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What is "[Embedded - Microcontrollers](#)"?

"[Embedded - Microcontrollers](#)" refer to small, integrated circuits designed to perform specific tasks within larger systems. These microcontrollers are essentially compact computers on a single chip, containing a processor core, memory, and programmable input/output peripherals. They are called "embedded" because they are embedded within electronic devices to control various functions, rather than serving as standalone computers. Microcontrollers are crucial in modern electronics, providing the intelligence and control needed for a wide range of applications.

Applications of "[Embedded - Microcontrollers](#)"

Details

Product Status	Discontinued at Digi-Key
Core Processor	ARM® Cortex®-M0+
Core Size	32-Bit Single-Core
Speed	48MHz
Connectivity	CANbus, I ² C, IrDA, LINbus, SmartCard, SPI, UART/USART
Peripherals	Brown-out Detect/Reset, DMA, I ² S, LCD, POR, PWM, WDT
Number of I/O	22
Program Memory Size	64KB (64K x 8)
Program Memory Type	FLASH
EEPROM Size	-
RAM Size	32K x 8
Voltage - Supply (Vcc/Vdd)	1.8V ~ 3.8V
Data Converters	A/D 12bit SAR; D/A 12bit
Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	32-WFQFN Exposed Pad
Supplier Device Package	32-QFN (5x5)
Purchase URL	https://www.e-xfl.com/product-detail/silicon-labs/efm32tg11b540f64gm32-a

1. Feature List

The EFM32TG11 highlighted features are listed below.

- **ARM Cortex-M0+ CPU platform**
 - High performance 32-bit processor @ up to 48 MHz
 - Memory Protection Unit
 - Wake-up Interrupt Controller
- **Flexible Energy Management System**
 - 37 μ A/MHz in Active Mode (EM0)
 - 1.30 μ A EM2 Deep Sleep current (8 kB RAM retention and RTCC running from LFRCO)
- **Integrated DC-DC buck converter**
- **Backup Power Domain**
 - RTCC and retention registers in a separate power domain, available in all energy modes
 - Operation from backup battery when main power absent/insufficient
- **Up to 128 kB flash program memory**
- **Up to 32 kB RAM data memory**
- **Communication Interfaces**
 - CAN Bus Controller
 - Version 2.0A and 2.0B up to 1 Mbps
 - 4 \times Universal Synchronous/Asynchronous Receiver/ Transmitter
 - UART/SPI/SmartCard (ISO 7816)/IrDA/I2S/LIN
 - Triple buffered full/half-duplex operation with flow control
 - Ultra high speed (24 MHz) operation on one instance
 - 1 \times Universal Asynchronous Receiver/ Transmitter
 - 1 \times Low Energy UART
 - Autonomous operation with DMA in Deep Sleep Mode
 - 2 \times I²C Interface with SMBus support
 - Address recognition in EM3 Stop Mode
- **Up to 67 General Purpose I/O Pins**
 - Configurable push-pull, open-drain, pull-up/down, input filter, drive strength
 - Configurable peripheral I/O locations
 - 5 V tolerance on select pins
 - Asynchronous external interrupts
 - Output state retention and wake-up from Shutoff Mode
- **Up to 8 Channel DMA Controller**
- **Up to 8 Channel Peripheral Reflex System (PRS) for autonomous inter-peripheral signaling**
- **Hardware Cryptography**
 - AES 128/256-bit keys
 - ECC B/K163, B/K233, P192, P224, P256
 - SHA-1 and SHA-2 (SHA-224 and SHA-256)
 - True Random Number Generator (TRNG)
- **Hardware CRC engine**
 - Single-cycle computation with 8/16/32-bit data and 16-bit (programmable)/32-bit (fixed) polynomial
- **Security Management Unit (SMU)**
 - Fine-grained access control for on-chip peripherals
- **Integrated Low-energy LCD Controller with up to 8 \times 32 segments**
 - Voltage boost, contrast and autonomous animation
 - Patented low-energy LCD driver
- **Ultra Low-Power Precision Analog Peripherals**
 - 12-bit 1 Msamples/s Analog to Digital Converter (ADC)
 - On-chip temperature sensor
 - 2 \times 12-bit 500 ksamples/s Digital to Analog Converter (VDAC)
 - Up to 2 \times Analog Comparator (ACMP)
 - Up to 4 \times Operational Amplifier (OPAMP)
 - Robust current-based capacitive sensing with up to 38 inputs and wake-on-touch (CSEN)
 - Up to 62 GPIO pins are analog-capable. Flexible analog peripheral-to-pin routing via Analog Port (APORT)
 - Supply Voltage Monitor

Ordering Code	Flash (kB)	RAM (kB)	DC-DC Converter	LCD	GPIO	Package	Temp Range
EFM32TG11B320F128GQ48-A	128	32	No	Yes	37	QFP48	-40 to +85°C
EFM32TG11B320F128IQ48-A	128	32	No	Yes	37	QFP48	-40 to +125°C
EFM32TG11B340F64GQ48-A	64	32	No	Yes	37	QFP48	-40 to +85°C
EFM32TG11B340F64IQ48-A	64	32	No	Yes	37	QFP48	-40 to +125°C
EFM32TG11B120F128GM64-A	128	32	No	No	56	QFN64	-40 to +85°C
EFM32TG11B120F128GQ64-A	128	32	No	No	53	QFP64	-40 to +85°C
EFM32TG11B120F128IM64-A	128	32	No	No	56	QFN64	-40 to +125°C
EFM32TG11B120F128IQ64-A	128	32	No	No	53	QFP64	-40 to +125°C
EFM32TG11B140F64GM64-A	64	32	No	No	56	QFN64	-40 to +85°C
EFM32TG11B140F64GQ64-A	64	32	No	No	53	QFP64	-40 to +85°C
EFM32TG11B140F64IM64-A	64	32	No	No	56	QFN64	-40 to +125°C
EFM32TG11B140F64IQ64-A	64	32	No	No	53	QFP64	-40 to +125°C
EFM32TG11B120F128GQ48-A	128	32	No	No	37	QFP48	-40 to +85°C
EFM32TG11B120F128IQ48-A	128	32	No	No	37	QFP48	-40 to +125°C
EFM32TG11B140F64GQ48-A	64	32	No	No	37	QFP48	-40 to +85°C
EFM32TG11B140F64IQ48-A	64	32	No	No	37	QFP48	-40 to +125°C
EFM32TG11B120F128GM32-A	128	32	No	No	24	QFN32	-40 to +85°C
EFM32TG11B120F128IM32-A	128	32	No	No	24	QFN32	-40 to +125°C
EFM32TG11B140F64GM32-A	64	32	No	No	24	QFN32	-40 to +85°C
EFM32TG11B140F64IM32-A	64	32	No	No	24	QFN32	-40 to +125°C

3.6.5 Controller Area Network (CAN)

The CAN peripheral provides support for communication at up to 1 Mbps over CAN protocol version 2.0 part A and B. It includes 32 message objects with independent identifier masks and retains message RAM in EM2. Automatic retransmission may be disabled in order to support Time Triggered CAN applications.

3.6.6 Peripheral Reflex System (PRS)

The Peripheral Reflex System provides a communication network between different peripheral modules without software involvement. Peripheral modules producing Reflex signals are called producers. The PRS routes Reflex signals from producers to consumer peripherals which in turn perform actions in response. Edge triggers and other functionality such as simple logic operations (AND, OR, NOT) can be applied by the PRS to the signals. The PRS allows peripheral to act autonomously without waking the MCU core, saving power.

3.6.7 Low Energy Sensor Interface (LESENSE)

The Low Energy Sensor Interface LESENSE™ is a highly configurable sensor interface with support for up to 16 individually configurable sensors. By controlling the analog comparators, ADC, and DAC, LESENSE is capable of supporting a wide range of sensors and measurement schemes, and can for instance measure LC sensors, resistive sensors and capacitive sensors. LESENSE also includes a programmable finite state machine which enables simple processing of measurement results without CPU intervention. LESENSE is available in energy mode EM2, in addition to EM0 and EM1, making it ideal for sensor monitoring in applications with a strict energy budget.

3.7 Security Features

3.7.1 GPCRC (General Purpose Cyclic Redundancy Check)

The GPCRC module implements a Cyclic Redundancy Check (CRC) function. It supports both 32-bit and 16-bit polynomials. The supported 32-bit polynomial is 0x04C11DB7 (IEEE 802.3), while the 16-bit polynomial can be programmed to any value, depending on the needs of the application.

3.7.2 Crypto Accelerator (CRYPTO)

The Crypto Accelerator is a fast and energy-efficient autonomous hardware encryption and decryption accelerator. Tiny Gecko Series 1 devices support AES encryption and decryption with 128- or 256-bit keys, ECC over both GF(P) and GF(2^m), and SHA-1 and SHA-2 (SHA-224 and SHA-256).

Supported block cipher modes of operation for AES include: ECB, CTR, CBC, PCBC, CFB, OFB, GCM, CBC-MAC, GMAC and CCM.

Supported ECC NIST recommended curves include P-192, P-224, P-256, K-163, K-233, B-163 and B-233.

The CRYPTO module allows fast processing of GCM (AES), ECC and SHA with little CPU intervention. CRYPTO also provides trigger signals for DMA read and write operations.

3.7.3 True Random Number Generator (TRNG)

The TRNG module is a non-deterministic random number generator based on a full hardware solution. The TRNG is validated with NIST800-22 and AIS-31 test suites as well as being suitable for FIPS 140-2 certification (for the purposes of cryptographic key generation).

3.7.4 Security Management Unit (SMU)

The Security Management Unit (SMU) allows software to set up fine-grained security for peripheral access, which is not possible in the Memory Protection Unit (MPU). Peripherals may be secured by hardware on an individual basis, such that only privileged accesses to the peripheral's register interface will be allowed. When an access fault occurs, the SMU reports the specific peripheral involved and can optionally generate an interrupt.

3.8 Analog

3.8.1 Analog Port (APORT)

The Analog Port (APORT) is an analog interconnect matrix allowing access to many analog modules on a flexible selection of pins. Each APORT bus consists of analog switches connected to a common wire. Since many clients can operate differentially, buses are grouped by X/Y pairs.

3.8.2 Analog Comparator (ACMP)

The Analog Comparator is used to compare the voltage of two analog inputs, with a digital output indicating which input voltage is higher. Inputs are selected from among internal references and external pins. The tradeoff between response time and current consumption is configurable by software. Two 6-bit reference dividers allow for a wide range of internally-programmable reference sources. The ACMP can also be used to monitor the supply voltage. An interrupt can be generated when the supply falls below or rises above the programmable threshold.

3.8.3 Analog to Digital Converter (ADC)

The ADC is a Successive Approximation Register (SAR) architecture, with a resolution of up to 12 bits at up to 1 Msps. The output sample resolution is configurable and additional resolution is possible using integrated hardware for averaging over multiple samples. The ADC includes integrated voltage references and an integrated temperature sensor. Inputs are selectable from a wide range of sources, including pins configurable as either single-ended or differential.

3.8.4 Capacitive Sense (CSEN)

The CSEN module is a dedicated Capacitive Sensing block for implementing touch-sensitive user interface elements such as switches and sliders. The CSEN module uses a charge ramping measurement technique, which provides robust sensing even in adverse conditions including radiated noise and moisture. The module can be configured to take measurements on a single port pin or scan through multiple pins and store results to memory through DMA. Several channels can also be shorted together to measure the combined capacitance or implement wake-on-touch from very low energy modes. Hardware includes a digital accumulator and an averaging filter, as well as digital threshold comparators to reduce software overhead.

3.8.5 Digital to Analog Converter (VDAC)

The Digital to Analog Converter (VDAC) can convert a digital value to an analog output voltage. The VDAC is a fully differential, 500 ksp/s, 12-bit converter. The opamps are used in conjunction with the VDAC, to provide output buffering. One opamp is used per single-ended channel, or two opamps are used to provide differential outputs. The VDAC may be used for a number of different applications such as sensor interfaces or sound output. The VDAC can generate high-resolution analog signals while the MCU is operating at low frequencies and with low total power consumption. Using DMA and a timer, the VDAC can be used to generate waveforms without any CPU intervention. The VDAC is available in all energy modes down to and including EM3.

3.8.6 Operational Amplifiers

The opamps are low power amplifiers with a high degree of flexibility targeting a wide variety of standard opamp application areas, and are available down to EM3. With flexible built-in programming for gain and interconnection they can be configured to support multiple common opamp functions. All pins are also available externally for filter configurations. Each opamp has a rail to rail input and a rail to rail output. They can be used in conjunction with the VDAC module or in stand-alone configurations. The opamps save energy, PCB space, and cost as compared with standalone opamps because they are integrated on-chip.

3.8.7 Liquid Crystal Display Driver (LCD)

The LCD driver is capable of driving a segmented LCD display with up to 8x32 segments. A voltage boost function enables it to provide the LCD display with higher voltage than the supply voltage for the device. A patented charge redistribution driver can reduce the LCD module supply current by up to 40%. In addition, an animation feature can run custom animations on the LCD display without any CPU intervention. The LCD driver can also remain active even in Energy Mode 2 and provides a Frame Counter interrupt that can wake-up the device on a regular basis for updating data.

3.9 Reset Management Unit (RMU)

The RMU is responsible for handling reset of the EFM32TG11. A wide range of reset sources are available, including several power supply monitors, pin reset, software controlled reset, core lockup reset, and watchdog reset.

4. Electrical Specifications

4.1 Electrical Characteristics

All electrical parameters in all tables are specified under the following conditions, unless stated otherwise:

- Typical values are based on $T_{AMB}=25^{\circ}\text{C}$ and $V_{DD}=3.3\text{ V}$, by production test and/or technology characterization.
- Minimum and maximum values represent the worst conditions across supply voltage, process variation, and operating temperature, unless stated otherwise.

Refer to [4.1.2.1 General Operating Conditions](#) for more details about operational supply and temperature limits.

4.1.1 Absolute Maximum Ratings

Stresses above those listed below may cause permanent damage to the device. This is a stress rating only and functional operation of the devices at those or any other conditions above those indicated in the operation listings of this specification is not implied. Exposure to maximum rating conditions for extended periods may affect device reliability. For more information on the available quality and reliability data, see the Quality and Reliability Monitor Report at <http://www.silabs.com/support/quality/pages/default.aspx>.

Table 4.1. Absolute Maximum Ratings

Parameter	Symbol	Test Condition	Min	Typ	Max	Unit
Storage temperature range	T_{STG}		-50	—	150	$^{\circ}\text{C}$
Voltage on any supply pin	V_{DDMAX}		-0.3	—	3.8	V
Voltage ramp rate on any supply pin	$V_{DDRAMPMAX}$		—	—	1	V / μs
DC voltage on any GPIO pin	V_{DIGPIN}	5V tolerant GPIO pins ^{1 2 3}	-0.3	—	Min of 5.25 and IOVDD +2	V
		LCD pins ³	-0.3	—	Min of 3.8 and IOVDD +2	V
		Standard GPIO pins	-0.3	—	IOVDD+0.3	V
Total current into VDD power lines	I_{VDDMAX}	Source	—	—	200	mA
Total current into VSS ground lines	I_{VSSMAX}	Sink	—	—	200	mA
Current per I/O pin	I_{IOMAX}	Sink	—	—	50	mA
		Source	—	—	50	mA
Current for all I/O pins	$I_{IOALLMAX}$	Sink	—	—	200	mA
		Source	—	—	200	mA
Junction temperature	T_J	-G grade devices	-40	—	105	$^{\circ}\text{C}$
		-I grade devices	-40	—	125	$^{\circ}\text{C}$

Note:

1. When a GPIO pin is routed to the analog module through the APORT, the maximum voltage = IOVDD.
2. Valid for IOVDD in valid operating range or when IOVDD is undriven (high-Z). If IOVDD is connected to a low-impedance source below the valid operating range (e.g. IOVDD shorted to VSS), the pin voltage maximum is IOVDD + 0.3 V, to avoid exceeding the maximum IO current specifications.
3. To operate above the IOVDD supply rail, over-voltage tolerance must be enabled according to the GPIO_Px_OVTDIS register. Pins with over-voltage tolerance disabled have the same limits as Standard GPIO.

4.1.2.1 General Operating Conditions

Table 4.2. General Operating Conditions

Parameter	Symbol	Test Condition	Min	Typ	Max	Unit
Operating ambient temperature range ⁶	T _A	-G temperature grade	-40	25	85	°C
		-I temperature grade	-40	25	125	°C
AVDD supply voltage ²	V _{AVDD}		1.8	3.3	3.8	V
VREGVDD operating supply voltage ^{2 1}	V _{VREGVDD}	DCDC in regulation	2.4	3.3	3.8	V
		DCDC in bypass, 50mA load	1.8	3.3	3.8	V
		DCDC not in use. DVDD externally shorted to VREGVDD	1.8	3.3	3.8	V
VREGVDD current	I _{VREGVDD}	DCDC in bypass, T ≤ 85 °C	—	—	200	mA
		DCDC in bypass, T > 85 °C	—	—	100	mA
DVDD operating supply voltage	V _{DVDD}		1.62	—	V _{VREGVDD}	V
IOVDD operating supply voltage	V _{IOVDD}	All IOVDD pins ⁵	1.62	—	V _{VREGVDD}	V
DECOUPLE output capacitor ^{3 4}	C _{DECOUPLE}		0.75	1.0	2.75	μF
HFCORECLK frequency	f _{CORE}	VSCALE2, MODE = WS1	—	—	48	MHz
		VSCALE2, MODE = WS0	—	—	25	MHz
		VSCALE0, MODE = WS1	—	—	20	MHz
		VSCALE0, MODE = WS0	—	—	10	MHz
HFCLK frequency	f _{HFCLK}	VSCALE2	—	—	48	MHz
		VSCALE0	—	—	20	MHz
HFSRCCLK frequency	f _{HFSRCCLK}	VSCALE2	—	—	48	MHz
		VSCALE0	—	—	20	MHz
HFBUSCLK frequency	f _{HFBUSCLK}	VSCALE2	—	—	48	MHz
		VSCALE0	—	—	20	MHz
HFPERCLK frequency	f _{HFPERCLK}	VSCALE2	—	—	48	MHz
		VSCALE0	—	—	20	MHz
HFPERBCLK frequency	f _{HFPERBCLK}	VSCALE2	—	—	48	MHz
		VSCALE0	—	—	20	MHz
HFPERCCLK frequency	f _{HFPERCCLK}	VSCALE2	—	—	48	MHz
		VSCALE0	—	—	20	MHz

Parameter	Symbol	Test Condition	Min	Typ	Max	Unit
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Note:

1. The minimum voltage required in bypass mode is calculated using R_{BYP} from the DCDC specification table. Requirements for other loads can be calculated as $V_{DVDD_min} + I_{LOAD} * R_{BYP_max}$.
2. VREGVDD must be tied to AVDD. Both VREGVDD and AVDD minimum voltages must be satisfied for the part to operate.
3. The system designer should consult the characteristic specs of the capacitor used on DECOUPLE to ensure its capacitance value stays within the specified bounds across temperature and DC bias.
4. VSCALE0 to VSCALE2 voltage change transitions occur at a rate of 10 mV / usec for approximately 20 usec. During this transition, peak currents will be dependent on the value of the DECOUPLE output capacitor, from 35 mA (with a 1 μ F capacitor) to 70 mA (with a 2.7 μ F capacitor).
5. When the CSEN peripheral is used with chopping enabled (CSEN_CTRL_CHOPEN = ENABLE), IOVDD must be equal to AVDD.
6. The maximum limit on T_A may be lower due to device self-heating, which depends on the power dissipation of the specific application. $T_A (max) = T_J (max) - (THETA_{JA} \times PowerDissipation)$. Refer to the Absolute Maximum Ratings table and the Thermal Characteristics table for T_J and $THETA_{JA}$.

4.1.3 Thermal Characteristics

Table 4.3. Thermal Characteristics

Parameter	Symbol	Test Condition	Min	Typ	Max	Unit
Thermal resistance, QFN32 Package	$THETA_{JA_QFN32}$	4-Layer PCB, Air velocity = 0 m/s	—	25.7	—	$^{\circ}C/W$
		4-Layer PCB, Air velocity = 1 m/s	—	23.2	—	$^{\circ}C/W$
		4-Layer PCB, Air velocity = 2 m/s	—	21.3	—	$^{\circ}C/W$
Thermal resistance, TQFP48 Package	$THE- TA_{JA_TQFP48}$	4-Layer PCB, Air velocity = 0 m/s	—	44.1	—	$^{\circ}C/W$
		4-Layer PCB, Air velocity = 1 m/s	—	43.5	—	$^{\circ}C/W$
		4-Layer PCB, Air velocity = 2 m/s	—	42.3	—	$^{\circ}C/W$
Thermal resistance, QFN64 Package	$THETA_{JA_QFN64}$	4-Layer PCB, Air velocity = 0 m/s	—	20.9	—	$^{\circ}C/W$
		4-Layer PCB, Air velocity = 1 m/s	—	18.2	—	$^{\circ}C/W$
		4-Layer PCB, Air velocity = 2 m/s	—	16.4	—	$^{\circ}C/W$
Thermal resistance, TQFP64 Package	$THE- TA_{JA_TQFP64}$	4-Layer PCB, Air velocity = 0 m/s	—	37.3	—	$^{\circ}C/W$
		4-Layer PCB, Air velocity = 1 m/s	—	35.6	—	$^{\circ}C/W$
		4-Layer PCB, Air velocity = 2 m/s	—	33.8	—	$^{\circ}C/W$
Thermal resistance, QFN80 Package	$THETA_{JA_QFN80}$	4-Layer PCB, Air velocity = 0 m/s	—	20.9	—	$^{\circ}C/W$
		4-Layer PCB, Air velocity = 1 m/s	—	18.2	—	$^{\circ}C/W$
		4-Layer PCB, Air velocity = 2 m/s	—	16.4	—	$^{\circ}C/W$
Thermal resistance, TQFP80 Package	$THE- TA_{JA_TQFP80}$	4-Layer PCB, Air velocity = 0 m/s	—	49.3	—	$^{\circ}C/W$
		4-Layer PCB, Air velocity = 1 m/s	—	44.5	—	$^{\circ}C/W$
		4-Layer PCB, Air velocity = 2 m/s	—	42.6	—	$^{\circ}C/W$

4.1.8 Brown Out Detector (BOD)

Table 4.10. Brown Out Detector (BOD)

Parameter	Symbol	Test Condition	Min	Typ	Max	Unit
DVDD BOD threshold	V _{DVddbOD}	DVDD rising	—	—	TBD	V
		DVDD falling (EM0/EM1)	TBD	—	—	V
		DVDD falling (EM2/EM3)	TBD	—	—	V
DVDD BOD hysteresis	V _{DVddbOD_HYST}		—	18	—	mV
DVDD BOD response time	t _{DVddbOD_DELAY}	Supply drops at 0.1V/μs rate	—	2.4	—	μs
AVDD BOD threshold	V _{AVddbOD}	AVDD rising	—	—	TBD	V
		AVDD falling (EM0/EM1)	TBD	—	—	V
		AVDD falling (EM2/EM3)	TBD	—	—	V
AVDD BOD hysteresis	V _{AVddbOD_HYST}		—	20	—	mV
AVDD BOD response time	t _{AVddbOD_DELAY}	Supply drops at 0.1V/μs rate	—	2.4	—	μs
EM4 BOD threshold	V _{EM4dBOD}	AVDD rising	—	—	TBD	V
		AVDD falling	TBD	—	—	V
EM4 BOD hysteresis	V _{EM4BOD_HYST}		—	25	—	mV
EM4 BOD response time	t _{EM4BOD_DELAY}	Supply drops at 0.1V/μs rate	—	300	—	μs

Parameter	Symbol	Test Condition	Min	Typ	Max	Unit
ADC clock frequency	f_{ADCCLK}		—	—	16	MHz
Throughput rate	f_{ADCRATE}		—	—	1	Msp/s
Conversion time ¹	t_{ADCCONV}	6 bit	—	7	—	cycles
		8 bit	—	9	—	cycles
		12 bit	—	13	—	cycles
Startup time of reference generator and ADC core	t_{ADCSTART}	WARMUPMODE ⁴ = NORMAL	—	—	5	μs
		WARMUPMODE ⁴ = KEEPIN-STANDBY	—	—	2	μs
		WARMUPMODE ⁴ = KEEPINSLOWACC	—	—	1	μs
SNDR at 1Msp/s and $f_{\text{IN}} = 10\text{kHz}$	SNDR_{ADC}	Internal reference ⁷ , differential measurement	TBD	67	—	dB
		External reference ⁶ , differential measurement	—	68	—	dB
Spurious-free dynamic range (SFDR)	SFDR_{ADC}	1 MSamples/s, 10 kHz full-scale sine wave	—	75	—	dB
Differential non-linearity (DNL)	DNL_{ADC}	12 bit resolution, No missing codes	TBD	—	TBD	LSB
Integral non-linearity (INL), End point method	INL_{ADC}	12 bit resolution	TBD	—	TBD	LSB
Offset error	$V_{\text{ADCOFFSETERR}}$		TBD	0	TBD	LSB
Gain error in ADC	V_{ADCGAIN}	Using internal reference	—	-0.2	TBD	%
		Using external reference	—	-1	—	%
Temperature sensor slope	$V_{\text{TS_SLOPE}}$		—	-1.84	—	mV/°C

Note:

1. Derived from ADCCLK.
2. PSRR is referenced to AVDD when ANASW=0 and to DVDD when ANASW=1 in EMU_PWRCTRL.
3. In ADCn_BIASPROG register.
4. In ADCn_CNTL register.
5. The absolute voltage allowed at any ADC input is dictated by the power rail supplied to on-chip circuitry, and may be lower than the effective full scale voltage. All ADC inputs are limited to the ADC supply (AVDD or DVDD depending on EMU_PWRCTRL_ANASW). Any ADC input routed through the APORT will further be limited by the IOVDD supply to the pin.
6. External reference is 1.25 V applied externally to ADCnEXTREFP, with the selection CONF in the SINGLECTRL_REF or SCANCTRL_REF register field and VREFP in the SINGLECTRLX_VREFSEL or SCANCTRLX_VREFSEL field. The differential input range with this configuration is $\pm 1.25\text{ V}$.
7. Internal reference option used corresponds to selection 2V5 in the SINGLECTRL_REF or SCANCTRL_REF register field. The differential input range with this configuration is $\pm 1.25\text{ V}$. Typical value is characterized using full-scale sine wave input. Minimum value is production-tested using sine wave input at 1.5 dB lower than full scale.

4.1.15 Digital to Analog Converter (VDAC)

DRIVESTRENGTH = 2 unless otherwise specified. Primary VDAC output.

Table 4.22. Digital to Analog Converter (VDAC)

Parameter	Symbol	Test Condition	Min	Typ	Max	Unit
Output voltage	V_{DACOUT}	Single-Ended	0	—	V_{VREF}	V
		Differential ²	$-V_{VREF}$	—	V_{VREF}	V
Current consumption including references (2 channels) ¹	I_{DAC}	500 ksps, 12-bit, DRIVESTRENGTH = 2, REFSEL = 4	—	396	—	μA
		44.1 ksps, 12-bit, DRIVESTRENGTH = 1, REFSEL = 4	—	72	—	μA
		200 Hz refresh rate, 12-bit Sample-Off mode in EM2, DRIVESTRENGTH = 2, BGRREQTIME = 1, EM2REFENTIME = 9, REFSEL = 4, SETTLETIME = 0x0A, WARMUPTIME = 0x02	—	2	—	μA
Current from HFPERCLK ⁴	I_{DAC_CLK}		—	5.8	—	$\mu A/MHz$
Sample rate	SR_{DAC}		—	—	500	ksps
DAC clock frequency	f_{DAC}		—	—	1	MHz
Conversion time	$t_{DACCONV}$	$f_{DAC} = 1MHz$	2	—	—	μs
Settling time	$t_{DACSETTLE}$	50% fs step settling to 5 LSB	—	2.5	—	μs
Startup time	$t_{DACSTARTUP}$	Enable to 90% fs output, settling to 10 LSB	—	—	12	μs
Output impedance	R_{OUT}	DRIVESTRENGTH = 2, $0.4 V \leq V_{OUT} \leq V_{OPA} - 0.4 V$, $-8 mA < I_{OUT} < 8 mA$, Full supply range	—	2	—	Ω
		DRIVESTRENGTH = 0 or 1, $0.4 V \leq V_{OUT} \leq V_{OPA} - 0.4 V$, $-400 \mu A < I_{OUT} < 400 \mu A$, Full supply range	—	2	—	Ω
		DRIVESTRENGTH = 2, $0.1 V \leq V_{OUT} \leq V_{OPA} - 0.1 V$, $-2 mA < I_{OUT} < 2 mA$, Full supply range	—	2	—	Ω
		DRIVESTRENGTH = 0 or 1, $0.1 V \leq V_{OUT} \leq V_{OPA} - 0.1 V$, $-100 \mu A < I_{OUT} < 100 \mu A$, Full supply range	—	2	—	Ω
Power supply rejection ratio ⁶	PSRR	$V_{out} = 50\% fs$, DC	—	65.5	—	dB

Parameter	Symbol	Test Condition	Min	Typ	Max	Unit
Supply current, continuous conversions, WARMUP-MODE=KEEPCSENWARM	I _{CSEN_ACTIVE}	SAR or Delta Modulation conversions of 33 pF capacitor, CS0CG=0 (Gain = 10x), always on	—	90.5	—	μA
HFPERCLK supply current	I _{CSEN_HFPERCLK}	Current contribution from HFPERCLK when clock to CSEN block is enabled.	—	2.25	—	μA/MHz

Note:

- Current is specified with a total external capacitance of 33 pF per channel. Average current is dependent on how long the module is actively sampling channels within the scan period, and scales with the number of samples acquired. Supply current for a specific application can be estimated by multiplying the current per sample by the total number of samples per period (total_current = single_sample_current * (number_of_channels * accumulation)).

4.2.1 Supply Current

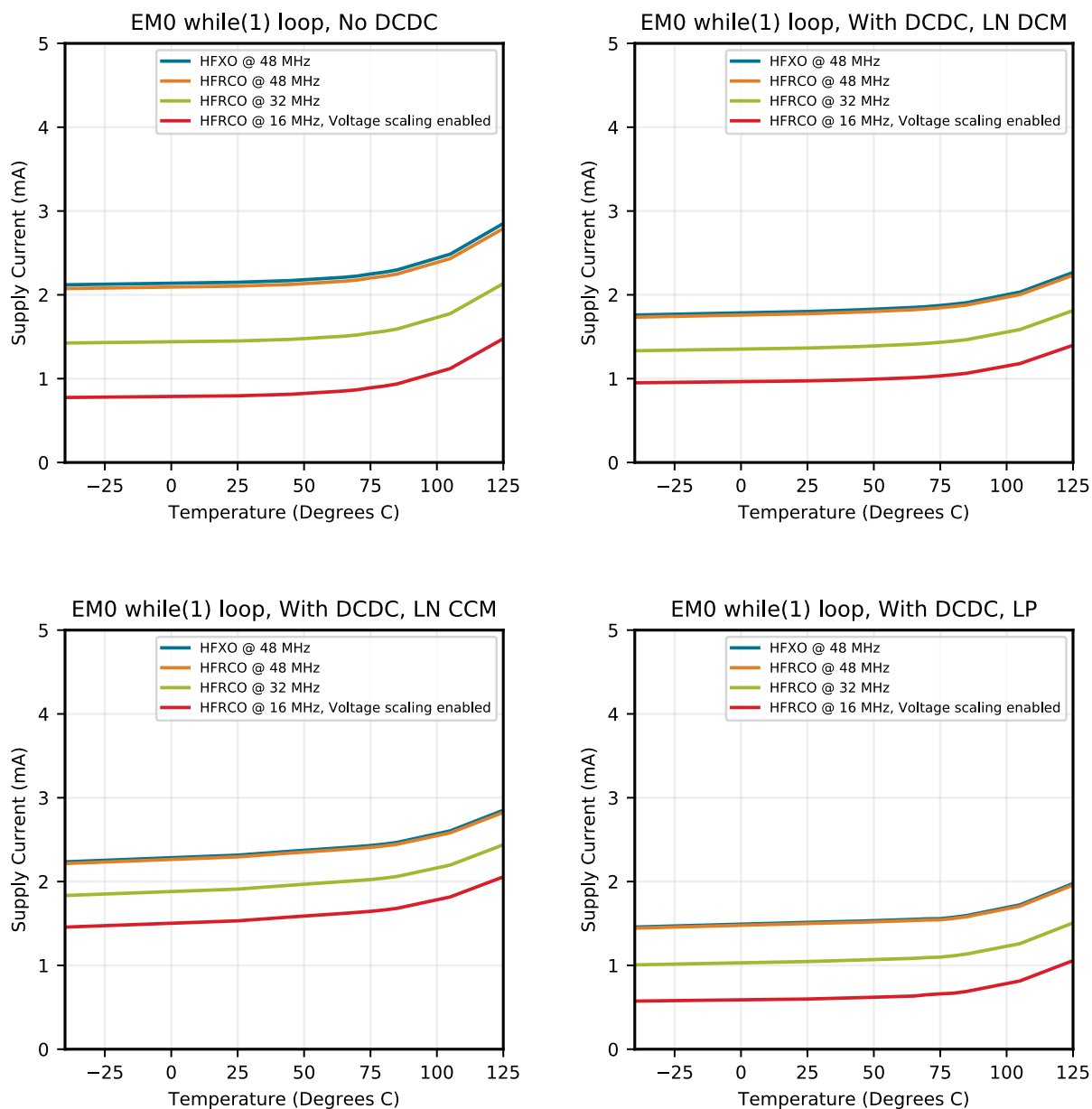


Figure 4.3. EM0 Active Mode Typical Supply Current vs. Temperature

Pin Name	Pin(s)	Description	Pin Name	Pin(s)	Description
VSS	9 24 51 70	Ground	PB3	10	GPIO
PB4	11	GPIO	PB5	12	GPIO
PB6	13	GPIO	PC1	14	GPIO (5V)
PC2	15	GPIO (5V)	PC3	16	GPIO (5V)
PC4	17	GPIO	PC5	18	GPIO
PB7	19	GPIO	PB8	20	GPIO
PA8	21	GPIO	PA9	22	GPIO
PA10	23	GPIO	PA12	25	GPIO
PA14	26	GPIO	RESETn	27	Reset input, active low. To apply an external reset source to this pin, it is required to only drive this pin low during reset, and let the internal pull-up ensure that reset is released.
PB11	28	GPIO	PB12	29	GPIO
AVDD	30 34	Analog power supply.	PB13	31	GPIO
PB14	32	GPIO	PD0	35	GPIO (5V)
PD1	36	GPIO	PD3	37	GPIO
PD4	38	GPIO	PD5	39	GPIO
PD6	40	GPIO	PD7	41	GPIO
PD8	42	GPIO	PC6	43	GPIO
PC7	44	GPIO	VREGVSS	45	Voltage regulator VSS
VREGSW	46	DCDC regulator switching node	VREGVDD	47	Voltage regulator VDD input
DVDD	48	Digital power supply.	DECOUPLE	49	Decouple output for on-chip voltage regulator. An external decoupling capacitor is required at this pin.
PE4	52	GPIO	PE5	53	GPIO
PE6	54	GPIO	PE7	55	GPIO
PC8	56	GPIO	PC9	57	GPIO
PC10	58	GPIO (5V)	PC11	59	GPIO (5V)
PC13	60	GPIO (5V)	PC14	61	GPIO (5V)
PC15	62	GPIO (5V)	PF0	63	GPIO (5V)
PF1	64	GPIO (5V)	PF2	65	GPIO
PF3	66	GPIO	PF4	67	GPIO
PF5	68	GPIO	PE8	71	GPIO
PE9	72	GPIO	PE10	73	GPIO
PE11	74	GPIO	BODEN	75	Brown-Out Detector Enable. This pin may be left disconnected or tied to AVDD.

Pin Name	Pin(s)	Description	Pin Name	Pin(s)	Description
PB8	8	GPIO	RESETn	9	Reset input, active low. To apply an external reset source to this pin, it is required to only drive this pin low during reset, and let the internal pull-up ensure that reset is released.
PB11	10	GPIO	AVDD	11 15	Analog power supply.
PB13	12	GPIO	PB14	13	GPIO
PD4	16	GPIO	PD5	17	GPIO
PD6	18	GPIO	PD7	19	GPIO
DVDD	20	Digital power supply.	DECOUPLE	21	Decouple output for on-chip voltage regulator. An external decoupling capacitor is required at this pin.
PC13	22	GPIO (5V)	PC14	23	GPIO (5V)
PC15	24	GPIO (5V)	PF0	25	GPIO (5V)
PF1	26	GPIO (5V)	PF2	27	GPIO
PE10	29	GPIO	PE11	30	GPIO
PE12	31	GPIO	PE13	32	GPIO
Note: 1. GPIO with 5V tolerance are indicated by (5V).					

Alternate	LOCATION		
Functionality	0 - 3	4 - 7	Description
CAN0_TX	0: PC1 1: PF2 2: PD1		CAN0 TX.
CMU_CLK0	0: PA2 1: PC12 2: PD7	4: PF2 5: PA12	Clock Management Unit, clock output number 0.
CMU_CLK1	0: PA1 1: PD8 2: PE12	4: PF3 5: PB11	Clock Management Unit, clock output number 1.
CMU_CLK2	0: PA0 1: PA3 2: PD6	4: PA3	Clock Management Unit, clock output number 2.
CMU_CLKI0	0: PD4 1: PA3 2: PB8 3: PB13	6: PE12 7: PB11	Clock Management Unit, clock input number 0.
DBG_SWCLKTCK	0: PF0		Debug-interface Serial Wire clock input and JTAG Test Clock. Note that this function is enabled to the pin out of reset, and has a built-in pull down.
DBG_SWDIOTMS	0: PF1		Debug-interface Serial Wire data input / output and JTAG Test Mode Select. Note that this function is enabled to the pin out of reset, and has a built-in pull up.
DBG_TDI	0: PF5		Debug-interface JTAG Test Data In. Note that this function becomes available after the first valid JTAG command is received, and has a built-in pull up when JTAG is active.
DBG_TDO	0: PF2		Debug-interface JTAG Test Data Out. Note that this function becomes available after the first valid JTAG command is received.
GPIO_EM4WU0	0: PA0		Pin can be used to wake the system up from EM4
GPIO_EM4WU1	0: PA6		Pin can be used to wake the system up from EM4
GPIO_EM4WU2	0: PC9		Pin can be used to wake the system up from EM4
GPIO_EM4WU3	0: PF1		Pin can be used to wake the system up from EM4

Alternate	LOCATION		
Functionality	0 - 3	4 - 7	Description
US2_CLK	0: PC4 1: PB5 2: PA9 3: PA15	5: PF2	USART2 clock input / output.
US2_CS	0: PC5 1: PB6 2: PA10 3: PB11	5: PF5	USART2 chip select input / output.
US2_CTS	0: PC1 1: PB12	4: PC12 5: PD6	USART2 Clear To Send hardware flow control input.
US2_RTS	0: PC0 2: PA12 3: PC14	4: PC13 5: PD8	USART2 Request To Send hardware flow control output.
US2_RX	0: PC3 1: PB4 2: PA8 3: PA14	5: PF1	USART2 Asynchronous Receive. USART2 Synchronous mode Master Input / Slave Output (MISO).
US2_TX	0: PC2 1: PB3 3: PA13	5: PF0	USART2 Asynchronous Transmit. Also used as receive input in half duplex communication. USART2 Synchronous mode Master Output / Slave Input (MOSI).
US3_CLK	0: PA2 1: PD7 2: PD4		USART3 clock input / output.
US3_CS	0: PA3 1: PE4 2: PC14 3: PC0		USART3 chip select input / output.
US3_CTS	0: PA4 1: PE5 2: PD6		USART3 Clear To Send hardware flow control input.
US3_RTS	0: PA5 1: PC1 2: PA14 3: PC15		USART3 Request To Send hardware flow control output.
US3_RX	0: PA1 1: PE7 2: PB7		USART3 Asynchronous Receive. USART3 Synchronous mode Master Input / Slave Output (MISO).
US3_TX	0: PA0 1: PE6 2: PB3		USART3 Asynchronous Transmit. Also used as receive input in half duplex communication. USART3 Synchronous mode Master Output / Slave Input (MOSI).
VDAC0_EXT	0: PD6		Digital to analog converter VDAC0 external reference input pin.

Alternate	LOCATION		
Functionality	0 - 3	4 - 7	Description
WTIM1_CC3	0: PD1 1: PD5 2: PC6	4: PE6	Wide timer 1 Capture Compare input / output channel 3.

Table 5.19. CSEN Bus and Pin Mapping

[illegible]

Table 6.1. TQFP80 Package Dimensions

Dimension	Min	Typ	Max
A	—	—	1.20
A1	0.05	—	0.15
A2	0.95	1.00	1.05
b	0.17	0.20	0.27
c	0.09	—	0.20
D	14.00 BSC		
D1	12.00 BSC		
e	0.50 BSC		
E	14.00 BSC		
E1	12.00 BSC		
L	0.45	0.60	0.75
L1	1.00 REF		
θ	0	3.5	7
aaa	0.20		
bbb	0.20		
ccc	0.08		
ddd	0.08		
eee	0.05		

Note:

1. All dimensions shown are in millimeters (mm) unless otherwise noted.
2. Dimensioning and Tolerancing per ANSI Y14.5M-1994.
3. This package outline conforms to JEDEC MS-026, variant ADD.
4. Recommended card reflow profile is per the JEDEC/IPC J-STD-020C specification for Small Body Components.

11.3 QFN32 Package Marking



Figure 11.3. QFN32 Package Marking

The package marking consists of:

- P P P P P P P P P P – The part number designation.
- T T T T T T – A trace or manufacturing code. The first letter is the device revision.
- Y Y – The last 2 digits of the assembly year.
- W W – The 2-digit workweek when the device was assembled.