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Understanding Embedded - FPGAs (Field Programmable Gate Array)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details

Product Status	Active
Number of LABs/CLBs	216
Number of Logic Elements/Cells	972
Total RAM Bits	24576
Number of I/O	92
Number of Gates	30000
Voltage - Supply	2.375V ~ 2.625V
Mounting Type	Surface Mount
Operating Temperature	-40°C ~ 100°C (TJ)
Package / Case	144-TFBGA, CSPBGA
Supplier Device Package	144-LCSBGA (12x12)
Purchase URL	https://www.e-xfl.com/product-detail/xilinx/xc2s30-5csg144i

General Overview

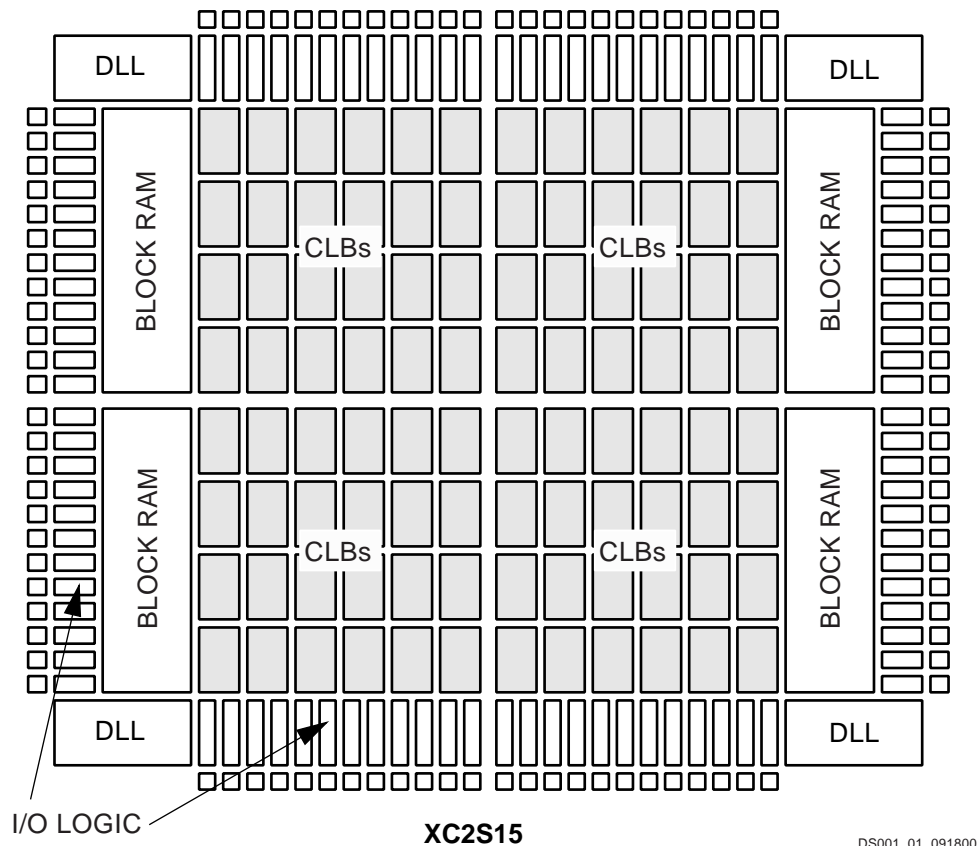
The Spartan-II family of FPGAs have a regular, flexible, programmable architecture of Configurable Logic Blocks (CLBs), surrounded by a perimeter of programmable Input/Output Blocks (IOBs). There are four Delay-Locked Loops (DLLs), one at each corner of the die. Two columns of block RAM lie on opposite sides of the die, between the CLBs and the IOB columns. These functional elements are interconnected by a powerful hierarchy of versatile routing channels (see Figure 1).

Spartan-II FPGAs are customized by loading configuration data into internal static memory cells. Unlimited reprogramming cycles are possible with this approach. Stored values in these cells determine logic functions and interconnections implemented in the FPGA. Configuration data can be read from an external serial PROM (master

serial mode), or written into the FPGA in slave serial, slave parallel, or Boundary Scan modes.

Spartan-II FPGAs are typically used in high-volume applications where the versatility of a fast programmable solution adds benefits. Spartan-II FPGAs are ideal for shortening product development cycles while offering a cost-effective solution for high volume production.

Spartan-II FPGAs achieve high-performance, low-cost operation through advanced architecture and semiconductor technology. Spartan-II devices provide system clock rates up to 200 MHz. In addition to the conventional benefits of high-volume programmable logic solutions, Spartan-II FPGAs also offer on-chip synchronous single-port and dual-port RAM (block and distributed form), DLL clock drivers, programmable set and reset on all flip-flops, fast carry logic, and many other features.



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Figure 1: Basic Spartan-II Family FPGA Block Diagram

Architectural Description

Spartan-II FPGA Array

The Spartan®-II field-programmable gate array, shown in [Figure 2](#), is composed of five major configurable elements:

- IOBs provide the interface between the package pins and the internal logic
- CLBs provide the functional elements for constructing most logic
- Dedicated block RAM memories of 4096 bits each
- Clock DLLs for clock-distribution delay compensation and clock domain control
- Versatile multi-level interconnect structure

As can be seen in [Figure 2](#), the CLBs form the central logic structure with easy access to all support and routing structures. The IOBs are located around all the logic and

memory elements for easy and quick routing of signals on and off the chip.

Values stored in static memory cells control all the configurable logic elements and interconnect resources. These values load into the memory cells on power-up, and can reload if necessary to change the function of the device.

Each of these elements will be discussed in detail in the following sections.

Input/Output Block

The Spartan-II FPGA IOB, as seen in [Figure 2](#), features inputs and outputs that support a wide variety of I/O signaling standards. These high-speed inputs and outputs are capable of supporting various state of the art memory and bus interfaces. [Table 3](#) lists several of the standards which are supported along with the required reference, output and termination voltages needed to meet the standard.

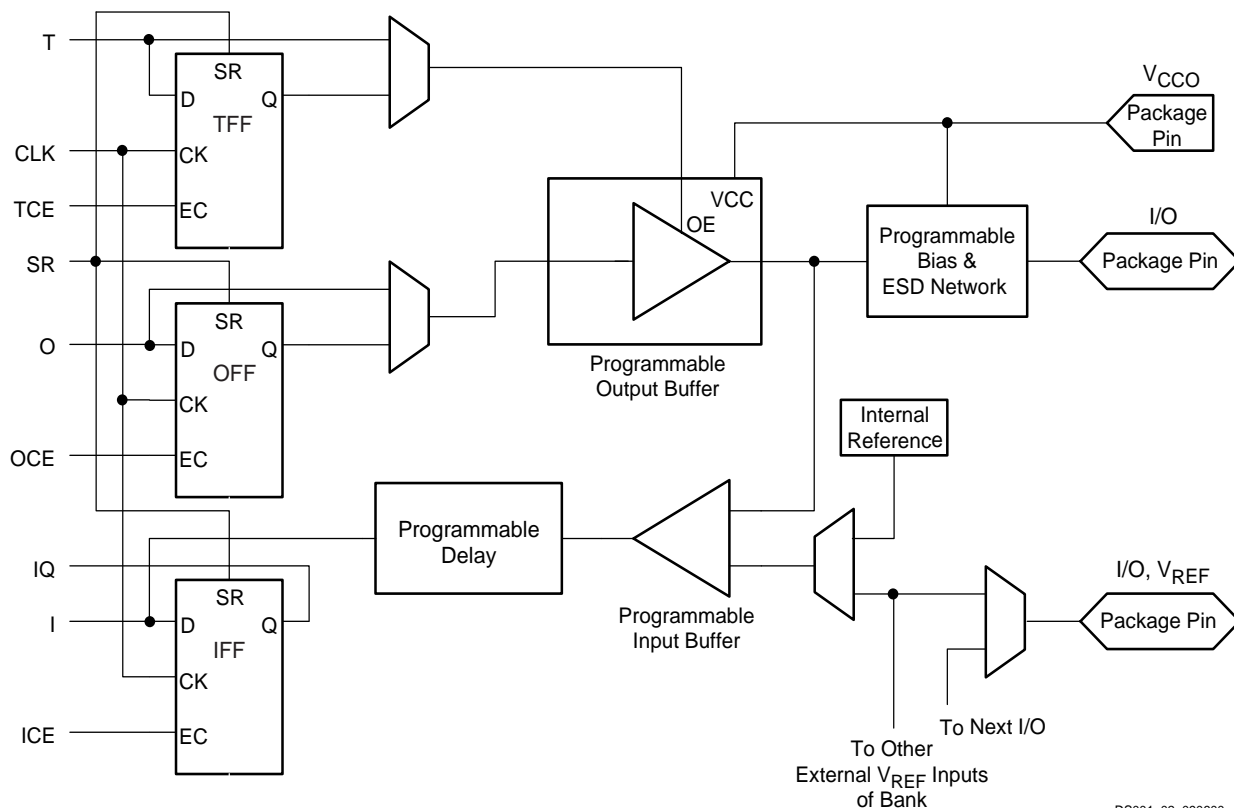


Figure 2: Spartan-II FPGA Input/Output Block (IOB)

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The three IOB registers function either as edge-triggered D-type flip-flops or as level-sensitive latches. Each IOB has a clock signal (CLK) shared by the three registers and independent Clock Enable (CE) signals for each register. In addition to the CLK and CE control signals, the three registers share a Set/Reset (SR). For each register, this signal can be independently configured as a synchronous Set, a synchronous Reset, an asynchronous Preset, or an asynchronous Clear.

A feature not shown in the block diagram, but controlled by the software, is polarity control. The input and output buffers and all of the IOB control signals have independent polarity controls.

Optional pull-up and pull-down resistors and an optional weak-keeper circuit are attached to each pad. Prior to configuration all outputs not involved in configuration are forced into their high-impedance state. The pull-down resistors and the weak-keeper circuits are inactive, but inputs may optionally be pulled up.

Table 3: Standards Supported by I/O (Typical Values)

I/O Standard	Input Reference Voltage (V_{REF})	Output Source Voltage (V_{CCO})	Board Termination Voltage (V_{TT})
LVTTTL (2-24 mA)	N/A	3.3	N/A
LVC MOS2	N/A	2.5	N/A
PCI (3V/5V, 33 MHz/66 MHz)	N/A	3.3	N/A
GTL	0.8	N/A	1.2
GTL+	1.0	N/A	1.5
HSTL Class I	0.75	1.5	0.75
HSTL Class III	0.9	1.5	1.5
HSTL Class IV	0.9	1.5	1.5
SSTL3 Class I and II	1.5	3.3	1.5
SSTL2 Class I and II	1.25	2.5	1.25
CTT	1.5	3.3	1.5
AGP-2X	1.32	3.3	N/A

The activation of pull-up resistors prior to configuration is controlled on a global basis by the configuration mode pins. If the pull-up resistors are not activated, all the pins will float. Consequently, external pull-up or pull-down resistors must be provided on pins required to be at a well-defined logic level prior to configuration.

All pads are protected against damage from electrostatic discharge (ESD) and from over-voltage transients. Two forms of over-voltage protection are provided, one that permits 5V compliance, and one that does not. For 5V compliance, a zener-like structure connected to ground turns on when the output rises to approximately 6.5V. When 5V compliance is not required, a conventional clamp diode may be connected to the output supply voltage, V_{CCO} . The type of over-voltage protection can be selected independently for each pad.

All Spartan-II FPGA IOBs support IEEE 1149.1-compatible boundary scan testing.

Input Path

A buffer in the Spartan-II FPGA IOB input path routes the input signal either directly to internal logic or through an optional input flip-flop.

An optional delay element at the D-input of this flip-flop eliminates pad-to-pad hold time. The delay is matched to the internal clock-distribution delay of the FPGA, and when used, assures that the pad-to-pad hold time is zero.

Each input buffer can be configured to conform to any of the low-voltage signaling standards supported. In some of these standards the input buffer utilizes a user-supplied threshold voltage, V_{REF} . The need to supply V_{REF} imposes constraints on which standards can be used in close proximity to each other. See "[I/O Banking](#)," page 9.

There are optional pull-up and pull-down resistors at each input for use after configuration.

Output Path

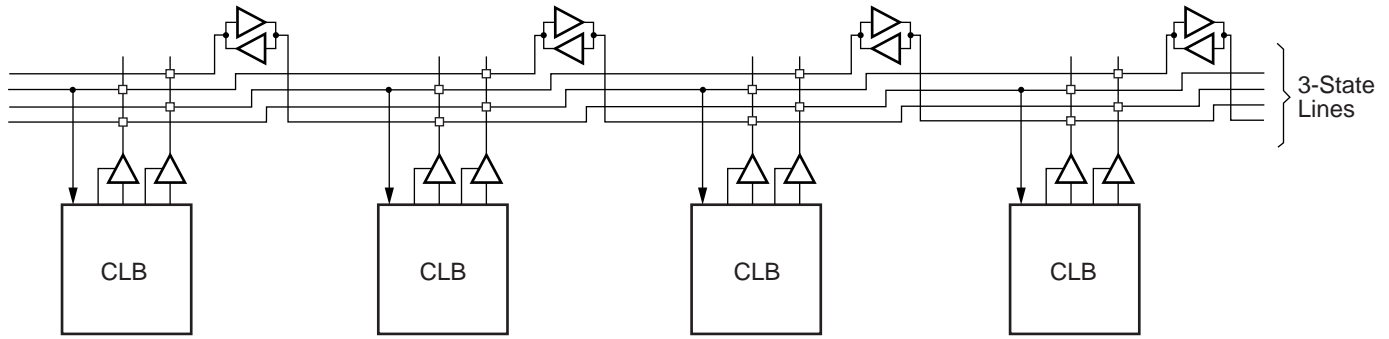
The output path includes a 3-state output buffer that drives the output signal onto the pad. The output signal can be routed to the buffer directly from the internal logic or through an optional IOB output flip-flop.

The 3-state control of the output can also be routed directly from the internal logic or through a flip-flop that provides synchronous enable and disable.

Each output driver can be individually programmed for a wide range of low-voltage signaling standards. Each output buffer can source up to 24 mA and sink up to 48 mA. Drive strength and slew rate controls minimize bus transients.

In most signaling standards, the output high voltage depends on an externally supplied V_{CCO} voltage. The need to supply V_{CCO} imposes constraints on which standards can be used in close proximity to each other. See "[I/O Banking](#)".

An optional weak-keeper circuit is connected to each output. When selected, the circuit monitors the voltage on the pad and weakly drives the pin High or Low to match the input signal. If the pin is connected to a multiple-source signal, the weak keeper holds the signal in its last state if all



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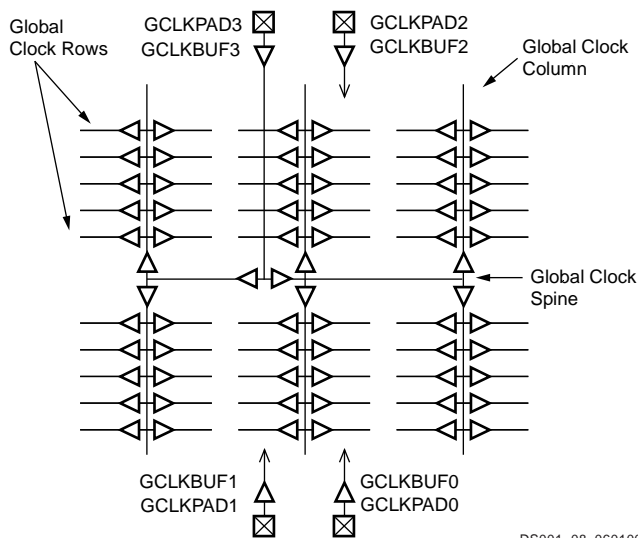
Figure 7: BUFT Connections to Dedicated Horizontal Bus Lines

Clock Distribution

The Spartan-II family provides high-speed, low-skew clock distribution through the primary global routing resources described above. A typical clock distribution net is shown in Figure 8.

Four global buffers are provided, two at the top center of the device and two at the bottom center. These drive the four primary global nets that in turn drive any clock pin.

Four dedicated clock pads are provided, one adjacent to each of the global buffers. The input to the global buffer is selected either from these pads or from signals in the general purpose routing. Global clock pins do not have the option for internal, weak pull-up resistors.



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Figure 8: Global Clock Distribution Network

Delay-Locked Loop (DLL)

Associated with each global clock input buffer is a fully digital Delay-Locked Loop (DLL) that can eliminate skew between the clock input pad and internal clock-input pins throughout the device. Each DLL can drive two global clock

networks. The DLL monitors the input clock and the distributed clock, and automatically adjusts a clock delay element. Additional delay is introduced such that clock edges reach internal flip-flops exactly one clock period after they arrive at the input. This closed-loop system effectively eliminates clock-distribution delay by ensuring that clock edges arrive at internal flip-flops in synchronism with clock edges arriving at the input.

In addition to eliminating clock-distribution delay, the DLL provides advanced control of multiple clock domains. The DLL provides four quadrature phases of the source clock, can double the clock, or divide the clock by 1.5, 2, 2.5, 3, 4, 5, 8, or 16. It has six outputs.

The DLL also operates as a clock mirror. By driving the output from a DLL off-chip and then back on again, the DLL can be used to deskew a board level clock among multiple Spartan-II devices.

In order to guarantee that the system clock is operating correctly prior to the FPGA starting up after configuration, the DLL can delay the completion of the configuration process until after it has achieved lock.

Boundary Scan

Spartan-II devices support all the mandatory boundary-scan instructions specified in the IEEE standard 1149.1. A Test Access Port (TAP) and registers are provided that implement the EXTEST, SAMPLE/PRELOAD, and BYPASS instructions. The TAP also supports two USERCODE instructions and internal scan chains.

The TAP uses dedicated package pins that always operate using LVTTTL. For TDO to operate using LVTTTL, the V_{CCO} for Bank 2 must be 3.3V. Otherwise, TDO switches rail-to-rail between ground and V_{CCO} . TDI, TMS, and TCK have a default internal weak pull-up resistor, and TDO has no default resistor. Bitstream options allow setting any of the four TAP pins to have an internal pull-up, pull-down, or neither.

Configuration

Configuration is the process by which the bitstream of a design, as generated by the Xilinx software, is loaded into the internal configuration memory of the FPGA. Spartan-II devices support both serial configuration, using the master/slave serial and JTAG modes, as well as byte-wide configuration employing the Slave Parallel mode.

Configuration File

Spartan-II devices are configured by sequentially loading frames of data that have been concatenated into a configuration file. [Table 8](#) shows how much nonvolatile storage space is needed for Spartan-II devices.

It is important to note that, while a PROM is commonly used to store configuration data before loading them into the FPGA, it is by no means required. Any of a number of different kinds of under populated nonvolatile storage already available either on or off the board (i.e., hard drives, FLASH cards, etc.) can be used. For more information on configuration without a PROM, refer to [XAPP098, The Low-Cost, Efficient Serial Configuration of Spartan FPGAs](#).

Table 8: Spartan-II Configuration File Size

Device	Configuration File Size (Bits)
XC2S15	197,696
XC2S30	336,768
XC2S50	559,200
XC2S100	781,216
XC2S150	1,040,096
XC2S200	1,335,840

Modes

Spartan-II devices support the following four configuration modes:

- Slave Serial mode
- Master Serial mode
- Slave Parallel mode
- Boundary-scan mode

The Configuration mode pins (M2, M1, M0) select among these configuration modes with the option in each case of having the IOB pins either pulled up or left floating prior to the end of configuration. The selection codes are listed in [Table 9](#).

Configuration through the boundary-scan port is always available, independent of the mode selection. Selecting the boundary-scan mode simply turns off the other modes. The three mode pins have internal pull-up resistors, and default to a logic High if left unconnected.

Table 9: Configuration Modes

Configuration Mode	Preconfiguration Pull-ups	M0	M1	M2	CCLK Direction	Data Width	Serial D _{OUT}
Master Serial mode	No	0	0	0	Out	1	Yes
	Yes	0	0	1			
Slave Parallel mode	Yes	0	1	0	In	8	No
	No	0	1	1			
Boundary-Scan mode	Yes	1	0	0	N/A	1	No
	No	1	0	1			
Slave Serial mode	Yes	1	1	0	In	1	Yes
	No	1	1	1			

Notes:

1. During power-on and throughout configuration, the I/O drivers will be in a high-impedance state. After configuration, all unused I/Os (those not assigned signals) will remain in a high-impedance state. Pins used as outputs may pulse High at the end of configuration (see [Answer 10504](#)).
2. If the Mode pins are set for preconfiguration pull-ups, those resistors go into effect once the rising edge of INIT samples the Mode pins. They will stay in effect until GTS is released during startup, after which the UnusedPin bitstream generator option will determine whether the unused I/Os have a pull-up, pull-down, or no resistor.

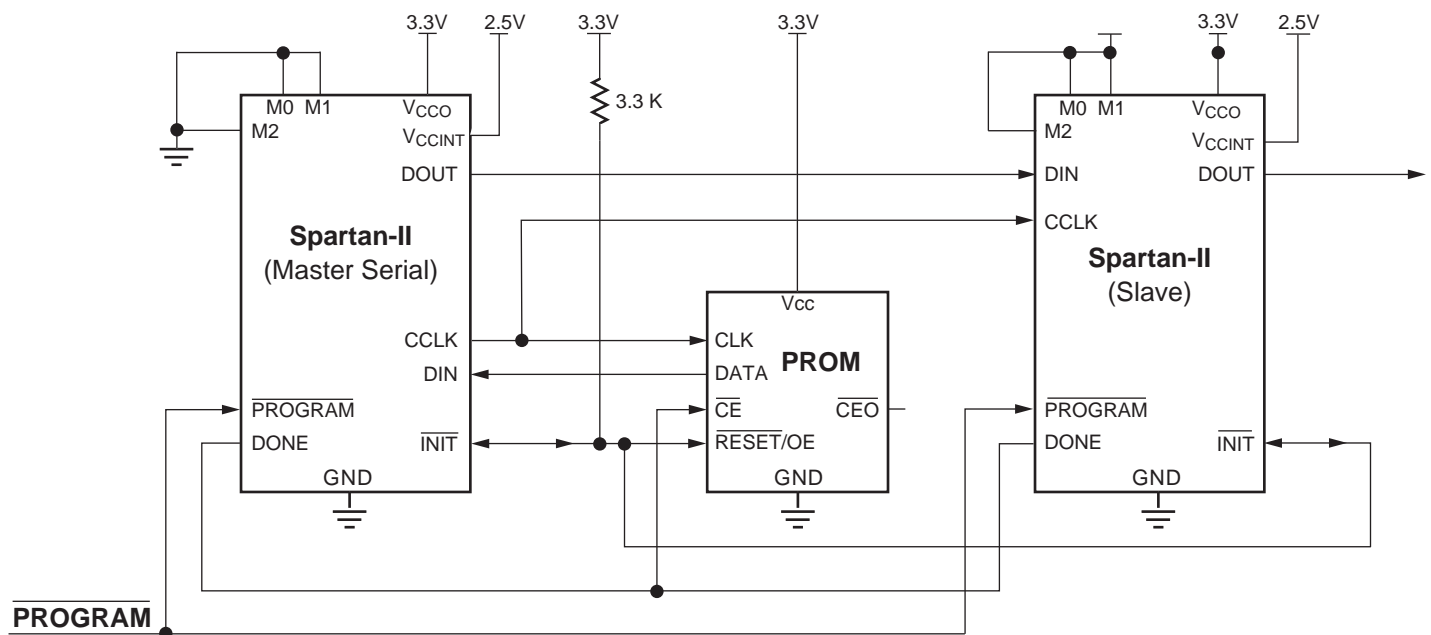
Slave Serial Mode

In Slave Serial mode, the FPGA's CCLK pin is driven by an external source, allowing FPGAs to be configured from other logic devices such as microprocessors or in a daisy-chain configuration. Figure 15 shows connections for a Master Serial FPGA configuring a Slave Serial FPGA from a PROM. A Spartan-II device in slave serial mode should be connected as shown for the third device from the left. Slave Serial mode is selected by a <11x> on the mode pins (M0, M1, M2).

Figure 16 shows the timing for Slave Serial configuration. The serial bitstream must be setup at the DIN input pin a short time before each rising edge of an externally generated CCLK.

Multiple FPGAs in Slave Serial mode can be daisy-chained for configuration from a single source. The maximum amount of data that can be sent to the DOUT pin for a serial daisy chain is $2^{20}-1$ (1,048,575) 32-bit words, or 33,554,400 bits, which is approximately 25 XC2S200 bitstreams. The configuration bitstream of downstream devices is limited to this size.

After an FPGA is configured, data for the next device is routed to the DOUT pin. Data on the DOUT pin changes on the rising edge of CCLK. Configuration must be delayed until INIT pins of all daisy-chained FPGAs are High. For more information, see "Start-up," page 19.



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Notes:

1. If the DriveDone configuration option is not active for any of the FPGAs, pull up DONE with a 330Ω resistor.

Figure 15: Master/Slave Serial Configuration Circuit Diagram

Startup Delay Property

This property, `STARTUP_WAIT`, takes on a value of `TRUE` or `FALSE` (the default value). When `TRUE` the Startup Sequence following device configuration is paused at a user-specified point until the DLL locks. [XAPP176: Configuration and Readback of the Spartan-II and Spartan-IIe Families](#) explains how this can result in delaying the assertion of the `DONE` pin until the DLL locks.

DLL Location Constraints

The DLLs are distributed such that there is one DLL in each corner of the device. The location constraint `LOC`, attached to the DLL primitive with the numeric identifier 0, 1, 2, or 3, controls DLL location. The orientation of the four DLLs and their corresponding clock resources appears in [Figure 27](#).

The `LOC` property uses the following form.

`LOC = DLL2`

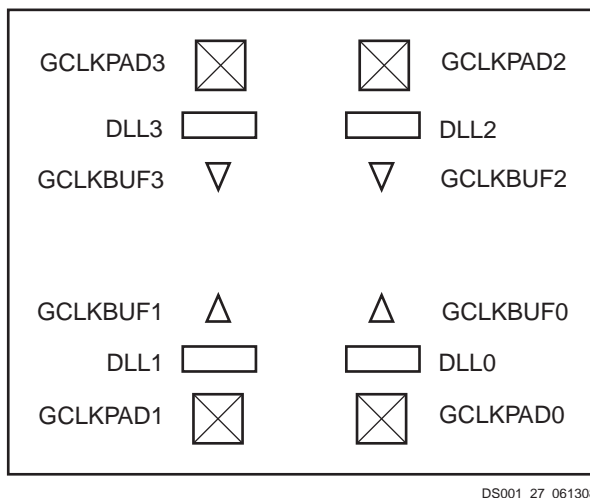


Figure 27: Orientation of DLLs

Design Considerations

Use the following design considerations to avoid pitfalls and improve success designing with Xilinx devices.

Input Clock

The output clock signal of a DLL, essentially a delayed version of the input clock signal, reflects any instability on the input clock in the output waveform. For this reason the quality of the DLL input clock relates directly to the quality of the output clock waveforms generated by the DLL. The DLL input clock requirements are specified in the ["DLL Timing Parameters"](#) section of the data sheet.

In most systems a crystal oscillator generates the system clock. The DLL can be used with any commercially available quartz crystal oscillator. For example, most crystal oscillators produce an output waveform with a frequency tolerance of 100 PPM, meaning 0.01 percent change in the

clock period. The DLL operates reliably on an input waveform with a frequency drift of up to 1 ns — orders of magnitude in excess of that needed to support any crystal oscillator in the industry. However, the cycle-to-cycle jitter must be kept to less than 300 ps in the low frequencies and 150 ps for the high frequencies.

Input Clock Changes

Changing the period of the input clock beyond the maximum drift amount requires a manual reset of the `CLKDLL`. Failure to reset the DLL will produce an unreliable lock signal and output clock.

It is possible to stop the input clock in a way that has little impact to the DLL. Stopping the clock should be limited to less than approximately 100 μ s to keep device cooling to a minimum and maintain the validity of the current tap setting. The clock should be stopped during a Low phase, and when restored the full High period should be seen. During this time `LOCKED` will stay High and remain High when the clock is restored. If these conditions may not be met in the design, apply a manual reset to the DLL after re-starting the input clock, even if the `LOCKED` signal has not changed.

When the clock is stopped, one to four more clocks will still be observed as the delay line is flushed. When the clock is restarted, the output clocks will not be observed for one to four clocks as the delay line is filled. The most common case will be two or three clocks.

In a similar manner, a phase shift of the input clock is also possible. The phase shift will propagate to the output one to four clocks after the original shift, with no disruption to the `CLKDLL` control.

Output Clocks

As mentioned earlier in the DLL pin descriptions, some restrictions apply regarding the connectivity of the output pins. The DLL clock outputs can drive an `OBUF`, a global clock buffer `BUFG`, or route directly to destination clock pins. The only `BUFGs` that the DLL clock outputs can drive are the two on the same edge of the device (top or bottom). One DLL output can drive more than one `OBUF`; however, this adds skew.

Do not use the DLL output clock signals until after activation of the `LOCKED` signal. Prior to the activation of the `LOCKED` signal, the DLL output clocks are not valid and can exhibit glitches, spikes, or other spurious movement.

the LOC property is described below. Table 16 summarizes the input standards compatibility requirements.

An optional delay element is associated with each IBUF. When the IBUF drives a flip-flop within the IOB, the delay element by default activates to ensure a zero hold-time requirement. The NODELAY=TRUE property overrides this default.

When the IBUF does not drive a flip-flop within the IOB, the delay element de-activates by default to provide higher performance. To delay the input signal, activate the delay element with the DELAY=TRUE property.

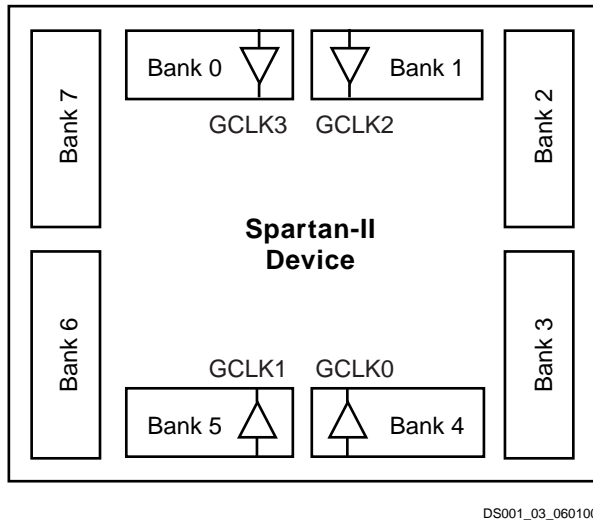


Figure 36: I/O Banks

Table 16: Xilinx Input Standards Compatibility Requirements

Rule 1	All differential amplifier input signals within a bank are required to be of the same standard.
Rule 2	There are no placement restrictions for inputs with standards that require a single-ended input buffer.

IBUFG

Signals used as high fanout clock inputs to the Spartan-II device should drive a global clock input buffer (IBUFG) via an external input port in order to take advantage of one of the four dedicated global clock distribution networks. The output of the IBUFG primitive can

only drive a CLKDLL, CLKDLLHF, or a BUFG primitive. The generic IBUFG primitive appears in Figure 37.

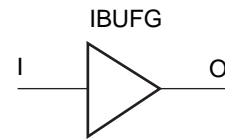


Figure 37: Global Clock Input Buffer (IBUFG) Primitive

With no extension or property specified for the generic IBUFG primitive, the assumed standard is LVTTTL.

The voltage reference signal is "banked" within the Spartan-II device on a half-edge basis such that for all packages there are eight independent V_{REF} banks internally. See Figure 36 for a representation of the I/O banks. Within each bank approximately one of every six I/O pins is automatically configured as a V_{REF} input.

IBUFG placement restrictions require any differential amplifier input signals within a bank be of the same standard. The LOC property can specify a location for the IBUFG.

As an added convenience, the BUFGP can be used to instantiate a high fanout clock input. The BUFGP primitive represents a combination of the LVTTTL IBUFG and BUFG primitives, such that the output of the BUFGP can connect directly to the clock pins throughout the design.

The Spartan-II FPGA BUFGP primitive can only be placed in a global clock pad location. The LOC property can specify a location for the BUFGP.

OBUF

An OBUF must drive outputs through an external output port. The generic output buffer (OBUF) primitive appears in Figure 38.

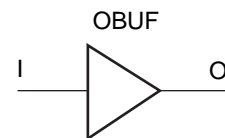


Figure 38: Output Buffer (OBUF) Primitive

With no extension or property specified for the generic OBUF primitive, the assumed standard is slew rate limited LVTTTL with 12 mA drive strength.

The LVTTTL OBUF additionally can support one of two slew rate modes to minimize bus transients. By default, the slew rate for each output buffer is reduced to minimize power bus transients when switching non-critical signals.

HSTL Class III

A sample circuit illustrating a valid termination technique for HSTL_III appears in Figure 45. DC voltage specifications appear in Table 23 for the HSTL_III standard. See "DC Specifications" in Module 3 for the actual FPGA characteristics.

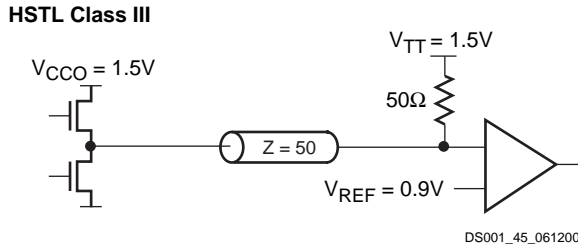


Figure 45: Terminated HSTL Class III

Table 23: HSTL Class III Voltage Specification

Parameter	Min	Typ	Max
V _{CCO}	1.40	1.50	1.60
V _{REF} ⁽¹⁾	-	0.90	-
V _{TT}	-	V _{CCO}	-
V _{IH}	V _{REF} + 0.1	-	-
V _{IL}	-	-	V _{REF} - 0.1
V _{OH}	V _{CCO} - 0.4	-	-
V _{OL}	-	-	0.4
I _{OH} at V _{OH} (mA)	-8	-	-
I _{OL} at V _{OL} (mA)	24	-	-

Notes:

- Per EIA/JESD8-6, "The value of V_{REF} is to be selected by the user to provide optimum noise margin in the use conditions specified by the user."

HSTL Class IV

A sample circuit illustrating a valid termination technique for HSTL_IV appears in Figure 46. DC voltage specifications appear in Table 23 for the HSTL_IV standard. See "DC Specifications" in Module 3 for the actual FPGA characteristics.

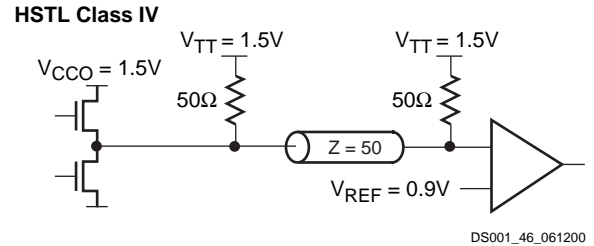


Figure 46: Terminated HSTL Class IV

Table 24: HSTL Class IV Voltage Specification

Parameter	Min	Typ	Max
V _{CCO}	1.40	1.50	1.60
V _{REF}	-	0.90	-
V _{TT}	-	V _{CCO}	-
V _{IH}	V _{REF} + 0.1	-	-
V _{IL}	-	-	V _{REF} - 0.1
V _{OH}	V _{CCO} - 0.4	-	-
V _{OL}	-	-	0.4
I _{OH} at V _{OH} (mA)	-8	-	-
I _{OL} at V _{OL} (mA)	48	-	-

Notes:

- Per EIA/JESD8-6, "The value of V_{REF} is to be selected by the user to provide optimum noise margin in the use conditions specified by the user."

Power-On Requirements

Spartan-II FPGAs require that a minimum supply current I_{CCPO} be provided to the V_{CCINT} lines for a successful power-on. If more current is available, the FPGA can consume more than I_{CCPO} minimum, though this cannot adversely affect reliability.

A maximum limit for I_{CCPO} is not specified. Therefore the use of foldback/crowbar supplies and fuses deserves special attention. In these cases, limit the I_{CCPO} current to a level below the trip point for over-current protection in order to avoid inadvertently shutting down the supply.

Symbol	Description	Conditions		New Requirements ⁽¹⁾ For Devices with Date Code 0321 or Later		Old Requirements ⁽¹⁾ For Devices with Date Code before 0321		Units
		Junction Temperature ⁽²⁾	Device Temperature Grade	Min	Max	Min	Max	
I_{CCPO} ⁽³⁾	Total V_{CCINT} supply current required during power-on	$-40^{\circ}\text{C} \leq T_J < -20^{\circ}\text{C}$	Industrial	1.50	-	2.00	-	A
		$-20^{\circ}\text{C} \leq T_J < 0^{\circ}\text{C}$	Industrial	1.00	-	2.00	-	A
		$0^{\circ}\text{C} \leq T_J \leq 85^{\circ}\text{C}$	Commercial	0.25	-	0.50	-	A
		$85^{\circ}\text{C} < T_J \leq 100^{\circ}\text{C}$	Industrial	0.50	-	0.50	-	A
T_{CCPO} ^(4,5)	V_{CCINT} ramp time	$-40^{\circ}\text{C} \leq T_J \leq 100^{\circ}\text{C}$	All	-	50	-	50	ms

Notes:

- The date code is printed on the top of the device's package. See the "Device Part Marking" section in Module 1.
- The expected T_J range for the design determines the I_{CCPO} minimum requirement. Use the applicable ranges in the junction temperature column to find the associated current values in the appropriate new or old requirements column according to the date code. Then choose the highest of these current values to serve as the minimum I_{CCPO} requirement that must be met. For example, if the junction temperature for a given design is $-25^{\circ}\text{C} \leq T_J \leq 75^{\circ}\text{C}$, then the new minimum I_{CCPO} requirement is 1.5A. If $5^{\circ}\text{C} \leq T_J \leq 90^{\circ}\text{C}$, then the new minimum I_{CCPO} requirement is 0.5A.
- The I_{CCPO} requirement applies for a brief time (commonly only a few milliseconds) when V_{CCINT} ramps from 0 to 2.5V.
- The ramp time is measured from GND to V_{CCINT} max on a fully loaded board.
- During power-on, the V_{CCINT} ramp must increase steadily in voltage with no dips.
- For more information on designing to meet the power-on specifications, refer to the application note [XAPP450 "Power-On Current Requirements for the Spartan-II and Spartan-IIe Families"](#)

DC Input and Output Levels

Values for V_{IL} and V_{IH} are recommended input voltages. Values for V_{OL} and V_{OH} are guaranteed output voltages over the recommended operating conditions. Only selected standards are tested. These are chosen to ensure that all

standards meet their specifications. The selected standards are tested at minimum V_{CCO} with the respective I_{OL} and I_{OH} currents shown. Other standards are sample tested.

Input/Output Standard	V_{IL}		V_{IH}		V_{OL}	V_{OH}	I_{OL}	I_{OH}
	V, Min	V, Max	V, Min	V, Max	V, Max	V, Min	mA	mA
LVTTL ⁽¹⁾	-0.5	0.8	2.0	5.5	0.4	2.4	24	-24
LVC MOS2	-0.5	0.7	1.7	5.5	0.4	1.9	12	-12
PCI, 3.3V	-0.5	44% V_{CCINT}	60% V_{CCINT}	$V_{CCO} + 0.5$	10% V_{CCO}	90% V_{CCO}	Note (2)	Note (2)
PCI, 5.0V	-0.5	0.8	2.0	5.5	0.55	2.4	Note (2)	Note (2)
GTL	-0.5	$V_{REF} - 0.05$	$V_{REF} + 0.05$	3.6	0.4	N/A	40	N/A
GTL+	-0.5	$V_{REF} - 0.1$	$V_{REF} + 0.1$	3.6	0.6	N/A	36	N/A
HSTL I	-0.5	$V_{REF} - 0.1$	$V_{REF} + 0.1$	3.6	0.4	$V_{CCO} - 0.4$	8	-8
HSTL III	-0.5	$V_{REF} - 0.1$	$V_{REF} + 0.1$	3.6	0.4	$V_{CCO} - 0.4$	24	-8
HSTL IV	-0.5	$V_{REF} - 0.1$	$V_{REF} + 0.1$	3.6	0.4	$V_{CCO} - 0.4$	48	-8
SSTL3 I	-0.5	$V_{REF} - 0.2$	$V_{REF} + 0.2$	3.6	$V_{REF} - 0.6$	$V_{REF} + 0.6$	8	-8
SSTL3 II	-0.5	$V_{REF} - 0.2$	$V_{REF} + 0.2$	3.6	$V_{REF} - 0.8$	$V_{REF} + 0.8$	16	-16
SSTL2 I	-0.5	$V_{REF} - 0.2$	$V_{REF} + 0.2$	3.6	$V_{REF} - 0.6$	$V_{REF} + 0.6$	7.6	-7.6
SSTL2 II	-0.5	$V_{REF} - 0.2$	$V_{REF} + 0.2$	3.6	$V_{REF} - 0.8$	$V_{REF} + 0.8$	15.2	-15.2

IOB Input Delay Adjustments for Different Standards⁽¹⁾

Input delays associated with the pad are specified for LVTTTL. For other standards, adjust the delays by the values shown. A delay adjusted in this way constitutes a worst-case limit.

Symbol	Description	Standard	Speed Grade		Units
			-6	-5	
Data Input Delay Adjustments					
T _{ILVTTL}	Standard-specific data input delay adjustments	LVTTTL	0	0	ns
T _{ILVCMOS2}		LVCMSO2	−0.04	−0.05	ns
T _{IPCI33_3}		PCI, 33 MHz, 3.3V	−0.11	−0.13	ns
T _{IPCI33_5}		PCI, 33 MHz, 5.0V	0.26	0.30	ns
T _{IPCI66_3}		PCI, 66 MHz, 3.3V	−0.11	−0.13	ns
T _{IGTL}		GTL	0.20	0.24	ns
T _{IGTLP}		GTL+	0.11	0.13	ns
T _{IHSTL}		HSTL	0.03	0.04	ns
T _{ISSTL2}		SSTL2	−0.08	−0.09	ns
T _{ISSTL3}		SSTL3	−0.04	−0.05	ns
T _{ICTT}		CTT	0.02	0.02	ns
T _{IAGP}		AGP	−0.06	−0.07	ns

Notes:

1. Input timing for LVTTTL is measured at 1.4V. For other I/O standards, see the table ["Delay Measurement Methodology," page 60](#).

Calculation of T_{IOOP} as a Function of Capacitance

T_{IOOP} is the propagation delay from the O Input of the IOB to the pad. The values for T_{IOOP} are based on the standard capacitive load (C_{SL}) for each I/O standard as listed in the table "Constants for Calculating T_{IOOP} ", below.

For other capacitive loads, use the formulas below to calculate an adjusted propagation delay, T_{IOOP1} .

$$T_{IOOP1} = T_{IOOP} + Adj + (C_{LOAD} - C_{SL}) * F_L$$

Where:

Adj is selected from "IOB Output Delay Adjustments for Different Standards", page 59, according to the I/O standard used

C_{LOAD} is the capacitive load for the design

F_L is the capacitance scaling factor

Delay Measurement Methodology

Standard	$V_L^{(1)}$	$V_H^{(1)}$	Meas. Point	V_{REF} Typ ⁽²⁾
LVTTL	0	3	1.4	-
LVC MOS2	0	2.5	1.125	-
PCI33_5	Per PCI Spec			-
PCI33_3	Per PCI Spec			-
PCI66_3	Per PCI Spec			-
GTL	$V_{REF} - 0.2$	$V_{REF} + 0.2$	V_{REF}	0.80
GTL+	$V_{REF} - 0.2$	$V_{REF} + 0.2$	V_{REF}	1.0
HSTL Class I	$V_{REF} - 0.5$	$V_{REF} + 0.5$	V_{REF}	0.75
HSTL Class III	$V_{REF} - 0.5$	$V_{REF} + 0.5$	V_{REF}	0.90
HSTL Class IV	$V_{REF} - 0.5$	$V_{REF} + 0.5$	V_{REF}	0.90
SSTL3 I and II	$V_{REF} - 1.0$	$V_{REF} + 1.0$	V_{REF}	1.5
SSTL2 I and II	$V_{REF} - 0.75$	$V_{REF} + 0.75$	V_{REF}	1.25
CTT	$V_{REF} - 0.2$	$V_{REF} + 0.2$	V_{REF}	1.5
AGP	$V_{REF} - (0.2 \times V_{CCO})$	$V_{REF} + (0.2 \times V_{CCO})$	V_{REF}	Per AGP Spec

Notes:

- Input waveform switches between V_L and V_H .
- Measurements are made at V_{REF} Typ, Maximum, and Minimum. Worst-case values are reported.
- I/O parameter measurements are made with the capacitance values shown in the table, "Constants for Calculating T_{IOOP} ". See Xilinx application note [XAPP179](#) for the appropriate terminations.
- I/O standard measurements are reflected in the IBIS model information except where the IBIS format precludes it.

Constants for Calculating T_{IOOP}

Standard	$C_{SL}^{(1)}$ (pF)	F_L (ns/pF)
LVTTL Fast Slew Rate, 2 mA drive	35	0.41
LVTTL Fast Slew Rate, 4 mA drive	35	0.20
LVTTL Fast Slew Rate, 6 mA drive	35	0.13
LVTTL Fast Slew Rate, 8 mA drive	35	0.079
LVTTL Fast Slew Rate, 12 mA drive	35	0.044
LVTTL Fast Slew Rate, 16 mA drive	35	0.043
LVTTL Fast Slew Rate, 24 mA drive	35	0.033
LVTTL Slow Slew Rate, 2 mA drive	35	0.41
LVTTL Slow Slew Rate, 4 mA drive	35	0.20
LVTTL Slow Slew Rate, 6 mA drive	35	0.100
LVTTL Slow Slew Rate, 8 mA drive	35	0.086
LVTTL Slow Slew Rate, 12 mA drive	35	0.058
LVTTL Slow Slew Rate, 16 mA drive	35	0.050
LVTTL Slow Slew Rate, 24 mA drive	35	0.048
LVC MOS2	35	0.041
PCI 33 MHz 5V	50	0.050
PCI 33 MHz 3.3V	10	0.050
PCI 66 MHz 3.3V	10	0.033
GTL	0	0.014
GTL+	0	0.017
HSTL Class I	20	0.022
HSTL Class III	20	0.016
HSTL Class IV	20	0.014
SSTL2 Class I	30	0.028
SSTL2 Class II	30	0.016
SSTL3 Class I	30	0.029
SSTL3 Class II	30	0.016
CTT	20	0.035
AGP	10	0.037

Notes:

- I/O parameter measurements are made with the capacitance values shown above. See Xilinx application note [XAPP179](#) for the appropriate terminations.
- I/O standard measurements are reflected in the IBIS model information except where the IBIS format precludes it.

CLB Arithmetic Switching Characteristics

Setup times not listed explicitly can be approximated by decreasing the combinatorial delays by the setup time adjustment listed. Precise values are provided by the timing analyzer.

Symbol	Description	Speed Grade				Units
		-6		-5		
		Min	Max	Min	Max	
Combinatorial Delays						
T _{OPX}	F operand inputs to X via XOR	-	0.8	-	0.9	ns
T _{OPXB}	F operand input to XB output	-	1.3	-	1.5	ns
T _{OPY}	F operand input to Y via XOR	-	1.7	-	2.0	ns
T _{OPYB}	F operand input to YB output	-	1.7	-	2.0	ns
T _{OPCYF}	F operand input to COUT output	-	1.3	-	1.5	ns
T _{OPGY}	G operand inputs to Y via XOR	-	0.9	-	1.1	ns
T _{OPGYB}	G operand input to YB output	-	1.6	-	2.0	ns
T _{OPCYG}	G operand input to COUT output	-	1.2	-	1.4	ns
T _{BXCY}	BX initialization input to COUT	-	0.9	-	1.0	ns
T _{CINX}	CIN input to X output via XOR	-	0.4	-	0.5	ns
T _{CINXB}	CIN input to XB	-	0.1	-	0.1	ns
T _{CINY}	CIN input to Y via XOR	-	0.5	-	0.6	ns
T _{CINYB}	CIN input to YB	-	0.6	-	0.7	ns
T _{BYP}	CIN input to COUT output	-	0.1	-	0.1	ns
Multiplier Operation						
T _{FANDXB}	F1/2 operand inputs to XB output via AND	-	0.5	-	0.5	ns
T _{FANDYB}	F1/2 operand inputs to YB output via AND	-	0.9	-	1.1	ns
T _{FANDCY}	F1/2 operand inputs to COUT output via AND	-	0.5	-	0.6	ns
T _{GANDYB}	G1/2 operand inputs to YB output via AND	-	0.6	-	0.7	ns
T _{GANDCY}	G1/2 operand inputs to COUT output via AND	-	0.2	-	0.2	ns
Setup/Hold Times with Respect to Clock CLK ⁽¹⁾						
T _{CCKX} / T _{CKCX}	CIN input to FFX	1.1 / 0	-	1.2 / 0	-	ns
T _{CCKY} / T _{CKCY}	CIN input to FFY	1.2 / 0	-	1.3 / 0	-	ns

Notes:

1. A zero hold time listing indicates no hold time or a negative hold time.

XC2S50 Device Pinouts

XC2S50 Pad Name		TQ144	PQ208	FG256	Bndry Scan
Function	Bank				
GND	-	P143	P1	GND*	-
TMS	-	P142	P2	D3	-
I/O	7	P141	P3	C2	149
I/O	7	-	-	A2	152
I/O	7	P140	P4	B1	155
I/O	7	-	-	E3	158
I/O	7	-	P5	D2	161
GND	-	-	-	GND*	-
I/O, V _{REF}	7	P139	P6	C1	164
I/O	7	-	P7	F3	167
I/O	7	-	-	E2	170
I/O	7	P138	P8	E4	173
I/O	7	P137	P9	D1	176
I/O	7	P136	P10	E1	179
GND	-	P135	P11	GND*	-
V _{CCO}	7	-	P12	V _{CCO} Bank 7*	-
V _{CCINT}	-	-	P13	V _{CCINT} *	-
I/O	7	P134	P14	F2	182
I/O	7	P133	P15	G3	185
I/O	7	-	-	F1	188
I/O	7	-	P16	F4	191
I/O	7	-	P17	F5	194
I/O	7	-	P18	G2	197
GND	-	-	P19	GND*	-
I/O, V _{REF}	7	P132	P20	H3	200
I/O	7	P131	P21	G4	203
I/O	7	-	-	H2	206
I/O	7	P130	P22	G5	209
I/O	7	-	P23	H4	212
I/O, IRDY ⁽¹⁾	7	P129	P24	G1	215
GND	-	P128	P25	GND*	-
V _{CCO}	7	P127	P26	V _{CCO} Bank 7*	-
V _{CCO}	6	P127	P26	V _{CCO} Bank 6*	-
I/O, TRDY ⁽¹⁾	6	P126	P27	J2	218
V _{CCINT}	-	P125	P28	V _{CCINT} *	-
I/O	6	P124	P29	H1	224
I/O	6	-	-	J4	227
I/O	6	P123	P30	J1	230
I/O, V _{REF}	6	P122	P31	J3	233

XC2S50 Device Pinouts (Continued)

XC2S50 Pad Name		TQ144	PQ208	FG256	Bndry Scan
Function	Bank				
GND	-	-	P32	GND*	-
I/O	6	-	P33	K5	236
I/O	6	-	P34	K2	239
I/O	6	-	P35	K1	242
I/O	6	-	-	K3	245
I/O	6	P121	P36	L1	248
I/O	6	P120	P37	L2	251
V _{CCINT}	-	-	P38	V _{CCINT} *	-
V _{CCO}	6	-	P39	V _{CCO} Bank 6*	-
GND	-	P119	P40	GND*	-
I/O	6	P118	P41	K4	254
I/O	6	P117	P42	M1	257
I/O	6	P116	P43	L4	260
I/O	6	-	-	M2	263
I/O	6	-	P44	L3	266
I/O, V _{REF}	6	P115	P45	N1	269
GND	-	-	-	GND*	-
I/O	6	-	P46	P1	272
I/O	6	-	-	L5	275
I/O	6	P114	P47	N2	278
I/O	6	-	-	M4	281
I/O	6	P113	P48	R1	284
I/O	6	P112	P49	M3	287
M1	-	P111	P50	P2	290
GND	-	P110	P51	GND*	-
M0	-	P109	P52	N3	291
V _{CCO}	6	P108	P53	V _{CCO} Bank 6*	-
V _{CCO}	5	P107	P53	V _{CCO} Bank 5*	-
M2	-	P106	P54	R3	292
I/O	5	-	-	N5	299
I/O	5	P103	P57	T2	302
I/O	5	-	-	P5	305
I/O	5	-	P58	T3	308
GND	-	-	-	GND*	-
I/O, V _{REF}	5	P102	P59	T4	311
I/O	5	-	P60	M6	314
I/O	5	-	-	T5	317
I/O	5	P101	P61	N6	320
I/O	5	P100	P62	R5	323

XC2S50 Device Pinouts (Continued)

XC2S50 Pad Name		TQ144	PQ208	FG256	Bndry Scan
Function	Bank				
I/O	3	-	-	J14	503
I/O	3	P56	P127	K15	506
V _{CCINT}	-	P55	P128	V _{CCINT} *	-
I/O, TRDY ⁽¹⁾	3	P54	P129	J15	512
V _{CCO}	3	P53	P130	V _{CCO} Bank 3*	-
V _{CCO}	2	P53	P130	V _{CCO} Bank 2*	-
GND	-	P52	P131	GND*	-
I/O, IRDY ⁽¹⁾	2	P51	P132	H16	515
I/O	2	-	P133	H14	518
I/O	2	P50	P134	H15	521
I/O	2	-	-	J13	524
I/O (D3)	2	P49	P135	G16	527
I/O, V _{REF}	2	P48	P136	H13	530
GND	-	-	P137	GND*	-
I/O	2	-	P138	G14	533
I/O	2	-	P139	G15	536
I/O	2	-	P140	G12	539
I/O	2	-	-	F16	542
I/O	2	P47	P141	G13	545
I/O (D2)	2	P46	P142	F15	548
V _{CCINT}	-	-	P143	V _{CCINT} *	-
V _{CCO}	2	-	P144	V _{CCO} Bank 2*	-
GND	-	P45	P145	GND*	-
I/O (D1)	2	P44	P146	E16	551
I/O	2	P43	P147	F14	554
I/O	2	P42	P148	D16	557
I/O	2	-	-	F12	560
I/O	2	-	P149	E15	563
I/O, V _{REF}	2	P41	P150	F13	566
GND	-	-	-	GND*	-
I/O	2	-	P151	E14	569
I/O	2	-	-	C16	572
I/O	2	P40	P152	E13	575
I/O	2	-	-	B16	578
I/O (DIN, D0)	2	P39	P153	D14	581
I/O (DOUT, BUSY)	2	P38	P154	C15	584
CCLK	2	P37	P155	D15	587
V _{CCO}	2	P36	P156	V _{CCO} Bank 2*	-

XC2S50 Device Pinouts (Continued)

XC2S50 Pad Name		TQ144	PQ208	FG256	Bndry Scan
Function	Bank				
V _{CCO}	1	P35	P156	V _{CCO} Bank 1*	-
TDO	2	P34	P157	B14	-
GND	-	P33	P158	GND*	-
TDI	-	P32	P159	A15	-
I/O (\overline{CS})	1	P31	P160	B13	0
I/O (\overline{WRITE})	1	P30	P161	C13	3
I/O	1	-	-	C12	6
I/O	1	P29	P162	A14	9
I/O	1	-	-	D12	12
I/O	1	-	P163	B12	15
GND	-	-	-	GND*	-
I/O, V _{REF}	1	P28	P164	C11	18
I/O	1	-	P165	A13	21
I/O	1	-	-	D11	24
I/O	1	-	P166	A12	27
I/O	1	P27	P167	E11	30
I/O	1	P26	P168	B11	33
GND	-	P25	P169	GND*	-
V _{CCO}	1	-	P170	V _{CCO} Bank 1*	-
V _{CCINT}	-	P24	P171	V _{CCINT} *	-
I/O	1	P23	P172	A11	36
I/O	1	P22	P173	C10	39
I/O	1	-	P174	B10	45
I/O	1	-	P175	D10	48
I/O	1	-	P176	A10	51
GND	-	-	P177	GND*	-
I/O, V _{REF}	1	P21	P178	B9	54
I/O	1	-	P179	E10	57
I/O	1	-	-	A9	60
I/O	1	P20	P180	D9	63
I/O	1	P19	P181	A8	66
I, GCK2	1	P18	P182	C9	72
GND	-	P17	P183	GND*	-
V _{CCO}	1	P16	P184	V _{CCO} Bank 1*	-
V _{CCO}	0	P16	P184	V _{CCO} Bank 0*	-
I, GCK3	0	P15	P185	B8	73
V _{CCINT}	-	P14	P186	V _{CCINT} *	-
I/O	0	P13	P187	A7	80

XC2S150 Device Pinouts (Continued)

XC2S150 Pad Name		PQ208	FG256	FG456	Bndry Scan
Function	Bank				
I/O	6	P46	P1	T4	404
I/O	6	-	L5	W1	407
I/O	6	-	-	V2	410
I/O	6	-	-	U4	413
I/O	6	P47	N2	Y1	416
GND	-	-	GND*	GND*	-
I/O	6	-	M4	W2	419
I/O	6	-	-	V3	422
I/O	6	-	-	V4	425
I/O	6	P48	R1	Y2	428
I/O	6	P49	M3	W3	431
M1	-	P50	P2	U5	434
GND	-	P51	GND*	GND*	-
M0	-	P52	N3	AB2	435
V _{CCO}	6	P53	V _{CCO} Bank 6*	V _{CCO} Bank 6*	-
V _{CCO}	5	P53	V _{CCO} Bank 5*	V _{CCO} Bank 5*	-
M2	-	P54	R3	Y4	436
I/O	5	-	-	W5	443
I/O	5	-	-	AB3	446
I/O	5	-	N5	V7	449
GND	-	-	GND*	GND*	-
I/O	5	P57	T2	Y6	452
I/O	5	-	-	AA4	455
I/O	5	-	-	AB4	458
I/O	5	-	P5	W6	461
I/O	5	P58	T3	Y7	464
GND	-	-	GND*	GND*	-
V _{CCO}	5	-	V _{CCO} Bank 5*	V _{CCO} Bank 5*	-
I/O, V _{REF}	5	P59	T4	AA5	467
I/O	5	P60	M6	AB5	470
I/O	5	-	-	V8	473
I/O	5	-	-	AA6	476
I/O	5	-	T5	AB6	479
I/O	5	P61	N6	AA7	482
I/O	5	-	-	W7	485
I/O, V _{REF}	5	P62	R5	W8	488
I/O	5	P63	P6	Y8	491
GND	-	P64	GND*	GND*	-

XC2S150 Device Pinouts (Continued)

XC2S150 Pad Name		PQ208	FG256	FG456	Bndry Scan
Function	Bank				
V _{CCO}	5	P65	V _{CCO} Bank 5*	V _{CCO} Bank 5*	-
V _{CCINT}	-	P66	V _{CCINT} *	V _{CCINT} *	-
I/O	5	P67	R6	AA8	494
I/O	5	P68	M7	V9	497
I/O	5	-	-	W9	503
I/O	5	-	-	AB9	506
I/O	5	P69	N7	Y9	509
I/O	5	-	-	V10	512
I/O	5	P70	T6	W10	518
I/O	5	P71	P7	AB10	521
GND	-	P72	GND*	GND*	-
V _{CCO}	5	-	V _{CCO} Bank 5*	V _{CCO} Bank 5*	-
I/O, V _{REF}	5	P73	P8	Y10	524
I/O	5	P74	R7	V11	527
I/O	5	-	T7	W11	530
I/O	5	P75	T8	AB11	533
I/O	5	-	-	U11	536
V _{CCINT}	-	P76	V _{CCINT} *	V _{CCINT} *	-
I, GCK1	5	P77	R8	Y11	545
V _{CCO}	5	P78	V _{CCO} Bank 5*	V _{CCO} Bank 5*	-
V _{CCO}	4	P78	V _{CCO} Bank 4*	V _{CCO} Bank 4*	-
GND	-	P79	GND*	GND*	-
I, GCK0	4	P80	N8	W12	546
I/O	4	P81	N9	U12	550
I/O	4	-	-	V12	553
I/O	4	P82	R9	Y12	556
I/O	4	-	N10	AA12	559
I/O	4	P83	T9	AB13	562
I/O, V _{REF}	4	P84	P9	AA13	565
V _{CCO}	4	-	V _{CCO} Bank 4*	V _{CCO} Bank 4*	-
GND	-	P85	GND*	GND*	-
I/O	4	P86	M10	Y13	568
I/O	4	P87	R10	V13	571
I/O	4	-	-	W14	577
I/O	4	P88	P10	AA14	580
I/O	4	-	-	V14	583
I/O	4	-	-	Y14	586
I/O	4	P89	T10	AB15	592

XC2S150 Device Pinouts (Continued)

XC2S150 Pad Name		PQ208	FG256	FG456	Bndry Scan
Function	Bank				
I/O	4	P90	R11	AA15	595
V _{CCINT}	-	P91	V _{CCINT} *	V _{CCINT} *	-
V _{CCO}	4	P92	V _{CCO} Bank 4*	V _{CCO} Bank 4*	-
GND	-	P93	GND*	GND*	-
I/O	4	P94	M11	Y15	598
I/O, V _{REF}	4	P95	T11	AB16	601
I/O	4	-	-	AB17	604
I/O	4	P96	N11	V15	607
I/O	4	-	R12	Y16	610
I/O	4	-	-	AA17	613
I/O	4	-	-	W16	616
I/O	4	P97	P11	AB18	619
I/O, V _{REF}	4	P98	T12	AB19	622
V _{CCO}	4	-	V _{CCO} Bank 4*	V _{CCO} Bank 4*	-
GND	-	-	GND*	GND*	-
I/O	4	P99	T13	Y17	625
I/O	4	-	N12	V16	628
I/O	4	-	-	AA18	631
I/O	4	-	-	W17	634
I/O	4	P100	R13	AB20	637
GND	-	-	GND*	GND*	-
I/O	4	-	P12	AA19	640
I/O	4	-	-	V17	643
I/O	4	-	-	Y18	646
I/O	4	P101	P13	AA20	649
I/O	4	P102	T14	W18	652
GND	-	P103	GND*	GND*	-
DONE	3	P104	R14	Y19	655
V _{CCO}	4	P105	V _{CCO} Bank 4*	V _{CCO} Bank 4*	-
V _{CCO}	3	P105	V _{CCO} Bank 3*	V _{CCO} Bank 3*	-
PROGRAM	-	P106	P15	W20	658
I/O (INIT)	3	P107	N15	V19	659
I/O (D7)	3	P108	N14	Y21	662
I/O	3	-	-	V20	665
I/O	3	-	-	AA22	668
I/O	3	-	T15	W21	671
GND	-	-	GND*	GND*	-
I/O	3	P109	M13	U20	674

XC2S150 Device Pinouts (Continued)

XC2S150 Pad Name		PQ208	FG256	FG456	Bndry Scan
Function	Bank				
I/O	3	-	-	U19	677
I/O	3	-	-	V21	680
I/O	3	-	R16	T18	683
I/O	3	P110	M14	W22	686
GND	-	-	GND*	GND*	-
V _{CCO}	3	-	V _{CCO} Bank 3*	V _{CCO} Bank 3*	-
I/O, V _{REF}	3	P111	L14	U21	689
I/O	3	P112	M15	T20	692
I/O	3	-	-	T19	695
I/O	3	-	-	V22	698
I/O	3	-	L12	T21	701
I/O	3	P113	P16	R18	704
I/O	3	-	-	U22	707
I/O, V _{REF}	3	P114	L13	R19	710
I/O (D6)	3	P115	N16	T22	713
GND	-	P116	GND*	GND*	-
V _{CCO}	3	P117	V _{CCO} Bank 3*	V _{CCO} Bank 3*	-
V _{CCINT}	-	P118	V _{CCINT} *	V _{CCINT} *	-
I/O (D5)	3	P119	M16	R21	716
I/O	3	P120	K14	P18	719
I/O	3	-	-	P19	725
I/O	3	-	L16	P20	728
I/O	3	P121	K13	P21	731
I/O	3	-	-	N19	734
I/O	3	P122	L15	N18	740
I/O	3	P123	K12	N20	743
GND	-	P124	GND*	GND*	-
V _{CCO}	3	-	V _{CCO} Bank 3*	V _{CCO} Bank 3*	-
I/O, V _{REF}	3	P125	K16	N21	746
I/O (D4)	3	P126	J16	N22	749
I/O	3	-	J14	M19	752
I/O	3	P127	K15	M20	755
I/O	3	-	-	M18	758
V _{CCINT}	-	P128	V _{CCINT} *	V _{CCINT} *	-
I/O, TRDY ⁽¹⁾	3	P129	J15	M22	764
V _{CCO}	3	P130	V _{CCO} Bank 3*	V _{CCO} Bank 3*	-
V _{CCO}	2	P130	V _{CCO} Bank 2*	V _{CCO} Bank 2*	-
GND	-	P131	GND*	GND*	-

Additional XC2S150 Package Pins

PQ208

Not Connected Pins					
P55	P56	-	-	-	-

11/02/00

FG256

V _{CCINT} Pins					
C3	C14	D4	D13	E5	E12
M5	M12	N4	N13	P3	P14
V _{CCO} Bank 0 Pins					
E8	F8	-	-	-	-
V _{CCO} Bank 1 Pins					
E9	F9	-	-	-	-
V _{CCO} Bank 2 Pins					
H11	H12	-	-	-	-
V _{CCO} Bank 3 Pins					
J11	J12	-	-	-	-
V _{CCO} Bank 4 Pins					
L9	M9	-	-	-	-
V _{CCO} Bank 5 Pins					
L8	M8	-	-	-	-
V _{CCO} Bank 6 Pins					
J5	J6	-	-	-	-
V _{CCO} Bank 7 Pins					
H5	H6	-	-	-	-
GND Pins					
A1	A16	B2	B15	F6	F7
F10	F11	G6	G7	G8	G9
G10	G11	H7	H8	H9	H10
J7	J8	J9	J10	K6	K7
K8	K9	K10	K11	L6	L7
L10	L11	R2	R15	T1	T16
Not Connected Pins					
P4	R4	-	-	-	-

11/02/00

Additional XC2S150 Package Pins (Continued)

FG456

V _{CCINT} Pins					
E5	E18	F6	F17	G7	G8
G9	G14	G15	G16	H7	H16
J7	J16	P7	P16	R7	R16
T7	T8	T9	T14	T15	T16
U6	U17	V5	V18	-	-
V _{CCO} Bank 0 Pins					
F7	F8	F9	F10	G10	G11
V _{CCO} Bank 1 Pins					
F13	F14	F15	F16	G12	G13
V _{CCO} Bank 2 Pins					
G17	H17	J17	K16	K17	L16
V _{CCO} Bank 3 Pins					
M16	N16	N17	P17	R17	T17
V _{CCO} Bank 4 Pins					
T12	T13	U13	U14	U15	U16
V _{CCO} Bank 5 Pins					
T10	T11	U7	U8	U9	U10
V _{CCO} Bank 6 Pins					
M7	N6	N7	P6	R6	T6
V _{CCO} Bank 7 Pins					
G6	H6	J6	K6	K7	L7
GND Pins					
A1	A22	B2	B21	C3	C20
J9	J10	J11	J12	J13	J14
K9	K10	K11	K12	K13	K14
L9	L10	L11	L12	L13	L14
M9	M10	M11	M12	M13	M14
N9	N10	N11	N12	N13	N14
P9	P10	P11	P12	P13	P14
Y3	Y20	AA2	AA21	AB1	AB22
Not Connected Pins					
A2	A6	A12	A13	A14	B11
B16	C2	C8	C9	D1	D4
D18	D19	E13	E17	E19	F11
G2	G22	H21	J1	J4	K2
K18	K19	L2	L19	M2	M17
M21	N1	P1	P5	P22	R3
R20	R22	U3	U18	V6	W4
W13	W15	W19	Y5	Y22	AA1
AA3	AA9	AA10	AA11	AA16	AB7
AB8	AB12	AB14	AB21	-	-

11/02/00

XC2S200 Device Pinouts (Continued)

XC2S200 Pad Name		PQ208	FG256	FG456	Bndry Scan
Function	Bank				
I/O	6	-	-	T2	449
I/O	6	P43	L4	U1	452
GND	-	-	GND*	GND*	-
I/O	6	-	M2	R5	455
I/O	6	-	-	V1	458
I/O	6	-	-	T5	461
I/O	6	P44	L3	U2	464
I/O, V _{REF}	6	P45	N1	T3	467
V _{CCO}	6	-	V _{CCO} Bank 6*	V _{CCO} Bank 6*	-
GND	-	-	GND*	GND*	-
I/O	6	P46	P1	T4	470
I/O	6	-	L5	W1	473
GND	-	-	GND*	GND*	-
I/O	6	-	-	V2	476
I/O	6	-	-	U4	482
I/O, V _{REF}	6	P47	N2	Y1	485
GND	-	-	GND*	GND*	-
I/O	6	-	M4	W2	488
I/O	6	-	-	V3	491
I/O	6	-	-	V4	494
I/O	6	P48	R1	Y2	500
I/O	6	P49	M3	W3	503
M1	-	P50	P2	U5	506
GND	-	P51	GND*	GND*	-
M0	-	P52	N3	AB2	507
V _{CCO}	6	P53	V _{CCO} Bank 6*	V _{CCO} Bank 6*	-
V _{CCO}	5	P53	V _{CCO} Bank 5*	V _{CCO} Bank 5*	-
M2	-	P54	R3	Y4	508
I/O	5	-	-	W5	518
I/O	5	-	-	AB3	521
I/O	5	-	N5	V7	524
GND	-	-	GND*	GND*	-
I/O, V _{REF}	5	P57	T2	Y6	527
I/O	5	-	-	AA4	530
I/O	5	-	-	AB4	536
I/O	5	-	P5	W6	539
I/O	5	P58	T3	Y7	542
GND	-	-	GND*	GND*	-

XC2S200 Device Pinouts (Continued)

XC2S200 Pad Name		PQ208	FG256	FG456	Bndry Scan
Function	Bank				
V _{CCO}	5	-	V _{CCO} Bank 5*	V _{CCO} Bank 5*	-
I/O, V _{REF}	5	P59	T4	AA5	545
I/O	5	P60	M6	AB5	548
I/O	5	-	-	V8	551
I/O	5	-	-	AA6	554
I/O	5	-	T5	AB6	557
GND	-	-	GND*	GND*	-
I/O	5	P61	N6	AA7	560
I/O	5	-	-	W7	563
I/O, V _{REF}	5	P62	R5	W8	569
I/O	5	P63	P6	Y8	572
GND	-	P64	GND*	GND*	-
V _{CCO}	5	P65	V _{CCO} Bank 5*	V _{CCO} Bank 5*	-
V _{CCINT}	-	P66	V _{CCINT} *	V _{CCINT} *	-
I/O	5	P67	R6	AA8	575
I/O	5	P68	M7	V9	578
I/O	5	-	-	AB8	581
I/O	5	-	-	W9	584
I/O	5	-	-	AB9	587
GND	-	-	GND*	GND*	-
I/O	5	P69	N7	Y9	590
I/O	5	-	-	V10	593
I/O	5	-	-	AA9	596
I/O	5	P70	T6	W10	599
I/O	5	P71	P7	AB10	602
GND	-	P72	GND*	GND*	-
V _{CCO}	5	-	V _{CCO} Bank 5*	V _{CCO} Bank 5*	-
I/O, V _{REF}	5	P73	P8	Y10	605
I/O	5	P74	R7	V11	608
I/O	5	-	-	AA10	614
I/O	5	-	T7	W11	617
I/O	5	P75	T8	AB11	620
I/O	5	-	-	U11	623
V _{CCINT}	-	P76	V _{CCINT} *	V _{CCINT} *	-
I, GCK1	5	P77	R8	Y11	635
V _{CCO}	5	P78	V _{CCO} Bank 5*	V _{CCO} Bank 5*	-
V _{CCO}	4	P78	V _{CCO} Bank 4*	V _{CCO} Bank 4*	-
GND	-	P79	GND*	GND*	-

XC2S200 Device Pinouts (Continued)

XC2S200 Pad Name		PQ208	FG256	FG456	Bndry Scan
Function	Bank				
V _{CCO}	1	P156	V _{CCO} Bank 1*	V _{CCO} Bank 1*	-
TDO	2	P157	B14	A21	-
GND	-	P158	GND*	GND*	-
TDI	-	P159	A15	B20	-
I/O (\overline{CS})	1	P160	B13	C19	0
I/O (\overline{WRITE})	1	P161	C13	A20	3
I/O	1	-	-	B19	9
I/O	1	-	-	C18	12
I/O	1	-	C12	D17	15
GND	-	-	GND*	GND*	-
I/O, V _{REF}	1	P162	A14	A19	18
I/O	1	-	-	B18	21
I/O	1	-	-	E16	27
I/O	1	-	D12	C17	30
I/O	1	P163	B12	D16	33
GND	-	-	GND*	GND*	-
V _{CCO}	1	-	V _{CCO} Bank 1*	V _{CCO} Bank 1*	-
I/O, V _{REF}	1	P164	C11	A18	36
I/O	1	P165	A13	B17	39
I/O	1	-	-	E15	42
I/O	1	-	-	A17	45
I/O	1	-	D11	D15	48
GND	-	-	GND*	GND*	-
I/O	1	P166	A12	C16	51
I/O	1	-	-	D14	54
I/O, V _{REF}	1	P167	E11	E14	60
I/O	1	P168	B11	A16	63
GND	-	P169	GND*	GND*	-
V _{CCO}	1	P170	V _{CCO} Bank 1*	V _{CCO} Bank 1*	-
V _{CCINT}	-	P171	V _{CCINT} *	V _{CCINT} *	-
I/O	1	P172	A11	C15	66
I/O	1	P173	C10	B15	69
I/O	1	-	-	E13	72
I/O	1	-	-	A15	75
I/O	1	-	-	F12	78
GND	-	-	GND*	GND*	-
I/O	1	P174	B10	C14	81
I/O	1	-	-	B14	84
I/O	1	-	-	A14	87

XC2S200 Device Pinouts (Continued)

XC2S200 Pad Name		PQ208	FG256	FG456	Bndry Scan
Function	Bank				
I/O	1	P175	D10	D13	90
I/O	1	P176	A10	C13	93
GND	-	P177	GND*	GND*	-
V _{CCO}	1	-	V _{CCO} Bank 1*	V _{CCO} Bank 1*	-
I/O, V _{REF}	1	P178	B9	B13	96
I/O	1	P179	E10	E12	99
I/O	1	-	-	A13	105
I/O	1	-	A9	B12	108
I/O	1	P180	D9	D12	111
I/O	1	-	-	C12	114
I/O	1	P181	A8	D11	120
I, GCK2	1	P182	C9	A11	126
GND	-	P183	GND*	GND*	-
V _{CCO}	1	P184	V _{CCO} Bank 1*	V _{CCO} Bank 1*	-
V _{CCO}	0	P184	V _{CCO} Bank 0*	V _{CCO} Bank 0*	-
I, GCK3	0	P185	B8	C11	127
V _{CCINT}	-	P186	V _{CCINT} *	V _{CCINT} *	-
I/O	0	-	-	E11	137
I/O	0	P187	A7	A10	140
I/O	0	-	D8	B10	143
I/O	0	-	-	F11	146
I/O	0	P188	A6	C10	152
I/O, V _{REF}	0	P189	B7	A9	155
V _{CCO}	0	-	V _{CCO} Bank 0*	V _{CCO} Bank 0*	-
GND	-	P190	GND*	GND*	-
I/O	0	P191	C8	B9	158
I/O	0	P192	D7	E10	161
I/O	0	-	-	C9	164
I/O	0	-	-	D10	167
I/O	0	P193	E7	A8	170
GND	-	-	GND*	GND*	-
I/O	0	-	-	D9	173
I/O	0	-	-	B8	176
I/O	0	-	-	C8	179
I/O	0	P194	C7	E9	182
I/O	0	P195	B6	A7	185
V _{CCINT}	-	P196	V _{CCINT} *	V _{CCINT} *	-
V _{CCO}	0	P197	V _{CCO} Bank 0*	V _{CCO} Bank 0*	-