AMD Xilinx - XC2S50-5FG256C Datasheet

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Understanding Embedded - FPGAs (Field Programmable Gate Array)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications,

Details	
Product Status	Active
Number of LABs/CLBs	384
Number of Logic Elements/Cells	1728
Total RAM Bits	32768
Number of I/O	176
Number of Gates	50000
Voltage - Supply	2.375V ~ 2.625V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 85°C (TJ)
Package / Case	256-BGA
Supplier Device Package	256-FBGA (17x17)
Purchase URL	https://www.e-xfl.com/product-detail/xilinx/xc2s50-5fg256c

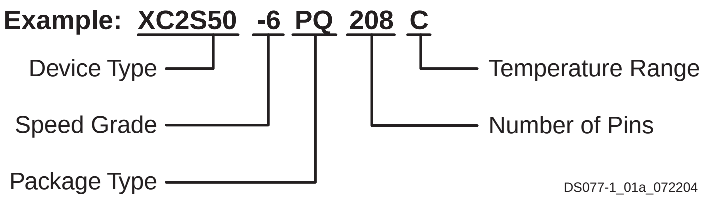
Email: info@E-XFL.COM

Address: Room A, 16/F, Full Win Commercial Centre, 573 Nathan Road, Mongkok, Hong Kong

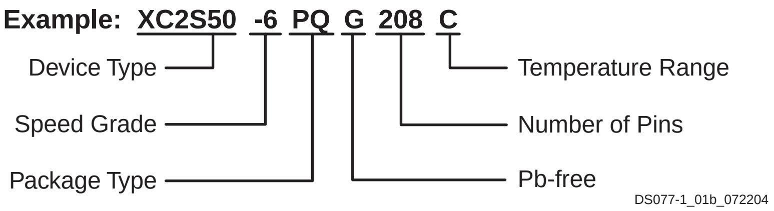
Ordering Information

Spartan-II devices are available in both standard and Pb-free packaging options for all device/package combinations. The Pb-free packages include a special "G" character in the ordering code.

Standard Packaging



Pb-Free Packaging



Device Ordering Options

Device
XC2S15
XC2S30
XC2S50
XC2S100
XC2S150
XC2S200

Speed Grade	
-5	Standard Performance
-6	Higher Performance[1]

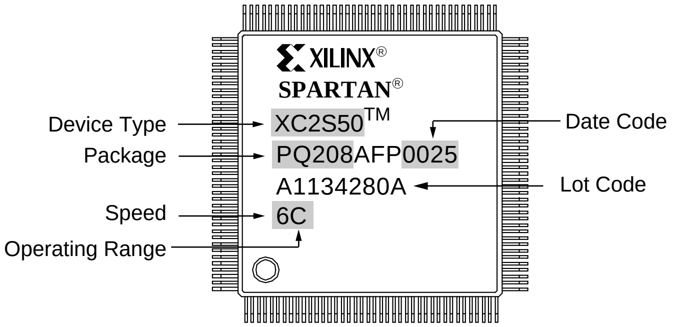
Number of Pins / Package Type	
VQ(G)100	100-pin Plastic Very Thin QFP
CS(G)144	144-ball Chip-Scale BGA
TQ(G)144	144-pin Plastic Thin QFP
PQ(G)208	208-pin Plastic QFP
FG(G)256	256-ball Fine Pitch BGA
FG(G)456	456-ball Fine Pitch BGA

Temperature Range (T_J)	
C = Commercial	0°C to +85°C
I = Industrial	–40°C to +100°C

Notes:

1. The -6 speed grade is exclusively available in the Commercial temperature range.

Device Part Marking



Sample package with part marking for XC2S50-6PQ208C.

ds001-1_02_090303

Local Routing

The local routing resources, as shown in Figure 6, provide the following three types of connections:

- Interconnections among the LUTs, flip-flops, and General Routing Matrix (GRM)
- Internal CLB feedback paths that provide high-speed connections to LUTs within the same CLB, chaining them together with minimal routing delay
- Direct paths that provide high-speed connections between horizontally adjacent CLBs, eliminating the delay of the GRM

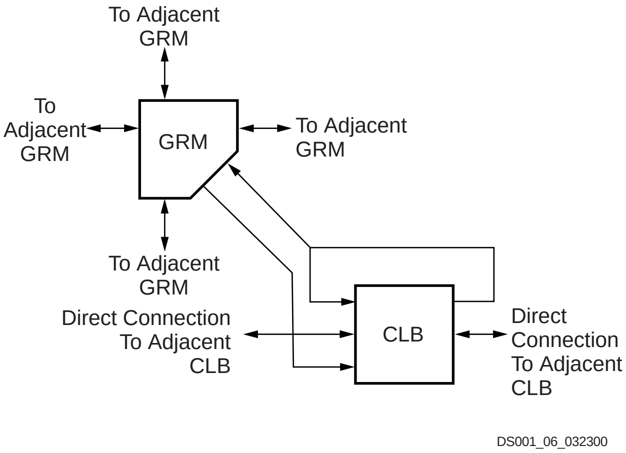


Figure 6: **Spartan-II Local Routing**

General Purpose Routing

Most Spartan-II FPGA signals are routed on the general purpose routing, and consequently, the majority of interconnect resources are associated with this level of the routing hierarchy. The general routing resources are located in horizontal and vertical routing channels associated with the rows and columns CLBs. The general-purpose routing resources are listed below.

- Adjacent to each CLB is a General Routing Matrix (GRM). The GRM is the switch matrix through which horizontal and vertical routing resources connect, and is also the means by which the CLB gains access to the general purpose routing.
- 24 single-length lines route GRM signals to adjacent GRMs in each of the four directions.
- 96 buffered Hex lines route GRM signals to other GRMs six blocks away in each one of the four directions. Organized in a staggered pattern, Hex lines may be driven only at their endpoints. Hex-line signals can be accessed either at the endpoints or at the midpoint (three blocks from the source). One third of the Hex lines are bidirectional, while the remaining ones are unidirectional.
- 12 Longlines are buffered, bidirectional wires that distribute signals across the device quickly and efficiently. Vertical Longlines span the full height of the device, and horizontal ones span the full width of the device.

I/O Routing

Spartan-II devices have additional routing resources around their periphery that form an interface between the CLB array and the IOBs. This additional routing, called the VersaRing, facilitates pin-swapping and pin-locking, such that logic redesigns can adapt to existing PCB layouts. Time-to-market is reduced, since PCBs and other system components can be manufactured while the logic design is still in progress.

Dedicated Routing

Some classes of signal require dedicated routing resources to maximize performance. In the Spartan-II architecture, dedicated routing resources are provided for two classes of signal.

- Horizontal routing resources are provided for on-chip 3-state busses. Four partitionable bus lines are provided per CLB row, permitting multiple busses within a row, as shown in Figure 7.
- Two dedicated nets per CLB propagate carry signals vertically to the adjacent CLB.

Global Routing

Global Routing resources distribute clocks and other signals with very high fanout throughout the device. Spartan-II devices include two tiers of global routing resources referred to as primary and secondary global routing resources.

- The primary global routing resources are four dedicated global nets with dedicated input pins that are designed to distribute high-fanout clock signals with minimal skew. Each global clock net can drive all CLB, IOB, and block RAM clock pins. The primary global nets may only be driven by global buffers. There are four global buffers, one for each global net.
- The secondary global routing resources consist of 24 backbone lines, 12 across the top of the chip and 12 across bottom. From these lines, up to 12 unique signals per column can be distributed via the 12 longlines in the column. These secondary resources are more flexible than the primary resources since they are not restricted to routing only to clock pins.

Configuration

Configuration is the process by which the bitstream of a design, as generated by the Xilinx software, is loaded into the internal configuration memory of the FPGA. Spartan-II devices support both serial configuration, using the master/slave serial and JTAG modes, as well as byte-wide configuration employing the Slave Parallel mode.

Configuration File

Spartan-II devices are configured by sequentially loading frames of data that have been concatenated into a configuration file. Table 8 shows how much nonvolatile storage space is needed for Spartan-II devices.

It is important to note that, while a PROM is commonly used to store configuration data before loading them into the FPGA, it is by no means required. Any of a number of different kinds of under populated nonvolatile storage already available either on or off the board (i.e., hard drives, FLASH cards, etc.) can be used. For more information on configuration without a PROM, refer to XAPP098, *The Low-Cost, Efficient Serial Configuration of Spartan FPGAs*.

Table 8: **Spartan-II Configuration File Size**

Device	Configuration File Size (Bits)
XC2S15	197,696
XC2S30	336,768
XC2S50	559,200
XC2S100	781,216
XC2S150	1,040,096
XC2S200	1,335,840

Modes

Spartan-II devices support the following four configuration modes:

- Slave Serial mode
- Master Serial mode
- Slave Parallel mode
- Boundary-scan mode

The Configuration mode pins (M2, M1, M0) select among these configuration modes with the option in each case of having the IOB pins either pulled up or left floating prior to the end of configuration. The selection codes are listed in Table 9.

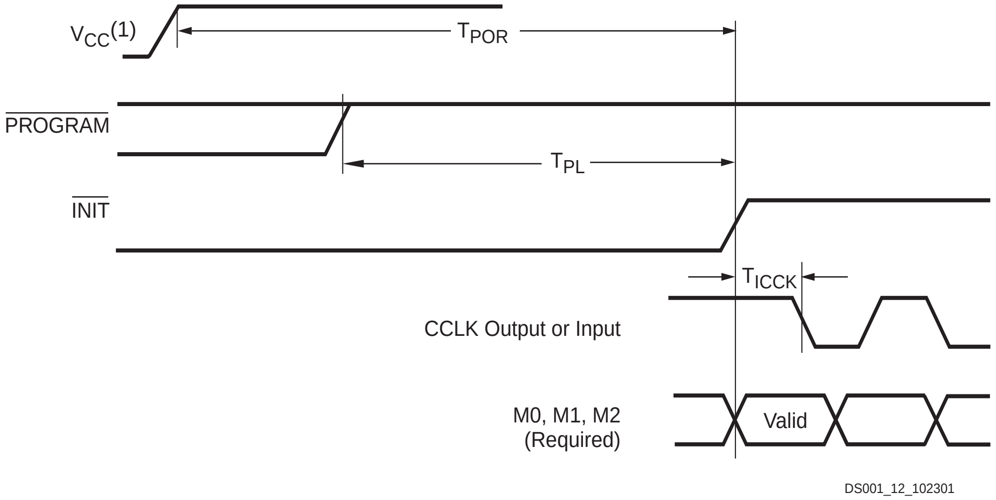
Configuration through the boundary-scan port is always available, independent of the mode selection. Selecting the boundary-scan mode simply turns off the other modes. The three mode pins have internal pull-up resistors, and default to a logic High if left unconnected.

Table 9: **Configuration Modes**

Configuration Mode	Preconfiguration Pull-ups	M0	M1	M2	CCLK Direction	Data Width	Serial D_{OUT}
Master Serial mode	No	0	0	0	Out	1	Yes
	Yes	0	0	1			
Slave Parallel mode	Yes	0	1	0	In	8	No
	No	0	1	1			
Boundary-Scan mode	Yes	1	0	0	N/A	1	No
	No	1	0	1			
Slave Serial mode	Yes	1	1	0	In	1	Yes
	No	1	1	1			

Notes:

1. During power-on and throughout configuration, the I/O drivers will be in a high-impedance state. After configuration, all unused I/Os (those not assigned signals) will remain in a high-impedance state. Pins used as outputs may pulse High at the end of configuration (see Answer 10504).
2. If the Mode pins are set for preconfiguration pull-ups, those resistors go into effect once the rising edge of INIT samples the Mode pins. They will stay in effect until GTS is released during startup, after which the UnusedPin bitstream generator option will determine whether the unused I/Os have a pull-up, pull-down, or no resistor.



Symbol	Description	Min	Max
T_{POR}	Power-on reset	-	2 ms
T_{PL}	Program latency	-	100 µs
T_{ICCK}	CCLK output delay (Master Serial mode only)	0.5 µs	4 µs
$T_{PROGRAM}$	Program pulse width	300 ns	-

Notes: (referring to waveform above:)

1. Before configuration can begin, V_{CCINT} must be greater than 1.6V and V_{CCO} Bank 2 must be greater than 1.0V.

Figure 12: **Configuration Timing on Power-Up**

Clearing Configuration Memory

The device indicates that clearing the configuration memory is in progress by driving INIT Low. At this time, the user can delay configuration by holding either PROGRAM or INIT Low, which causes the device to remain in the memory clearing phase. Note that the bidirectional INIT line is driving a Low logic level during memory clearing. To avoid contention, use an open-drain driver to keep INIT Low.

With no delay in force, the device indicates that the memory is completely clear by driving INIT High. The FPGA samples its mode pins on this Low-to-High transition.

Loading Configuration Data

Once INIT is High, the user can begin loading configuration data frames into the device. The details of loading the configuration data are discussed in the sections treating the configuration modes individually. The sequence of operations necessary to load configuration data using the serial modes is shown in Figure 14. Loading data using the Slave Parallel mode is shown in Figure 19, page 25.

CRC Error Checking

During the loading of configuration data, a CRC value embedded in the configuration file is checked against a CRC value calculated within the FPGA. If the CRC values do not match, the FPGA drives INIT Low to indicate that a frame error has occurred and configuration is aborted.

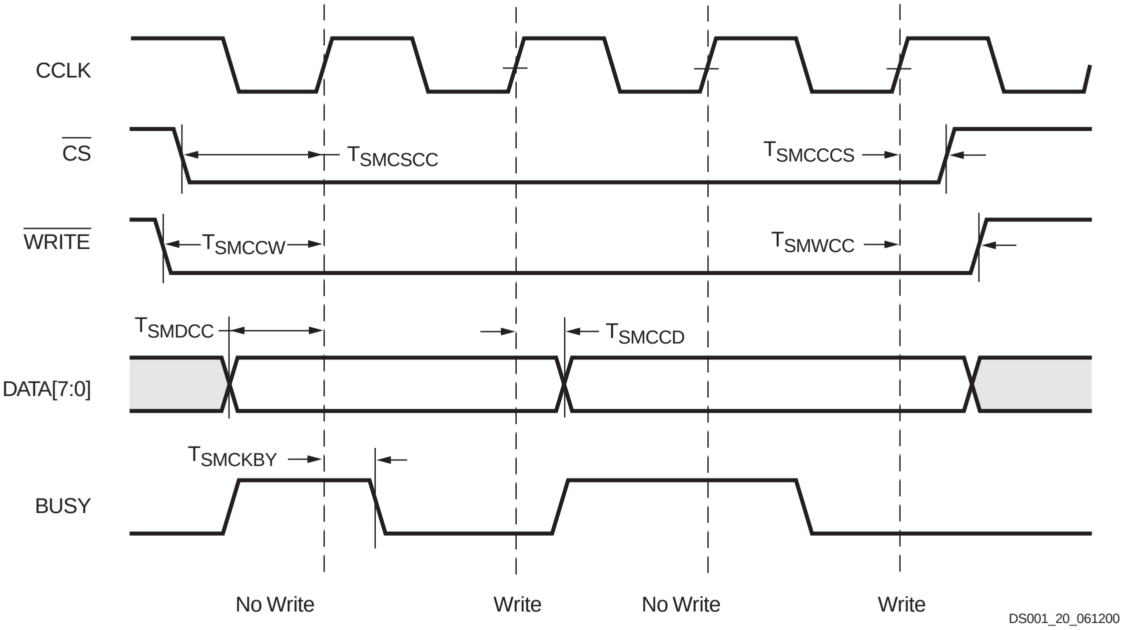
To reconfigure the device, the PROGRAM pin should be asserted to reset the configuration logic. Recycling power also resets the FPGA for configuration. See "Clearing Configuration Memory".

Start-up

The start-up sequence oversees the transition of the FPGA from the configuration state to full user operation. A match of CRC values, indicating a successful loading of the configuration data, initiates the sequence.

During start-up, the device performs four operations:

1. The assertion of DONE. The failure of DONE to go High may indicate the unsuccessful loading of configuration data.
2. The release of the Global Three State net. This activates I/Os to which signals are assigned. The remaining I/Os stay in a high-impedance state with internal weak pull-down resistors present.
3. Negates Global Set Reset (GSR). This allows all flip-flops to change state.
4. The assertion of Global Write Enable (GWE). This allows all RAMs and flip-flops to change state.



Symbol		Description		Units
T_{SMDCC}	CCLK	D0-D7 setup/hold	5	ns, min
T_{SMCCD}		D0-D7 hold	0	ns, min
T_{SMCSCC}		$\overline{\text{CS}}$ setup	7	ns, min
T_{SMCCCS}		$\overline{\text{CS}}$ hold	0	ns, min
T_{SMCCW}		$\overline{\text{WRITE}}$ setup	7	ns, min
T_{SMWCC}		$\overline{\text{WRITE}}$ hold	0	ns, min
T_{SMCKBY}		BUSY propagation delay	12	ns, max
F_{CC}		Maximum frequency	66	MHz, max
F_{CCNH}		Maximum frequency with no handshake	50	MHz, max

Figure 20: **Slave Parallel Write Timing**

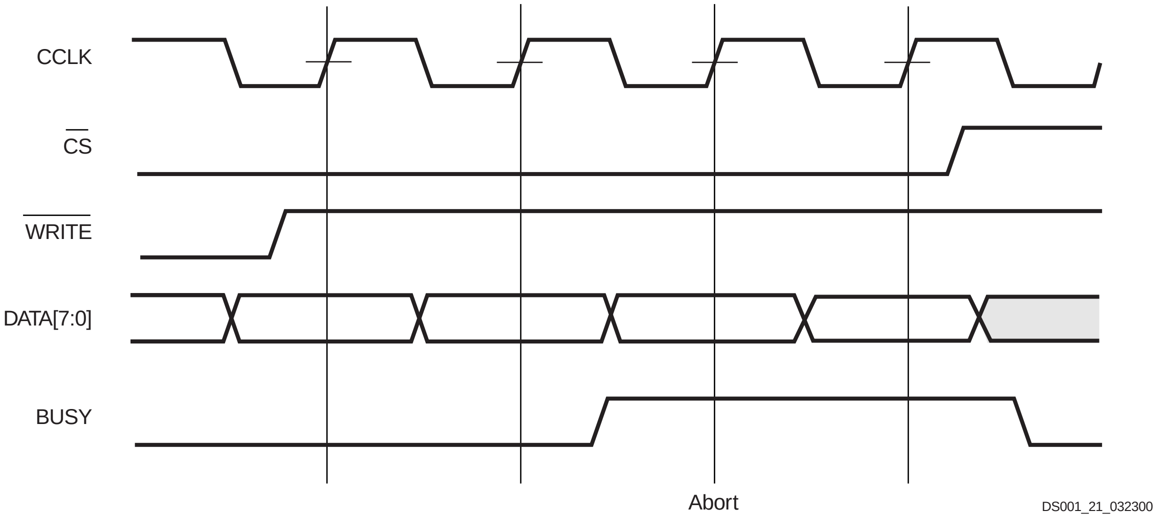


Figure 21: **Slave Parallel Write Abort Waveforms**

Design Considerations

This section contains more detailed design information on the following features:

- Delay-Locked Loop . . . see page 27
- Block RAM . . . see page 32
- Versatile I/O . . . see page 36

Using Delay-Locked Loops

The Spartan-II FPGA family provides up to four fully digital dedicated on-chip Delay-Locked Loop (DLL) circuits which provide zero propagation delay, low clock skew between output clock signals distributed throughout the device, and advanced clock domain control. These dedicated DLLs can be used to implement several circuits that improve and simplify system level design.

Introduction

Quality on-chip clock distribution is important. Clock skew and clock delay impact device performance and the task of managing clock skew and clock delay with conventional clock trees becomes more difficult in large devices. The Spartan-II family of devices resolve this potential problem by providing up to four fully digital dedicated on-chip Delay-Locked Loop (DLL) circuits which provide zero propagation delay and low clock skew between output clock signals distributed throughout the device.

Each DLL can drive up to two global clock routing networks within the device. The global clock distribution network minimizes clock skews due to loading differences. By monitoring a sample of the DLL output clock, the DLL can compensate for the delay on the routing network, effectively eliminating the delay from the external input port to the individual clock loads within the device.

In addition to providing zero delay with respect to a user source clock, the DLL can provide multiple phases of the source clock. The DLL can also act as a clock doubler or it can divide the user source clock by up to 16.

Clock multiplication gives the designer a number of design alternatives. For instance, a 50 MHz source clock doubled by the DLL can drive an FPGA design operating at 100 MHz. This technique can simplify board design because the clock path on the board no longer distributes such a high-speed signal. A multiplied clock also provides designers the option of time-domain-multiplexing, using one circuit twice per clock cycle, consuming less area than two copies of the same circuit.

The DLL can also act as a clock mirror. By driving the DLL output off-chip and then back in again, the DLL can be used to de-skew a board level clock between multiple devices.

In order to guarantee the system clock establishes prior to the device "waking up," the DLL can delay the completion of the device configuration process until after the DLL achieves lock.

By taking advantage of the DLL to remove on-chip clock delay, the designer can greatly simplify and improve system level design involving high-fanout, high-performance clocks.

Library DLL Primitives

Figure 22 shows the simplified Xilinx library DLL macro, BUFGDLL. This macro delivers a quick and efficient way to provide a system clock with zero propagation delay throughout the device. Figure 23 and Figure 24 show the two library DLL primitives. These primitives provide access to the complete set of DLL features when implementing more complex applications.

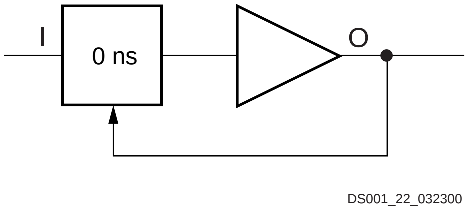


Figure 22: **Simplified DLL Macro BUFGDLL**

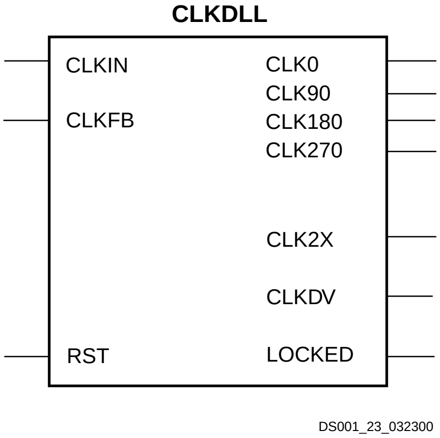


Figure 23: **Standard DLL Primitive CLKDLL**

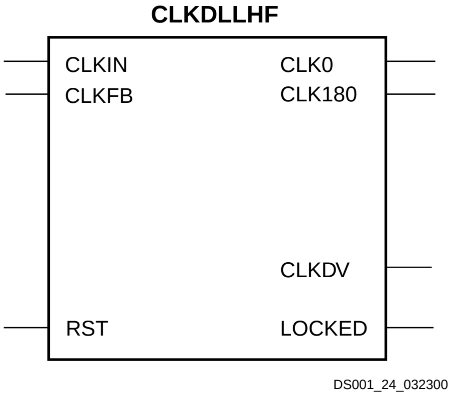


Figure 24: **High-Frequency DLL Primitive CLKDLLHF**

Startup Delay Property

This property, STARTUP_WAIT, takes on a value of TRUE or FALSE (the default value). When TRUE the Startup Sequence following device configuration is paused at a user-specified point until the DLL locks. **XAPP176**: *Configuration and Readback of the Spartan-II and Spartan-IIE Families* explains how this can result in delaying the assertion of the DONE pin until the DLL locks.

DLL Location Constraints

The DLLs are distributed such that there is one DLL in each corner of the device. The location constraint LOC, attached to the DLL primitive with the numeric identifier 0, 1, 2, or 3, controls DLL location. The orientation of the four DLLs and their corresponding clock resources appears in Figure 27.

The LOC property uses the following form.

LOC = DLL2

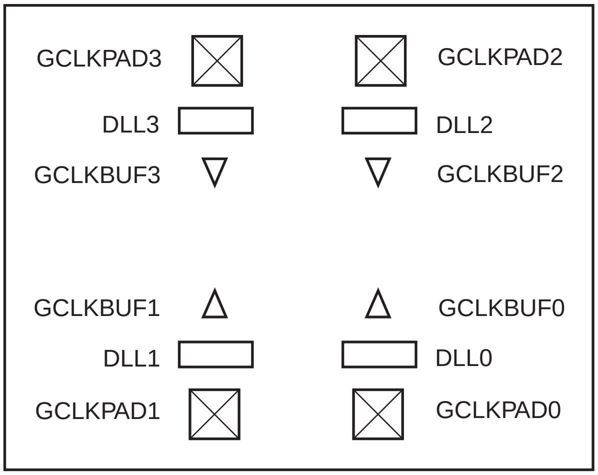


Figure 27: **Orientation of DLLs**

Design Considerations

Use the following design considerations to avoid pitfalls and improve success designing with Xilinx devices.

Input Clock

The output clock signal of a DLL, essentially a delayed version of the input clock signal, reflects any instability on the input clock in the output waveform. For this reason the quality of the DLL input clock relates directly to the quality of the output clock waveforms generated by the DLL. The DLL input clock requirements are specified in the "DLL Timing Parameters" section of the data sheet.

In most systems a crystal oscillator generates the system clock. The DLL can be used with any commercially available quartz crystal oscillator. For example, most crystal oscillators produce an output waveform with a frequency tolerance of 100 PPM, meaning 0.01 percent change in the clock period. The DLL operates reliably on an input waveform with a frequency drift of up to 1 ns — orders of magnitude in excess of that needed to support any crystal oscillator in the industry. However, the cycle-to-cycle jitter must be kept to less than 300 ps in the low frequencies and 150 ps for the high frequencies.

Input Clock Changes

Changing the period of the input clock beyond the maximum drift amount requires a manual reset of the CLKDLL. Failure to reset the DLL will produce an unreliable lock signal and output clock.

It is possible to stop the input clock in a way that has little impact to the DLL. Stopping the clock should be limited to less than approximately 100 μs to keep device cooling to a minimum and maintain the validity of the current tap setting. The clock should be stopped during a Low phase, and when restored the full High period should be seen. During this time LOCKED will stay High and remain High when the clock is restored. If these conditions may not be met in the design, apply a manual reset to the DLL after re-starting the input clock, even if the LOCKED signal has not changed.

When the clock is stopped, one to four more clocks will still be observed as the delay line is flushed. When the clock is restarted, the output clocks will not be observed for one to four clocks as the delay line is filled. The most common case will be two or three clocks.

In a similar manner, a phase shift of the input clock is also possible. The phase shift will propagate to the output one to four clocks after the original shift, with no disruption to the CLKDLL control.

Output Clocks

As mentioned earlier in the DLL pin descriptions, some restrictions apply regarding the connectivity of the output pins. The DLL clock outputs can drive an OBUF, a global clock buffer BUFG, or route directly to destination clock pins. The only BUFGs that the DLL clock outputs can drive are the two on the same edge of the device (top or bottom). One DLL output can drive more than one OBUF; however, this adds skew.

Do not use the DLL output clock signals until after activation of the LOCKED signal. Prior to the activation of the LOCKED signal, the DLL output clocks are not valid and can exhibit glitches, spikes, or other spurious movement.

Creating Larger RAM Structures

The block RAM columns have specialized routing to allow cascading blocks together with minimal routing delays. This achieves wider or deeper RAM structures with a smaller timing penalty than when using normal routing channels.

Location Constraints

Block RAM instances can have LOC properties attached to them to constrain the placement. The block RAM placement locations are separate from the CLB location naming convention, allowing the LOC properties to transfer easily from array to array.

The LOC properties use the following form:

LOC = RAMB4_R#C#

RAMB4_R0C0 is the upper left RAMB4 location on the device.

Conflict Resolution

The block RAM memory is a true dual-read/write port RAM that allows simultaneous access of the same memory cell from both ports. When one port writes to a given memory cell, the other port must not address that memory cell (for a write or a read) within the clock-to-clock setup window. The following lists specifics of port and memory cell write conflict resolution.

- If both ports write to the same memory cell simultaneously, violating the clock-to-clock setup requirement, consider the data stored as invalid.
- If one port attempts a read of the same memory cell the other simultaneously writes, violating the clock-to-clock setup requirement, the following occurs.
  - The write succeeds
  - The data out on the writing port accurately reflects the data written.
  - The data out on the reading port is invalid.

Conflicts do not cause any physical damage.

Single Port Timing

Figure 33 shows a timing diagram for a single port of a block RAM memory. The block RAM AC switching characteristics are specified in the data sheet. The block RAM memory is initially disabled.

At the first rising edge of the CLK pin, the ADDR, DI, EN, WE, and RST pins are sampled. The EN pin is High and the WE pin is Low indicating a read operation. The DO bus contains the contents of the memory location, 0x00, as indicated by the ADDR bus.

At the second rising edge of the CLK pin, the ADDR, DI, EN, WR, and RST pins are sampled again. The EN and WE pins are High indicating a write operation. The DO bus mirrors the DI bus. The DI bus is written to the memory location 0x0F.

At the third rising edge of the CLK pin, the ADDR, DI, EN, WR, and RST pins are sampled again. The EN pin is High and the WE pin is Low indicating a read operation. The DO bus contains the contents of the memory location 0x7E as indicated by the ADDR bus.

At the fourth rising edge of the CLK pin, the ADDR, DI, EN, WR, and RST pins are sampled again. The EN pin is Low indicating that the block RAM memory is now disabled. The DO bus retains the last value.

Dual Port Timing

Figure 34 shows a timing diagram for a true dual-port read/write block RAM memory. The clock on port A has a longer period than the clock on Port B. The timing parameter T_{BCCS}, (clock-to-clock setup) is shown on this diagram. The parameter, T_{BCCS} is violated once in the diagram. All other timing parameters are identical to the single port version shown in Figure 33.

T_{BCCS} is only of importance when the address of both ports are the same and at least one port is performing a write operation. When the clock-to-clock set-up parameter is violated for a WRITE-WRITE condition, the contents of the memory at that location will be invalid. When the clock-to-clock set-up parameter is violated for a WRITE-READ condition, the contents of the memory will be correct, but the read port will have invalid data. At the first rising edge of the CLKA, memory location 0x00 is to be written with the value 0xAAAA and is mirrored on the DOA bus. The last operation of Port B was a read to the same memory location 0x00. The DOB bus of Port B does not change with the new value on Port A, and retains the last read value. A short time later, Port B executes another read to memory location 0x00, and the DOB bus now reflects the new memory value written by Port A.

At the second rising edge of CLKA, memory location 0x7E is written with the value 0x9999 and is mirrored on the DOA bus. Port B then executes a read operation to the same memory location without violating the T_{BCCS} parameter and the DOB reflects the new memory values written by Port A.

Recommended Operating Conditions

Symbol	Description		Min	Max	Units
T_J	Junction temperature[1]	Commercial	0	85	°C
		Industrial	–40	100	°C
V_{CCINT}	Supply voltage relative to GND[2,5]	Commercial	2.5 – 5%	2.5 + 5%	V
		Industrial	2.5 – 5%	2.5 + 5%	V
V_{CCO}	Supply voltage relative to GND[3,5]	Commercial	1.4	3.6	V
		Industrial	1.4	3.6	V
T_{IN}	Input signal transition time[4]		-	250	ns

Notes:

1. At junction temperatures above those listed as Operating Conditions, all delay parameters increase by 0.35% per °C.
2. Functional operation is guaranteed down to a minimum V_{CCINT} of 2.25V (Nominal V_{CCINT} – 10%). For every 50 mV reduction in V_{CCINT} below 2.375V (nominal V_{CCINT} – 5%), all delay parameters increase by 3%.
3. Minimum and maximum values for V_{CCO} vary according to the I/O standard selected.
4. Input and output measurement threshold is ~50% of V_{CCO}. See "Delay Measurement Methodology," page 60 for specific levels.
5. Supply voltages may be applied in any order desired.

DC Characteristics Over Operating Conditions

Symbol	Description			Min	Typ	Max	Units
V_{DRINT}	Data Retention V_{CCINT} voltage (below which configuration data may be lost)			2.0	-	-	V
V_{DRIO}	Data Retention V_{CCO} voltage (below which configuration data may be lost)			1.2	-	-	V
I_{CCINTQ}	Quiescent V_{CCINT} supply current[1]	XC2S15	Commercial	-	10	30	mA
			Industrial	-	10	60	mA
		XC2S30	Commercial	-	10	30	mA
			Industrial	-	10	60	mA
		XC2S50	Commercial	-	12	50	mA
			Industrial	-	12	100	mA
		XC2S100	Commercial	-	12	50	mA
			Industrial	-	12	100	mA
		XC2S150	Commercial	-	15	50	mA
			Industrial	-	15	100	mA
		XC2S200	Commercial	-	15	75	mA
			Industrial	-	15	150	mA
I_{CCOQ}	Quiescent V_{CCO} supply current[1]			-	-	2	mA
I_{REF}	V_{REF} current per V_{REF} pin			-	-	20	µA
I_L	Input or output leakage current[2]			–10	-	+10	µA
C_{IN}	Input capacitance (sample tested)	VQ, CS, TQ, PQ, FG packages		-	-	8	pF
I_{RPU}	Pad pull-up (when selected) @ V_{IN} = 0V, V_{CCO} = 3.3V (sample tested)[3]			-	-	0.25	mA
I_{RPD}	Pad pull-down (when selected) @ V_{IN} = 3.6V (sample tested)[3]			-	-	0.15	mA

Notes:

1. With no output current loads, no active input pull-up resistors, all I/O pins 3-stated and floating.
2. The I/O leakage current specification applies only when the V_{CCINT} and V_{CCO} supply voltages have reached their respective minimum Recommended Operating Conditions.
3. Internal pull-up and pull-down resistors guarantee valid logic levels at unconnected input pins. These pull-up and pull-down resistors do not provide valid logic levels when input pins are connected to other circuits.

Global Clock Setup and Hold for LVTTL Standard, *with* DLL (Pin-to-Pin)

Symbol	Description	Device	Speed Grade -6 Min	Speed Grade -5 Min	Units
T_{PSDLL} / T_{PHDLL}	Input setup and hold time relative to global clock input signal for LVTTL standard, no delay, IFF,(1) with DLL	All	1.7 / 0	1.9 / 0	ns

Notes:

1. IFF = Input Flip-Flop or Latch
2. Setup time is measured relative to the Global Clock input signal with the fastest route and the lightest load. Hold time is measured relative to the Global Clock input signal with the slowest route and heaviest load.
3. DLL output jitter is already included in the timing calculation.
4. A zero hold time listing indicates no hold time or a negative hold time.
5. For data input with different standards, adjust the setup time delay by the values shown in "IOB Input Delay Adjustments for Different Standards," page 57. For a global clock input with standards other than LVTTL, adjust delays with values from the "I/O Standard Global Clock Input Adjustments," page 61.

Global Clock Setup and Hold for LVTTL Standard, *without* DLL (Pin-to-Pin)

Symbol	Description	Device	Speed Grade -6 Min	Speed Grade -5 Min	Units
T_{PSFD} / T_{PHFD}	Input setup and hold time relative to global clock input signal for LVTTL standard, no delay, IFF,(1) without DLL	XC2S15	2.2 / 0	2.7 / 0	ns
		XC2S30	2.2 / 0	2.7 / 0	ns
		XC2S50	2.2 / 0	2.7 / 0	ns
		XC2S100	2.3 / 0	2.8 / 0	ns
		XC2S150	2.4 / 0	2.9 / 0	ns
		XC2S200	2.4 / 0	3.0 / 0	ns

Notes:

1. IFF = Input Flip-Flop or Latch
2. Setup time is measured relative to the Global Clock input signal with the fastest route and the lightest load. Hold time is measured relative to the Global Clock input signal with the slowest route and heaviest load.
3. A zero hold time listing indicates no hold time or a negative hold time.
4. For data input with different standards, adjust the setup time delay by the values shown in "IOB Input Delay Adjustments for Different Standards," page 57. For a global clock input with standards other than LVTTL, adjust delays with values from the "I/O Standard Global Clock Input Adjustments," page 61.

IOB Input Switching Characteristics(1)

Input delays associated with the pad are specified for LVTTL levels. For other standards, adjust the delays with the values shown in "IOB Input Delay Adjustments for Different Standards," page 57.

Symbol	Description	Device	Speed Grade				Units
			-6		-5		
			Min	Max	Min	Max	
Propagation Delays							
T_{IOPI}	Pad to I output, no delay	All	-	0.8	-	1.0	ns
T_{IOPID}	Pad to I output, with delay	All	-	1.5	-	1.8	ns
T_{IOPLI}	Pad to output IQ via transparent latch, no delay	All	-	1.7	-	2.0	ns
T_{IOPLID}	Pad to output IQ via transparent latch, with delay	XC2S15	-	3.8	-	4.5	ns
		XC2S30	-	3.8	-	4.5	ns
		XC2S50	-	3.8	-	4.5	ns
		XC2S100	-	3.8	-	4.5	ns
		XC2S150	-	4.0	-	4.7	ns
		XC2S200	-	4.0	-	4.7	ns
Sequential Delays							
T_{IOCKIQ}	Clock CLK to output IQ	All	-	0.7	-	0.8	ns
Setup/Hold Times with Respect to Clock CLK(2)							
T_{IOPICK} / T_{IOICKP}	Pad, no delay	All	1.7 / 0	-	1.9 / 0	-	ns
$T_{IOPICKD}$ / $T_{IOICKPD}$	Pad, with delay(1)	XC2S15	3.8 / 0	-	4.4 / 0	-	ns
		XC2S30	3.8 / 0	-	4.4 / 0	-	ns
		XC2S50	3.8 / 0	-	4.4 / 0	-	ns
		XC2S100	3.8 / 0	-	4.4 / 0	-	ns
		XC2S150	3.9 / 0	-	4.6 / 0	-	ns
		XC2S200	3.9 / 0	-	4.6 / 0	-	ns
$T_{IOICECK}$ / $T_{IOCKICE}$	ICE input	All	0.9 / 0.01	-	0.9 / 0.01	-	ns
Set/Reset Delays							
$T_{IOSRCKI}$	SR input (IFF, synchronous)	All	-	1.1	-	1.2	ns
T_{IOSRIQ}	SR input to IQ (asynchronous)	All	-	1.5	-	1.7	ns
T_{GSRQ}	GSR to output IQ	All	-	9.9	-	11.7	ns

Notes:

1. Input timing for LVTTL is measured at 1.4V. For other I/O standards, see the table "Delay Measurement Methodology," page 60.
2. A zero hold time listing indicates no hold time or a negative hold time.

DLL Timing Parameters

All devices are 100 percent functionally tested. Because of the difficulty in directly measuring many internal timing parameters, those parameters are derived from benchmark timing patterns. The following guidelines reflect worst-case values across the recommended operating conditions.

Symbol	Description	Speed Grade				Units
		-6		-5		
		Min	Max	Min	Max	
$F_{CLKINHF}$	Input clock frequency (CLKDLLHF)	60	200	60	180	MHz
$F_{CLKINLF}$	Input clock frequency (CLKDLL)	25	100	25	90	MHz
$T_{DLLPWHF}$	Input clock pulse width (CLKDLLHF)	2.0	-	2.4	-	ns
$T_{DLLPWLF}$	Input clock pulse width (CLKDLL)	2.5	-	3.0	-	ns

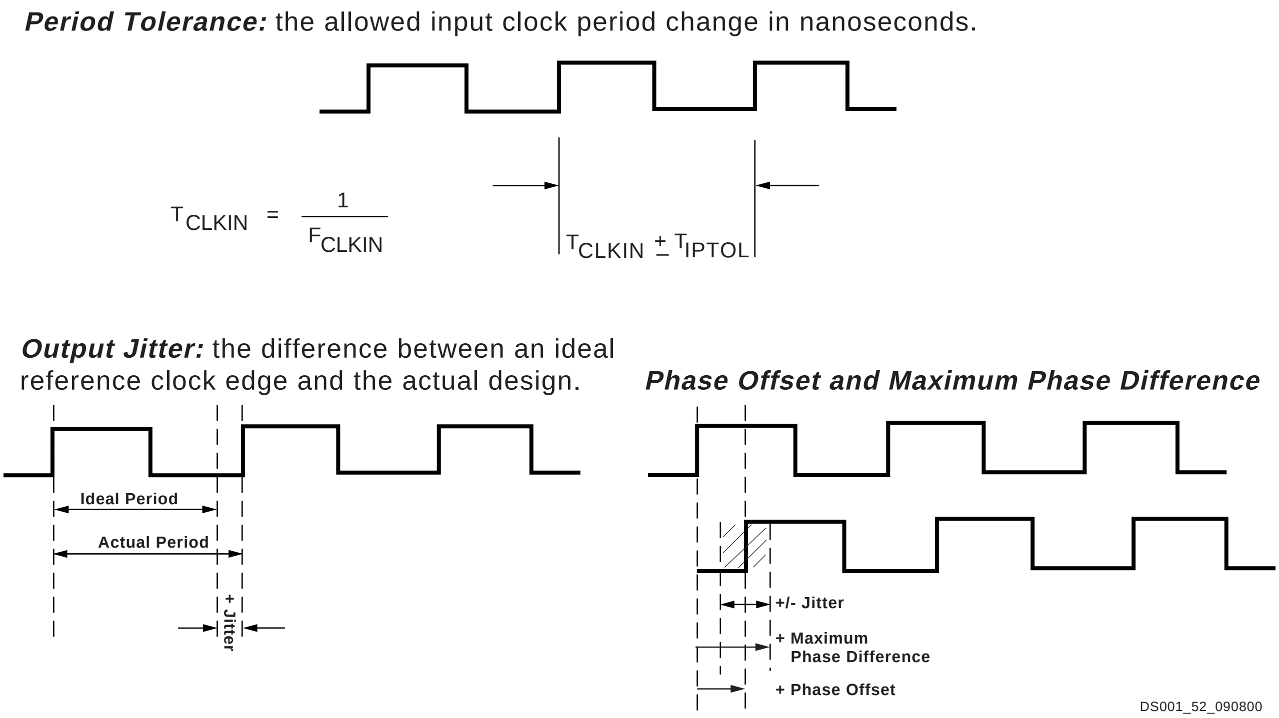
DLL Clock Tolerance, Jitter, and Phase Information

All DLL output jitter and phase specifications were determined through statistical measurement at the package pins using a clock mirror configuration and matched drivers. Figure 52, page 63, provides definitions for various parameters in the table below.

Symbol	Description	F_{CLKIN}	CLKDLLHF		CLKDLL		Units
			Min	Max	Min	Max	
T_{IPTOL}	Input clock period tolerance		-	1.0	-	1.0	ns
T_{IJITCC}	Input clock jitter tolerance (cycle-to-cycle)		-	±150	-	±300	ps
T_{LOCK}	Time required for DLL to acquire lock	> 60 MHz	-	20	-	20	µs
		50-60 MHz	-	-	-	25	µs
		40-50 MHz	-	-	-	50	µs
		30-40 MHz	-	-	-	90	µs
		25-30 MHz	-	-	-	120	µs
T_{OJITCC}	Output jitter (cycle-to-cycle) for any DLL clock output(1)		-	±60	-	±60	ps
T_{PHIO}	Phase offset between CLKIN and CLKO(2)		-	±100	-	±100	ps
T_{PHOO}	Phase offset between clock outputs on the DLL(3)		-	±140	-	±140	ps
T_{PHIOM}	Maximum phase difference between CLKIN and CLKO(4)		-	±160	-	±160	ps
T_{PHOOM}	Maximum phase difference between clock outputs on the DLL(5)		-	±200	-	±200	ps

Notes:

1. **Output Jitter** is cycle-to-cycle jitter measured on the DLL output clock, *excluding* input clock jitter.
2. **Phase Offset between CLKIN and CLKO** is the worst-case fixed time difference between rising edges of CLKIN and CLKO, *excluding* output jitter and input clock jitter.
3. **Phase Offset between Clock Outputs on the DLL** is the worst-case fixed time difference between rising edges of any two DLL outputs, *excluding* Output Jitter and input clock jitter.
4. **Maximum Phase Difference between CLKIN an CLKO** is the sum of Output Jitter and Phase Offset between CLKIN and CLKO, or the greatest difference between CLKIN and CLKO rising edges due to DLL alone (*excluding* input clock jitter).
5. **Maximum Phase Difference between Clock Outputs on the DLL** is the sum of Output JItter and Phase Offset between any DLL clock outputs, or the greatest difference between any two DLL output rising edges due to DLL alone (*excluding* input clock jitter).



Period Tolerance: the allowed input clock period change in nanoseconds.

Output Jitter: the difference between an ideal reference clock edge and the actual design.

Phase Offset and Maximum Phase Difference

Figure 52: **Period Tolerance and Clock Jitter**

Revision History

Date	Version No.	Description
09/18/00	2.0	Sectioned the Spartan-II Family data sheet into four modules. Updated timing to reflect the latest speed files. Added current supply numbers and XC2S200 -5 timing numbers. Approved -5 timing numbers as preliminary information with exceptions as noted.
11/02/00	2.1	Removed Power Down feature.
01/19/01	2.2	DC and timing numbers updated to Preliminary for the XC2S50 and XC2S100. Industrial power-on current specifications and -6 DLL timing numbers added. Power-on specification clarified.
03/09/01	2.3	Added note on power sequencing. Clarified power-on current requirement.
08/28/01	2.4	Added -6 preliminary timing. Added typical and industrial standby current numbers. Specified min. power-on current by junction temperature instead of by device type (Commercial vs. Industrial). Eliminated minimum V_{CCINT} ramp time requirement. Removed footnote limiting DLL operation to the Commercial temperature range.
07/26/02	2.5	Clarified that I/O leakage current is specified over the Recommended Operating Conditions for V_{CCINT} and V_{CCO}.
08/26/02	2.6	Added references for XAPP450 to Power-On Current Specification.
09/03/03	2.7	Added relaxed minimum power-on current (I_{CCPO}) requirements to page 53. On page 64, moved T_{RPW} values from maximum to minimum column.
06/13/08	2.8	Updated I/O measurement thresholds. Updated description and links. Updated all modules for continuous page, figure, and table numbering. Synchronized all modules to v2.8.

XC2S50 Device Pinouts *(Continued)*

XC2S50 Pad Name		TQ144	PQ208	FG256	Bndry Scan
Function	**Bank**				
I/O	0	-	-	D8	83
I/O	0	-	P188	A6	86
I/O, V_{REF}	0	P12	P189	B7	89
GND	-	-	P190	GND*	-
I/O	0	-	P191	C8	92
I/O	0	-	P192	D7	95
I/O	0	-	P193	E7	98
I/O	0	P11	P194	C7	104
I/O	0	P10	P195	B6	107
V_{CCINT}	-	P9	P196	V_{CCINT}*	-
V_{CCO}	0	-	P197	V_{CCO} Bank 0*	-
GND	-	P8	P198	GND*	-
I/O	0	P7	P199	A5	110
I/O	0	P6	P200	C6	113
I/O	0	-	P201	B5	116
I/O	0	-	-	D6	119
I/O	0	-	P202	A4	122
I/O, V_{REF}	0	P5	P203	B4	125
GND	-	-	-	GND*	-
I/O	0	-	P204	E6	128
I/O	0	-	-	D5	131
I/O	0	P4	P205	A3	134
I/O	0	-	-	C5	137
I/O	0	P3	P206	B3	140
TCK	-	P2	P207	C4	-
V_{CCO}	0	P1	P208	V_{CCO} Bank 0*	-
V_{CCO}	7	P144	P208	V_{CCO} Bank 7*	-

04/18/01

Notes:

1. IRDY and TRDY can only be accessed when using Xilinx PCI cores.
2. Pads labelled GND*, V_{CCINT}*, V_{CCO} Bank 0*, V_{CCO} Bank 1*, V_{CCO} Bank 2*, V_{CCO} Bank 3*, V_{CCO} Bank 4*, V_{CCO} Bank 5*, V_{CCO} Bank 6*, V_{CCO} Bank 7* are internally bonded to independent ground or power planes within the package.
3. See "VCCO Banks" for details on V_{CCO} banking.

Additional XC2S50 Package Pins

TQ144

Not Connected Pins					
P104	P105	-	-	-	-

11/02/00

XC2S150 Device Pinouts

XC2S150 Pad Name		PQ208	FG256	FG456	Bndry Scan
Function	Bank				
GND	-	P1	GND*	GND*	-
TMS	-	P2	D3	D3	-
I/O	7	P3	C2	B1	221
I/O	7	-	-	E4	224
I/O	7	-	-	C1	227
I/O	7	-	A2	F5	230
GND	-	-	GND*	GND*	-
I/O	7	P4	B1	D2	233
I/O	7	-	-	E3	236
I/O	7	-	-	F4	239
I/O	7	-	E3	G5	242
I/O	7	P5	D2	F3	245
GND	-	-	GND*	GND*	-
V_{CCO}	7	-	V_{CCO} Bank 7*	V_{CCO} Bank 7*	-
I/O, V_{REF}	7	P6	C1	E2	248
I/O	7	P7	F3	E1	251
I/O	7	-	-	G4	254
I/O	7	-	-	G3	257
I/O	7	-	E2	H5	260
I/O	7	P8	E4	F2	263
I/O	7	-	-	F1	266
I/O, V_{REF}	7	P9	D1	H4	269
I/O	7	P10	E1	G1	272
GND	-	P11	GND*	GND*	-
V_{CCO}	7	P12	V_{CCO} Bank 7*	V_{CCO} Bank 7*	-
V_{CCINT}	-	P13	V_{CCINT}*	V_{CCINT}*	-
I/O	7	P14	F2	H3	275
I/O	7	P15	G3	H2	278
I/O	7	-	-	H1	284
I/O	7	-	F1	J5	287
I/O	7	P16	F4	J2	290
I/O	7	-	-	J3	293
I/O	7	P17	F5	K5	299
I/O	7	P18	G2	K1	302
GND	-	P19	GND*	GND*	-
V_{CCO}	7	-	V_{CCO} Bank 7*	V_{CCO} Bank 7*	-
I/O, V_{REF}	7	P20	H3	K3	305
I/O	7	P21	G4	K4	308
I/O	7	-	H2	L6	311

XC2S150 Device Pinouts *(Continued)*

XC2S150 Pad Name		PQ208	FG256	FG456	Bndry Scan
Function	Bank				
I/O	7	P22	G5	L1	314
I/O	7	-	-	L5	317
I/O	7	P23	H4	L4	320
I/O, IRDY(1)	7	P24	G1	L3	323
GND	-	P25	GND*	GND*	-
V_{CCO}	7	P26	V_{CCO} Bank 7*	V_{CCO} Bank 7*	-
V_{CCO}	6	P26	V_{CCO} Bank 6*	V_{CCO} Bank 6*	-
I/O, TRDY(1)	6	P27	J2	M1	326
V_{CCINT}	-	P28	V_{CCINT}*	V_{CCINT}*	-
I/O	6	-	-	M6	332
I/O	6	P29	H1	M3	335
I/O	6	-	J4	M4	338
I/O	6	P30	J1	M5	341
I/O, V_{REF}	6	P31	J3	N2	344
V_{CCO}	6	-	V_{CCO} Bank 6*	V_{CCO} Bank 6*	-
GND	-	P32	GND*	GND*	-
I/O	6	P33	K5	N3	347
I/O	6	P34	K2	N4	350
I/O	6	-	-	N5	356
I/O	6	P35	K1	P2	359
I/O	6	-	K3	P4	362
I/O	6	-	-	R1	365
I/O	6	P36	L1	P3	371
I/O	6	P37	L2	R2	374
V_{CCINT}	-	P38	V_{CCINT}*	V_{CCINT}*	-
V_{CCO}	6	P39	V_{CCO} Bank 6*	V_{CCO} Bank 6*	-
GND	-	P40	GND*	GND*	-
I/O	6	P41	K4	T1	377
I/O, V_{REF}	6	P42	M1	R4	380
I/O	6	-	-	T2	383
I/O	6	P43	L4	U1	386
I/O	6	-	M2	R5	389
I/O	6	-	-	V1	392
I/O	6	-	-	T5	395
I/O	6	P44	L3	U2	398
I/O, V_{REF}	6	P45	N1	T3	401
V_{CCO}	6	-	V_{CCO} Bank 6*	V_{CCO} Bank 6*	-
GND	-	-	GND*	GND*	-

XC2S150 Device Pinouts *(Continued)*

XC2S150 Pad Name					Bndry
Function	**Bank**	**PQ208**	**FG256**	**FG456**	**Scan**
I/O	4	P90	R11	AA15	595
V_{CCINT}	-	P91	V_{CCINT}*	V_{CCINT}*	-
V_{CCO}	4	P92	V_{CCO} Bank 4*	V_{CCO} Bank 4*	-
GND	-	P93	GND*	GND*	-
I/O	4	P94	M11	Y15	598
I/O, V_{REF}	4	P95	T11	AB16	601
I/O	4	-	-	AB17	604
I/O	4	P96	N11	V15	607
I/O	4	-	R12	Y16	610
I/O	4	-	-	AA17	613
I/O	4	-	-	W16	616
I/O	4	P97	P11	AB18	619
I/O, V_{REF}	4	P98	T12	AB19	622
V_{CCO}	4	-	V_{CCO} Bank 4*	V_{CCO} Bank 4*	-
GND	-	-	GND*	GND*	-
I/O	4	P99	T13	Y17	625
I/O	4	-	N12	V16	628
I/O	4	-	-	AA18	631
I/O	4	-	-	W17	634
I/O	4	P100	R13	AB20	637
GND	-	-	GND*	GND*	-
I/O	4	-	P12	AA19	640
I/O	4	-	-	V17	643
I/O	4	-	-	Y18	646
I/O	4	P101	P13	AA20	649
I/O	4	P102	T14	W18	652
GND	-	P103	GND*	GND*	-
DONE	3	P104	R14	Y19	655
V_{CCO}	4	P105	V_{CCO} Bank 4*	V_{CCO} Bank 4*	-
V_{CCO}	3	P105	V_{CCO} Bank 3*	V_{CCO} Bank 3*	-
$\overline{\text{PROGRAM}}$	-	P106	P15	W20	658
I/O ($\overline{\text{INIT}}$)	3	P107	N15	V19	659
I/O (D7)	3	P108	N14	Y21	662
I/O	3	-	-	V20	665
I/O	3	-	-	AA22	668
I/O	3	-	T15	W21	671
GND	-	-	GND*	GND*	-
I/O	3	P109	M13	U20	674

XC2S150 Device Pinouts *(Continued)*

XC2S150 Pad Name					Bndry
Function	**Bank**	**PQ208**	**FG256**	**FG456**	**Scan**
I/O	3	-	-	U19	677
I/O	3	-	-	V21	680
I/O	3	-	R16	T18	683
I/O	3	P110	M14	W22	686
GND	-	-	GND*	GND*	-
V_{CCO}	3	-	V_{CCO} Bank 3*	V_{CCO} Bank 3*	-
I/O, V_{REF}	3	P111	L14	U21	689
I/O	3	P112	M15	T20	692
I/O	3	-	-	T19	695
I/O	3	-	-	V22	698
I/O	3	-	L12	T21	701
I/O	3	P113	P16	R18	704
I/O	3	-	-	U22	707
I/O, V_{REF}	3	P114	L13	R19	710
I/O (D6)	3	P115	N16	T22	713
GND	-	P116	GND*	GND*	-
V_{CCO}	3	P117	V_{CCO} Bank 3*	V_{CCO} Bank 3*	-
V_{CCINT}	-	P118	V_{CCINT}*	V_{CCINT}*	-
I/O (D5)	3	P119	M16	R21	716
I/O	3	P120	K14	P18	719
I/O	3	-	-	P19	725
I/O	3	-	L16	P20	728
I/O	3	P121	K13	P21	731
I/O	3	-	-	N19	734
I/O	3	P122	L15	N18	740
I/O	3	P123	K12	N20	743
GND	-	P124	GND*	GND*	-
V_{CCO}	3	-	V_{CCO} Bank 3*	V_{CCO} Bank 3*	-
I/O, V_{REF}	3	P125	K16	N21	746
I/O (D4)	3	P126	J16	N22	749
I/O	3	-	J14	M19	752
I/O	3	P127	K15	M20	755
I/O	3	-	-	M18	758
V_{CCINT}	-	P128	V_{CCINT}*	V_{CCINT}*	-
I/O, TRDY(1)	3	P129	J15	M22	764
V_{CCO}	3	P130	V_{CCO} Bank 3*	V_{CCO} Bank 3*	-
V_{CCO}	2	P130	V_{CCO} Bank 2*	V_{CCO} Bank 2*	-
GND	-	P131	GND*	GND*	-

XC2S150 Device Pinouts *(Continued)*

XC2S150 Pad Name					
Function	Bank	PQ208	FG256	FG456	Bndry Scan
I/O, IRDY(1)	2	P132	H16	L20	767
I/O	2	P133	H14	L17	770
I/O	2	-	-	L18	773
I/O	2	P134	H15	L21	776
I/O	2	-	J13	L22	779
I/O (D3)	2	P135	G16	K20	782
I/O, V_{REF}	2	P136	H13	K21	785
V_{CCO}	2	-	V_{CCO} Bank 2*	V_{CCO} Bank 2*	-
GND	-	P137	GND*	GND*	-
I/O	2	P138	G14	K22	788
I/O	2	P139	G15	J21	791
I/O	2	-	-	J20	797
I/O	2	P140	G12	J18	800
I/O	2	-	F16	J22	803
I/O	2	-	-	J19	806
I/O	2	P141	G13	H19	812
I/O (D2)	2	P142	F15	H20	815
V_{CCINT}	-	P143	V_{CCINT}*	V_{CCINT}*	-
V_{CCO}	2	P144	V_{CCO} Bank 2*	V_{CCO} Bank 2*	-
GND	-	P145	GND*	GND*	-
I/O (D1)	2	P146	E16	H22	818
I/O, V_{REF}	2	P147	F14	H18	821
I/O	2	-	-	G21	824
I/O	2	P148	D16	G18	827
I/O	2	-	F12	G20	830
I/O	2	-	-	G19	833
I/O	2	-	-	F22	836
I/O	2	P149	E15	F19	839
I/O, V_{REF}	2	P150	F13	F21	842
V_{CCO}	2	-	V_{CCO} Bank 2*	V_{CCO} Bank 2*	-
GND	-	-	GND*	GND*	-
I/O	2	P151	E14	F20	845
I/O	2	-	C16	F18	848
I/O	2	-	-	E22	851
I/O	2	-	-	E21	854
I/O	2	P152	E13	D22	857
GND	-	-	GND*	GND*	-
I/O	2	-	B16	E20	860
I/O	2	-	-	D21	863

XC2S150 Device Pinouts *(Continued)*

XC2S150 Pad Name					
Function	Bank	PQ208	FG256	FG456	Bndry Scan
I/O	2	-	-	C22	866
I/O (DIN, D0)	2	P153	D14	D20	869
I/O (DOUT, BUSY)	2	P154	C15	C21	872
CCLK	2	P155	D15	B22	875
V_{CCO}	2	P156	V_{CCO} Bank 2*	V_{CCO} Bank 2*	-
V_{CCO}	1	P156	V_{CCO} Bank 1*	V_{CCO} Bank 1*	-
TDO	2	P157	B14	A21	-
GND	-	P158	GND*	GND*	-
TDI	-	P159	A15	B20	-
I/O ($\overline{CS}$)	1	P160	B13	C19	0
I/O ($\overline{WRITE}$)	1	P161	C13	A20	3
I/O	1	-	-	B19	6
I/O	1	-	-	C18	9
I/O	1	-	C12	D17	12
GND	-	-	GND*	GND*	-
I/O	1	P162	A14	A19	15
I/O	1	-	-	B18	18
I/O	1	-	-	E16	21
I/O	1	-	D12	C17	24
I/O	1	P163	B12	D16	27
GND	-	-	GND*	GND*	-
V_{CCO}	1	-	V_{CCO} Bank 1*	V_{CCO} Bank 1*	-
I/O, V_{REF}	1	P164	C11	A18	30
I/O	1	P165	A13	B17	33
I/O	1	-	-	E15	36
I/O	1	-	-	A17	39
I/O	1	-	D11	D15	42
I/O	1	P166	A12	C16	45
I/O	1	-	-	D14	48
I/O, V_{REF}	1	P167	E11	E14	51
I/O	1	P168	B11	A16	54
GND	-	P169	GND*	GND*	-
V_{CCO}	1	P170	V_{CCO} Bank 1*	V_{CCO} Bank 1*	-
V_{CCINT}	-	P171	V_{CCINT}*	V_{CCINT}*	-
I/O	1	P172	A11	C15	57
I/O	1	P173	C10	B15	60
I/O	1	-	-	A15	66
I/O	1	-	-	F12	69

XC2S150 Device Pinouts *(Continued)*

XC2S150 Pad Name					
Function	Bank	PQ208	FG256	FG456	Bndry Scan
I/O	1	P174	B10	C14	72
I/O	1	-	-	B14	75
I/O	1	P175	D10	D13	81
I/O	1	P176	A10	C13	84
GND	-	P177	GND*	GND*	-
V_{CCO}	1	-	V_{CCO} Bank 1*	V_{CCO} Bank 1*	-
I/O, V_{REF}	1	P178	B9	B13	87
I/O	1	P179	E10	E12	90
I/O	1	-	A9	B12	93
I/O	1	P180	D9	D12	96
I/O	1	-	-	C12	99
I/O	1	P181	A8	D11	102
I, GCK2	1	P182	C9	A11	108
GND	-	P183	GND*	GND*	-
V_{CCO}	1	P184	V_{CCO} Bank 1*	V_{CCO} Bank 1*	-
V_{CCO}	0	P184	V_{CCO} Bank 0*	V_{CCO} Bank 0*	-
I, GCK3	0	P185	B8	C11	109
V_{CCINT}	-	P186	V_{CCINT}*	V_{CCINT}*	-
I/O	0	-	-	E11	116
I/O	0	P187	A7	A10	119
I/O	0	-	D8	B10	122
I/O	0	P188	A6	C10	125
I/O, V_{REF}	0	P189	B7	A9	128
V_{CCO}	0	-	V_{CCO} Bank 0*	V_{CCO} Bank 0*	-
GND	-	P190	GND*	GND*	-
I/O	0	P191	C8	B9	131
I/O	0	P192	D7	E10	134
I/O	0	-	-	D10	140
I/O	0	P193	E7	A8	143
I/O	0	-	-	D9	146
I/O	0	-	-	B8	149
I/O	0	P194	C7	E9	155
I/O	0	P195	B6	A7	158

XC2S150 Device Pinouts *(Continued)*

XC2S150 Pad Name					
Function	Bank	PQ208	FG256	FG456	Bndry Scan
V_{CCINT}	-	P196	V_{CCINT}*	V_{CCINT}*	-
V_{CCO}	0	P197	V_{CCO} Bank 0*	V_{CCO} Bank 0*	-
GND	-	P198	GND*	GND*	-
I/O	0	P199	A5	B7	161
I/O, V_{REF}	0	P200	C6	E8	164
I/O	0	-	-	D8	167
I/O	0	P201	B5	C7	170
I/O	0	-	D6	D7	173
I/O	0	-	-	B6	176
I/O	0	-	-	A5	179
I/O	0	P202	A4	D6	182
I/O, V_{REF}	0	P203	B4	C6	185
V_{CCO}	0	-	V_{CCO} Bank 0*	V_{CCO} Bank 0*	-
GND	-	-	GND*	GND*	-
I/O	0	P204	E6	B5	188
I/O	0	-	D5	E7	191
I/O	0	-	-	A4	194
I/O	0	-	-	E6	197
I/O	0	P205	A3	B4	200
GND	-	-	GND*	GND*	-
I/O	0	-	C5	A3	203
I/O	0	-	-	B3	206
I/O	0	-	-	D5	209
I/O	0	P206	B3	C5	212
TCK	-	P207	C4	C4	-
V_{CCO}	0	P208	V_{CCO} Bank 0*	V_{CCO} Bank 0*	-
V_{CCO}	7	P208	V_{CCO} Bank 7*	V_{CCO} Bank 7*	-

04/18/01

Notes:

1. IRDY and TRDY can only be accessed when using Xilinx PCI cores.
2. Pads labelled GND*, V_{CCINT}*, V_{CCO} Bank 0*, V_{CCO} Bank 1*, V_{CCO} Bank 2*, V_{CCO} Bank 3*, V_{CCO} Bank 4*, V_{CCO} Bank 5*, V_{CCO} Bank 6*, V_{CCO} Bank 7* are internally bonded to independent ground or power planes within the package.
3. See "VCCO Banks" for details on V_{CCO} banking.

XC2S200 Device Pinouts

XC2S200 Pad Name					
Function	Bank	PQ208	FG256	FG456	Bndry Scan
GND	-	P1	GND*	GND*	-
TMS	-	P2	D3	D3	-
I/O	7	P3	C2	B1	257
I/O	7	-	-	E4	263
I/O	7	-	-	C1	266
I/O	7	-	A2	F5	269
GND	-	-	GND*	GND*	-
I/O, V_{REF}	7	P4	B1	D2	272
I/O	7	-	-	E3	275
I/O	7	-	-	F4	281
GND	-	-	GND*	GND*	-
I/O	7	-	E3	G5	284
I/O	7	P5	D2	F3	287
GND	-	-	GND*	GND*	-
V_{CCO}	7	-	V_{CCO} Bank 7*	V_{CCO} Bank 7*	-
I/O, V_{REF}	7	P6	C1	E2	290
I/O	7	P7	F3	E1	293
I/O	7	-	-	G4	296
I/O	7	-	-	G3	299
I/O	7	-	E2	H5	302
GND	-	-	GND*	GND*	-
I/O	7	P8	E4	F2	305
I/O	7	-	-	F1	308
I/O, V_{REF}	7	P9	D1	H4	314
I/O	7	P10	E1	G1	317
GND	-	P11	GND*	GND*	-
V_{CCO}	7	P12	V_{CCO} Bank 7*	V_{CCO} Bank 7*	-
V_{CCINT}	-	P13	V_{CCINT}*	V_{CCINT}*	-
I/O	7	P14	F2	H3	320
I/O	7	P15	G3	H2	323
I/O	7	-	-	J4	326
I/O	7	-	-	H1	329
I/O	7	-	F1	J5	332
GND	-	-	GND*	GND*	-
I/O	7	P16	F4	J2	335
I/O	7	-	-	J3	338
I/O	7	-	-	J1	341
I/O	7	P17	F5	K5	344
I/O	7	P18	G2	K1	347
GND	-	P19	GND*	GND*	-

XC2S200 Device Pinouts *(Continued)*

XC2S200 Pad Name					
Function	Bank	PQ208	FG256	FG456	Bndry Scan
V_{CCO}	7	-	V_{CCO} Bank 7*	V_{CCO} Bank 7*	-
I/O, V_{REF}	7	P20	H3	K3	350
I/O	7	P21	G4	K4	353
I/O	7	-	-	K2	359
I/O	7	-	H2	L6	362
I/O	7	P22	G5	L1	365
I/O	7	-	-	L5	368
I/O	7	P23	H4	L4	374
I/O, IRDY(1)	7	P24	G1	L3	377
GND	-	P25	GND*	GND*	-
V_{CCO}	7	P26	V_{CCO} Bank 7*	V_{CCO} Bank 7*	-
V_{CCO}	6	P26	V_{CCO} Bank 6*	V_{CCO} Bank 6*	-
I/O, TRDY(1)	6	P27	J2	M1	380
V_{CCINT}	-	P28	V_{CCINT}*	V_{CCINT}*	-
I/O	6	-	-	M6	389
I/O	6	P29	H1	M3	392
I/O	6	-	J4	M4	395
I/O	6	-	-	N1	398
I/O	6	P30	J1	M5	404
I/O, V_{REF}	6	P31	J3	N2	407
V_{CCO}	6	-	V_{CCO} Bank 6*	V_{CCO} Bank 6*	-
GND	-	P32	GND*	GND*	-
I/O	6	P33	K5	N3	410
I/O	6	P34	K2	N4	413
I/O	6	-	-	P1	416
I/O	6	-	-	N5	419
I/O	6	P35	K1	P2	422
GND	-	-	GND*	GND*	-
I/O	6	-	K3	P4	425
I/O	6	-	-	R1	428
I/O	6	-	-	P5	431
I/O	6	P36	L1	P3	434
I/O	6	P37	L2	R2	437
V_{CCINT}	-	P38	V_{CCINT}*	V_{CCINT}*	-
V_{CCO}	6	P39	V_{CCO} Bank 6*	V_{CCO} Bank 6*	-
GND	-	P40	GND*	GND*	-
I/O	6	P41	K4	T1	440
I/O, V_{REF}	6	P42	M1	R4	443