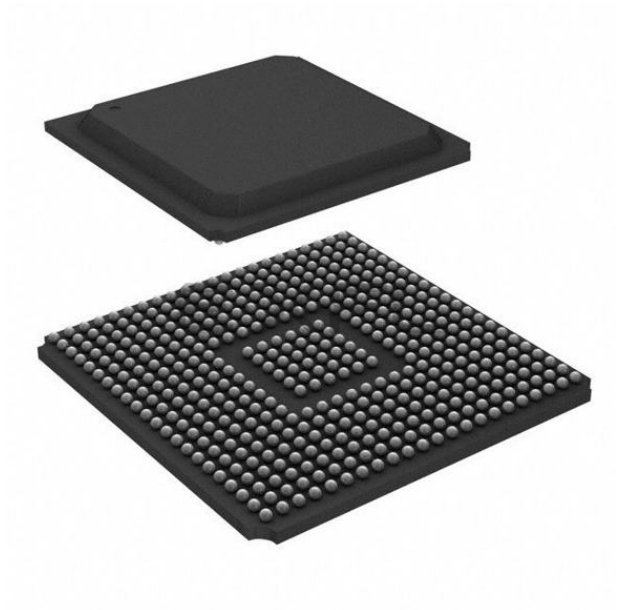


Welcome to E-XFL.COM

Understanding Embedded - FPGAs (Field Programmable Gate Array)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details	
Product Status	Active
Number of LABs/CLBs	1176
Number of Logic Elements/Cells	5292
Total RAM Bits	57344
Number of I/O	284
Number of Gates	200000
Voltage - Supply	2.375V ~ 2.625V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 85°C (TJ)
Package / Case	456-BBGA
Supplier Device Package	456-FBGA (23x23)
Purchase URL	https://www.e-xfl.com/product-detail/xilinx/xc2s200-6fg456c

Spartan-II Product Availability

Table 2 shows the maximum user I/Os available on the device and the number of user I/Os available for each device/package combination. The four global clock pins are usable as additional user I/Os when not used as a global clock pin. These pins are not included in user I/O counts.

Table 2: **Spartan-II FPGA User I/O Chart(1)**

Device	Maximum User I/O	Available User I/O According to Package Type					
		VQ100 VQG100	TQ144 TQG144	CS144 CSG144	PQ208 PQG208	FG256 FGG256	FG456 FGG456
XC2S15	86	60	86	(Note 2)	-	-	-
XC2S30	92	60	92	92	(Note 2)	-	-
XC2S50	176	-	92	-	140	176	-
XC2S100	176	-	92	-	140	176	(Note 2)
XC2S150	260	-	-	-	140	176	260
XC2S200	284	-	-	-	140	176	284

Notes:
1. All user I/O counts do not include the four global clock/user input pins.
2. Discontinued by PDN2004-01.

By default, these operations are synchronized to CCLK. The entire start-up sequence lasts eight cycles, called C0-C7, after which the loaded design is fully functional. The default timing for start-up is shown in the top half of Figure 13. The four operations can be selected to switch on any CCLK cycle C1-C6 through settings in the Xilinx software. Heavy lines show default settings.

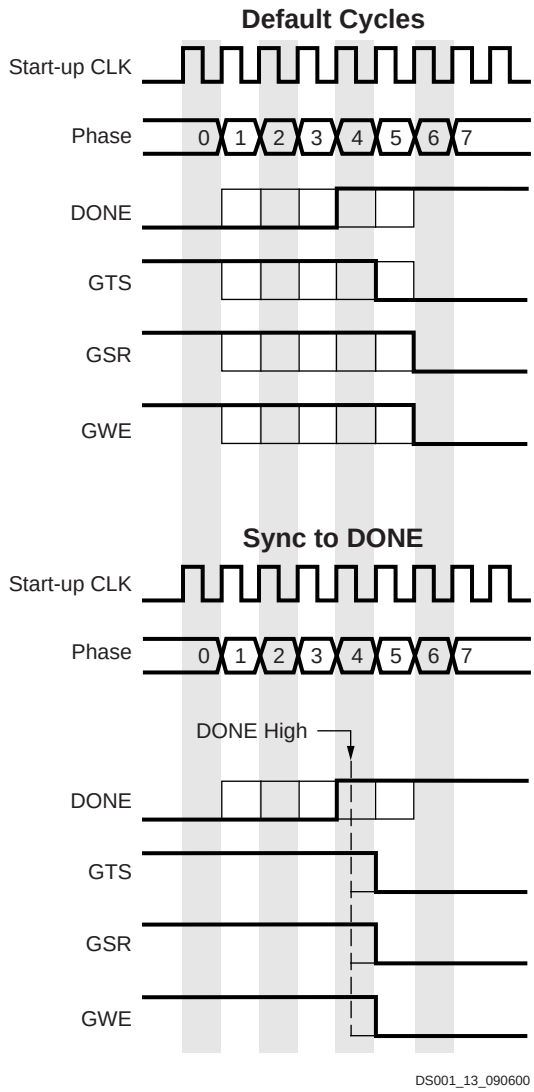


Figure 13: **Start-Up Waveforms**

The bottom half of Figure 13 shows another commonly used version of the start-up timing known as Sync-to-DONE. This version makes the GTS, GSR, and GWE events conditional upon the DONE pin going High. This timing is important for a daisy chain of multiple FPGAs in serial mode, since it ensures that all FPGAs go through start-up together, after all their DONE pins have gone High.

Sync-to-DONE timing is selected by setting the GTS, GSR, and GWE cycles to a value of DONE in the configuration options. This causes these signals to transition one clock cycle after DONE externally transitions High.

Serial Modes

There are two serial configuration modes: In Master Serial mode, the FPGA controls the configuration process by driving CCLK as an output. In Slave Serial mode, the FPGA passively receives CCLK as an input from an external agent (e.g., a microprocessor, CPLD, or second FPGA in master mode) that is controlling the configuration process. In both modes, the FPGA is configured by loading one bit per CCLK cycle. The MSB of each configuration data byte is always written to the DIN pin first.

See Figure 14 for the sequence for loading data into the Spartan-II FPGA serially. This is an expansion of the "Load Configuration Data Frames" block in Figure 11. Note that $\overline{CS}$ and $\overline{WRITE}$ normally are not used during serial configuration. To ensure successful loading of the FPGA, do not toggle $\overline{WRITE}$ with $\overline{CS}$ Low during serial configuration.

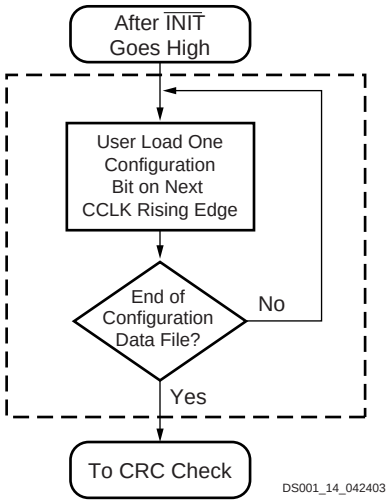
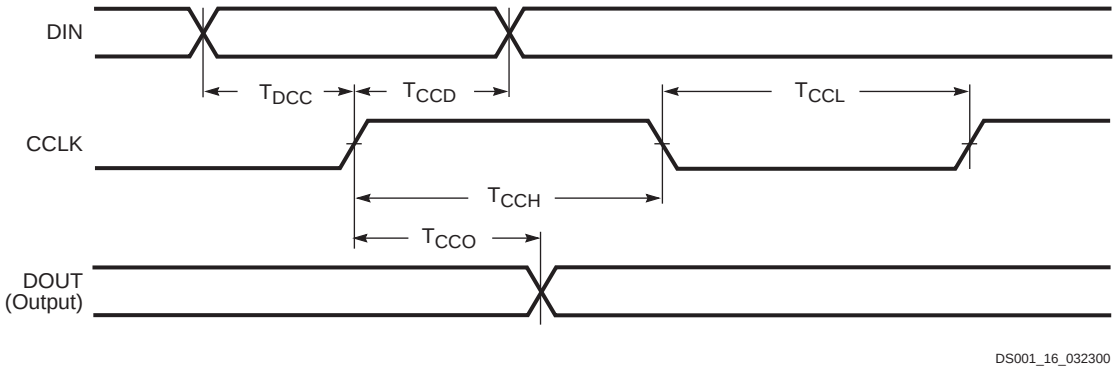


Figure 14: **Loading Serial Mode Configuration Data**



Symbol		Description		Units
T_{DCC}	CCLK	DIN setup	5	ns, min
T_{CCD}		DIN hold	0	ns, min
T_{CCO}		DOUT	12	ns, max
T_{CCH}		High time	5	ns, min
T_{CCL}		Low time	5	ns, min
F_{CC}		Maximum frequency	66	MHz, max

Figure 16: **Slave Serial Mode Timing**

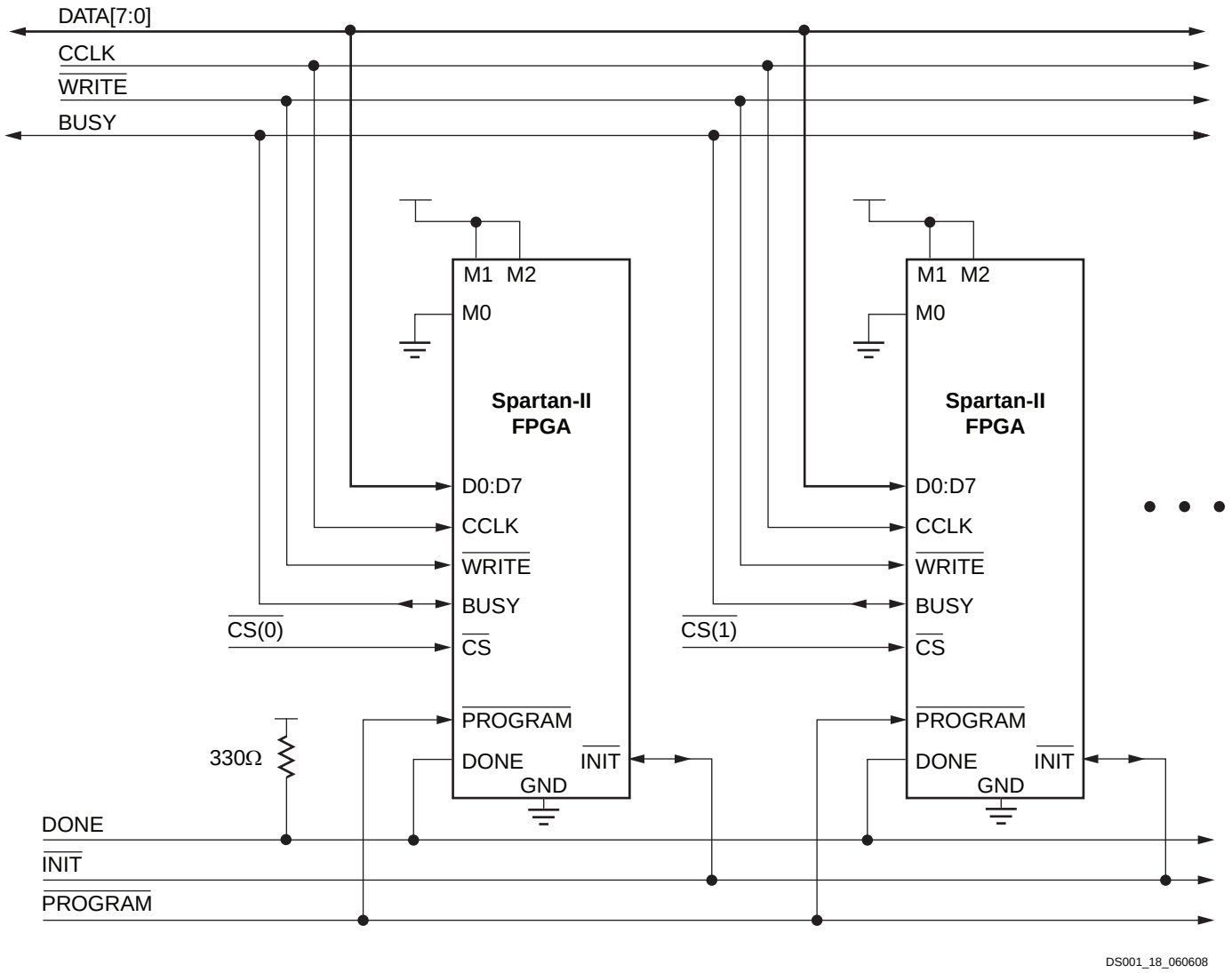


Figure 18: **Slave Parallel Configuration Circuit Diagram**

Multiple Spartan-II FPGAs can be configured using the Slave Parallel mode, and be made to start-up simultaneously. To configure multiple devices in this way, wire the individual CCLK, Data, $\overline{\text{WRITE}}$, and BUSY pins of all the devices in parallel. The individual devices are loaded separately by asserting the $\overline{\text{CS}}$ pin of each device in turn and writing the appropriate data. Sync-to-DONE start-up timing is used to ensure that the start-up sequence does not begin until all the FPGAs have been loaded. See "Start-up," page 19.

Write

When using the Slave Parallel Mode, write operations send packets of byte-wide configuration data into the FPGA. Figure 19, page 25 shows a flowchart of the write sequence used to load data into the Spartan-II FPGA. This is an expansion of the "Load Configuration Data Frames" block in Figure 11, page 18. The timing for write operations is shown in Figure 20, page 26.

For the present example, the user holds $\overline{\text{WRITE}}$ and $\overline{\text{CS}}$ Low throughout the sequence of write operations. Note that when $\overline{\text{CS}}$ is asserted on successive CCLKs, $\overline{\text{WRITE}}$ must remain either asserted or de-asserted. Otherwise an abort will be initiated, as in the next section.

1. Drive data onto D0-D7. Note that to avoid contention, the data source should not be enabled while $\overline{\text{CS}}$ is Low and $\overline{\text{WRITE}}$ is High. Similarly, while $\overline{\text{WRITE}}$ is High, no more than one device's $\overline{\text{CS}}$ should be asserted.
2. On the rising edge of CCLK: If BUSY is Low, the data is accepted on this clock. If BUSY is High (from a previous write), the data is not accepted. Acceptance will instead occur on the first clock after BUSY goes Low, and the data must be held until this happens.
3. Repeat steps 1 and 2 until all the data has been sent.
4. De-assert $\overline{\text{CS}}$ and $\overline{\text{WRITE}}$.

BUFGDLL Pin Descriptions

Use the BUFGDLL macro as the simplest way to provide zero propagation delay for a high-fanout on-chip clock from an external input. This macro uses the IBUFG, CLKDLL and BUFG primitives to implement the most basic DLL application as shown in Figure 25.

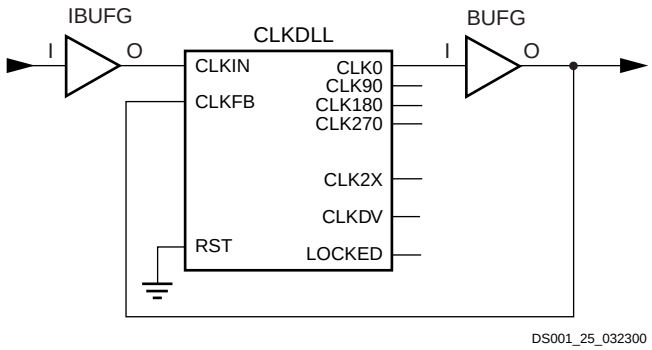


Figure 25: **BUFGDLL Block Diagram**

This macro does not provide access to the advanced clock domain controls or to the clock multiplication or clock division features of the DLL. This macro also does not provide access to the RST or LOCKED pins of the DLL. For access to these features, a designer must use the DLL primitives described in the following sections.

Source Clock Input — I

The I pin provides the user source clock, the clock signal on which the DLL operates, to the BUFGDLL. For the BUFGDLL macro the source clock frequency must fall in the low frequency range as specified in the data sheet. The BUFGDLL requires an external signal source clock. Therefore, only an external input port can source the signal that drives the BUFGDLL I pin.

Clock Output — O

The clock output pin O represents a delay-compensated version of the source clock (I) signal. This signal, sourced by a global clock buffer BUFG primitive, takes advantage of the dedicated global clock routing resources of the device.

The output clock has a 50/50 duty cycle unless you deactivate the duty cycle correction property.

CLKDLL Primitive Pin Descriptions

The library CLKDLL primitives provide access to the complete set of DLL features needed when implementing more complex applications with the DLL.

Source Clock Input — CLKIN

The CLKIN pin provides the user source clock (the clock signal on which the DLL operates) to the DLL. The CLKIN frequency must fall in the ranges specified in the data sheet. A global clock buffer (BUFG) driven from another CLKDLL or one of the global clock input buffers (IBUFG) on the same edge of the device (top or bottom) must source this clock signal.

Feedback Clock Input — CLKFB

The DLL requires a reference or feedback signal to provide the delay-compensated output. Connect only the CLK0 or CLK2X DLL outputs to the feedback clock input (CLKFB) pin to provide the necessary feedback to the DLL. Either a global clock buffer (BUFG) or one of the global clock input buffers (IBUFG) on the same edge of the device (top or bottom) must source this clock signal.

If an IBUFG sources the CLKFB pin, the following special rules apply.

1. An external input port must source the signal that drives the IBUFG I pin.
2. The CLK2X output must feed back to the device if both the CLK0 and CLK2X outputs are driving off chip devices.
3. That signal must directly drive only OBUFs and nothing else.

These rules enable the software to determine which DLL clock output sources the CLKFB pin.

Reset Input — RST

When the reset pin RST activates, the LOCKED signal deactivates within four source clock cycles. The RST pin, active High, must either connect to a dynamic signal or be tied to ground. As the DLL delay taps reset to zero, glitches can occur on the DLL clock output pins. Activation of the RST pin can also severely affect the duty cycle of the clock output pins. Furthermore, the DLL output clocks no longer deskew with respect to one another. The DLL must be reset when the input clock frequency changes, if the device is reconfigured in Boundary-Scan mode, if the device undergoes a hot swap, and after the device is configured if the input clock is not stable during the startup sequence.

2x Clock Output — CLK2X

The output pin CLK2X provides a frequency-doubled clock with an automatic 50/50 duty-cycle correction. Until the CLKDLL has achieved lock, the CLK2X output appears as a 1x version of the input clock with a 25/75 duty cycle. This behavior allows the DLL to lock on the correct edge with respect to source clock. This pin is not available on the CLKDLLHF primitive.

Clock Divide Output — CLKDV

The clock divide output pin CLKDV provides a lower frequency version of the source clock. The CLKDV_DIVIDE property controls CLKDV such that the source clock is divided by N where N is either 1.5, 2, 2.5, 3, 4, 5, 8, or 16.

This feature provides automatic duty cycle correction. The CLKDV output pin has a 50/50 duty cycle for all values of the

division factor N except for non-integer division in High Frequency (HF) mode. For division factor 1.5 the duty cycle in the HF mode is 33.3% High and 66.7% Low. For division factor 2.5, the duty cycle in the HF mode is 40.0% High and 60.0% Low.

1x Clock Outputs — CLK[0|90|180|270]

The 1x clock output pin CLK0 represents a delay-compensated version of the source clock (CLKIN) signal. The CLKDLL primitive provides three phase-shifted versions of the CLK0 signal while CLKDLLHF provides only the 180 degree phase-shifted version. The relationship between phase shift and the corresponding period shift appears in Table 10.

The timing diagrams in Figure 26 illustrate the DLL clock output characteristics.

Table 10: **Relationship of Phase-Shifted Output Clock to Period Shift**

Phase (degrees)	Period Shift (percent)
0	0%
90	25%
180	50%
270	75%

The DLL provides duty cycle correction on all 1x clock outputs such that all 1x clock outputs by default have a 50/50 duty cycle. The DUTY_CYCLE_CORRECTION property (TRUE by default), controls this feature. In order to deactivate the DLL duty cycle correction, attach the DUTY_CYCLE_CORRECTION=FALSE property to the DLL primitive. When duty cycle correction deactivates, the output clock has the same duty cycle as the source clock.

The DLL clock outputs can drive an OBUF, a BUFG, or they can route directly to destination clock pins. The DLL clock outputs can only drive the BUFGs that reside on the same edge (top or bottom).

Locked Output — LOCKED

In order to achieve lock, the DLL may need to sample several thousand clock cycles. After the DLL achieves lock the LOCKED signal activates. The "DLL Timing Parameters" section of Module 3 provides estimates for locking times.

In order to guarantee that the system clock is established prior to the device "waking up," the DLL can delay the completion of the device configuration process until after the DLL locks. The STARTUP_WAIT property activates this feature.

Until the LOCKED signal activates, the DLL output clocks are not valid and can exhibit glitches, spikes, or other spurious movement. In particular the CLK2X output will appear as a 1x clock with a 25/75 duty cycle.

DLL Properties

Properties provide access to some of the Spartan-II family DLL features, (for example, clock division and duty cycle correction).

Duty Cycle Correction Property

The 1x clock outputs, CLK0, CLK90, CLK180, and CLK270, use the duty-cycle corrected default, such that they exhibit a 50/50 duty cycle. The DUTY_CYCLE_CORRECTION property (by default TRUE) controls this feature. To deactivate the DLL duty-cycle correction for the 1x clock outputs, attach the DUTY_CYCLE_CORRECTION=FALSE property to the DLL primitive.

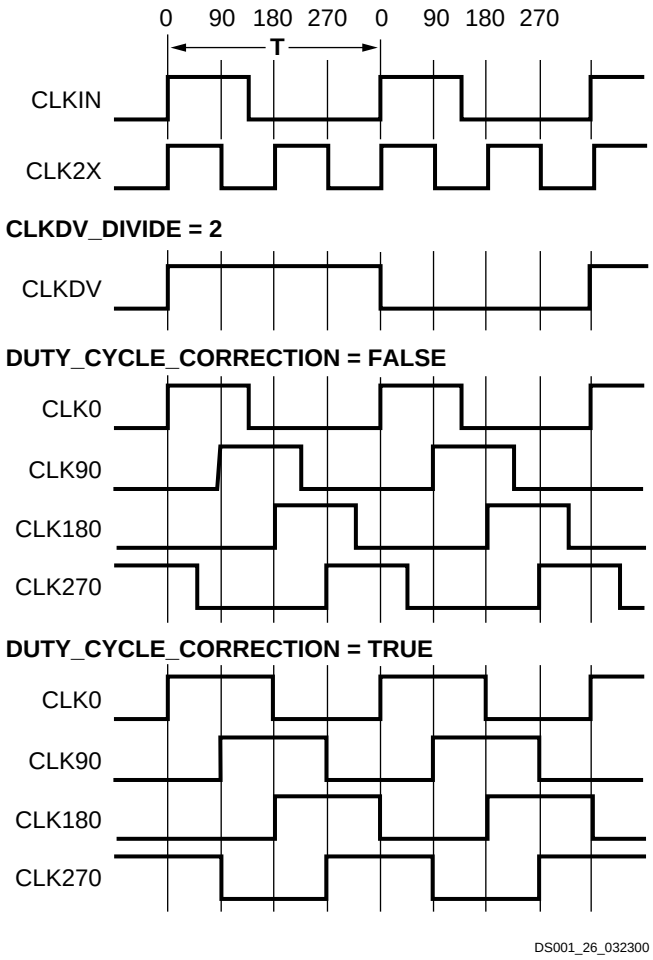


Figure 26: **DLL Output Characteristics**

Clock Divide Property

The CLKDV_DIVIDE property specifies how the signal on the CLKDV pin is frequency divided with respect to the CLK0 pin. The values allowed for this property are 1.5, 2, 2.5, 3, 4, 5, 8, or 16; the default value is 2.

Startup Delay Property

This property, STARTUP_WAIT, takes on a value of TRUE or FALSE (the default value). When TRUE the Startup Sequence following device configuration is paused at a user-specified point until the DLL locks. **XAPP176**: *Configuration and Readback of the Spartan-II and Spartan-IIE Families* explains how this can result in delaying the assertion of the DONE pin until the DLL locks.

DLL Location Constraints

The DLLs are distributed such that there is one DLL in each corner of the device. The location constraint LOC, attached to the DLL primitive with the numeric identifier 0, 1, 2, or 3, controls DLL location. The orientation of the four DLLs and their corresponding clock resources appears in Figure 27.

The LOC property uses the following form.

LOC = DLL2

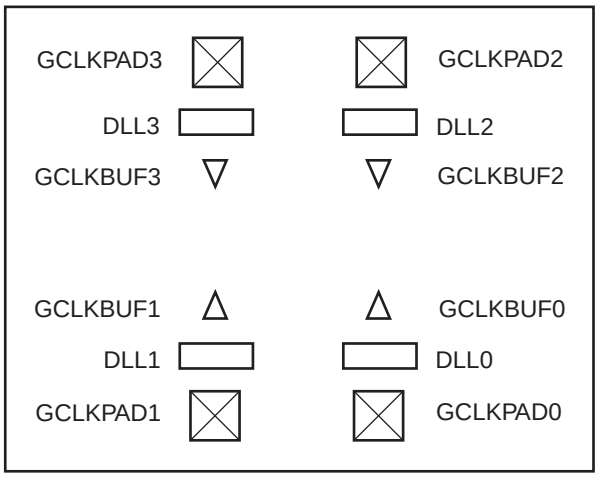


Figure 27: **Orientation of DLLs**

Design Considerations

Use the following design considerations to avoid pitfalls and improve success designing with Xilinx devices.

Input Clock

The output clock signal of a DLL, essentially a delayed version of the input clock signal, reflects any instability on the input clock in the output waveform. For this reason the quality of the DLL input clock relates directly to the quality of the output clock waveforms generated by the DLL. The DLL input clock requirements are specified in the "DLL Timing Parameters" section of the data sheet.

In most systems a crystal oscillator generates the system clock. The DLL can be used with any commercially available quartz crystal oscillator. For example, most crystal oscillators produce an output waveform with a frequency tolerance of 100 PPM, meaning 0.01 percent change in the clock period. The DLL operates reliably on an input waveform with a frequency drift of up to 1 ns — orders of magnitude in excess of that needed to support any crystal oscillator in the industry. However, the cycle-to-cycle jitter must be kept to less than 300 ps in the low frequencies and 150 ps for the high frequencies.

Input Clock Changes

Changing the period of the input clock beyond the maximum drift amount requires a manual reset of the CLKDLL. Failure to reset the DLL will produce an unreliable lock signal and output clock.

It is possible to stop the input clock in a way that has little impact to the DLL. Stopping the clock should be limited to less than approximately 100 µs to keep device cooling to a minimum and maintain the validity of the current tap setting. The clock should be stopped during a Low phase, and when restored the full High period should be seen. During this time LOCKED will stay High and remain High when the clock is restored. If these conditions may not be met in the design, apply a manual reset to the DLL after re-starting the input clock, even if the LOCKED signal has not changed.

When the clock is stopped, one to four more clocks will still be observed as the delay line is flushed. When the clock is restarted, the output clocks will not be observed for one to four clocks as the delay line is filled. The most common case will be two or three clocks.

In a similar manner, a phase shift of the input clock is also possible. The phase shift will propagate to the output one to four clocks after the original shift, with no disruption to the CLKDLL control.

Output Clocks

As mentioned earlier in the DLL pin descriptions, some restrictions apply regarding the connectivity of the output pins. The DLL clock outputs can drive an OBUF, a global clock buffer BUFG, or route directly to destination clock pins. The only BUFGs that the DLL clock outputs can drive are the two on the same edge of the device (top or bottom). One DLL output can drive more than one OBUF; however, this adds skew.

Do not use the DLL output clock signals until after activation of the LOCKED signal. Prior to the activation of the LOCKED signal, the DLL output clocks are not valid and can exhibit glitches, spikes, or other spurious movement.

Creating Larger RAM Structures

The block RAM columns have specialized routing to allow cascading blocks together with minimal routing delays. This achieves wider or deeper RAM structures with a smaller timing penalty than when using normal routing channels.

Location Constraints

Block RAM instances can have LOC properties attached to them to constrain the placement. The block RAM placement locations are separate from the CLB location naming convention, allowing the LOC properties to transfer easily from array to array.

The LOC properties use the following form:

LOC = RAMB4_R#C#

RAMB4_R0C0 is the upper left RAMB4 location on the device.

Conflict Resolution

The block RAM memory is a true dual-read/write port RAM that allows simultaneous access of the same memory cell from both ports. When one port writes to a given memory cell, the other port must not address that memory cell (for a write or a read) within the clock-to-clock setup window. The following lists specifics of port and memory cell write conflict resolution.

- If both ports write to the same memory cell simultaneously, violating the clock-to-clock setup requirement, consider the data stored as invalid.
- If one port attempts a read of the same memory cell the other simultaneously writes, violating the clock-to-clock setup requirement, the following occurs.
  - The write succeeds
  - The data out on the writing port accurately reflects the data written.
  - The data out on the reading port is invalid.

Conflicts do not cause any physical damage.

Single Port Timing

Figure 33 shows a timing diagram for a single port of a block RAM memory. The block RAM AC switching characteristics are specified in the data sheet. The block RAM memory is initially disabled.

At the first rising edge of the CLK pin, the ADDR, DI, EN, WE, and RST pins are sampled. The EN pin is High and the WE pin is Low indicating a read operation. The DO bus contains the contents of the memory location, 0x00, as indicated by the ADDR bus.

At the second rising edge of the CLK pin, the ADDR, DI, EN, WR, and RST pins are sampled again. The EN and WE pins are High indicating a write operation. The DO bus mirrors the DI bus. The DI bus is written to the memory location 0x0F.

At the third rising edge of the CLK pin, the ADDR, DI, EN, WR, and RST pins are sampled again. The EN pin is High and the WE pin is Low indicating a read operation. The DO bus contains the contents of the memory location 0x7E as indicated by the ADDR bus.

At the fourth rising edge of the CLK pin, the ADDR, DI, EN, WR, and RST pins are sampled again. The EN pin is Low indicating that the block RAM memory is now disabled. The DO bus retains the last value.

Dual Port Timing

Figure 34 shows a timing diagram for a true dual-port read/write block RAM memory. The clock on port A has a longer period than the clock on Port B. The timing parameter T_{BCCS}, (clock-to-clock setup) is shown on this diagram. The parameter, T_{BCCS} is violated once in the diagram. All other timing parameters are identical to the single port version shown in Figure 33.

T_{BCCS} is only of importance when the address of both ports are the same and at least one port is performing a write operation. When the clock-to-clock set-up parameter is violated for a WRITE-WRITE condition, the contents of the memory at that location will be invalid. When the clock-to-clock set-up parameter is violated for a WRITE-READ condition, the contents of the memory will be correct, but the read port will have invalid data. At the first rising edge of the CLKA, memory location 0x00 is to be written with the value 0xAAAA and is mirrored on the DOA bus. The last operation of Port B was a read to the same memory location 0x00. The DOB bus of Port B does not change with the new value on Port A, and retains the last read value. A short time later, Port B executes another read to memory location 0x00, and the DOB bus now reflects the new memory value written by Port A.

At the second rising edge of CLKA, memory location 0x7E is written with the value 0x9999 and is mirrored on the DOA bus. Port B then executes a read operation to the same memory location without violating the T_{BCCS} parameter and the DOB reflects the new memory values written by Port A.

PCI — Peripheral Component Interface

The Peripheral Component Interface (PCI) standard specifies support for both 33 MHz and 66 MHz PCI bus applications. It uses a LVTTL input buffer and a push-pull output buffer. This standard does not require the use of a reference voltage (V_{REF}) or a board termination voltage (V_{TT}), however, it does require a 3.3V output source voltage (V_{CCO}). I/Os configured for the PCI, 33 MHz, 5V standard are also 5V-tolerant.

GTL — Gunning Transceiver Logic Terminated

The Gunning Transceiver Logic (GTL) standard is a high-speed bus standard (JESD8.3). Xilinx has implemented the terminated variation of this standard. This standard requires a differential amplifier input buffer and an open-drain output buffer.

GTL+ — Gunning Transceiver Logic Plus

The Gunning Transceiver Logic Plus (GTL+) standard is a high-speed bus standard (JESD8.3).

HSTL — High-Speed Transceiver Logic

The High-Speed Transceiver Logic (HSTL) standard is a general purpose high-speed, 1.5V bus standard (EIA/JESD 8-6). This standard has four variations or classes. Versatile I/O devices support Class I, III, and IV. This standard requires a Differential Amplifier input buffer and a Push-Pull output buffer.

SSTL3 — Stub Series Terminated Logic for 3.3V

The Stub Series Terminated Logic for 3.3V (SSTL3) standard is a general purpose 3.3V memory bus standard (JESD8-8). This standard has two classes, I and II. Versatile I/O devices support both classes for the SSTL3 standard. This standard requires a Differential Amplifier input buffer and an Push-Pull output buffer.

SSTL2 — Stub Series Terminated Logic for 2.5V

The Stub Series Terminated Logic for 2.5V (SSTL2) standard is a general purpose 2.5V memory bus standard (JESD8-9). This standard has two classes, I and II. Versatile I/O devices support both classes for the SSTL2 standard. This standard requires a Differential Amplifier input buffer and an Push-Pull output buffer.

CTT — Center Tap Terminated

The Center Tap Terminated (CTT) standard is a 3.3V memory bus standard (JESD8-4). This standard requires a Differential Amplifier input buffer and a Push-Pull output buffer.

AGP-2X — Advanced Graphics Port

The AGP standard is a 3.3V Advanced Graphics Port-2X bus standard used with processors for graphics applications. This standard requires a Push-Pull output buffer and a Differential Amplifier input buffer.

Library Primitives

The Xilinx library includes an extensive list of primitives designed to provide support for the variety of Versatile I/O features. Most of these primitives represent variations of the five generic Versatile I/O primitives:

- IBUF (input buffer)
- IBUFG (global clock input buffer)
- OBUF (output buffer)
- OBUFT (3-state output buffer)
- IOBUF (input/output buffer)

These primitives are available with various extensions to define the desired I/O standard. However, it is recommended that customers use a a property or attribute on the generic primitive to specify the I/O standard. See "Versatile I/O Properties".

IBUF

Signals used as inputs to the Spartan-II device must source an input buffer (IBUF) via an external input port. The generic IBUF primitive appears in Figure 35. The assumed standard is LVTTL when the generic IBUF has no specified extension or property.

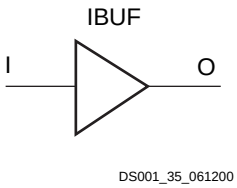


Figure 35: **Input Buffer (IBUF) Primitive**

When the IBUF primitive supports an I/O standard such as LVTTL, LVCMOS, or PCI33_5, the IBUF automatically configures as a 5V tolerant input buffer unless the V_{CCO} for the bank is less than 2V. If the single-ended IBUF is placed in a bank with an HSTL standard (V_{CCO} < 2V), the input buffer is not 5V tolerant.

The voltage reference signal is "banked" within the Spartan-II device on a half-edge basis such that for all packages there are eight independent V_{REF} banks internally. See Figure 36 for a representation of the I/O banks. Within each bank approximately one of every six I/O pins is automatically configured as a V_{REF} input.

IBUF placement restrictions require that any differential amplifier input signals within a bank be of the same standard. How to specify a specific location for the IBUF via

Recommended Operating Conditions

Symbol	Description		Min	Max	Units
T_J	Junction temperature(1)	Commercial	0	85	°C
		Industrial	–40	100	°C
V_{CCINT}	Supply voltage relative to GND(2,5)	Commercial	2.5 – 5%	2.5 + 5%	V
		Industrial	2.5 – 5%	2.5 + 5%	V
V_{CCO}	Supply voltage relative to GND(3,5)	Commercial	1.4	3.6	V
		Industrial	1.4	3.6	V
T_{IN}	Input signal transition time(4)		-	250	ns

Notes:
1. At junction temperatures above those listed as Operating Conditions, all delay parameters increase by 0.35% per °C.
2. Functional operation is guaranteed down to a minimum V_{CCINT} of 2.25V (Nominal V_{CCINT} – 10%). For every 50 mV reduction in V_{CCINT} below 2.375V (nominal V_{CCINT} – 5%), all delay parameters increase by 3%.
3. Minimum and maximum values for V_{CCO} vary according to the I/O standard selected.
4. Input and output measurement threshold is ~50% of V_{CCO}. See "Delay Measurement Methodology," page 60 for specific levels.
5. Supply voltages may be applied in any order desired.

DC Characteristics Over Operating Conditions

Symbol	Description			Min	Typ	Max	Units
V_{DRINT}	Data Retention V_{CCINT} voltage (below which configuration data may be lost)			2.0	-	-	V
V_{DRIO}	Data Retention V_{CCO} voltage (below which configuration data may be lost)			1.2	-	-	V
I_{CCINTQ}	Quiescent V_{CCINT} supply current(1)	XC2S15	Commercial	-	10	30	mA
			Industrial	-	10	60	mA
		XC2S30	Commercial	-	10	30	mA
			Industrial	-	10	60	mA
		XC2S50	Commercial	-	12	50	mA
			Industrial	-	12	100	mA
		XC2S100	Commercial	-	12	50	mA
			Industrial	-	12	100	mA
		XC2S150	Commercial	-	15	50	mA
			Industrial	-	15	100	mA
		XC2S200	Commercial	-	15	75	mA
			Industrial	-	15	150	mA
I_{CCOQ}	Quiescent V_{CCO} supply current(1)			-	-	2	mA
I_{REF}	V_{REF} current per V_{REF} pin			-	-	20	µA
I_L	Input or output leakage current(2)			–10	-	+10	µA
C_{IN}	Input capacitance (sample tested)	VQ, CS, TQ, PQ, FG packages		-	-	8	pF
I_{RPU}	Pad pull-up (when selected) @ V_{IN} = 0V, V_{CCO} = 3.3V (sample tested)(3)			-	-	0.25	mA
I_{RPD}	Pad pull-down (when selected) @ V_{IN} = 3.6V (sample tested)(3)			-	-	0.15	mA

Notes:
1. With no output current loads, no active input pull-up resistors, all I/O pins 3-stated and floating.
2. The I/O leakage current specification applies only when the V_{CCINT} and V_{CCO} supply voltages have reached their respective minimum Recommended Operating Conditions.
3. Internal pull-up and pull-down resistors guarantee valid logic levels at unconnected input pins. These pull-up and pull-down resistors do not provide valid logic levels when input pins are connected to other circuits.

Input/Output Standard	V_{IL}		V_{IH}		V_{OL}	V_{OH}	I_{OL}	I_{OH}
	V, Min	V, Max	V, Min	V, Max	V, Max	V, Min	mA	mA
CTT	–0.5	V_{REF} – 0.2	V_{REF} + 0.2	3.6	V_{REF} – 0.4	V_{REF} + 0.4	8	–8
AGP	–0.5	V_{REF} – 0.2	V_{REF} + 0.2	3.6	10% V_{CCO}	90% V_{CCO}	Note (2)	Note (2)

Notes:
1. V_{OL} and V_{OH} for lower drive currents are sample tested.
2. Tested according to the relevant specifications.

Switching Characteristics

All devices are 100% functionally tested. Internal timing parameters are derived from measuring internal test patterns. Listed below are representative values. For more specific, more precise, and worst-case guaranteed data, use the values reported by the static timing analyzer (TRCE in the Xilinx Development System) and back-annotated to the simulation netlist. All timing parameters assume worst-case operating conditions (supply voltage and junction temperature). Values apply to all Spartan-II devices unless otherwise noted.

Global Clock Input to Output Delay for LVTTL, *with* DLL (Pin-to-Pin)(1)

Symbol	Description	Device	Speed Grade			Units
			All	-6	-5	
			Min	Max	Max	
$T_{ICKOFDLL}$	Global clock input to output delay using output flip-flop for LVTTL, 12 mA, fast slew rate, *with* DLL.	All		2.9	3.3	ns

Notes:
1. Listed above are representative values where one global clock input drives one vertical clock line in each accessible column, and where all accessible IOB and CLB flip-flops are clocked by the global clock net.
2. Output timing is measured at 1.4V with 35 pF external capacitive load for LVTTL. The 35 pF load does not apply to the Min values. For other I/O standards and different loads, see the tables "Constants for Calculating TIOOP" and "Delay Measurement Methodology," page 60.
3. DLL output jitter is already included in the timing calculation.
4. For data *output* with different standards, adjust delays with the values shown in "IOB Output Delay Adjustments for Different Standards," page 59. For a global clock input with standards other than LVTTL, adjust delays with values from the "I/O Standard Global Clock Input Adjustments," page 61.

Global Clock Input to Output Delay for LVTTL, *without* DLL (Pin-to-Pin)(1)

Symbol	Description	Device	Speed Grade			Units
			All	-6	-5	
			Min	Max	Max	
T_{ICKOF}	Global clock input to output delay using output flip-flop for LVTTL, 12 mA, fast slew rate, *without* DLL.	XC2S15		4.5	5.4	ns
		XC2S30		4.5	5.4	ns
		XC2S50		4.5	5.4	ns
		XC2S100		4.6	5.5	ns
		XC2S150		4.6	5.5	ns
		XC2S200		4.7	5.6	ns

Notes:
1. Listed above are representative values where one global clock input drives one vertical clock line in each accessible column, and where all accessible IOB and CLB flip-flops are clocked by the global clock net.
2. Output timing is measured at 1.4V with 35 pF external capacitive load for LVTTL. The 35 pF load does not apply to the Min values. For other I/O standards and different loads, see the tables "Constants for Calculating TIOOP" and "Delay Measurement Methodology," page 60.
3. For data *output* with different standards, adjust delays with the values shown in "IOB Output Delay Adjustments for Different Standards," page 59. For a global clock input with standards other than LVTTL, adjust delays with values from the "I/O Standard Global Clock Input Adjustments," page 61.

Global Clock Setup and Hold for LVTTL Standard, *with* DLL (Pin-to-Pin)

Symbol	Description	Device	Speed Grade		Units
			-6	-5	
			Min	Min	
T_{PSDLL} / T_{PHDLL}	Input setup and hold time relative to global clock input signal for LVTTL standard, no delay, IFF,(1) with DLL	All	1.7 / 0	1.9 / 0	ns

Notes:
1. IFF = Input Flip-Flop or Latch
2. Setup time is measured relative to the Global Clock input signal with the fastest route and the lightest load. Hold time is measured relative to the Global Clock input signal with the slowest route and heaviest load.
3. DLL output jitter is already included in the timing calculation.
4. A zero hold time listing indicates no hold time or a negative hold time.
5. For data input with different standards, adjust the setup time delay by the values shown in "IOB Input Delay Adjustments for Different Standards," page 57. For a global clock input with standards other than LVTTL, adjust delays with values from the "I/O Standard Global Clock Input Adjustments," page 61.

Global Clock Setup and Hold for LVTTL Standard, *without* DLL (Pin-to-Pin)

Symbol	Description	Device	Speed Grade		Units
			-6	-5	
			Min	Min	
T_{PSFD} / T_{PHFD}	Input setup and hold time relative to global clock input signal for LVTTL standard, no delay, IFF,(1) without DLL	XC2S15	2.2 / 0	2.7 / 0	ns
		XC2S30	2.2 / 0	2.7 / 0	ns
		XC2S50	2.2 / 0	2.7 / 0	ns
		XC2S100	2.3 / 0	2.8 / 0	ns
		XC2S150	2.4 / 0	2.9 / 0	ns
		XC2S200	2.4 / 0	3.0 / 0	ns

Notes:
1. IFF = Input Flip-Flop or Latch
2. Setup time is measured relative to the Global Clock input signal with the fastest route and the lightest load. Hold time is measured relative to the Global Clock input signal with the slowest route and heaviest load.
3. A zero hold time listing indicates no hold time or a negative hold time.
4. For data input with different standards, adjust the setup time delay by the values shown in "IOB Input Delay Adjustments for Different Standards," page 57. For a global clock input with standards other than LVTTL, adjust delays with values from the "I/O Standard Global Clock Input Adjustments," page 61.

Clock Distribution Guidelines(1)

Symbol	Description	Speed Grade		Units
		-6	-5	
		Max	Max	
GCLK Clock Skew				
$T_{GSKEWIOB}$	Global clock skew between IOB flip-flops	0.13	0.14	ns

Notes:
1. These clock distribution delays are provided for guidance only. They reflect the delays encountered in a typical design under worst-case conditions. Precise values for a particular design are provided by the timing analyzer.

Clock Distribution Switching Characteristics

T_{GPIO} is specified for LVTTL levels. For other standards, adjust T_{GPIO} with the values shown in "I/O Standard Global Clock Input Adjustments".

Symbol	Description	Speed Grade		Units
		-6	-5	
		Max	Max	
GCLK IOB and Buffer				
T_{GPIO}	Global clock pad to output	0.7	0.8	ns
T_{GIO}	Global clock buffer I input to O output	0.7	0.8	ns

I/O Standard Global Clock Input Adjustments

Delays associated with a global clock input pad are specified for LVTTL levels. For other standards, adjust the delays by the values shown. A delay adjusted in this way constitutes a worst-case limit.

Symbol	Description	Standard	Speed Grade		Units
			-6	-5	
Data Input Delay Adjustments					
$T_{GPLVTTL}$	Standard-specific global clock input delay adjustments	LVTTL	0	0	ns
$T_{GPLVCMOS2}$		LVCMOS2	–0.04	–0.05	ns
$T_{GPPCI33\_3}$		PCI, 33 MHz, 3.3V	–0.11	–0.13	ns
$T_{GPPCI33\_5}$		PCI, 33 MHz, 5.0V	0.26	0.30	ns
$T_{GPPCI66\_3}$		PCI, 66 MHz, 3.3V	–0.11	–0.13	ns
T_{GPGTL}		GTL	0.80	0.84	ns
T_{GPGTLP}		GTL+	0.71	0.73	ns
T_{GPHSTL}		HSTL	0.63	0.64	ns
$T_{GPSSTL2}$		SSTL2	0.52	0.51	ns
$T_{GPSSTL3}$		SSTL3	0.56	0.55	ns
T_{GPCTT}		CTT	0.62	0.62	ns
T_{GPAGP}		AGP	0.54	0.53	ns

Notes:
1. Input timing for GPLVTTL is measured at 1.4V. For other I/O standards, see the table "Delay Measurement Methodology," page 60.

CLB Arithmetic Switching Characteristics

Setup times not listed explicitly can be approximated by decreasing the combinatorial delays by the setup time adjustment listed. Precise values are provided by the timing analyzer.

Symbol	Description	Speed Grade				Units
		-6		-5		
		Min	Max	Min	Max	
Combinatorial Delays						
T_{OPX}	F operand inputs to X via XOR	-	0.8	-	0.9	ns
T_{OPXB}	F operand input to XB output	-	1.3	-	1.5	ns
T_{OPY}	F operand input to Y via XOR	-	1.7	-	2.0	ns
T_{OPYB}	F operand input to YB output	-	1.7	-	2.0	ns
T_{OPCYF}	F operand input to COUT output	-	1.3	-	1.5	ns
T_{OPGY}	G operand inputs to Y via XOR	-	0.9	-	1.1	ns
T_{OPGYB}	G operand input to YB output	-	1.6	-	2.0	ns
T_{OPCYG}	G operand input to COUT output	-	1.2	-	1.4	ns
T_{BXCY}	BX initialization input to COUT	-	0.9	-	1.0	ns
T_{CINX}	CIN input to X output via XOR	-	0.4	-	0.5	ns
T_{CINXB}	CIN input to XB	-	0.1	-	0.1	ns
T_{CINY}	CIN input to Y via XOR	-	0.5	-	0.6	ns
T_{CINYB}	CIN input to YB	-	0.6	-	0.7	ns
T_{BYP}	CIN input to COUT output	-	0.1	-	0.1	ns
Multiplier Operation						
T_{FANDXB}	F1/2 operand inputs to XB output via AND	-	0.5	-	0.5	ns
T_{FANDYB}	F1/2 operand inputs to YB output via AND	-	0.9	-	1.1	ns
T_{FANDCY}	F1/2 operand inputs to COUT output via AND	-	0.5	-	0.6	ns
T_{GANDYB}	G1/2 operand inputs to YB output via AND	-	0.6	-	0.7	ns
T_{GANDCY}	G1/2 operand inputs to COUT output via AND	-	0.2	-	0.2	ns
Setup/Hold Times with Respect to Clock CLK(1)						
T_{CCKX} / T_{CKCX}	CIN input to FFX	1.1 / 0	-	1.2 / 0	-	ns
T_{CCKY} / T_{CKCY}	CIN input to FFY	1.2 / 0	-	1.3 / 0	-	ns

Notes:
1. A zero hold time listing indicates no hold time or a negative hold time.

Revision History

Date	Version No.	Description
09/18/00	2.0	Sectioned the Spartan-II Family data sheet into four modules. Updated timing to reflect the latest speed files. Added current supply numbers and XC2S200 -5 timing numbers. Approved -5 timing numbers as preliminary information with exceptions as noted.
11/02/00	2.1	Removed Power Down feature.
01/19/01	2.2	DC and timing numbers updated to Preliminary for the XC2S50 and XC2S100. Industrial power-on current specifications and -6 DLL timing numbers added. Power-on specification clarified.
03/09/01	2.3	Added note on power sequencing. Clarified power-on current requirement.
08/28/01	2.4	Added -6 preliminary timing. Added typical and industrial standby current numbers. Specified min. power-on current by junction temperature instead of by device type (Commercial vs. Industrial). Eliminated minimum V_{CCINT} ramp time requirement. Removed footnote limiting DLL operation to the Commercial temperature range.
07/26/02	2.5	Clarified that I/O leakage current is specified over the Recommended Operating Conditions for V_{CCINT} and V_{CCO}.
08/26/02	2.6	Added references for XAPP450 to Power-On Current Specification.
09/03/03	2.7	Added relaxed minimum power-on current (I_{CCPO}) requirements to page 53. On page 64, moved T_{RPW} values from maximum to minimum column.
06/13/08	2.8	Updated I/O measurement thresholds. Updated description and links. Updated all modules for continuous page, figure, and table numbering. Synchronized all modules to v2.8.

Package Thermal Characteristics

Table 39 provides the thermal characteristics for the various Spartan-II FPGA package offerings. This information is also available using the Thermal Query tool on xilinx.com (www.xilinx.com/cgi-bin/thermal/thermal.pl).

The junction-to-case thermal resistance (θ_{JC}) indicates the difference between the temperature measured on the package body (case) and the die junction temperature per watt of power consumption. The junction-to-board (θ_{JB}) value similarly reports the difference between the board and junction temperature. The junction-to-ambient (θ_{JA}) value reports the temperature difference between the ambient environment and the junction temperature. The θ_{JA} value is reported at different air velocities, measured in linear feet per minute (LFM). The "Still Air (0 LFM)" column shows the θ_{JA} value in a system without a fan. The thermal resistance drops with increasing air flow.

Table 39: **Spartan-II Package Thermal Characteristics**

Package	Device	Junction-to-Case (θ_{JC})	Junction-to-Board (θ_{JB})	Junction-to-Ambient (θ_{JA}) at Different Air Flows				Units
				Still Air (0 LFM)	250 LFM	500 LFM	750 LFM	
VQ100 VQG100	XC2S15	11.3	N/A	44.1	36.7	34.2	33.3	°C/Watt
	XC2S30	10.1	N/A	40.7	33.9	31.5	30.8	°C/Watt
TQ144 TQG144	XC2S15	7.3	N/A	38.6	30.0	25.7	24.1	°C/Watt
	XC2S30	6.7	N/A	34.7	27.0	23.1	21.7	°C/Watt
	XC2S50	5.8	N/A	32.2	25.1	21.4	20.1	°C/Watt
	XC2S100	5.3	N/A	31.4	24.4	20.9	19.6	°C/Watt
CS144 CSG144	XC2S30	2.8	N/A	34.0	26.0	23.9	23.2	°C/Watt
PQ208 PQG208	XC2S50	6.7	N/A	25.2	18.6	16.4	15.2	°C/Watt
	XC2S100	5.9	N/A	24.6	18.1	16.0	14.9	°C/Watt
	XC2S150	5.0	N/A	23.8	17.6	15.6	14.4	°C/Watt
	XC2S200	4.1	N/A	23.0	17.0	15.0	13.9	°C/Watt
FG256 FGG256	XC2S50	7.1	17.6	27.2	21.4	20.3	19.8	°C/Watt
	XC2S100	5.8	15.1	25.1	19.5	18.3	17.8	°C/Watt
	XC2S150	4.6	12.7	23.0	17.6	16.3	15.8	°C/Watt
	XC2S200	3.5	10.7	21.4	16.1	14.7	14.2	°C/Watt
FG456 FGG456	XC2S150	2.0	N/A	21.9	17.3	15.8	15.2	°C/Watt
	XC2S200	2.0	N/A	21.0	16.6	15.1	14.5	°C/Watt

XC2S30 Device Pinouts

XC2S30 Pad Name						
Function	Bank	VQ100	TQ144	CS144	PQ208	Bndry Scan
GND	-	P1	P143	A1	P1	-
TMS	-	P2	P142	B1	P2	-
I/O	7	P3	P141	C2	P3	113
I/O	7	-	P140	C1	P4	116
I/O	7	-	-	-	P5	119
I/O, V_{REF}	7	P4	P139	D4	P6	122
I/O	7	-	P138	D3	P8	125
I/O	7	P5	P137	D2	P9	128
I/O	7	P6	P136	D1	P10	131
GND	-	-	P135	E4	P11	-
V_{CCO}	7	-	-	-	P12	-
I/O	7	P7	P134	E3	P14	134
I/O	7	-	P133	E2	P15	137
I/O	7	-	-	-	P16	140
I/O	7	-	-	-	P17	143
I/O	7	-	-	-	P18	146
GND	-	-	-	-	P19	-
I/O, V_{REF}	7	P8	P132	E1	P20	149
I/O	7	P9	P131	F4	P21	152
I/O	7	-	P130	F3	P22	155
I/O	7	-	-	-	P23	158
I/O, IRDY(1)	7	P10	P129	F2	P24	161
GND	-	P11	P128	F1	P25	-
V_{CCO}	7	P12	P127	G2	P26	-
V_{CCO}	6	P12	P127	G2	P26	-
I/O, TRDY(1)	6	P13	P126	G1	P27	164
V_{CCINT}	-	P14	P125	G3	P28	-
I/O	6	-	P124	G4	P29	170
I/O	6	P15	P123	H1	P30	173
I/O, V_{REF}	6	P16	P122	H2	P31	176
GND	-	-	-	-	P32	-
I/O	6	-	-	-	P33	179
I/O	6	-	-	-	P34	182
I/O	6	-	-	-	P35	185
I/O	6	-	P121	H3	P36	188
I/O	6	P17	P120	H4	P37	191
V_{CCO}	6	-	-	-	P39	-
GND	-	-	P119	J1	P40	-
I/O	6	P18	P118	J2	P41	194
I/O	6	P19	P117	J3	P42	197
I/O	6	-	P116	J4	P43	200

XC2S30 Device Pinouts *(Continued)*

XC2S30 Pad Name						
Function	Bank	VQ100	TQ144	CS144	PQ208	Bndry Scan
I/O, V_{REF}	6	P20	P115	K1	P45	203
I/O	6	-	-	-	P46	206
I/O	6	-	P114	K2	P47	209
I/O	6	P21	P113	K3	P48	212
I/O	6	P22	P112	L1	P49	215
M1	-	P23	P111	L2	P50	218
GND	-	P24	P110	L3	P51	-
M0	-	P25	P109	M1	P52	219
V_{CCO}	6	P26	P108	M2	P53	-
V_{CCO}	5	P26	P107	N1	P53	-
M2	-	P27	P106	N2	P54	220
I/O	5	-	P103	K4	P57	227
I/O	5	-	-	-	P58	230
I/O, V_{REF}	5	P30	P102	L4	P59	233
I/O	5	-	P101	M4	P61	236
I/O	5	P31	P100	N4	P62	239
I/O	5	P32	P99	K5	P63	242
GND	-	-	P98	L5	P64	-
V_{CCO}	5	-	-	-	P65	-
V_{CCINT}	-	P33	P97	M5	P66	-
I/O	5	-	P96	N5	P67	245
I/O	5	-	P95	K6	P68	248
I/O	5	-	-	-	P69	251
I/O	5	-	-	-	P70	254
I/O	5	-	-	-	P71	257
GND	-	-	-	-	P72	-
I/O, V_{REF}	5	P34	P94	L6	P73	260
I/O	5	-	-	-	P74	263
I/O	5	-	P93	M6	P75	266
V_{CCINT}	-	P35	P92	N6	P76	-
I, GCK1	5	P36	P91	M7	P77	275
V_{CCO}	5	P37	P90	N7	P78	-
V_{CCO}	4	P37	P90	N7	P78	-
GND	-	P38	P89	L7	P79	-
I, GCK0	4	P39	P88	K7	P80	276
I/O	4	P40	P87	N8	P81	280
I/O	4	-	P86	M8	P82	283
I/O	4	-	-	-	P83	286
I/O, V_{REF}	4	P41	P85	L8	P84	289
GND	-	-	-	-	P85	-
I/O	4	-	-	-	P86	292

XC2S50 Device Pinouts *(Continued)*

XC2S50 Pad Name					
Function	Bank	TQ144	PQ208	FG256	Bndry Scan
I/O	0	-	-	D8	83
I/O	0	-	P188	A6	86
I/O, V_{REF}	0	P12	P189	B7	89
GND	-	-	P190	GND*	-
I/O	0	-	P191	C8	92
I/O	0	-	P192	D7	95
I/O	0	-	P193	E7	98
I/O	0	P11	P194	C7	104
I/O	0	P10	P195	B6	107
V_{CCINT}	-	P9	P196	V_{CCINT}*	-
V_{CCO}	0	-	P197	V_{CCO} Bank 0*	-
GND	-	P8	P198	GND*	-
I/O	0	P7	P199	A5	110
I/O	0	P6	P200	C6	113
I/O	0	-	P201	B5	116
I/O	0	-	-	D6	119
I/O	0	-	P202	A4	122
I/O, V_{REF}	0	P5	P203	B4	125
GND	-	-	-	GND*	-
I/O	0	-	P204	E6	128
I/O	0	-	-	D5	131
I/O	0	P4	P205	A3	134
I/O	0	-	-	C5	137
I/O	0	P3	P206	B3	140
TCK	-	P2	P207	C4	-
V_{CCO}	0	P1	P208	V_{CCO} Bank 0*	-
V_{CCO}	7	P144	P208	V_{CCO} Bank 7*	-

04/18/01

Notes:

1. IRDY and TRDY can only be accessed when using Xilinx PCI cores.
2. Pads labelled GND*, V_{CCINT}*, V_{CCO} Bank 0*, V_{CCO} Bank 1*, V_{CCO} Bank 2*, V_{CCO} Bank 3*, V_{CCO} Bank 4*, V_{CCO} Bank 5*, V_{CCO} Bank 6*, V_{CCO} Bank 7* are internally bonded to independent ground or power planes within the package.
3. See "VCCO Banks" for details on V_{CCO} banking.

Additional XC2S50 Package Pins

TQ144

Not Connected Pins					
P104	P105	-	-	-	-

11/02/00

Additional XC2S50 Package Pins *(Continued)*

PQ208

Not Connected Pins					
P55	P56	-	-	-	-

11/02/00

FG256

V_{CCINT} Pins					
C3	C14	D4	D13	E5	E12
M5	M12	N4	N13	P3	P14
V_{CCO} Bank 0 Pins					
E8	F8	-	-	-	-
V_{CCO} Bank 1 Pins					
E9	F9	-	-	-	-
V_{CCO} Bank 2 Pins					
H11	H12	-	-	-	-
V_{CCO} Bank 3 Pins					
J11	J12	-	-	-	-
V_{CCO} Bank 4 Pins					
L9	M9	-	-	-	-
V_{CCO} Bank 5 Pins					
L8	M8	-	-	-	-
V_{CCO} Bank 6 Pins					
J5	J6	-	-	-	-
V_{CCO} Bank 7 Pins					
H5	H6	-	-	-	-
GND Pins					
A1	A16	B2	B15	F6	F7
F10	F11	G6	G7	G8	G9
G10	G11	H7	H8	H9	H10
J7	J8	J9	J10	K6	K7
K8	K9	K10	K11	L6	L7
L10	L11	R2	R15	T1	T16
Not Connected Pins					
P4	R4	-	-	-	-

11/02/00

XC2S100 Device Pinouts

XC2S100 Pad Name		TQ144	PQ208	FG256	FG456	Bndry Scan
Function	Bank					
GND	-	P143	P1	GND*	GND*	-
TMS	-	P142	P2	D3	D3	-
I/O	7	P141	P3	C2	B1	185
I/O	7	-	-	A2	F5	191
I/O	7	P140	P4	B1	D2	194
I/O	7	-	-	-	E3	197
I/O	7	-	-	E3	G5	200
I/O	7	-	P5	D2	F3	203
GND	-	-	-	GND*	GND*	-
V_{CCO}	7	-	-	V_{CCO} Bank 7*	V_{CCO} Bank 7*	-
I/O, V_{REF}	7	P139	P6	C1	E2	206

XC2S100 Device Pinouts *(Continued)*

XC2S100 Pad Name		TQ144	PQ208	FG256	FG456	Bndry Scan
Function	Bank					
I/O	7	-	P7	F3	E1	209
I/O	7	-	-	E2	H5	215
I/O	7	P138	P8	E4	F2	218
I/O	7	-	-	-	F1	221
I/O, V_{REF}	7	P137	P9	D1	H4	224
I/O	7	P136	P10	E1	G1	227
GND	-	P135	P11	GND*	GND*	-
V_{CCO}	7	-	P12	V_{CCO} Bank 7*	V_{CCO} Bank 7*	-
V_{CCINT}	-	-	P13	V_{CCINT}*	V_{CCINT}*	-
I/O	7	P134	P14	F2	H3	230
I/O	7	P133	P15	G3	H2	233
I/O	7	-	-	F1	J5	236
I/O	7	-	P16	F4	J2	239
I/O	7	-	P17	F5	K5	245
I/O	7	-	P18	G2	K1	248
GND	-	-	P19	GND*	GND*	-
I/O, V_{REF}	7	P132	P20	H3	K3	251
I/O	7	P131	P21	G4	K4	254
I/O	7	-	-	H2	L6	257
I/O	7	P130	P22	G5	L1	260
I/O	7	-	P23	H4	L4	266
I/O, IRDY(1)	7	P129	P24	G1	L3	269
GND	-	P128	P25	GND*	GND*	-
V_{CCO}	7	P127	P26	V_{CCO} Bank 7*	V_{CCO} Bank 7*	-
V_{CCO}	6	P127	P26	V_{CCO} Bank 6*	V_{CCO} Bank 6*	-
I/O, TRDY(1)	6	P126	P27	J2	M1	272
V_{CCINT}	-	P125	P28	V_{CCINT}*	V_{CCINT}*	-
I/O	6	P124	P29	H1	M3	281
I/O	6	-	-	J4	M4	284
I/O	6	P123	P30	J1	M5	287
I/O, V_{REF}	6	P122	P31	J3	N2	290
GND	-	-	P32	GND*	GND*	-
I/O	6	-	P33	K5	N3	293
I/O	6	-	P34	K2	N4	296
I/O	6	-	P35	K1	P2	302
I/O	6	-	-	K3	P4	305
I/O	6	P121	P36	L1	P3	308
I/O	6	P120	P37	L2	R2	311

XC2S100 Device Pinouts *(Continued)*

XC2S100 Pad Name		TQ144	PQ208	FG256	FG456	Bndry Scan
Function	Bank					
I/O	2	-	-	F12	G20	695
I/O	2	-	P149	E15	F19	701
I/O, V_{REF}	2	P41	P150	F13	F21	704
V_{CCO}	2	-	-	V_{CCO} Bank 2*	V_{CCO} Bank 2*	-
GND	-	-	-	GND*	GND*	-
I/O	2	-	P151	E14	F20	707
I/O	2	-	-	C16	F18	710
I/O	2	-	-	-	E21	713
I/O	2	P40	P152	E13	D22	716
I/O	2	-	-	B16	E20	719
I/O (DIN, D0)	2	P39	P153	D14	D20	725
I/O (DOUT, BUSY)	2	P38	P154	C15	C21	728
CCLK	2	P37	P155	D15	B22	731
V_{CCO}	2	P36	P156	V_{CCO} Bank 2*	V_{CCO} Bank 2*	-
V_{CCO}	1	P35	P156	V_{CCO} Bank 1*	V_{CCO} Bank 1*	-
TDO	2	P34	P157	B14	A21	-
GND	-	P33	P158	GND*	GND*	-
TDI	-	P32	P159	A15	B20	-
I/O ($\overline{CS}$)	1	P31	P160	B13	C19	0
I/O ($\overline{WRITE}$)	1	P30	P161	C13	A20	3
I/O	1	-	-	C12	D17	9
I/O	1	P29	P162	A14	A19	12
I/O	1	-	-	-	B18	15
I/O	1	-	-	D12	C17	18
I/O	1	-	P163	B12	D16	21
GND	-	-	-	GND*	GND*	-
V_{CCO}	1	-	-	V_{CCO} Bank 1*	V_{CCO} Bank 1*	-
I/O, V_{REF}	1	P28	P164	C11	A18	24
I/O	1	-	P165	A13	B17	27
I/O	1	-	-	D11	D15	33
I/O	1	-	P166	A12	C16	36
I/O	1	-	-	-	D14	39
I/O, V_{REF}	1	P27	P167	E11	E14	42
I/O	1	P26	P168	B11	A16	45
GND	-	P25	P169	GND*	GND*	-

XC2S100 Device Pinouts *(Continued)*

XC2S100 Pad Name		TQ144	PQ208	FG256	FG456	Bndry Scan
Function	Bank					
V_{CCO}	1	-	P170	V_{CCO} Bank 1*	V_{CCO} Bank 1*	-
V_{CCINT}	-	P24	P171	V_{CCINT}*	V_{CCINT}*	-
I/O	1	P23	P172	A11	C15	48
I/O	1	P22	P173	C10	B15	51
I/O	1	-	-	-	F12	54
I/O	1	-	P174	B10	C14	57
I/O	1	-	P175	D10	D13	63
I/O	1	-	P176	A10	C13	66
GND	-	-	P177	GND*	GND*	-
I/O, V_{REF}	1	P21	P178	B9	B13	69
I/O	1	-	P179	E10	E12	72
I/O	1	-	-	A9	B12	75
I/O	1	P20	P180	D9	D12	78
I/O	1	P19	P181	A8	D11	84
I, GCK2	1	P18	P182	C9	A11	90
GND	-	P17	P183	GND*	GND*	-
V_{CCO}	1	P16	P184	V_{CCO} Bank 1*	V_{CCO} Bank 1*	-
V_{CCO}	0	P16	P184	V_{CCO} Bank 0*	V_{CCO} Bank 0*	-
I, GCK3	0	P15	P185	B8	C11	91
V_{CCINT}	-	P14	P186	V_{CCINT}*	V_{CCINT}*	-
I/O	0	P13	P187	A7	A10	101
I/O	0	-	-	D8	B10	104