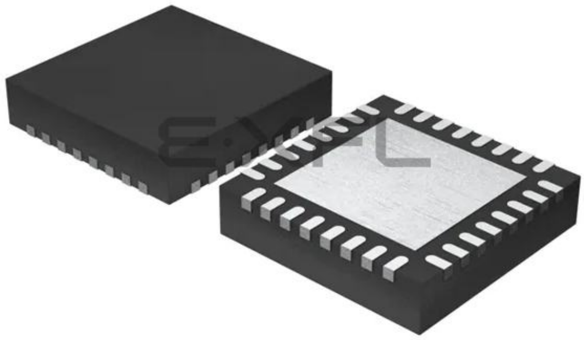

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"Embedded - Microcontrollers" refer to small, integrated circuits designed to perform specific tasks within larger systems. These microcontrollers are essentially compact computers on a single chip, containing a processor core, memory, and programmable input/output peripherals. They are called "embedded" because they are embedded within electronic devices to control various functions, rather than serving as standalone computers. Microcontrollers are crucial in modern electronics, providing the intelligence and control needed for a wide range of applications.

Applications of "Embedded - Microcontrollers"

Details	
Product Status	Discontinued at Digi-Key
Core Processor	ARM® Cortex®-M0+
Core Size	32-Bit Single-Core
Speed	48MHz
Connectivity	CANbus, I²C, IrDA, LINbus, SmartCard, SPI, UART/USART
Peripherals	Brown-out Detect/Reset, DMA, I²S, LCD, POR, PWM, WDT
Number of I/O	22
Program Memory Size	64KB (64K x 8)
Program Memory Type	FLASH
EEPROM Size	-
RAM Size	32K x 8
Voltage - Supply (Vcc/Vdd)	1.8V ~ 3.8V
Data Converters	A/D 12bit SAR; D/A 12bit
Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	32-WFQFN Exposed Pad
Supplier Device Package	32-QFN (5x5)
Purchase URL	https://www.e-xfl.com/product-detail/silicon-labs/efm32tg11b540f64gm32-ar

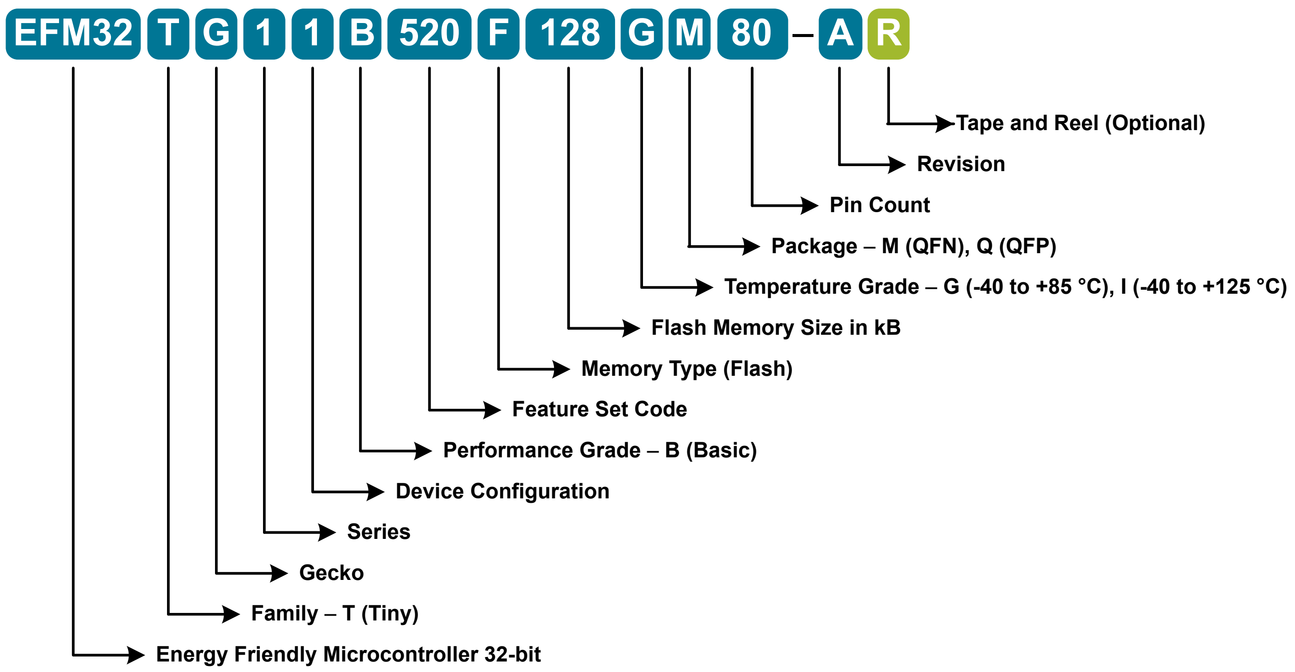

Figure 2.1. Ordering Code Key

3.6.5 Controller Area Network (CAN)

The CAN peripheral provides support for communication at up to 1 Mbps over CAN protocol version 2.0 part A and B. It includes 32 message objects with independent identifier masks and retains message RAM in EM2. Automatic retransmittion may be disabled in order to support Time Triggered CAN applications.

3.6.6 Peripheral Reflex System (PRS)

The Peripheral Reflex System provides a communication network between different peripheral modules without software involvement. Peripheral modules producing Reflex signals are called producers. The PRS routes Reflex signals from producers to consumer peripherals which in turn perform actions in response. Edge triggers and other functionality such as simple logic operations (AND, OR, NOT) can be applied by the PRS to the signals. The PRS allows peripheral to act autonomously without waking the MCU core, saving power.

3.6.7 Low Energy Sensor Interface (LESENSE)

The Low Energy Sensor Interface LESENSE™ is a highly configurable sensor interface with support for up to 16 individually configurable sensors. By controlling the analog comparators, ADC, and DAC, LESENSE is capable of supporting a wide range of sensors and measurement schemes, and can for instance measure LC sensors, resistive sensors and capacitive sensors. LESENSE also includes a programmable finite state machine which enables simple processing of measurement results without CPU intervention. LESENSE is available in energy mode EM2, in addition to EM0 and EM1, making it ideal for sensor monitoring in applications with a strict energy budget.

3.7 Security Features

3.7.1 GPCRC (General Purpose Cyclic Redundancy Check)

The GPCRC module implements a Cyclic Redundancy Check (CRC) function. It supports both 32-bit and 16-bit polynomials. The supported 32-bit polynomial is 0x04C11DB7 (IEEE 802.3), while the 16-bit polynomial can be programmed to any value, depending on the needs of the application.

3.7.2 Crypto Accelerator (CRYPTO)

The Crypto Accelerator is a fast and energy-efficient autonomous hardware encryption and decryption accelerator. Tiny Gecko Series 1 devices support AES encryption and decryption with 128- or 256-bit keys, ECC over both GF(P) and GF(2^m), and SHA-1 and SHA-2 (SHA-224 and SHA-256).

Supported block cipher modes of operation for AES include: ECB, CTR, CBC, PCBC, CFB, OFB, GCM, CBC-MAC, GMAC and CCM.

Supported ECC NIST recommended curves include P-192, P-224, P-256, K-163, K-233, B-163 and B-233.

The CRYPTO module allows fast processing of GCM (AES), ECC and SHA with little CPU intervention. CRYPTO also provides trigger signals for DMA read and write operations.

3.7.3 True Random Number Generator (TRNG)

The TRNG module is a non-deterministic random number generator based on a full hardware solution. The TRNG is validated with NIST800-22 and AIS-31 test suites as well as being suitable for FIPS 140-2 certification (for the purposes of cryptographic key generation).

3.7.4 Security Management Unit (SMU)

The Security Management Unit (SMU) allows software to set up fine-grained security for peripheral access, which is not possible in the Memory Protection Unit (MPU). Peripherals may be secured by hardware on an individual basis, such that only priveleged accesses to the peripheral's register interface will be allowed. When an access fault occurs, the SMU reports the specific peripheral involved and can optionally generate an interrupt.

3.8 Analog

4.1.6.2 Current Consumption 3.3 V using DC-DC Converter

Unless otherwise indicated, typical conditions are: VREGVDD = AVDD = IOVDD = 3.3 V, DVDD = 1.8 V DC-DC output. T = 25 °C. Minimum and maximum values in this table represent the worst conditions across supply voltage and process variation at T = 25 °C.

Table 4.7. Current Consumption 3.3 V using DC-DC Converter

Parameter	Symbol	Test Condition	Min	Typ	Max	Unit
Current consumption in EM0 mode with all peripherals disabled, DCDC in Low Noise DCM mode[2]	I_{ACTIVE_DCM}	48 MHz crystal, CPU running while loop from flash	—	38	—	μA/MHz
		48 MHz HFRCO, CPU running while loop from flash	—	37	—	μA/MHz
		48 MHz HFRCO, CPU running Prime from flash	—	45	—	μA/MHz
		48 MHz HFRCO, CPU running CoreMark loop from flash	—	53	—	μA/MHz
		32 MHz HFRCO, CPU running while loop from flash	—	43	—	μA/MHz
		26 MHz HFRCO, CPU running while loop from flash	—	47	—	μA/MHz
		16 MHz HFRCO, CPU running while loop from flash	—	61	—	μA/MHz
		1 MHz HFRCO, CPU running while loop from flash	—	587	—	μA/MHz
Current consumption in EM0 mode with all peripherals disabled, DCDC in Low Noise CCM mode[1]	I_{ACTIVE_CCM}	48 MHz crystal, CPU running while loop from flash	—	49	—	μA/MHz
		48 MHz HFRCO, CPU running while loop from flash	—	48	—	μA/MHz
		48 MHz HFRCO, CPU running Prime from flash	—	55	—	μA/MHz
		48 MHz HFRCO, CPU running CoreMark loop from flash	—	63	—	μA/MHz
		32 MHz HFRCO, CPU running while loop from flash	—	60	—	μA/MHz
		26 MHz HFRCO, CPU running while loop from flash	—	68	—	μA/MHz
		16 MHz HFRCO, CPU running while loop from flash	—	96	—	μA/MHz
		1 MHz HFRCO, CPU running while loop from flash	—	1157	—	μA/MHz
Current consumption in EM0 mode with all peripherals disabled, DCDC in LP mode[3]	I_{ACTIVE_LPM}	32 MHz HFRCO, CPU running while loop from flash	—	32	—	μA/MHz
		26 MHz HFRCO, CPU running while loop from flash	—	33	—	μA/MHz
		16 MHz HFRCO, CPU running while loop from flash	—	36	—	μA/MHz
		1 MHz HFRCO, CPU running while loop from flash	—	156	—	μA/MHz

4.1.9 Oscillators

4.1.9.1 Low-Frequency Crystal Oscillator (LFXO)

Table 4.11. Low-Frequency Crystal Oscillator (LFXO)

Parameter	Symbol	Test Condition	Min	Typ	Max	Unit
Crystal frequency	f_{LFXO}		—	32.768	—	kHz
Supported crystal equivalent series resistance (ESR)	ESR_{LFXO}		—	—	70	kΩ
Supported range of crystal load capacitance [1]	C_{LFXO_CL}		6	—	18	pF
On-chip tuning cap range [2]	C_{LFXO_T}	On each of LFXTAL_N and LFXTAL_P pins	8	—	40	pF
On-chip tuning cap step size	SS_{LFXO}		—	0.25	—	pF
Current consumption after startup [3]	I_{LFXO}	ESR = 70 kOhm, C_L = 7 pF, GAIN[4] = 2, AGC[4] = 1	—	273	—	nA
Start- up time	t_{LFXO}	ESR = 70 kOhm, C_L = 7 pF, GAIN[4] = 2	—	308	—	ms

Note:
1. Total load capacitance as seen by the crystal.
2. The effective load capacitance seen by the crystal will be C_{LFXO_T} /2. This is because each XTAL pin has a tuning cap and the two caps will be seen in series by the crystal.
3. Block is supplied by AVDD if ANASW = 0, or DVDD if ANASW=1 in EMU_PWRCTRL register.
4. In CMU_LFXOCTRL register.

4.1.9.2 High-Frequency Crystal Oscillator (HFXO)

Table 4.12. High-Frequency Crystal Oscillator (HFXO)

Parameter	Symbol	Test Condition	Min	Typ	Max	Unit
Crystal frequency	f_{HFXO}		4	—	48	MHz
Supported crystal equivalent series resistance (ESR)	ESR_{HFXO}	48 MHz crystal	—	—	50	Ω
		24 MHz crystal	—	—	150	Ω
		4 MHz crystal	—	—	180	Ω
Supported range of crystal load capacitance[1]	C_{HFXO_CL}		TBD	—	TBD	pF
Nominal on-chip tuning cap range[2]	C_{HFXO_T}	On each of HFXTAL_N and HFXTAL_P pins	8.7	—	51.7	pF
On-chip tuning capacitance step	SS_{HFXO}		—	0.08	—	pF
Startup time	t_{HFXO}	48 MHz crystal, ESR = 50 Ohm, C_L = 8 pF	—	350	—	µs
		24 MHz crystal, ESR = 150 Ohm, C_L = 6 pF	—	700	—	µs
		4 MHz crystal, ESR = 180 Ohm, C_L = 18 pF	—	3	—	ms
Current consumption after startup	I_{HFXO}	48 MHz crystal	—	880	—	µA
		24 MHz crystal	—	420	—	µA
		4 MHz crystal	—	80	—	µA

Note:

1. Total load capacitance as seen by the crystal.
2. The effective load capacitance seen by the crystal will be C_{HFXO_T} /2. This is because each XTAL pin has a tuning cap and the two caps will be seen in series by the crystal.

4.1.13 Analog to Digital Converter (ADC)

Specified at 1 Msps, ADCCLK = 16 MHz, BIASPROG = 0, GPBIASACC = 0, unless otherwise indicated.

Table 4.20. Analog to Digital Converter (ADC)

Parameter	Symbol	Test Condition	Min	Typ	Max	Unit
Resolution	$V_{RESOLUTION}$		6	—	12	Bits
Input voltage range[5]	V_{ADCIN}	Single ended	—	—	V_{FS}	V
		Differential	$-V_{FS}/2$	—	$V_{FS}/2$	V
Input range of external reference voltage, single ended and differential	$V_{ADCREFIN\_P}$		1	—	V_{AVDD}	V
Power supply rejection[2]	$PSRR_{ADC}$	At DC	—	80	—	dB
Analog input common mode rejection ratio	$CMRR_{ADC}$	At DC	—	80	—	dB
Current from all supplies, using internal reference buffer. Continous operation. WARMUPMODE[4] = KEEPADCWARM	$I_{ADC\_CONTINOUS\_LP}$	1 Msps / 16 MHz ADCCLK, BIASPROG = 0, GPBIASACC = 1 [3]	—	270	TBD	μA
		250 ksps / 4 MHz ADCCLK, BIASPROG = 6, GPBIASACC = 1 [3]	—	125	—	μA
		62.5 ksps / 1 MHz ADCCLK, BIASPROG = 15, GPBIASACC = 1 [3]	—	80	—	μA
Current from all supplies, using internal reference buffer. Duty-cycled operation. WARMUPMODE[4] = NORMAL	$I_{ADC\_NORMAL\_LP}$	35 ksps / 16 MHz ADCCLK, BIASPROG = 0, GPBIASACC = 1 [3]	—	45	—	μA
		5 ksps / 16 MHz ADCCLK BIASPROG = 0, GPBIASACC = 1 [3]	—	8	—	μA
Current from all supplies, using internal reference buffer. Duty-cycled operation. AWARMUPMODE[4] = KEEPINSTANDBY or KEEPINSLOWACC	$I_{ADC\_STANDBY\_LP}$	125 ksps / 16 MHz ADCCLK, BIASPROG = 0, GPBIASACC = 1 [3]	—	105	—	μA
		35 ksps / 16 MHz ADCCLK, BIASPROG = 0, GPBIASACC = 1 [3]	—	70	—	μA
Current from all supplies, using internal reference buffer. Continous operation. WARMUPMODE[4] = KEEPADCWARM	$I_{ADC\_CONTINOUS\_HP}$	1 Msps / 16 MHz ADCCLK, BIASPROG = 0, GPBIASACC = 0 [3]	—	325	—	μA
		250 ksps / 4 MHz ADCCLK, BIASPROG = 6, GPBIASACC = 0 [3]	—	175	—	μA
		62.5 ksps / 1 MHz ADCCLK, BIASPROG = 15, GPBIASACC = 0 [3]	—	125	—	μA
Current from all supplies, using internal reference buffer. Duty-cycled operation. WARMUPMODE[4] = NORMAL	$I_{ADC\_NORMAL\_HP}$	35 ksps / 16 MHz ADCCLK, BIASPROG = 0, GPBIASACC = 0 [3]	—	85	—	μA
		5 ksps / 16 MHz ADCCLK BIASPROG = 0, GPBIASACC = 0 [3]	—	16	—	μA
Current from all supplies, using internal reference buffer. Duty-cycled operation. AWARMUPMODE[4] = KEEPINSTANDBY or KEEPINSLOWACC	$I_{ADC\_STANDBY\_HP}$	125 ksps / 16 MHz ADCCLK, BIASPROG = 0, GPBIASACC = 0 [3]	—	160	—	μA
		35 ksps / 16 MHz ADCCLK, BIASPROG = 0, GPBIASACC = 0 [3]	—	125	—	μA
Current from HFPERCLK	I_{ADC_CLK}	HFPERCLK = 16 MHz	—	166	—	μA

4.1.15 Digital to Analog Converter (VDAC)

DRIVESTRENGTH = 2 unless otherwise specified. Primary VDAC output.

Table 4.22. Digital to Analog Converter (VDAC)

Parameter	Symbol	Test Condition	Min	Typ	Max	Unit
Output voltage	V_{DACOUT}	Single-Ended	0	—	V_{VREF}	V
		Differential[2]	$-V_{VREF}$	—	V_{VREF}	V
Current consumption including references (2 channels)[1]	I_{DAC}	500 ksps, 12-bit, DRIVESTRENGTH = 2, REFSEL = 4	—	396	—	µA
		44.1 ksps, 12-bit, DRIVESTRENGTH = 1, REFSEL = 4	—	72	—	µA
		200 Hz refresh rate, 12-bit Sample-Off mode in EM2, DRIVESTRENGTH = 2, BGRREQTIME = 1, EM2REFENTIME = 9, REFSEL = 4, SETTLETIME = 0x0A, WARMUPTIME = 0x02	—	2	—	µA
Current from HFPERCLK[4]	I_{DAC_CLK}		—	5.8	—	µA/MHz
Sample rate	SR_{DAC}		—	—	500	ksps
DAC clock frequency	f_{DAC}		—	—	1	MHz
Conversion time	$t_{DACCONV}$	f_{DAC} = 1MHz	2	—	—	µs
Settling time	$t_{DACSETTLE}$	50% fs step settling to 5 LSB	—	2.5	—	µs
Startup time	$t_{DACSTARTUP}$	Enable to 90% fs output, settling to 10 LSB	—	—	12	µs
Output impedance	R_{OUT}	DRIVESTRENGTH = 2, 0.4 V ≤ V_{OUT} ≤ V_{OPA} - 0.4 V, -8 mA < I_{OUT} < 8 mA, Full supply range	—	2	—	Ω
		DRIVESTRENGTH = 0 or 1, 0.4 V ≤ V_{OUT} ≤ V_{OPA} - 0.4 V, -400 µA < I_{OUT} < 400 µA, Full supply range	—	2	—	Ω
		DRIVESTRENGTH = 2, 0.1 V ≤ V_{OUT} ≤ V_{OPA} - 0.1 V, -2 mA < I_{OUT} < 2 mA, Full supply range	—	2	—	Ω
		DRIVESTRENGTH = 0 or 1, 0.1 V ≤ V_{OUT} ≤ V_{OPA} - 0.1 V, -100 µA < I_{OUT} < 100 µA, Full supply range	—	2	—	Ω
Power supply rejection ratio[6]	PSRR	Vout = 50% fs. DC	—	65.5	—	dB

Parameter	Symbol	Test Condition	Min	Typ	Max	Unit
Signal to noise and distortion ratio (1 kHz sine wave), Noise band limited to 250 kHz	$SNDR_{DAC}$	500 ksps, single-ended, internal 1.25V reference	—	60.4	—	dB
		500 ksps, single-ended, internal 2.5V reference	—	61.6	—	dB
		500 ksps, single-ended, 3.3V VDD reference	—	64.0	—	dB
		500 ksps, differential, internal 1.25V reference	—	63.3	—	dB
		500 ksps, differential, internal 2.5V reference	—	64.4	—	dB
		500 ksps, differential, 3.3V VDD reference	—	65.8	—	dB
Signal to noise and distortion ratio (1 kHz sine wave), Noise band limited to 22 kHz	$SNDR_{DAC\_BAND}$	500 ksps, single-ended, internal 1.25V reference	—	65.3	—	dB
		500 ksps, single-ended, internal 2.5V reference	—	66.7	—	dB
		500 ksps, differential, 3.3V VDD reference	—	68.5	—	dB
		500 ksps, differential, internal 1.25V reference	—	67.8	—	dB
		500 ksps, differential, internal 2.5V reference	—	69.0	—	dB
		500 ksps, single-ended, 3.3V VDD reference	—	70.0	—	dB
Total harmonic distortion	THD		—	70.2	—	dB
Differential non-linearity[3]	DNL_{DAC}		TBD	—	TBD	LSB
Intergral non-linearity	INL_{DAC}		TBD	—	TBD	LSB
Offset error[5]	V_{OFFSET}	T = 25 °C	TBD	—	TBD	mV
		Across operating temperature range	TBD	—	TBD	mV
Gain error[5]	V_{GAIN}	T = 25 °C, Low-noise internal reference (REFSEL = 1V25LN or 2V5LN)	TBD	—	TBD	%
		Across operating temperature range, Low-noise internal reference (REFSEL = 1V25LN or 2V5LN)	TBD	—	TBD	%
External load capacitance, OUTSCALE=0	C_{LOAD}		—	—	75	pF

Parameter	Symbol	Test Condition	Min	Typ	Max	Unit

Note:

1. Supply current specifications are for VDAC circuitry operating with static output only and do not include current required to drive the load.
2. In differential mode, the output is defined as the difference between two single-ended outputs. Absolute voltage on each output is limited to the single-ended range.
3. Entire range is monotonic and has no missing codes.
4. Current from HFPERCLK is dependent on HFPERCLK frequency. This current contributes to the total supply current used when the clock to the DAC module is enabled in the CMU.
5. Gain is calculated by measuring the slope from 10% to 90% of full scale. Offset is calculated by comparing actual VDAC output at 10% of full scale to ideal VDAC output at 10% of full scale with the measured gain.
6. PSRR calculated as 20 * $\log_{10}(\Delta VDD / \Delta V_{OUT})$, VDAC output at 90% of full scale

4.1.21.3 I2C Fast-mode Plus (Fm+)[1]

Table 4.30. I2C Fast-mode Plus (Fm+)[1]

Parameter	Symbol	Test Condition	Min	Typ	Max	Unit
SCL clock frequency[2]	f_{SCL}		0	—	1000	kHz
SCL clock low time	t_{LOW}		0.5	—	—	µs
SCL clock high time	t_{HIGH}		0.26	—	—	µs
SDA set-up time	t_{SU_DAT}		50	—	—	ns
SDA hold time	t_{HD_DAT}		100	—	—	ns
Repeated START condition set-up time	t_{SU_STA}		0.26	—	—	µs
(Repeated) START condition hold time	t_{HD_STA}		0.26	—	—	µs
STOP condition set-up time	t_{SU_STO}		0.26	—	—	µs
Bus free time between a STOP and START condition	t_{BUF}		0.5	—	—	µs

Note:
1. For CLHR set to 0 or 1 in the I2Cn_CTRL register.
2. For the minimum HFPERCLK frequency required in Fast-mode Plus, refer to the I2C chapter in the reference manual.

4.2.1 Supply Current

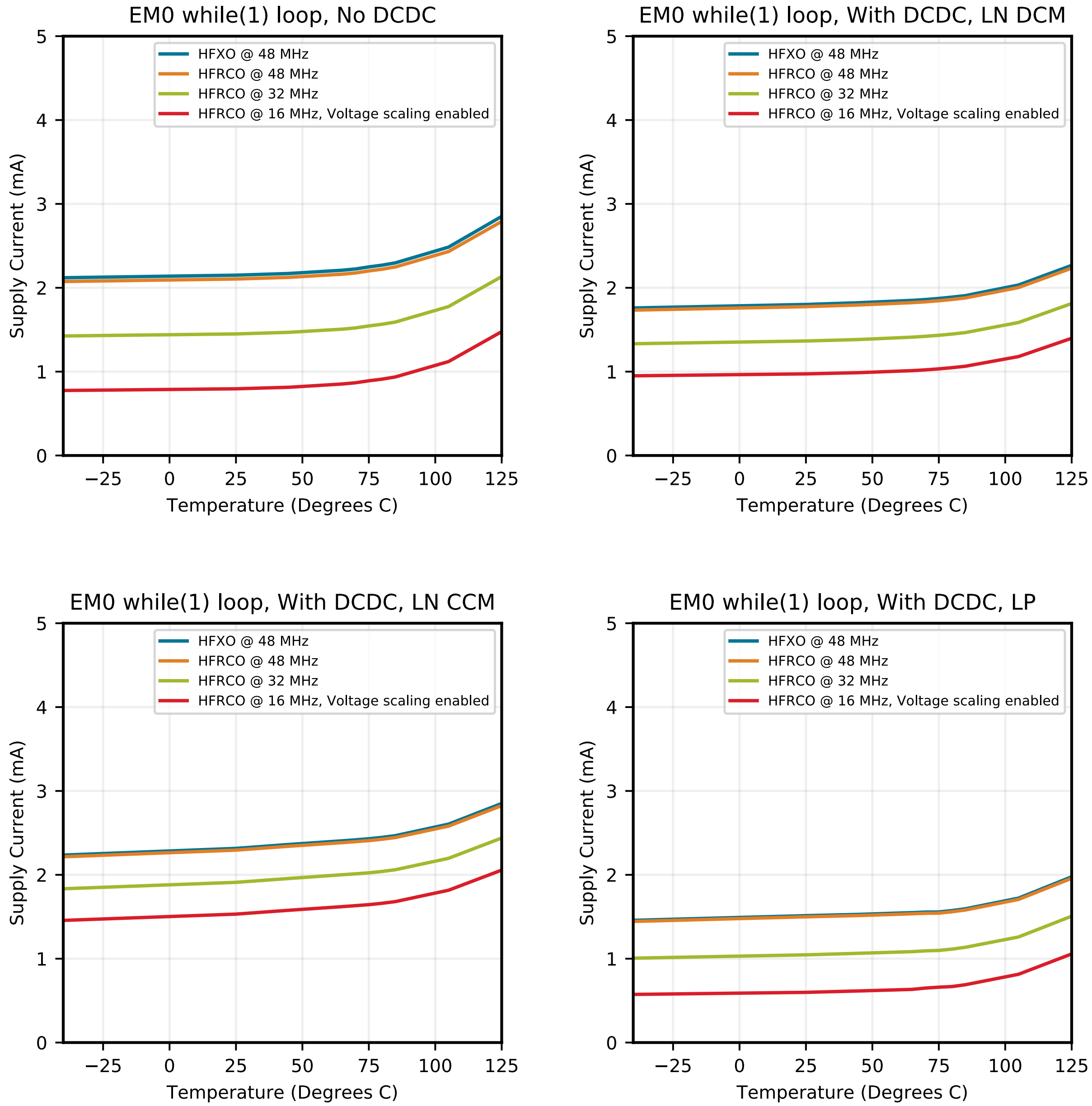

Figure 4.3. EM0 Active Mode Typical Supply Current vs. Temperature

Pin Name	Pin(s)	Description	Pin Name	Pin(s)	Description
PB4	10	GPIO	PB5	11	GPIO
PB6	12	GPIO	PC0	13	GPIO (5V)
PC1	14	GPIO (5V)	PC2	15	GPIO (5V)
PC3	16	GPIO (5V)	PC4	17	GPIO
PC5	18	GPIO	PB7	19	GPIO
PB8	20	GPIO	PA8	21	GPIO
PA9	22	GPIO	PA10	23	GPIO
PA12	24	GPIO	PA13	25	GPIO (5V)
PA14	26	GPIO	RESETn	27	Reset input, active low. To apply an external reset source to this pin, it is required to only drive this pin low during reset, and let the internal pull-up ensure that reset is released.
PB11	28	GPIO	PB12	29	GPIO
AVDD	30 34	Analog power supply.	PB13	31	GPIO
PB14	32	GPIO	PD0	35	GPIO (5V)
PD1	36	GPIO	PD2	37	GPIO (5V)
PD3	38	GPIO	PD4	39	GPIO
PD5	40	GPIO	PD6	41	GPIO
PD7	42	GPIO	PD8	43	GPIO
PC6	44	GPIO	PC7	45	GPIO
VREGSW	47	DCDC regulator switching node	VREGVDD	48	Voltage regulator VDD input
DVDD	49	Digital power supply.	DECOUPLE	50	Decouple output for on-chip voltage regulator. An external decoupling capacitor is required at this pin.
PE4	52	GPIO	PE5	53	GPIO
PE6	54	GPIO	PE7	55	GPIO
PC8	56	GPIO	PC9	57	GPIO
PC10	58	GPIO (5V)	PC11	59	GPIO (5V)
PC12	60	GPIO (5V)	PC13	61	GPIO (5V)
PC14	62	GPIO (5V)	PC15	63	GPIO (5V)
PF0	64	GPIO (5V)	PF1	65	GPIO (5V)
PF2	66	GPIO	PF3	67	GPIO
PF4	68	GPIO	PF5	69	GPIO
PE8	71	GPIO	PE9	72	GPIO
PE10	73	GPIO	PE11	74	GPIO
BODEN	75	Brown-Out Detector Enable. This pin may be left disconnected or tied to AVDD.	PE12	76	GPIO
PE13	77	GPIO	PE14	78	GPIO

Pin Name	Pin(s)	Description	Pin Name	Pin(s)	Description
PE15	79	GPIO	PA15	80	GPIO

Note:

1. GPIO with 5V tolerance are indicated by (5V).

GPIO Name	Pin Alternate Functionality / Description			
	Analog	Timers	Communication	Other
PD5	BUSADC0Y BUSADC0X OPA2_OUT	WTIM0_CDTI1 #4 WTIM1_CC3 #1	US1_RTS #1 U0_CTS #5 LEU0_RX #0 I2C1_SCL #3	
PD6	BUSADC0Y BUSADC0X ADC0_EXTP VDAC0_EXT OPA1_P	TIM1_CC0 #4 WTIM0_CDTI2 #4 WTIM1_CC0 #2 LE-TIM0_OUT0 #0 PCNT0_S0IN #3	US0_RTS #5 US1_RX #2 US2_CTS #5 US3_CTS #2 U0_RTS #5 I2C0_SDA #1	CMU_CLK2 #2 LES_AL-TEX0 PRS_CH5 #2 ACMP0_O #2
PD7	BUSADC0Y BUSADC0X ADC0_EXTN OPA1_N	TIM1_CC1 #4 WTIM1_CC1 #2 LE-TIM0_OUT1 #0 PCNT0_S1IN #3	US1_TX #2 US3_CLK #1 U0_TX #6 I2C0_SCL #1	CMU_CLK0 #2 LES_AL-TEX1 ACMP1_O #2
PD8	BU_VIN	WTIM1_CC2 #2	US2_RTS #5	CMU_CLK1 #1
PC6	BUSACMP0Y BU-SACMP0X OPA3_P LCD_SEG32	WTIM1_CC3 #2	US0_RTS #2 US1_CTS #3 I2C0_SDA #2	LES_CH6
PC7	BUSACMP0Y BU-SACMP0X OPA3_N LCD_SEG33	WTIM1_CC0 #3	US0_CTS #2 US1_RTS #3 I2C0_SCL #2	LES_CH7
PE4	BUSDY BUSCX LCD_COM0	WTIM0_CC0 #0 WTIM1_CC1 #4	US0_CS #1 US1_CS #5 US3_CS #1 U0_RX #6 I2C0_SDA #7	
PE5	BUSCY BUSDX LCD_COM1	WTIM0_CC1 #0 WTIM1_CC2 #4	US0_CLK #1 US1_CLK #6 US3_CTS #1 I2C0_SCL #7	
PE6	BUSDY BUSCX LCD_COM2	WTIM0_CC2 #0 WTIM1_CC3 #4	US0_RX #1 US3_TX #1	PRS_CH6 #2
PE7	BUSCY BUSDX LCD_COM3	WTIM1_CC0 #5	US0_TX #1 US3_RX #1	PRS_CH7 #2
PC8	BUSACMP1Y BU-SACMP1X LCD_SEG34		US0_CS #2	LES_CH8 PRS_CH4 #0
PC9	BUSACMP1Y BU-SACMP1X LCD_SEG35		US0_CLK #2	LES_CH9 PRS_CH5 #0 GPIO_EM4WU2
PC10	BUSACMP1Y BU-SACMP1X		US0_RX #2	LES_CH10
PC11	BUSACMP1Y BU-SACMP1X		US0_TX #2 I2C1_SDA #4	LES_CH11
PC12	VDAC0_OUT1ALT / OPA1_OUTALT #0 BU-SACMP1Y BUSACMP1X	TIM1_CC3 #0	US0_RTS #3 US1_CTS #4 US2_CTS #4 U0_RTS #3	CMU_CLK0 #1 LES_CH12
PC13	VDAC0_OUT1ALT / OPA1_OUTALT #1 BU-SACMP1Y BUSACMP1X	TIM0_CDTI0 #1 TIM1_CC0 #0 TIM1_CC2 #4 PCNT0_S0IN #0	US0_CTS #3 US1_RTS #4 US2_RTS #4 U0_CTS #3	LES_CH13
PC14	VDAC0_OUT1ALT / OPA1_OUTALT #2 BU-SACMP1Y BUSACMP1X	TIM0_CDTI1 #1 TIM1_CC1 #0 TIM1_CC3 #4 LETIM0_OUT0 #5 PCNT0_S1IN #0	US0_CS #3 US1_CS #3 US2_RTS #3 US3_CS #2 U0_TX #3 LEU0_TX #5	LES_CH14 PRS_CH0 #2

Alternate	LOCATION		
Functionality	0 - 3	4 - 7	Description
U0_TX	2: PA3 3: PC14	4: PC4 5: PF1 6: PD7	UART0 Transmit output. Also used as receive input in half duplex communication.
US0_CLK	0: PE12 1: PE5 2: PC9 3: PC15	4: PB13 5: PA12	USART0 clock input / output.
US0_CS	0: PE13 1: PE4 2: PC8 3: PC14	4: PB14 5: PA13	USART0 chip select input / output.
US0_CTS	0: PE14 2: PC7 3: PC13	4: PB6 5: PB11	USART0 Clear To Send hardware flow control input.
US0_RTS	0: PE15 2: PC6 3: PC12	4: PB5 5: PD6	USART0 Request To Send hardware flow control output.
US0_RX	0: PE11 1: PE6 2: PC10 3: PE12	4: PB8 5: PC1	USART0 Asynchronous Receive. USART0 Synchronous mode Master Input / Slave Output (MISO).
US0_TX	0: PE10 1: PE7 2: PC11 3: PE13	4: PB7 5: PC0	USART0 Asynchronous Transmit. Also used as receive input in half duplex communication. USART0 Synchronous mode Master Output / Slave Input (MOSI).
US1_CLK	0: PB7 1: PD2 2: PF0 3: PC15	4: PC3 5: PB11 6: PE5	USART1 clock input / output.
US1_CS	0: PB8 1: PD3 2: PF1 3: PC14	4: PC0 5: PE4	USART1 chip select input / output.
US1_CTS	1: PD4 2: PF3 3: PC6	4: PC12 5: PB13	USART1 Clear To Send hardware flow control input.
US1_RTS	1: PD5 2: PF4 3: PC7	4: PC13 5: PB14	USART1 Request To Send hardware flow control output.
US1_RX	0: PC1 1: PD1 2: PD6	4: PC2 5: PA0 6: PA2	USART1 Asynchronous Receive. USART1 Synchronous mode Master Input / Slave Output (MISO).
US1_TX	0: PC0 1: PD0 2: PD7	4: PC1 5: PF2 6: PA14	USART1 Asynchronous Transmit. Also used as receive input in half duplex communication. USART1 Synchronous mode Master Output / Slave Input (MOSI).

Port	Bus	CH31	CH30	CH29	CH28	CH27	CH26	CH25	CH24	CH23	CH22	CH21	CH20	CH19	CH18	CH17	CH16	CH15	CH14	CH13	CH12	CH11	CH10	CH9	CH8	CH7	CH6	CH5	CH4	CH3	CH2	CH1	CH0
OPA3_OUT																																	
APORT1Y	BUSAY			PB13		PB11						PB5		PB3				PA15		PA13				PA9				PA5		PA3		PA1	
APORT2Y	BUSBY		PB14		PB12						PB6		PB4						PA14				PA10				PA6		PA4		PA2		PA0
APORT3Y	BUSCY											PF5		PF3		PF1		PE15		PE13		PE11		PE9		PE7		PE5					
APORT4Y	BUSDY												PF4		PF2		PF0		PE14		PE12		PE10		PE8		PE6		PE4				
OPA3_P																																	
APORT1X	BUSAX		PB14		PB12						PB6		PB4						PA14				PA10				PA6		PA4		PA2		PA0
APORT2X	BUSBX			PB13		PB11						PB5		PB3				PA15		PA13				PA9				PA5		PA3		PA1	
APORT3X	BUSCX												PF4		PF2		PF0		PE14		PE12		PE10		PE8		PE6		PE4				
APORT4X	BUSDX											PF5		PF3		PF1		PE15		PE13		PE11		PE9		PE7		PE5					
VDAC0_OUT0 / OPA0_OUT																																	
APORT1Y	BUSAY			PB13		PB11						PB5		PB3				PA15		PA13				PA9				PA5		PA3		PA1	
APORT2Y	BUSBY		PB14		PB12						PB6		PB4						PA14				PA10				PA6		PA4		PA2		PA0
APORT3Y	BUSCY											PF5		PF3		PF1		PE15		PE13		PE11		PE9		PE7		PE5					
APORT4Y	BUSDY												PF4		PF2		PF0		PE14		PE12		PE10		PE8		PE6		PE4				

Port	Bus	CH31	CH30	CH29	CH28	CH27	CH26	CH25	CH24	CH23	CH22	CH21	CH20	CH19	CH18	CH17	CH16	CH15	CH14	CH13	CH12	CH11	CH10	CH9	CH8	CH7	CH6	CH5	CH4	CH3	CH2	CH1	CH0
VDAC0_OUT1 / OPA1_OUT																																	
APORT1Y	BUSAY			PB13		PB11						PB5		PB3				PA15		PA13				PA9				PA5		PA3		PA1	
APORT2Y	BUSBY		PB14		PB12						PB6		PB4						PA14				PA10				PA6		PA4		PA2		PA0
APORT3Y	BUSCY											PF5		PF3		PF1		PE15		PE13		PE11		PE9		PE7		PE5					
APORT4Y	BUSDY												PF4		PF2		PF0		PE14		PE12		PE10		PE8		PE6		PE4				

Table 6.1. TQFP80 Package Dimensions

Dimension	Min	Typ	Max
A	—	—	1.20
A1	0.05	—	0.15
A2	0.95	1.00	1.05
b	0.17	0.20	0.27
c	0.09	—	0.20
D		14.00 BSC	
D1		12.00 BSC	
e		0.50 BSC	
E		14.00 BSC	
E1		12.00 BSC	
L	0.45	0.60	0.75
L1		1.00 REF	
θ	0	3.5	7
aaa		0.20	
bbb		0.20	
ccc		0.08	
ddd		0.08	
eee		0.05	

Note:
1. All dimensions shown are in millimeters (mm) unless otherwise noted.
2. Dimensioning and Tolerancing per ANSI Y14.5M-1994.
3. This package outline conforms to JEDEC MS-026, variant ADD.
4. Recommended card reflow profile is per the JEDEC/IPC J-STD-020C specification for Small Body Components.

7.3 QFN80 Package Marking

Figure 7.3. QFN80 Package Marking

The package marking consists of:

- PPPPPPPPPP – The part number designation.
- TTTTTT – A trace or manufacturing code. The first letter is the device revision.
- YY – The last 2 digits of the assembly year.
- WW – The 2-digit workweek when the device was assembled.

8. TQFP64 Package Specifications

8.1 TQFP64 Package Dimensions

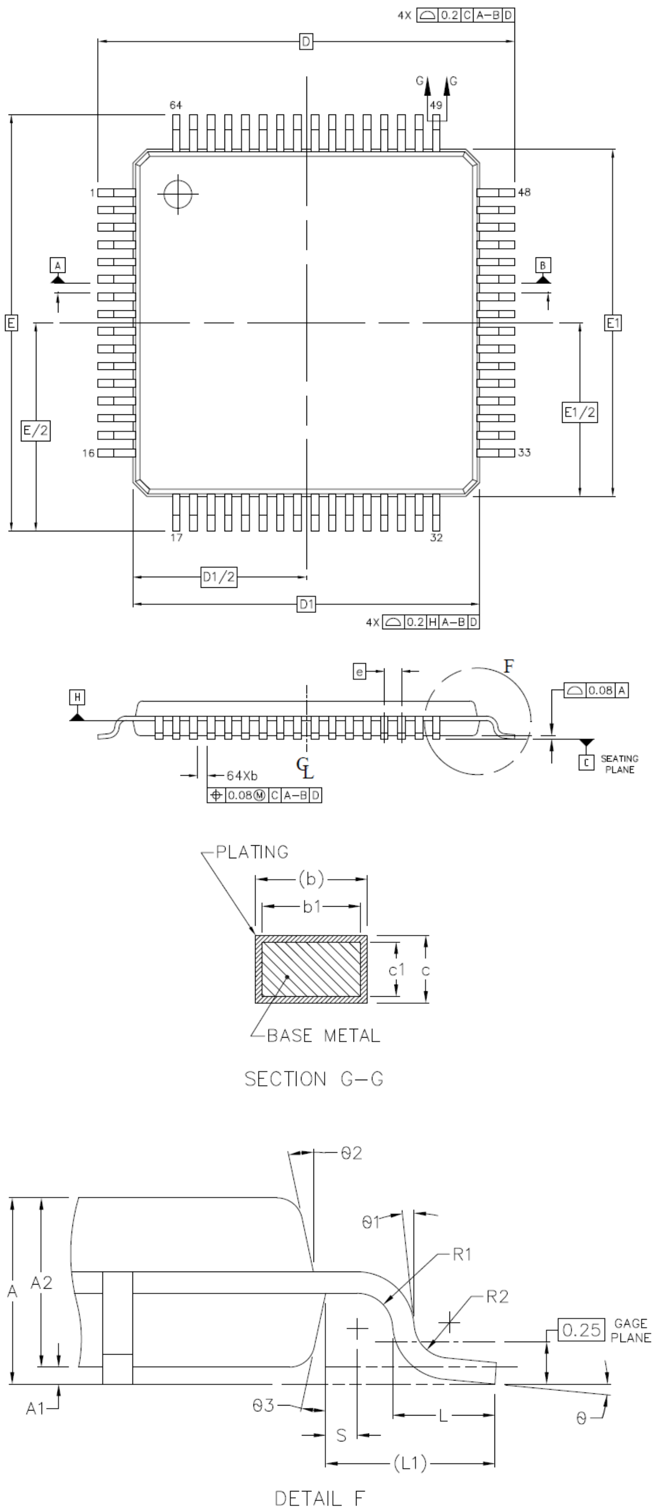

Figure 8.1. TQFP64 Package Drawing