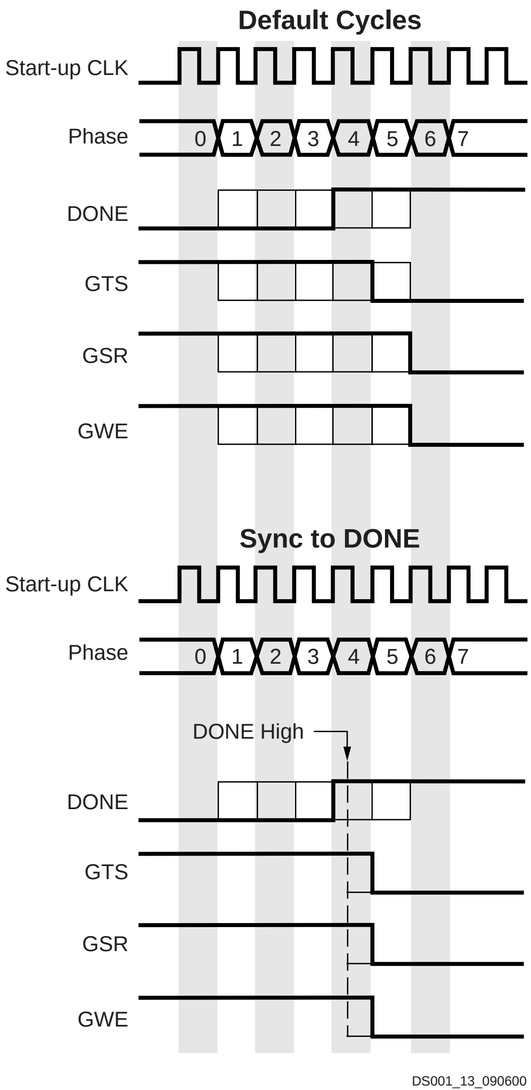


Figure 13: **Start-Up Waveforms**

The bottom half of Figure 13 shows another commonly used version of the start-up timing known as Sync-to-DONE. This version makes the GTS, GSR, and GWE events conditional upon the DONE pin going High. This timing is important for a daisy chain of multiple FPGAs in serial mode, since it ensures that all FPGAs go through start-up together, after all their DONE pins have gone High.

Sync-to-DONE timing is selected by setting the GTS, GSR, and GWE cycles to a value of DONE in the configuration options. This causes these signals to transition one clock cycle after DONE externally transitions High.

Serial Modes

There are two serial configuration modes: In Master Serial mode, the FPGA controls the configuration process by driving CCLK as an output. In Slave Serial mode, the FPGA passively receives CCLK as an input from an external agent (e.g., a microprocessor, CPLD, or second FPGA in master mode) that is controlling the configuration process. In both modes, the FPGA is configured by loading one bit per CCLK cycle. The MSB of each configuration data byte is always written to the DIN pin first.

See Figure 14 for the sequence for loading data into the Spartan-II FPGA serially. This is an expansion of the "Load Configuration Data Frames" block in Figure 11. Note that $\overline{CS}$ and $\overline{WRITE}$ normally are not used during serial configuration. To ensure successful loading of the FPGA, do not toggle $\overline{WRITE}$ with $\overline{CS}$ Low during serial configuration.

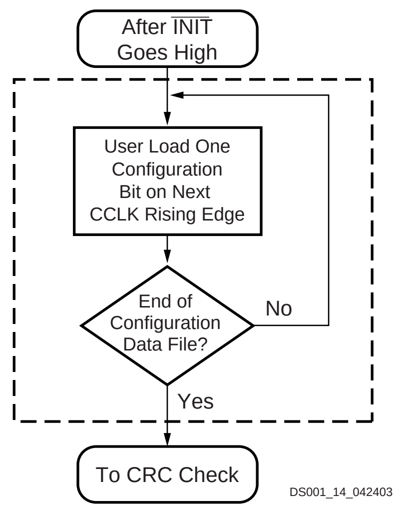


Figure 14: **Loading Serial Mode Configuration Data**

Design Considerations

This section contains more detailed design information on the following features:

- Delay-Locked Loop . . . see page 27
- Block RAM . . . see page 32
- Versatile I/O . . . see page 36

Using Delay-Locked Loops

The Spartan-II FPGA family provides up to four fully digital dedicated on-chip Delay-Locked Loop (DLL) circuits which provide zero propagation delay, low clock skew between output clock signals distributed throughout the device, and advanced clock domain control. These dedicated DLLs can be used to implement several circuits that improve and simplify system level design.

Introduction

Quality on-chip clock distribution is important. Clock skew and clock delay impact device performance and the task of managing clock skew and clock delay with conventional clock trees becomes more difficult in large devices. The Spartan-II family of devices resolve this potential problem by providing up to four fully digital dedicated on-chip Delay-Locked Loop (DLL) circuits which provide zero propagation delay and low clock skew between output clock signals distributed throughout the device.

Each DLL can drive up to two global clock routing networks within the device. The global clock distribution network minimizes clock skews due to loading differences. By monitoring a sample of the DLL output clock, the DLL can compensate for the delay on the routing network, effectively eliminating the delay from the external input port to the individual clock loads within the device.

In addition to providing zero delay with respect to a user source clock, the DLL can provide multiple phases of the source clock. The DLL can also act as a clock doubler or it can divide the user source clock by up to 16.

Clock multiplication gives the designer a number of design alternatives. For instance, a 50 MHz source clock doubled by the DLL can drive an FPGA design operating at 100 MHz. This technique can simplify board design because the clock path on the board no longer distributes such a high-speed signal. A multiplied clock also provides designers the option of time-domain-multiplexing, using one circuit twice per clock cycle, consuming less area than two copies of the same circuit.

The DLL can also act as a clock mirror. By driving the DLL output off-chip and then back in again, the DLL can be used to de-skew a board level clock between multiple devices.

In order to guarantee the system clock establishes prior to the device "waking up," the DLL can delay the completion of the device configuration process until after the DLL achieves lock.

By taking advantage of the DLL to remove on-chip clock delay, the designer can greatly simplify and improve system level design involving high-fanout, high-performance clocks.

Library DLL Primitives

Figure 22 shows the simplified Xilinx library DLL macro, BUFGDLL. This macro delivers a quick and efficient way to provide a system clock with zero propagation delay throughout the device. Figure 23 and Figure 24 show the two library DLL primitives. These primitives provide access to the complete set of DLL features when implementing more complex applications.

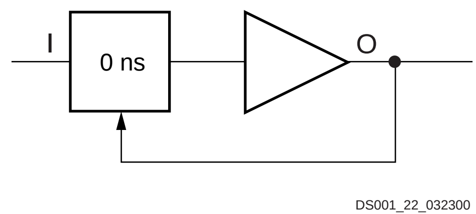


Figure 22: **Simplified DLL Macro BUFGDLL**

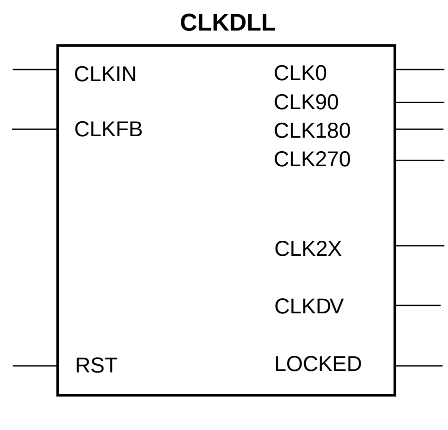


Figure 23: **Standard DLL Primitive CLKDLL**

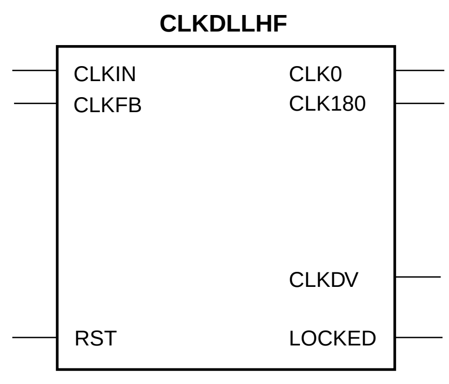


Figure 24: **High-Frequency DLL Primitive CLKDLLHF**

division factor N except for non-integer division in High Frequency (HF) mode. For division factor 1.5 the duty cycle in the HF mode is 33.3% High and 66.7% Low. For division factor 2.5, the duty cycle in the HF mode is 40.0% High and 60.0% Low.

1x Clock Outputs — CLK[0|90|180|270]

The 1x clock output pin CLK0 represents a delay-compensated version of the source clock (CLKIN) signal. The CLKDLL primitive provides three phase-shifted versions of the CLK0 signal while CLKDLLHF provides only the 180 degree phase-shifted version. The relationship between phase shift and the corresponding period shift appears in Table 10.

The timing diagrams in Figure 26 illustrate the DLL clock output characteristics.

Table 10: **Relationship of Phase-Shifted Output Clock to Period Shift**

Phase (degrees)	Period Shift (percent)
0	0%
90	25%
180	50%
270	75%

The DLL provides duty cycle correction on all 1x clock outputs such that all 1x clock outputs by default have a 50/50 duty cycle. The DUTY_CYCLE_CORRECTION property (TRUE by default), controls this feature. In order to deactivate the DLL duty cycle correction, attach the DUTY_CYCLE_CORRECTION=FALSE property to the DLL primitive. When duty cycle correction deactivates, the output clock has the same duty cycle as the source clock.

The DLL clock outputs can drive an OBUF, a BUFG, or they can route directly to destination clock pins. The DLL clock outputs can only drive the BUFGs that reside on the same edge (top or bottom).

Locked Output — LOCKED

In order to achieve lock, the DLL may need to sample several thousand clock cycles. After the DLL achieves lock the LOCKED signal activates. The "DLL Timing Parameters" section of Module 3 provides estimates for locking times.

In order to guarantee that the system clock is established prior to the device "waking up," the DLL can delay the completion of the device configuration process until after the DLL locks. The STARTUP_WAIT property activates this feature.

Until the LOCKED signal activates, the DLL output clocks are not valid and can exhibit glitches, spikes, or other spurious movement. In particular the CLK2X output will appear as a 1x clock with a 25/75 duty cycle.

DLL Properties

Properties provide access to some of the Spartan-II family DLL features, (for example, clock division and duty cycle correction).

Duty Cycle Correction Property

The 1x clock outputs, CLK0, CLK90, CLK180, and CLK270, use the duty-cycle corrected default, such that they exhibit a 50/50 duty cycle. The DUTY_CYCLE_CORRECTION property (by default TRUE) controls this feature. To deactivate the DLL duty-cycle correction for the 1x clock outputs, attach the DUTY_CYCLE_CORRECTION=FALSE property to the DLL primitive.

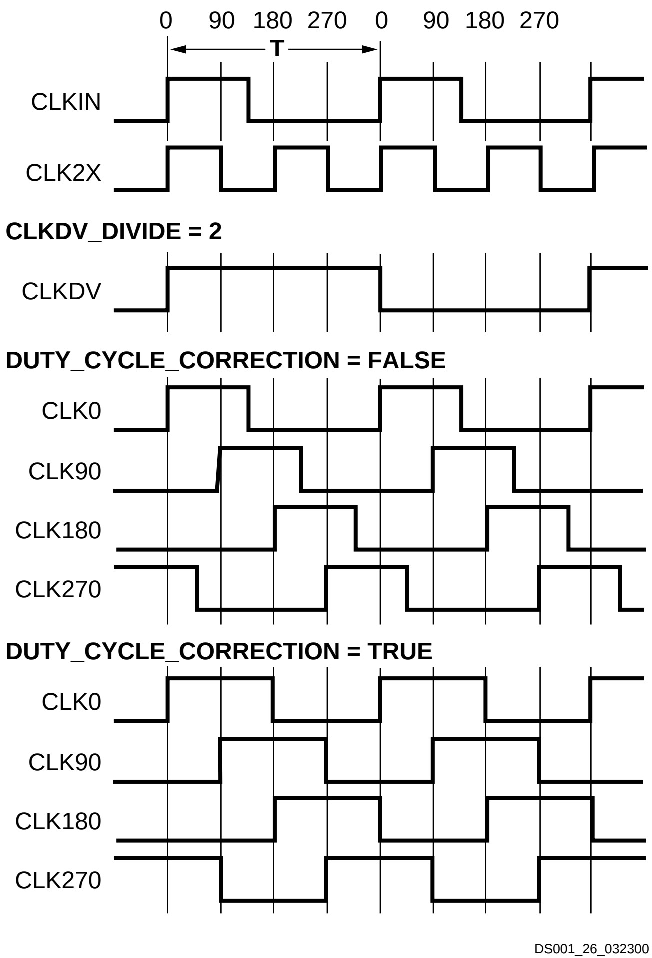


Figure 26: **DLL Output Characteristics**

Clock Divide Property

The CLKDV_DIVIDE property specifies how the signal on the CLKDV pin is frequency divided with respect to the CLK0 pin. The values allowed for this property are 1.5, 2, 2.5, 3, 4, 5, 8, or 16; the default value is 2.

Useful Application Examples

The Spartan-II FPGA DLL can be used in a variety of creative and useful applications. The following examples show some of the more common applications.

Standard Usage

The circuit shown in Figure 28 resembles the BUFGDLL macro implemented to provide access to the RST and LOCKED pins of the CLKDLL.

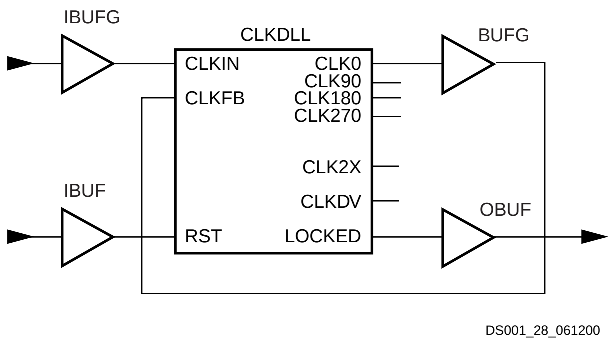


Figure 28: **Standard DLL Implementation**

Deskew of Clock and Its 2x Multiple

The circuit shown in Figure 29 implements a 2x clock multiplier and also uses the CLK0 clock output with zero ns skew between registers on the same chip. A clock divider circuit could alternatively be implemented using similar connections.

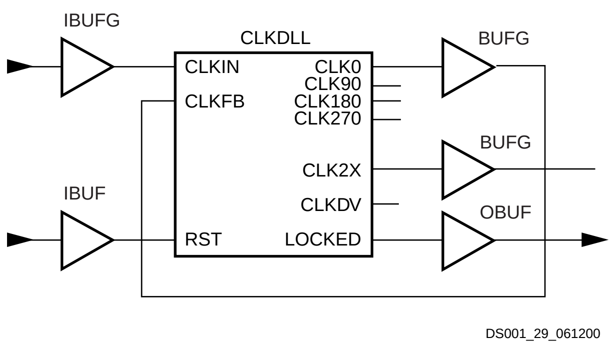


Figure 29: **DLL Deskew of Clock and 2x Multiple**

Because any single DLL can only access at most two BUFGs, any additional output clock signals must be routed from the DLL in this example on the high speed backbone routing.

Generating a 4x Clock

By connecting two DLL circuits each implementing a 2x clock multiplier in series as shown in Figure 30, a 4x clock multiply can be implemented with zero skew between registers in the same device.

If other clock output is needed, the clock could access a BUFG only if the DLLs are constrained to exist on opposite edges (Top or Bottom) of the device.

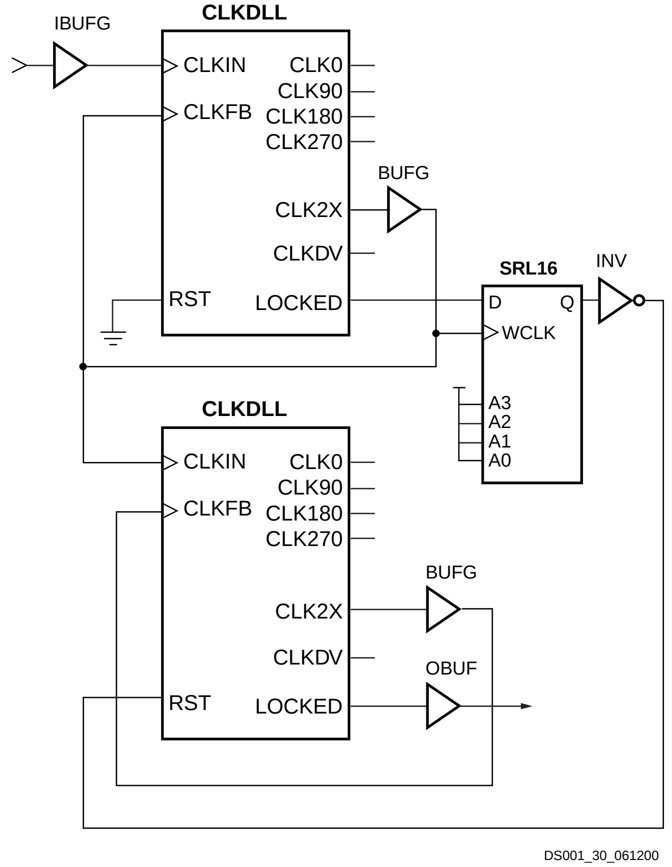


Figure 30: **DLL Generation of 4x Clock**

When using this circuit it is vital to use the SRL16 cell to reset the second DLL after the initial chip reset. If this is not done, the second DLL may not recognize the change of frequencies from when the input changes from a 1x (25/75) waveform to a 2x (50/50) waveform. It is not recommended to cascade more than two DLLs.

For design examples and more information on using the DLL, see XAPP174, *Using Delay-Locked Loops in Spartan-II FPGAs*.

Using Block RAM Features

The Spartan-II FPGA family provides dedicated blocks of on-chip, true dual-read/write port synchronous RAM, with 4096 memory cells. Each port of the block RAM memory can be independently configured as a read/write port, a read port, a write port, and can be configured to a specific data width. The block RAM memory offers new capabilities allowing the FPGA designer to simplify designs.

Operating Modes

Block RAM memory supports two operating modes.

- Read Through
- Write Back

Read Through (One Clock Edge)

The read address is registered on the read port clock edge and data appears on the output after the RAM access time. Some memories may place the latch/register at the outputs depending on the desire to have a faster clock-to-out versus setup time. This is generally considered to be an inferior solution since it changes the read operation to an asynchronous function with the possibility of missing an address/control line transition during the generation of the read pulse clock.

Write Back (One Clock Edge)

The write address is registered on the write port clock edge and the data input is written to the memory and mirrored on the write port input.

Block RAM Characteristics

1. All inputs are registered with the port clock and have a setup to clock timing specification.
2. All outputs have a read through or write back function depending on the state of the port WE pin. The outputs relative to the port clock are available after the clock-to-out timing specification.
3. The block RAM are true SRAM memories and do not have a combinatorial path from the address to the output. The LUT cells in the CLBs are still available with this function.
4. The ports are completely independent from each other (*i.e.,* clocking, control, address, read/write function, and data width) without arbitration.
5. A write operation requires only one clock edge.
6. A read operation requires only one clock edge.

The output ports are latched with a self timed circuit to guarantee a glitch free read. The state of the output port will not change until the port executes another read or write operation.

Library Primitives

Figure 31 and Figure 32 show the two generic library block RAM primitives. Table 11 describes all of the available primitives for synthesis and simulation.

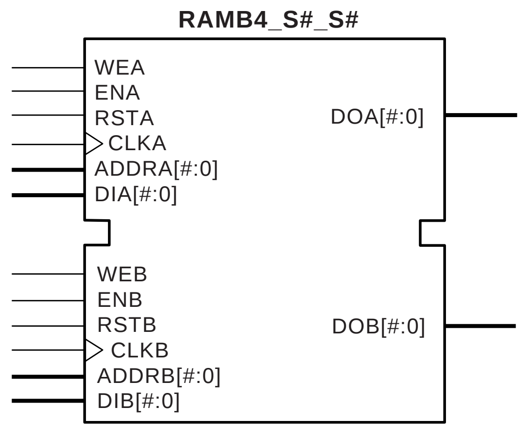


Figure 31: **Dual-Port Block RAM Memory**

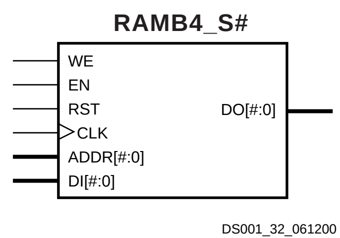


Figure 32: **Single-Port Block RAM Memory**

Table 11: **Available Library Primitives**

Primitive	Port A Width	Port B Width
RAMB4_S1	1	N/A
RAMB4_S1_S1		1
RAMB4_S1_S2		2
RAMB4_S1_S4		4
RAMB4_S1_S8		8
RAMB4_S1_S16		16
RAMB4_S2	2	N/A
RAMB4_S2_S2		2
RAMB4_S2_S4		4
RAMB4_S2_S8		8
RAMB4_S2_S16		16

support of a wide variety of applications, from general purpose standard applications to high-speed low-voltage memory busses.

Versatile I/O blocks also provide selectable output drive strengths and programmable slew rates for the LVTTL output buffers, as well as an optional, programmable weak pull-up, weak pull-down, or weak "keeper" circuit ideal for use in external bussing applications.

Each Input/Output Block (IOB) includes three registers, one each for the input, output, and 3-state signals within the IOB. These registers are optionally configurable as either a D-type flip-flop or as a level sensitive latch.

The input buffer has an optional delay element used to guarantee a zero hold time requirement for input signals registered within the IOB.

The Versatile I/O features also provide dedicated resources for input reference voltage (V_{REF}) and output source voltage (V_{CCO}), along with a convenient banking system that simplifies board design.

By taking advantage of the built-in features and wide variety of I/O standards supported by the Versatile I/O features, system-level design and board design can be greatly simplified and improved.

Fundamentals

Modern bus applications, pioneered by the largest and most influential companies in the digital electronics industry, are commonly introduced with a new I/O standard tailored specifically to the needs of that application. The bus I/O standards provide specifications to other vendors who create products designed to interface with these applications. Each standard often has its own specifications for current, voltage, I/O buffering, and termination techniques.

The ability to provide the flexibility and time-to-market advantages of programmable logic is increasingly dependent on the capability of the programmable logic device to support an ever increasing variety of I/O standards

The Versatile I/O resources feature highly configurable input and output buffers which provide support for a wide variety of I/O standards. As shown in Table 15, each buffer type can support a variety of voltage requirements.

Table 15: **Versatile I/O Supported Standards (Typical Values)**

I/O Standard	Input Reference Voltage (V_{REF})	Output Source Voltage (V_{CCO})	Board Termination Voltage (V_{TT})
LVTTL (2-24 mA)	N/A	3.3	N/A
LVCMOS2	N/A	2.5	N/A
PCI (3V/5V, 33 MHz/66 MHz)	N/A	3.3	N/A
GTL	0.8	N/A	1.2
GTL+	1.0	N/A	1.5
HSTL Class I	0.75	1.5	0.75
HSTL Class III	0.9	1.5	1.5
HSTL Class IV	0.9	1.5	1.5
SSTL3 Class I and II	1.5	3.3	1.5
SSTL2 Class I and II	1.25	2.5	1.25
CTT	1.5	3.3	1.5
AGP-2X	1.32	3.3	N/A

Overview of Supported I/O Standards

This section provides a brief overview of the I/O standards supported by all Spartan-II devices.

While most I/O standards specify a range of allowed voltages, this document records typical voltage values only. Detailed information on each specification may be found on the Electronic Industry Alliance JEDEC website at http://www.jedec.org. For more details on the I/O standards and termination application examples, see XAPP179, "*Using SelectIO Interfaces in Spartan-II and Spartan-IIE FPGAs.*"

LVTTL — Low-Voltage TTL

The Low-Voltage TTL (LVTTL) standard is a general purpose EIA/JESDSA standard for 3.3V applications that uses an LVTTL input buffer and a Push-Pull output buffer. This standard requires a 3.3V output source voltage (V_{CCO}), but does not require the use of a reference voltage (V_{REF}) or a termination voltage (V_{TT}).

LVCMOS2 — Low-Voltage CMOS for 2.5V

The Low-Voltage CMOS for 2.5V or lower (LVCMOS2) standard is an extension of the LVCMOS standard (JESD 8.5) used for general purpose 2.5V applications. This standard requires a 2.5V output source voltage (V_{CCO}), but does not require the use of a reference voltage (V_{REF}) or a board termination voltage (V_{TT}).

HSTL Class III

A sample circuit illustrating a valid termination technique for HSTL_III appears in Figure 45. DC voltage specifications appear in Table 23 for the HSTL_III standard. See "DC Specifications" in Module 3 for the actual FPGA characteristics.

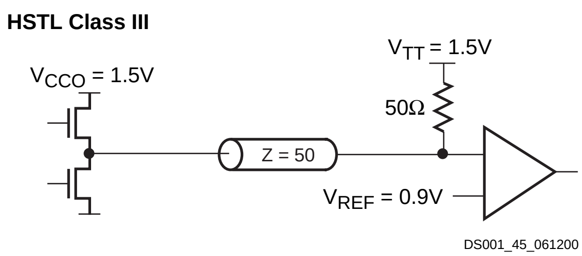


Figure 45: **Terminated HSTL Class III**

Table 23: **HSTL Class III Voltage Specification**

Parameter	Min	Typ	Max
V_{CCO}	1.40	1.50	1.60
V_{REF} (1)	-	0.90	-
V_{TT}	-	V_{CCO}	-
V_{IH}	V_{REF} + 0.1	-	-
V_{IL}	-	-	V_{REF} – 0.1
V_{OH}	V_{CCO} – 0.4	-	-
V_{OL}	-	-	0.4
I_{OH} at V_{OH} (mA)	–8	-	-
I_{OL} at V_{OL} (mA)	24	-	-

Notes:
1. Per EIA/JESD8-6, "The value of V_{REF} is to be selected by the user to provide optimum noise margin in the use conditions specified by the user."

HSTL Class IV

A sample circuit illustrating a valid termination technique for HSTL_IV appears in Figure 46.DC voltage specifications appear in Table 23 for the HSTL_IV standard. See "DC Specifications" in Module 3 for the actual FPGA characteristics

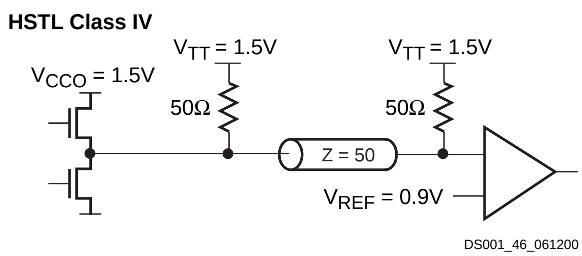


Figure 46: **Terminated HSTL Class IV**

Table 24: **HSTL Class IV Voltage Specification**

Parameter	Min	Typ	Max
V_{CCO}	1.40	1.50	1.60
V_{REF}	-	0.90	-
V_{TT}	-	V_{CCO}	-
V_{IH}	V_{REF} + 0.1	-	-
V_{IL}	-	-	V_{REF} – 0.1
V_{OH}	V_{CCO} – 0.4	-	-
V_{OL}	-	-	0.4
I_{OH} at V_{OH} (mA)	–8	-	-
I_{OL} at V_{OL} (mA)	48	-	-

Notes:
1. Per EIA/JESD8-6, "The value of V_{REF} is to be selected by the user to provide optimum noise margin in the use conditions specified by the user."

Revision History

Date	Version	Description
09/18/00	2.0	Sectioned the Spartan-II Family data sheet into four modules. Corrected banking description.
03/05/01	2.1	Clarified guidelines for applying power to V_{CCINT} and V_{CCO}
09/03/03	2.2	The following changes were made: • "Serial Modes," page 20 cautions about toggling $\overline{\text{WRITE}}$ during serial configuration. • Maximum V_{IH} values in Table 32 and Table 33 changed to 5.5V. • In "Boundary Scan," page 13, removed sentence about lack of INTEST support. • In Table 9, page 17, added note about the state of I/Os after power-on. • In "Slave Parallel Mode," page 23, explained configuration bit alignment to SelectMap port.
06/13/08	2.8	Added note that TDI, TMS, and TCK have a default pull-up resistor. Added note on maximum daisy chain limit. Updated Figure 15 and Figure 18 since Mode pins can be pulled up to either 2.5V or 3.3V. Updated DLL section. Recommended using property or attribute instead of primitive to define I/O properties. Updated description and links. Updated all modules for continuous page, figure, and table numbering. Synchronized all modules to v2.8.

IOB Input Delay Adjustments for Different Standards[1]

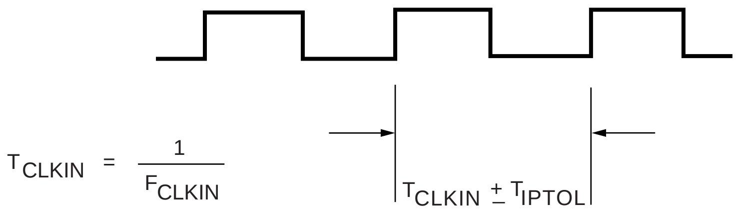
Input delays associated with the pad are specified for LVTTL. For other standards, adjust the delays by the values shown. A delay adjusted in this way constitutes a worst-case limit.

Symbol	Description	Standard	Speed Grade -6	Speed Grade -5	Units
Data Input Delay Adjustments					
T_{ILVTTL}	Standard-specific data input delay adjustments	LVTTL	0	0	ns
$T_{ILVCMOS2}$		LVCMOS2	–0.04	–0.05	ns
T_{IPCI33_3}		PCI, 33 MHz, 3.3V	–0.11	–0.13	ns
T_{IPCI33_5}		PCI, 33 MHz, 5.0V	0.26	0.30	ns
T_{IPCI66_3}		PCI, 66 MHz, 3.3V	–0.11	–0.13	ns
T_{IGTL}		GTL	0.20	0.24	ns
T_{IGTLP}		GTL+	0.11	0.13	ns
T_{IHSTL}		HSTL	0.03	0.04	ns
T_{ISSTL2}		SSTL2	–0.08	–0.09	ns
T_{ISSTL3}		SSTL3	–0.04	–0.05	ns
T_{ICTT}		CTT	0.02	0.02	ns
T_{IAGP}		AGP	–0.06	–0.07	ns

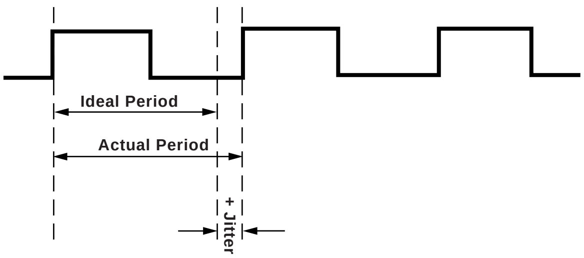
Notes:

1. Input timing for LVTTL is measured at 1.4V. For other I/O standards, see the table "Delay Measurement Methodology," page 60.

Period Tolerance: the allowed input clock period change in nanoseconds.



Output Jitter: the difference between an ideal reference clock edge and the actual design.



Phase Offset and Maximum Phase Difference

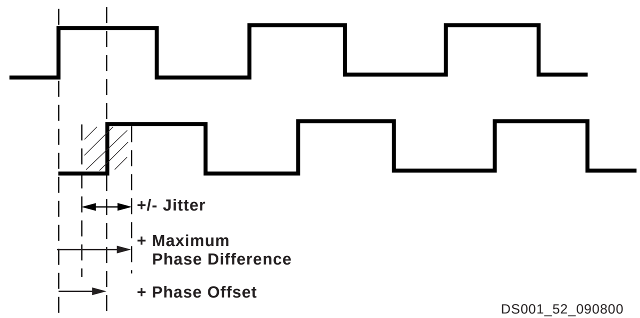


Figure 52: **Period Tolerance and Clock Jitter**

CLB Distributed RAM Switching Characteristics

Symbol	Description	Speed Grade -6 Min	Speed Grade -6 Max	Speed Grade -5 Min	Speed Grade -5 Max	Units
Sequential Delays						
$T_{SHCKO16}$	Clock CLK to X/Y outputs (WE active, 16 x 1 mode)	-	2.2	-	2.6	ns
$T_{SHCKO32}$	Clock CLK to X/Y outputs (WE active, 32 x 1 mode)	-	2.5	-	3.0	ns
Setup/Hold Times with Respect to Clock CLK(1)						
T_{AS} / T_{AH}	F/G address inputs	0.7 / 0	-	0.7 / 0	-	ns
T_{DS} / T_{DH}	BX/BY data inputs (DIN)	0.8 / 0	-	0.9 / 0	-	ns
T_{WS} / T_{WH}	CE input (WS)	0.9 / 0	-	1.0 / 0	-	ns
Clock CLK						
T_{WPH}	Minimum pulse width, High	-	2.9	-	2.9	ns
T_{WPL}	Minimum pulse width, Low	-	2.9	-	2.9	ns
T_{WC}	Minimum clock period to meet address write cycle time	-	5.8	-	5.8	ns

Notes:
1. A zero hold time listing indicates no hold time or a negative hold time.

CLB Shift Register Switching Characteristics

Symbol	Description	Speed Grade -6 Min	Speed Grade -6 Max	Speed Grade -5 Min	Speed Grade -5 Max	Units
Sequential Delays						
T_{REG}	Clock CLK to X/Y outputs	-	3.47	-	3.88	ns
Setup Times with Respect to Clock CLK						
T_{SHDICK}	BX/BY data inputs (DIN)	0.8	-	0.9	-	ns
T_{SHCECK}	CE input (WS)	0.9	-	1.0	-	ns
Clock CLK						
T_{SRPH}	Minimum pulse width, High	-	2.9	-	2.9	ns
T_{SRPL}	Minimum pulse width, Low	-	2.9	-	2.9	ns

XC2S50 Device Pinouts *(Continued)*

XC2S50 Pad Name					
Function	**Bank**	**TQ144**	**PQ208**	**FG256**	**Bndry Scan**
I/O	5	P99	P63	P6	326
GND	-	P98	P64	GND*	-
V_{CCO}	5	-	P65	V_{CCO} Bank 5*	-
V_{CCINT}	-	P97	P66	V_{CCINT}*	-
I/O	5	P96	P67	R6	329
I/O	5	P95	P68	M7	332
I/O	5	-	P69	N7	338
I/O	5	-	P70	T6	341
I/O	5	-	P71	P7	344
GND	-	-	P72	GND*	-
I/O, V_{REF}	5	P94	P73	P8	347
I/O	5	-	P74	R7	350
I/O	5	-	-	T7	353
I/O	5	P93	P75	T8	356
V_{CCINT}	-	P92	P76	V_{CCINT}*	-
I, GCK1	5	P91	P77	R8	365
V_{CCO}	5	P90	P78	V_{CCO} Bank 5*	-
V_{CCO}	4	P90	P78	V_{CCO} Bank 4*	-
GND	-	P89	P79	GND*	-
I, GCK0	4	P88	P80	N8	366
I/O	4	P87	P81	N9	370
I/O	4	P86	P82	R9	373
I/O	4	-	-	N10	376
I/O	4	-	P83	T9	379
I/O, V_{REF}	4	P85	P84	P9	382
GND	-	-	P85	GND*	-
I/O	4	-	P86	M10	385
I/O	4	-	P87	R10	388
I/O	4	-	P88	P10	391
I/O	4	P84	P89	T10	397
I/O	4	P83	P90	R11	400
V_{CCINT}	-	P82	P91	V_{CCINT}*	-
V_{CCO}	4	-	P92	V_{CCO} Bank 4*	-
GND	-	P81	P93	GND*	-
I/O	4	P80	P94	M11	403
I/O	4	P79	P95	T11	406
I/O	4	P78	P96	N11	409
I/O	4	-	-	R12	412

XC2S50 Device Pinouts *(Continued)*

XC2S50 Pad Name					
Function	**Bank**	**TQ144**	**PQ208**	**FG256**	**Bndry Scan**
I/O	4	-	P97	P11	415
I/O, V_{REF}	4	P77	P98	T12	418
GND	-	-	-	GND*	-
I/O	4	-	P99	T13	421
I/O	4	-	-	N12	424
I/O	4	P76	P100	R13	427
I/O	4	-	-	P12	430
I/O	4	P75	P101	P13	433
I/O	4	P74	P102	T14	436
GND	-	P73	P103	GND*	-
DONE	3	P72	P104	R14	439
V_{CCO}	4	P71	P105	V_{CCO} Bank 4*	-
V_{CCO}	3	P70	P105	V_{CCO} Bank 3*	-
$\overline{\text{PROGRAM}}$	-	P69	P106	P15	442
I/O ($\overline{\text{INIT}}$)	3	P68	P107	N15	443
I/O (D7)	3	P67	P108	N14	446
I/O	3	-	-	T15	449
I/O	3	P66	P109	M13	452
I/O	3	-	-	R16	455
I/O	3	-	P110	M14	458
GND	-	-	-	GND*	-
I/O, V_{REF}	3	P65	P111	L14	461
I/O	3	-	P112	M15	464
I/O	3	-	-	L12	467
I/O	3	P64	P113	P16	470
I/O	3	P63	P114	L13	473
I/O (D6)	3	P62	P115	N16	476
GND	-	P61	P116	GND*	-
V_{CCO}	3	-	P117	V_{CCO} Bank 3*	-
V_{CCINT}	-	-	P118	V_{CCINT}*	-
I/O (D5)	3	P60	P119	M16	479
I/O	3	P59	P120	K14	482
I/O	3	-	-	L16	485
I/O	3	-	P121	K13	488
I/O	3	-	P122	L15	491
I/O	3	-	P123	K12	494
GND	-	-	P124	GND*	-
I/O, V_{REF}	3	P58	P125	K16	497
I/O (D4)	3	P57	P126	J16	500

XC2S100 Device Pinouts *(Continued)*

XC2S100 Pad Name						
Function	**Bank**	**TQ144**	**PQ208**	**FG256**	**FG456**	**Bndry Scan**
I/O	2	-	-	F12	G20	695
I/O	2	-	P149	E15	F19	701
I/O, V_{REF}	2	P41	P150	F13	F21	704
V_{CCO}	2	-	-	V_{CCO} Bank 2*	V_{CCO} Bank 2*	-
GND	-	-	-	GND*	GND*	-
I/O	2	-	P151	E14	F20	707
I/O	2	-	-	C16	F18	710
I/O	2	-	-	-	E21	713
I/O	2	P40	P152	E13	D22	716
I/O	2	-	-	B16	E20	719
I/O (DIN, D0)	2	P39	P153	D14	D20	725
I/O (DOUT, BUSY)	2	P38	P154	C15	C21	728
CCLK	2	P37	P155	D15	B22	731
V_{CCO}	2	P36	P156	V_{CCO} Bank 2*	V_{CCO} Bank 2*	-
V_{CCO}	1	P35	P156	V_{CCO} Bank 1*	V_{CCO} Bank 1*	-
TDO	2	P34	P157	B14	A21	-
GND	-	P33	P158	GND*	GND*	-
TDI	-	P32	P159	A15	B20	-
I/O ($\overline{CS}$)	1	P31	P160	B13	C19	0
I/O ($\overline{WRITE}$)	1	P30	P161	C13	A20	3
I/O	1	-	-	C12	D17	9
I/O	1	P29	P162	A14	A19	12
I/O	1	-	-	-	B18	15
I/O	1	-	-	D12	C17	18
I/O	1	-	P163	B12	D16	21
GND	-	-	-	GND*	GND*	-
V_{CCO}	1	-	-	V_{CCO} Bank 1*	V_{CCO} Bank 1*	-
I/O, V_{REF}	1	P28	P164	C11	A18	24
I/O	1	-	P165	A13	B17	27
I/O	1	-	-	D11	D15	33
I/O	1	-	P166	A12	C16	36
I/O	1	-	-	-	D14	39
I/O, V_{REF}	1	P27	P167	E11	E14	42
I/O	1	P26	P168	B11	A16	45
GND	-	P25	P169	GND*	GND*	-

XC2S100 Device Pinouts *(Continued)*

XC2S100 Pad Name						
Function	**Bank**	**TQ144**	**PQ208**	**FG256**	**FG456**	**Bndry Scan**
V_{CCO}	1	-	P170	V_{CCO} Bank 1*	V_{CCO} Bank 1*	-
V_{CCINT}	-	P24	P171	V_{CCINT}*	V_{CCINT}*	-
I/O	1	P23	P172	A11	C15	48
I/O	1	P22	P173	C10	B15	51
I/O	1	-	-	-	F12	54
I/O	1	-	P174	B10	C14	57
I/O	1	-	P175	D10	D13	63
I/O	1	-	P176	A10	C13	66
GND	-	-	P177	GND*	GND*	-
I/O, V_{REF}	1	P21	P178	B9	B13	69
I/O	1	-	P179	E10	E12	72
I/O	1	-	-	A9	B12	75
I/O	1	P20	P180	D9	D12	78
I/O	1	P19	P181	A8	D11	84
I, GCK2	1	P18	P182	C9	A11	90
GND	-	P17	P183	GND*	GND*	-
V_{CCO}	1	P16	P184	V_{CCO} Bank 1*	V_{CCO} Bank 1*	-
V_{CCO}	0	P16	P184	V_{CCO} Bank 0*	V_{CCO} Bank 0*	-
I, GCK3	0	P15	P185	B8	C11	91
V_{CCINT}	-	P14	P186	V_{CCINT}*	V_{CCINT}*	-
I/O	0	P13	P187	A7	A10	101
I/O	0	-	-	D8	B10	104

XC2S150 Device Pinouts *(Continued)*

XC2S150 Pad Name					
Function	Bank	PQ208	FG256	FG456	Bndry Scan
I/O	6	P46	P1	T4	404
I/O	6	-	L5	W1	407
I/O	6	-	-	V2	410
I/O	6	-	-	U4	413
I/O	6	P47	N2	Y1	416
GND	-	-	GND*	GND*	-
I/O	6	-	M4	W2	419
I/O	6	-	-	V3	422
I/O	6	-	-	V4	425
I/O	6	P48	R1	Y2	428
I/O	6	P49	M3	W3	431
M1	-	P50	P2	U5	434
GND	-	P51	GND*	GND*	-
M0	-	P52	N3	AB2	435
V_{CCO}	6	P53	V_{CCO} Bank 6*	V_{CCO} Bank 6*	-
V_{CCO}	5	P53	V_{CCO} Bank 5*	V_{CCO} Bank 5*	-
M2	-	P54	R3	Y4	436
I/O	5	-	-	W5	443
I/O	5	-	-	AB3	446
I/O	5	-	N5	V7	449
GND	-	-	GND*	GND*	-
I/O	5	P57	T2	Y6	452
I/O	5	-	-	AA4	455
I/O	5	-	-	AB4	458
I/O	5	-	P5	W6	461
I/O	5	P58	T3	Y7	464
GND	-	-	GND*	GND*	-
V_{CCO}	5	-	V_{CCO} Bank 5*	V_{CCO} Bank 5*	-
I/O, V_{REF}	5	P59	T4	AA5	467
I/O	5	P60	M6	AB5	470
I/O	5	-	-	V8	473
I/O	5	-	-	AA6	476
I/O	5	-	T5	AB6	479
I/O	5	P61	N6	AA7	482
I/O	5	-	-	W7	485
I/O, V_{REF}	5	P62	R5	W8	488
I/O	5	P63	P6	Y8	491
GND	-	P64	GND*	GND*	-

XC2S150 Device Pinouts *(Continued)*

XC2S150 Pad Name					
Function	Bank	PQ208	FG256	FG456	Bndry Scan
V_{CCO}	5	P65	V_{CCO} Bank 5*	V_{CCO} Bank 5*	-
V_{CCINT}	-	P66	V_{CCINT}*	V_{CCINT}*	-
I/O	5	P67	R6	AA8	494
I/O	5	P68	M7	V9	497
I/O	5	-	-	W9	503
I/O	5	-	-	AB9	506
I/O	5	P69	N7	Y9	509
I/O	5	-	-	V10	512
I/O	5	P70	T6	W10	518
I/O	5	P71	P7	AB10	521
GND	-	P72	GND*	GND*	-
V_{CCO}	5	-	V_{CCO} Bank 5*	V_{CCO} Bank 5*	-
I/O, V_{REF}	5	P73	P8	Y10	524
I/O	5	P74	R7	V11	527
I/O	5	-	T7	W11	530
I/O	5	P75	T8	AB11	533
I/O	5	-	-	U11	536
V_{CCINT}	-	P76	V_{CCINT}*	V_{CCINT}*	-
I, GCK1	5	P77	R8	Y11	545
V_{CCO}	5	P78	V_{CCO} Bank 5*	V_{CCO} Bank 5*	-
V_{CCO}	4	P78	V_{CCO} Bank 4*	V_{CCO} Bank 4*	-
GND	-	P79	GND*	GND*	-
I, GCK0	4	P80	N8	W12	546
I/O	4	P81	N9	U12	550
I/O	4	-	-	V12	553
I/O	4	P82	R9	Y12	556
I/O	4	-	N10	AA12	559
I/O	4	P83	T9	AB13	562
I/O, V_{REF}	4	P84	P9	AA13	565
V_{CCO}	4	-	V_{CCO} Bank 4*	V_{CCO} Bank 4*	-
GND	-	P85	GND*	GND*	-
I/O	4	P86	M10	Y13	568
I/O	4	P87	R10	V13	571
I/O	4	-	-	W14	577
I/O	4	P88	P10	AA14	580
I/O	4	-	-	V14	583
I/O	4	-	-	Y14	586
I/O	4	P89	T10	AB15	592

Additional XC2S150 Package Pins

PQ208

Not Connected Pins					
P55	P56	-	-	-	-

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FG256

V_{CCINT} Pins					
C3	C14	D4	D13	E5	E12
M5	M12	N4	N13	P3	P14
V_{CCO} Bank 0 Pins					
E8	F8	-	-	-	-
V_{CCO} Bank 1 Pins					
E9	F9	-	-	-	-
V_{CCO} Bank 2 Pins					
H11	H12	-	-	-	-
V_{CCO} Bank 3 Pins					
J11	J12	-	-	-	-
V_{CCO} Bank 4 Pins					
L9	M9	-	-	-	-
V_{CCO} Bank 5 Pins					
L8	M8	-	-	-	-
V_{CCO} Bank 6 Pins					
J5	J6	-	-	-	-
V_{CCO} Bank 7 Pins					
H5	H6	-	-	-	-
GND Pins					
A1	A16	B2	B15	F6	F7
F10	F11	G6	G7	G8	G9
G10	G11	H7	H8	H9	H10
J7	J8	J9	J10	K6	K7
K8	K9	K10	K11	L6	L7
L10	L11	R2	R15	T1	T16
Not Connected Pins					
P4	R4	-	-	-	-

11/02/00

Additional XC2S150 Package Pins *(Continued)*

FG456

V_{CCINT} Pins					
E5	E18	F6	F17	G7	G8
G9	G14	G15	G16	H7	H16
J7	J16	P7	P16	R7	R16
T7	T8	T9	T14	T15	T16
U6	U17	V5	V18	-	-
V_{CCO} Bank 0 Pins					
F7	F8	F9	F10	G10	G11
V_{CCO} Bank 1 Pins					
F13	F14	F15	F16	G12	G13
V_{CCO} Bank 2 Pins					
G17	H17	J17	K16	K17	L16
V_{CCO} Bank 3 Pins					
M16	N16	N17	P17	R17	T17
V_{CCO} Bank 4 Pins					
T12	T13	U13	U14	U15	U16
V_{CCO} Bank 5 Pins					
T10	T11	U7	U8	U9	U10
V_{CCO} Bank 6 Pins					
M7	N6	N7	P6	R6	T6
V_{CCO} Bank 7 Pins					
G6	H6	J6	K6	K7	L7
GND Pins					
A1	A22	B2	B21	C3	C20
J9	J10	J11	J12	J13	J14
K9	K10	K11	K12	K13	K14
L9	L10	L11	L12	L13	L14
M9	M10	M11	M12	M13	M14
N9	N10	N11	N12	N13	N14
P9	P10	P11	P12	P13	P14
Y3	Y20	AA2	AA21	AB1	AB22
Not Connected Pins					
A2	A6	A12	A13	A14	B11
B16	C2	C8	C9	D1	D4
D18	D19	E13	E17	E19	F11
G2	G22	H21	J1	J4	K2
K18	K19	L2	L19	M2	M17
M21	N1	P1	P5	P22	R3
R20	R22	U3	U18	V6	W4
W13	W15	W19	Y5	Y22	AA1
AA3	AA9	AA10	AA11	AA16	AB7
AB8	AB12	AB14	AB21	-	-

11/02/00

XC2S200 Device Pinouts *(Continued)*

XC2S200 Pad Name					Bndry
Function	Bank	PQ208	FG256	FG456	Scan
I, GCK0	4	P80	N8	W12	636
I/O	4	P81	N9	U12	640
I/O	4	-	-	V12	646
I/O	4	P82	R9	Y12	649
I/O	4	-	N10	AA12	652
I/O	4	-	-	W13	655
I/O	4	P83	T9	AB13	661
I/O, V_{REF}	4	P84	P9	AA13	664
V_{CCO}	4	-	V_{CCO} Bank 4*	V_{CCO} Bank 4*	-
GND	-	P85	GND*	GND*	-
I/O	4	P86	M10	Y13	667
I/O	4	P87	R10	V13	670
I/O	4	-	-	AB14	673
I/O	4	-	-	W14	676
I/O	4	P88	P10	AA14	679
GND	-	-	GND*	GND*	-
I/O	4	-	-	V14	682
I/O	4	-	-	Y14	685
I/O	4	-	-	W15	688
I/O	4	P89	T10	AB15	691
I/O	4	P90	R11	AA15	694
V_{CCINT}	-	P91	V_{CCINT}*	V_{CCINT}*	-
V_{CCO}	4	P92	V_{CCO} Bank 4*	V_{CCO} Bank 4*	-
GND	-	P93	GND*	GND*	-
I/O	4	P94	M11	Y15	697
I/O, V_{REF}	4	P95	T11	AB16	700
I/O	4	-	-	AB17	706
I/O	4	P96	N11	V15	709
GND	-	-	GND*	GND*	-
I/O	4	-	R12	Y16	712
I/O	4	-	-	AA17	715
I/O	4	-	-	W16	718
I/O	4	P97	P11	AB18	721
I/O, V_{REF}	4	P98	T12	AB19	724
V_{CCO}	4	-	V_{CCO} Bank 4*	V_{CCO} Bank 4*	-
GND	-	-	GND*	GND*	-
I/O	4	P99	T13	Y17	727
I/O	4	-	N12	V16	730
I/O	4	-	-	AA18	733

XC2S200 Device Pinouts *(Continued)*

XC2S200 Pad Name					Bndry
Function	Bank	PQ208	FG256	FG456	Scan
I/O	4	-	-	W17	739
I/O, V_{REF}	4	P100	R13	AB20	742
GND	-	-	GND*	GND*	-
I/O	4	-	P12	AA19	745
I/O	4	-	-	V17	748
I/O	4	-	-	Y18	751
I/O	4	P101	P13	AA20	757
I/O	4	P102	T14	W18	760
GND	-	P103	GND*	GND*	-
DONE	3	P104	R14	Y19	763
V_{CCO}	4	P105	V_{CCO} Bank 4*	V_{CCO} Bank 4*	-
V_{CCO}	3	P105	V_{CCO} Bank 3*	V_{CCO} Bank 3*	-
$\overline{\text{PROGRAM}}$	-	P106	P15	W20	766
I/O ($\overline{\text{INIT}}$)	3	P107	N15	V19	767
I/O (D7)	3	P108	N14	Y21	770
I/O	3	-	-	V20	776
I/O	3	-	-	AA22	779
I/O	3	-	T15	W21	782
GND	-	-	GND*	GND*	-
I/O, V_{REF}	3	P109	M13	U20	785
I/O	3	-	-	U19	788
I/O	3	-	-	V21	794
GND	-	-	GND*	GND*	-
I/O	3	-	R16	T18	797
I/O	3	P110	M14	W22	800
GND	-	-	GND*	GND*	-
V_{CCO}	3	-	V_{CCO} Bank 3*	V_{CCO} Bank 3*	-
I/O, V_{REF}	3	P111	L14	U21	803
I/O	3	P112	M15	T20	806
I/O	3	-	-	T19	809
I/O	3	-	-	V22	812
I/O	3	-	L12	T21	815
GND	-	-	GND*	GND*	-
I/O	3	P113	P16	R18	818
I/O	3	-	-	U22	821
I/O, V_{REF}	3	P114	L13	R19	827
I/O (D6)	3	P115	N16	T22	830
GND	-	P116	GND*	GND*	-